

Wireless Communication Systems

@CS.NCTU

Lecture 5: Multiple-Input Multiple-Output (MIMO)

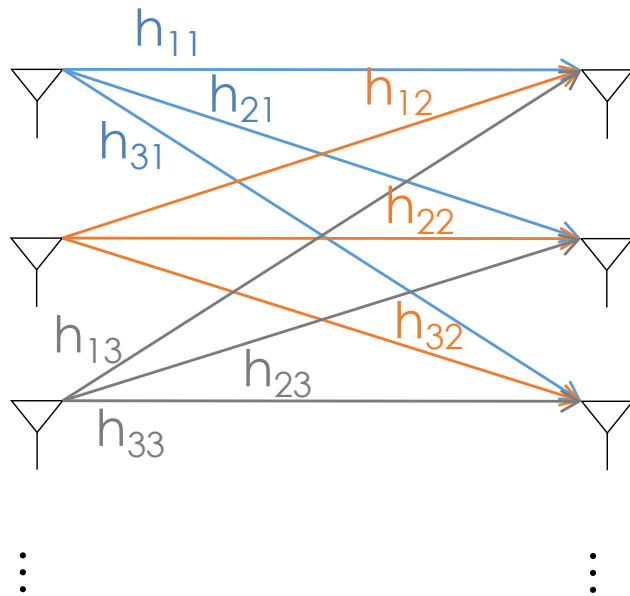
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Agenda

- Channel model
- MIMO decoding
- Degrees of freedom
- Multiplexing and Diversity

MIMO

- Each node has multiple antennas
 - Capable of transmitting (receiving) multiple streams concurrently
 - Exploit antenna diversity to increase the capacity

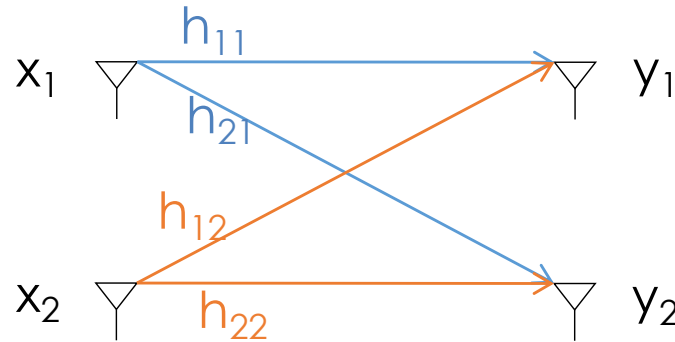


$$\mathbf{H}_{N \times M} = \begin{bmatrix} h_{11} & h_{12} & h_{13} \\ h_{21} & h_{22} & h_{23} \\ h_{31} & h_{32} & h_{33} \end{bmatrix}$$

N: number of antennas at Rx
M: number of antennas at Tx
 H_{ij} : channel from the j -th Tx antenna to the i -th Rx antenna

Channel Model (2x2)

- Say a 2-antenna transmitter sends 2 streams simultaneously to a 2-antenna receiver



