

## MIXED SIGNAL MICROCONTROLLER

#### **FEATURES**

- Low Supply Voltage Range:
   3.6 V Down to 1.8 V
- Ultralow Power Consumption
  - Active Mode (AM):
     All System Clocks Active
     230 μA/MHz at 8 MHz, 3.0 V, Flash Program
     Execution (Typical)
     110 μA/MHz at 8 MHz, 3.0 V, RAM Program
     Execution (Typical)
  - Standby Mode (LPM3):
     Real-Time Clock With Crystal , Watchdog, and Supply Supervisor Operational, Full RAM Retention, Fast Wake-Up:
     1.7 μA at 2.2 V, 2.1 μA at 3.0 V (Typical)
     Low-Power Oscillator (VLO), General-Purpose Counter, Watchdog, and Supply Supervisor Operational, Full RAM Retention, Fast Wake-Up:
     1.2 μA at 3.0 V (Typical)
  - Off Mode (LPM4):
     Full RAM Retention, Supply Supervisor
     Operational, Fast Wake-Up:
     1.2 μA at 3.0 V (Typical)
  - Shutdown Mode (LPM4.5):
     0.1 μA at 3.0 V (Typical)
- Wake-Up From Standby Mode in 3.5 μs (Typical)
- 16-Bit RISC Architecture
  - Extended Memory
  - Up to 25-MHz System Clock
- Flexible Power Management System
  - Fully Integrated LDO With Programmable Regulated Core Supply Voltage
  - Supply Voltage Supervision, Monitoring, and Brownout
- Unified Clock System
  - FLL Control Loop for Frequency Stabilization
  - Low-Power/Low-Frequency Internal Clock Source (VLO)

- Low-Frequency Trimmed Internal Reference Source (REFO)
- 32-kHz Crystals
- High-Frequency Crystals up to 32 MHz
- 16-Bit Timer TA0, Timer\_A With Five Capture/Compare Registers
- 16-Bit Timer TA1, Timer\_A With Three Capture/Compare Registers
- 16-Bit Timer TB0, Timer\_B With Seven Capture/Compare Shadow Registers
- Up to Four Universal Serial Communication Interfaces
  - USCI\_A0, USCI\_A1, USCI\_A2, and USCI\_A3
     Each Supporting
    - Enhanced UART supporting Auto-Baudrate Detection
    - IrDA Encoder and Decoder
    - Synchronous SPI
  - USCI\_B0, USCI\_B1, USCI\_B2, and USCI\_B3
     Each Supporting
    - $-I^2C^{TM}$
    - Synchronous SPI
- 12-Bit Analog-to-Digital (A/D) Converter
  - Internal Reference
  - Sample-and-Hold
  - Autoscan Feature
  - 14 External Channels, 2 Internal Channels
- Hardware Multiplier Supporting 32-Bit Operations
- Serial Onboard Programming, No External Programming Voltage Needed
- Three Channel Internal DMA
- Basic Timer With Real-Time Clock Feature
- Family Members are Summarized in Table 1
- For Complete Module Descriptions, See the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208)

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#### DESCRIPTION

The Texas Instruments MSP430 family of ultralow-power microcontrollers consists of several devices featuring different sets of peripherals targeted for various applications. The architecture, combined with extensive low-power modes, is optimized to achieve extended battery life in portable measurement applications. The device features a powerful 16-bit RISC CPU, 16-bit registers, and constant generators that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows wake-up from low-power modes to active mode in 3.5 µs (typical).

The MSP430F543xA and MSP430F541xA series are microcontroller configurations with three 16-bit timers, a high performance 12-bit analog-to-digital (A/D) converter, up to four universal serial communication interfaces (USCI), hardware multiplier, DMA, real-time clock module with alarm capabilities, and up to 87 I/O pins.

Typical applications for this device include analog and digital sensor systems, digital motor control, remote controls, thermostats, digital timers, and hand-held meters.

Family members available are summarized in Table 1.

