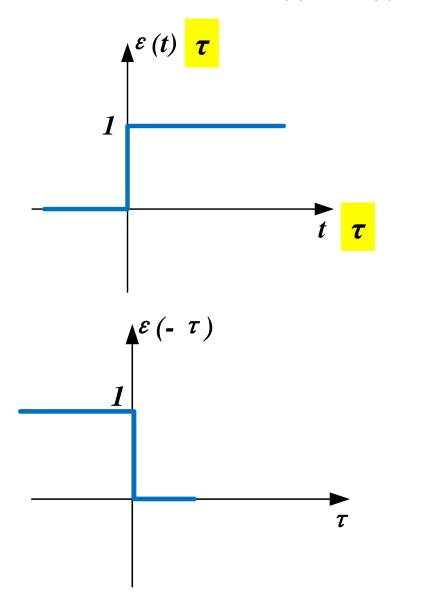
本次课学习内容

- 卷积积分性质及应用
- 正弦交流电基本概念
- 正弦量的相量表示

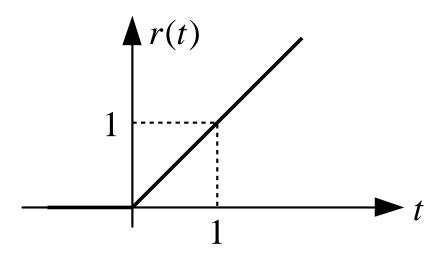
3.7 卷积积分

- 3.卷积的计算
- (1) 利用定义式求解
- (2) 利用性质求解

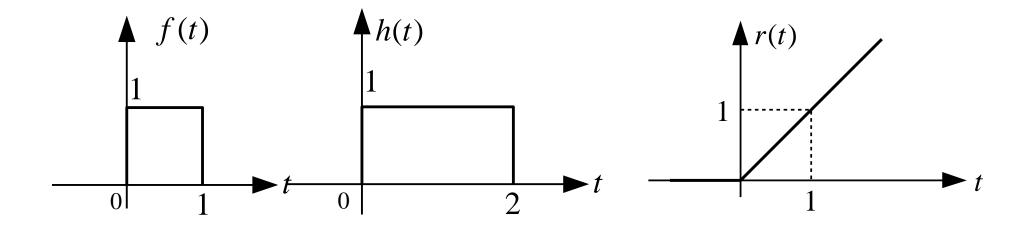
例: 计算 $r(t) = \varepsilon(t) * \varepsilon(t)$ 。



$$r(t) = \int_{-\infty}^{+\infty} \varepsilon(\tau) \varepsilon(t - \tau) d\tau$$
$$= \int_{0}^{t} d\tau = t, \quad t > 0$$



例: 利用时移特性及 $\varepsilon(t)$ * $\varepsilon(t)$ = r(t) , 计算y(t) = f(t) * h(t)。



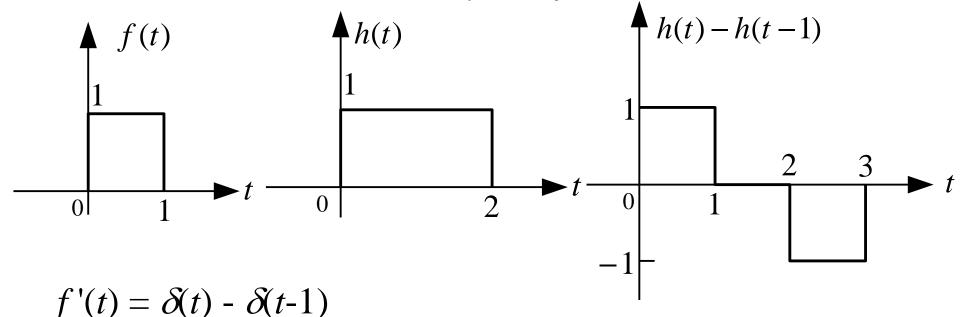
$$y(t) = f(t) * h(t) = [\varepsilon(t) - \varepsilon(t-1)] * [\varepsilon(t) - \varepsilon(t-2)]$$

$$= \varepsilon(t) * \varepsilon(t) - \varepsilon(t-1) * \varepsilon(t) - \varepsilon(t) * \varepsilon(t-2) - \varepsilon(t-1) * \varepsilon(t-2)$$

