

1

$$y(n) = x(n) + 2x(n-1) + x(n-2)$$

a) $y(n) = h(n)$ when $x(n) = \delta(n)$

$$h(n) = \delta(n) + 2\delta(n-1) + 1\delta(n-2)$$

$$h(n) = [1, 2, 1]$$

BIBO stable if: $\sum_{n=-\infty}^{\infty} |h(n)| < \infty$

So the system is stable. (FIR-system!)

b) $H(\omega) = \sum_{n=0}^2 h(n) e^{-jn\omega}$

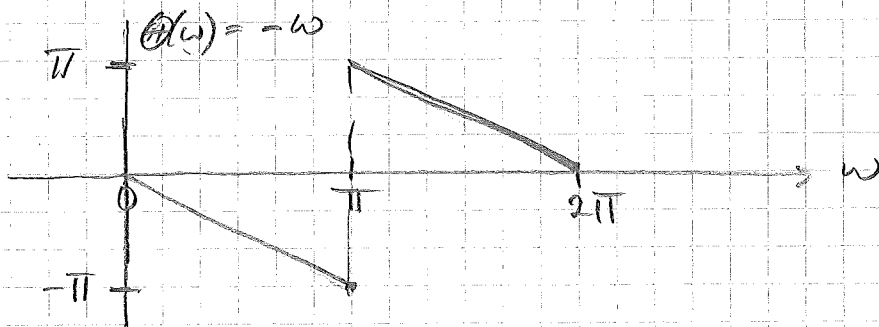
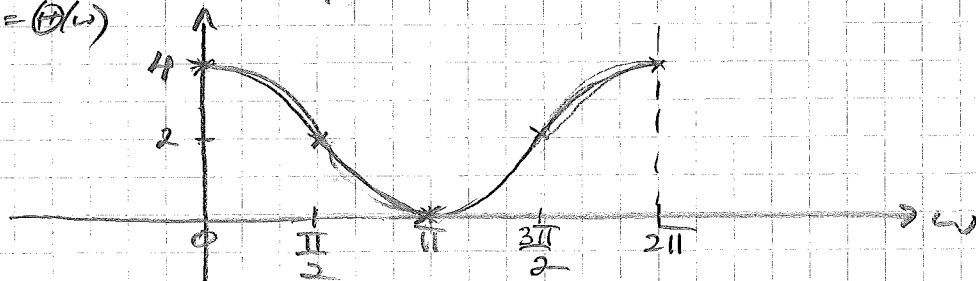
$$= 1 + 2e^{-j\omega} + e^{-j2\omega}$$

$$= e^{j\omega} (e^{j\omega} + e^{-j\omega}) + 2e^{-j\omega}$$

$$= e^{j\omega} (2 + 2\cos(\omega)) = e^{j\omega} 2(1 + \cos(\omega))$$

Sketch
 $|H(\omega)|$ and
 $\angle H(\omega) = \phi(\omega)$

$$|H(\omega)| = 2(1 + \cos(\omega)), \geq 0 \forall \omega$$



① cont.

c) $H_1(\omega) = H(\omega + \pi)$

$$\begin{aligned} H_1(\omega) = H(\omega + \pi) &= \sum_{n=0}^2 h(n) e^{-j(\omega + \pi) \cdot n} \\ &= \sum_{n=0}^2 h(n) e^{-j\omega n} \cdot e^{-j\pi \cdot n} \\ &= \sum_{n=0}^2 h(n) \cdot \underbrace{e^{-j\pi \cdot n}}_{= h_p(n)} \cdot e^{-j\omega n} \end{aligned}$$

So $h_p(n) = h(n) \cdot e^{-j\pi \cdot n} = h(n) (e^{-j\pi})^n = h(n) \underline{\underline{(-1)^n}}$

$$h_p(0) = h(0) (-1)^0 = h(0)$$

$$h_p(1) = h(1) (-1)^1 = -h(1)$$

$$h_p(2) = h(2) (-1)^2 = h(2)$$

$$\underline{\underline{h_p(n) = [1, -2, 1]}}$$

2

a) $x_1(n) = 0.5 \cos\left(\frac{\pi}{5}n\right) + 0.8 \sin\left(\frac{\pi}{5}n\right) \quad -10 < n < 10$

Periodic signal is repr by $\{c_k\}$ in the freq. domain.