Equations

$$y_1 = h_{11}x_1 + h_{12}x_2 + n_1$$

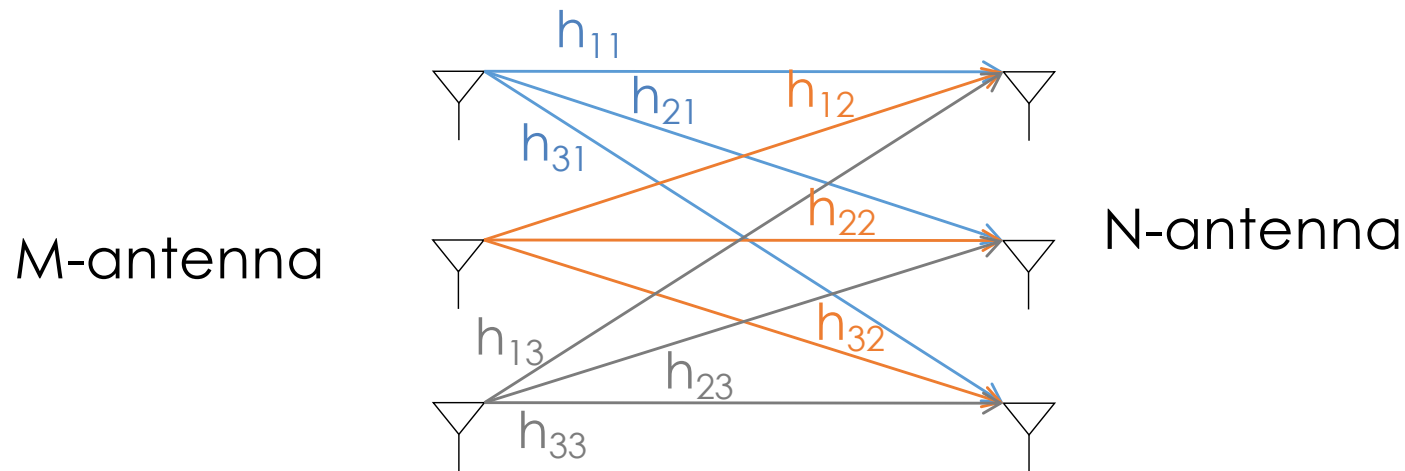
$$y_2 = h_{21}x_1 + h_{22}x_2 + n_2$$

Matrix form: $\mathbf{y} = \mathbf{H}\mathbf{x} + \mathbf{n}$

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} n_1 \\ n_2 \end{pmatrix}$$

MIMO (MxN)

- An M-antenna Tx sends to an N-antenna Rx



$$\mathbf{y} = \mathbf{H}\mathbf{x} + \mathbf{n} \quad \vdots$$

$$\rightarrow \begin{pmatrix} y_1 \\ y_2 \\ \vdots \\ y_N \end{pmatrix} = \begin{pmatrix} h_{11} & h_{12} & \cdots & h_{1M} \\ h_{21} & h_{22} & \cdots & h_{2M} \\ \vdots & \vdots & \ddots & \vdots \\ h_{N1} & h_{N2} & \cdots & h_{NM} \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \\ \vdots \\ x_M \end{pmatrix} + \begin{pmatrix} n_1 \\ n_2 \\ \vdots \\ n_N \end{pmatrix}$$

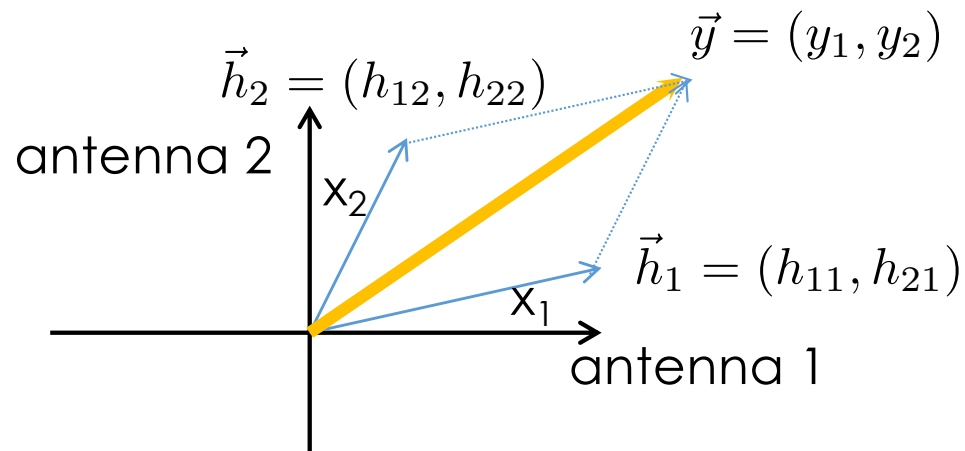
Antenna Space (2x2, 3x3)

N-antenna node receives in N-dimensional space

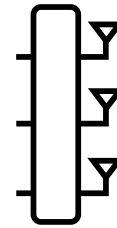
2 x 2

$$\begin{bmatrix} \text{Antenna} \end{bmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} h_{11} \\ h_{21} \end{pmatrix} x_1 + \begin{pmatrix} h_{12} \\ h_{22} \end{pmatrix} x_2 + \begin{pmatrix} n_1 \\ n_2 \end{pmatrix}$$

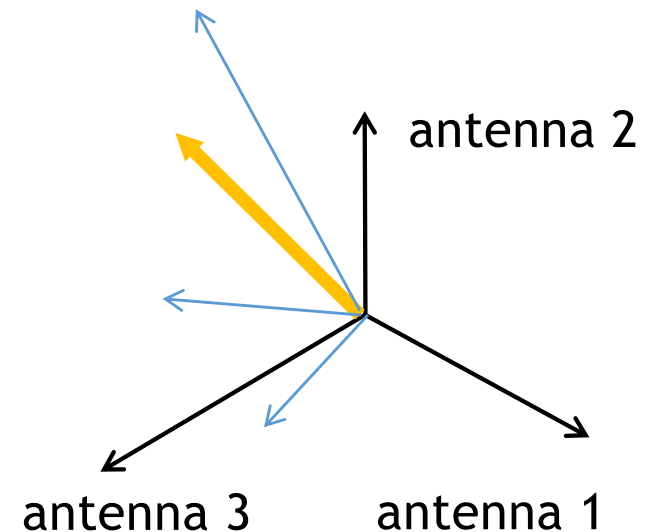
$$\vec{y} = \vec{h}_1 x_1 + \vec{h}_2 x_2 + \vec{n}$$



