1

Fourier Series

Yash Rajput BM20BTECH11018

1

4

CONTENTS

1 Periodic Function	
---------------------	--

- **2** Fourier Series 1
- **3** Fourier Transform 2
- 4 Filter
- 5 Filter Design 5

Abstract—This manual provides a simple introduction to Fourier Series.

1 Periodic Function

Let

$$x(t) = A_0 |\sin(2\pi f_0 t)| \tag{1.1}$$

1.1 Plot x(t).

Solution: The Python code codes/1_1.py plots x(t) below.

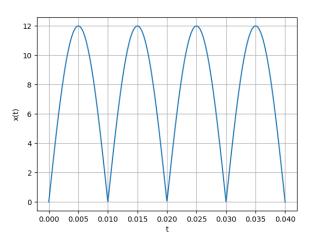


Fig. 1.1: *x*(*t*)

Fig. 1.1. x(t)

Solution: From Fig. (1.1), we see that x(t) is periodic. Further,

$$x\left(t + \frac{1}{2f_0}\right) = A_0 \left| \sin\left(2\pi f_0 \left(t + \frac{1}{2f_0}\right)\right) \right|$$
 (1.2)

$$= A_0 |\sin(2\pi f_0 t + \pi)| \tag{1.3}$$

$$= A_0 |\sin(2\pi f_0 t)| \tag{1.4}$$

Hence the period of x(t) is $\frac{1}{2f_0}$.

2 Fourier Series

Consider $A_0 = 12$ and $f_0 = 50$ for all numerical calculations.

2.1 If

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (2.1)

show that

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi k f_0 t} dt \qquad (2.2)$$

Solution: We have for some $n \in \mathbb{Z}$,

$$x(t)e^{-J2\pi nf_0t} = \sum_{k=-\infty}^{\infty} c_k e^{J2\pi(k-n)f_0t}$$
 (2.3)

But we know from the periodicity of $e^{j2\pi k f_0 t}$,

$$\int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{j2\pi k f_0 t} dt = \frac{1}{f_0} \delta(k)$$
 (2.4)

Thus,

$$\int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi nf_0t} dt = \frac{c_n}{f_0}$$
 (2.5)

$$\implies c_n = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t) e^{-j2\pi n f_0 t} dt \qquad (2.6)$$

1.2 Show that x(t) is periodic and find its period.

2.2 Find c_k for (1.1)

Solution: Using (2.2),

$$c_{n} = f_{0} \int_{-\frac{1}{2f_{0}}}^{\frac{1}{2f_{0}}} A_{0} |\sin(2\pi f_{0}t)| e^{-j2\pi n f_{0}t} dt \qquad (2.7)$$

$$= f_{0} \int_{-\frac{1}{2f_{0}}}^{\frac{1}{2f_{0}}} A_{0} |\sin(2\pi f_{0}t)| \cos(2\pi n f_{0}t) dt$$

$$+ \int_{-\frac{1}{2f_{0}}}^{\frac{1}{2f_{0}}} A_{0} |\sin(2\pi f_{0}t)| \sin(2\pi n f_{0}t) dt$$

$$+ \int_{-\frac{1}{2f_{0}}}^{\frac{1}{2f_{0}}} A_{0} |\sin(2\pi f_{0}t)| \sin(2\pi n f_{0}t) dt$$

$$= 2f_{0} \int_{0}^{\frac{1}{2f_{0}}} A_{0} \sin(2\pi f_{0}t) \cos(2\pi n f_{0}t) dt$$

$$= f_{0}A_{0} \int_{0}^{\frac{1}{2f_{0}}} (\sin(2\pi (n+1) f_{0}t)) dt \qquad (2.9)$$

$$= A_{0} \int_{0}^{\frac{1}{2f_{0}}} (\sin(2\pi (n-1) f_{0}t)) dt \qquad (2.10)$$

$$= A_{0} \frac{1 + (-1)^{n}}{2\pi} \left(\frac{1}{n+1} - \frac{1}{n-1}\right) \qquad (2.11)$$

$$= \begin{cases} \frac{2A_{0}}{\pi(1-n^{2})} & n \text{ even} \\ 0 & n \text{ odd} \end{cases}$$

2.3 Verify (2.1) using python.

Solution: The Python code codes/2_3.py verifies (2.13).

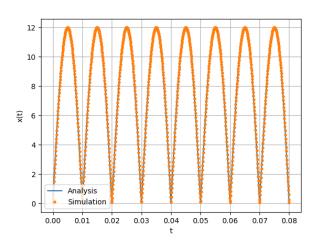


Fig. 2.3: Verification of (2.1).

