



# EE140 Introduction to Communication Systems

## Lecture 13

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# Vector Space and Signal Space

Importance: Signal space (or vector) representation of signal is a very effective and useful tools in the analysis of digitally modulated signals.

- Many insights about signals depend not on time or frequency, but on the vector relationships
- Orthogonal expansions are best viewed in vector space terms
- Questions of limits and approximation are easily treated in vector space terms

↳ function  $\underline{u}(t)$   $\rightarrow$  orthogonal expansion  $\rightarrow \{\dots \underline{u}_k \dots\}$   
 $\int |u(t)|^2 dt < \infty$   
 $\rightarrow$  Vector space  $\rightarrow$  signal space

Vector space + Inner product  $\Rightarrow$  Inner product space  
EE140: Introduction to Communication Systems

2

Signal space + Inner product  $\Rightarrow$  Inner product space of  $\underline{f}$ -functions

# Axioms and Basic Properties of Vector Space

**Vector:** An  $n$ -dimensional column vector  $\mathbf{v} = (v_1, v_2, \dots, v_n)^T$

**Vector space:** A vector space  $\mathcal{V}$  is a set of vectors  $\mathbf{v} \in \mathcal{V}$ , along with a set of rules for manipulating those objects and along with a set of ancillary elements  $\alpha \in \mathbb{F}$ .

- $\alpha \in \mathbb{F}$  is called scalar (real or complex)
- Real vector space if  $\alpha$  is real
- Complex vector space if  $\alpha$  is complex

## Properties

- $\mathbf{u} + \mathbf{v} = \mathbf{v} + \mathbf{u}$  Commutativity of addition
  - for each scalar  $\alpha, \beta$ ,  $\alpha(\beta\mathbf{v}) = (\alpha\beta)\mathbf{v}$  Scalar associativity
  - $\alpha(\mathbf{u} + \mathbf{v}) = \alpha\mathbf{u} + \alpha\mathbf{v}$
  - $(\alpha + \beta)\mathbf{v} = \alpha\mathbf{v} + \beta\mathbf{v}$
  - $\|\mathbf{v}\| = \sqrt{\sum_{i=1}^n |v_i|^2}$
- } Distributive laws

# Inner Product Space

Let  $\mathbf{v}_1 = (v_{11}, v_{12}, \dots, v_{1n})^t$  and  $\mathbf{v}_2 = (v_{21}, v_{22}, \dots, v_{2n})^t$

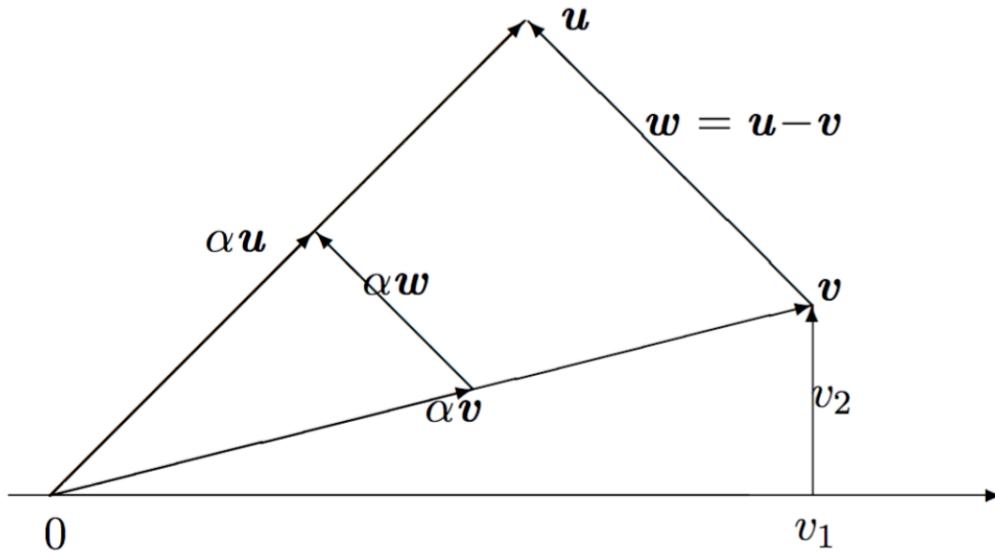
- Inner (dot) product:

