

**ECES-352**  
**Homework #5**  
**Solutions**

Reading: Chapter 4 on Sampling

**PROBLEM 1**

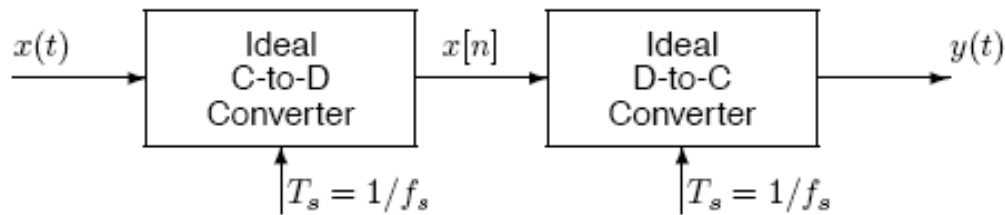


Figure 1: Ideal sampling and reconstruction system.

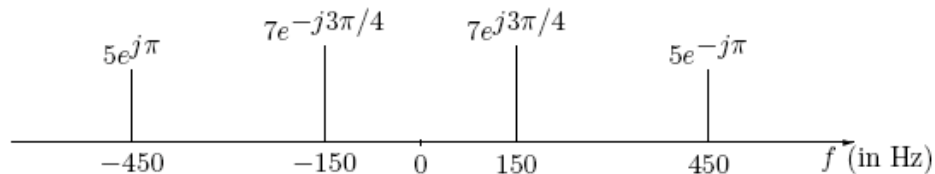
Again consider the ideal sampling and reconstruction system shown in Figure 1 of the previous problem.

- (a) Suppose that the discrete-time signal  $x[n]$  in Figure 1 is given by the formula

$$x[n] = 3 \cos(0.25\pi n + \pi/5)$$

If the sampling rate of the C-to-D converter is  $f_s = 11000$  samples/second, many *different* continuous-time signals  $x(t) = x_\ell(t)$  could have been inputs to the above system. Determine two such inputs with frequency less than 11000 Hz; i.e., find  $x_1(t) = A_1 \cos(\omega_1 t + \phi_1)$  and  $x_2(t) = A_2 \cos(\omega_2 t + \phi_2)$  such that  $x[n] = x_1(nT_s) = x_2(nT_s)$  if  $T_s = 1/10000$  secs.

- (b) Now if the input  $x(t)$  to the system in Figure 1 of Problem 5.1 has the two-sided spectrum representation shown below, what is the *minimum* sampling rate  $f_s$  such that the output  $y(t)$  is equal to the input  $x(t)$ ?



- (c) Determine the spectrum for  $x[n]$  when  $f_s = 450$  samples/sec. Make a plot for your answer, but label the frequency, amplitude and phase of each spectral component.

$$(a) \quad x[n] = 3 \cos(0.25\pi n + \pi/5)$$

$$f_s = 11000 \text{ Hz}$$

$$\hat{\omega} = 0.25\pi = \frac{\omega}{f_s}$$

$$\omega_1 = (11000 \times 0.25\pi) = 2750\pi \text{ rad/s}, \quad \boxed{f_1 = 1375 \text{ Hz}}$$

$$\hat{\omega} = (2\pi - 0.25\pi) = 1.75\pi = \frac{\omega}{f_s}$$

$$\omega_2 = (11000 \times 1.75\pi) = 19250\pi \text{ rad/s}, \quad \boxed{f_2 = 9625 \text{ Hz (folded alias)}}$$