$$= r(t) - r(t-1) - r(t-2) + r(t-3)$$

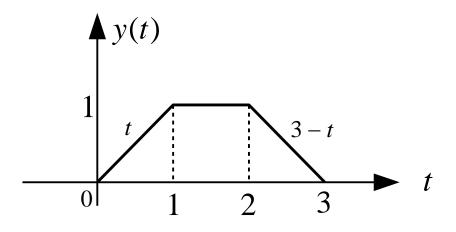
$$t = r(t) - r(t-1) - r(t-2) + r(t-3)$$

例3: 利用微积分特性,计算y(t) = f(t) * h(t)。



$$f'(t) * h^{(-1)}(t) = h^{(-1)}(t) - h^{(-1)}(t-1)$$

$$y(t) = \int_0^t [h(\tau) - h(\tau - 1)] d\tau$$



3.7 卷积积分

例3. 7-3 求例3.6-2所示电路在 $u_{\rm s}(t)={\rm e}^{-2t}\varepsilon(t){\rm V}$ 时的零状态响应

解 已在例3.6-2中求得该电路的冲激响应为 $h(t)=e^{-\frac{2}{3}t}\varepsilon(t)$ V

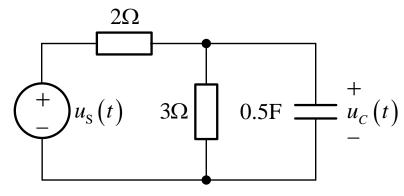
因此, 该电路在 $u_{\rm S}(t)={\rm e}^{-2t}\varepsilon(t){\rm V}$ 时的零状态响应为

$$u_{Czs}(t) = u_{S}(t) * h(t) = e^{-2t} \varepsilon(t) * e^{-\frac{5}{3}t} \varepsilon(t)$$

$$= \int_{-\infty}^{+\infty} e^{-2\tau} \varepsilon(\tau) \cdot e^{-\frac{5}{3}(t-\tau)} \varepsilon(t-\tau) d\tau$$

$$= e^{-\frac{5}{3}t} \int_{0}^{t} e^{-\frac{1}{3}\tau} d\tau \varepsilon(t) = -\frac{1}{3} e^{-\frac{5}{3}t} e^{-\frac{1}{3}\tau} \Big|_{\tau=0}^{t} \varepsilon(t)$$

$$= -\frac{1}{3} e^{-\frac{5}{3}t} \left(e^{-\frac{1}{3}t} - 1 \right) \varepsilon(t) = \frac{1}{3} \left(e^{-\frac{5}{3}t} - e^{-2t} \right) \varepsilon(t) V$$

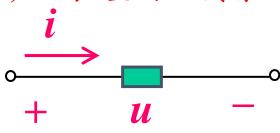


第四章 线性电路的正弦稳态分析

- 4.1 正弦交流电基本概念
- 4.2 正弦量的相量表示
- 4.3 基尔霍夫定律的相量形式
- 4.4 无源单口网络的阻抗、导纳及等效变换
- 4.5 正弦稳态电路的相量分析法
- 4.6正弦稳态电路的功率
- 4.7 磁耦合电路的正弦稳态分析

1 正弦量的基本概念

(1) 正弦量的三要素



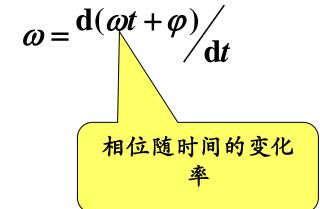
$$i(t)=I_{\mathbf{m}}\cos(\omega t + \varphi)$$
 相位