$$\langle \mathbf{v}_1, \mathbf{v}_2 \rangle = \mathbf{v}_1 \cdot \mathbf{v}_2 = \sum_{i=1}^n v_{1i} v_{2i}^* = \mathbf{v}_2^H \mathbf{v}_1$$

where  $A^H$  denotes the Hermitian transpose of matrix  $A$

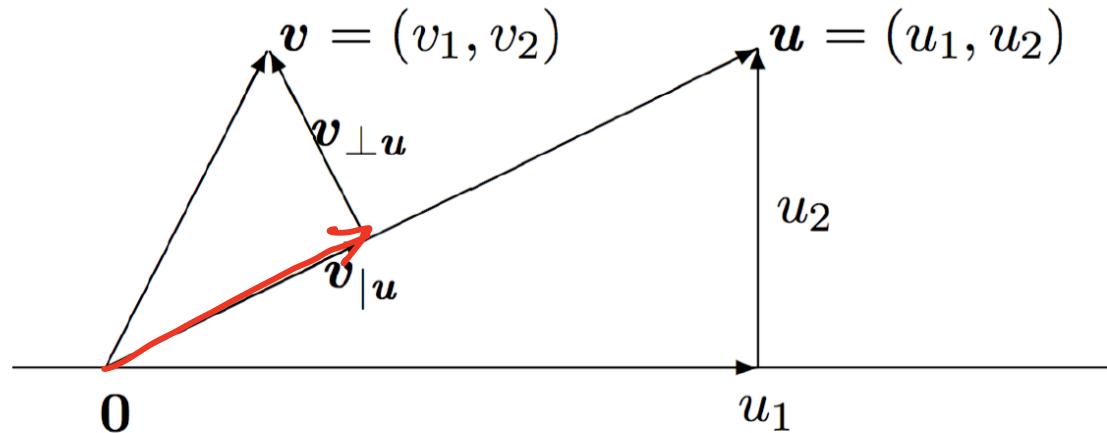
- $\langle \mathbf{v}_1, \mathbf{v}_2 \rangle = \langle \mathbf{v}_2, \mathbf{v}_1 \rangle^*$   $\| \vec{v} \| = \sqrt{\langle \vec{v}, \vec{v} \rangle}$  Hermitian Symmetry
- $\langle \mathbf{v}_1, \mathbf{v}_2 \rangle + \langle \mathbf{v}_2, \mathbf{v}_1 \rangle = 2\text{Re}[\langle \mathbf{v}_1, \mathbf{v}_2 \rangle]$
- $\mathbf{v}_1$  and  $\mathbf{v}_2$  are orthogonal if  $\langle \mathbf{v}_1, \mathbf{v}_2 \rangle = 0$
- A set of vectors  $\mathbf{v}_k$  are orthogonal if  $\langle \mathbf{v}_i, \mathbf{v}_j \rangle = 0$  for all  $i \neq j$ .
- A set of vectors are orthonormal if it is orthogonal and all  $\| \mathbf{v}_i \| = 1$ .

# 2-Dimensional Vector Space



- $v = (v_1, v_2)$
- $v + w = u, u - w = v$
- $\cos(\angle(v, u)) = \frac{v_1 u_1 + v_2 u_2}{\sqrt{v_1^2 + v_2^2} \sqrt{u_1^2 + u_2^2}} = \frac{\langle v, u \rangle}{\|v\| \|u\|}$
- proof: hint  $\cos A = (b^2 + c^2 - a^2)/2bc$