F-series expansion:

$$x(n) = \sum_{k=0}^{N-1} c_k e^{j\left(\frac{2\pi}{N} \cdot k\right) \cdot n} \quad ; N \text{ is the period of } x(n)$$

$$x_1(n) = 0.5 \cos\left(\frac{2\pi}{6} \cdot n\right) + 0.8 \sin\left(\frac{2\pi}{10} \cdot n\right)$$

$$= 0.5 \cos\left(\frac{2\pi \cdot 5}{30} \cdot n\right) + 0.8 \sin\left(\frac{2\pi \cdot 3}{30} \cdot n\right)$$

$N=30$ (Minimum common period)

F-series expansion when $N=30$

$$x(n) = \sum_{k=0}^{29} c_k e^{j\left(\frac{2\pi}{30} \cdot k\right) \cdot n}$$

$$\mathcal{F}\{x_1(n)\} = \mathcal{F}\left\{0.5 \cos\left(\frac{2\pi \cdot 5}{30} n\right)\right\} + \mathcal{F}\left\{0.8 \sin\left(\frac{2\pi \cdot 3}{30} n\right)\right\}$$

$c_k = c_k^{s1} \quad c_k^{s2}$

Use Euler id:

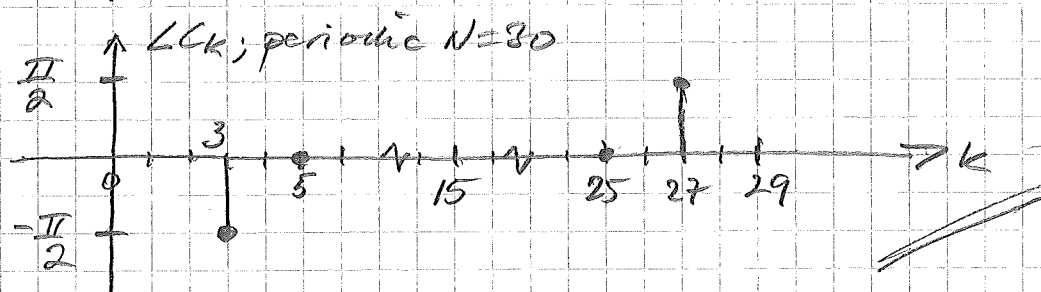
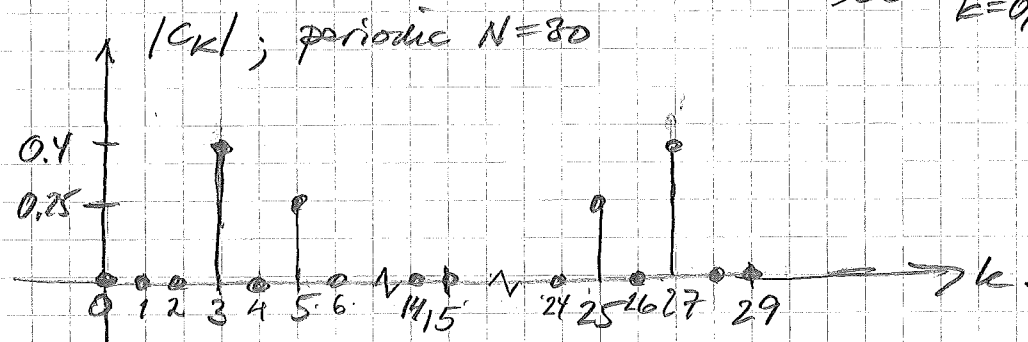
$$\begin{aligned} 0.5 \cos\left(\frac{2\pi}{30} \cdot 5 \cdot n\right) &= 0.5 \cdot \frac{1}{2} \left[e^{j\frac{2\pi}{30} \cdot 5 \cdot n} + e^{-j\frac{2\pi}{30} \cdot 5 \cdot n} \right] \\ &= \left(\frac{1}{4}\right) e^{j\frac{2\pi}{30} \cdot 5 \cdot n} + \left(\frac{1}{4}\right) e^{j\frac{2\pi}{30} \cdot (-5) \cdot n} \\ &= \underline{\underline{c_5^{s1}}} \quad c_{-5}^{s1} = c_{30-5}^{s1} = \underline{\underline{c_{25}^{s1}}} \end{aligned}$$

periodic $N=30$.