- (a) 幅值 (amplitude) (振幅、 最大值) I_m
- (b) 角频率(angular frequency) ω

$$\omega = 2\pi f = \frac{2\pi}{T}$$
 单位: rad/s

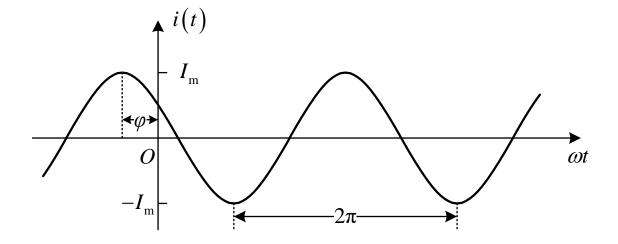
(c) 初相位/角(initial phase angle) φ

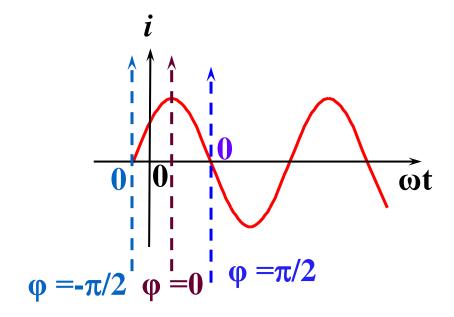
$$i(t)|_{t=0}=I_{m}\cos \varphi$$



t=0时的相位

$$i(t) = I_{\rm m} \cos(\omega t + \varphi)$$





一般
$$|\phi| \le \pi$$

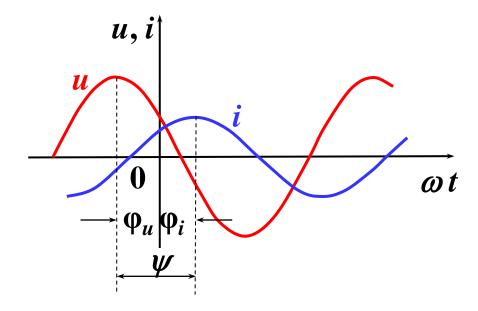
(2) 同频率正弦量的相位差 (phase difference)。

设
$$u(t)=U_{m}\cos(\omega t+\varphi_{u}), i(t)=I_{m}\cos(\omega t+\varphi_{i})$$

相位差
$$\psi = (\omega t + \varphi_u) - (\omega t + \varphi_i) = \varphi_u - \varphi_i$$

 $\psi > 0$, u 领先(lead)(超前)i, 或i 落后(lag)(滞后)u

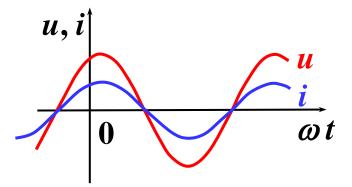
 $\psi < 0$, i 领先(超前) u, 或u 落后(滞后) i

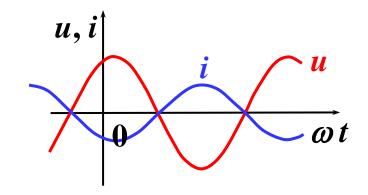


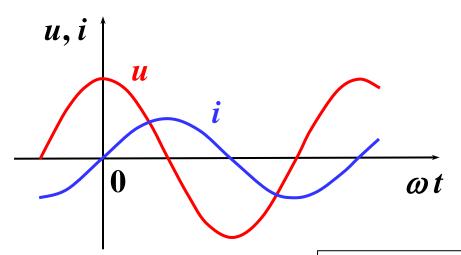
特殊相位关系

$$\psi = 0$$
, 同相

$$\Psi = \pm \pi \ (\pm 180^{\circ})$$
,反相







$$\psi = 90^{\circ}$$
 u 领先 i 90°
 \dot{a} i 落后 u 90°
 \ddot{a} i 落后 i 270°
 \dot{a} i 领先 u 270°

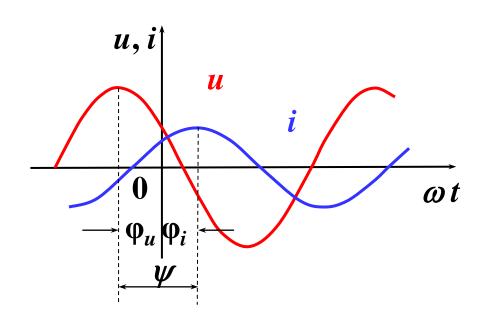




图中u和i之间的相位关系是



- B u 滞后 $i |\varphi_u| |\varphi_i|$
- u 超前 $i | \varphi_u | + | \varphi_i |$
- u 滞后 $i |\varphi_u| + |\varphi_i|$



例4.1-1 正弦电压

$$u_1(t) = 5\sin(314t + 15^{\circ})V$$
 $u_2(t) = -4\cos(314t + 45^{\circ})V$

求相位差,并比较相位关系。

解两正弦电压频率相同,因此可以进行相位比较。将 u_1 和 u_2

化成标准形式,可得

$$u_1(t) = 5\sin(314t + 15^\circ) = 5\cos(314t + 15^\circ - 90^\circ) = 5\cos(314t - 75^\circ)V$$

$$u_2(t) = -4\cos(314t + 45^\circ) = 4\cos(314t + 45^\circ - 180^\circ) = 4\cos(314t - 135^\circ)V$$

相位差为
$$\psi = -75^{\circ} - (-135^{\circ}) = 60^{\circ}$$

u₁ 超前 u₂ 60°, 或者 u₂ 滯后 u₁ 60°

$$\sin \alpha = \cos(\frac{\pi}{2} - \alpha) = \cos(\alpha - \frac{\pi}{2})$$
$$-\cos \alpha = \cos(\pi - \alpha) = \cos(\alpha - \pi)$$

$$-\sin\alpha = -\cos(\frac{\pi}{2} - \alpha) = \cos(\alpha + \frac{\pi}{2})$$



(3) 周期量的有效值(effective value)

