

誠,樸,精,勤



Matlab HW9: MMSE and Bayesian Estimation

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1. Problem:

• Extract a random signal from noise. The signal is assumed to be the outcome of the random process.

2. Application area example:

 A typical use is to enhance images corrupted by noise, which requires a straightforward extension to a two-dimensional signal. An example is in high resolution electron microscope images.

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Estimation of Random Signal in Noise

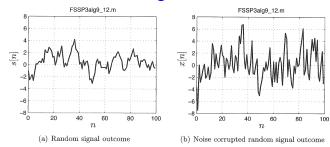


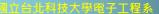
Figure 1: Outcome of random signal and observed noisy data. The points have been connected by straight lines for easier viewing.

3. Data model/assumptions:

The Gaussian signal to be extracted s[n] is assumed to be embedded in WGN with known variance σ^2 ,

$$x[n] = s[n] + w[n]$$
 $n = 0, 1, ..., N - 1$

• s[n] is assumed to be the outcome of a zero mean Gaussian random process with a known $N \times N$ covariance matrix C_s .





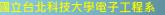
4. Estimator: The MMSE estimator is given as

$$\hat{\mathbf{s}} = \mathbf{C}_{\mathbf{s}} (\mathbf{C}_{\mathbf{s}} + \sigma^2 \mathbf{I})^{-1} \mathbf{x} \tag{1}$$

- $\hat{\mathbf{s}} = [\hat{\mathbf{s}}[0] \; \hat{\mathbf{s}}[1] \; ... \; \hat{\mathbf{s}}[N-1]]^T$
- If the signal is from a stationary random process, then the covariance matrix becomes an autocorrelation matrix \mathbf{R}_s given by

$$\mathbf{C}_{s} = \mathbf{R}_{s} = \begin{bmatrix} r_{s}[0] & r_{s}[1] & \dots & r_{s}[N-1] \\ r_{s}[1] & r_{s}[0] & \dots & r_{s}[N-2] \\ \vdots & \vdots & \ddots & \vdots \\ r_{s}[(N-1)] & r_{s}[N-2] & \dots & r_{s}[0] \end{bmatrix}$$

where $r_s[k]$ is the autocorrelation sequence for s[n].





Theorem 1 (MMSE estimator)

$$\mathbf{x} = \mathbf{H}\theta + \mathbf{w}$$

where $\mathbf{x} = [x[0] \ x[1] \ ... \ x[N-1]]^T$. The Bayesian linear model the posterior PDF is given by

$$\rho(\boldsymbol{\theta}|\mathbf{x}) = \frac{1}{(2\pi)^{\textit{N}/2}\textit{det}^{1/2}(\mathbf{C}_{\boldsymbol{\theta}|\mathbf{x}})} \textit{exp} \bigg[-\frac{1}{2} (\boldsymbol{\theta} - \textit{E}[\boldsymbol{\theta}|\mathbf{x}])^T \mathbf{C}_{\boldsymbol{\theta}|\mathbf{x}}^{-1} (\boldsymbol{\theta} - \textit{E}[\boldsymbol{\theta}|\mathbf{x}]) \bigg]$$

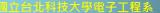
where the mean is

$$E[\theta|\mathbf{x}] = \mu_{\theta} + \left(\mathbf{C}_{\theta}^{-1} + \frac{1}{\sigma^{2}}\mathbf{H}^{T}\mathbf{H}\right)^{-1}\frac{\mathbf{H}^{T}}{\sigma^{2}}(\mathbf{x} - \mathbf{H}\mu_{\theta})$$
$$= \mu_{\theta} + \mathbf{C}_{\theta}\mathbf{H}^{T}\left(\mathbf{H}\mathbf{C}_{\theta}\mathbf{H}^{T} + \mathbf{C}_{\mathbf{w}}\right)^{-1}(\mathbf{x} - \mathbf{H}\mu_{\theta})$$

and the covariance matrix is

$$\mathbf{C}_{oldsymbol{ heta}|\mathbf{x}} = \mathbf{C}_{oldsymbol{ heta}} - \mathbf{C}_{oldsymbol{ heta}} \mathbf{H}^{\mathsf{T}} (\mathbf{H} \mathbf{C}_{oldsymbol{ heta}} \mathbf{H}^{\mathsf{T}} + \mathbf{C}_{oldsymbol{w}})^{-1} \mathbf{H} \mathbf{C}_{oldsymbol{ heta}}$$

The MMSE estimator is $\theta = E[\theta|\mathbf{x}]$.