$$\begin{aligned} 0.8 \sin\left(\frac{2\pi}{30} \cdot 3 \cdot n\right) &= 0.8 \cdot \frac{1}{2j} \left[e^{j\frac{2\pi}{30} \cdot 3 \cdot n} - e^{-j\frac{2\pi}{30} \cdot 3 \cdot n} \right] \\ &= \left(\frac{0.4}{j}\right) e^{j\frac{2\pi}{30} \cdot 3 \cdot n} - \left(\frac{0.4}{j}\right) e^{j\frac{2\pi}{30} \cdot (-3) \cdot n} \\ &= \underline{\underline{c_3^{s2}}} \quad c_{-3}^{s2} = c_{30-3}^{s2} = \underline{\underline{c_{27}^{s2}}} \end{aligned}$$

② cont.

$$C_k = C_k^{s1} + C_k^{s2} \Rightarrow \begin{cases} C_3 = \frac{0.4}{j} = 0.4 e^{-j\pi/2} \\ C_5 = \frac{1}{4} = 0.25 \\ C_{25} = \frac{1}{4} = 0.25 \\ C_{27} = -\frac{0.4}{j} = 0.4 e^{j\pi/2} \\ C_k = 0 \text{ for "other" } k \text{ values} \\ k=0,1,2,\dots,29 \end{cases}$$



b)

$$x_2(n) = x_1(n) \cdot w(n) = w(n) \cdot 0.5 \cos\left(\frac{\pi}{3}n\right) + w(n) \cdot 0.8 \sin\left(\frac{\pi}{5}n\right)$$

$$X_2(\omega) = 0.5 \cdot \frac{1}{2} \left[W(\omega - \frac{\pi}{3}) + W(\omega + \frac{\pi}{3}) \right] + 0.8 \cdot \frac{1}{2j} \left[W(\omega - \frac{\pi}{5}) - W(\omega + \frac{\pi}{5}) \right]$$

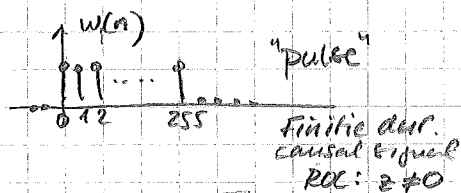
So determine

$$W(\omega) = \mathcal{F}\{w(n)\}$$

$$\text{where } w(n) = \begin{cases} 1 & 0 \leq n \leq 255 \\ 0 & \text{otherwise} \end{cases}$$

②
cont.

$$w(n) = u(n) - u(n-256)$$



$z \downarrow$

$$W(z) = \frac{1}{(1-z^{-1})} - \frac{z^{-256}}{(1-z^{-1})} = \frac{1-z^{-256}}{(1-z^{-1})}$$

