

EE3025 Assignment-1

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Download all python codes from

[https://github.com/yaswanthNaidu99/
ASSIGNMENT1/tree/main/codes](https://github.com/yaswanthNaidu99/ASSIGNMENT1/tree/main/codes)

and latex-tikz codes from

[https://github.com/yaswanthNaidu99/
ASSIGNMENT1/blob/main/
EE18BTECH11024.tex](https://github.com/yaswanthNaidu99/ASSIGNMENT1/blob/main/EE18BTECH11024.tex)

$$h(n) = \frac{-1^n}{2} u(n) + \left(\frac{-1}{2}\right)^{n-2} u(n-2) \quad (2.1.5)$$

2.2. Computing y (for N samples) using FFT and IFFT :

$$X = FFT(x) \quad (2.2.1)$$

$$H = FFT(h) \quad (2.2.2)$$

$$Y = X.H \quad (2.2.3)$$

$$y = IFFT(Y) \quad (2.2.4)$$

1 PROBLEM

1.1. Let

$$x(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4, 2, 1 \right\} \quad (1.1.1)$$

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2) \quad (1.1.2)$$

1.2. Compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (1.2.1)$$

and $H(k)$ using $h(n)$.

1.3. Compute $X(k)$, $H(k)$ and $y(n)$ using FFT and IFFT methods.

2 SOLUTION

2.1. Computing $h(n)$ using Z-transform of $y(n)$ as follows :

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z) \quad (2.1.1)$$

$$\Rightarrow Y(z) = \frac{2(z^2 + 1)}{z(2z + 1)}X(z) \quad (2.1.2)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (2.1.3)$$

applying inverse Z-transform to compute $h(n)$

$$h(n) = Z^{-1} \left(\frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}} \right) \quad (2.1.4)$$

2.3. If desired output is real :

$$y = IFFT(Y) = \frac{1}{N} * FFT(Y^*) \quad (2.3.1)$$

where Y^* = complex conjugate(Y)

Thus IFFT can be implemented using the FFT function itself, which can save memory in a hardware setup.

2.4. Implementation and results :

y is computed through above steps (by padding x so that $N = 8$), while a recursive FFT algorithm is implemented to compute the 8 point FFT.

The code for it is as follows :

[https://github.com/yaswanthNaidu99/
ASSIGNMENT1/blob/main/codes/fft.py](https://github.com/yaswanthNaidu99/ASSIGNMENT1/blob/main/codes/fft.py)

$$\bar{x} = \begin{bmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \\ 0 \\ 0 \end{bmatrix} \quad \bar{h} = \begin{bmatrix} 1. \\ -0.5 \\ 1.25 \\ -0.625 \\ 0.3125 \\ -0.15625 \\ 0.0625 \\ -0.03125 \end{bmatrix} \quad (2.4.1)$$

$$\bar{X} = \begin{bmatrix} 13 \\ -3.121 - 6.536j \\ 1.j \\ 1.121 - 0.536j \\ -1. \\ 1.121 + 0.536j \\ -1.j \\ -3.121 + 6.536j \end{bmatrix} \quad \bar{H} = \begin{bmatrix} 1.312 + 0.j \\ 0.864 - 0.525j \\ 0. \\ 0.511 + 1.85j \\ 3.938 \\ 0.511 - 1.85j \\ 0. \\ 0.864 + 0.525j \end{bmatrix} \quad (2.4.2)$$

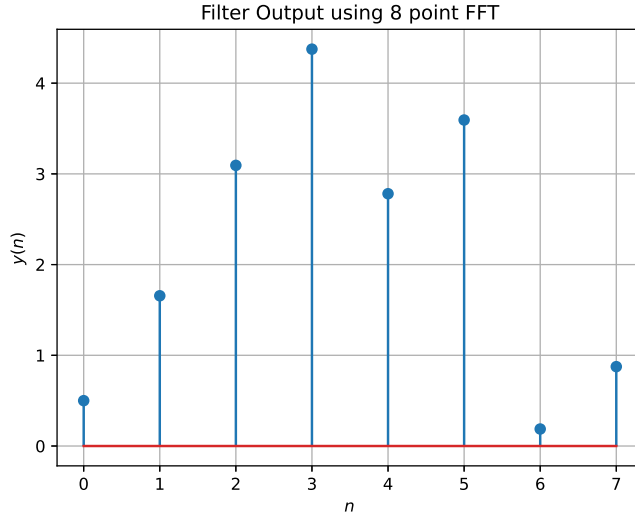


Fig. 2.4: $y(n)$ obtained using 8 point recursive FFT

2.5. Formulating a recursive N-point FFT Algorithm

($N = 2^\gamma$; γ is an integer):