2.4 Show that

$$x(t) = \sum_{k=0}^{\infty} (a_k \cos j 2\pi k f_0 t + b_k \sin j 2\pi k f_0 t)$$
(2.13)

and obtain the formulae for a_k and b_k . **Solution:** From (2.1),

$$x(t) = \sum_{k = -\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (2.14)

$$= c_0 + \sum_{k=1}^{\infty} c_k e^{j2\pi k f_0 t} + c_{-k} e^{-j2\pi k f_0 t}$$
 (2.15)

$$= c_0 + \sum_{k=1}^{\infty} (c_k + c_{-k}) \cos(2\pi k f_0 t)$$

$$+\sum_{k=0}^{\infty} (c_k - c_{-k}) \sin(2\pi k f_0 t)$$
 (2.16)

Hence, for $k \ge 0$,

$$a_k = \begin{cases} c_0 & k = 0 \\ c_k + c_{-k} & k > 0 \end{cases}$$
 (2.17)

$$b_k = c_k - c_{-k} (2.18)$$

2.5 Find a_k and b_k for (1.1)

Solution: From (2.1), we see that since x(t) is even,

$$x(-t) = \sum_{k=-\infty}^{\infty} c_k e^{-j2\pi k f_0 t}$$
 (2.19)

$$= \sum_{k=-\infty}^{\infty} c_{-k} e^{j2\pi k f_0 t}$$
 (2.20)

$$=\sum_{k=-\infty}^{\infty}c_ke^{\mathrm{J}^2\pi kf_0t}$$
 (2.21)

where we substitute k := -k in (2.20). Hence, we see that $c_k = c_{-k}$. So, from (2.17) and (2.18), for $k \ge 0$,

$$a_k = \begin{cases} \frac{2A_0}{\pi} & k = 0\\ \frac{4A_0}{\pi(1-k^2)} & k > 0, \ k \text{ even} \\ 0 & \text{otherwise} \end{cases}$$
 (2.22)

$$b_k = 0 (2.23)$$

2.6 Verify (2.13) using python.

Solution: The Python code codes/2_6.py verifies (2.13).

3 Fourier Transform

3.1

$$\delta(t) = 0, \quad t \neq 0 \tag{3.1}$$

$$\int_{-\infty}^{\infty} \delta(t) \, dt = 1 \tag{3.2}$$

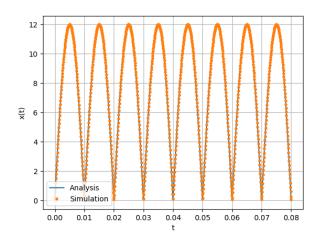


Fig. 2.6: Verification of (2.13).

3.2 The Fourier Transform of g(t) is

$$G(f) = \int_{-\infty}^{\infty} g(t)e^{-j2\pi ft} dt \qquad (3.3)$$

3.3 Show that

$$g(t-t_0) \stackrel{\mathcal{F}}{\longleftrightarrow} G(f)e^{-j2\pi ft_0}$$
 (3.4)

Solution: We write, substituting $u := t - t_0$,

$$g(t-t_0) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} g(t-t_0)e^{-j2\pi ft} dt$$
 (3.5)

$$= \int_{-\infty}^{\infty} g(u)e^{-j2\pi f(u+t_0)} du \qquad (3.6)$$

$$=G(f)e^{-j2\pi ft_0} \tag{3.7}$$

where the last equality follows from (3.3).

3.4 Show that

$$G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} g(-f)$$
 (3.8)

Solution: Using the definition of the Inverse Fourier Transform,

$$g(t) = \int_{-\infty}^{\infty} G(f)e^{j2\pi ft} df \qquad (3.9)$$

Hence, setting t := -f and f := t, which implies df = dt,

$$g(-f) = \int_{-\infty}^{\infty} G(t)e^{-j2\pi ft} dt \qquad (3.10)$$

$$\implies G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} g(-f)$$
 (3.11)

Solution: We have, from the definition of $\delta(t)$,

$$\delta(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} \delta(t) e^{-j2\pi f t} dt$$
 (3.12)

$$= \int_{-\infty}^{\infty} \delta(0) dt \tag{3.13}$$

$$= \int_{-\infty}^{\infty} \delta(t) dt = 1 \tag{3.14}$$

3.6 $e^{-j2\pi f_0 t} \stackrel{\mathcal{F}}{\longleftrightarrow} ?$

Solution: Suppose $g(t) \stackrel{\mathcal{F}}{\longleftrightarrow} G(f)$. Then,

$$g(t)e^{j2\pi f_0 t} \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} g(t)e^{-j2\pi (f-f_0)t} dt \qquad (3.15)$$

$$= F(f - f_0) (3.16)$$

Using (3.9) in (3.14), $1 \stackrel{\mathcal{F}}{\longleftrightarrow} \delta(-f)$. Hence, applying (3.16),

$$e^{-J^2\pi f_0 t} \stackrel{\mathcal{F}}{\longleftrightarrow} \delta(-(f+f_0)) = \delta(f+f_0)$$
 (3.17)

