
VE320 – Summer 2024

Introduction to Semiconductor Devices

Instructor: Yaping Dan (但亚平)
yaping.dan@sjtu.edu.cn

Chapter 12 Bipolar Junction Transistor

Outline

12.1 Review and example

12.2 Bipolar Junction transistor

12.3 Early Effect

12.4 Summary

12.5 Quantitative analysis of BJT gain

Outline

12.1 Review and example

12.2 Bipolar Junction transistor

12.3 Early Effect

12.4 Summary

12.5 Quantitative analysis of BJT gain

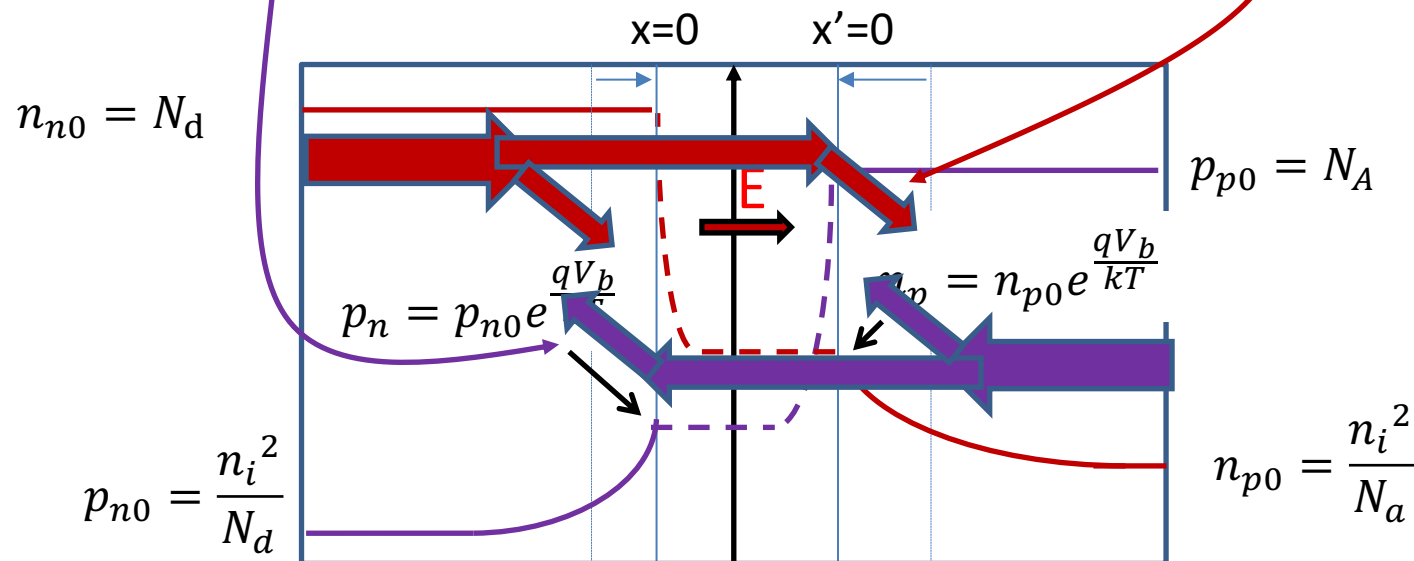
12.1 Previously: pn Junction Current

- charge carrier transport: forward bias: current ratio

$$J_n = -qD_n \frac{dn_p}{dx} = -\frac{qD_n n_{p0}}{L_n} (e^{\frac{qV_b}{kT}} - 1)e^{-x/L_n}$$

$$J_p = -qD_p \frac{dp_n}{dx} = -\frac{qD_p p_{n0}}{L_p} (e^{\frac{qV_b}{kT}} - 1)e^{x/L_p}$$

$$\frac{J_n}{J_p} = \frac{D_n N_d / L_n}{D_p N_a / L_p}$$



Assumption: No recombination-generation in depletion region.

12.1 Example: pn Junction Current

Finding L_n, τ_n in **p-type** region because electrons are minority carriers.

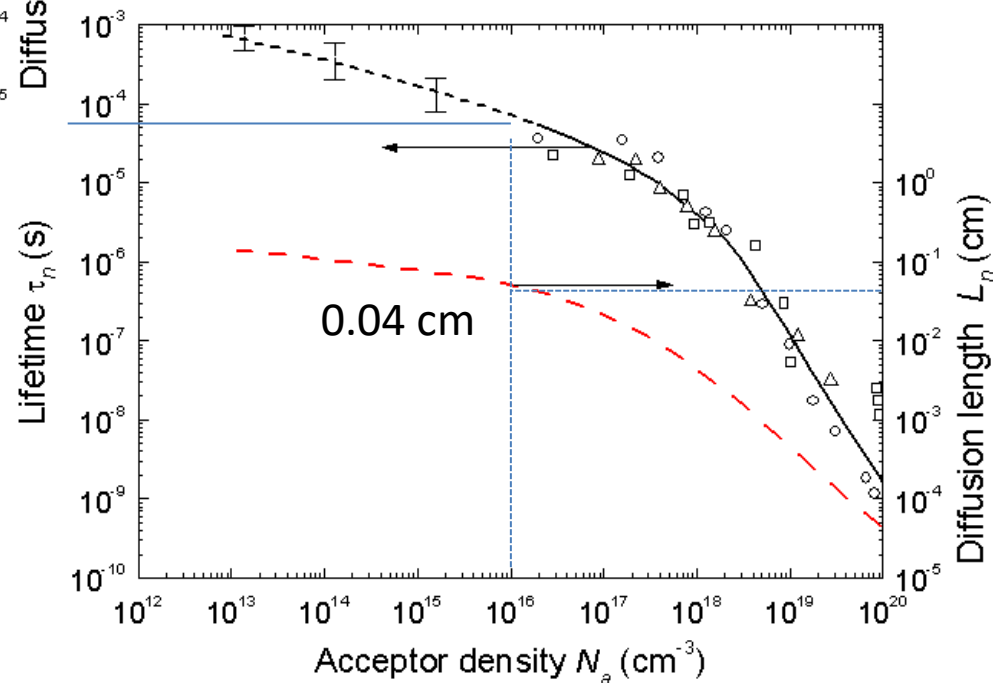
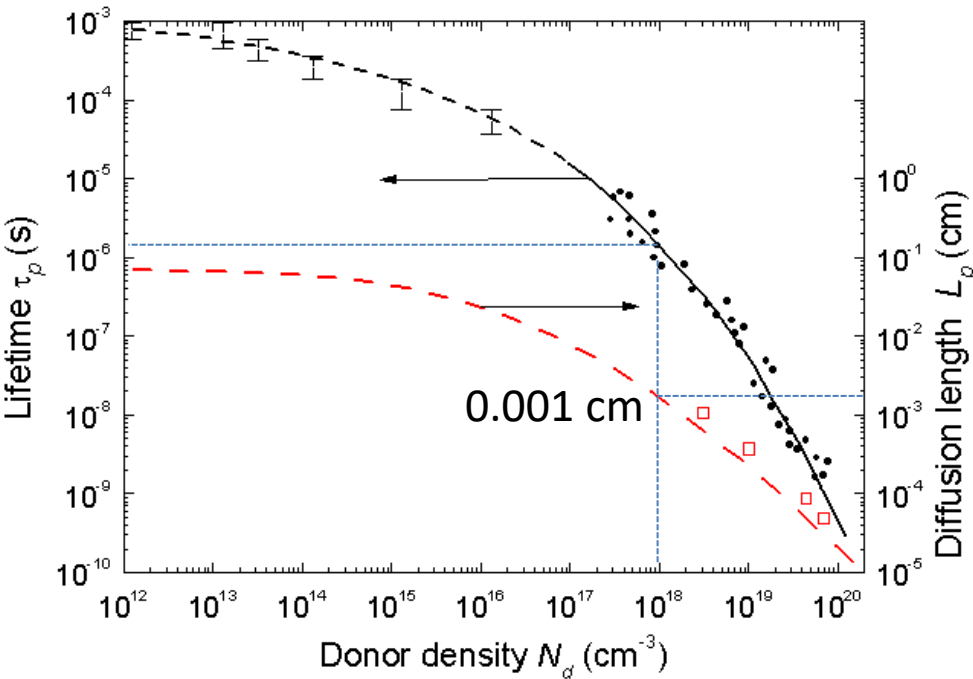
$$\text{For } N_a = 10^{16} \text{ cm}^{-3} \quad L_n = 0.04 \text{ cm} \quad \tau_n = 5 \times 10^{-5} \text{ s}$$

Finding L_p, τ_p in **n-type** region because holes are minority carriers.

$$\text{For } N_d = 10^{18} \text{ cm}^{-3} \quad L_p = 0.0015 \text{ cm} \quad \tau_p = 1.5 \times 10^{-6} \text{ s}$$

$$\frac{J_n}{J_p} = \frac{D_n N_d / L_n}{D_p N_a / L_p} = \frac{L_n / \tau_n N_d}{L_p / \tau_p N_a} \approx \frac{\frac{4 \times 10^{-2}}{5 \times 10^{-5}}}{\frac{1.5 \times 10^{-3}}{1.5 \times 10^{-6}}} \times \frac{10^{18}}{10^{16}} = 80$$