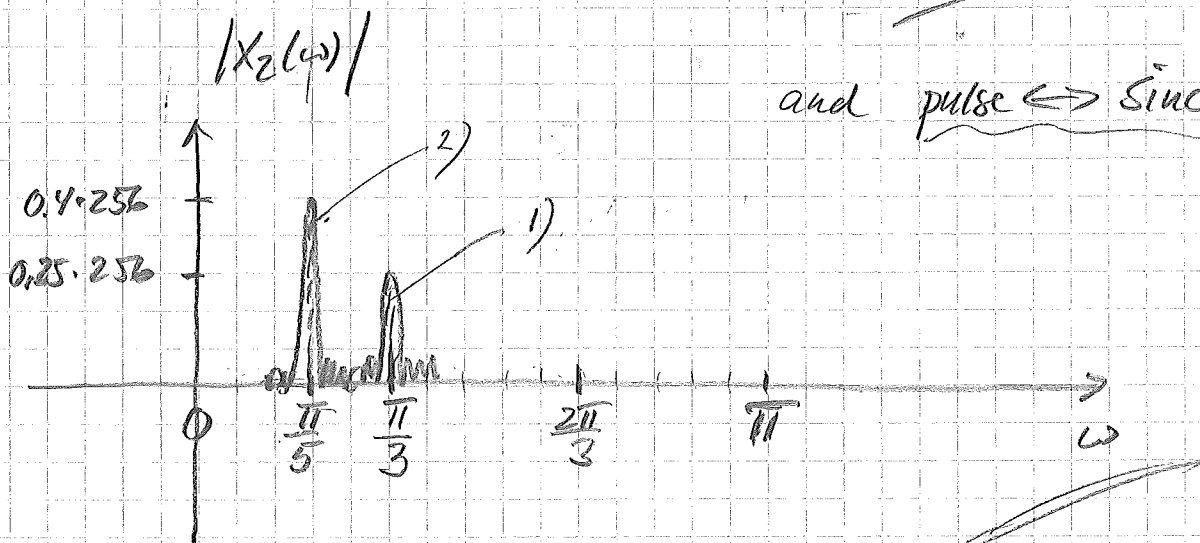
$$\left\{ \begin{aligned} W(\omega) &= W(z) \Big|_{z=e^{j\omega}} = \frac{1-e^{-j\omega 256}}{1-e^{-j\omega}} \\ &= \frac{e^{-j128\omega} (e^{j128\omega} - e^{-j128\omega})}{e^{-j\omega/2} (e^{j\omega/2} - e^{-j\omega/2})} \\ &= e^{-j\frac{255}{2}\omega} \frac{\sin(128\omega)}{\sin(\omega/2)} \quad \omega \neq 0 \\ W(0) &= \sum_{n=0}^{255} 1 = 256 \end{aligned} \right.$$

zero-crossings when

$$128\omega = k \cdot \pi \quad k = \pm 1, \pm 2, \dots$$

$$\omega = \frac{\pi}{128} \cdot k$$

and pulse \leftrightarrow Sinc



$$1) \quad \left| \frac{0.5}{2} W(\omega - \frac{\pi}{3}) \right| = \frac{0.5}{2} \left| W(\omega - \frac{\pi}{3}) \right|$$

$$2) \quad \left| \frac{0.8}{2j} W(\omega - \frac{\pi}{5}) \right| = \frac{0.8}{2} \left| W(\omega - \frac{\pi}{5}) \right|$$

3.

$$H(z) = \frac{0.1(1 - z^{-2})}{1 - 0.9z^{-1} + 0.81z^{-2}}$$

$$a) \quad H(z) \frac{z^2}{z^2} = \frac{0.1(z^2 - 1)}{z^2 - 0.9z + 0.81} = \frac{0.1(z+1)(z-1)}{(z-p_1)(z-p_2)}$$

p_1 and p_2 ?

$$z^2 - 0.9z + 0.81 = 0$$

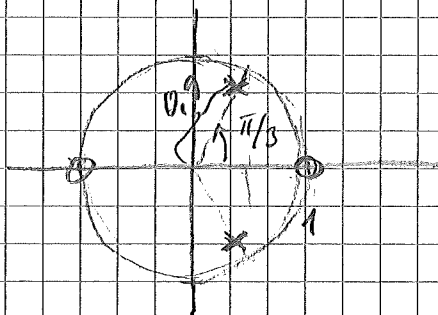
$$z_{1,2} = \frac{0.9}{2} \pm \sqrt{\left(\frac{0.9}{2}\right)^2 - 0.81}$$

$$= \frac{0.9}{2} \pm \sqrt{-\frac{3 \cdot 0.81}{4}} = \frac{0.9}{2} \pm j \frac{\sqrt{3}}{2} \cdot 0.9$$

$$= 0.9 \left(\frac{1}{2} \pm j \frac{\sqrt{3}}{2} \right)$$

$$\begin{cases} p_1 = 0.9 e^{j\pi/3} \\ p_2 = p_1^* = 0.9 e^{-j\pi/3} \end{cases}$$

