

# VE320

## Intro to Semiconductor Devices

### HOMEWORK 9

December 4, 2020

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1. (a) Since  $V_{SG} > -V_T$  and  $V_{SD} < V_{SG} + V_T$ , the drain current  $I_D$  for  $V_{SG} = 0.8$  V,  $V_{SD} = 0.25$  V is

$$\begin{aligned} I_D &= \frac{k'_p}{2} \cdot \frac{W}{L} \cdot [2(V_{SG} + V_T)V_{SD} - V_{SD}^2] \\ &= \frac{1.0 \times 10^{-4}}{2} \times 15 \times [2 \times (0.8 - 0.4) \times 0.25 - 0.25^2] \\ &= 1.031 \times 10^{-4} \text{ A.} \end{aligned}$$

- (b) Since  $V_{SD} > V_{SG} + V_T$ , the drain current  $I_D$  for  $V_{SG} = 0.8$  V,  $V_{SD} = 1.0$  V is

$$I_D = \frac{k'_p}{2} \cdot \frac{W}{L} \cdot (V_{SG} + V_T)^2 = \frac{1.0 \times 10^{-4}}{2} \times 15 \times (0.8 - 0.4)^2 = 1.2 \times 10^{-4} \text{ A.}$$

- (c) Since  $V_{SD} > V_{SG} + V_T$ , the drain current  $I_D$  for  $V_{SG} = 1.2$  V,  $V_{SD} = 1.0$  V is

$$I_D = \frac{k'_p}{2} \cdot \frac{W}{L} \cdot (V_{SG} + V_T)^2 = \frac{1.0 \times 10^{-4}}{2} \times 15 \times (1.2 - 0.4)^2 = 4.8 \times 10^{-4} \text{ A.}$$

- (d) Since  $V_{SD} > V_{SG} + V_T$ , the drain current  $I_D$  for  $V_{SG} = 1.2$  V,  $V_{SD} = 2.0$  V is

$$I_D = \frac{k'_p}{2} \cdot \frac{W}{L} \cdot (V_{SG} + V_T)^2 = \frac{1.0 \times 10^{-4}}{2} \times 15 \times (1.2 - 0.4)^2 = 4.8 \times 10^{-4} \text{ A.}$$

2. (a) Assume the transistor operates in the saturation region, then the drain current is

$$\begin{aligned} I_D &= \frac{k'_p}{2} \cdot \frac{W}{L} \cdot (V_{SG} + V_T)^2 \\ 10^{-4} &= \frac{1.2 \times 10^{-4}}{2} \times 20 \times (0 + V_T)^2 \end{aligned}$$

Solving the equation, the  $V_T$  value is

$$V_T = 0.2887 \text{ V.}$$

Then we can check that

$$V_{SD} = 1 \text{ V} > V_{SG} + V_T = 0.2887 \text{ V}$$

so the transistor does operate in the saturation region.

- (b) Since  $V_{SD} > V_{SG} + V_T$ , the drain current  $I_D$  for  $V_{SG} = 0.4$  V,  $V_{SB} = 0$  V, and  $V_{SD} = 1.5$  V is

$$I_D = \frac{k'_p}{2} \cdot \frac{W}{L} \cdot (V_{SG} + V_T)^2 = \frac{1.2 \times 10^{-4}}{2} \times 20 \times (0.4 + 0.2887)^2 = 5.69 \times 10^{-4} \text{ A.}$$

(c) Since  $V_{SG} > -V_T$  and  $V_{SD} < V_{SG} + V_T$ , the value of  $I_D$  for  $V_{SG} = 0.6$  V,  $V_{SB} = 0$  V, and  $V_{SD} = 0.15$  V is

$$\begin{aligned} I_D &= \frac{k'_p}{2} \cdot \frac{W}{L} \cdot [2(V_{SG} + V_T)V_{SD} - V_{SD}^2] \\ &= \frac{1.2 \times 10^{-4}}{2} \times 20 \times [2 \times (0.6 + 0.2887) \times 0.15 - 0.15^2] \\ &= 2.929 \times 10^{-4} \text{ A.} \end{aligned}$$

3. (a) The difference between  $E_{Fi}$  and  $E_F$  is

$$\phi_{fp} = \frac{kT}{e} \ln \left( \frac{N_a}{n_i} \right) = \frac{300k}{e} \ln \left( \frac{10^{15}}{1.5 \times 10^{10}} \right) = 0.287 \text{ V.}$$