$$(b) \quad f_{\max} = 450 \text{ Hz}$$

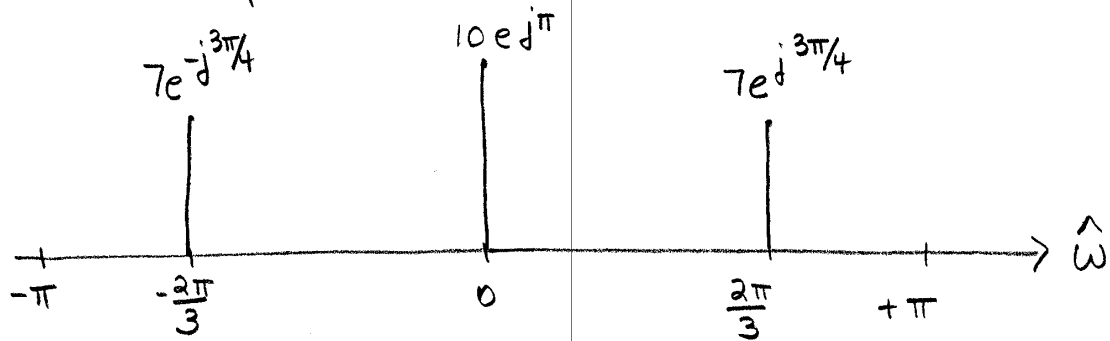
$$\text{minimum } f_s = 2 \cdot f_{\max} = \boxed{900 \text{ Hz}}$$

$$(c) \quad f_s = 450 \text{ Hz}$$

$$\hat{\omega}_1 = \frac{150 \cdot 2\pi}{450} = \frac{2\pi}{3}$$

$$\hat{\omega}_2 = \frac{450 \cdot 2\pi}{450} = 2\pi$$

Note that the plus & minus components at  $f = 450 \text{ Hz}$  both map to  $\hat{\omega} = 0$  and thus reinforce.



## PROBLEM 2

A non-ideal D-to-C converter takes a sequence  $y[n]$  as input and produces a continuous-time output  $y(t)$  according to the relation

$$y(t) = \sum_{n=-\infty}^{\infty} y[n]p(t - nT_s)$$

where  $T_s = 0.1$  second. The input sequence is given by the formula

$$y[n] = \begin{cases} 32 & 0 \leq n \leq 4 \\ 32(.5)^{n-4} & 5 \leq n \leq 9 \\ 0 & \text{otherwise} \end{cases}$$

(a) Plot  $y[n]$  versus  $n$ .

(b) For the pulse shape

$$p(t) = \begin{cases} 1 & -0.05 \leq t \leq 0.05 \\ 0 & \text{otherwise} \end{cases}$$

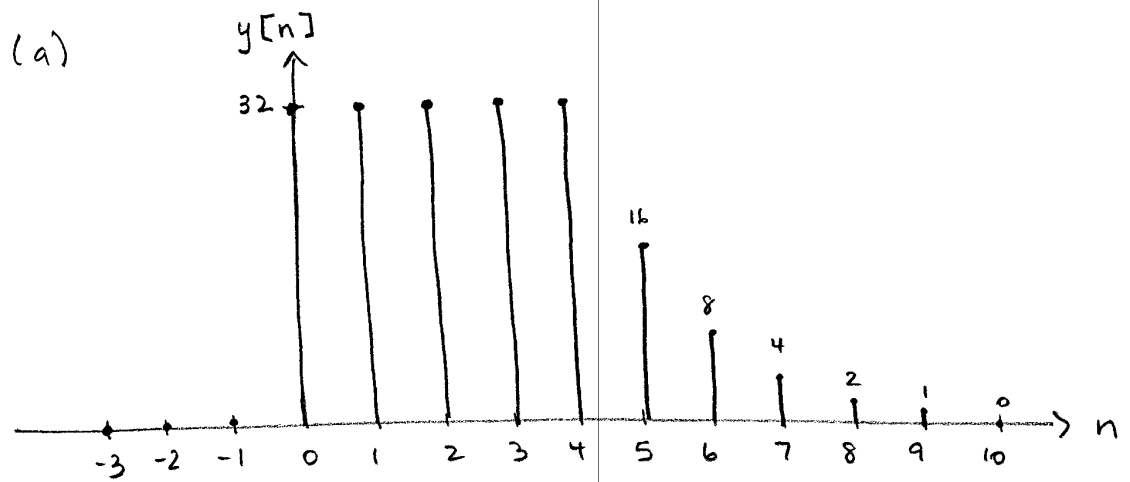
carefully sketch the output waveform  $y(t)$  over its non-zero region.