3 x 3



$$\vec{y} = \vec{h}_1 x_1 + \vec{h}_2 x_2 + \vec{h}_3 x_3 + \vec{n}$$



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Zero-Forcing (ZF) Decoding

- Decode x_1

$$\begin{array}{rcl} \left(\begin{array}{c} y_1 \\ y_2 \end{array} \right) & = & \left(\begin{array}{c} h_{11} \\ h_{21} \end{array} \right) x_1 + \left(\begin{array}{c} h_{12} \\ h_{22} \end{array} \right) x_2 + \left(\begin{array}{c} n_1 \\ n_2 \end{array} \right) \end{array}$$

orthogonal vectors

* h_{22}
* $-h_{12}$

$$y_1 h_{22} - y_2 h_{12} = (h_{11} h_{22} - h_{21} h_{12}) x_1 + n'$$

$$\begin{aligned} x'_1 &= \frac{y_1 h_{22} - y_2 h_{12}}{h_{11} h_{22} - h_{21} h_{12}} \\ &= x_1 + \frac{n'}{h_{11} h_{22} - h_{21} h_{12}} \\ &= x_1 + \frac{n'}{\vec{h}_1 \cdot \vec{h}_2^\perp} \end{aligned}$$

Zero-Forcing (ZF) Decoding

- Decode x_2

orthogonal vectors

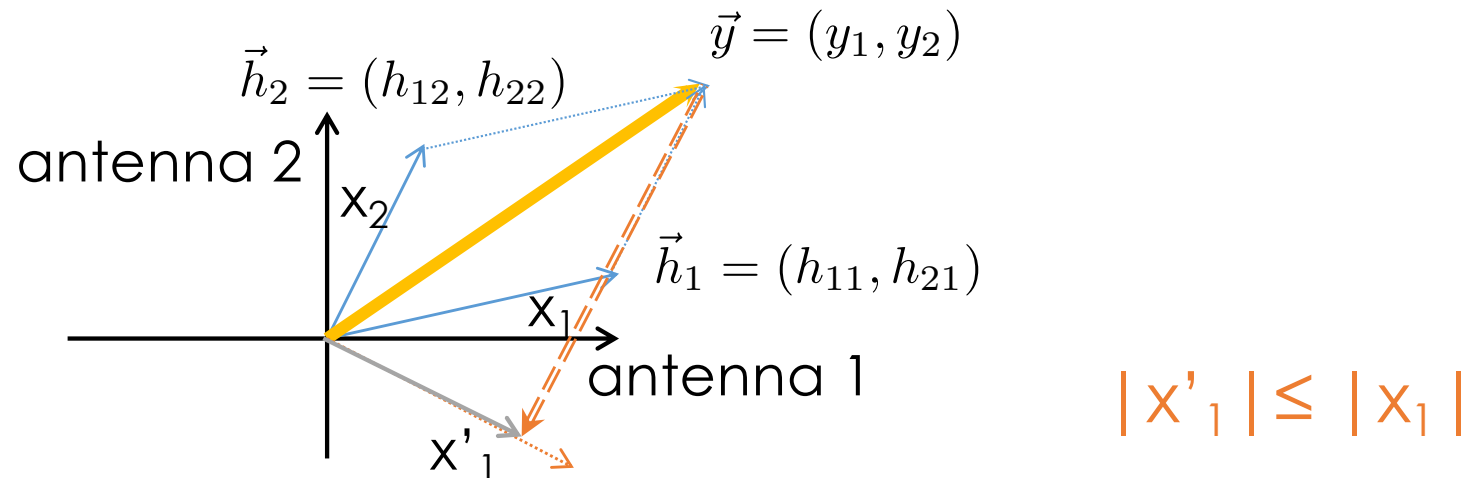
$$+) \quad \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} h_{11} \\ h_{21} \end{pmatrix} x_1 + \begin{pmatrix} h_{12} \\ h_{22} \end{pmatrix} x_2 + \begin{pmatrix} n_1 \\ n_2 \end{pmatrix}$$