**Table 1. Family Members** 

					US	USCI			
Device	Flash (KB)	SRAM (KB)	Timer_A <sup>(1)</sup>	Timer_B <sup>(2)</sup>	Channel A: UART, IrDA, SPI	Channel B: SPI, I <sup>2</sup> C	ADC12_A (Ch)	I/O	Package Type
MSP430F5438A	256	16	5, 3	7	4	4	14 ext, 2 int	87	100 PZ, 113 ZQW
MSP430F5437A	256	16	5, 3	7	2	2	14 ext, 2 int	67	80 PN
MSP430F5436A	192	16	5, 3	7	4	4	14 ext, 2 int	87	100 PZ, 113 ZQW
MSP430F5435A	192	16	5, 3	7	2	2	14 ext, 2 int	67	80 PN
MSP430F5419A	128	16	5, 3	7	4	4	14 ext, 2 int	87	100 PZ, 113 ZQW
MSP430F5418A	128	16	5, 3	7	2	2	14 ext, 2 int	67	80 PN

<sup>(1)</sup> Each number in the sequence represents an instantiation of Timer\_A with its associated number of capture compare registers and PWM output generators available. For example, a number sequence of 3, 5 would represent two instantiations of Timer\_A, the first instantiation having 3 and the second instantiation having 5 capture compare registers and PWM output generators, respectively.

Table 2. Ordering Information<sup>(1)</sup>

	PACKAGED DEVICES <sup>(2)</sup>									
T <sub>A</sub>	PLASTIC 100-PIN LQFP (PZ)	PLASTIC 80-PIN LQFP (PN)	PLASTIC 113-BALL BGA (ZQW)							
	MSP430F5438AIPZ	MSP430F5437AIPN	MSP430F5438AIZQW							
-40°C to 85°C	MSP430F5436AIPZ	MSP430F5435AIPN	MSP430F5436AIZQW							
	MSP430F5419AIPZ	MSP430F5418AIPN	MSP430F5419AIZQW							

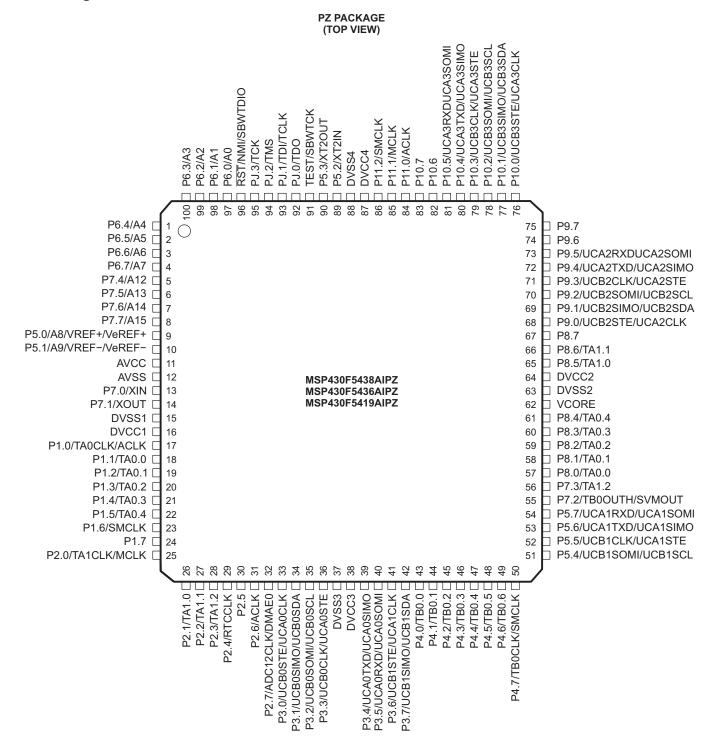
<sup>(1)</sup> For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

<sup>(2)</sup> Each number in the sequence represents an instantiation of Timer\_B with its associated number of capture compare registers and PWM output generators available. For example, a number sequence of 3, 5 would represent two instantiations of Timer\_B, the first instantiation having 3 and the second instantiation having 5 capture compare registers and PWM output generators, respectively.

<sup>(2)</sup> Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.