(c) For the pulse shape

$$p(t) = \begin{cases} 1 - 10|t| & -0.1 \leq t \leq 0.1 \\ 0 & \text{otherwise} \end{cases}$$

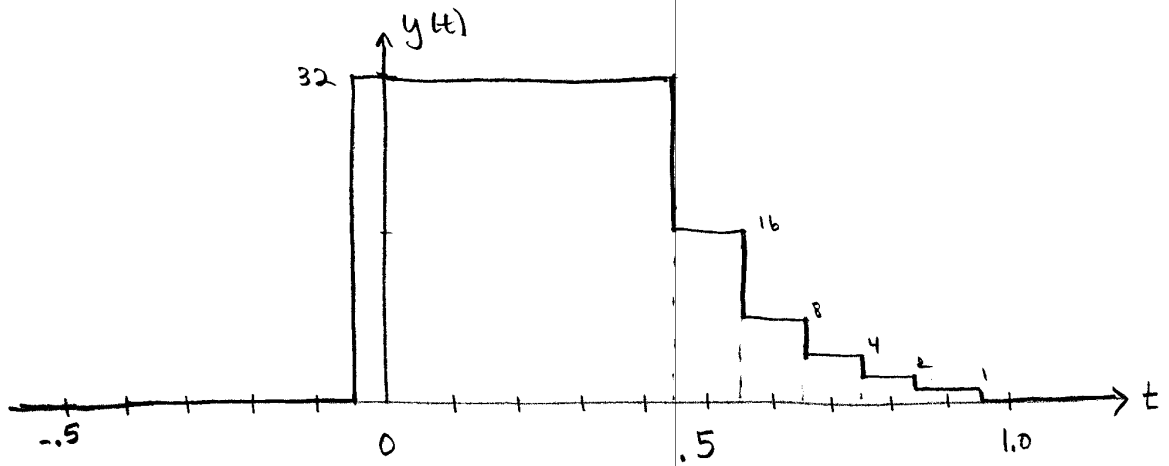
carefully sketch the output waveform  $y(t)$  over its non-zero region.

$$y[n] = \begin{cases} 32 & 0 \leq n \leq 4 \\ 32(.5)^{n-4} & 5 \leq n \leq 9 \\ 0 & \text{otherwise} \end{cases}$$



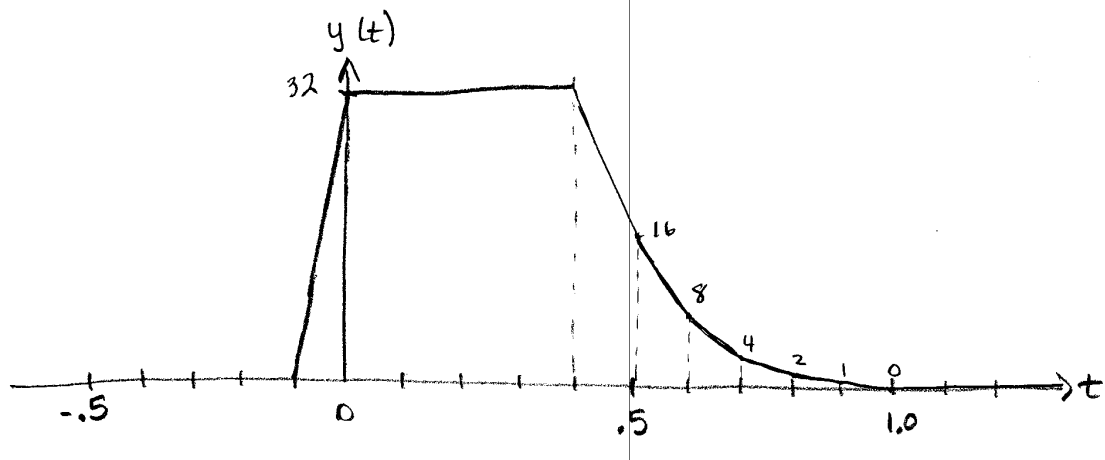
(b)

$$p(t) = \begin{cases} 1 & -0.05 \leq t \leq 0.05 \\ 0 & \text{otherwise} \end{cases}$$