12.1 Example: pn Junction Current



<http://www.ioffe.ru/SVA/NSM/Semicond/Si/>

Outline

12.1 Review and example

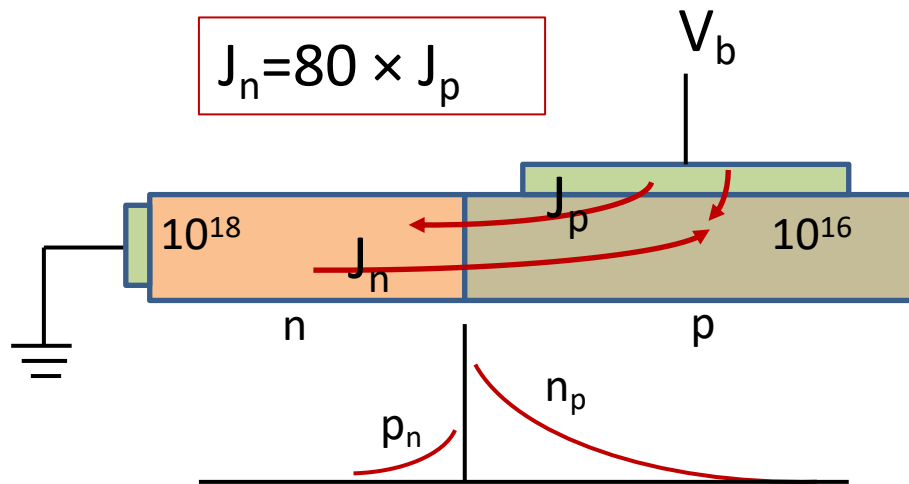
12.2 Bipolar Junction transistor

12.3 Early Effect

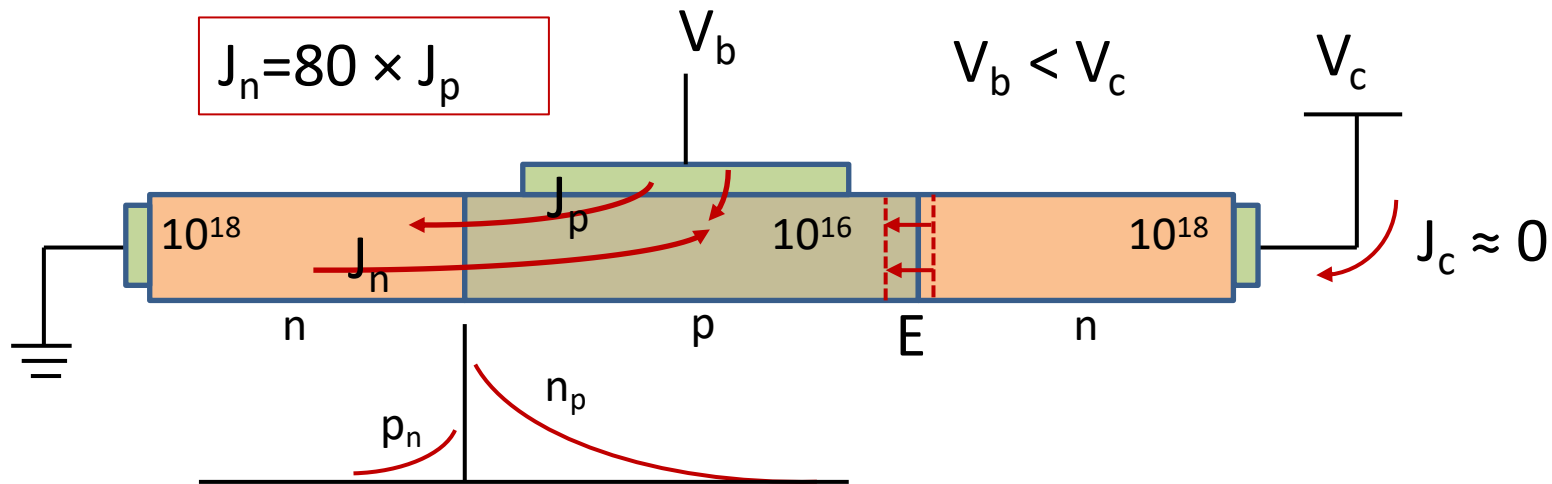
12.4 Summary

12.5 Quantitative analysis of BJT gain

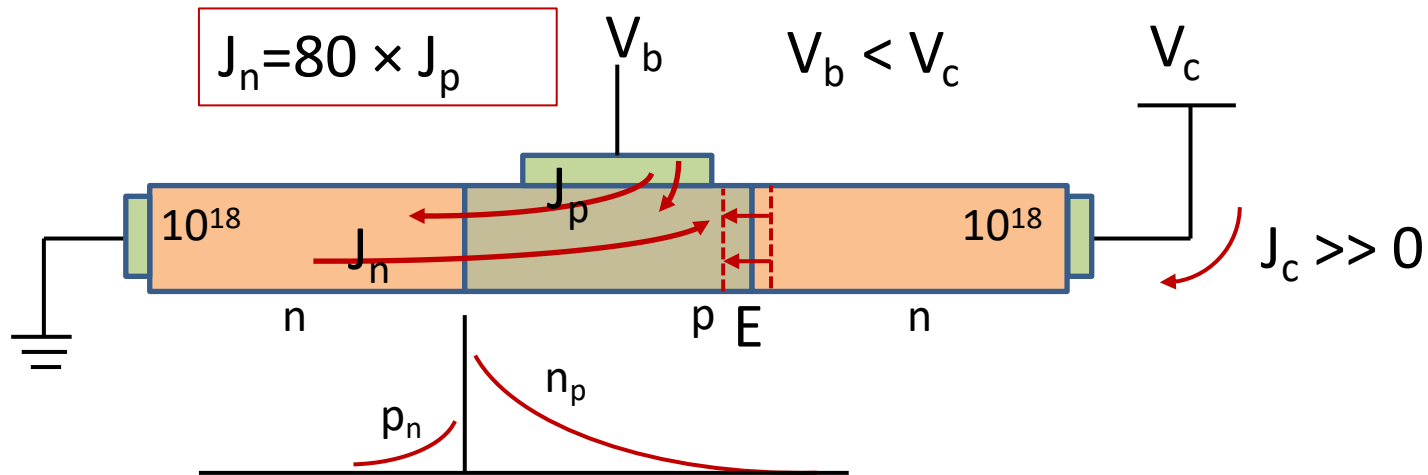
12.2 Bipolar Junction transistor



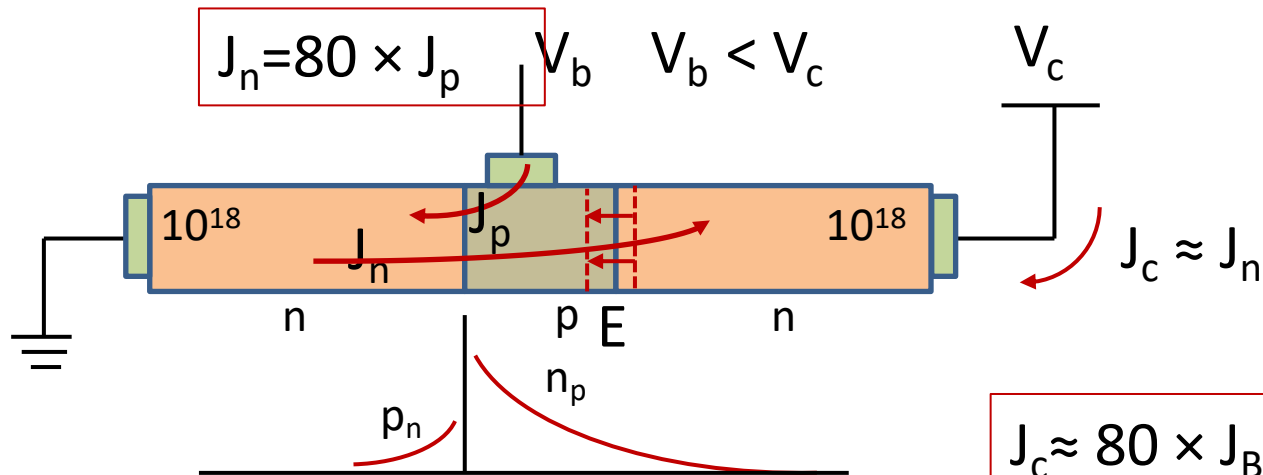
12.2 Bipolar Junction transistor



12.2 Bipolar Junction transistor

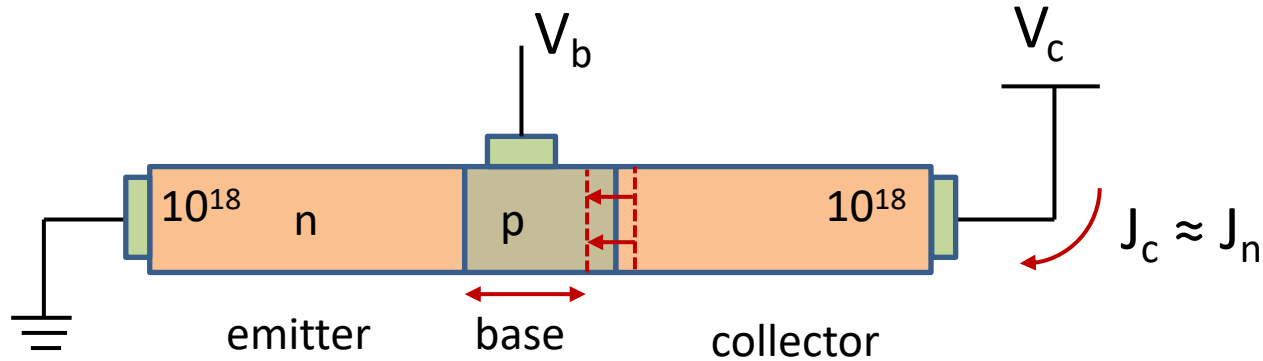


12.2 Bipolar Junction transistor



$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

12.2 Bipolar Junction transistor

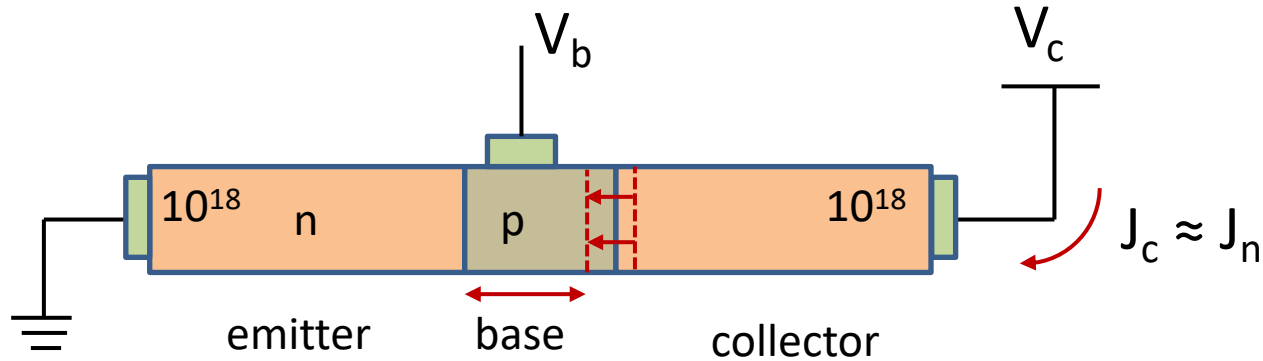


$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

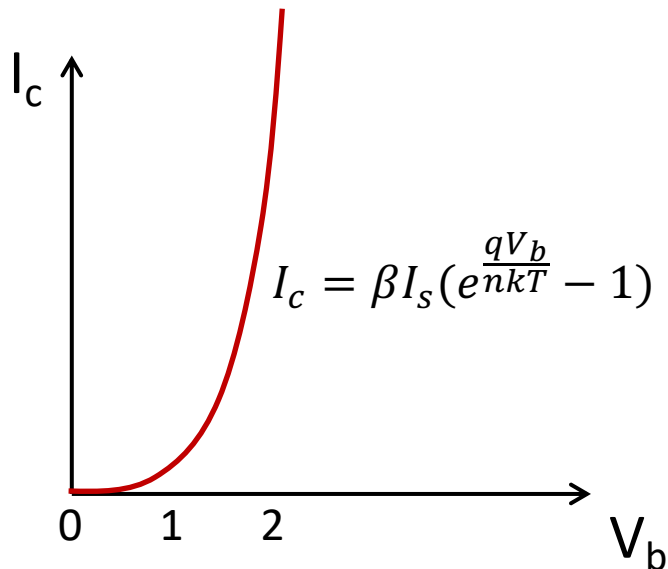
BJT Characteristics:

