

## Multi-Pole Feedback Network OP-Amp Circuit

### Objectives

1. To analyze the theory of *feedback network* in the multi-pole OP-Amp circuit.
2. To discuss the issue of *stability* for the feedback amplifier.
3. To understand the physical meaning of *sinusoidal vibration*.

### Overview

The shunt-shunt feedback circuit to be explored is shown in Fig. 1.

Here the underlying assumption is that the poles associated with each follower are very high in frequency, and that those associated with the single inverter (without  $C_1$ ) are high enough to be ignored. With this assumption, the circuit has three controllable poles for which the associated time constants are  $RC_1$ ,  $RC_2$  and  $RC_3$ . The total (open loop) gain at low frequencies is  $-R/r$ . Resistors  $R_1$  and  $R_2$  establish a nominal closed-loop minimum gain of  $-1$ . Potentiometer  $R_p$  allows the closed-loop gain to be adjusted from  $-1$  to  $-R/r$  continuously. Accordingly, reference will likewise be broad, covering Sections 8.1, 8.6~8.10 of the text [Reference 1].

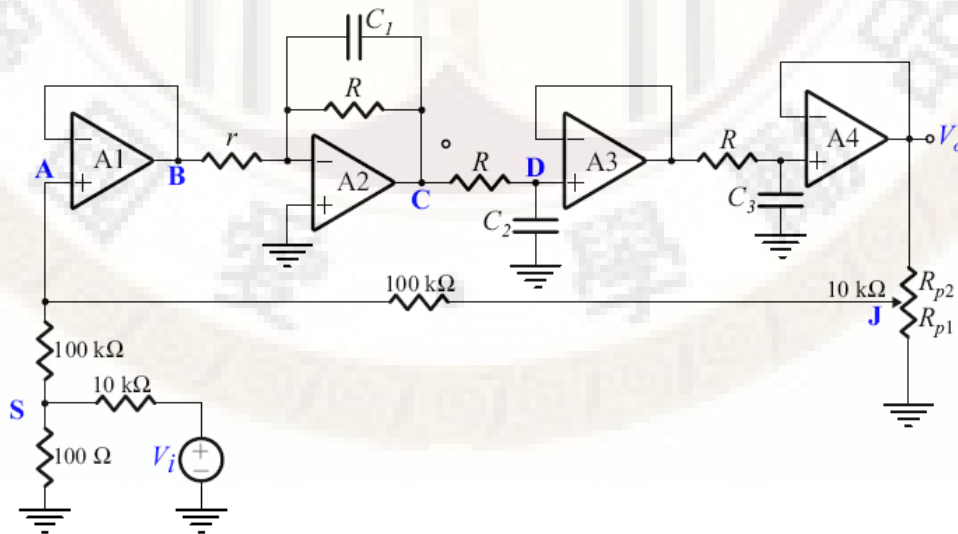


Fig. 1 Multi-pole feedback network OP-Amp circuit

## Components and Instrumentation

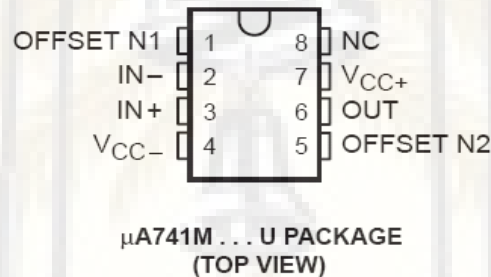
Instrument	Quantity	Component	Quantity
Oscilloscope	1	$\mu A$ 741	4
Multi-meter	1	VR (10 k $\Omega$ )	1
Power supplier	1	100 $\Omega$	2
Function Gen.	1	10 k $\Omega$	4
0.1 $\mu F$	3	100 k $\Omega$	2

## Instrument confirmation

Before you proceed to any part of the experiment, please remember to do the **Instrument Examinations** to the instruments before performing any experiment. The examining procedures are shown in experiment 1.

