

Communication acoustics Ch 3: Signal processing and signals

Ville Pulkki and Matti Karjalainen

Department of Signal Processing and Acoustics Aalto University, Finland

September 13, 2017

Sound as signal

In signal representations a physical or abstract variable is typically represented as a function of time, such as:

- Signal as a mathematical function:
 - Pure tone:

$$y(t) = A\sin(2\pi ft) = A\sin(\omega t)$$

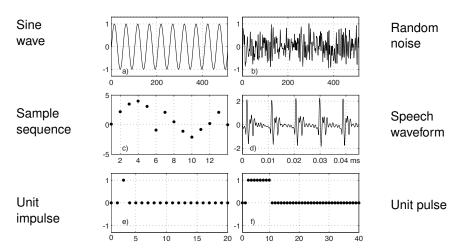
■ Random signal:

$$n(t) = \operatorname{rand}(t)$$

■ Discrete-time numeric sequence

$$x(n) = [0.1 \ 2.2 \ 3.5 \ 4.0 \ 3.1 \ -0.9 \ 2.1 \ 0.5 \ -1.1 \ -2.1 \ -0.8 \ 0.2]$$

Graphical presentations of signals



Linear and time-invariant (LTI) systems

$$\begin{array}{c|cccc} x(t) & X(\omega) & h(t) & <-> H(\omega) & y(t) & Y(\omega) \\ \hline & & & \\ x(n) & X(k) & h(n) & <-> H(k) & y(n) & Y(k) \end{array}$$

Properties of LTI systems

- Any (stable) LTI system can be fully represented by its impulse response
- Output cannot include any frequencies that are not in the input (no nonlinear distortion)
- Any bandlimited LTI system can be approximated by digital filters with arbitrary accuracy (theoretically)

Signal processing algorithms

Convolution

$$y(t) = x(t) * h(t) = \int_{-\infty}^{+\infty} x(\tau) h(t - \tau) d\tau$$
$$y(n) = x(n) * h(n) = \sum_{i=-\infty}^{+\infty} x(i) h(n - i)$$

Fourier analysis

$$X(\omega) = \mathcal{F}\{x(t)\} = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$
$$X(k) = \mathcal{F}_{d}\{x(n)\} = \sum_{n=1}^{N-1} x(n) e^{-jk(2\pi/N)n}$$

Signal processing algorithms

Fourier synthesis

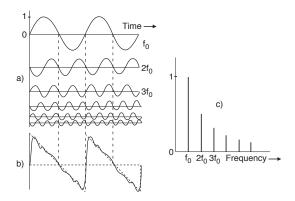
$$x(t) = \mathcal{F}^{-1}\{X(\omega)\} = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(\omega) e^{j\omega t} d\omega$$
$$x(n) = \mathcal{F}_{d}^{-1}\{X(k)\} = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{jk(2\pi/N)n}$$

Convolution and Fourier transform

$$\mathcal{F}\{x(t) * y(t)\} = X(\omega) \cdot Y(\omega)$$

$$\mathcal{F}_{\mathrm{d}}\{x(n)*y(n)\}=X(k)\cdot Y(k)$$

Decomposition of sawtooth waveform



Spectrum analysis

Magnitude spectrum

$$|X(\omega)|_{\mathrm{dB}} = 20 \log_{10} |X(\omega)|$$

$$|X(k)|_{\mathrm{dB}} = 20 \log_{10} |X(k)|$$

Phase spectrum

$$\varphi(\omega) = \angle X(\omega) = \arg\{X(\omega)\}\$$

$$\varphi(k) = \angle X(k) = \arg\{X(k)\}\$$

- lacksquare Phase delay $au_{
 m p}(\omega) = -arphi(\omega)/\omega$
- lacktriangle Group delay $au_{
 m g}(\omega) = -{
 m d} arphi(\omega)/{
 m d} \omega$

Fourier analysis with windowing

$$X(\omega) = \int_{t_b}^{t_e} w(t) x(t) e^{-j\omega t} dt$$

$$X(k) = \sum_{n_e}^{n_e} w(n) x(n) e^{-jk(2\pi/N)n}$$

- Rectangular window
- Hamming window
- Hann(ing) window
- Kaiser window
- Blackman (Blackman-Harris) window

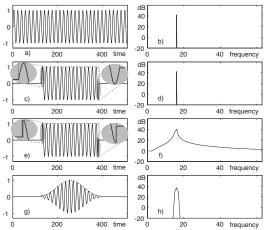
Spectrum analysis using Fourier analysis with windowing