1. Base width smaller \rightarrow higher gain
2. Larger emitter-base concentration ratio \rightarrow higher gain

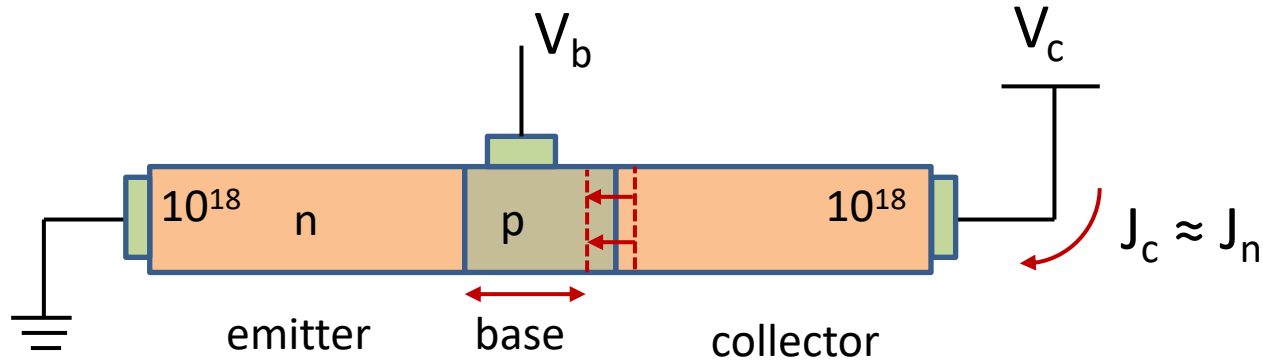
12.2 Bipolar Junction transistor: I-V



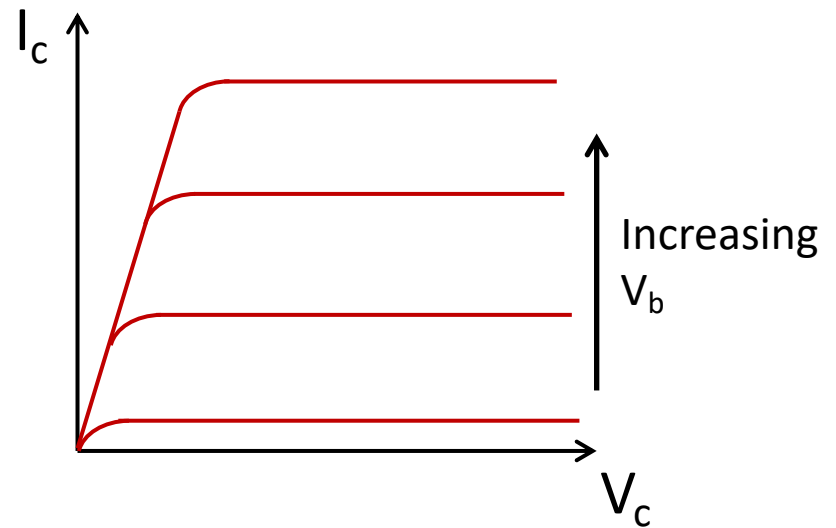
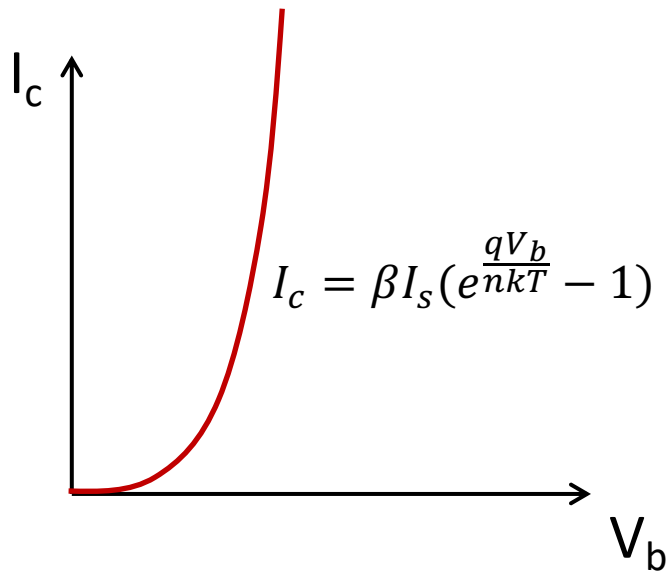
$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$



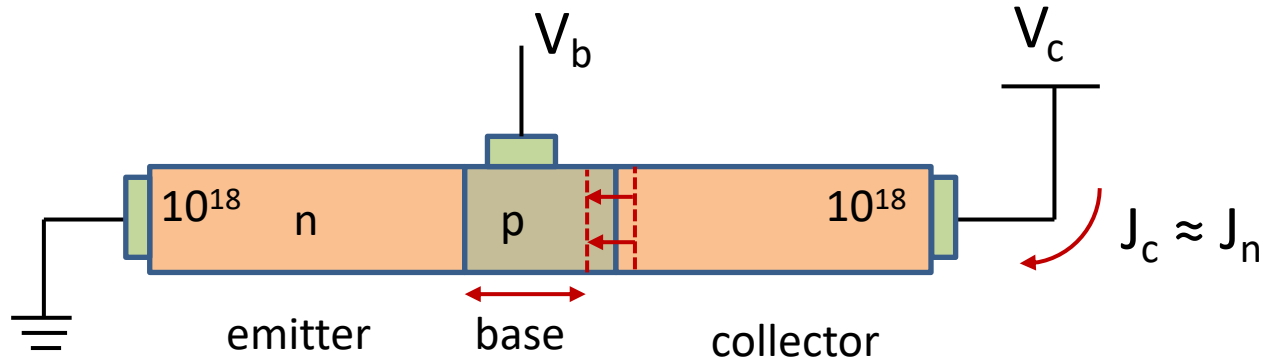
12.2 Bipolar Junction transistor: I-V



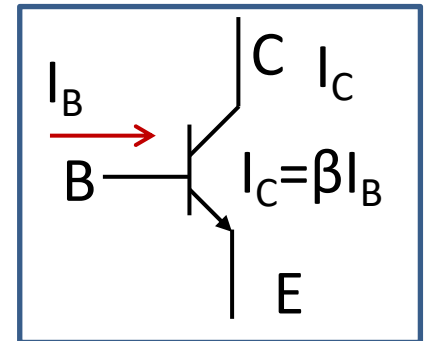
$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$



12.2 Bipolar Junction transistor: I-V



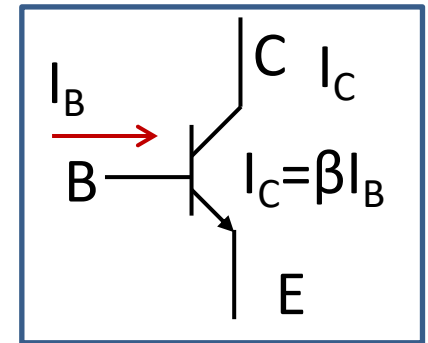
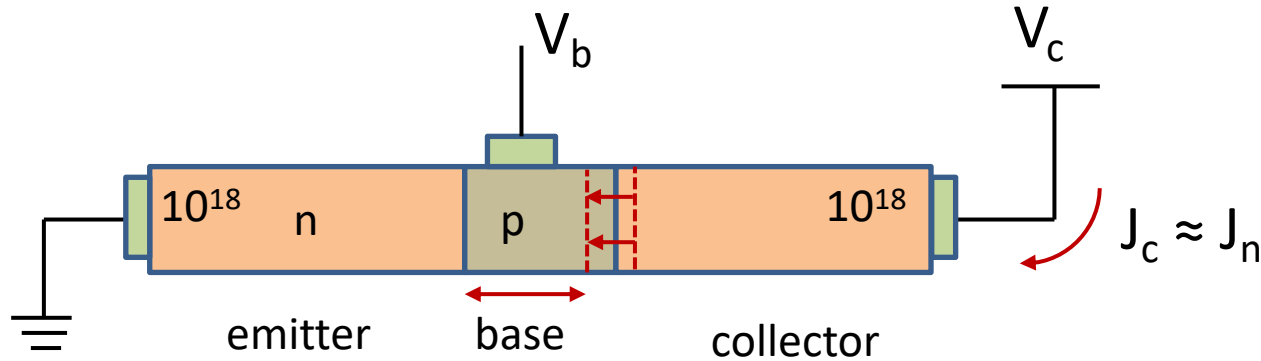
$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$



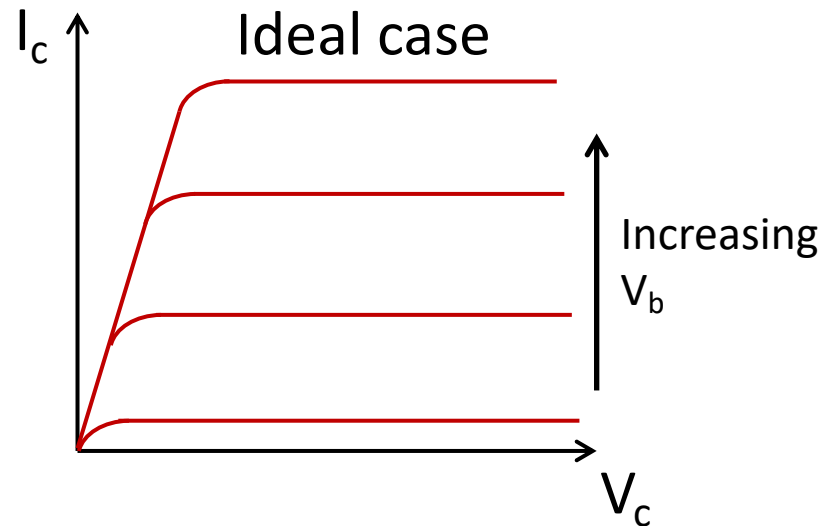
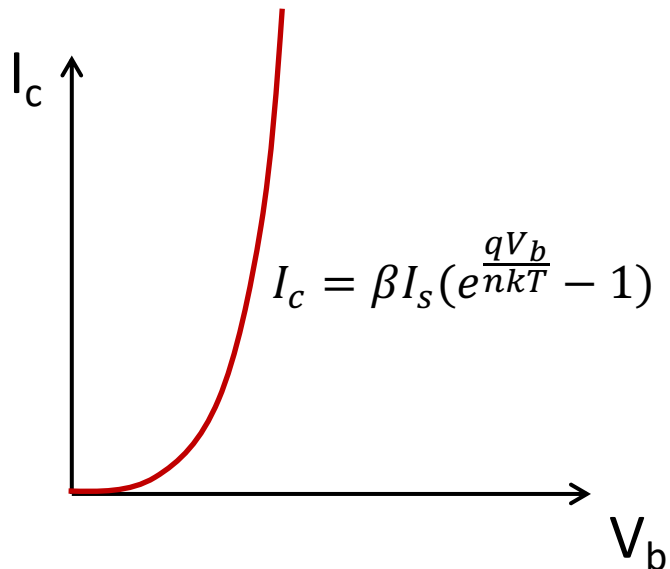
Basic facts:

1. Narrower base \rightarrow larger gain
2. $\beta \approx N_D/N_A$, higher emitter-to-base doping ratio \rightarrow higher gain

