Outline Digital control introduction Difference equations Stability Sysquake

# Difference equations

- Digital control introduction
  - Discrete-time control of discrete-time systems
  - Discrete-time control of continuous-time systems
- 2 Difference equations
  - Operator descriptions
  - Poles and zeros
- Stability
  - Criterion
  - Steady-state gain
  - Transient responses
- Sysquake
  - Simulation of a difference equation
  - Interactive analysis

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### Discrete-time control

### Digital system

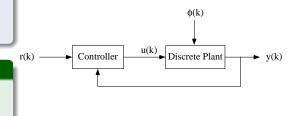
Discrete-time u(k), y(k),

$$k = 0, 1, 2, \dots$$

Discrete values a signal can take (finite precision)

### Example

Car engine control u(k) = fuel injection time y(k) = air/fuel ratio k = combustion cycle



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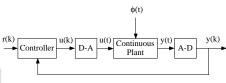
# Discrete-time control of continuous-time systems

- analogue signals u(t), y(t)
- real time  $t \in \Re$
- discrete-time signals u(k), y(k)
- discrete time  $k = 0, 1, 2, \dots$

#### Example

Tank level control u(k) = pump voltage y(k) = tank level

k = index of sampling instant



Discrete-time system includes D-A and A-D converters

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## Linear difference equations

A linear difference equation describing a signal y

$$y(k) = -a_1y(k-1) - a_2y(k-2) - \ldots - a_ny(k-n)$$

With input u and output y

$$y(k) = -a_1y(k-1) - \dots + a_ny(k-n) + b_0u(k) + b_1u(k-1) + \dots$$

With two inputs,  $u_1$  and  $u_2$ , and one output y

$$y(k) = -a_1y(k-1)\dots b_0u_1(k) + b_1u_1(k-1) + \dots c_0u_2(k) + c_1u_2(k-1) + \dots$$

where  $a_1, a_2, \ldots, b_0, b_1, \ldots, c_0, c_1, \ldots$  are real constants

# Example

$$y(k) = -0.9y(k-1) + 0.1u(k-1)$$

Initial conditions y(0) = u(0) = 0Response of the system for the input u(1) = 1, u(2) = -1

$$y(1) = -0.9y(0) + 0.1u(0) = -0.9 \cdot 0 + 0.1 \cdot 0 = 0$$
  

$$y(2) = -0.9y(1) + 0.1u(1) = -0.9 \cdot 0 + 0.1 \cdot 1 = 0.1$$
  

$$y(3) = -0.9y(2) + 0.1u(2) = -0.9 \cdot 0.1 + 0.1 \cdot (-1) = -0.19$$

# Shift operators

### Backward-shift operator

$$q^{-1}y(k) = y(k-1)$$
  
implementable (memory)

### Example

$$y(k) = -0.9y(k-1) + 0.1u(k-1)$$
  

$$y(k) = -0.9q^{-1}y(k) + 0.1q^{-1}u(k)$$
  

$$(1 + 0.9q^{-1})y(k) = 0.1q^{-1}u(k)$$
  

$$A(q^{-1})y(k) = B(q^{-1})u(k)$$

# Shift operators

### Backward-shift operator

$$q^{-1}y(k) = y(k-1)$$
 implementable (memory)

### Forward-shift operator

$$qy(k) = y(k+1)$$
  
not implementable (future)

### Example

$$y(k) = -0.9y(k-1) + 0.1u(k-1)$$

$$y(k) = -0.9q^{-1}y(k) + 0.1q^{-1}u(k)$$

$$(1 + 0.9q^{-1})y(k) = 0.1q^{-1}u(k)$$

$$A(q^{-1})y(k) = B(q^{-1})u(k)$$

#### Example

$$y(k) = -0.9y(k-1) + 0.1u(k-1)$$

$$y(k+1) = -0.9y(k) + 0.1u(k)$$

$$(q+0.9)y(k) = 0.1u(k)$$

$$\bar{A}(q)y(k) = \bar{B}(q)u(k)$$

## Polynomial description

A difference equation in polynomial form

$$A(q^{-1})y(k) = B(q^{-1})u(k),$$
 
$$\begin{cases} A(q^{-1}) = 1 + a_1q^{-1} + a_2q^{-2} + \dots \\ B(q^{-1}) = b_0 + b_1q^{-1} + b_2q^{-2} + \dots \end{cases}$$

or, for brevity,

$$Ay = Bu$$

useful for analysis.