# One-Dimensional Projection



- $v_{\perp u}, v_{|u}$ : Projection of  $v$  onto  $u$
- $\underline{v} = \boxed{v_{\perp u}} + \boxed{v_{|u}}$  ( $v_{\perp u}$  and  $v_{|u}$  are independent and orthogonal)
- $\cos(\angle(v, u)) = \frac{\langle v, u \rangle}{\|v\| \|u\|}$  ↓
- $\|v_{|u}\| = \|v\| \cos(\angle(v, u))$ ; Unite vector:  $\frac{u}{\|u\|}$
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$$\frac{\langle \vec{v}, \vec{u} \rangle}{\|\vec{v}\| \|\vec{u}\|} \frac{\vec{u}}{\|\vec{u}\|}$$

$$v_{|u} = \frac{\langle v, u \rangle}{\|u\|^2} u$$

# Inner Product Spaces

Let  $\mathbf{v}_1 = (v_{11}, v_{12}, \dots, v_{1n})^t$  and  $\mathbf{v}_2 = (v_{21}, v_{22}, \dots, v_{2n})^t$

- Vectors  $\mathbf{v}_1, \mathbf{v}_2$  satisfy the triangle inequality:

$$\|\mathbf{v}_1 + \mathbf{v}_2\| \leq \|\mathbf{v}_1\| + \|\mathbf{v}_2\|$$

- The Cauchy-Schwarz inequality:

$$|\langle \mathbf{v}_1, \mathbf{v}_2 \rangle| \leq \|\mathbf{v}_1\| \cdot \|\mathbf{v}_2\|$$

with equality if  $\mathbf{v}_1 = \overline{a} \mathbf{v}_2$  for some complex scalar  $a$ .

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$$\|\mathbf{v}_1 + \mathbf{v}_2\|^2 = \|\mathbf{v}_1\|^2 + \|\mathbf{v}_2\|^2 + 2\text{Re}[\langle \mathbf{v}_1, \mathbf{v}_2 \rangle]$$

If  $\mathbf{v}_1$  and  $\mathbf{v}_2$  are orthogonal, then  $\|\mathbf{v}_1 + \mathbf{v}_2\|^2 = \|\mathbf{v}_1\|^2 + \|\mathbf{v}_2\|^2$

# Linear Independent and Dependent

A set of vectors  $\mathbf{v}_1, \mathbf{v}_2, \dots, \mathbf{v}_n$  is

- **linearly dependent** if  $\sum_{j=1}^n \alpha_j \mathbf{v}_j = 0$ , where not all  $\alpha_i = 0$ . Thus, when  $\alpha_k \neq 0$ ,

$$\mathbf{v}_k = \sum_{j \neq k} \frac{\alpha_j}{\alpha_k} \mathbf{v}_j$$

(Each  $\mathbf{v}_k$  for which  $\alpha_k \neq 0$  is a linear combination of the others)

- **linearly independent** if it's Not linearly dependent, i.e., if  $\sum_{j=1}^n \alpha_j \mathbf{v}_j = 0$ , it implies that each  $\alpha_j = 0$ .

(No one vector can be represented as a linear combination of the remaining vectors.)

The word "linear" is often omitted.

# Basis for Vector Space

- A set of vectors  $v_1, v_2, \dots, v_n$  spans a vector space  $\mathcal{V}$  if every vector  $v \in \mathcal{V}$  is a linear combination of  $v_1, v_2, \dots, v_n$
- For  $\mathbb{R}^n$ , let

$$\begin{cases} e_1 = (1, 0, 0, \dots, 0) \\ e_2 = (0, 1, 0, \dots, 0) \\ \vdots \\ e_n = (0, 0, \dots, 1) \end{cases}$$

be the  $n$  unit vectors for  $\mathbb{R}^n$ . The unit vectors span  $\mathbb{R}^n$  since every  $v \in \mathcal{V}$  is

$$v = (\alpha_1, \dots, \alpha_n) = \sum_{j=1}^n \alpha_j e_j$$

- A set of vectors  $v_1, v_2, \dots, v_n$  is a basis for  $\mathcal{V}$  if the set is linearly independent and spans  $\mathcal{V}$ , e.g.,  $e_1, \dots, e_n$  form basis for  $\mathbb{R}^n$  and  $\mathbb{C}^n$

# Finite-Dimensional Projections

Given

- a vector  $v \in \mathcal{V}$
- an orthonormal basis  $\phi_1, \phi_2, \dots, \phi_n$  for vector space  $S$