An N-point DFT can be written as :

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (2.5.1)$$

$$(2.5.2)$$

By dividing the inputs into even and odd indices ; where $W_N = e^{-\frac{j2\pi}{N}}$

$$\begin{aligned} X_k &= \sum_{m=0}^{N/2-1} x_{2m} e^{-\frac{j2\pi k 2m}{N}} + \sum_{m=0}^{N/2-1} x_{2m+1} e^{-\frac{j2\pi k(2m+1)}{N}} \\ &= \underbrace{\sum_{m=0}^{N/2-1} x_{2m} e^{-\frac{j2\pi k m}{N/2}}}_{N/2 \text{ DFT with even inputs}} + W_N^k \underbrace{\sum_{m=0}^{N/2-1} x_{2m+1} e^{-\frac{j2\pi k m}{N/2}}}_{N/2 \text{ DFT with odd inputs}} \end{aligned} \quad (2.5.3)$$

While exploiting symmetry of W_N as :

$$W_N^{k+N/2} = -W_N^k \quad (2.5.4)$$

We can transform the iterative problem to a Divide-Conquer algorithm, where :

$$X_{0 \rightarrow \frac{N}{2}-1} = X_{\text{even}} + \bar{W}_{N/2} * X_{\text{odd}} \quad (2.5.5)$$

$$X_{\frac{N}{2} \rightarrow N-1} = X_{\text{even}} - \bar{W}_{N/2} X_{\text{odd}} \quad (2.5.6)$$

$$\bar{W}_{N/2}(i) = W_N^i$$

; for $i = 0, 1, 2, \dots, (N/2) - 1$

Where X_{even} and X_{odd} are again recursively computed using $(N/2)$ -FFT thus halving its computation time; until $N = 2$ (the base case of the recursion) for which a 2-point DFT is computed as follows.

$$X = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} * x \quad (2.5.7)$$

Thus , the time complexity of the algorithm is $O(N \log_2 N)$

2.6. Vector representation of the FFT algorithm

An 8-point DFT can be represented as a Matrix product as follows:

$$\bar{X} = \bar{W} \bar{x}$$

$$\bar{x} = \begin{bmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \\ x(4) \\ x(5) \\ x(6) \\ x(7) \end{bmatrix} \quad \bar{X} = \begin{bmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \\ X(4) \\ X(5) \\ X(6) \\ X(7) \end{bmatrix} \quad (2.6.1)$$

$$\bar{W} = \begin{bmatrix} W^0 & W^0 & W^0 & W^0 & W^0 & W^0 & W^0 & W^0 \\ W^0 & W^1 & W^2 & W^3 & W^4 & W^5 & W^6 & W^7 \\ W^0 & W^2 & W^4 & W^6 & W^0 & W^2 & W^4 & W^6 \\ W^0 & W^3 & W^6 & W^1 & W^4 & W^7 & W^2 & W^5 \\ W^0 & W^4 & W^0 & W^4 & W^0 & W^4 & W^0 & W^4 \\ W^0 & W^5 & W^2 & W^7 & W^4 & W^1 & W^6 & W^3 \\ W^0 & W^6 & W^4 & W^2 & W^0 & W^6 & W^4 & W^2 \\ W^0 & W^7 & W^6 & W^5 & W^4 & W^3 & W^2 & W^1 \end{bmatrix} \quad (2.6.2)$$

where $W = W_8 = e^{-j2\pi/8}$

The FFT algorithm exploits the inherent symmetry in \bar{W} matrix by permuting \bar{x} in a bit-reversed fashion.

A 8 point FFT can be represented as :

$$\bar{X} = \bar{W}_p \bar{x}_p$$

$$\bar{x}_p = P \bar{x}$$

The P matrix rearranges the input x vector in a bit-reversed fashion as in :

$$x_p(i) = x(\text{bit reverse}(i)) : \quad (2.6.3)$$

For Eg ;

$$x_p(4) = x_p(\text{bin}(100)) = x(\text{bin}(001)) = x(1) \quad (2.6.4)$$

$$P = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix} \quad (2.6.5)$$

For such a rearrangement of \bar{x} , we can exploit the symmetry in \bar{W}_p , and thus factorise it into

3 sparse matrices.