For calculation of responses, use recursive form

$$y(k) = -a_1y(k-1)... + b_0u(k) + b_1u(k-1)...$$

## Transfer operator

Input-output description

$$u \longrightarrow G \longrightarrow Y$$
 input output

$$y(k) = G(q^{-1})u(k), \quad G(q^{-1}) = \frac{B(q^{-1})}{A(q^{-1})}$$

Transfer operator G interpreted as difference equation

$$y(k) = -a_1y(k-1) - a_2y(k-2) - \ldots + b_0u(k) + b_1u(k-1)\ldots$$

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## Poles and zeros

$$y = G(\mathbf{q}^{-1})u = \bar{G}(\mathbf{q})u$$

Forward-shift description

#### zero

z is a zero if  $\bar{G}(z) = 0$ 

#### pole

 $\lambda$  is a pole if  $\bar{G}(\lambda) \to \infty$ 

### Example

$$y(k) = 0.9y(k-1) + u(k) + u(k-1)$$
  
 $\bar{G}(q) = \frac{q+1}{q-0.9} \Rightarrow \begin{cases} \lambda_1 = 0.9 \\ z_1 = -1 \end{cases}$ 

# Example

Backward-shift description

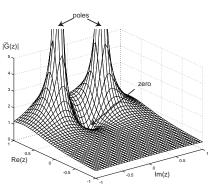
$$A(\mathbf{q}^{-1})y(k) = B(\mathbf{q}^{-1})u(k)$$

with 
$$\lambda_{1,2} = 0.7 \pm 0.4i$$
,  $z_1 = 0.5$ 

$$\begin{cases} A(q^{-1}) = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1}) \\ B(q^{-1}) = q^{-1}(1 - z_1 q^{-1}) \end{cases}$$

Forward-shift description

$$ar{G} = rac{\mathrm{q} - z_1}{(\mathrm{q} - \lambda_1)(\mathrm{q} - \lambda_2)}$$



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## Stability definition

$$y(k) + a_1y(k-1) + \dots a_ny(k-n) = 0$$

The system is stable if

$$y(k) \to 0, k \to \infty$$

for all initial conditions

## Stability definition

$$y(k) + a_1y(k-1) + \dots a_ny(k-n) = 0$$

### The system is stable if

$$y(k) \rightarrow 0, k \rightarrow \infty$$

for all initial conditions

#### Example

If 
$$A(q^{-1}) = 1 - \lambda q^{-1}$$
, the general solution is  $y(k) = y(0)\lambda^k$ 

$$A(q^{-1})y(k) = y(k) - \lambda y(k-1) = y(0)\lambda^k - \lambda y(0)\lambda^{k-1} = 0$$

Stable if and only if  $|\lambda| < 1$ 

## Second order system

$$A(q^{-1})y(k) = y(k) + a_1y(k-1) + a_2y(k-2) = 0$$

Factorize in poles

$$A(q^{-1}) = 1 + a_1 q^{-1} + a_2 q^{-2} = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1})$$

General solution

$$y(k) = c_1 \lambda_1^k + c_2 \lambda_2^k$$

since 
$$A_i = 1 - \lambda_i q^{-1}$$
 and  $y_i(k) = c_i \lambda_i^k$  satisfies  $A_i y_i = 0$ 

$$Ay = A_1A_2(y_1 + y_2) = A_2\underbrace{[A_1y_1]}_{=0} + A_1\underbrace{[A_2y_2]}_{=0} = 0$$

Stable if and only if  $|\lambda_i| < 1$ , i = 1, 2

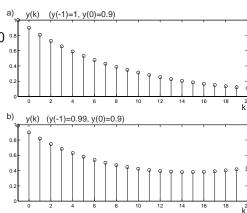
# Example

$$y(k)-2y(k-1)+0.99y(k-2)=0$$

Initial conditions y(0) = 0.9 and

a) 
$$y(-1) = 1$$

b) 
$$y(-1) = 0.99$$



Thus, not stable!