* h_{21}
* $-h_{11}$

$$y_1 h_{21} - y_2 h_{11} = (h_{12} h_{21} - h_{22} h_{11}) x_2 + n'$$

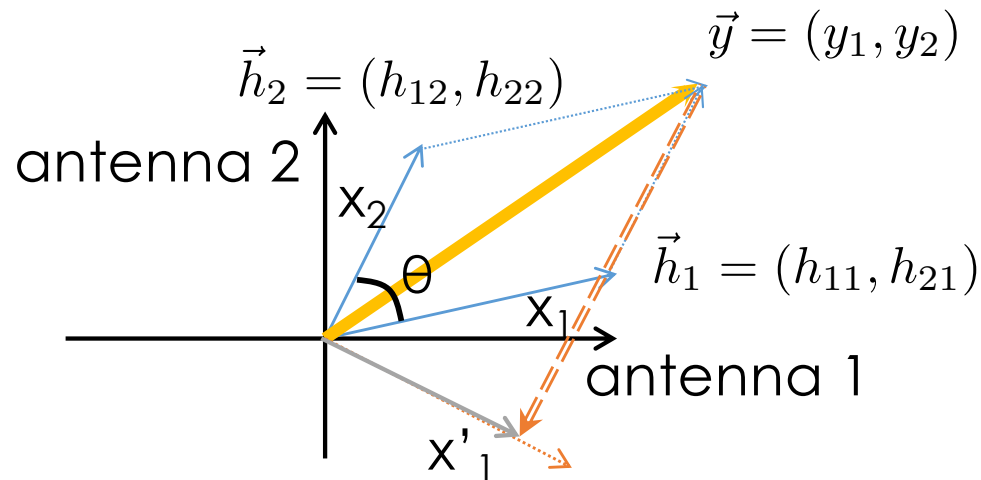
$$\begin{aligned} x'_2 &= \frac{y_1 h_{21} - y_2 h_{11}}{h_{12} h_{21} - h_{22} h_{11}} \\ &= x_2 + \frac{n'}{h_{12} h_{21} - h_{22} h_{11}} \\ &= x_2 + \frac{n'}{\vec{h}_2 \cdot \vec{h}_1^\perp} \end{aligned}$$

ZF Decoding (antenna space)



- To decode x_1 , project the received signal y onto the **interference-free** direction h_2^\perp
- To decode x_2 , project the received signal y onto the **interference-free** direction h_1^\perp
- SNR reduces if the channels h_1 and h_2 are correlated, i.e., not perfect orthogonal ($h_1 \cdot h_2 \neq 0$)

SNR Loss due to ZF Detection



$$|x'_1|^2 = |x_1|^2 \cos^2(90 - \theta) = |x_1|^2 \sin^2(\theta)$$

- From equation: $x'_1 = x_1 + \frac{n}{\vec{h}_1 \cdot \vec{h}_2^\perp}$

$$\text{SNR}_{\text{ZF}} = \text{SNR}_{\text{SISO}} \text{ when } h_1 \perp h_2$$

$$\text{SNR}' = \frac{|x_1|^2}{N_0 / (\vec{h}_1 \cdot \vec{h}_2^\perp)^2} = \frac{|x_1|^2 \sin^2(\theta)}{N_0} = \text{SNR} * \sin^2(\theta)$$

- The more correlated the channels (the smaller angles), the larger SNR reduction

When will MIMO Fail?

- In the worst case, SNR might drop down to 0 if the channels are strongly correlated to each other, e.g., $h_1 \parallel h_2$ in the 2x2 MIMO
- To ensure channel independency, should guarantee the full rank of H
 - Antenna spacing at the transmitter and receiver must **exceed half of the wavelength**

ZF Decoding – General Eq.

- For a $N \times M$ MIMO system,

$$\mathbf{y} = \mathbf{H}\mathbf{x} + \mathbf{n}$$

- To solve \mathbf{x} , find a decoder \mathbf{W} satisfying the constraint

$$\mathbf{W}\mathbf{H} = \mathbf{I}, \text{ then } \mathbf{x}' = \mathbf{W}\mathbf{y} = \mathbf{x} + \mathbf{W}\mathbf{n}$$

→ \mathbf{W} is the pseudo inverse of \mathbf{H}

$$\mathbf{W} = (\mathbf{H}^*\mathbf{H})^{-1}\mathbf{H}^*$$

→ Noise may be amplified! $\mathbf{n}' = \mathbf{W}\mathbf{n}$

ZF-SIC Decoding

- Combine ZF with SIC to improve SNR
 - Decode one stream and subtract it from the received signal
 - Repeat until all the streams are recovered
 - Example: after decoding x_2 , we have $y_1 = h_1x_1 + n_1$
→ decode x_1 using standard SISO decoder