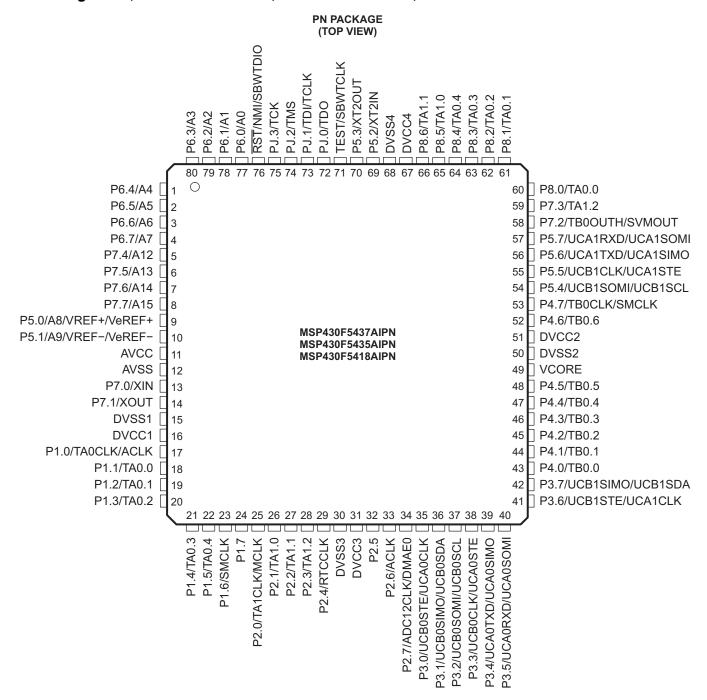


### Pin Designation, MSP430F5438AIPZ, MSP430F5436AIPZ, MSP430F5419AIPZ





## Pin Designation, MSP430F5437AIPN, MSP430F5435AIPN, MSP430F5418AIPN





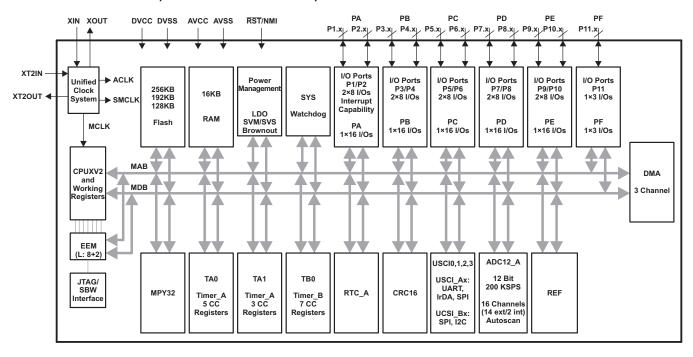
## Pin Designation, MSP430F5438AIZQW, MSP430F5436AIZQW, MSP430F5419AIZQW

## ZQW PACKAGE (TOP VIEW)

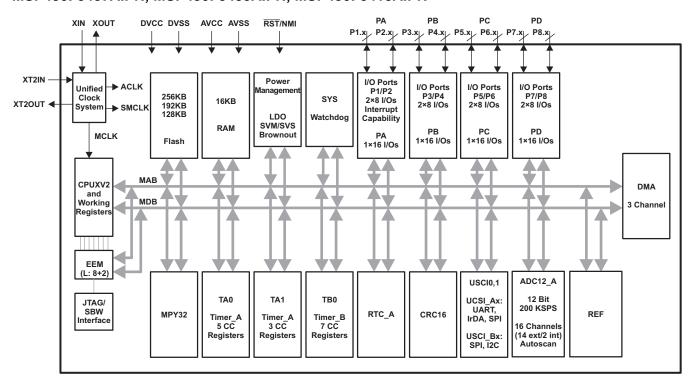
Г	P6.4	P6.2	RST	PJ.1	P5.3	P5.2	P11.2	P11.0	P10.6	P10.4	P10.1	P9.7
	(A1)	(A2)	(A3)	(A4)	(A5)	(A6)	(A7)	(A8)	(A9)	(A10)	(A11)	(A12)
	P6.6	P6.3	P6.1	PJ.3	PJ.0	DVSS4	DVCC4	P10.7	P10.5	P10.3	P9.6	P9.5
	(B1)	(B2)	(B3)	(B4)	(B5)	(B6)	(B7)	(B8)	(B9)	(B10)	(B11)	(B12)
	P7.5	P6.7	_								P9.4	P9.2
	(C1)	(C2)	(C3)								(C11)	(C12)
	P5.0	P7.6		P6.0	PJ.2	TEST	P11.1	P10.2	P10.0		P9.0	P8.7
	(D1)	(D2)		(D4)	(D5)	(D6)	(D7)	(D8)	(D9)		(D11)	(D12)
	P5.1	AVCC		P6.5	_	_	_	_	P9.3		P8.6	DVCC2
	(E1)	(E2)		(E4)	(E5)	(E6)	(E7)	(E8)	(E9)		(E11)	(E12)
	P7.0	AVSS		P7.4	_			_	P9.1		P8.5	DVSS2
	(F1)	(F2)		(F4)	(F5)			(F8)	(F9)		(F11)	(F12)
	P7.1	DVSS1		P7.7	_			_	P8.3		P8.4	VCORE
	(G1)	(G2)		(G4)	(G5)			(G8)	(G9)		(G11)	(G12)
	P1.0	DVCC1		P1.1	_	_	. ~	_	P8.0		P8.1	P8.2
	(H1)	(H2)		(H4)	(H5)	(H6)	(H7)	(H8)	(H9)		(H11)	(H12)
	P1.3	P1.4		P1.2	P2.7	P3.2	P3.5	P4.0	P5.5		P7.2	P7.3
	$(J_1)$	(J2)		(J4)	(J5)	(16)	(J7)	(18)	(19)		(J11)	(J12)
	P1.5	P1.6									P5.6	P5.7
	(K1)	(K2)									(K11)	(K12)
	P1.7	P2.1	P2.3	P2.5	P3.0	P3.3	P3.4	P3.7	P4.2	P4.3	P4.5	P5.4
	(L1)	(L2)	(F3)	(L4)	(L5)	(L6)	(L7)	(F8)	(F8)	(L10)	(L11)	(L12)
1	P2.0	P2.2	P2.4	P2.6	P3.1		DVCC3	P3.6	P4.1	P4.4	P4.6	P4.7
	(M1)	(M2)	(M3)	(M4)	(M5)	(M6)	(M7)	(M8)	(M9)	(M10)	(M11)	(M12)