$v_{|S}$ : projection onto  $S$ :

$$\underline{v}_{|S} = \sum_{j=1}^n \langle v, \phi_j \rangle \underline{\phi_j}$$

When  $n = 1$ , then  $\boxed{v_{|S} = v_{|\phi_1} = \langle v, \phi_1 \rangle \phi_1}$

# Gram-Schmidt Procedure

Given an arbitrary basis  $\{s_1, \dots, s_n\}$  spanning a space  $\mathcal{S}$ , generate an orthonormal basis  $\{\phi_1, \phi_2, \dots, \phi_n\}$  for  $\mathcal{S}$ . Let  $\mathcal{S}_k$  be the subspace spanned by  $\{\phi_1, \dots, \phi_k\}$

1  $\phi_1 = s_1 / \|s_1\|.$

$$\phi_1 \rightarrow S_1$$

2  $(s_2)_{\perp \mathcal{S}_1} = s_2 - (s_2)_{|\mathcal{S}_1}$

3  $\phi_2 = (s_2)_{\perp \mathcal{S}_1} / \|(s_2)_{\perp \mathcal{S}_1}\|.$   $\phi_2$

4 Using induction,

5  $(s_{k+1})_{|\mathcal{S}_k} = \sum_{j=1}^k \langle s_{k+1}, \phi_j \rangle \phi_j$

$$\boxed{\{\phi_1, \dots, \phi_k\} \rightarrow \mathcal{S}_k}$$

6  $(s_{k+1})_{\perp \mathcal{S}_k} = s_{k+1} - \sum_{j=1}^k \langle s_{k+1}, \phi_j \rangle \phi_j$

$$s_{k+1} \mid \underline{\mathcal{S}_k}$$

7

$$\underline{\phi_{k+1}} = \frac{(s_{k+1})_{\perp \mathcal{S}_k}}{\|(s_{k+1})_{\perp \mathcal{S}_k}\|}$$

# Signal Space

- Norm  $|x(t)|$ . Energy  $\mathcal{E}_x$

$$\underline{\|x(t)\|} = \left( \int_{-\infty}^{\infty} |x(t)|^2 dt \right)^{1/2} = \sqrt{\mathcal{E}_x}$$

- Inner produce:

$$\langle x_1(t), x_2(t) \rangle = \int_{-\infty}^{\infty} x_1(t) x_2^*(t) dt$$

- $x_1(t), x_2(t)$  are **orthogonal** if their inner product is 0.
- Signals are **orthonormal** if they are orthogonal and their norm are unity
- Signals are **linearly independent** if no signal can be represented as a linear combination of remaining signals
- $\|x_1(t) + x_2(t)\| \leq \|x_1(t)\| + \|x_2(t)\|$
- $\|\langle x_1(t), x_2(t) \rangle\| \leq \|x_1(t)\| \|x_2(t)\|$

# Gram-Schmidt for Signals

Given an finite energy signal waveforms  $\{s_m(t)\}_{m=1}^M$ , generate an orthonormal waveforms  $\{\phi_1(t), \phi_2(t), \dots, \phi_M(t)\}$

