

# Mid2 RC\_part2

Zhiyu Zhou

UM-SJTU JI

July 10, 2024

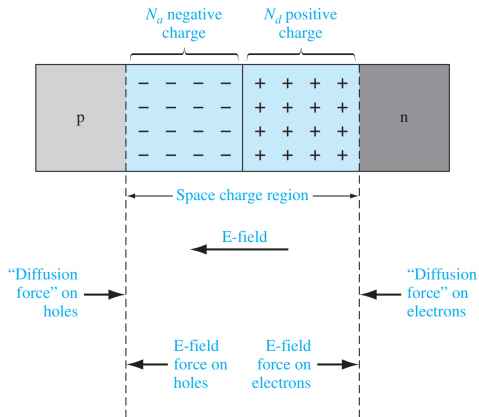
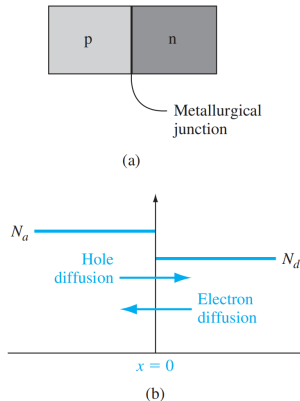
## 1 Chapter 7

- Basic Structure
- Zero Applied Bias
- Reverse Applied Bias

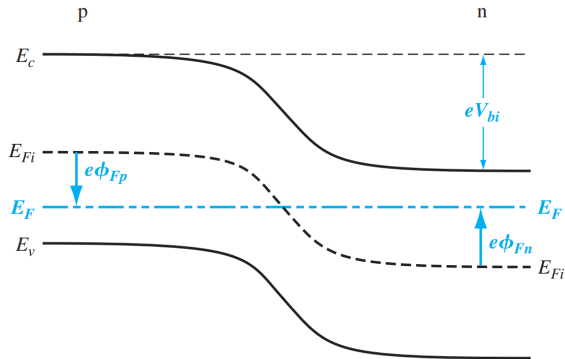
## 2 Chapter 8

- pn Junction Current
- Generation-Recombination Currents

# Basic Structure



# Zero Applied Bias



# Built-in Potential Barrier

$$\phi_{Fn} = -\frac{kT}{e} \ln\left(\frac{N_d}{n_i}\right)$$

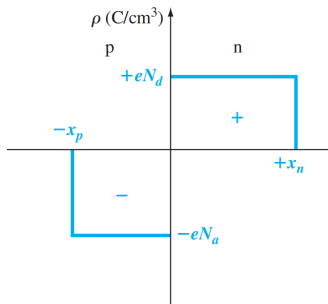
$$\phi_{Fp} = \frac{kT}{e} \ln\left(\frac{N_a}{n_i}\right)$$

$$\begin{aligned} V_{bi} &= |\phi_{Fn}| + |\phi_{Fp}| \\ &= \frac{kT}{e} \ln\left(\frac{N_a N_d}{n_i^2}\right) = V_t \ln\left(\frac{N_a N_d}{n_i^2}\right) \end{aligned}$$

$V_{bi}$ : built-in potential barrier;  $V_t = kT/e$ : thermal voltage

# Charge Density

- Poisson's equation:  $\frac{d^2\phi(x)}{dx^2} = \frac{-\rho(x)}{\epsilon_s} = -\frac{dE(x)}{dx}$ ,  $\epsilon_s = \epsilon_0\epsilon_r$



- volume charge density:

$$\rho(x) = \begin{cases} -eN_a, & -x_p \leq x \leq 0 \\ +eN_d, & 0 \leq x \leq x_n \end{cases}$$

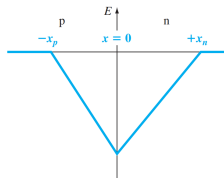
# Electric Field

- electric field:

$$E(x) = \begin{cases} -\frac{eN_a}{\epsilon_s}(x + x_p), & -x_p \leq x \leq 0 \\ -\frac{eN_d}{\epsilon_s}(x_n - x), & 0 \leq x \leq x_n \end{cases}$$