## Lab Work

### 1. $\mu A741$ – PINOUT & Functional confirmation



### 2. DC Functional Confirmation of $A_1$

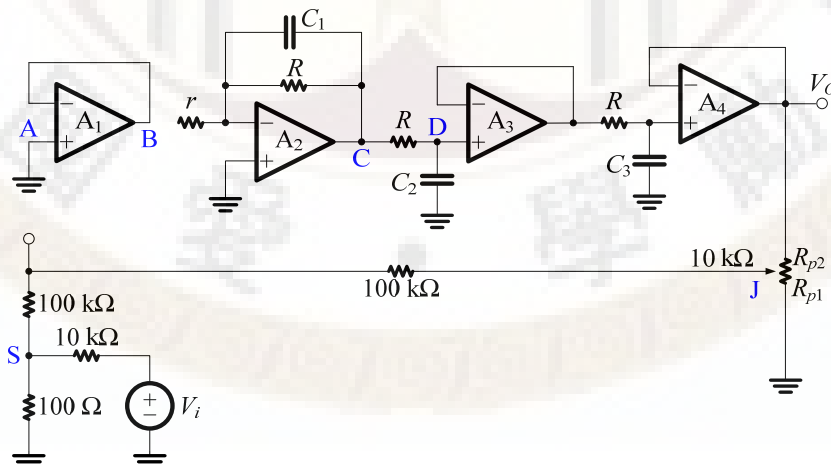
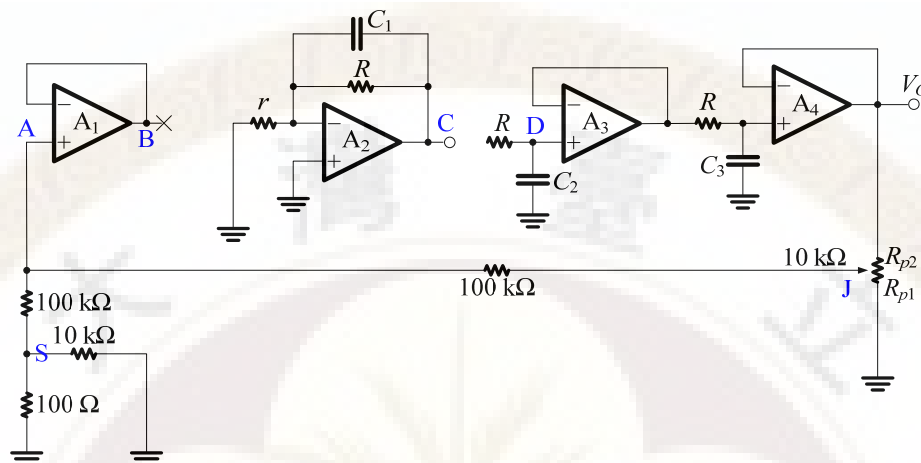


Fig. 2 DC Functional Confirmation of  $A_1$

- (1) Reference pin voltage for  $A_1$ ,
- (2)  $V_{pin7} \doteq +15V$ ,  $V_{pin4} \doteq -15V$ ,  $V_{pin2} = V_{pin3} = V_{pin6} \doteq 0V$ .

### 3. DC Functional Confirmation of A<sub>2</sub>



(4) Use  $R = 10 \text{ k}\Omega$ ,  $r = 100 \text{ }\Omega$ ,  $C_1 = 0.1 \text{ }\mu\text{F}$  (104) for  $A_2$  (Single-time-constant Low-pass Filter, STC LPF) in Fig. 3.

(6) Reference pin voltage for  $A_2$ .

(8) Record the measured pin voltage for A<sub>2</sub>,  $V_{pin7}$  = \_\_\_\_\_ V,  $V_{pin4}$  = \_\_\_\_\_ V,  
 $V_{pin2}$  = \_\_\_\_\_ V,  $V_{pin3}$  = \_\_\_\_\_ V,  $V_{pin6}$  = \_\_\_\_\_ V.

**Note:** If the measured values are far different from those shown in the reference pin voltage, try to change a chip of  $\mu A741$  and repeat step 2 until they are correct.