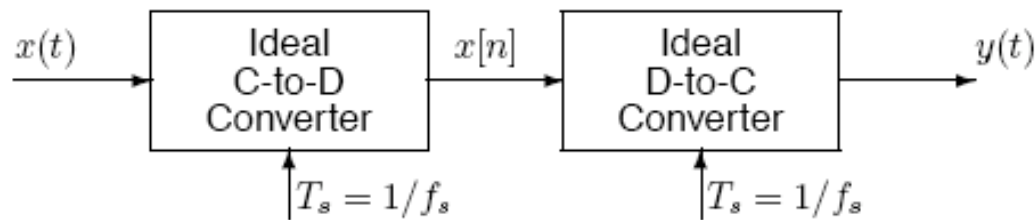


(c)

$$p(t) = \begin{cases} 1 - 10|t| & -0.1 \leq t \leq 0.1 \\ 0 & \text{otherwise} \end{cases}$$



### PROBLEM 3

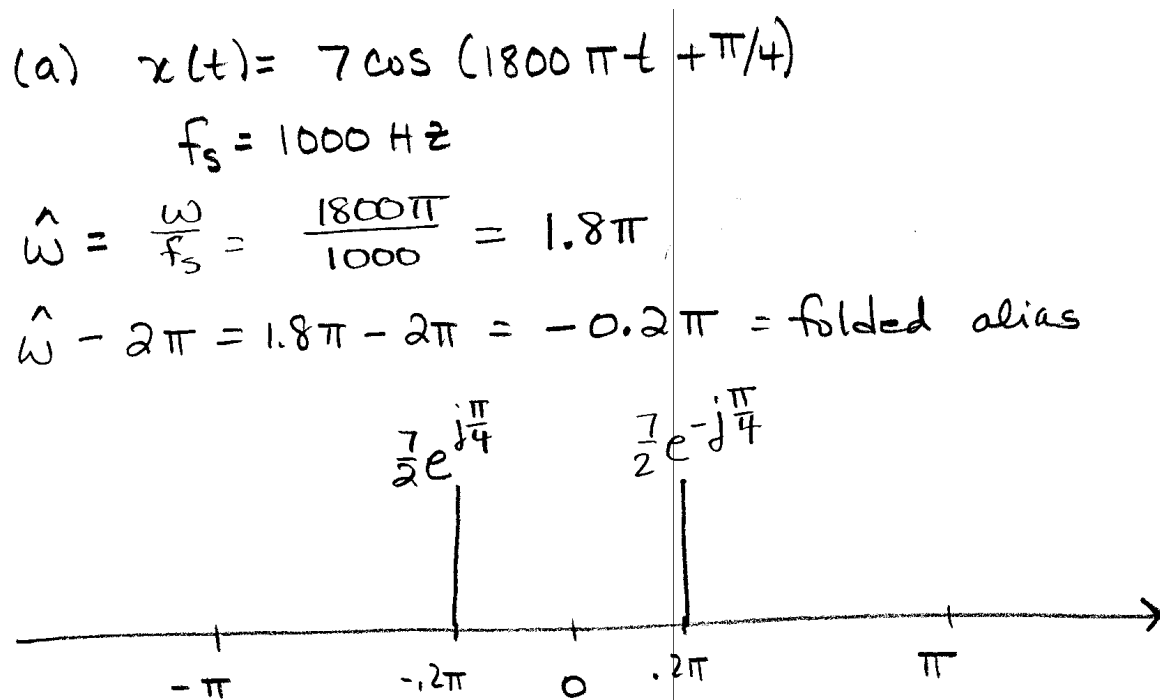


Chirps are very useful signals for probing the behavior of sampling operations and illustrating the “folding” type of aliasing (see Fig. 4.4 in the book).