- continuity at  $x = 0$ :

$$N_a x_p = N_d x_n$$



- maximum electric field intensity at  $x = 0$ :

$$E_{\max} = -\frac{eN_a x_p}{\epsilon_s} = -\frac{eN_d x_n}{\epsilon_s}$$

# Electric Potential

- Set  $\phi(x = -x_p) = 0$ :

$$\phi(x) = \begin{cases} \frac{eN_a}{2\epsilon_s}(x + x_p)^2, & -x_p \leq x \leq 0 \\ \frac{eN_d}{\epsilon_s}(x_n \cdot x - \frac{x^2}{2}) + \frac{eN_d}{2\epsilon_s}x_p^2, & 0 \leq x \leq x_n \end{cases}$$

- when  $x = x_n$ , the potential is equal to the built-in potential barrier:

$$V_{bi} = |\phi(x = x_n)| = \frac{e}{2\epsilon_s}(N_dx_n^2 + N_ax_p^2)$$



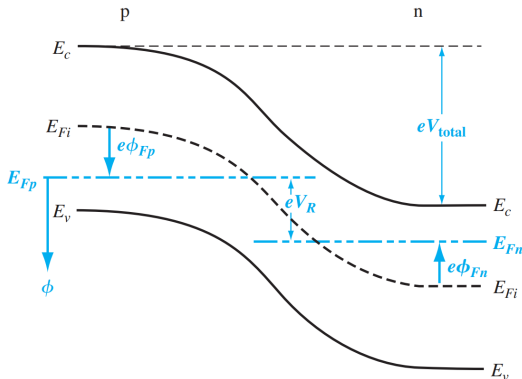
# Space Charge Width

$$x_n = \sqrt{\frac{2\epsilon_s V_{bi}}{e} \left(\frac{N_a}{N_d}\right) \left(\frac{1}{N_a + N_d}\right)}$$
$$x_p = \sqrt{\frac{2\epsilon_s V_{bi}}{e} \left(\frac{N_d}{N_a}\right) \left(\frac{1}{N_a + N_d}\right)}$$

- total depletion region width  $W$ :

$$\begin{aligned} W &= x_n + x_p \\ &= \sqrt{\frac{2\epsilon_s V_{bi}}{e} \left(\frac{N_a + N_d}{N_a N_d}\right)} \end{aligned}$$

# Reverse Applied Bias



- total potential barrier  $V_{total}$ :

$$V_{total} = |\phi_{Fn}| + |\phi_{Fp}| + V_R = V_{bi} + V_R$$

# Space Charge Width and Electric Field

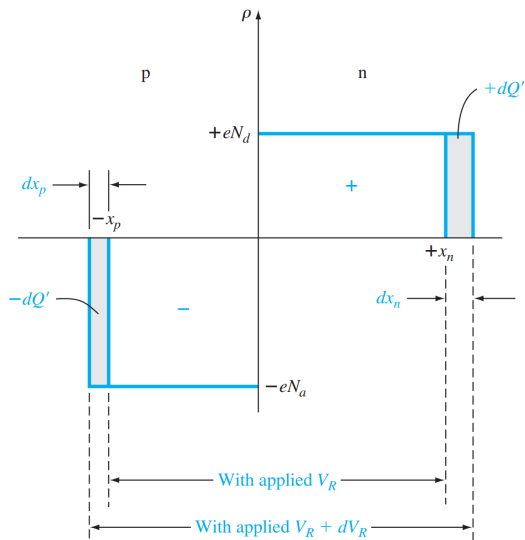
- In all of the previous equations, the built-in potential barrier can be replaced by the total potential barrier.
- total space charge width:

$$W = \sqrt{\frac{2\epsilon_s(V_{bi} + V_R)}{e} \left( \frac{N_a + N_d}{N_a N_d} \right)}$$

- maximum electric field:

$$\begin{aligned} E_{\max} &= -\sqrt{\frac{2e(V_{bi} + V_R)}{\epsilon_s} \left( \frac{N_a N_d}{N_a + N_d} \right)} \\ &= \frac{-2(V_{bi} + V_R)}{W} \end{aligned}$$

# Junction Capacitance



# Junction Capacitance

- separated  $+/-$  charges in depletion region: capacitance behavior
- The junction capacitance is defined as

$$C' = \frac{dQ'}{dV_R}$$

where

$$dQ' = eN_d dx_n = eN_a dx_p$$

$$C' = \sqrt{\frac{e\epsilon_s N_a N_d}{2(V_{bi} + V_R)(N_a + N_d)}} = \frac{\epsilon_s}{W}$$