# Example

$$y(k)-2y(k-1)+0.99y(k-2)=0$$

Initial conditions y(0) = 0.9 and

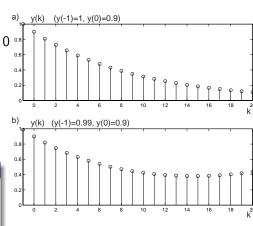
a) 
$$y(-1) = 1$$

b) 
$$y(-1) = 0.99$$

Poles: 
$$\lambda_1 = 0.9$$
,  $\lambda_2 = 1.1$ 

$$y(k) = c_1 \lambda_1^k + c_2 \lambda_2^k$$

In a)  $c_2 = 0$ , but not in b)



Thus, not stable!

### General case

Difference equation

$$A(\mathbf{q}^{-1})y(k) = 0$$

factorized in poles

$$A(q^{-1}) = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1}) \dots (1 - \lambda_n q^{-1})$$

### Stability criterion

$$|\lambda_i| < 1, \quad i = 1, 2, \dots, n$$

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# Steady-state (stationary) gain

Unit step

$$u(k) = \begin{cases} 1 & k \ge 0 \\ 0 & k < 0 \end{cases}$$

Step response

$$y(k) + a_1y(k-1) + \dots + a_ny(k-n) = b_0 \underbrace{u(k)}_{=1} + b_1 \underbrace{u(k-1)}_{=1} +$$

If stable  $y(k) \to y_{\infty}$ ,  $k \to \infty$ 

$$y_{\infty} = \frac{b_0 + b_1 + \dots}{1 + a_1 + a_2 + \dots} = \frac{B(1)}{A(1)} = G(1)$$

#### Steady-state gain

If stable, then 
$$\frac{y_{\infty}}{u_{\infty}} = G(1)$$

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# Example: FIR

Unit pulse (impulse)

$$u(k) = \begin{cases} 1 & k = 0 \\ 0 & k \neq 0 \end{cases}$$

Pulse response for y(k) = 5u(k) + 2u(k-1)

$$y(0) = 5u(0) + 2u(-1) = 5 \cdot 1 + 2 \cdot 0 = 5$$
  

$$y(1) = 5u(1) + 2u(0) = 5 \cdot 0 + 2 \cdot 1 = 2$$
  

$$y(2) = 5u(2) + 2u(1) = 5 \cdot 0 + 2 \cdot 0 = 0$$
  

$$y(k) = 0, \quad k > 1$$

Finite Impulse Response (FIR)

Pulse response for 
$$y(k) = b_0 u(k) + b_1 u(k-1) + \ldots + b_m u(k-m)$$

$$y(k) = b_k$$

## Example: Step response

#### Example

Step response for 
$$y(k) = b_0 u(k) + b_1 u(k-1) + \ldots + b_m u(k-m)$$

$$y(0) = b_0$$

$$y(1) = b_0 + b_1$$

$$\vdots$$

$$y(m) = b_0 + b_1 + \ldots + b_m = B(1)$$

$$y(k) = B(1) \quad k \ge m$$

# Example: First order system

### Example

Step response for 
$$y(k) = 0.9y(k-1) + 0.1u(k)$$

$$y(0) = 0.1$$

$$y(1) = 0.9 \cdot 0.1 + 0.1 = 0.19$$

$$y(2) = 0.9 \cdot 0.19 + 0.1 = 0.271$$

$$\vdots$$

$$y(k) \to \frac{B(1)}{A(1)} = 1, \quad k \to \infty$$

# Example: First order system

### Example

Step response for 
$$y(k) = 0.9y(k-1) + 0.1u(k)$$
  

$$y(0) = 0.1$$

$$y(1) = 0.9 \cdot 0.1 + 0.1 = 0.19$$

$$y(2) = 0.9 \cdot 0.19 + 0.1 = 0.271$$

$$\vdots$$

### Compare analytical solution

$$y(k) = c_1 \lambda_1^k + 1$$
  
 $y(0) = c_1 0.9^0 + 1 = 0.1 \rightarrow c_1 = -0.9$   
 $\rightarrow y(k) = -0.9 \cdot 0.9^k + 1$ 

 $y(k) \to \frac{B(1)}{A(1)} = 1, \quad k \to \infty$ 

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## Simulation of a difference equation

Implement difference equation

$$y(k) = -a_1y(k-1) + b_1u(k-1)$$

#### Edit the text file:

and open it in Sysquake

## Simulation of a difference equation

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$$y(k) = -a_1y(k-1) + b_1u(k-1)$$

