

Signals and Systems Internal Assessment (4)

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IIIrd semester - VLSI Design & Technology

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yes

Q1] Apply Laplace T,

$$Y(s) + 2s Y(s) = 2X(s) + 3s X(s) + s^2 X(s)$$

$$Y(s) (1+2s) = X(s) (2+3s+s^2)$$

$$H(s) = \frac{Y(s)}{X(s)} = \frac{2+3s+s^2}{1+2s}$$

* now, for causality / non-causality :

$$\text{numerator} = 1+2s \rightarrow \text{degree} = 1$$

$$\text{denominator} = s^2 + 3s + 2 \rightarrow \text{degree} = 2$$

Degree of Denominator > Degree of Numerator

\therefore Inverse system is Causal

* for stability / non-stability,
we find roots of denominator

$$s^2 + 3s + 2$$

$$\text{roots: } s = -1$$

$$s = -2$$

Since roots are negative and real, it is
STABLE system.

MATLAB R2024a - academic use

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Clean Data

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Clear Workspace

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>> %Define the numerator and denominator coefficients of H(s)
numerator = [1, 3, 2];
denominator = [2, 1];
% Find zeros by solving numerator = 0
zeros = roots (numerator);
% Find poles by solving denominator = 0 roots (denominator);
poles = roots (denominator);
% Display the results
disp('Poles of H(s):');
disp (poles);
disp('Zeros of H(s):');
disp (zeros);
% Details: Yashvardhan Singh (230959136) VLSI-46 Signals and Systems IA4
Poles of H(s):
-0.5000

Zeros of H(s):
-2
-1
|
fx >>

Workspace

Name	Value
denominator	[2,1]
numerator	[1,3,2]
poles	-0.5000
zeros	[-2,-1]

Details

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Q2]