Sine wave

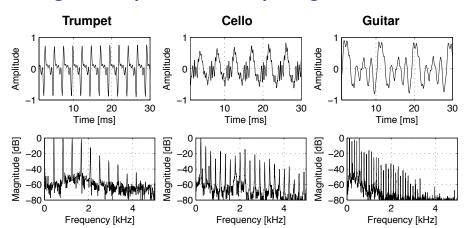
Sine wave windowed synchronously

Sine wave windowed non-synchronously

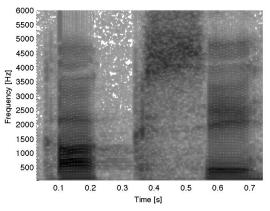
Sine wave Hammingwindowed



Magnitude spectra of example signals

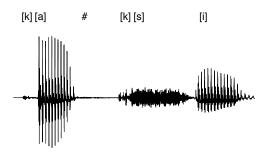


Spectrogram



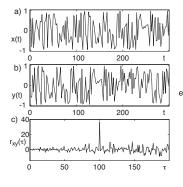
Windowed Fourier magnitude spectrum shown for each time position

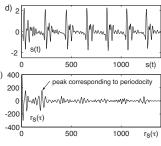
Corresponding waveform



Auto- and cross-correlation

$$r_{xy}(\tau) = \int_{-\infty}^{+\infty} x(t) y(t+\tau) d\tau$$
$$r_{xy}(k) = \sum_{i=0}^{N-1} x(i) y(i+k)$$





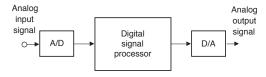
Cepstrum

$$c_{x}(t) = \mathcal{F}^{-1}\{\log |\mathcal{F}\{x(t)\}|\}$$

Commonly used in speech recognition as feature vector.

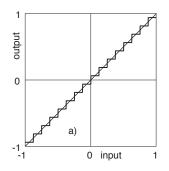
- Compute Fourier transform
- Logarithm of magnitude spectrum
- Inverse Fourier transform
- "Spectrum of the curve of magnitude spectrum"

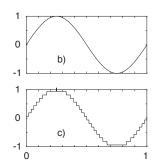
Digital audio signal processing



- Analog-to-digital (A/D) converter
- Digital signal processor
- Digital signal processing (DSP) software
- Digital-to-analog (D/A) converter

Signal quantization in A/D conversion





- Linear quantization (PCM-coding)
- Discrete levels: 2ⁿ (n= number of bits)
- 16–24 bits/sample in audio (≥ 96 dB SNR)
- Sample rate: e.g. 44100 or 48000 samples/s

Filtering

Filters are DSP components that have frequency-dependent magnitude and/or phase response. Needed often in audio techniques.

- low-pass filter, attenuate high frequencies above cutoff frequency
- high-pass filter, attenuate low frequencies
- band-pass filter, attenuate low and high, and leave a band unmodified
- band-reject filter, correspondingly
- all-pass filter, modify only phase response
- arbitrary-response filter, design the response as needed for each frequency

Z-transform

Linear transform of sequence

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

Unit delay as building element:

$$\mathcal{Z}\{x(n-1)\}=z^{-1}X(z)$$

Digital filtering can be expressed as a polynomial of z^{-1}

Digital filtering: FIR filters

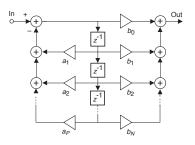
FIR = Finite impulse response

$$H_{FIR}(z) = \sum_{n=0}^{N-1} b_n z^{-n} = b_0 + b_1 z^{-1} + \dots + b_{N-1} z^{-(N-1)}$$

Digital filtering: IIR filters

IIR = Infinite impulse response

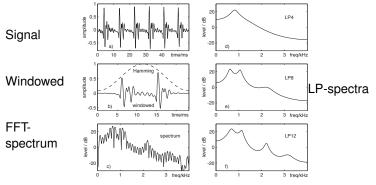
$$H_{\rm IIR}(z) = \frac{\sum_{n=0}^{N-1} b_n z^{-n}}{1 + \sum_{p=1}^{P-1} a_p z^{-p}} = \frac{b_0 + b_1 z^{-1} + \dots + b_{N-1} z^{-(N-1)}}{1 + a_1 z^{-1} + \dots + a_{P-1} z^{-(P-1)}}$$



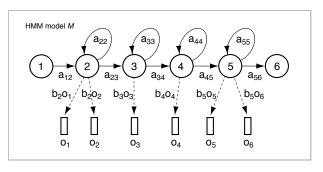
Linear prediction (AR-modeling)

Modeling of signal generation with flat spectrum excitation (impulse or noise) and IIR (all-pole) filter.

Speech example:



Hidden Markov models (HMM)



- For probabilistic modeling of state sequences
- Used especially in speech recognition and synthesis

References

These slides follow corresponding chapter in: Pulkki, V. and Karjalainen, M. Communication Acoustics: An Introduction to Speech, Audio and Psychoacoustics. John Wiley & Sons, 2015, where also a more complete list of references can be found.