Pole-zero pattern:



$$\begin{cases} p_1 = 0.9 e^{j\pi/3} \\ p_2 = p_1^* \end{cases}$$

poles

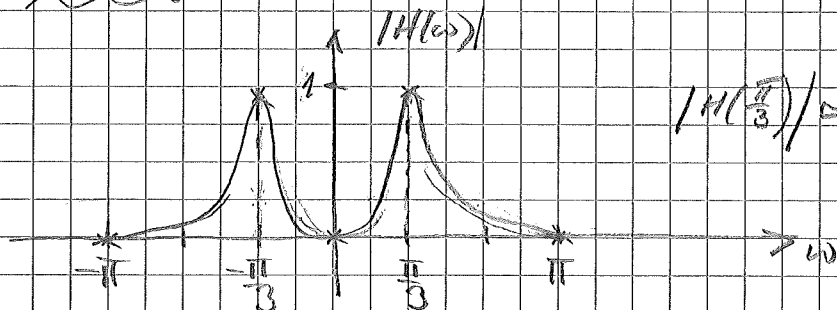
$$\begin{cases} z_1 = 1 \\ z_2 = -1 \end{cases}$$

zeros

Magnitude function $H(\omega) = H(z)_{z=e^{j\omega}}$

$H(z)$ peaks close to a pole, i.e. at $\omega = \pm \frac{\pi}{3}$

$H(z) = 0$ for $z=1$ and $z=-1$ (zeros), i.e. at $\omega=0$ and $\pm\pi$



$$|H(\pi/3)| = 1$$

③ cont.

$$x(n) = 0.4 + 0.4 \cos\left(\frac{\pi}{3}(n-2)\right) \quad -\infty < n < \infty$$

Steady-state response of the output:

$$y(n) = \underbrace{0.4|H(0)|} + 0.4 \underbrace{|H(\frac{\pi}{3})|} \cos\left(\frac{\pi}{3}(n-2) + \underbrace{\angle H(\frac{\pi}{3})}\right)$$

$$H(0) = H(z)|_{z=1} = 0 \quad \begin{array}{l} \text{zero at } \omega=0 \\ \text{(check the pole-zero pattern)} \end{array}$$

$$H\left(\frac{\pi}{3}\right) = H(z)|_{z=e^{j\frac{\pi}{3}}} = \frac{0.1(e^{j\frac{2\pi}{3}} - 1)}{(e^{j\frac{\pi}{3}} - 0.9e^{j\frac{\pi}{3}})(e^{j\frac{\pi}{3}} - 0.9e^{-j\frac{\pi}{3}})}$$

$$= \frac{0.1(-0.5 + j\frac{\sqrt{3}}{2} - 1)}{e^{j\frac{\pi}{3}}(1-0.9)(e^{j\frac{\pi}{3}} - 0.9e^{-j\frac{\pi}{3}})} =$$

$$= \frac{0.1(-\frac{3}{2} + j\frac{\sqrt{3}}{2})}{0.1(e^{j\frac{2\pi}{3}} - 0.9)} = \frac{(-\frac{3}{2} + j\frac{\sqrt{3}}{2})}{(-\frac{1}{2} + j\frac{\sqrt{3}}{2} - 0.9)}$$

$$= \frac{(-1.5 + j\frac{\sqrt{3}}{2})}{(-1.4 + j\frac{\sqrt{3}}{2})} \approx 1.05 e^{j1.7^\circ}$$

So,

$$y(n) = \underbrace{0.4 \cdot 1.05}_{\approx 0.42} \cos\left(\frac{\pi}{3}(n-2) + 1.7^\circ\right)$$