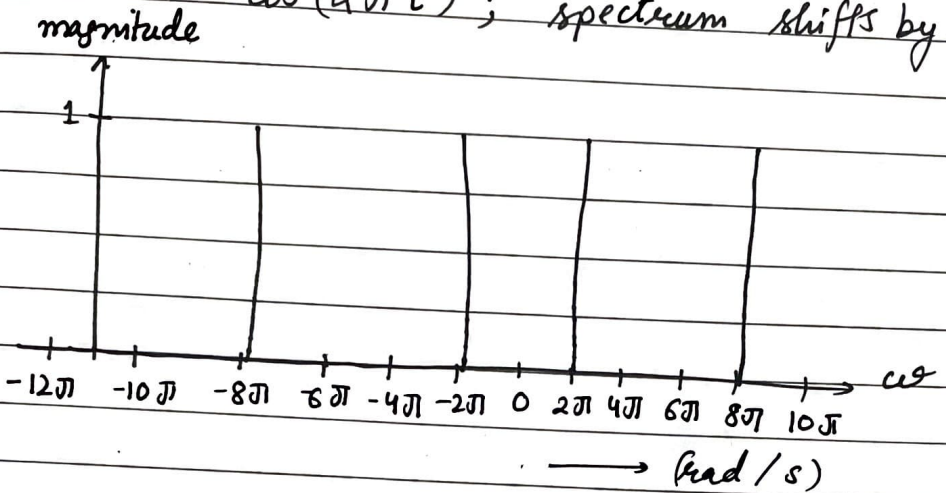
$\cos \omega_0 t$



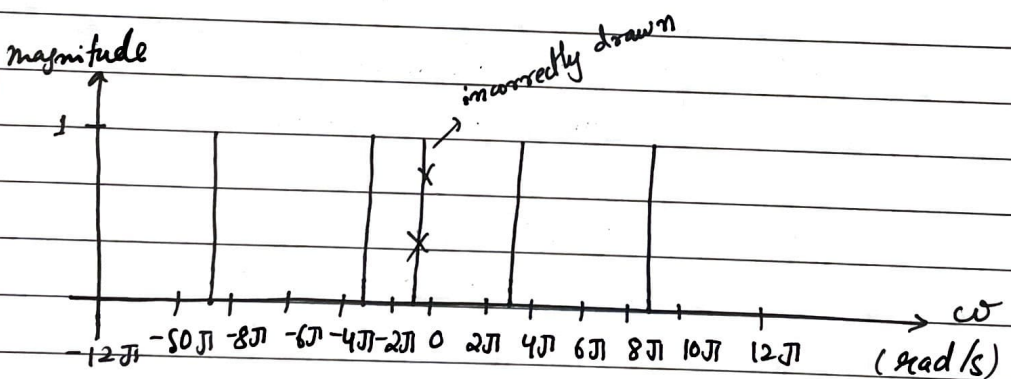
$x p(t)$

$$p(t) = \sum_{n=-\infty}^{\infty} s(t - nT) ; T = 1/3$$

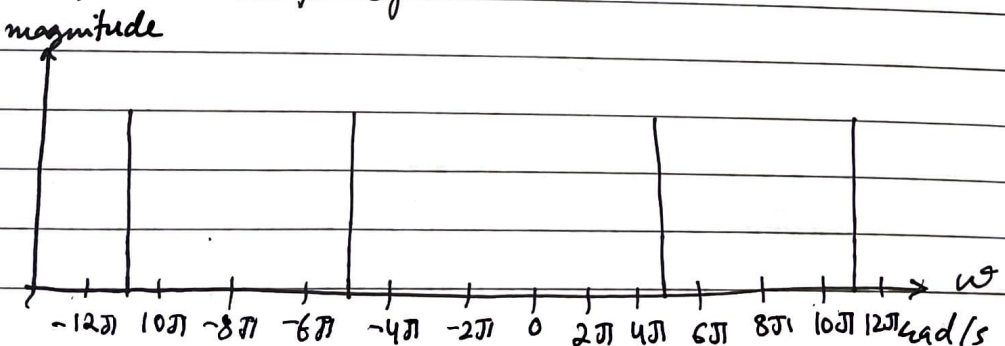
- ① case one: base band signal is modulated by $\cos(2\pi t)$; spectrum shifts by $\pm 2\pi$



- ② case 2: modulation with $\omega_0 = 3\pi$, causing spectrum to shift by $\pm 3\pi$



- ③ case 3: modulation with $\omega_0 = 5\pi$, causing spectrum shift by $\pm 5\pi$



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ys

q2-7
continued

signal can be reconstructed by ideal reconstruction by using formula:

$$X_s(j\omega) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(j\omega - jk\omega_s)$$

Here, the sampling is taken equal to or greater than nyquist's rate:

$$\omega_s \geq 2\omega_0$$

q3] ROC of a Z-transform is range of values in complex plane for which Z-transform of a discrete time signal converges.

PROPERTIES:

- 1) ROC does not have any poles
- 2) If signal is right-sided (causal), ROC is outside outermost pole.
- 3) If signal is left-sided (anti-causal), ROC is inside innermost pole.
- 4) ROC is single and continuous region.

$$a[n] = [n(-0.25)^n u(n)] * [(0.5)^{-n} u(-n)]$$

\therefore Z-transform of $n a^n u(n) = \frac{z a}{(z-a)^2}$
at $a = -0.25$,

$$X_1(z) = \frac{z(-0.25)}{(z+0.25)^2} \Rightarrow \text{ROC: } |z| > 0.25$$

\therefore Z-transform of $a^{-n} u(-n) = \frac{z}{z-a}$

$$X_2(z) = \frac{z}{z-0.5} \Rightarrow \text{ROC: } |z| < 0.5$$

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q3. continued :

convolution in time domain
is multiplication in Z domain

$$\frac{-0.25z}{(z+0.25)^2} \times \frac{z}{(z-0.5)}$$

∴ ROC : $0.25 < |z| < 0.5$

- i) poles : at $z = -0.25$; $z = 0.5$
- ii) zeroes : at $z = 0$
- iii) $0.25 < |z| < 0.5$