#### 4. Small signal analysis

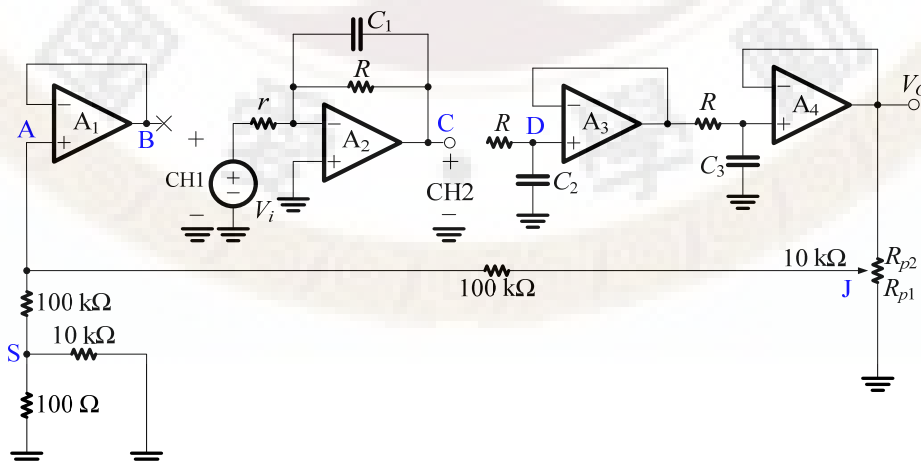


Fig. 4 Small signal analysis of  $A_2$

- (1) Use  $R = 10\text{ k}\Omega$ ,  $r = 100\text{ }\Omega$ ,  $C_1 = 0.1\text{ }\mu\text{F}$  (104) for  $A_2$  in Fig. 4.
- (2) Supply voltage source  $V_{CC} = +15\text{V}$ , and  $-V_{CC} = -15\text{V}$  to the circuit.
- (3) Apply the input small signal  $V_i$  to the breadboard by using function generator to generate  $v_i = v_{ac} \times \sin(2\pi ft)$ ,  $2v_{ac} = 100\text{mV}_{(p-p)}$ ,  $f = 100\text{ Hz}$ .

$$T(j\omega) = \frac{V_C(j\omega)}{V_B(j\omega)} = -\frac{R}{r} \cdot \frac{1}{1 + j\omega \cdot (RC_1)} = \frac{-A_M}{1 + \frac{j\omega}{1/(RC_1)}}$$

$$\text{where: } \omega_{3dB} = \frac{1}{RC_1}, A_M = \frac{R}{r} = \frac{10\text{k}\Omega}{0.1\text{k}\Omega} = 100$$

$$|T(j\omega_{3dB})|_{\omega=\omega_{3dB}} = \frac{-A_M}{\sqrt{2}} = -0.707 \times A_M = -70.7$$

- (4) Make sure that the  $v_i$  is measured from the breadboard by using the probe from **CH1** in oscilloscope.
- (5) Oscilloscope ► Press the **CH1** and **CH2 MENU ► Coupling ► AC**.
- (6) Observe  $V_{i(p-p)}$  and  $V_{o(p-p)}$  in CH1 and CH2, respectively.
- (7) Keep the previous adjustment of  $V_i$  constantly.
- (8) Record the voltage gain  $A_M = \underline{\hspace{2cm}} \text{ V/V}$  in the oscilloscope.
- (9) Function generator ► Adjust Frequency and observe the voltage gain  $A_V$  in oscilloscope until  $A_V = 0.707 \times A_M$ .
- (10) Record the frequency  $f_{3dB} = \underline{\hspace{2cm}} \text{ Hz}$ . (Reference value  $f_{3dB} = 160\text{Hz}$ .)