12.2 Bipolar Junction transistor: I-V



$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$



Outline

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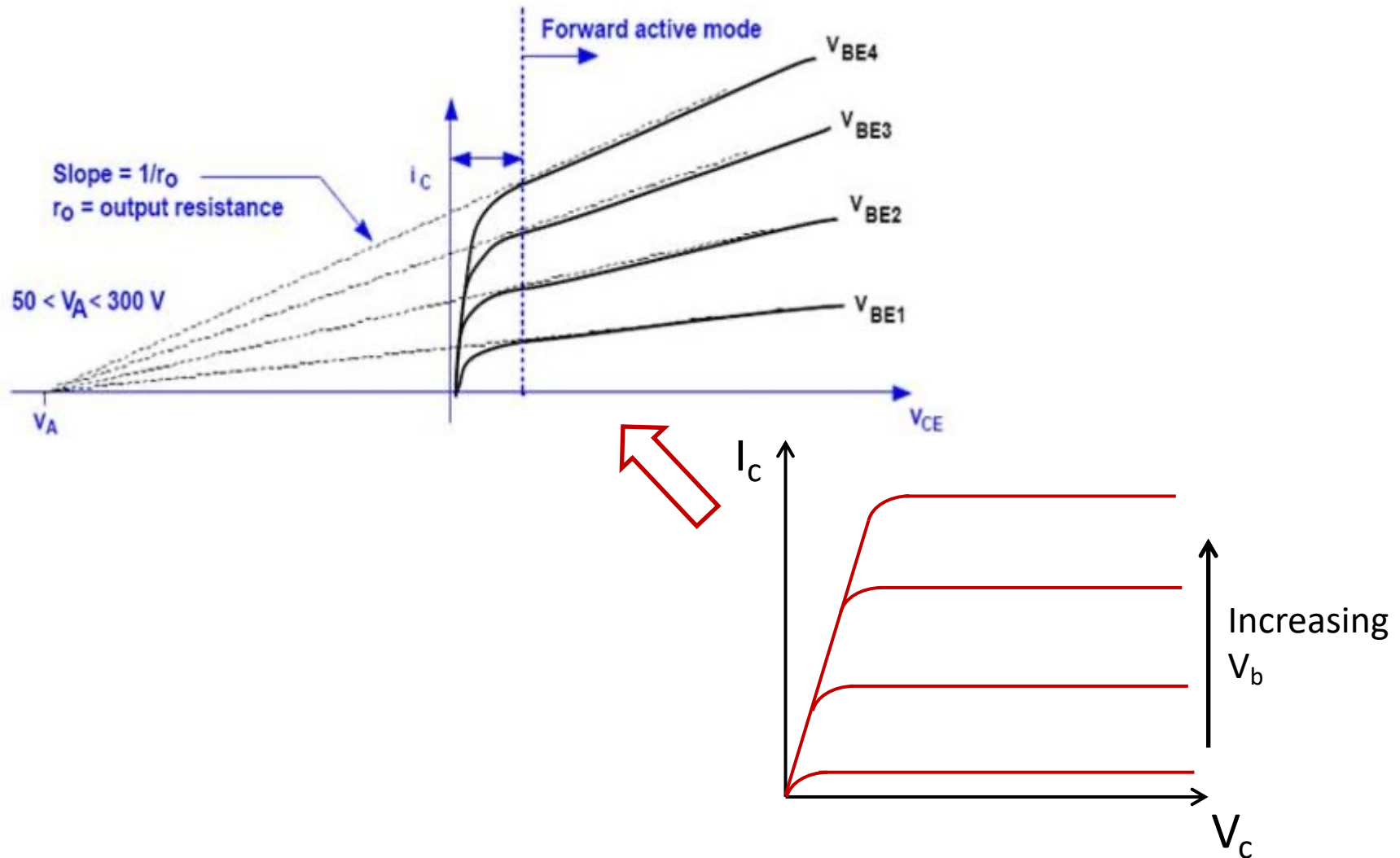
12.2 Bipolar Junction transistor