周期交流电信号的有效值是根据相同平均热效应来定义的,对同 (a) 定义 一电阻分别通过交流电流i(t)和直流电流I,若一个周期内电阻消耗的能量相等,则把 I 称为i(t)的有效值

$$I = \sqrt{\frac{1}{T} \int_0^T i^2(t) \mathrm{d}t}$$

$$U \stackrel{\text{def}}{=} \sqrt{\frac{1}{T} \int_0^T u^2(t) dt}$$

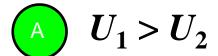
有效值也称均方根值(Root-Mean-Square,简记为RMS)

(b) 正弦电流、电压的有效值

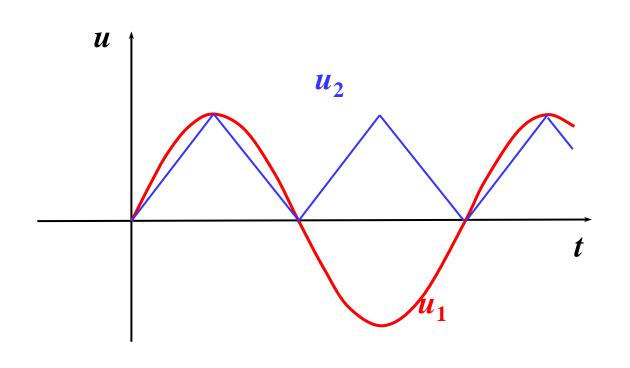
$$i(t) = I_{\rm m} \cos(\omega t + \varphi) = \sqrt{2}I \cos(\omega t + \varphi) \qquad I_{\rm m} = \sqrt{2}I$$
$$u(t) = U_{\rm m} \cos(\omega t + \varphi) = \sqrt{2} U \cos(\omega t + \varphi) \qquad U_{\rm m} = \sqrt{2}U$$



标准正弦波电压信号 \mathbf{u}_1 和整流三角波电压信号 \mathbf{u}_2 之间的有效值大小关系是()

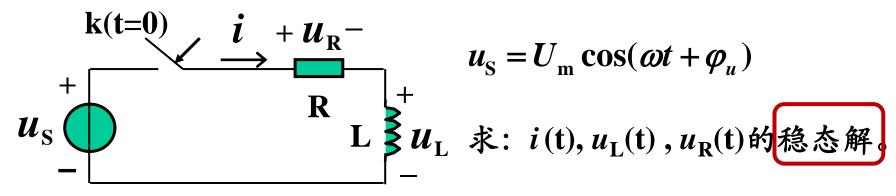


- $U_1 < U_2$

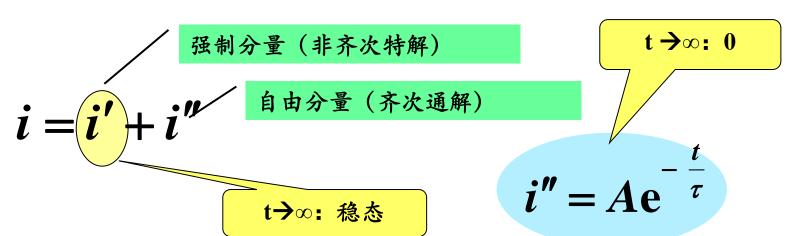


2 正弦稳态分析的关键 → 相量(phasor)

(1) 问题



$$L\frac{\mathrm{d}i}{\mathrm{d}t} + Ri = U_{\mathrm{m}}\cos(\omega t + \varphi_{u})$$
 一阶常系数线性微分方程



求特解/稳态解

查表寻找特解的函数类型

激励 特解类型 $\cos \omega t \longrightarrow C_1 \sin \omega t + C_2 \cos \omega t$ 或 $A \cos(\omega t + B)$ 查表

$$L\frac{\mathrm{d}i}{\mathrm{d}t} + Ri = U_{\mathrm{m}}\cos(\omega t + \varphi_{u})$$
 设特解为
$$i = A\cos(\omega t + B)$$
 代入

$$-LA\omega\sin(\omega t + B) + RA\cos(\omega t + B) = U_{\rm m}\cos(\omega t + \varphi_{\rm u})$$

$$A\sqrt{R^{2} + (\omega L)^{2}} \left(\frac{R}{\sqrt{R^{2} + (\omega L)^{2}}} \cos(\omega t + B) - \frac{\omega L}{\sqrt{R^{2} + (\omega L)^{2}}} \sin(\omega t + B) \right)$$

$$= U_{m} \cos(\omega t + \varphi_{u})$$

$RA\cos(\omega t + B) - LA\omega\sin(\omega t + B) = U_{\rm m}\cos(\omega t + \varphi_{\rm u})$