The threshold voltage is

$$\begin{aligned} V_{TN} &= \frac{Q'_{SD}(\text{max})}{C_{ox}} + V_{FB} + 2\phi_{fp} \\ &= ((Q'_{SD}(\text{max}) - Q'_{ss}) \left( \frac{t_{ox}}{\epsilon_{ox}} \right) + \phi_{ms} + 2\phi_{fp} \\ &= ((eN_a x_{dT} - Q'_{ss}) \left( \frac{t_{ox}}{\epsilon_{ox}} \right) + \phi_{ms} + 2\phi_{fp} \\ &= \left( (\sqrt{4eN_a \epsilon_s \phi_{fp}} - Q'_{ss}) \left( \frac{t_{ox}}{\epsilon_{ox}} \right) + \phi_{ms} + 2\phi_{fp} \right. \\ &= \left( \sqrt{4e \times 10^{15} \times 0.01 \times 11.7 \epsilon_0 \phi_{fp}} - 5 \times 10^{10} e \right) \left( \frac{4 \times 10^{-6}}{0.01 \times 3.9 \epsilon_0} \right) - 1 + 2\phi_{fp} \\ &= -0.359 \text{ V.} \end{aligned}$$

(b) It is possible to apply a  $V_{SB}$  voltage such that  $V_T = 0$ . The body effect coefficient is

$$\gamma = \frac{2\epsilon_s e N_a}{C_{ox}} = \sqrt{2e \times 11.7 \times 0.01 \epsilon_0 \times 10^{15}} \frac{4 \times 10^{-6}}{3.9 \times 0.01 \epsilon_0} = 0.211 \text{ V}^{1/2}.$$

The value of  $\Delta V_T$  is

$$\Delta V_T = -V_{TN} = \gamma \left( \sqrt{2\phi_{fp} + V_{SB}} - \sqrt{2\phi_{fp}} \right)$$

Solving the equation, the value of  $V_{SB}$  is

$$V_{SB} = \left( \frac{-V_{TN}}{\gamma} + \sqrt{2\phi_{fp}} \right)^2 - 2\phi_{fp} = 5.46 \text{ V.}$$

4. (a) The transconductance is

$$\begin{aligned} g_{ms} &= \frac{k'_n W (V_{GS} - V_T)}{L} \\ &= \mu_n \frac{\epsilon}{t_{ox}} \frac{W}{L} (V_{GS} - V_T) \\ &= 400 \times \frac{3.9 \times 0.01 \epsilon_0}{4.75 \times 10^{-6}} \times 10 \times (5 - 0.65) \\ &= 1.265 \times 10^{-3} \text{ A/V.} \end{aligned}$$

We have the relation

$$g'_m = 0.8g_m = \frac{g_m}{1 + g_m r_s}.$$

Solving the equation, the value of source resistance is

$$r_s = 197.6 \Omega.$$

(b) The ratio is

$$\begin{aligned} \frac{g'_m}{g_m} &= \frac{1}{1 + \mu_n \frac{\epsilon}{t_{ox}} \frac{W}{L} (V_{GS} - V_T) r_s} \\ &= \frac{1}{1 + 400 \times \frac{3.9 \times 0.01 \epsilon_0}{4.75 \times 10^{-6}} \times 10 \times (3 - 0.65) r_s} \\ &= 0.881. \end{aligned}$$

Therefore,  $g_{ms}$  is reduced from its ideal value when  $V_G = 3$  V by 11.9 %.

5. (a) The total current is

$$I = 10^6 \times 10^{-15} \exp\left(\frac{V_{GS}}{2.1V_t}\right) = 10^{-9} \exp\left(\frac{eV_{GS}}{630k}\right).$$

At  $V_{GS} = 0.5$  V, the total current is

$$I = 10^{-9} \exp\left(\frac{0.5e}{630k}\right) = 9.996 \times 10^{-6} \text{ A}.$$

At  $V_{GS} = 0.7$  V, the total current is

$$I = 10^{-9} \exp\left(\frac{0.7e}{630k}\right) = 3.979 \times 10^{-4} \text{ A}.$$

At  $V_{GS} = 0.9$  V, the total current is

$$I = 10^{-9} \exp\left(\frac{0.9e}{630k}\right) = 1.584 \times 10^{-2} \text{ A}.$$