12.3 Early Effect

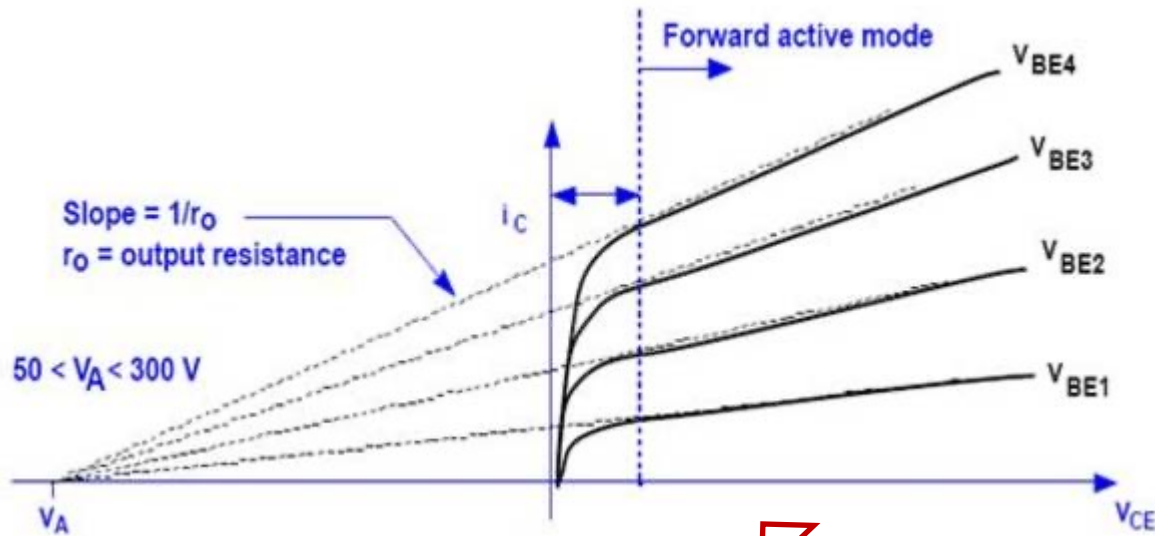
12.4 Summary

12.5 Quantitative analysis of BJT gain

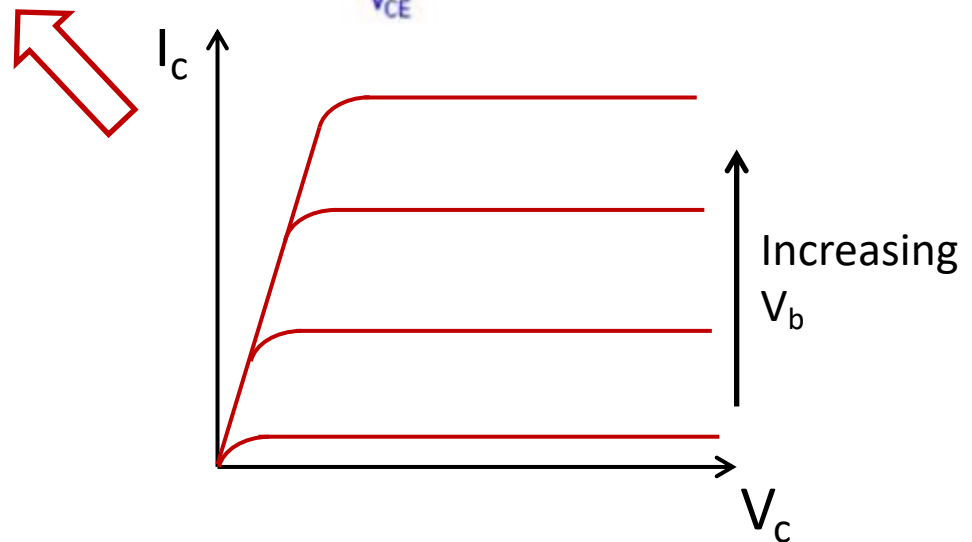
12.3 Early Effect



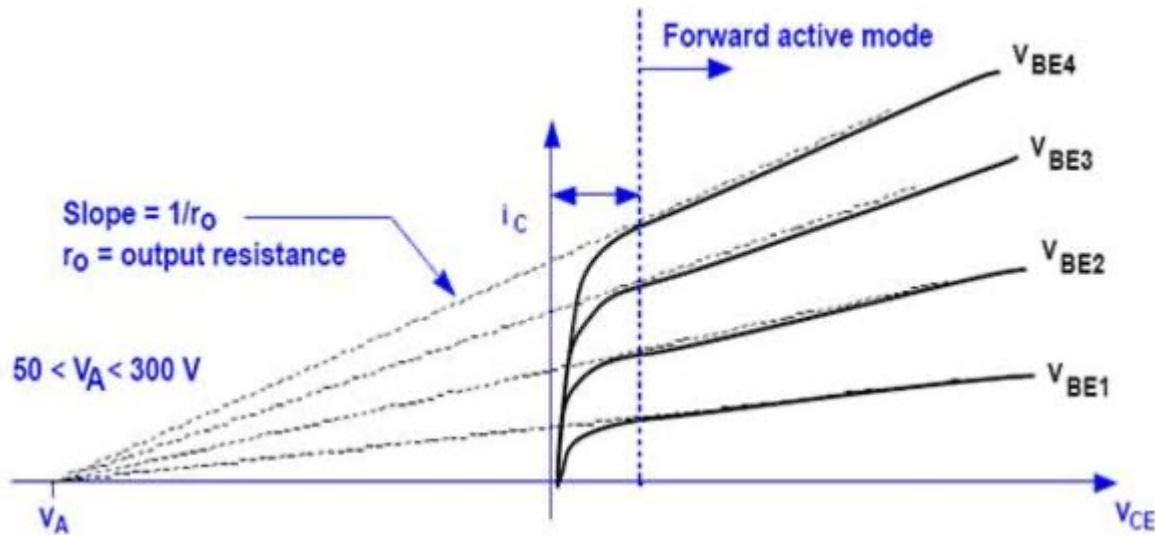
12.3 Early Effect



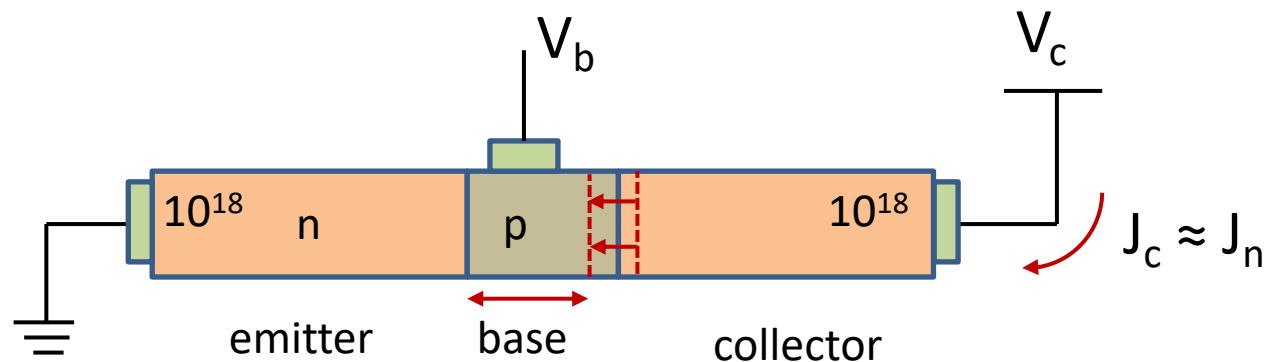
James M. Early



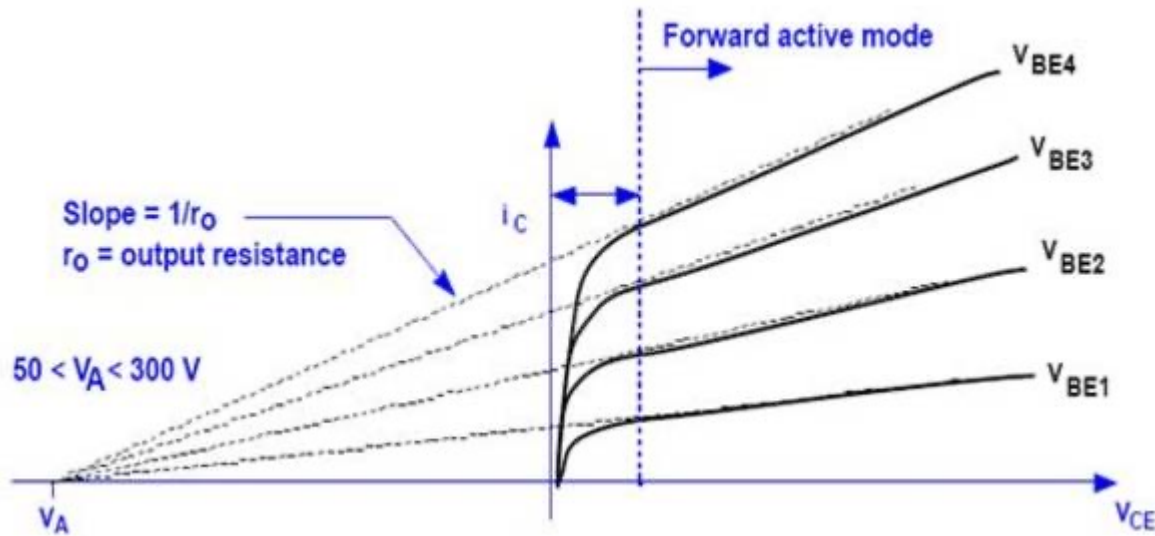
12.3 Early Effect



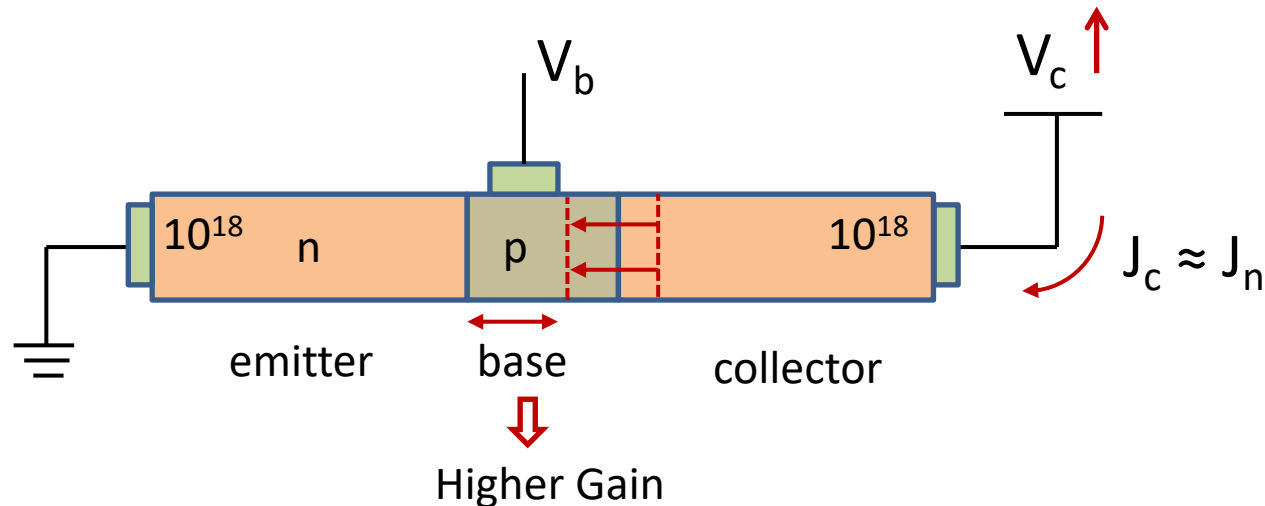
[James M. Early](#)



12.3 Early Effect



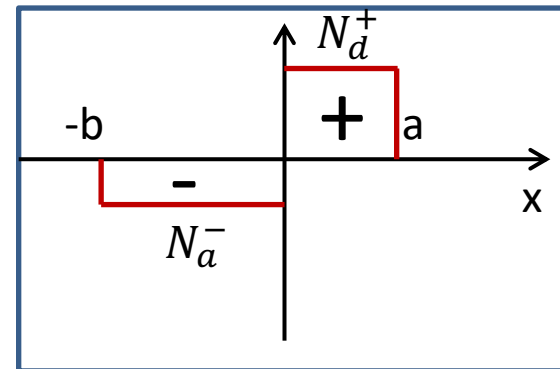
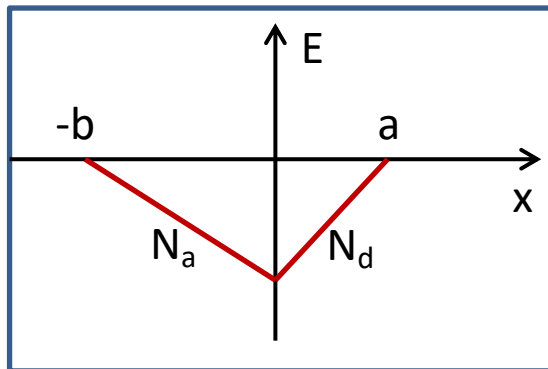
[James M. Early](#)



Previously...

$$a = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{N_a}{N_d} \frac{1}{N_a + N_d}}$$

$$N_A^- b = N_D^+ a \Rightarrow b = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{N_d}{N_a} \frac{1}{N_a + N_d}}$$



Previously...

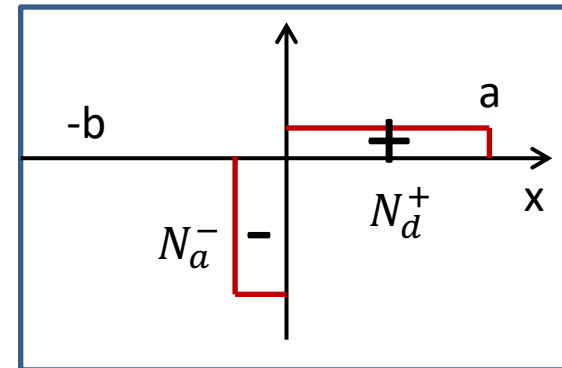
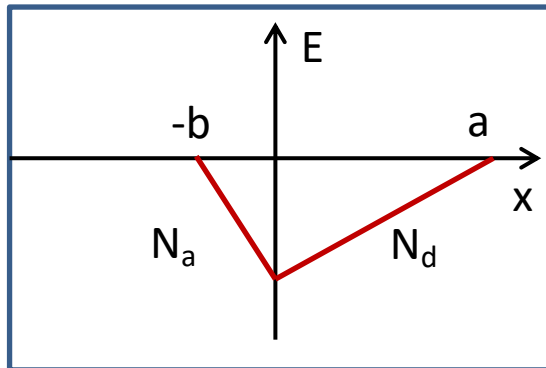
$$a = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{N_a}{N_d} \frac{1}{N_a + N_d}}$$

$$N_A^- b = N_D^+ a \Rightarrow b = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{N_d}{N_a} \frac{1}{N_a + N_d}}$$

$$N_a = 100 N_d$$

$$a = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{\cancel{100N_a}}{N_d} \frac{1}{\cancel{100N_a}}}$$

$$b = \sqrt{\frac{2\varepsilon(V_{bi}-V_R)}{q} \frac{\cancel{N_a}}{100\cancel{N_a}} \frac{1}{100N_d}}$$



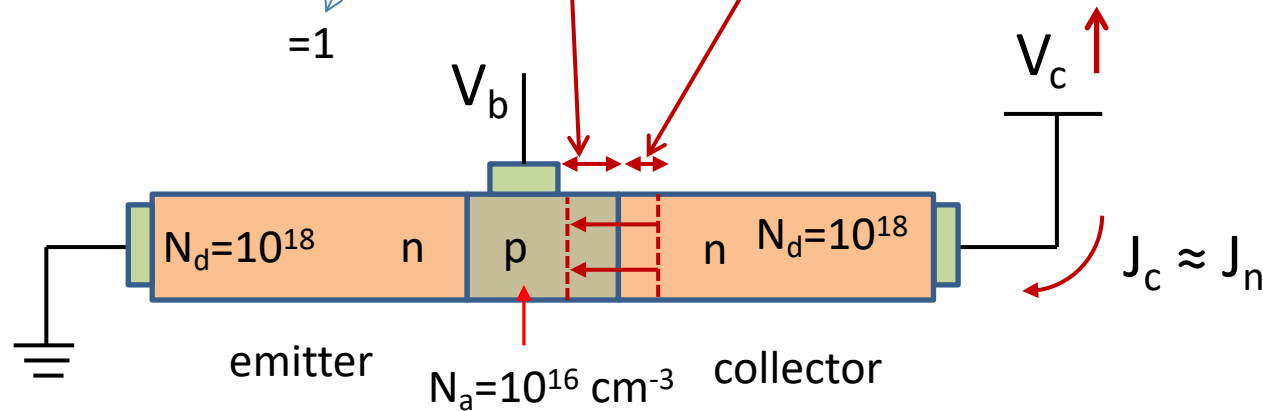
12.3 Early Effect

$$a = \sqrt{\frac{2\epsilon(V_{bi}-V_R)}{q} \frac{10^{16}}{10^{18}} \frac{1}{10^{16} + 10^{18}}} = 1/10000$$

$$b = \sqrt{\frac{2\epsilon(V_{bi}-V_R)}{q} \frac{10^{18}}{10^{16}} \frac{1}{10^{16} + 10^{18}}} = 1$$



James M. Early



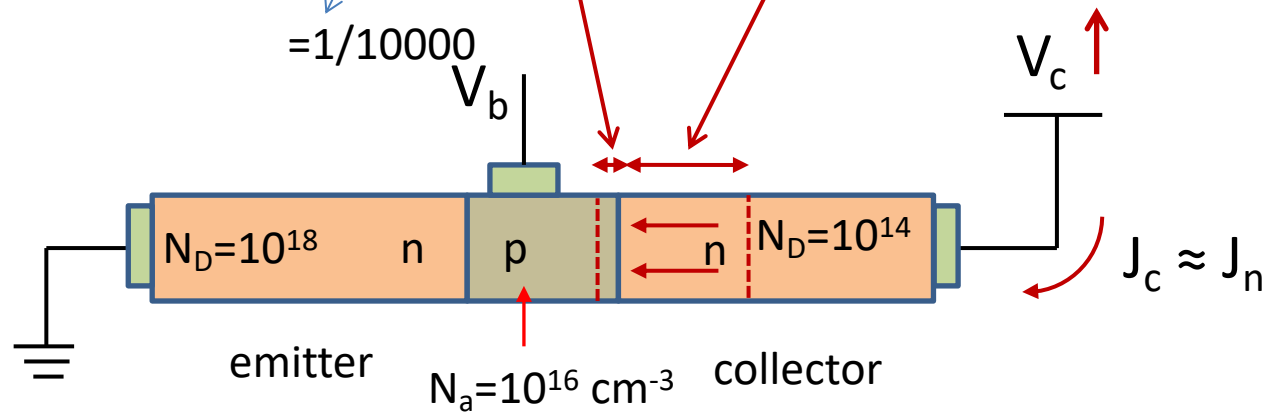
12.3 Early Effect

$$a = \sqrt{\frac{2\epsilon(V_{bi}-V_R)}{q} \frac{10^{16}}{10^{14}} \frac{1}{10^{16} + 10^{14}}} = 1$$

$$b = \sqrt{\frac{2\epsilon(V_{bi}-V_R)}{q} \frac{10^{14}}{10^{16}} \frac{1}{10^{16} + 10^{14}}}$$



James M. Early



Outline

12.1 Review and example

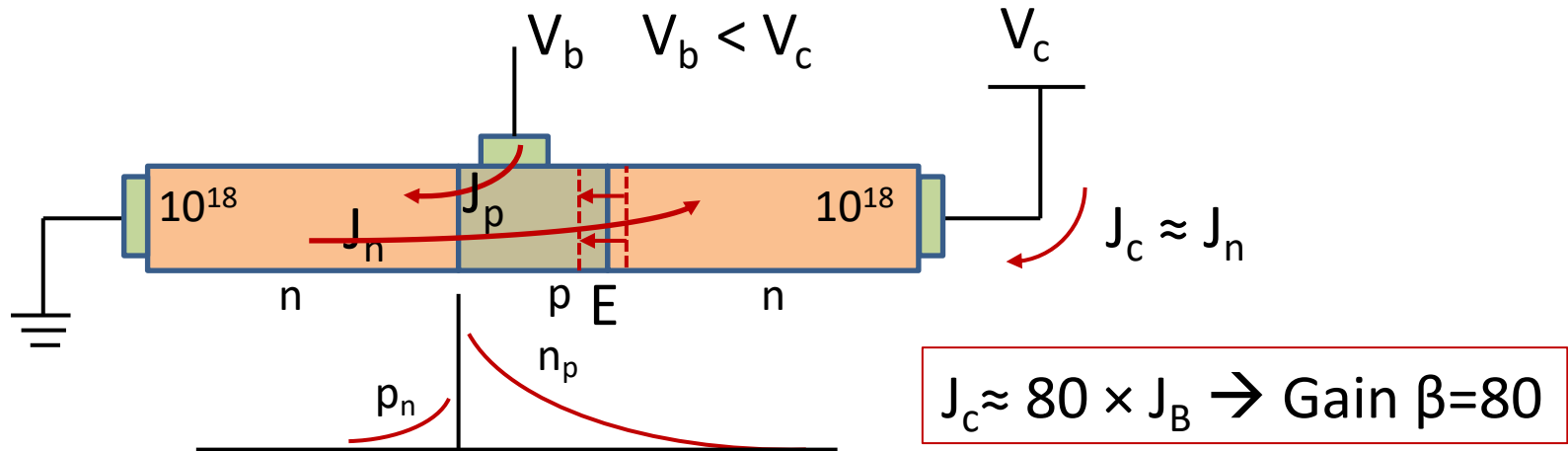
12.2 Bipolar Junction transistor

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12.4 Summary

12.5 Quantitative analysis of BJT gain

12.4 Summary



1. highest doping concentration is limited by solubility ($<10^{20}$)
2. Lowest doping concentration is limited by n_i and fabrication process