$$① \quad \underline{\phi_1(t) = s_1(t)/\sqrt{E_1}}.$$

$$② \quad \underline{c_{21} = \langle s_2(t), \phi_1(t) \rangle}$$

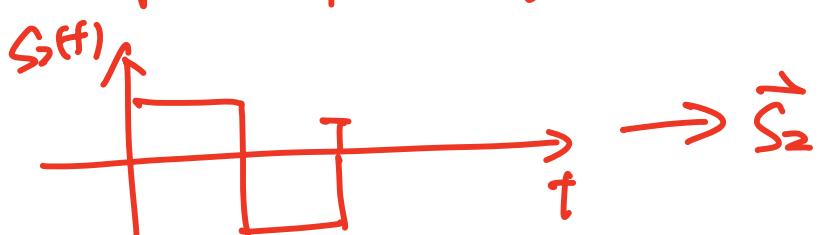
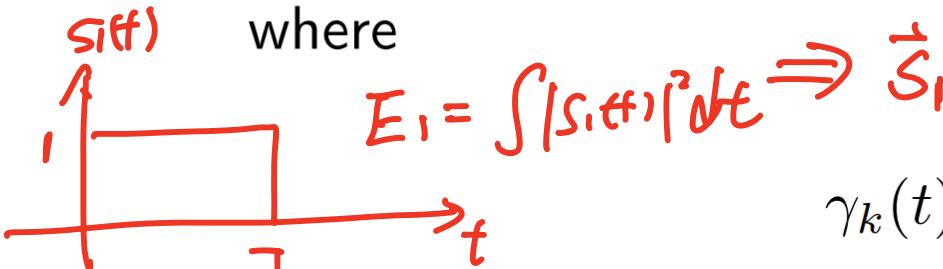
$$③ \quad \underline{\gamma_2(t) = s_2(t) - c_{21}\phi_1(t)}, \text{ and let } \underline{E_2 = \int_{-\infty}^{\infty} \gamma_2^2(t) dt}$$

$$④ \quad \underline{\phi_2(t) = \gamma_2(t)/\sqrt{E_2}}$$

$$⑤ \quad \underline{\phi_k(t) = \gamma_k(t)/\sqrt{E_k}}$$

$$S_{k-1} = \{\phi_1(t) \dots \phi_{k-1}(t)\}$$

where



$$\gamma_k(t) = \underline{s_k(t)} - \sum_{i=1}^{k-1} c_{ki} \phi_i(t)$$

$$c_{ki} = \langle s_k(t), \phi_i(t) \rangle$$

$$E_k = \int_{-\infty}^{\infty} \gamma_k^2(t) dt$$

# Orthonormal Expansions in $\mathcal{L}_2$

- Given an orthonormal basis  $\{\phi_1(t), \phi_2(t), \dots\}$  of  $\mathcal{L}_2$ , any  $\mathcal{L}_2$  signal  $x(t)$  can be represented as  $x(t) = \sum_j x_j \phi_j(t)$ .
  - $x_j = \langle x(t), \phi_j(t) \rangle = \int x(t) \phi_j^*(t) dt$ .
  - $x = (x_1, x_2, \dots)$
  - Equality holds in the sense of MSE=0.

$$\begin{aligned} & \bullet \int_{-\infty}^{\infty} |x(t) - \sum_j x_j \phi_j(t)|^2 dt = 0 \\ & - E_x = \int_{-\infty}^{\infty} |x(t)|^2 dt = \sum_j |x_j|^2 = \|x\|^2 \\ & - \langle x(t), s(t) \rangle = \langle x, s \rangle \end{aligned}$$

- Example

- Fourier Series
- Sampling Theorem

orthonormal basis for some subspace of  $\mathcal{L}_2$

{ Time-limited  $\mathcal{L}_2$  functions  $\rightarrow$

Band-limited  $\mathcal{L}_2$  functions  $\rightarrow$

# Orthonormal Expansions in $\mathcal{L}_2$

- Fourier Series

Given a function  $v(t)$ , with duration  $[-T/2, T/2]$

- Define  $\theta_k(t) = e^{2\pi i k t / T} \text{rect}(t/T)$
- $\{\theta_k(t)\}_k$  are orthogonal functions, with  $\|\theta_k(t)\|^2 = T$
- Let  $\phi_k(t) = \theta_k(t)/\sqrt{T}$ :

$$\phi_k(t) = \frac{1}{\sqrt{T}} e^{2\pi i k t / T} \text{rect}(t/T)$$

- By FSE,  $v(t)$  can be written as

$$v(t) = \sum_k \alpha_k \phi_k(t)$$

where  $\alpha_k = \langle v(t), \phi_k(t) \rangle$ .