(b) The total power dissipated in the chip is

$$P = I_D V_D D = 5I_D.$$

The total power dissipated in the chip at  $V_{GS} = 0.5$  V is

$$P = 5I_D = 4.998 \times 10^{-5} \text{ W}.$$

The total power dissipated in the chip at  $V_{GS} = 0.7$  V is

$$P = 5I_D = 1.989 \times 10^{-3} \text{ W}.$$

The total power dissipated in the chip at  $V_{GS} = 0.9$  V is

$$P = 5I_D = 7.919 \times 10^{-2} \text{ W}.$$

6. (a) (i) The difference between  $E_{Fi}$  and  $E_F$  is

$$\phi_{fp} = \frac{kT}{e} \ln\left(\frac{N_a}{n_i}\right) = \frac{300k}{e} \ln\left(\frac{4 \times 10^{16}}{1.5 \times 10^{10}}\right) = 0.3825 \text{ V}.$$

The threshold voltage is

$$\begin{aligned} V_{TN} &= \left( (\sqrt{4eN_a\epsilon_s\phi_{fp}} - Q'_{ss}) \left( \frac{t_{ox}}{\epsilon_{ox}} \right) + \phi_{ms} + 2\phi_{fp} \right) \\ &= \left( \sqrt{4e \times 4 \times 10^{16} \times 0.01 \times 11.7\epsilon_0\phi_{fp}} - 4 \times 10^{10}e \right) \left( \frac{1.2 \times 10^{-6}}{0.01 \times 3.9\epsilon_0} \right) - 0.5 + 2\phi_{fp} \\ &= 0.593 \text{ V}. \end{aligned}$$

The drain-source saturation voltage is

$$V_{DS} = V_{GS} - V_{TN} = 1.25 - 0.593 = 0.657 \text{ V}.$$

The value of  $\Delta L$  is

$$\Delta L = \sqrt{\frac{2\epsilon}{eN_a}} \left( \sqrt{\phi_{fp} + V_{DS} + \Delta V_{DS}} - \sqrt{\phi_{fp} + V_{DS}} \right).$$

$\Delta L$  for  $\Delta V_{DS} = 1 \text{ V}$  is

$$\Delta L = \sqrt{\frac{2 \times 0.117\epsilon_0}{4 \times 10^{16}e}} \left( \sqrt{\phi_{fp} + V_{DS} + 1} - \sqrt{\phi_{fp} + V_{DS}} \right) = 7.346 \times 10^{-8} \text{ m}.$$

(ii)  $\Delta L$  for  $\Delta V_{DS} = 2 \text{ V}$  is

$$\Delta L = \sqrt{\frac{2 \times 0.117\epsilon_0}{4 \times 10^{16}e}} \left( \sqrt{\phi_{fp} + V_{DS} + 1} - \sqrt{\phi_{fp} + V_{DS}} \right) = 1.302 \times 10^{-7} \text{ m}.$$

(iii)  $\Delta L$  for  $\Delta V_{DS} = 4 \text{ V}$  is

$$\Delta L = \sqrt{\frac{2 \times 0.117\epsilon_0}{4 \times 10^{16}e}} \left( \sqrt{\phi_{fp} + V_{DS} + 1} - \sqrt{\phi_{fp} + V_{DS}} \right) = 2.203 \times 10^{-7} \text{ m}.$$

(b) We want

$$\frac{\Delta L}{L} = 0.12,$$

using result from (a) (iii), the minimum channel length is

$$L = 1.836 \times 10^{-6} \text{ m}.$$

7. (a) (i) The ideal drain current for  $V_{GS} = 0.8 \text{ V}$  is

$$I_D = \frac{k'_n}{2} \cdot \frac{W}{L} \cdot (V_{GS} - V_T)^2 = \frac{7.5 \times 10^{-5}}{2} \times 10 \times (0.8 - 0.35)^2 = 7.594 \times 10^{-5} \text{ A}.$$

(ii) The drain current if  $\lambda = 0.02 \text{ V}^{-1}$  is

$$I'_D = I_D(1 + \lambda V_{DS}) = I_D(1 + 0.02 \times 1.5) = 7.822 \times 10^{-5} \text{ A}.$$

(iii) The output resistance for  $\lambda = 0.02 \text{ V}^{-1}$  is

$$r_o = \frac{1}{\lambda I_D} = \frac{1}{0.02 I_D} = 6.584 \times 10^5 \Omega.$$

(b) (i) The ideal drain current for  $V_{GS} = 1.25 \text{ V}$  is

$$I_D = \frac{k'_n}{2} \cdot \frac{W}{L} \cdot (V_{GS} - V_T)^2 = \frac{7.5 \times 10^{-5}}{2} \times 10 \times (1.25 - 0.35)^2 = 3.0375 \times 10^{-4} \text{ A}.$$