Basic facts:

1. Narrower base \rightarrow larger gain
2. $\beta \approx N_D/N_A$, higher emitter-to-base doping ratio \rightarrow higher gain
3. Trade-off for base doping concentration (gain and Early effect)

Outline

12.1 Review and example

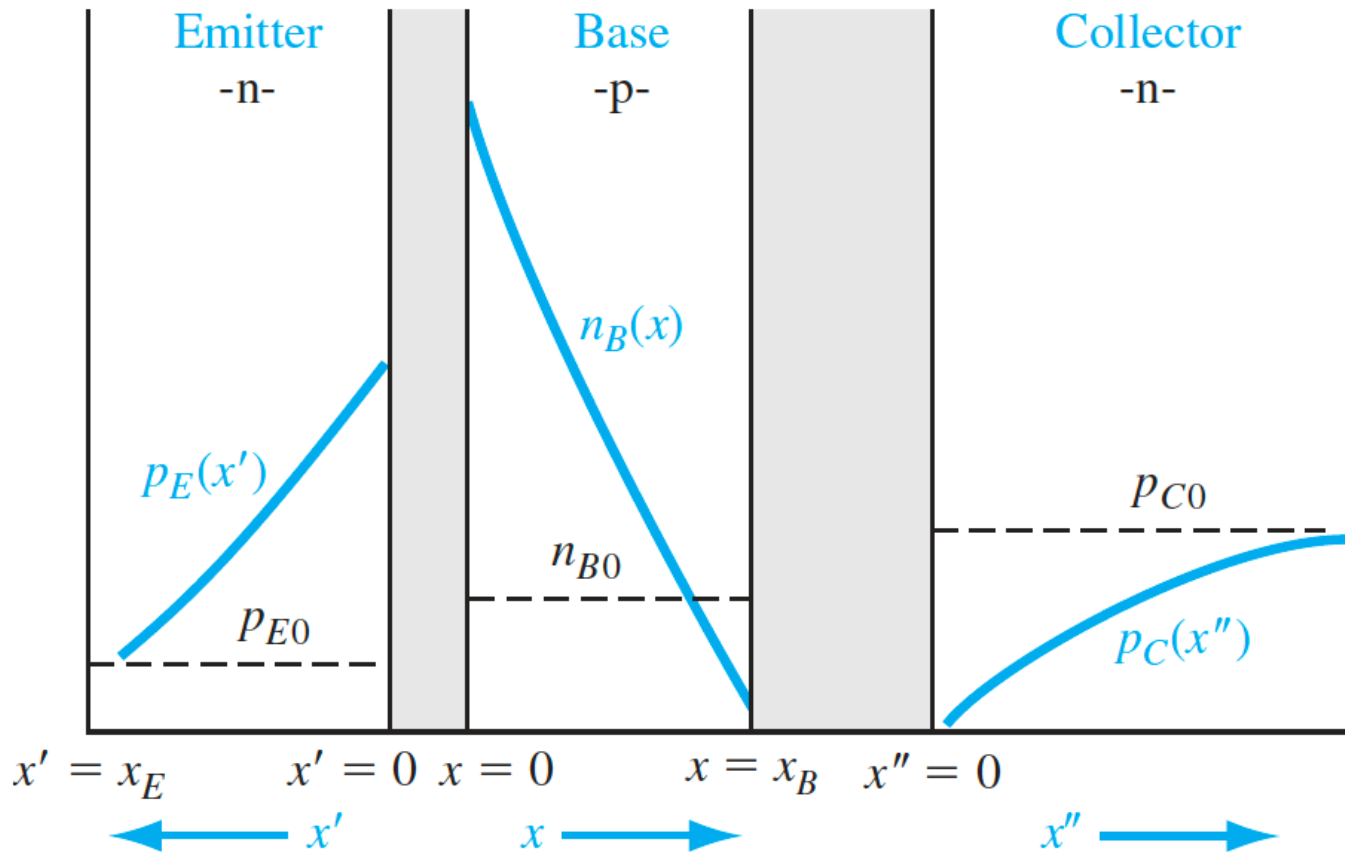
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12.3 Early Effect

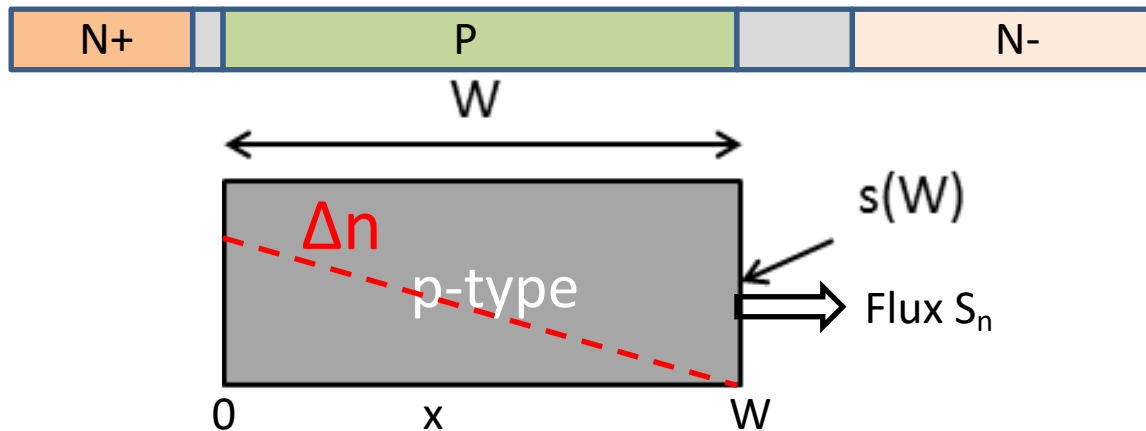
12.4 Summary

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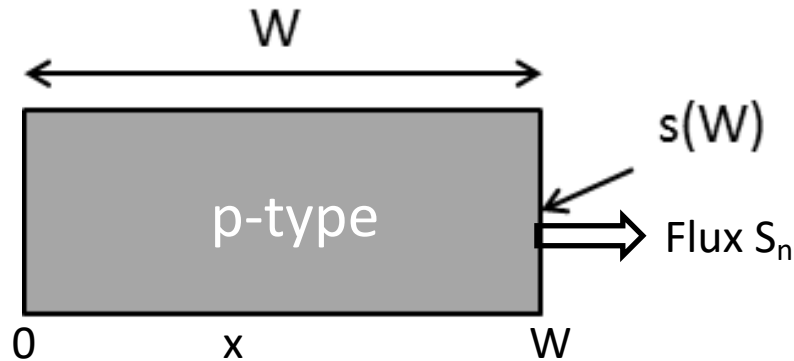


$$N_a = 10^{17} \text{ cm}^{-3}, D_n = 10 \text{ cm}^2/\text{s}, \tau_n = 10^{-7} \text{ s}, \text{SRV } s(x=W) = \infty$$
$$\Delta n(x=0) = 10^{14} \text{ cm}^{-3}$$

Find the electron flux S_n at $x=0$ and W , if

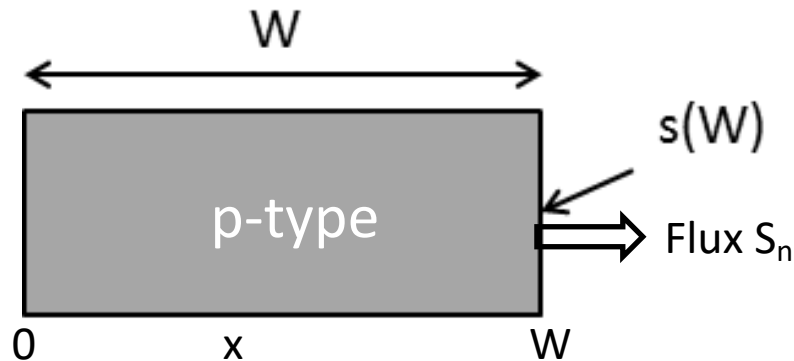
- 1) $W=20\mu\text{m}$
- 2) $W=2\mu\text{m}$

12.5 Quantitative analysis of BJT gain



$$0 = D_n \frac{\partial^2 \Delta n}{\partial x^2} - \frac{\Delta n}{\tau} \quad \Rightarrow \Delta n = A \exp\left(-\frac{x}{L_n}\right) + B \exp\left(\frac{x}{L_n}\right)$$

12.5 Quantitative analysis of BJT gain

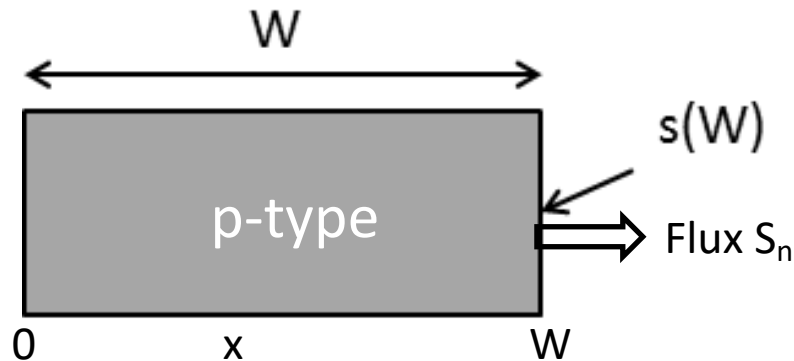


$$0 = D_n \frac{\partial^2 \Delta n}{\partial x^2} - \frac{\Delta n}{\tau}$$

$$\Rightarrow \Delta n = A \exp\left(-\frac{x}{L_n}\right) + B \exp\left(\frac{x}{L_n}\right)$$

$$\begin{cases} x = 0 \Rightarrow \Delta n_0 = \Delta n(x = 0) = A + B \\ x = W \Rightarrow \Delta n = A \exp\left(-\frac{W}{L_n}\right) + B \exp\left(\frac{W}{L_n}\right) = 0 \end{cases}$$