# Orthonormal Expansions in $\mathcal{L}_2$

- Sampling Theorem

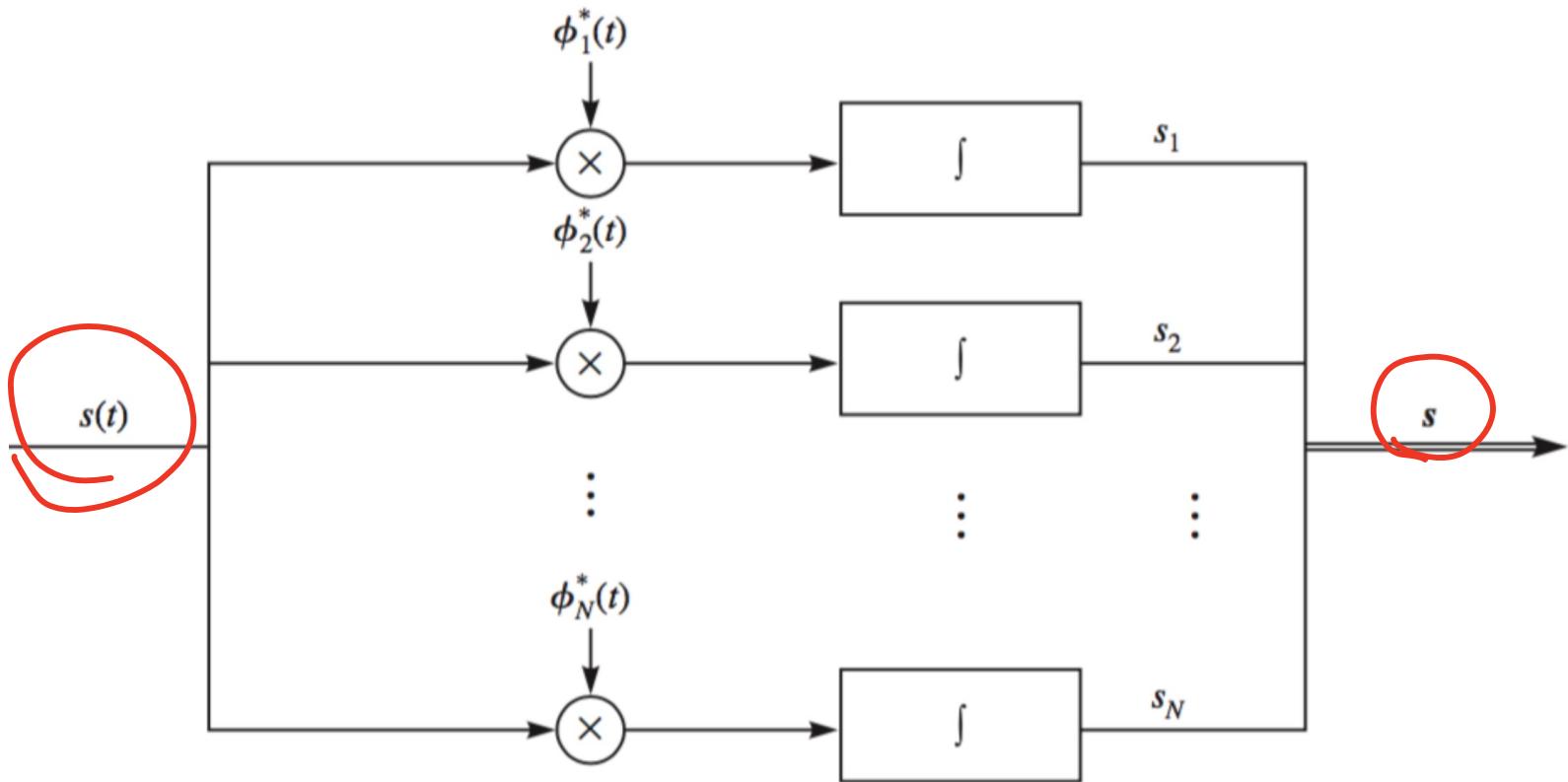
Sampling Theorem: If  $x(t)$  is a band-limited in  $W$ , then

$$x(t) = \sum_{n=-\infty}^{\infty} x(nT) \text{sinc} \left[ \left( \frac{t}{T} - n \right) \right]$$

where  $\text{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$ ,  $T = 1/2W$ .