(ii) The drain current if  $\lambda = 0.02 \text{ V}^{-1}$  is

$$I'_D = I_D(1 + \lambda V_{DS}) = I_D(1 + 0.02 \times 1.5) = 3.129 \times 10^{-4} \text{ A}.$$

(iii) The output resistance for  $\lambda = 0.02 \text{ V}^{-1}$  is

$$r_o = \frac{1}{\lambda I_D} = \frac{1}{0.02 I_D} = 1.646 \times 10^5 \Omega.$$

8. (a) (i) The maximum drain current in the original device is

$$I_D = \frac{k'_n}{2} \cdot \frac{W}{L} \cdot (V_{GS} - V_T)^2 = \frac{1.5 \times 10^{-4}}{2} \times \frac{6.0}{1.2} \times (3 - 0.45)^2 = 2.438 \times 10^{-3} \text{ A}.$$

(ii) The maximum drain current in the scaled device is

$$\begin{aligned} I_D &= \frac{k'_n}{2k} \cdot \frac{kW}{kL} \cdot (kV_{GS} - V_T)^2 \\ &= \frac{1.5 \times 10^{-4}}{2 \times 0.65} \times \frac{6.0}{1.2} \times (3 \times 0.65 - 0.45)^2 \\ &= 1.298 \times 10^{-3} \text{ A}. \end{aligned}$$

(b) (i) The maximum power dissipation in the original device is

$$P = I_D V_D = 3I_D = 7.315 \times 10^{-3} \text{ W}.$$

(ii) The maximum power dissipation in the scaled device is

$$P = I_D k V_D = 1.95 I_D = 2.531 \times 10^{-3} \text{ W}.$$

9. The difference between  $E_{Fi}$  and  $E_F$  is

$$\phi_{fp} = \frac{kT}{e} \ln \left( \frac{N_a}{n_i} \right) = \frac{300k}{e} \ln \left( \frac{2 \times 10^{16}}{1.5 \times 10^{10}} \right) = 0.3646 \text{ V}.$$

The maximum charge width is

$$x_{dT} = \sqrt{\frac{4\epsilon_s \phi_{fp}}{eN_a}} = \sqrt{\frac{4 \times 11.7 \times 0.01\epsilon_0 \phi_{fp}}{2 \times 10^{16}e}} = 2.171 \times 10^{-5} \text{ cm}.$$

The threshold voltage shift is

$$\begin{aligned} \Delta V_T &= -\frac{eN_a x_{dT}}{C_{ox}} \left[ \frac{r_j}{L} \left( \sqrt{1 + \frac{2x_{dT}}{r_j}} - 1 \right) \right] \\ &= -\frac{2 \times 10^{16} e x_{dT} \times 8 \times 10^{-7}}{3.9 \times 0.01\epsilon_0} \left[ \frac{0.30}{0.70} \left( \sqrt{1 + \frac{2x_{dT}}{3 \times 10^{-5}}} - 1 \right) \right] \\ &= -0.0390 \text{ V}. \end{aligned}$$

The equivalent long-channel threshold voltage is

$$V_{TO} = V_T - \Delta V_T = 0.3890 \text{ V}.$$

10. The difference between  $E_{Fi}$  and  $E_F$  is

$$\phi_{fp} = \frac{kT}{e} \ln \left( \frac{N_a}{n_i} \right) = \frac{300k}{e} \ln \left( \frac{3 \times 10^{16}}{1.5 \times 10^{10}} \right) = 0.3751 \text{ V}.$$

The maximum charge width is

$$x_{dT} = \sqrt{\frac{4\epsilon_s \phi_{fp}}{eN_a}} = \sqrt{\frac{4 \times 11.7 \times 0.01\epsilon_0 \phi_{fp}}{3 \times 10^{16}e}} = 1.798 \times 10^{-5} \text{ cm}.$$

The shift in threshold voltage due to narrow-channel effects is

$$\Delta V_T = \frac{eN_a x_{dT}}{C_{ox}} \left( \frac{\xi x_{dT}}{W} \right) = \frac{3 \times 10^{16} e x_{dT} \times 8 \times 10^{-7}}{3.9 \times 0.01\epsilon_0} \left( \frac{\frac{\pi}{2} x_{dT}}{2.2 \times 10^{-4}} \right) = 0.02571 \text{ V}.$$