12.5 Quantitative analysis of BJT gain

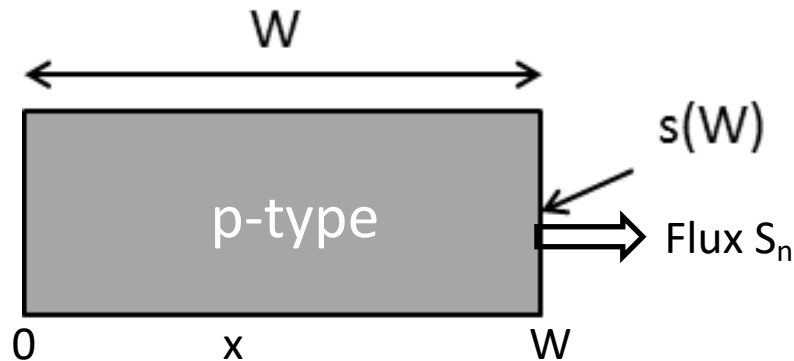


$$0 = D_n \frac{\partial^2 \Delta n}{\partial x^2} - \frac{\Delta n}{\tau} \quad \Rightarrow \Delta n = A \exp\left(-\frac{x}{L_n}\right) + B \exp\left(\frac{x}{L_n}\right)$$

$$A = (\Delta n)_0 \frac{\exp\left(\frac{W}{L_n}\right)}{\exp\left(\frac{W}{L_n}\right) - \exp\left(-\frac{W}{L_n}\right)}$$

$$B = (\Delta n)_0 \frac{\exp\left(-\frac{W}{L_n}\right)}{\exp\left(\frac{W}{L_n}\right) - \exp\left(-\frac{W}{L_n}\right)}$$

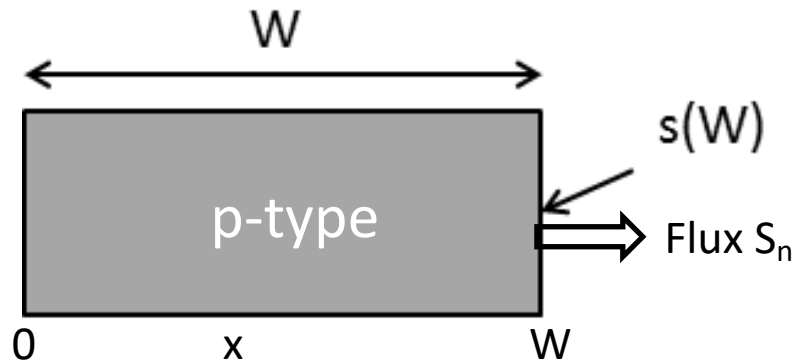
12.5 Quantitative analysis of BJT gain



$$0 = D_n \frac{\partial^2 \Delta n}{\partial x^2} - \frac{\Delta n}{\tau} \quad \Rightarrow \Delta n = A \exp\left(-\frac{x}{L_n}\right) + B \exp\left(\frac{x}{L_n}\right)$$

$$\Delta n(x) = (\Delta n)_0 \frac{\text{sh}\left(\frac{W-x}{L_n}\right)}{\text{sh}\left(\frac{W}{L_n}\right)}$$

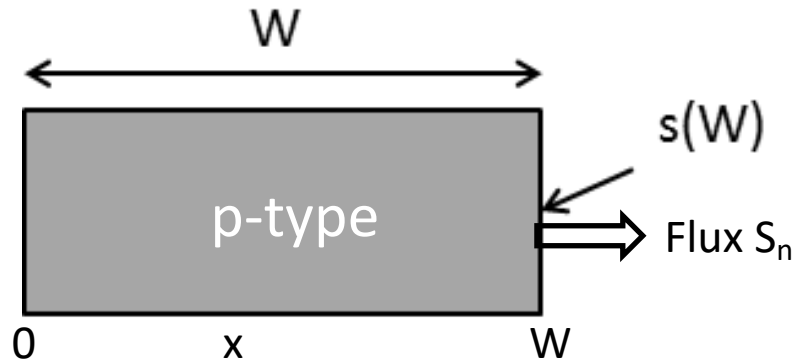
12.5 Quantitative analysis of BJT gain



$$0 = D_n \frac{\partial^2 \Delta n}{\partial x^2} - \frac{\Delta n}{\tau} \Rightarrow \Delta n = A \exp\left(-\frac{x}{L_n}\right) + B \exp\left(\frac{x}{L_n}\right)$$

$$\Delta n(x) = (\Delta n)_0 \frac{\text{sh}\left(\frac{W-x}{L_n}\right)}{\text{sh}\left(\frac{W}{L_n}\right)} \quad S_n = -D_n \frac{d\Delta n(x)}{dx} = \frac{D_n (\Delta n)_0}{L_n} \frac{\text{ch}\left(\frac{W-x}{L_n}\right)}{\text{sh}\left(\frac{W}{L_n}\right)}$$

12.5 Quantitative analysis of BJT gain

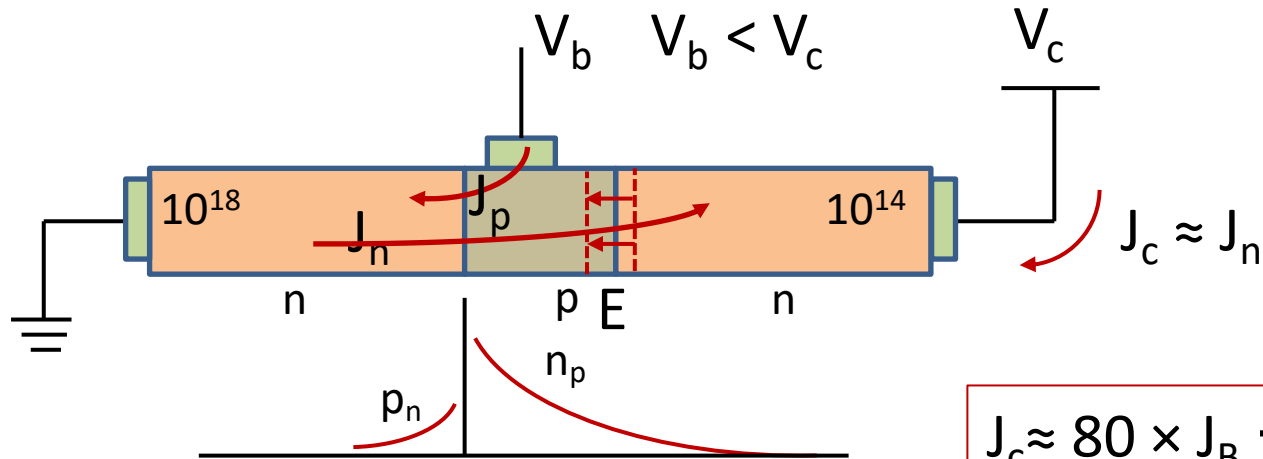


$$S_n = -D_n \frac{d\Delta n(x)}{dx} = \frac{D_n(\Delta n)_{B0}}{L_n} \frac{ch(\frac{W-x}{L_p})}{sh(\frac{W}{L_p})}$$

$$S_n(0) = \frac{D_n(\Delta n)_0}{L_n} \frac{ch(\frac{W}{L_p})}{sh(\frac{W}{L_p})}$$

$$S_n(W) = \frac{D_n(\Delta n)_0}{L_n} \frac{1}{sh(\frac{W}{L_p})}$$

12.5 Quantitative analysis of BJT gain



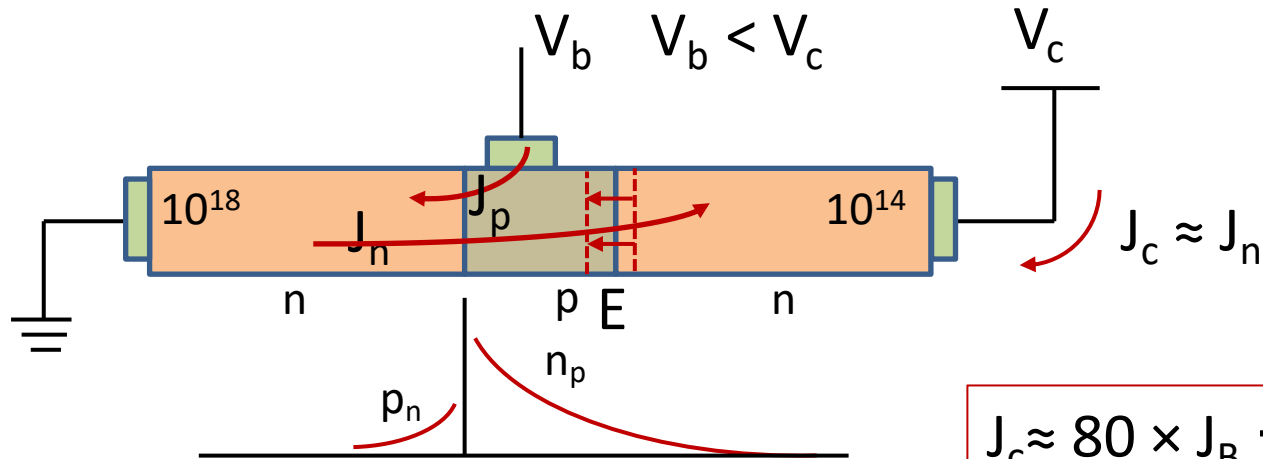
$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

$$S_n(0) = \frac{D_B \cdot (\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})} \quad S_n(W_b) = \frac{D_B \cdot (\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}$$

Electron flux from emitter to base: $S_n(0)$

Electron flux from base to collector: $S_n(W_b)$

12.5 Quantitative analysis of BJT gain



$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

$$S_n(0) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})}$$

$$S_n(W_b) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}$$

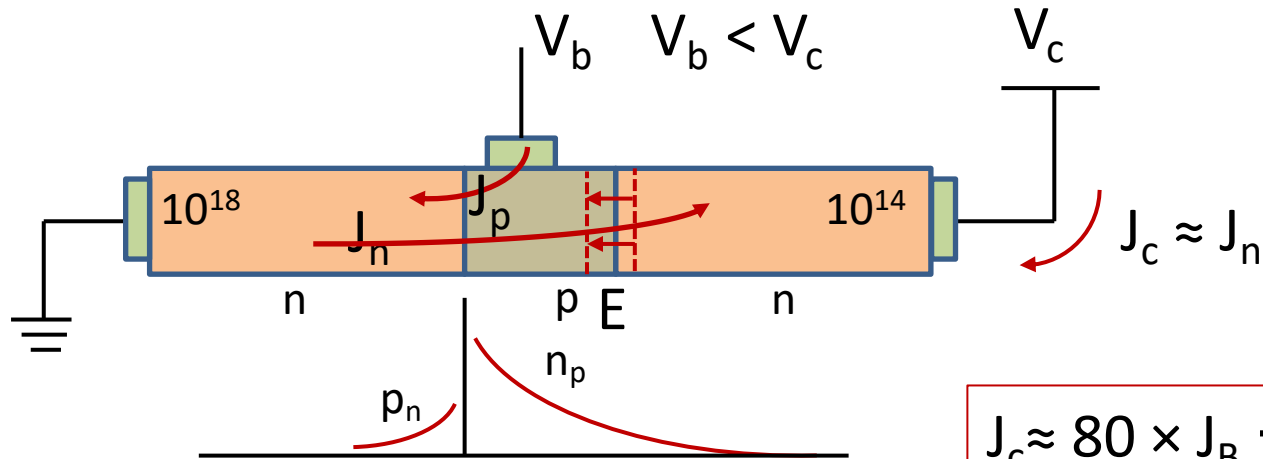
Electron flux from emitter to base: $S_n(0)$