#### 5. DC Functional Confirmation of $A_3$

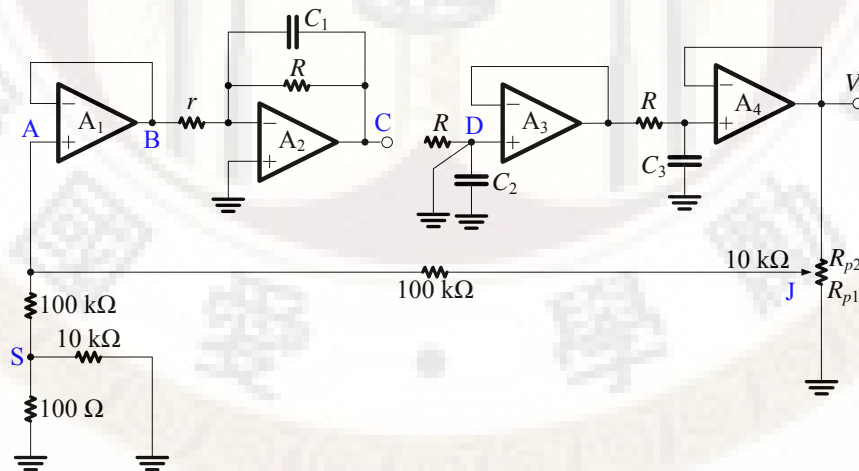


Fig. 5 DC Functional Confirmation of  $A_3$

- (1) In the Fig. 5, short terminal **D** to the ground.
- (2) Reference pin voltage for  $A_3$ ,  $V_{pin7} \doteq +15\text{V}$ ,  $V_{pin4} \doteq -15\text{V}$ ,  $V_{pin2} = V_{pin3} = V_{pin6} \doteq 0\text{V}$ .
- (3) Record the measured pin voltage for  $A_3$ ,  $V_{pin7} = \underline{\hspace{1cm}} \text{V}$ ,  $V_{pin4} = \underline{\hspace{1cm}} \text{V}$ ,

$$V_{pin2} = \underline{\hspace{1cm}} \text{ V}, V_{pin3} = \underline{\hspace{1cm}} \text{ V}, V_{pin6} = \underline{\hspace{1cm}} \text{ V}.$$

#### 6. DC Functional Confirmation of $A_4$

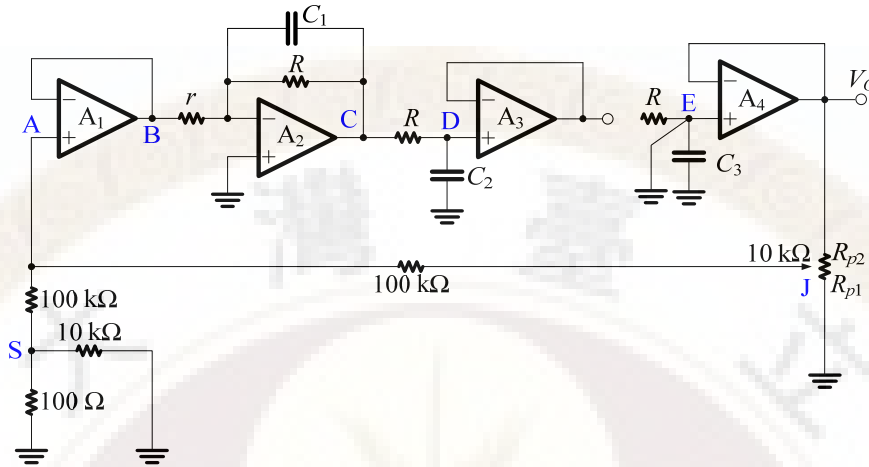


Fig. 6 DC Functional Confirmation of  $A_4$

(4) In the Fig. 6, short terminal **E** to the ground.

(5) Reference pin voltage for  $A_4$ ,

(6)  $V_{pin7} \doteq +15\text{V}$ ,  $V_{pin4} \doteq -15\text{V}$ ,  $V_{pin2}=V_{pin3}=V_{pin6} \doteq 0\text{V}$ .

(7) Record the measured pin voltage for  $A_4$ ,  $V_{pin7} = \underline{\hspace{1cm}} \text{ V}$ ,  $V_{pin4} = \underline{\hspace{1cm}} \text{ V}$ ,  
 $V_{pin2} = \underline{\hspace{1cm}} \text{ V}$ ,  $V_{pin3} = \underline{\hspace{1cm}} \text{ V}$ ,  $V_{pin6} = \underline{\hspace{1cm}} \text{ V}$ .