 $3.7 \cos(2\pi f_0 t) \stackrel{\mathcal{F}}{\longleftrightarrow} ?$

Solution: Using the linearity of the Fourier Transform and (??),

$$\cos(2\pi f_0 t) = \frac{1}{2} \left(e^{j2\pi f_0 t} + e^{-j2\pi f_0 t} \right)$$
 (3.18)

$$\stackrel{\mathcal{F}}{\longleftrightarrow} \frac{1}{2} \left(\delta \left(f + f_0 \right) + \delta \left(f - f_0 \right) \right) \tag{3.19}$$

3.8 Find the Fourier Transform of x(t) and plot it. Verify using python.

Solution: Substituting (2.12) in (2.1),

$$x(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \sum_{k=-\infty}^{\infty} c_k \delta(f + kf_0)$$
 (3.20)

$$= \frac{2A_0}{\pi} \sum_{k=-\infty}^{\infty} \frac{\delta(f + 2kf_0)}{1 - 4k^2}$$
 (3.21)

The python code codes/3_8.py verifies (3.21).

3.9 Show that

$$\operatorname{rect} t \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{sinc} t \tag{3.22}$$

Verify using python.

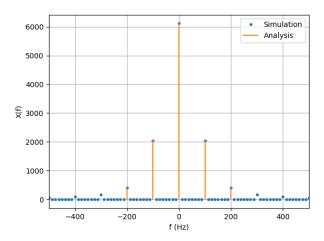


Fig. 3.8: Fourier Transform of x(t).

Solution: We write

$$\operatorname{rect} t \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} \operatorname{rect} t e^{-j2\pi f t} dt \tag{3.23}$$

$$= \int_{-\frac{1}{2}}^{\frac{1}{2}} e^{-j2\pi ft} dt \tag{3.24}$$

$$= \frac{e^{j\pi f} - e^{-j\pi f}}{j2\pi f} = \frac{\sin \pi f}{\pi f} = \operatorname{sinc} f \quad (3.25)$$

The python code codes/3_9.py verifies (3.25).

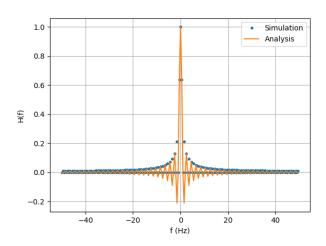


Fig. 3.9: Fourier Transform of rect(t).

3.10 sinc $t \stackrel{\mathcal{F}}{\longleftrightarrow}$? Verify using python. **Solution:** From (3.9), we have

$$\operatorname{sinc} t \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{rect}(-f) = \operatorname{rect} f \tag{3.26}$$

Since rect f is an even function. The python code codes/3 10.py verifies (3.26).

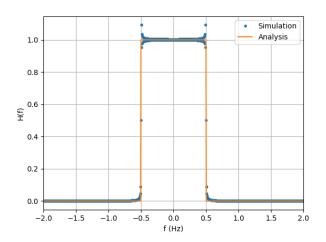


Fig. 3.10: Fourier Transform of sinc(t).

4 FILTER

4.1 Find H(f) which transforms x(t) to DC 5V.

Solution: The function H(f) is a low pass filter which filters out even harmonics and leaves the zero frequency component behind. The rectangular function represents an ideal low pass filter. Suppose the cutoff frequency is $f_c = 50$ Hz, then

$$H(f) = \operatorname{rect} \frac{f}{2f_c} = \begin{cases} 1 & |f| < f_c \\ 0 & \text{otherwise} \end{cases}$$
 (4.1)

Multiplying by a scaling factor to get DC 5V,

$$H(f) = \frac{\pi V_0}{2A_0} \operatorname{rect}\left(\frac{f}{2f_c}\right) \tag{4.2}$$

where $V_0 = 5$ V.

4.2 Find h(t).

Solution: Suppose $g(t) \stackrel{\mathcal{F}}{\longleftrightarrow} G(f)$. Then, for some nonzero $a \in \mathbb{R}$

$$g(at) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} g(at)e^{-J2\pi ft} dt$$
 (4.3)

$$= \frac{1}{a} \int_{-\infty}^{\infty} g(u)e^{\left(-j2\pi\frac{f}{a}t\right)} dt \tag{4.4}$$

$$=\frac{1}{a}G\left(\frac{f}{a}\right) \tag{4.5}$$

where we have substituted u := at. Using (4.5) of the Fourier Transform in (4.1),

$$h(t) = 2 * f_c \operatorname{sinc}(2 * f_c t)$$
 (4.6)

4.3 Verify your result using convolution.

Solution: The Python code codes/4_3.py verifies the result by plotting the graph below.