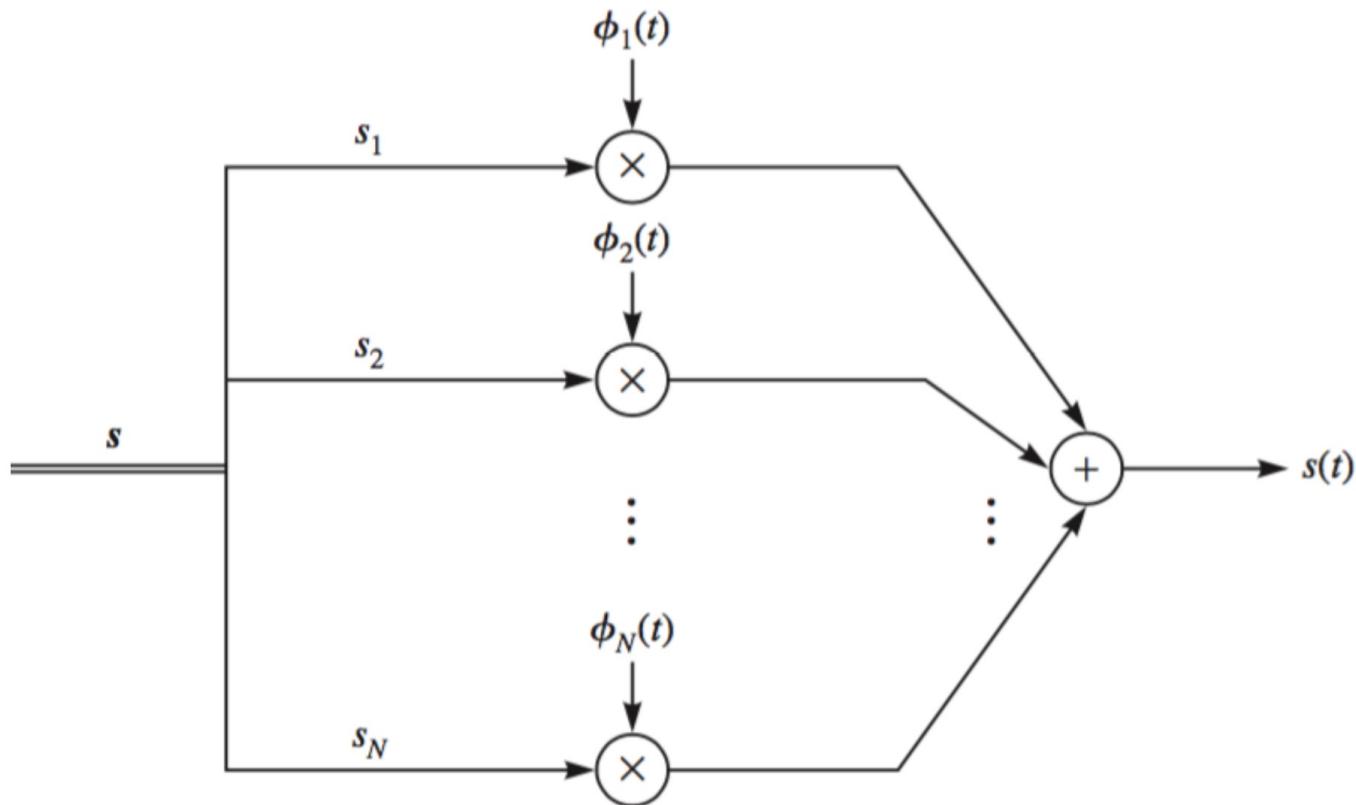
- $x(nT)$  is coefficient ( $x_k$ )
- $\{\text{sinc} \left[ \left( \frac{t}{T} - n \right) \right]\}_{n=-\infty}^{\infty}$  are orthogonal function
- Equality holds in sense of  $E_e = \int_{-\infty}^{\infty} |x(t) - \hat{x}(t)|^2 dt = 0$
- View signal  $x(t)$  as vectors  $x(nT)$

# Map Signal to Vector



$$s_k = \int_{-\infty}^{\infty} s(t) \phi_k^*(t) dt = \langle s(t), \phi_k(t) \rangle$$

# Map Vector to Signal



$$s(t) = \sum_{k=1}^N s_k \phi_k(t)$$



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Thanks for your kind attention!

Questions?