$$\bar{W}_p = \bar{W}_3 \bar{W}_2 \bar{W}_1 \quad (2.6.6)$$

$$\bar{W}_3 = \begin{bmatrix} 1 & . & . & . & W^0 & . & . & . \\ . & 1 & . & . & . & W^1 & . & . \\ . & . & 1 & . & . & . & W^1 & . \\ . & . & . & 1 & . & . & . & W^1 \\ 1 & . & . & . & -W^0 & . & . & . \\ . & 1 & . & . & . & -W^1 & . & . \\ . & . & 1 & . & . & . & -W^1 & . \\ . & . & . & 1 & . & . & . & -W^1 \end{bmatrix} \quad (2.6.7)$$

$$\bar{W}_2 = \begin{bmatrix} 1 & . & W^0 & . & . & . & . & . \\ . & 1 & . & W^2 & . & . & . & . \\ 1 & . & -W^0 & . & . & . & . & . \\ . & 1 & . & -W^2 & . & . & . & . \\ . & . & . & . & 1 & . & W^0 & . \\ . & . & . & . & . & 1 & . & W^2 \\ . & . & . & . & . & 1 & -W^0 & . \\ . & . & . & . & . & . & 1 & -W^2 \end{bmatrix} \quad (2.6.8)$$

$$\bar{W}_1 = \begin{bmatrix} 1 & W^0 & . & . & . & . & . & . \\ 1 & -W^0 & . & . & . & . & . & . \\ . & . & 1 & W^0 & . & . & . & . \\ . & . & 1 & -W^0 & . & . & . & . \\ . & . & . & . & 1 & W^0 & . & . \\ . & . & . & . & 1 & -W^0 & . & . \\ . & . & . & . & . & . & 1 & W^0 \\ . & . & . & . & . & . & 1 & -W^0 \end{bmatrix} \quad (2.6.9)$$

where (.) refers to zero.

Similarly a N-point DFT's W matrix can be factorised into γ sparse matrices, ($N = 2^\gamma$), with each row containing a 1 and a complex no. These γ sparse matrices represent the γ -stages in the butterfly diagram of an N-point FFT.

2.7. Run time analysis of FFT

Considering no. of multiplications as a metric for time complexity:

1. In N-point DFT, the dense matrix multiplication consist of $2N^2$ real multiplications. Hence time complexity of DFT is $O(N^2)$

2. While in FFT, there are $\log N$ (stages) sparse

matrices, each stage requires $4*(N/2)$ real unique multiplications.

Thus, the total multiplications for N-FFT is $2*N*\log N$ which implies a time complexity of $O(N\log N)$

The below code compares time-complexities of DFT and FFT :

https://github.com/yaswanthNaidu99/ASSIGNMENT1/blob/main/codes/dft_fft.py

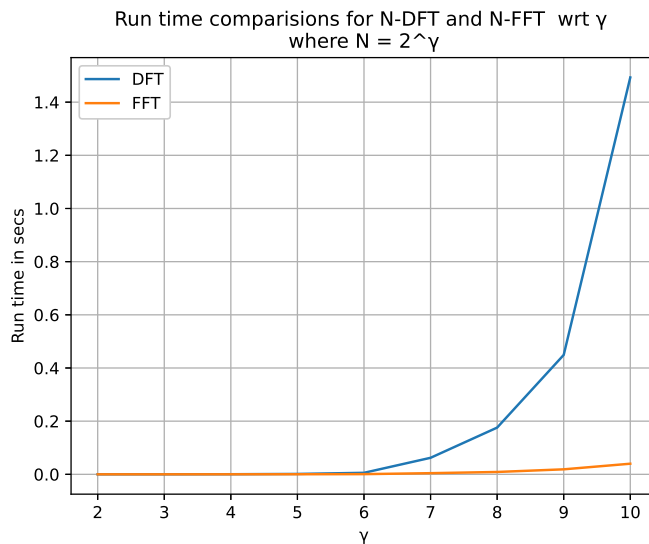


Fig. 2.7: Time complexity comparison

2.8. Convolution vs FFT

A convolution takes N^2 operations $\approx O(N^2)$.

While the same output can be achieved using FFT and IFFT within :

$$\underbrace{O(N\log N)}_{\substack{x \rightarrow X \\ h \rightarrow H}} + \underbrace{O(N)}_{Y = X * H} + \underbrace{O(N\log N)}_{Y \rightarrow y} \approx O(N\log N) \quad (2.8.1)$$