Why ZF-SIC is Better

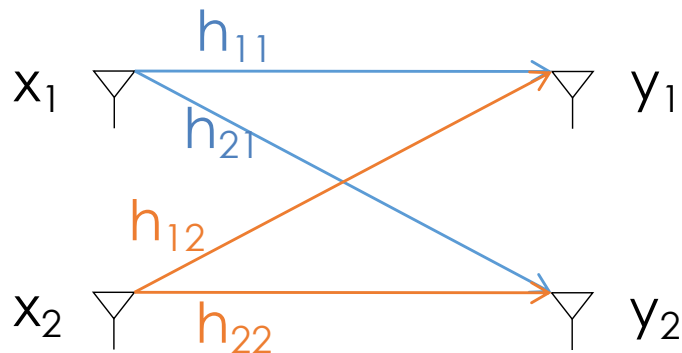
- Why it achieves a higher SNR?
 - The streams recovered after SIC can be projected to a smaller subspace
→ lower SNR reduction
 - In the 2x2 example, x_1 can be decoded as usual without ZF
→ no SNR reduction (though x_2 still experience SNR loss)

Other Detection Schemes

- Maximum-Likelihood (ML) decoding
 - Measure the distance between the received signal and all the possible symbol vectors
 - Optimal Decoding
 - High complexity (exhaustive search)
- Minimum Mean Square Error (MMSE) decoding
 - Minimize the mean square error
 - Bayesian approach: conditional expectation of \mathbf{x} given the known observed value of the measurements
- ML-SIC, MMSE-SIC

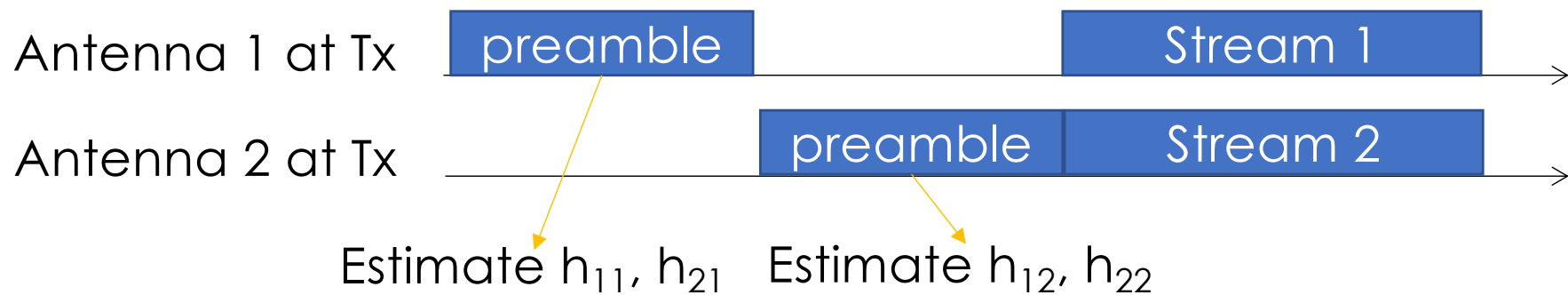
Channel Estimation

- Estimate $N \times M$ matrix H



$$y_1 = h_{11}x_1 + h_{12}x_2 + n_1$$
$$y_2 = h_{21}x_1 + h_{22}x_2 + n_2$$

Two equations, but four unknowns



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Degree of Freedom

For $N \times M$ MIMO channel

- Degree of Freedom (DoF): $\min \{N, M\}$
 - Can transmit at most DoF streams
- Maximum diversity: NM
 - There exist NM paths among Tx and Rx

MIMO Gains

- Multiplex Gain

- Exploit DoF to deliver multiple streams concurrently

- Diversity Gain

- Exploit path diversity to increase the SNR of a single stream
- Receive diversity and transmit diversity

Multiplexing-Diversity Tradeoff

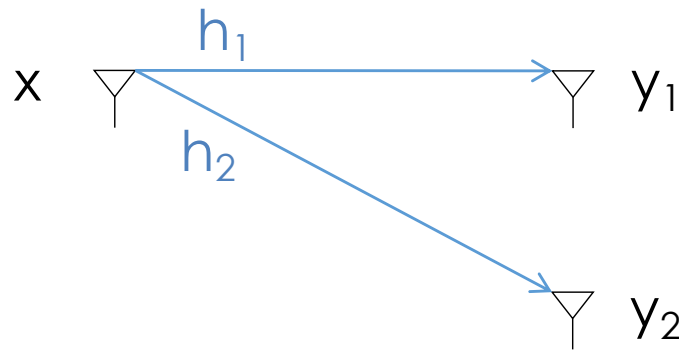
- Tradeoff between the diversity gain and the multiplex gain
- Say we have a $N \times N$ system
 - Degree of freedom: N
 - The transmitter can send k streams concurrently, where $k \leq N$
 - If $k < N$, leverage partial multiplexing gains, while each stream gets some diversity
 - The optimal value of k maximizing the capacity should be determined by the tradeoff between the diversity gain and multiplex gain