Back to the problem of estimation of random signal in noise

$$\mathbf{x} = \mathbf{s} + \mathbf{w}$$

Compare to the $\mathbf{x} = \mathbf{H}\theta + \mathbf{w}$, the MMSE estimator is

$$E[\theta|\mathbf{x}] = \mu_{\theta} + \mathbf{C}_{\theta}\mathbf{H}^{T}\Big(\mathbf{H}\mathbf{C}_{\theta}\mathbf{H}^{T} + \mathbf{C}_{w}\Big)^{-1}(\mathbf{x} - \mathbf{H}\mu_{\theta})$$

Therefore, we let

$$\mathbf{H} = \mathbf{I}$$

$$\boldsymbol{\mu}_{\theta} = \begin{bmatrix} 0 \ 0 \ \dots \ 0 \end{bmatrix}^{\mathsf{T}} = \mathbf{0}$$

$$\mathbf{C}_{\theta} = \mathbf{C}_{\mathsf{S}}$$

$$\mathbf{C}_{\mathsf{W}} = \sigma^{2} \mathbf{I}$$

Then.

$$\hat{\mathbf{s}} = \boldsymbol{E}[\boldsymbol{\theta}|\mathbf{x}] = \mathbf{C}_{\boldsymbol{s}}(\mathbf{C}_{\boldsymbol{s}} + \sigma^2 \mathbf{I})^{-1}\mathbf{x}$$





For this case, if the PSD is given by P_s(f), then an approximate estimate (which avoids the inversion of the N × N autocorrelation matrix) is to filter the data x[n] with the noncausual filter with frequency response

$$H(f) = \frac{P_s(f)}{P_s(f) + \sigma^2}$$

It is called a Wiener smoothing filter.





5. Performance:

- The MMSE estimator is optimal in that minimizes the Bayesian mean square error of the estimator, assuming the signal and noise are Gaussian
- The minimum mean square error for the estimator \$\hat{s}[n]\$ in (1) for \$n = 0, 1, ..., N-1\$ is the \$[n, n]\$ diagonal element of the \$N \times N\$ mean square error matrix

$$\mathbf{M}_{\hat{\mathbf{s}}} = \mathbf{C}_{\mathbf{s}} - \mathbf{C}_{\mathbf{s}} (\mathbf{C}_{\mathbf{s}} + \sigma^2 \mathbf{I})^{-1} \mathbf{C}_{\mathbf{s}}$$





6. Example:

 Consider a random signal that is an outcome of a Gaussian AR random process of order one

$$s[n] = -a[1]s[n-1] + u[n]$$

- where a[1] = -0.9 and $\sigma_u^2 = 1$
- It is embedded in WGN w[n] with variance $\sigma^2 = 5$

$$x[n] = s[n] + w[n] = (-a[1]s[n-1] + u[n]) + w[n]$$

- The signal and the noise corrupted signal are shown as Fig. 1
- The signal is an outcome of a stationary random process so that we can estimate it using the (1) with the covariance matrix replaced by the autocorrelation matrix. The autocorrelation sequence is given by

$$r_s[k] = \frac{\sigma_u^2}{1 - a^2[1]} (-a[1])^{|k|}$$





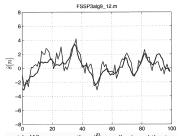
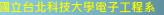


Figure 2: Estimated signal using the matrix Wiener smoother (Heavy line) and the true signal (light signal). The points have been connected by straight lines for easier viewing

- The Wiener smoothed signal along with the true signal is shown in Fig. 2
- It is closer to the true signal than noise corrupted signal but exhibits some smoothing





6. Explanation:

- The Wiener smoother cannot separate out the signal from the noise perfectly since the PSDs overlap in frequency.
- It attempts to reduce the WGN as much as possible in an effort to reduce the overall Bayesian mean square error.
- This result in reducing the power in some of the high frequency signal bands, resulting in a smoothed estimate.





Example 2 (Estimation of time delay)

Try to plot Fig. 2 using the program **FSSP3alg9_12.m** and understand how to implement MMSE for the random signal estimation problem. Try to use different noise variance and compare the result:

- $\sigma^2 = 5$ as in Fig. 2
- A higher SNR case: $\sigma^2 = 1$