#### 7. Initial state of the feedback network circuit

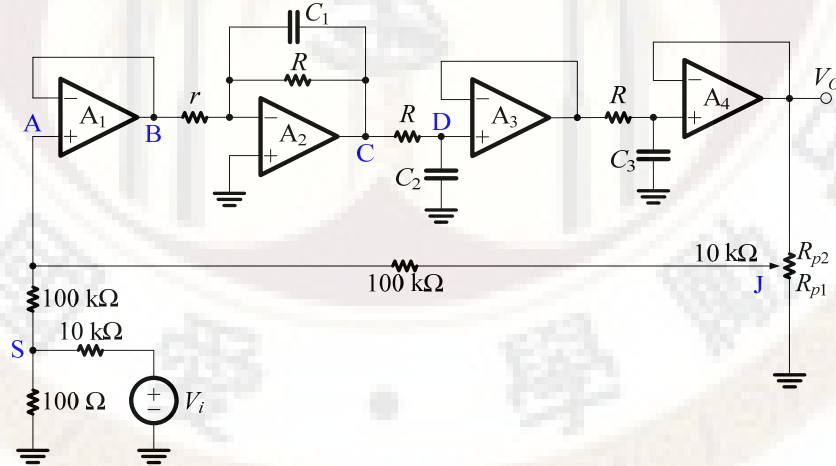


Fig. 7 Multi-pole feedback network OP-Amp circuit

(1) In Fig. 7, use  $R = 10 \text{ k}\Omega$ ,  $r = 100 \Omega$ ,  $C_1 = C_2 = C_3 = 0.1 \mu\text{F}$  (104),  $VR = 10 \text{ k}\Omega$  for  $R_{p1}(R_{p2})$ .

(2) Adjust  $VR$  to have  $R_{p1} = 0 \Omega$ ,  $R_{p2} = 10 \text{ k}\Omega$ .

(3) Apply the input signal  $V_i$  to the breadboard by using function generator

to generate  $v_i = v_{ac} \times \text{square}(2\pi ft)$ ,  $2v_{ac} = 5V_{(p-p)}$ ,  $f = 0 \sim 10 \text{ Hz}$ . (即  $0 \sim 10 \text{ Hz}$  之方波)。

- (4) Make sure that the  $v_i$  is measured from the breadboard by using the probe from **CH1** in oscilloscope.
- (5) Oscilloscope ► Press the **CH1** and **CH2 MENU ► Coupling ► DC**.
- (6) Observe whether the waveform shown in CH1 and CH2 distort (Y/N)?

*Homework #1:* Why should we use *DC Coupling* to observe both the waveforms in oscilloscope? Try to explain it in the conclusive report by electronic or mathematical expression.

#### 8. Vibration observation of the circuit

- (1) Keep the previous adjustment in step 7 constantly.
- (2) Observe the waveform of  $V_{o(p-p)}$  in CH2 when slowly increasing the value of  $R_{p1}$  (decrease  $R_{p2}$ ) until the sinusoidal vibration occur.
- (3) As the sinusoidal vibration occur, record  $V_{S(p-p)} = \underline{\hspace{1cm}} V$ ,  $V_{J(p-p)} = \underline{\hspace{1cm}} V$ ,  $V_{o(p-p)} = \underline{\hspace{1cm}} V$ ,  $f_o = \underline{\hspace{1cm}} \text{ Hz}$  from the oscilloscope,  $R_{p1} = \underline{\hspace{1cm}} \text{ k}\Omega$ ,  $R_{p2} = \underline{\hspace{1cm}} \text{ k}\Omega$ .
- (4) During the adjustment of appearing sin-vibration, observe whether the waveform of  $V_{o(p-p)}$  occur damping phenomenon (Y/N).
- (5) ※ Observe the waveform of  $V_{o(p-p)}$  in CH2, what if Disconnect the input signal  $V_i$  to the breadboard? Does it disappear? Why?

*Homework #2:* Try to explain it in the conclusive report by electronic or mathematical expression.

## Reference

1. A.S. Sedra and K.C. Smith, *Microelectronic Circuits*, 6th ed., Oxford University Press publishing, New York, August 2011.
2. A.S. Sedra and K.C. Smith, *Laboratory Manual for Microelectronic Circuits*, 3<sup>rd</sup> ed., Oxford University Press publishing, New York, 1997.
3. Paul Horowitz, Winfield Hill, *The art of electronics*, 2<sup>nd</sup> ed., Cambridge University Press, New York, 1989.