- unit of  $C'$ :  $F/cm^2$

# One-Sided Junctions

- If, for example,  $N_a \gg N_d$ , this junction is referred to as a  $p^+n$  junction.
- total space charge width reduces to

$$W \approx \sqrt{\frac{2\epsilon_s(V_{bi} + V_R)}{eN_d}}$$

- Consider expressions for  $x_n$  and  $x_p$ , for  $p^+n$  junction:

$$x_p \ll x_n$$

$$W \approx x_n$$

- junction capacitance for  $p^+n$  junction reduces to:

$$C' \approx \sqrt{\frac{e\epsilon_s N_d}{2(V_{bi} + V_R)}}$$

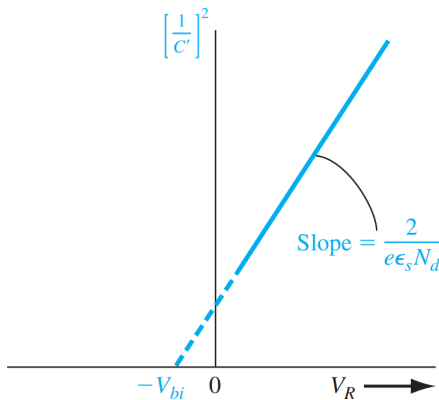
# One-Sided Junctions

- The depletion layer capacitance of a one-sided junction is a function of the doping concentration in the low-doped region:

$$\left(\frac{1}{C'}\right)^2 = \frac{2(V_{bi} + V_R)}{e\epsilon_s N_d}$$

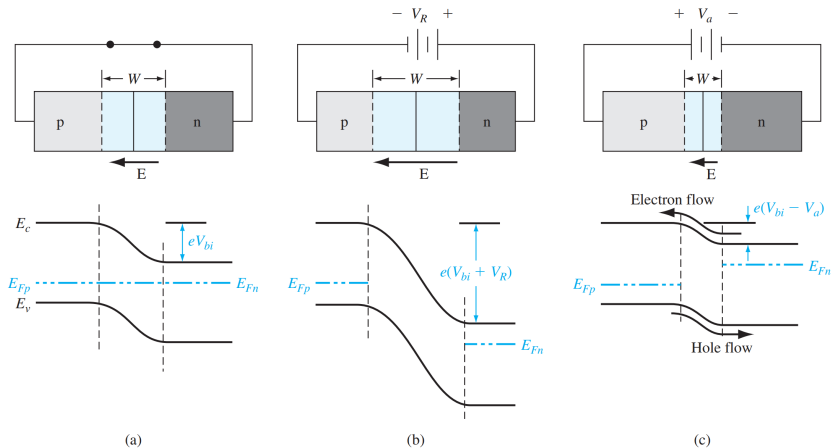
- the inverse capacitance squared is a linear function of applied reverse-biased voltage: an experimental way to determine doping concentration

# One-Sided Junctions





# pn Junction Current



**Figure 8.1** | A pn junction and its associated energy-band diagram for (a) zero bias, (b) reverse bias, and (c) forward bias.

# Ideal Current-Voltage Relationship

The ideal current–voltage relationship of a pn junction is derived on the basis of four assumptions. (The last assumption has three parts, but each part deals with current.) They are:

1. The abrupt depletion layer approximation applies. The space charge regions have abrupt boundaries, and the semiconductor is neutral outside of the depletion region.
2. The Maxwell–Boltzmann approximation applies to carrier statistics.
3. The concepts of low injection and complete ionization apply.
- 4a. The total current is a constant throughout the entire pn structure.
- 4b. The individual electron and hole currents are continuous functions through the pn structure.
- 4c. The individual electron and hole currents are constant throughout the depletion region.