# Functional Block Diagram MSP430F5438AIPZ, MSP430F5419AIPZ, MSP430F5438AIZQW, MSP430F5436AIZQW, MSP430F5419AIZQW



## Functional Block Diagram MSP430F5437AIPN, MSP430F5418AIPN





## **Table 3. Terminal Functions**

TERMINAL					. Terminal Functions
		I/O <sup>(1)</sup>	DESCRIPTION		
NAME	PZ	PN	ZQW		DEGGINI NON
P6.4/A4	1	1	A1	I/O	General-purpose digital I/O Analog input A4 – ADC
P6.5/A5	2	2	E4	I/O	General-purpose digital I/O Analog input A5 – ADC
P6.6/A6	3	3	B1	I/O	General-purpose digital I/O Analog input A6 – ADC
P6.7/A7	4	4	C2	I/O	General-purpose digital I/O Analog input A7 – ADC
P7.4/A12	5	5	F4	I/O	General-purpose digital I/O Analog input A12 –ADC
P7.5/A13	6	6	C1	I/O	General-purpose digital I/O Analog input A13 – ADC
P7.6/A14	7	7	D2	I/O	General-purpose digital I/O Analog input A14 – ADC
P7.7/A15	8	8	G4	I/O	General-purpose digital I/O Analog input A15 – ADC
P5.0/A8/VREF+/VeREF+	9	9	D1	I/O	General-purpose digital I/O Analog input A8 – ADC Output of reference voltage to the ADC Input for an external reference voltage to the ADC
P5.1/A9/VREF-/VeREF-	10	10	E1	I/O	General-purpose digital I/O Analog input A9 – ADC Negative terminal for the ADC's reference voltage for both sources, the internal reference voltage, or an external applied reference voltage
AVCC	11	11	E2		Analog power supply
AVSS	12	12	F2		Analog ground supply
P7.0/XIN	13	13	F1	I/O	General-purpose digital I/O Input terminal for crystal oscillator XT1
P7.1/XOUT	14	14	G1	I/O	General-purpose digital I/O Output terminal of crystal oscillator XT1
DVSS1	15	15	G2		Digital ground supply
DVCC1	16	16	H2		Digital power supply
P1.0/TA0CLK/ACLK	17	17	H1	I/O	General-purpose digital I/O with port interrupt TA0 clock signal TACLK input ACLK output (divided by 1, 2, 4, 8, 16, or 32)
P1.1/TA0.0	18	18	H4	I/O	General-purpose digital I/O with port interrupt TA0 CCR0 capture: CCI0A input, compare: Out0 output BSL transmit output
P1.2/TA0.1	19	19	J4	I/O	General-purpose digital I/O with port interrupt TA0 CCR1 capture: CCl1A input, compare: Out1 output BSL receive input
P1.3/TA0.2	20	20	J1	I/O	General-purpose digital I/O with port interrupt TA0 CCR2 capture: CCI2A input, compare: Out2 output
P1.4/TA0.3	21	21	J2	I/O	General-purpose digital I/O with port interrupt TA0 CCR3 capture: CCl3A input compare: Out3 output
P1.5/TA0.4	22	22	K1	I/O	General-purpose digital I/O with port interrupt TA0 CCR4 capture: CCI4A input, compare: Out4 output
P1.6/SMCLK	23	23	K2	I/O	General-purpose digital I/O with port interrupt SMCLK output
P1.7	24	24	L1	I/O	General-purpose digital I/O with port interrupt
P2.0/TA1CLK/MCLK	25	25	M1	I/O	General-purpose digital I/O with port interrupt TA1 clock signal TA1CLK input MCLK output