- If the input to the ideal C/D converter is  $x(t) = 7 \cos(1800\pi t + \pi/4)$ , and the sampling frequency is 1000 Hz, then the output  $y(t)$  is a sinusoid. Determine the formula for the output signal.
- Suppose that the input signal is a chirp signal defined as follows:

$$x(t) = \cos(2000\pi t - 400\pi t^2) \quad \text{for } 0 \leq t \leq 5 \text{ sec.}$$

If the sampling rate is  $f_s = 1000$  Hz, then the output signal  $y(t)$  will have time-varying frequency content. Draw a graph of the resulting analog *instantaneous* frequency (in Hz) versus time of the signal  $y(t)$  **after reconstruction**. Hint: this could be done in MATLAB by putting a sampled chirp signal into the MATLAB function `specgram()`, or the DSP-First function `plotspec()`.



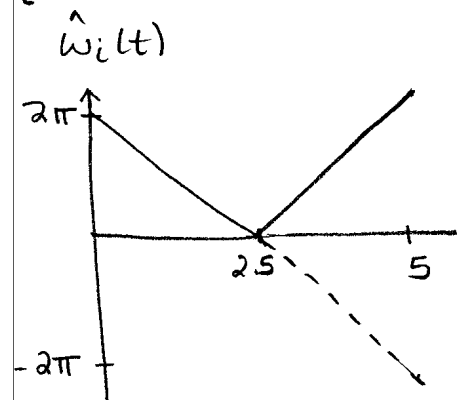
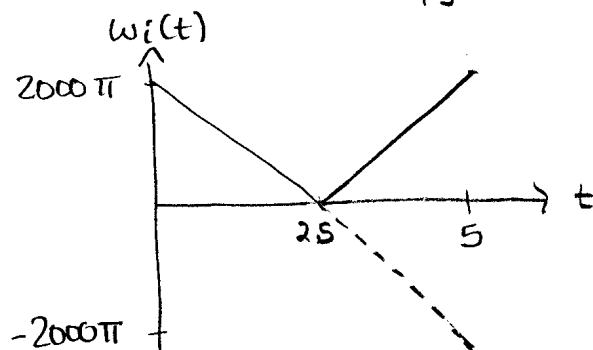
Reconstruction:  $\omega = \hat{\omega} \cdot f_s = 0.2\pi(1000) = 200\pi$