**Table 8.1** | Commonly used terms and notation for this chapter

Term	Meaning
$N_a$	Acceptor concentration in the p region of the pn junction
$N_d$	Donor concentration in the n region of the pn junction
$n_{n0} = N_d$	Thermal-equilibrium majority carrier electron concentration in the n region
$p_{p0} = N_a$	Thermal-equilibrium majority carrier hole concentration in the p region
$n_{p0} = n_i^2/N_a$	Thermal-equilibrium minority carrier electron concentration in the p region
$p_{n0} = n_i^2/N_d$	Thermal-equilibrium minority carrier hole concentration in the n region
$n_p$	Total minority carrier electron concentration in the p region
$p_n$	Total minority carrier hole concentration in the n region
$n_p(-x_p)$	Minority carrier electron concentration in the p region at the space charge edge
$p_n(x_n)$	Minority carrier hole concentration in the n region at the space charge edge
$\delta n_p = n_p - n_{p0}$	Excess minority carrier electron concentration in the p region
$\delta p_n = p_n - p_{n0}$	Excess minority carrier hole concentration in the n region

# Boundary Conditions

- Assume complete ionization:

$$n_{n0} \approx N_d$$

- in the p region:

$$n_{p0} \approx \frac{n_i^2}{N_a}$$

$$n_{p0} = n_{n0} \exp\left(\frac{-eV_{bi}}{kT}\right)$$

This equation relates the minority carrier electron concentration on the p side of the junction to the majority carrier electron concentration on the n side of the junction in thermal equilibrium.

# Boundary Conditions

- When a forward bias  $V_a$  is applied to the pn junction, the boundary conditions for total minority carrier concentrations are:

$$p_n(x_n) = p_{n0} \exp\left(\frac{eV_a}{kT}\right)$$

$$n_p(-x_p) = n_{p0} \exp\left(\frac{eV_a}{kT}\right)$$

$$p_n(x \rightarrow +\infty) = p_{n0}$$

$$n_p(x \rightarrow -\infty) = n_{p0}$$

# Minority Carrier Distribution

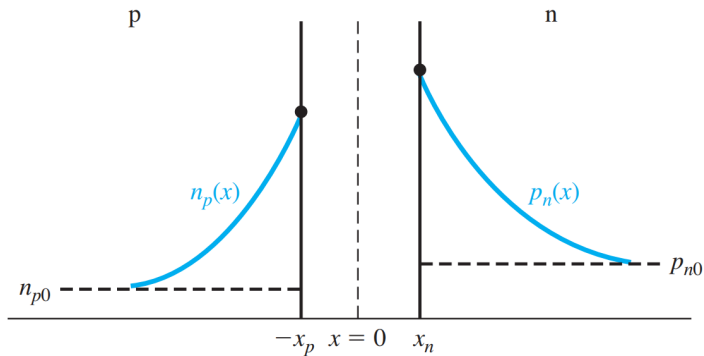
- Assume long pn junction.
- excess carrier concentrations for  $x \geq x_n$ :

$$\delta p_n(x) = p_n(x) - p_{n0} = p_{n0} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right] \exp\left(\frac{x_n - x}{L_p}\right)$$

- for  $x \leq -x_p$ :

$$\delta n_p(x) = n_p(x) - n_{p0} = n_{p0} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right] \exp\left(\frac{x_p + x}{L_n}\right)$$

# Minority Carrier Distribution



# Ideal pn Junction Current

- hole diffusion current density:

$$J_p(x_n) = \frac{eD_p p_{n0}}{L_p} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

- electron diffusion current density:

$$J_n(-x_p) = \frac{eD_n n_{p0}}{L_n} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

- for  $x \geq x_n$ :

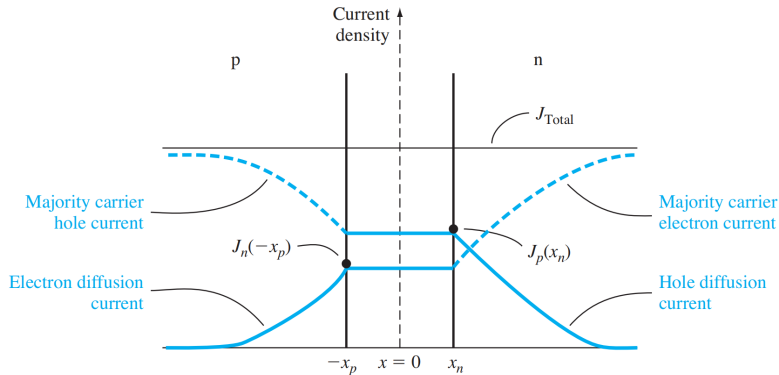
$$J_p(x) = \frac{eD_p p_{n0}}{L_p} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right] \exp\left(\frac{x_n - x}{L_p}\right)$$