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#### Test in command window

## Example: compare with built-in function

```
Test in command window

> global y1 u1 a1 b1

> y1=0;u1=0;a1=-0.9;b1=0.1;

> k=0:19;

> for n=1:20, y(n)=mydiff(1); end;

> plot(k,y,'ro')

> A=[1 a1]; B=[0 b1];

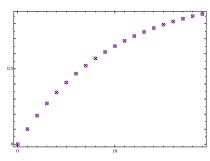
> yf=filter(B,A,ones(1,20));

> plot(k,yf,'bx')
```

# Example: compare with built-in function

```
Test in command window
> global y1 u1 a1 b1
> y1=0;u1=0;a1=-0.9;b1=0.1;
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> plot(k,yf,'bx')
```

### Result in **plot window**



## Example: step response of second order system

#### Test in command window

```
> lambda1=0.9*exp(i*pi/6)
Lambda1 =
 0.7794 + 0.451
> lambda2=conj(lambda1)
lambda2 =
 0.7794 - 0.451
> A=poly([lambda1,lambda2])
A =
             -1.5588
                          и.81
> conv([1 -lambda1],[1 -lambda2])
ans =
                  -1.5588
                                      0.81
> y=filter(1,A,ones(1,20));
> clf
> plot(y,'o')
> 1/sum(A) % steady-state gain
ans =
   3.9816
```

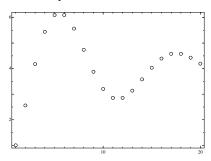
## Example: step response of second order system

#### Test in command window

ans = 3.9816

```
> lambda1=0.9*exp(i*pi/6)
Lambda1 =
 0.7794 + 0.45i
> lambda2=conj(lambda1)
lambda2 =
 0.7794 - 0.45i
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A =
             -1.5588
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ans =
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                                      0.81
> y=filter(1,A,ones(1,20));
> clf
> plot(y,'o')
> 1/sum(A) % steady-state gain
```

#### Result in **plot window**



Notice:

$$\arg \lambda_1 = \frac{\pi}{6} = \omega = \frac{2\pi}{T}$$

Period T = 12 samples

### Outline

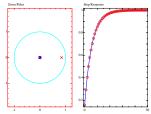
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## Example: real positive poles

#### Open zeropole.sq and drag poles

$$G(\mathbf{q}^{-1}) = \frac{(1-z_1\mathbf{q}^{-1})(1-z_2\mathbf{q}^{-1})(1-z_3\mathbf{q}^{-1})}{(1-\lambda_1\mathbf{q}^{-1})(1-\lambda_2\mathbf{q}^{-1})(1-\lambda_3\mathbf{q}^{-1})}$$

Move  $\lambda_1$ , then  $\lambda_2$  (other 0)

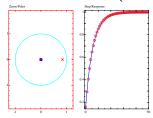


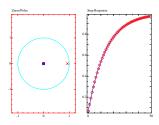
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### Move $\lambda_1$ , then $\lambda_2$ (other 0)



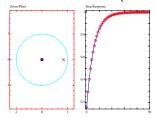


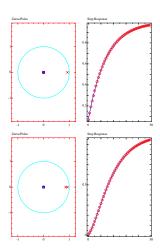
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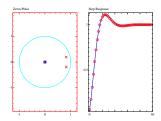
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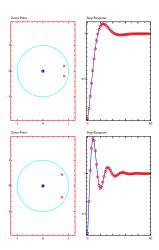
$$G(q^{-1}) = \frac{(1 - z_1 q^{-1})(1 - z_2 q^{-1})(1 - z_3 q^{-1})}{(1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1})(1 - \lambda_3 q^{-1})}$$

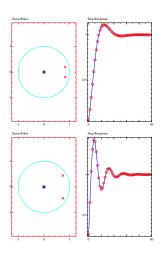
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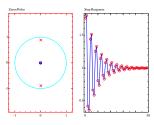


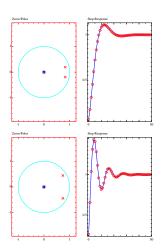


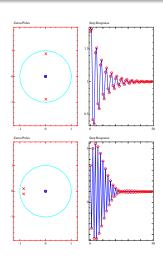




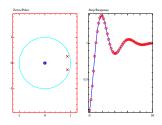






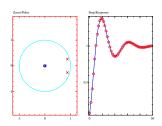


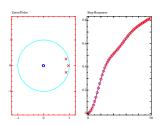
## Example: dominant poles



Pole closest to 1 *dominates* step response

### Example: dominant poles





Pole closest to 1 *dominates* step response