$$y(t) = 7 \cos(200\pi t - \pi/4)$$

(b)  $x(t) = \cos(2000\pi t - 400\pi t^2)$   $0 \leq t \leq 5$

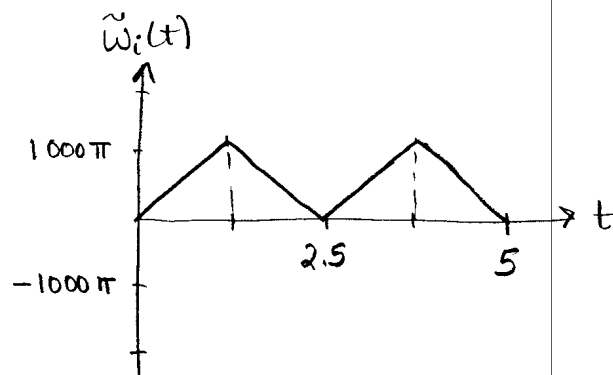
$$\omega_i(t) = 2000\pi - 800\pi t$$

$$\hat{\omega}_i(t) = \frac{\omega_i(t)}{f_s} = 2\pi - 0.8\pi t$$

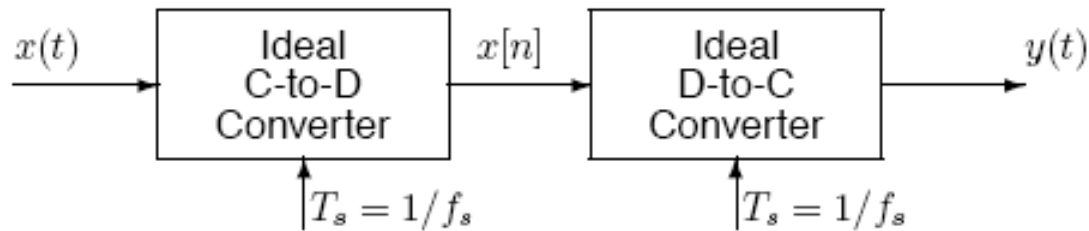


Since  $|\hat{\omega}_i| > \pi$ , aliasing occurs.

After reconstruction:



#### PROBLEM 4



In all parts below, the sampling rates of the C/D and D/C converters are equal, and the input to the Ideal C/D converter is

$$x(t) = 3 \cos(2\pi(50)t - \pi/2) + 2 \cos(2\pi(300)t).$$

- (a) If the output of the ideal D-to-C Converter is

$$y(t) = x(t) = 3 \cos(2\pi(50)t - \pi/2) + 2 \cos(2\pi(300)t),$$

what general statement can you make about the sampling frequency  $f_s$  in this case?

- (b) If the sampling rate is  $f_s = 200$  samples/sec., determine the discrete-time signal  $x[n]$ , and give an expression for  $x[n]$  as a sum of cosines. *Make sure that all frequencies in your answer are positive and less than  $\pi$  radians.* Plot the spectrum of this signal over the range of frequencies  $-\pi \leq \omega \leq \pi$ . Make a plot for your answer, but label the frequency, amplitude and phase of each spectral component.

- (c) If the output of the Ideal D-to-C Converter is

$$y(t) = 3 \cos(2\pi(50)t - \pi/2) + 2,$$

determine the value of the sampling frequency  $f_s$ . (Remember that the input  $x(t)$  is as defined above.)

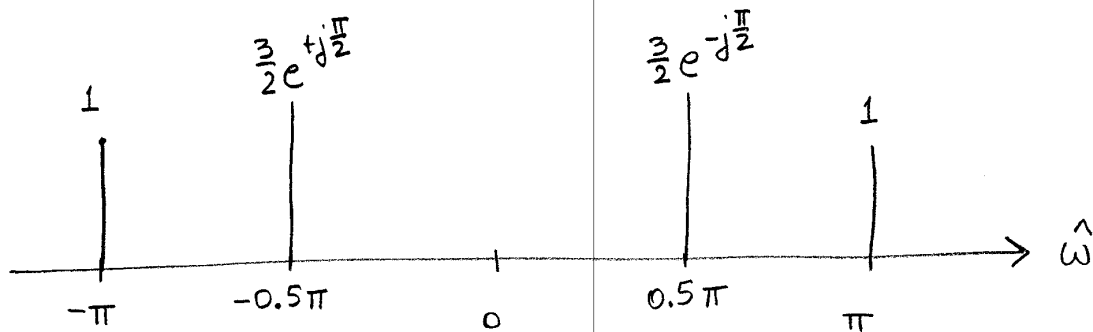
$$x(t) = 3 \cos(2\pi(50)t - \pi/2) + 2 \cos(2\pi(300)t)$$

(a) output is identical to input; therefore  $f_s > 600 \text{ Hz}$

(b)  $f_s = 200 \text{ Hz}$

$$\begin{aligned} x[n] &= 3 \cos\left(2\pi \cdot 50 \cdot \frac{n}{200} - \frac{\pi}{2}\right) + 2 \cos\left(2\pi \cdot 300 \cdot \frac{n}{200}\right) \\ &= 3 \cos\left(\frac{\pi}{2}n - \frac{\pi}{2}\right) + 2 \cos(3\pi n) \end{aligned}$$

$$\boxed{x[n] = 3 \cos\left(\frac{\pi}{2}n - \frac{\pi}{2}\right) + 2 \cos(\pi n)}$$



(c) The DC term is an alias of the original 300 Hz term.