$$A\sqrt{R^{2} + (\omega L)^{2}} \frac{\cos(\omega t + B) - \omega L}{\cos(\arctan \omega L)} \frac{\sin(\omega t + B)}{\cos(\arctan \omega L)}$$

$$= U_{m} \cos(\omega t + \varphi_{u}) \frac{\omega L}{\cos(\arctan \omega L)} \frac{\cos(\omega t + B + \arctan \omega L)}{\cos(\alpha t + B + \arctan \omega L)}$$

$$\begin{cases} A\sqrt{R^2 + (\omega L)^2} = U_{\rm m} & \longrightarrow \\ B + \arctan(\frac{\omega L}{R}) = \varphi_u & \longrightarrow \end{cases} B = \varphi_u - \arctan(\frac{\omega L}{R}) = \varphi_u - \psi$$

$$i'(t) = \frac{U_{\rm m}}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan \frac{\omega L}{R})$$

 $\cos(\alpha + \beta) = \cos\alpha \cdot \cos\beta - \sin\alpha \cdot \sin\beta$

$$\mathbf{u}_{S} = U_{m} \cos(\omega t + \varphi_{u})$$

$$\mathbf{u}_{S} = U_{m} \cos(\omega t + \varphi_{u})$$

$$\mathbf{u}_{S} = \mathbf{u}_{L} \quad \dot{\mathbf{x}} : \mathbf{i}(t), \mathbf{u}_{L}(t), \mathbf{u}_{R}(t) 的 稳 态 解 .$$

$$u_{\rm S} = U_{\rm m} \cos(\omega t + \varphi_{\rm u})$$

求微分方程特解
$$i'(t) = \frac{U_{\rm m}}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan \frac{\omega L}{R})$$
 麻烦1: 求特解的待定系数

求于 $u'_L(t) = L \frac{\mathrm{d}i'(t)}{\mathrm{d}t} = \frac{\omega L U_{\mathrm{m}}}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan \frac{\omega L}{R} + 90^\circ)$

KCL, KVL 元件约束

麻烦2: 正弦量的微分/积分计算



$$u_R'(t) = u_S - u_L'(t) = \frac{RU_m}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan\frac{\omega L}{R})$$

搞定!!!

麻烦3: 正弦量的土计算

$$\begin{array}{c|c} & i & + u_R - \\ & & & \\ u_S & & & \\ \end{array}$$

$$i'(t) = \frac{U_{\rm m}}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan \frac{\omega L}{R})$$

$$u'_{L}(t) = \frac{\omega L U_{m}}{\sqrt{R^{2} + (\omega L)^{2}}} \cos(\omega t + \varphi_{u} - \arctan \frac{\omega L}{R} + 90^{\circ})$$

$$u_R'(t) = \frac{RU_{pl}}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \varphi_u - \arctan\frac{\omega L}{R})$$



Charles Steinmetz (1865-1923)

3个支路量有何特点?

所有支路电压电流均以相同频率变化!!

接下来……

$$i(t)=I_{\rm m}\cos(\omega t + \varphi)$$

所有支路电压电流均以 相同频率变化!!

(b) 幅值 (I_m)
 (c) 初相角(φ)

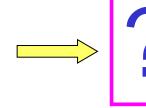
用什么可以同时表示幅值和相位?

问题1: 如何用它来表示正弦量?

求微分方程特解

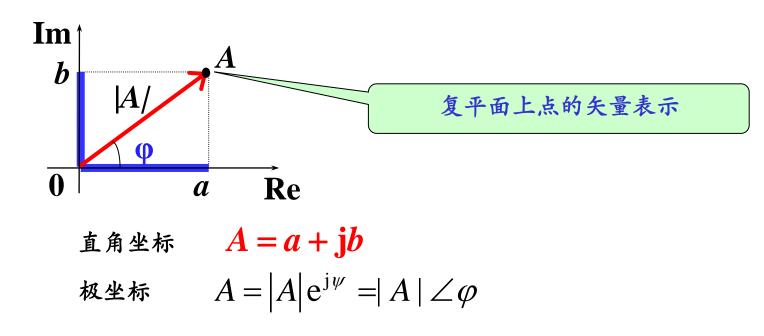
问题2: 正弦量微分/积分

正弦量加减



(2) 复数的复习

(a) 复数的表示形式



特换关系
$$a=|A|\cos arphi$$
 $b=|A|\sin arphi$ $|A|=\sqrt{a^2+b^2}$ $\psi=\arctan \frac{b}{a}$