- for  $x \leq -x_p$ :

$$J_n(x) = \frac{eD_n n_{p0}}{L_n} \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right] \exp\left(\frac{x_p + x}{L_n}\right)$$



# Ideal pn Junction Current



# Ideal pn Junction Current

- total current density:

$$J = J_p(x_n) + J_n(-x_p) = \left[ \frac{eD_p p_{n0}}{L_p} + \frac{eD_n n_{p0}}{L_n} \right] \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

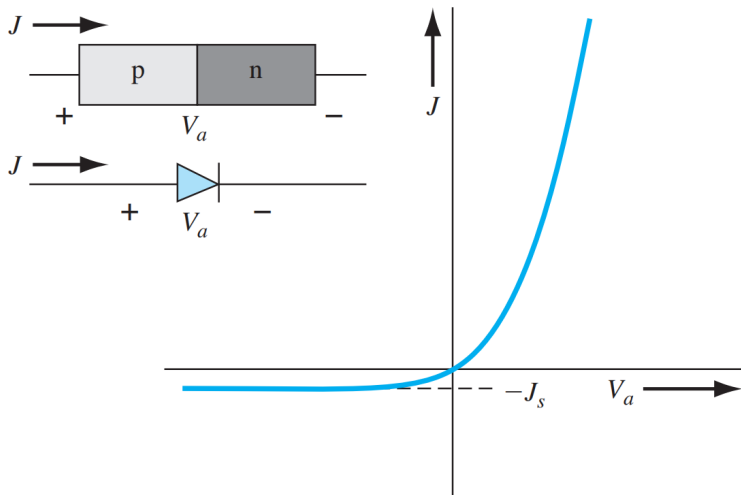
- define a parameter  $J_s$  as

$$J_s = \frac{eD_p p_{n0}}{L_p} + \frac{eD_n n_{p0}}{L_n}$$

- so that the total current density can be written as:

$$J = J_s \left[ \exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

# Ideal pn Junction Current



# Reverse-Biased Generation Current

- mobile electrons and holes are swept out of the space charge width:

$$n \approx p \approx 0$$

- recombination rate:

$$R = \frac{-C_n C_p N_t n_i^2}{C_n n' + C_p p'}$$

- to simplify calculation, assume  $E_t = E_i$ ,  $\tau_{n0} = \tau_{p0} = \tau_0$ :

$$R = \frac{-n_i}{2\tau_0} = -G$$

G: generation rate of electrons and holes in the depletion region

# Reverse-Biased Generation Current

- generation current density:

$$J_{\text{gen}} = \int_0^W eG dx$$

- If we assume that the generation rate is constant throughout the space charge region, then we obtain:

$$J_{\text{gen}} = \frac{en_i W}{2\tau_0}$$

- The total reverse-biased current density is the sum of the ideal reverse saturation current density and the generation current density:

$$J_R = J_s + J_{\text{gen}}$$

# Forward-Biased Recombination Current

- The recombination current density may be calculated from

$$J_{\text{rec}} = \int_0^W eR dx$$

- maximum recombination rate at the center of the space charge region:

$$R_{\text{max}} = \frac{n_i}{2\tau_0} \left[ \exp\left(\frac{eV_a}{2kT}\right) - 1 \right]$$

- recombination current density can be written as

$$J_{\text{rec}} = \frac{eWn_i}{2\tau_0} \left[ \exp\left(\frac{eV_a}{2kT}\right) - 1 \right]$$

# Forward-Biased Recombination Current

- total forward-bias current:

$$I = I_s \left[ \exp\left(\frac{eV_a}{nkT}\right) - 1 \right]$$

- n: ideality factor
- diffusion dominates:  $n \approx 1$
- recombination dominates:  $n \approx 2$

- ① Semiconductor Physics and Devices: Basic Principles 4th ed. Donald A. Neamen.
- ② 2023Summer Ve320Mid2\_RC\_part2, Jiajun Sun
- ③ 2022Summer RC\_Mid2\_part1, Xingyuan Wang
- ④ 2022Summer RC\_Mid2\_part2, Yucheng Huang