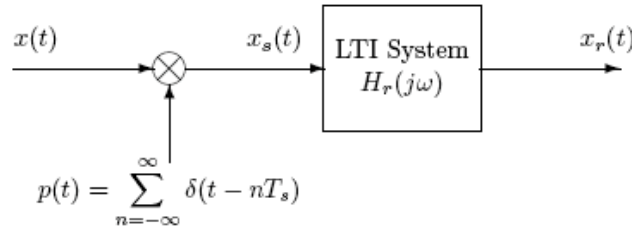
$$\hat{\omega} = 2\pi \text{ for } \omega = 300 \text{ Hz}$$

$$\hat{\omega} = 2\pi = \frac{2\pi \cdot 300}{f_s} ; \quad \boxed{f_s = 300 \text{ Hz}}$$



## PROBLEM 5

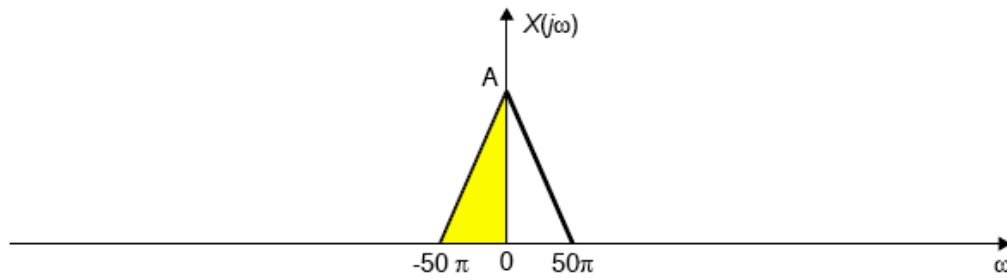
The derivation of the Sampling Theorem involves the operations of impulse train sampling and reconstruction as shown in the following system:



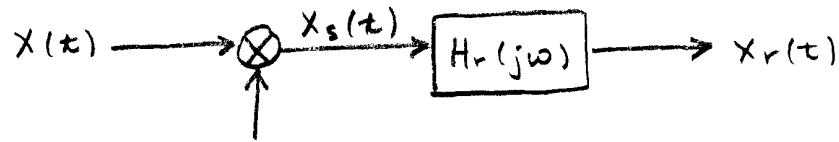
The LTI system is the ideal bandlimited reconstruction filter with frequency response given by

$$H_r(j\omega) = \begin{cases} T_s & |\omega| \leq \pi/T_s \\ 0 & |\omega| > \pi/T_s. \end{cases}$$

The “typical” bandlimited Fourier transform of the input is depicted below:

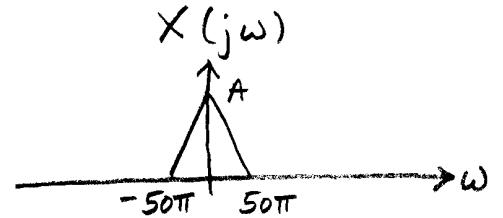


- For the input with Fourier transform depicted above, use the Sampling Theorem to choose the sampling rate  $\omega_s = 2\pi/T_s$  so that  $x_r(t) = x(t)$ . Plot  $X_s(j\omega)$  for the value of  $\omega_s = 2\pi/T_s$  that is equal to the *Nyquist* rate.<sup>1</sup>
- If  $\omega_s = 2\pi/T_s = 80\pi$  in the above system and  $X(j\omega)$  is as depicted above, plot the Fourier transform  $X_s(j\omega)$  and show that aliasing occurs. There will be an infinite number of shifted copies of  $X(j\omega)$ , so indicate what the pattern is versus  $\omega$ .
- For the conditions of part (b), determine and sketch the Fourier transform of the output  $X_r(j\omega)$ .