Electron flux from base to collector: $S_n(W_b)$

Hole flux from base to emitter: S_p

$$S_p = D_E \frac{dp_E}{dx} = \frac{D_E p_{E0}}{L_E}$$

12.5 Quantitative analysis of BJT gain



$$S_n(0) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})}$$

$$S_n(W_b) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}$$

Electron flux from emitter to base: $S_n(0)$

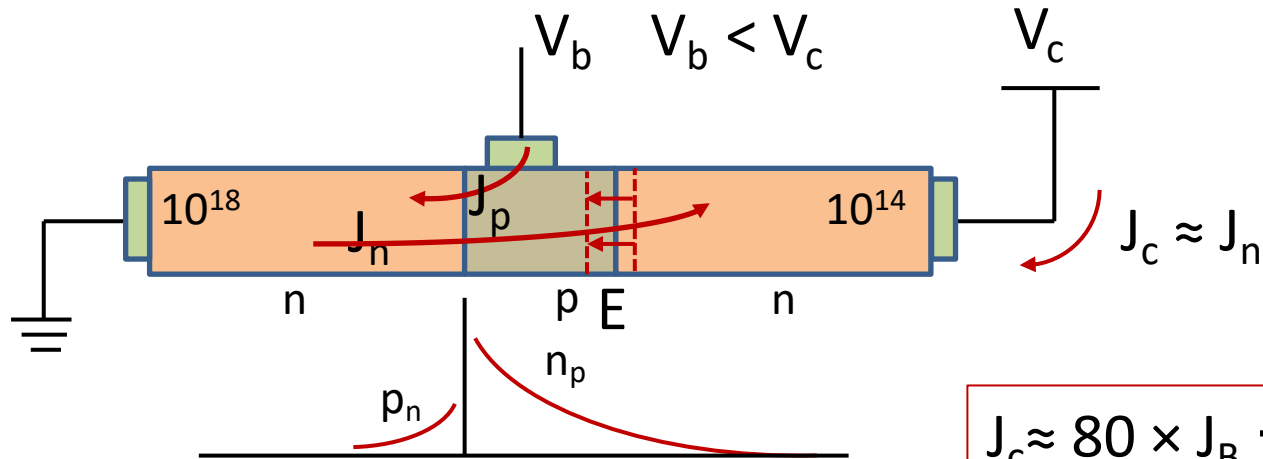
Electron flux from base to collector: $S_n(W_b)$

Hole flux from base to emitter: S_p

Base electrode flux: $S_p + S_n(0) - S_n(W_b)$

$$S_p = D_E \frac{dp_E}{dx} = \frac{D_E p_{E0}}{L_E}$$

12.5 Quantitative analysis of BJT gain



$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

$$S_n(0) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})}$$

$$S_n(W_b) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}$$

Electron flux from emitter to base: $S_n(0)$

Electron flux from base to collector: $S_n(W_b)$

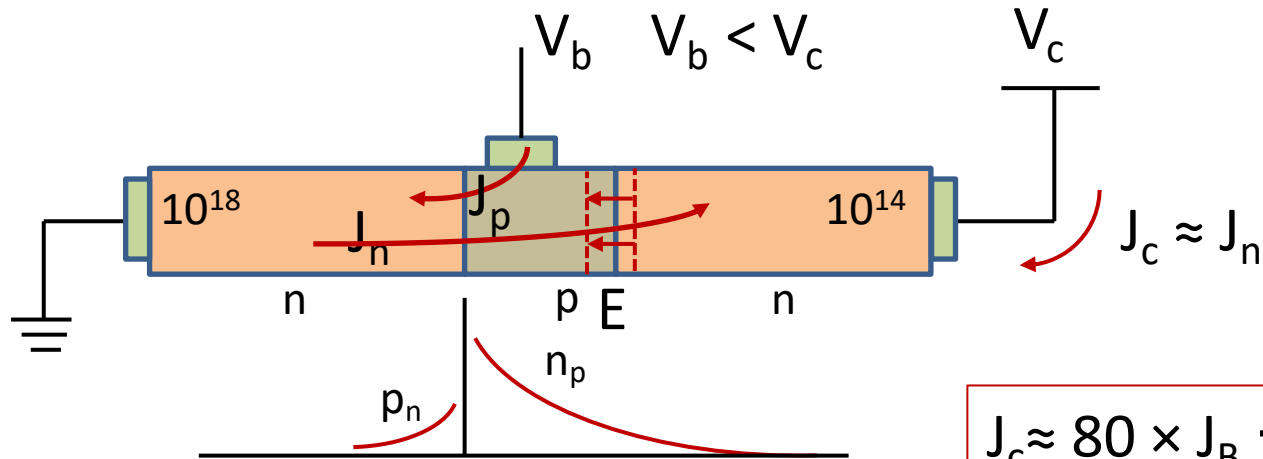
Hole flux from base to emitter: S_p

Base electrode flux: $S_p + S_n(0) - S_n(W_b)$

Gain β = collector flux / base electrode flux = $S_n(W_b) / (S_p + S_n(0) - S_n(W_b))$

$$S_p = D_E \frac{dp_E}{dx} = \frac{D_E p_{E0}}{L_E}$$

12.5 Quantitative analysis of BJT gain

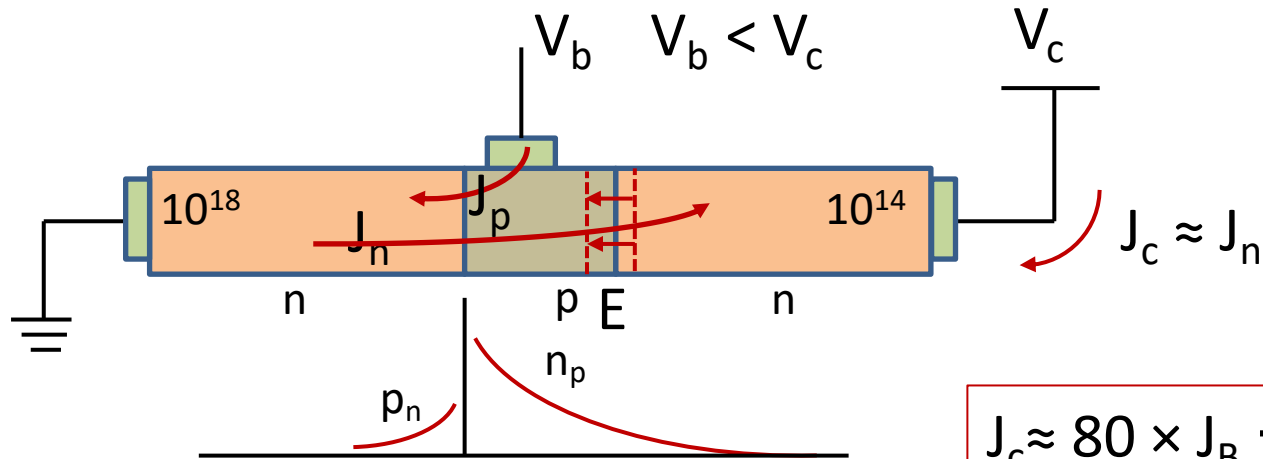


$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

$$S_n(0) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})} \quad S_n(W_b) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}$$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}}{\frac{D_E(\Delta p)_{E0}}{L_E} + \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch(\frac{W}{L_B})}{sh(\frac{W}{L_B})} - \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh(\frac{W}{L_B})}}$$

12.5 Quantitative analysis of BJT gain



$$J_c \approx 80 \times J_B \rightarrow \text{Gain } \beta = 80$$

$$S_n(0) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{ch\left(\frac{W}{L_p}\right)}{sh\left(\frac{W}{L_p}\right)}$$

$$S_n(W_b) = \frac{D_B(\Delta n)_{B0}}{L_B} \frac{1}{sh\left(\frac{W}{L_p}\right)}$$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} sh\left(\frac{W}{L_B}\right) + \frac{D_B(\Delta n)_{B0}}{L_B} [ch\left(\frac{W}{L_p}\right) - 1]}$$

12.5 Quantitative analysis of BJT gain

$$\operatorname{sh}\left(\frac{W}{L_B}\right) = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) - \exp\left(-\frac{W}{L_B}\right) \right]$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} \operatorname{sh}\left(\frac{W}{L_B}\right) + \frac{D_B(\Delta n)_{B0}}{L_B} \left[\operatorname{ch}\left(\frac{W}{L_p}\right) - 1 \right]}$$

12.5 Quantitative analysis of BJT gain

$$\operatorname{sh}\left(\frac{W}{L_B}\right) = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) - \exp\left(-\frac{W}{L_B}\right) \right] = \frac{W}{L_B}$$

$$\begin{aligned} \exp\left(\frac{W}{L_B}\right) &= 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots \\ \exp\left(-\frac{W}{L_B}\right) &= 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots \end{aligned} \quad \begin{array}{l} \nearrow \\ \nearrow \end{array} \quad \text{if } \frac{W}{L_p} < 1$$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} \operatorname{sh}\left(\frac{W}{L_B}\right) + \frac{D_B(\Delta n)_{B0}}{L_B} \left[\operatorname{ch}\left(\frac{W}{L_p}\right) - 1 \right]}$$

12.5 Quantitative analysis of BJT gain

$$\cosh\left(\frac{W}{L_B}\right) - 1 = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) + \exp\left(-\frac{W}{L_B}\right) \right] - 1$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} \frac{W}{L_B} + \frac{D_B(\Delta n)_{B0}}{L_B} \left[\cosh\left(\frac{W}{L_B}\right) - 1 \right]}$$

12.5 Quantitative analysis of BJT gain

$$\cosh\left(\frac{W}{L_B}\right) - 1 = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) + \exp\left(-\frac{W}{L_B}\right) \right] - 1 = \frac{1}{2} \left(\frac{W}{L_B}\right)^2$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

if $\frac{W}{L_B} < 1$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} \frac{W}{L_B} + \frac{D_B(\Delta n)_{B0}}{L_B} \left[\frac{1}{2} \left(\frac{W}{L_B}\right)^2 \right]}$$

12.5 Quantitative analysis of BJT gain

$$\cosh\left(\frac{W}{L_B}\right) - 1 = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) + \exp\left(-\frac{W}{L_B}\right) \right] - 1 = \frac{1}{2} \left(\frac{W}{L_B}\right)^2$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

if $\frac{W}{L_B} < 1$

$$\beta = \frac{\frac{D_B(\Delta n)_{B0}}{L_B}}{\frac{D_E(\Delta p)_{E0}}{L_E} \frac{W}{L_B} + \frac{D_B(\Delta n)_{B0}}{L_B} \left[\frac{1}{2} \left(\frac{W}{L_B}\right)^2 \right]} = \frac{1}{\frac{D_E(\Delta p)_{E0} W}{D_B(\Delta n)_{B0} L_E} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2}$$

12.5 Quantitative analysis of BJT gain

$$\cosh\left(\frac{W}{L_B}\right) - 1 = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) + \exp\left(-\frac{W}{L_B}\right) \right] - 1 = \frac{1}{2} \left(\frac{W}{L_B}\right)^2$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

if $\frac{W}{L_B} < 1$

$$\beta = \frac{1}{\frac{N_B D_E W}{N_E D_B L_E} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2}$$

12.5 Quantitative analysis of BJT gain

$$\cosh\left(\frac{W}{L_B}\right) - 1 = \frac{1}{2} \left[\exp\left(\frac{W}{L_B}\right) + \exp\left(-\frac{W}{L_B}\right) \right] - 1 = \frac{1}{2} \left(\frac{W}{L_B}\right)^2$$

$$\exp\left(\frac{W}{L_B}\right) = 1 + \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 + \dots$$

$$\exp\left(-\frac{W}{L_B}\right) = 1 - \frac{W}{L_B} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2 - \dots$$

if $\frac{W}{L_B} < 1$

$$\beta = \frac{1}{\frac{N_B D_E W}{N_E D_B L_E} + \frac{1}{2} \left(\frac{W}{L_B}\right)^2}$$

$$\beta = S_n(W_b) / (S_p + S_n(0) - S_n(W_b))$$