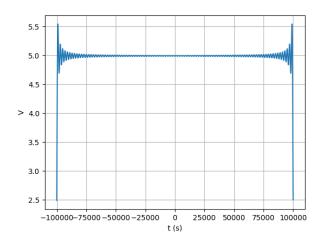


Fig. 4.3: Convolution of the two signals.

5 FILTER DESIGN

5.1 Design a Butterworth filter for H(f).

Solution: The Butterworth filter has an amplitude response given by

$$|H(f)|^2 = \frac{1}{\left(1 + \left(\frac{f}{f_c}\right)^{2n}\right)}$$
 (5.1)

where n is the order of the filter and f_c is the cutoff frequency. The loss at frequency f is given by

Loss =
$$-10\log_{10}|H(f)|^2$$
 (5.2)

$$= -20\log_{10}|H(f)| \tag{5.3}$$

Assuming the cutoff frequency $f_c = 10$ Hz, the Python code codes/5_1.py plots the magnitude response for different values of n as shown below. At $f = 10f_c = 2f_0$, we obtain a 40 dB loss for n = 2, which is ideal. Hence, we choose n = 2. Further, note that the DC gain is $\frac{5}{12}$.

5.2 Design a Chebyshev filter for H(f).

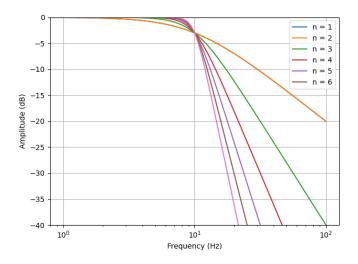


Fig. 5.1: Butterworth response for various n.

Solution: The Chebyshev filter has an amplitude response given by

$$|H(f)|^2 = \frac{1}{\left(1 + \epsilon^2 C_n^2 \left(\frac{f}{f_c}\right)^2\right)}$$
 (5.4)

where n is the order of the filter, ϵ is the ripple, f_c is the cutoff frequency and C_n denotes the nth order Chebyshev polynomial. Assuming the cutoff frequency to be at f_0 , the Python codes codes/5_2.py and codes/5_3.py plot the magnitude response for different values of n.

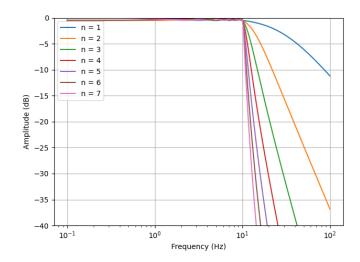


Fig. 5.2: Chebyshev response for various n.

At $f = 10f_c$, we obtain close to 40 dB loss for n = 2, which is ideal. Hence, we choose n = 2. Further, note that the DC gain is $\frac{5}{12}$.

5.3 Design a circuit for your Butterworth filter.

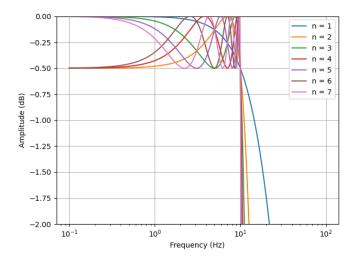


Fig. 5.2: Chebyshev response for various *n*, in detail.

Solution: Looking at the table of normalized *g* values of the Butterworth filter for order 2, we see that

$$g_1 = C_1 = 1.5774$$
 $g_2 = L_1 = 0.4226$ (5.5)

De-normalizing these values,

$$C_1' = \frac{C_1}{\omega_c} = 2.51 \times 10^{-2} \text{ F}$$
 (5.6)

$$L_1' = \frac{L_1}{\omega_c} = 6.72 \text{ mH}$$
 (5.7)

The L-C network is shown below

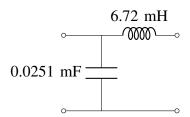


Fig. 5.3: L-C Butterworth Filter

5.4 Design a circuit for your Chebyshev filter.

Solution: Looking at the table of normalized *g* values of the Chebyshev filter for order 2 and 0.5 dB ripple, we see that

$$g_1 = C_1 = 1.4029$$
 $g_2 = L_1 = 0.7071$ (5.8)

De-normalizing these values,

$$C_1' = \frac{C_1}{\omega_c} = 2.23 \times 10^{-2} \text{ F}$$
 (5.9)

$$L_1' = \frac{L_1}{\omega_c} = 1.12 \times 10^{-2} \text{ H}$$
 (5.10)

The L-C network is shown below

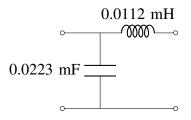


Fig. 5.4: L-C Chebyshev Filter