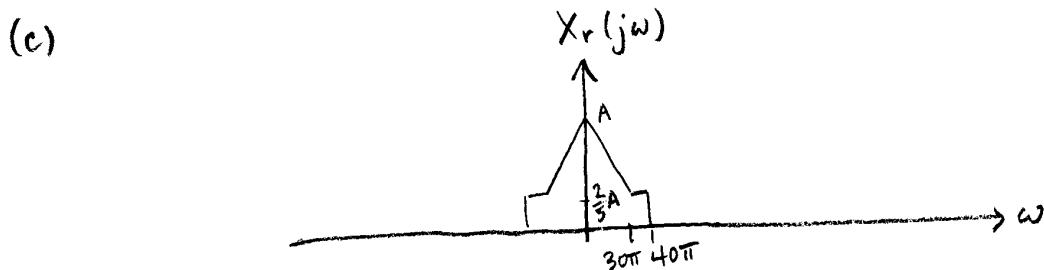
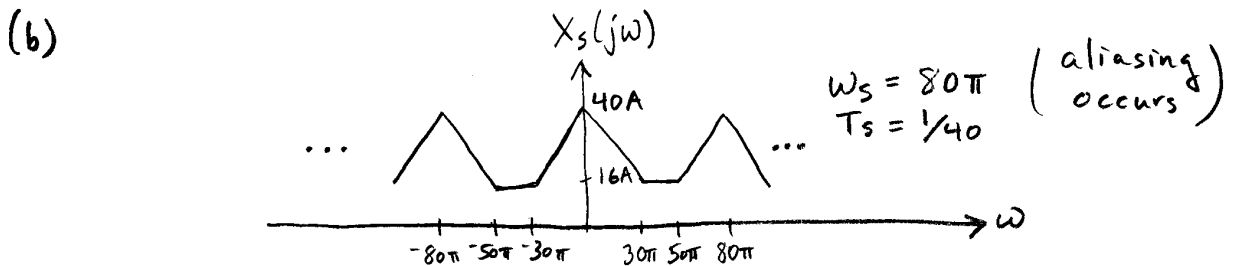
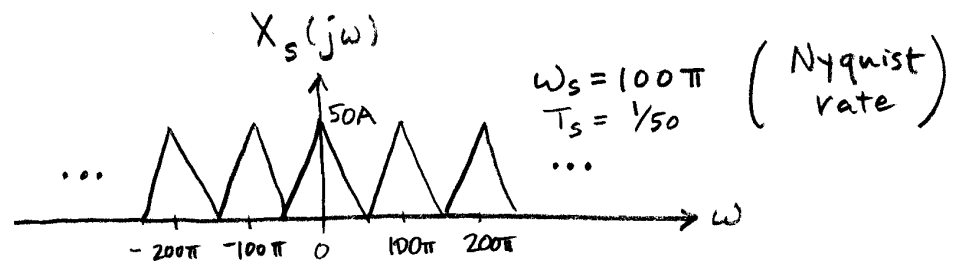


$$p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$$

$$H_r(j\omega) = \begin{cases} T_s, & |\omega| \leq \pi/T_s \\ 0, & \text{else} \end{cases}$$



(a) To obtain  $X_r(t) = X(t)$ , choose  $\omega_s \geq 100\pi$ .



## PROBLEM 6

- a) Determine the DTFT of the following signals:
- i)  $x[n] = \delta[n-3]$
  - ii)  $x[n] = \frac{1}{2} \delta[n+1] + \delta[n] + \frac{1}{2} \delta[n-1]$
  - iii)  $x[n] = (1/4)^{n-3} u[n-3]$
  - iv)  $x[n] = u[n+3] - u[n-4]$
- b) Consider a radix-2 16-point FFT. Please answer the following questions assuming a butterfly-type decimation-in-time implementation discussed in class.
- i) How many butterfly stages are required?
  - ii) How many butterflies per stage are required?
  - iii) How many 2-pt butterflies are required in total?
  - iv) Assuming a decimation-in-time FFT, what is the sequential order of the input samples to compute the fft?
  - v) Extra credit: What is the sequential order of input samples for decimation-infrequency?