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Receive Diversity

- 1 x 2 example



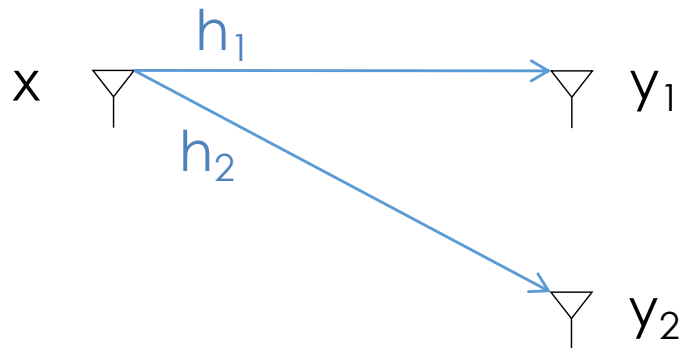
$$y_1 = h_1x + n_1$$

$$y_2 = h_2x + n_2$$

- Uncorrelated whit Gaussian noise with zero mean
- Packet can be delivered through at least one of the many diverse paths

Theoretical SNR of Receive Diversity

- 1 x 2 example



- Increase SNR by 3dB
- Especially beneficial for the low SNR link

$$\begin{aligned}\text{SNR} &= \frac{P(2X)}{P(n_1 + n_2)}, \text{ where } P \text{ refers to the power} \\ &= \frac{E[(2X)^2]}{E[n_1^2 + n_2^2]} \\ &= \frac{4E[X^2]}{2\sigma}, \text{ where } \sigma \text{ is the variance of AWGN} \\ &= 2 * \text{SNR}_{\text{single antenna}}\end{aligned}$$

Maximal Ratio Combining (MRC)

- Extract receive diversity via MRC decoding
- Multiply each \mathbf{y} with the conjugate of the channel

$$\begin{aligned} y_1 &= h_1 x + n_1 \\ y_2 &= h_2 x + n_2 \end{aligned} \implies \begin{aligned} h_1^* y_1 &= |h_1|^2 x + h_1^* n_1 \\ h_2^* y_2 &= |h_2|^2 x + h_2^* n_2 \end{aligned}$$

- Combine two signals constructively

$$h_1^* y_1 + h_2^* y_2 = (|h_1|^2 + |h_2|^2)x + (h_1^* + h_2^*)n$$

- Decode using the standard SISO decoder

$$x' = \frac{h_1^* y_1 + h_2^* y_2}{(|h_1|^2 + |h_2|^2)} + n'$$

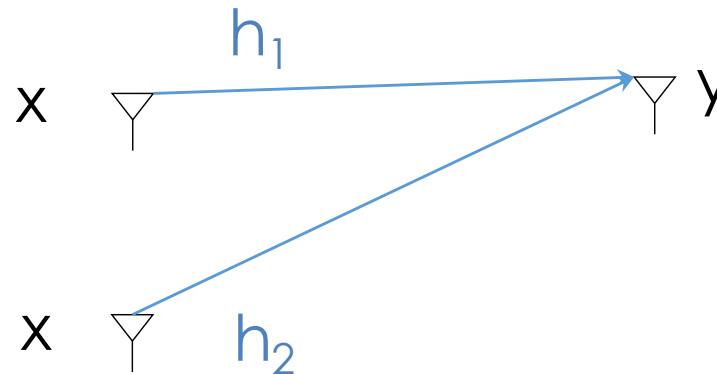
Achievable SNR of MRC


$$h_1^* y_1 + h_2^* y_2 = (|h_1|^2 + |h_2|^2)x + (h_1^* + h_2^*)n$$

$$\begin{aligned} \text{SNR}_{\text{MRC}} &= \frac{E[((|h_1|^2 + |h_2|^2)X)^2]}{(h_1^* + h_2^*)^2 n^2} & \text{SNR}_{\text{single}} &= \frac{E[|h_1|^2 X^2]}{n^2} \\ &= \frac{(|h_1|^2 + |h_2|^2)^2 E[X^2]}{(|h_1|^2 + |h_2|^2) \sigma^2} & &= \frac{|h_1|^2 E[X^2]}{\sigma^2} \\ &= \frac{(|h_1|^2 + |h_2|^2) E[X^2]}{\sigma^2} \end{aligned}$$