(b) 复数的运算

$$A_1 \pm A_2 = (a_1 \pm a_2) + j(b_1 \pm b_2)$$

乘除运算——极坐标

$$A_1 \cdot A_2 = |A_1| |A_2| \angle (\varphi_1 + \varphi_2)$$

$$Ae^{j\psi} = |A|e^{j\theta}e^{j\varphi} |A|e^{j(\theta+\varphi)}$$
 二 A逆时针旋转一个角度 φ ,模不变

$$e^{j\frac{\pi}{2}} = \cos\frac{\pi}{2} + j\sin\frac{\pi}{2} = +j$$

$$e^{j(-\frac{\pi}{2})} = \cos(-\frac{\pi}{2}) + j\sin(-\frac{\pi}{2}) = -j$$

$$e^{j(\pm\pi)} = \cos(\pm\pi) + j\sin(\pm\pi) = -1$$

$$-A$$

$$\mathbf{j}A$$
 $+\mathbf{j}A$
 $+\mathbf{j}A$
 \mathbf{k}
 \mathbf{k}
 \mathbf{k}
 \mathbf{k}
 \mathbf{k}
 \mathbf{k}
 \mathbf{k}

+j,-j,-1都可以看成旋转因子

"一乘(j/-j/-1)就转"

(3) 用复数来表示正弦量

复函数
$$A(t) = \sqrt{2}Ie^{j(\omega t + \varphi)}$$

$$= \sqrt{2}I\cos(\omega t + \varphi) + j\sqrt{2}I\sin(\omega t + \varphi)$$

$$Re[A(t)] = \sqrt{2}I\cos(\omega t + \varphi) = i(t)$$

$$A(t) = \sqrt{2}Ie^{j(\omega t + \varphi)}$$

$$A(t) = \sqrt{2} \mathbf{I} e^{j\varphi} e^{j\omega t} = \sqrt{2} \mathbf{\dot{I}} e^{j\omega t}$$

相量Phasor (phase vector)

也可以定义
$$\dot{I}_{\rm m} = \sqrt{2}I \,\mathrm{e}^{\mathrm{j}\varphi}$$

$$u(t) = \sqrt{2}U\cos(\omega t + \theta) \iff \dot{U} = U\angle\theta$$

例1
$$i(t) = 141.4\cos(314t + 30^{\circ})$$
A 用相量表示 $i(t)$, $u(t)$ 。 $u(t) = 311.1\cos(314t - 60^{\circ})$ V



$$\dot{I} = 100 \angle 30^{\circ} \text{A}$$
, $\dot{U} = 220 \angle -60^{\circ} \text{V}$

例2
$$\dot{I}=50\angle15^{\circ}A$$
, $f=50Hz$
写出该相量对应的时间表达式。

$$i(t) = 50\sqrt{2}\cos(314t + 15^\circ) A$$

已知
$$u(t) = 311\cos(314t - 60^{\circ})V$$

求电压相量 Ü

$$\dot{U} = 311 \angle 60^{\circ} \text{ V}$$

$$\dot{U} = 311 \angle -60^{\circ} \text{ V}$$

$$\dot{U} = 220 \angle -60^{\circ} \text{ V}$$

$$\dot{U} = 220 \angle 30^{\circ} \text{ V}$$

(4) 相量的计算

(a) 同频正弦量的"+"和"-"
$$u_1(t) = \sqrt{2} U_1 \cos(\omega t + \varphi_1) = \text{Re}(\sqrt{2}\dot{U}_1 \text{e}^{\text{j}\omega t})$$

$$u_2(t) = \sqrt{2} U_2 \cos(\omega t + \varphi_2) = \text{Re}(\sqrt{2}\dot{U}_2 e^{j\omega t})$$

$$u(t) = u_1(t) + u_2(t) = \operatorname{Re}(\sqrt{2}\dot{U}_1 e^{j\omega t}) + \operatorname{Re}(\sqrt{2}\dot{U}_2 e^{j\omega t})$$
$$= \operatorname{Re}(\sqrt{2}\dot{U}_1 e^{j\omega t} + \sqrt{2}\dot{U}_2 e^{j\omega t}) = \operatorname{Re}[\sqrt{2}(\dot{U}_1 + \dot{U}_2) e^{j\omega t}]$$

$$\dot{\boldsymbol{U}} = \dot{\boldsymbol{U}}_1 + \dot{\boldsymbol{U}}_2$$

时间域
$$\mathbf{u}_1(t) \pm \mathbf{u}_2(t) = \mathbf{u}_3(t)$$

$$x^{1.5} = 4$$

$$x = 2.52$$

变换域



类比





相量域
$$\dot{U}_1 \pm \dot{U}_2 = \dot{U}_3$$

$$1.5 * \lg x = \lg 4 \implies \lg x = 0.40$$

(2) 正弦量的微分和积分

微分/积分 关系



代数关系

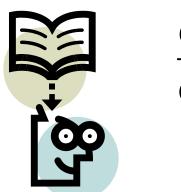
$$i(t) = \sqrt{2}I\cos(\omega t + \psi_i) = \text{Re}(\sqrt{2}\dot{I}e^{j\omega t})$$