(4)

a)

$$\begin{aligned} F_1 &= 1500 \text{ Hz} \\ F_2 &= 4600 \text{ Hz} \\ F_3 &= 5800 \text{ Hz} \end{aligned}$$

$$\left. \begin{aligned} F_1 &= 1500 \text{ Hz} \\ F_2 &= 4600 \text{ Hz} \\ F_3 &= 5800 \text{ Hz} \end{aligned} \right\} \Rightarrow \begin{aligned} F_s &= 8 \text{ kHz} \\ f &= \frac{F}{F_s} \end{aligned}$$

$$\begin{aligned} f_1 &= \frac{15}{80} \\ f_2 &= \frac{46}{80} > \frac{1}{2} \\ f_3 &= \frac{58}{80} > \frac{1}{2} \end{aligned}$$

$$f_2 = \frac{46}{80} = -\frac{34}{80} + 1 \Rightarrow f_{2 \text{ alias}} = \frac{34}{80}$$

$$f_3 = \frac{58}{80} = -\frac{22}{80} + 1 \Rightarrow f_{3 \text{ alias}} = \frac{22}{80}$$

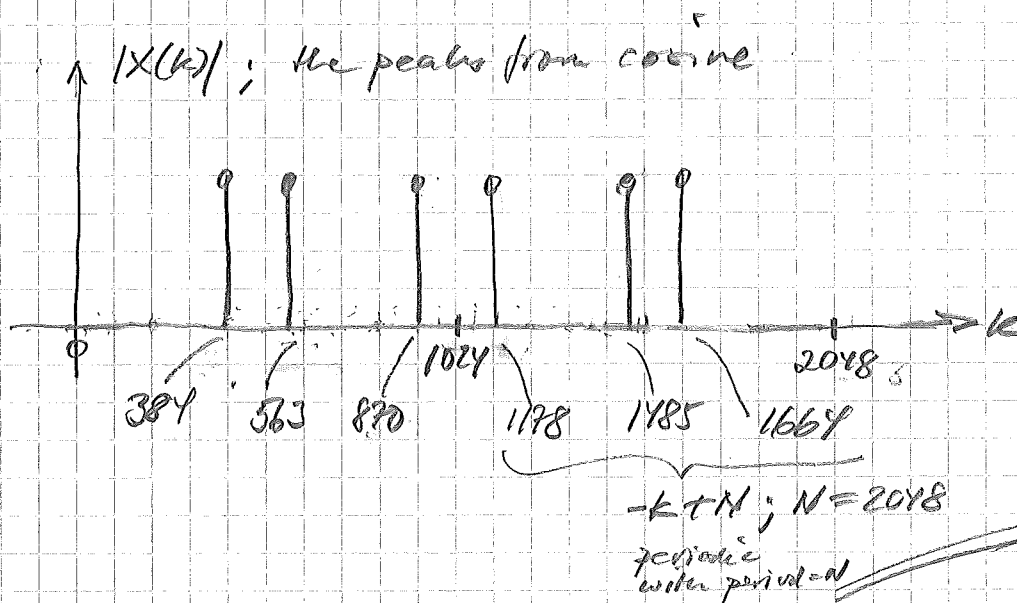
DFT in $N=2048$ points is computed.

$$f=1 \leftrightarrow k=2048$$

$$f_1 = \frac{15}{80} \Rightarrow k = \frac{15}{80} \cdot 2048 \approx 384$$

$$f_{2 \text{ alias}} = \frac{34}{80} \Rightarrow k = \frac{34}{80} \cdot 2048 = 870$$

$$f_{3 \text{ alias}} = \frac{22}{80} \Rightarrow k = \frac{22}{80} \cdot 2048 = 563$$



④ cont.

b)

$$F_s > 2F_{\max} = 2 \cdot 5800 = 11.6 \text{ kHz}$$

Select $F_s = 14 \text{ kHz}$.

$$f = \frac{F}{F_s} \Rightarrow f_1 = \frac{15}{140} < \frac{1}{2}$$

$$f_2 = \frac{46}{140} < \frac{1}{2}$$

$$f_3 = \frac{58}{140} < \frac{1}{2}$$

no aliasing!