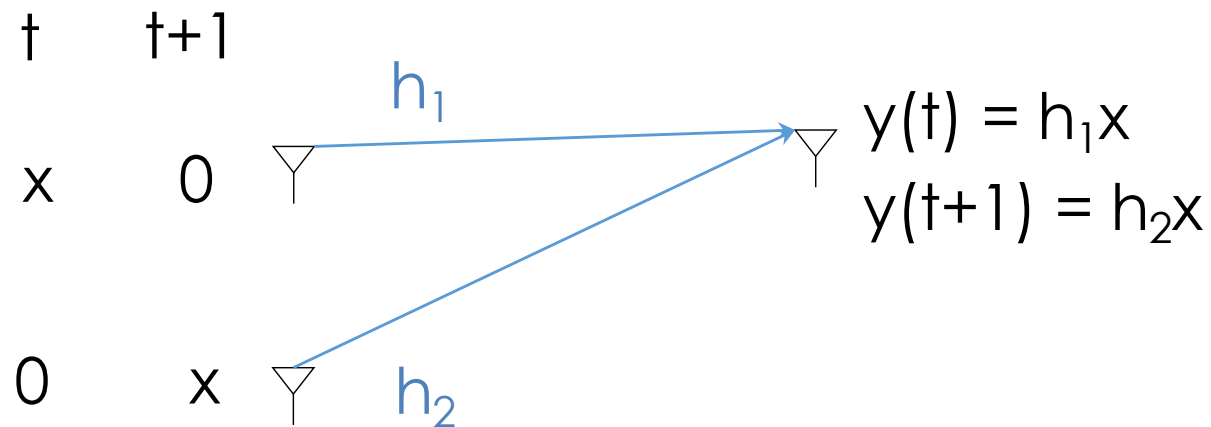
- $\text{gain} = \frac{|h_1|^2 + |h_2|^2}{|h_1|^2}$
- $\sim 2\times$ gain if $|h_1| \sim |h_2|$

Transmit Diversity



- Signals go through two diverse paths
- Theoretical SNR gain: similar to receive diversity
- How to extract the SNR gain?
 - Simply transmit from two antennas simultaneously? 
 - No! Again, h_1 and h_2 might be destructive

Transmit Diversity: Repetitive Code



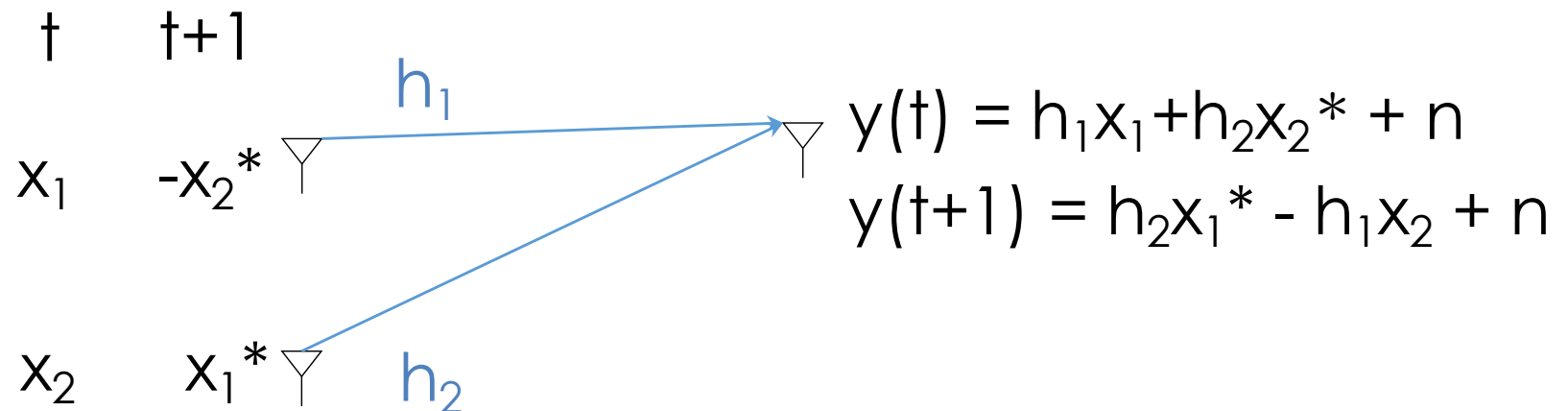
- Deliver a symbol twice in two consecutive time slots
- Repetitive code

$$\mathbf{X} = \begin{matrix} & \xrightarrow{\text{time}} \\ \text{space} \downarrow & \begin{pmatrix} x & 0 \\ 0 & x \end{pmatrix} \end{matrix}$$

- Diversity: 2
- Data rate: $1/2$ symbols/s/Hz

- Decode and extract the diversity gain via MRC
- Improve SNR, but reduce the data rate!!