微分

$$\frac{di}{dt} = \frac{d}{dt} \left(\text{Re}(\sqrt{2} \dot{I} e^{j\omega t}) \right)$$

$$= \text{Re}\left(\frac{d}{dt} (\sqrt{2} \dot{I} e^{j\omega t}) \right)$$

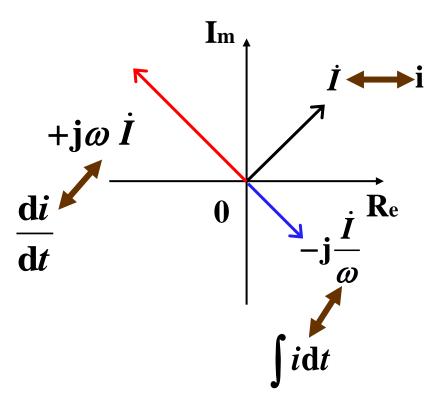
$$= \text{Re}(\sqrt{2} j\omega \dot{I} e^{j\omega t})$$



$$\frac{\mathrm{d}i}{\mathrm{d}t} \to \mathrm{j}\omega\dot{I}$$

"一乘就转"

积分 $\int i dt \to \frac{I}{j\omega}$



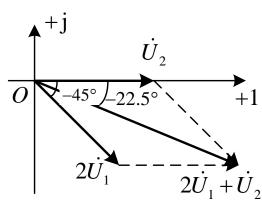
例4. 2-1 已知 $u_1(t) = \sqrt{2}\sin(2t + 45^\circ)V$ $u_2(t) = 2\sqrt{2}\cos(2t)V$ 求 $2u_1(t) + u_2(t)$, $u_1(t) - u_2(t)$, $\frac{d}{dt}u_1(t) + u_2(t)$, 并画出各相量图 u_1 和 u_2 为同频率正弦量,因此可以采用相量法进行分析

$$u_1(t) = \sqrt{2}\sin(2t + 45^\circ) = \sqrt{2}\cos(2t - 45^\circ) V \leftrightarrow \dot{U}_1 = 1\angle - 45^\circ V$$
$$u_2(t) = 2\sqrt{2}\cos(2t) V \leftrightarrow \dot{U}_2 = 2\angle 0^\circ V$$

(1)
$$2u_1 + u_2 \leftrightarrow 2\dot{U}_1 + \dot{U}_2 = 2\angle -45^{\circ}V + 2\angle 0^{\circ}$$

= $3.414 - j1.414 = 3.695\angle -22.5^{\circ}V$

$$2u_1 + u_2 = 3.695\sqrt{2}\cos(2t - 22.5^\circ)V$$



例4. 2-1 已知
$$u_1(t) = \sqrt{2}\sin(2t + 45^\circ)V$$
 $u_2(t) = 2\sqrt{2}\cos(2t)V$ 求 $2u_1(t) + u_2(t)$, $u_1(t) - u_2(t)$, $\frac{d}{dt}u_1(t) + u_2(t)$,并画出各相量图

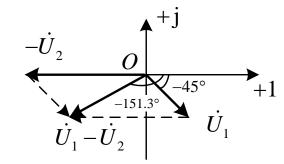
 \mathbf{m} u_1 和 u_2 为同频率正弦量,因此可以采用相量法进行分析

$$u_1(t) = \sqrt{2}\sin(2t + 45^\circ) = \sqrt{2}\cos(2t - 45^\circ) \text{V} \leftrightarrow \dot{U}_1 = 1\angle -45^\circ \text{V}$$
$$u_2(t) = 2\sqrt{2}\cos(2t) \text{V} \leftrightarrow \dot{U}_2 = 2\angle 0^\circ \text{V}$$

(2)

$$u_1 - u_2 \leftrightarrow \dot{U}_1 - \dot{U}_2 = 1 \angle -45^{\circ} - 2 \angle 0^{\circ} = -1.293 - j0.707 = 1.474 \angle -151.3^{\circ} V$$

$$u_1 - u_2 = 1.474\sqrt{2}\cos(2t - 151.325^\circ)V$$



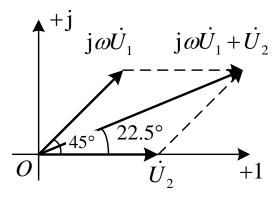
例4. 2-1 已知 $u_1(t) = \sqrt{2}\sin(2t + 45^\circ)V$ $u_2(t) = 2\sqrt{2}\cos(2t)V$ 求 $2u_1(t) + u_2(t)$, $u_1(t) - u_2(t)$, $\frac{d}{dt}u_1(t) + u_2(t)$,并画出各相量图