"

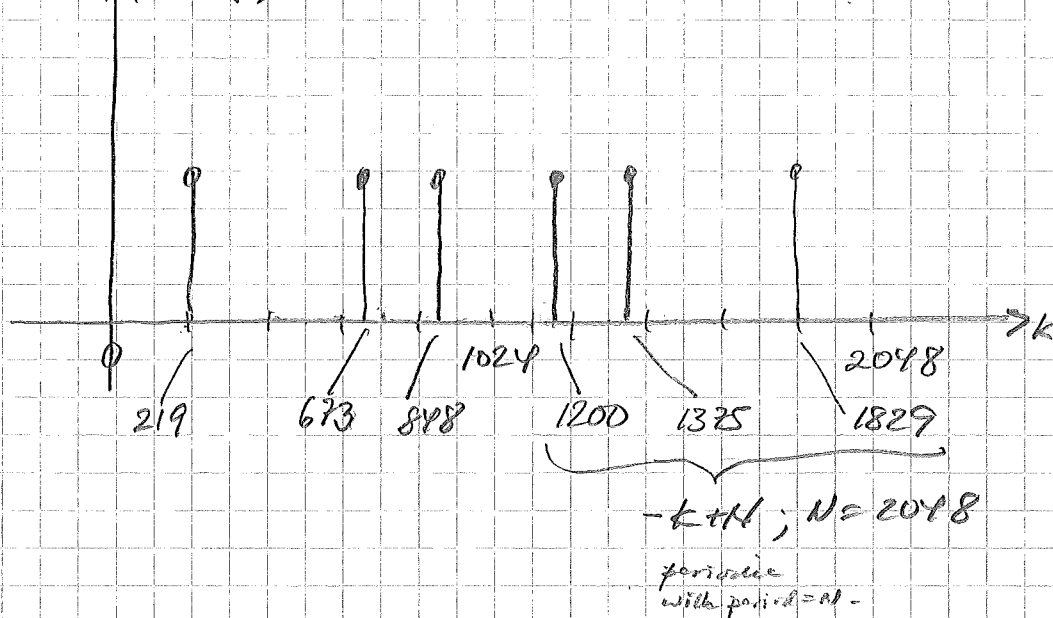
DFT in 2048 points

$$\Rightarrow f_1 = \frac{15}{140} \rightarrow k = \frac{15}{140} \cdot 2048 \approx 219$$

$$f_2 = \frac{46}{140} \rightarrow k = \frac{46}{140} \cdot 2048 \approx 673$$

$$f_3 = \frac{58}{140} \rightarrow k = \frac{58}{140} \cdot 2048 \approx 848$$

$|X(k)|$; the peaks from cosine



④ cont. c)

$$h(n) = \begin{bmatrix} \frac{1}{2} \\ -\frac{1}{2} \end{bmatrix} ; \text{length} = 2$$

$$x(n) = \begin{bmatrix} -1, -1, -1, 1, 1, 1 \end{bmatrix} ; \text{length} = 6$$

$$y = x(n) * h(n) \text{ is of length } (6+2-1) = 7$$

So DFT must be computed in at least 7 points ; $N \geq 7$

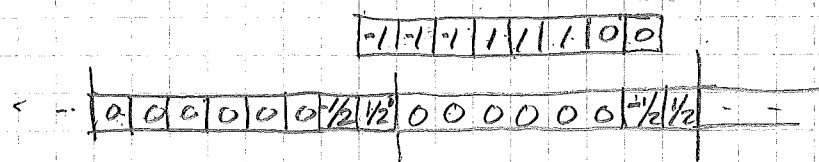
Select $N = 8$

$$Y(k) = \underset{(8)}{\text{DFT}\{X(k)\}} \cdot \underset{(8)}{\text{DFT}\{H(k)\}}$$

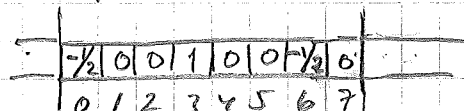
$$y(n) = \underset{(8)}{\text{IDFT}\{Y(k)\}}$$

can be
computed
in the time
domain
by circ. conv.

Graphical solution: (zero-padding to length=8)
in the time domain



periodic; period=8



periodic; period=8

$= y(n)$

$$y(n) = \begin{bmatrix} -\frac{1}{2}, 0, 0, 1, 0, 0, -\frac{1}{2} \end{bmatrix}$$