TERMINAL			,		
NO.		I/O <sup>(1)</sup>	DESCRIPTION		
NAME	PZ	PN	ZQW		
P2.1/TA1.0	26	26	L2	I/O	General-purpose digital I/O with port interrupt TA1 CCR0 capture: CCl0A input, compare: Out0 output
P2.2/TA1.1	27	27	M2	I/O	General-purpose digital I/O with port interrupt TA1 CCR1 capture: CCl1A input, compare: Out1 output
P2.3/TA1.2	28	28	L3	I/O	General-purpose digital I/O with port interrupt TA1 CCR2 capture: CCl2A input, compare: Out2 output
P2.4/RTCCLK	29	29	М3	I/O	General-purpose digital I/O with port interrupt RTCCLK output
P2.5	30	32	L4	I/O	General-purpose digital I/O with port interrupt
P2.6/ACLK	31	33	M4	I/O	General-purpose digital I/O with port interrupt ACLK output (divided by 1, 2, 4, 8, 16, or 32)
P2.7/ADC12CLK/DMAE0	32	34	J5	I/O	General-purpose digital I/O with port interrupt Conversion clock output ADC DMA external trigger input
P3.0/UCB0STE/UCA0CLK	33	35	L5	I/O	General-purpose digital I/O Slave transmit enable – USCI_B0 SPI mode Clock signal input – USCI_A0 SPI slave mode Clock signal output – USCI_A0 SPI master mode
P3.1/UCB0SIMO/UCB0SDA	34	36	M5	I/O	General-purpose digital I/O Slave in, master out – USCI_B0 SPI mode I2C data – USCI_B0 I2C mode
P3.2/UCB0SOMI/UCB0SCL	35	37	J6	I/O	General-purpose digital I/O Slave out, master in – USCI_B0 SPI mode I2C clock – USCI_B0 I2C mode
P3.3/UCB0CLK/UCA0STE	36	38	L6	I/O	General-purpose digital I/O Clock signal input – USCI_B0 SPI slave mode Clock signal output – USCI_B0 SPI master mode Slave transmit enable – USCI_A0 SPI mode
DVSS3	37	30	M6		Digital ground supply
DVCC3	38	31	M7		Digital power supply
P3.4/UCA0TXD/UCA0SIMO	39	39	L7	I/O	General-purpose digital I/O Transmit data – USCI_A0 UART mode Slave in, master out – USCI_A0 SPI mode
P3.5/UCA0RXD/UCA0SOMI	40	40	J7	I/O	General-purpose digital I/O Receive data – USCI_A0 UART mode Slave out, master in – USCI_A0 SPI mode
P3.6/UCB1STE/UCA1CLK	41	41	M8	I/O	General-purpose digital I/O Slave transmit enable – USCI_B1 SPI mode Clock signal input – USCI_A1 SPI slave mode Clock signal output – USCI_A1 SPI master mode
P3.7/UCB1SIMO/UCB1SDA	42	42	L8	I/O	General-purpose digital I/O Slave in, master out – USCI_B1 SPI mode I2C data – USCI_B1 I2C mode
P4.0/TB0.0	43	43	J8	I/O	General-purpose digital I/O TB0 capture CCR0: CCI0A/CCI0B input, compare: Out0 output
P4.1/TB0.1	44	44	M9	I/O	General-purpose digital I/O TB0 capture CCR1: CCI1A/CCI1B input, compare: Out1 output
P4.2/TB0.2	45	45	L9	I/O	General-purpose digital I/O TB0 capture CCR2: CCI2A/CCI2B input, compare: Out2 output
P4.3/TB0.3	46	46	L10	I/O	General-purpose digital I/O TB0 capture CCR3: CCI3A/CCI3B input, compare: Out3 output
P4.4/TB0.4	47	47	M10	I/O	General-purpose digital I/O TB0 capture CCR4: CCI4A/CCI4B input, compare: Out4 output
P4.5/TB0.5	48	48	L11	I/O	General-purpose digital I/O TB0 capture CCR5: CCI5A/CCI5B input, compare: Out5 output