 \mathbf{m} u_1 和 u_2 为同频率正弦量,因此可以采用相量法进行分析

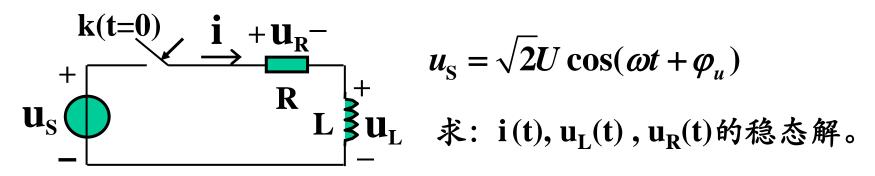
$$u_1(t) = \sqrt{2}\sin(2t + 45^\circ) = \sqrt{2}\cos(2t - 45^\circ) \text{V} \leftrightarrow \dot{U}_1 = 1\angle - 45^\circ \text{V}$$
$$u_2(t) = 2\sqrt{2}\cos(2t) \text{V} \leftrightarrow \dot{U}_2 = 2\angle 0^\circ \text{V}$$

(3)
$$\frac{d}{dt}u_1 + u_2 \leftrightarrow j\omega \dot{U}_1 + \dot{U}_2 = j2 \times 1 \angle -45^\circ + 2\angle 0^\circ = 3.695\angle 22.5^\circ V$$

$$\frac{d}{dt}u_1 + u_2 = 3.695\sqrt{2}\cos(2t + 22.5^\circ)V$$



(5) 相量的应用





$$u_{\rm S}(t) = Ri(t) + L \frac{\mathrm{d}i(t)}{\mathrm{d}t} \implies \dot{U} = R\dot{I} + \mathrm{j}\omega L\dot{I}$$

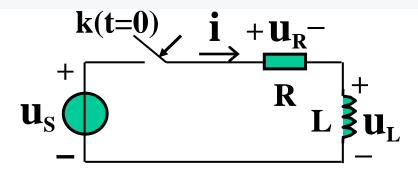
$$\dot{I} = \frac{\dot{U}}{R + j\omega L} = \frac{U \angle \varphi_u}{\sqrt{R^2 + \omega^2 L^2} \angle \arctan \frac{\omega L}{R}} = \frac{U}{\sqrt{R^2 + \omega^2 L^2}} \angle (\varphi_u - \arctan \frac{\omega L}{R})$$

$$u_L = L \frac{\mathrm{d}i_L}{\mathrm{d}t}$$
 $\Longrightarrow \dot{U}_L = \mathbf{j}\omega L\dot{I} = \frac{\omega LU}{\sqrt{R^2 + \omega^2 L^2}} \angle (\varphi_u - \arctan\frac{\omega L}{R} + 90^\circ)$



$$u_R = Ri_L \Longrightarrow \dot{U}_R = R\dot{I} = \frac{RU}{\sqrt{R^2 + \omega^2 L^2}} \angle (\varphi_u - \arctan \frac{\omega L}{R})$$

搞定!!!



已知 $u_S(t) = \cos 2t \text{ V}$ L=0.5H, R=1 Ω 求i(t) 的稳态解。

- $i(t) = \cos 2t A$
- $i(t) = \cos(2t 45^{\circ}) A$
- $i(t) = \frac{1}{\sqrt{2}}\cos(2t 45^{\circ}) A$
- $i(t) = \frac{1}{\sqrt{2}}\cos(2t + 45^{\circ}) A$

求解顺序

- ●列写 ODE
- ●将ODE变换为复系数代数方程
- ●求解复系数代数方程
- ●反变换得到时间表达式

$$u(t) = Ri(t) + L\frac{di(t)}{dt} \qquad i(t) = \frac{\sqrt{2}U}{\sqrt{R^2 + \omega^2 L^2}} \cos(\omega t + \varphi_u - \arctan\frac{\omega L}{R})$$

$$\dot{U} = R\dot{I} + j\omega L\dot{I} \qquad \rightarrow \dot{I} = \frac{\dot{U}}{R + j\omega L} = \frac{U}{\sqrt{R^2 + \omega^2 L^2}} \angle(\varphi_u - \arctan\frac{\omega L}{R})$$

•下节课讨论如何直接列写复系数代数方程!!

习题: 4-1, 4-2, 4-3; 提交截止时间: 周五(4月30日) 早8点