Transmit Diversity: Alamouti Code



- Deliver 2 symbols in two consecutive time slots, but switch the antennas
- Alamouti code (space-time block code)

$$\mathbf{x} = \begin{matrix} \text{space} \downarrow & \begin{pmatrix} x_1 & -x_2 \\ x_2^* & x_1^* \end{pmatrix} & \xrightarrow{\text{time}} \end{matrix}$$

- Diversity: 2
- Data rate: 1 symbols/s/Hz

- Improve SNR, while, meanwhile, maintain the data rate

Transmit Diversity: Alamouti Code

- Decoding

$$\begin{aligned} y(t) &= h_1 x_1 + h_2 x_2^* + n \\ y(t+1) &= h_2 x_1^* - h_1 x_2 + n \end{aligned}$$

$$h_1^* y(t) = |h_1|^2 x_1 + h_1^* h_2 x_2^* + h_1^* n$$

$$y^*(t+1) = h_2^* x_1 - h_1^* x_2^* + n^*$$

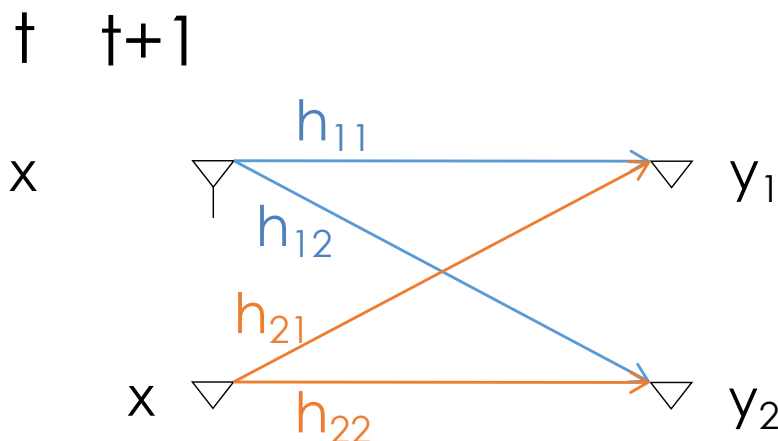
$$h_2 y^*(t+1) = |h_2|^2 x_1 - h_1^* h_2 x_2^* + h_2 n^*$$

$$\Rightarrow h_1^* y(t) + h_2 y^*(t+1) = (|h_1|^2 + |h_2|^2) x_1 + h_1^* n + h_2 n^*$$

- Achievable SNR

$$\begin{aligned} & \frac{(|h_1|^2 + |h_2|^2)^2 E[X^2]}{(h_1^* n + h_2 n^*)} \\ &= \frac{(|h_1|^2 + |h_2|^2)^2 E[X^2]}{(|h_1|^2 + |h_2|^2) \sigma^2} = \frac{(|h_1|^2 + |h_2|^2) E[X^2]}{\sigma^2} \end{aligned}$$

Multiplexing-Diversity Tradeoff

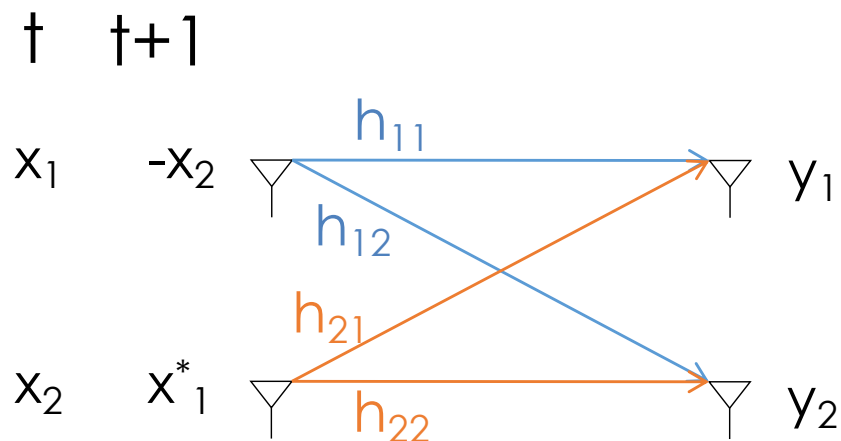


Repetitive scheme

$$\mathbf{X} = \begin{pmatrix} x & 0 \\ 0 & x \end{pmatrix}$$

Diversity: 4

Data rate: 1/2 sym/s/Hz



Alamouti scheme

$$\mathbf{X} = \begin{pmatrix} x_1 & -x_2 \\ x_2^* & x_1^* \end{pmatrix}$$

Diversity: 4

Data rate: 1 sym/s/Hz

But 2x2 MIMO has 2 degrees of freedom