TERMINAL						
NAME	NO.			I/O <sup>(1)</sup>	DESCRIPTION	
NAME	PZ	PN	ZQW			
P4.6/TB0.6	49	52	M11	I/O	General-purpose digital I/O TB0 capture CCR6: CCI6A/CCI6B input, compare: Out6 output	
P4.7/TB0CLK/SMCLK	50	53	M12	I/O	General-purpose digital I/O TB0 clock input SMCLK output	
P5.4/UCB1SOMI/UCB1SCL	51	54	L12	I/O	General-purpose digital I/O Slave out, master in – USCI_B1 SPI mode I2C clock – USCI_B1 I2C mode	
P5.5/UCB1CLK/UCA1STE	52	55	J9	I/O	General-purpose digital I/O Clock signal input – USCI_B1 SPI slave mode Clock signal output – USCI_B1 SPI master mode Slave transmit enable – USCI_A1 SPI mode	
P5.6/UCA1TXD/UCA1SIMO	53	56	K11	I/O	General-purpose digital I/O Transmit data – USCI_A1 UART mode Slave in, master out – USCI_A1 SPI mode	
P5.7/UCA1RXD/UCA1SOMI	54	57	K12	I/O	General-purpose digital I/O Receive data – USCI_A1 UART mode Slave out, master in – USCI_A1 SPI mode	
P7.2/TB0OUTH/SVMOUT	55	58	J11	I/O	General-purpose digital I/O Switch all PWM outputs high impedance – Timer TB0 SVM output	
P7.3/TA1.2	56	59	J12	I/O	General-purpose digital I/O TA1 CCR2 capture: CCl2B input, compare: Out2 output	
P8.0/TA0.0	57	60	Н9	I/O	General-purpose digital I/O TA0 CCR0 capture: CCl0B input, compare: Out0 output	
P8.1/TA0.1	58	61	H11	I/O	General-purpose digital I/O TA0 CCR1 capture: CCl1B input, compare: Out1 output	
P8.2/TA0.2	59	62	H12	I/O	General-purpose digital I/O TA0 CCR2 capture: CCl2B input, compare: Out2 output	
P8.3/TA0.3	60	63	G9	I/O	General-purpose digital I/O TA0 CCR3 capture: CCl3B input, compare: Out3 output	
P8.4/TA0.4	61	64	G11	I/O	General-purpose digital I/O TA0 CCR4 capture: CCI4B input, compare: Out4 output	
VCORE (2)	62	49	G12		Regulated core power supply output (internal use only, no external current loading)	
DVSS2	63	50	F12		Digital ground supply	
DVCC2	64	51	E12		Digital power supply	
P8.5/TA1.0	65	65	F11	I/O	General-purpose digital I/O TA1 CCR0 capture: CCl0B input, compare: Out0 output	
P8.6/TA1.1	66	66	E11	I/O	General-purpose digital I/O TA1 CCR1 capture: CCl1B input, compare: Out1 output	
P8.7	67	N/A	D12	I/O	General-purpose digital I/O	
P9.0/UCB2STE/UCA2CLK	68	N/A	D11	I/O	General-purpose digital I/O Slave transmit enable – USCI_B2 SPI mode Clock signal input – USCI_A2 SPI slave mode Clock signal output – USCI_A2 SPI master mode	
P9.1/UCB2SIMO/UCB2SDA	69	N/A	F9	I/O	General-purpose digital I/O Slave in, master out – USCI_B2 SPI mode I2C data – USCI_B2 I2C mode	
P9.2/UCB2SOMI/UCB2SCL	70	N/A	C12	I/O	General-purpose digital I/O Slave out, master in – USCI_B2 SPI mode I2C clock – USCI_B2 I2C mode	

<sup>(2)</sup> VCORE is for internal use only. No external current loading is possible. VCORE should only be connected to the recommended capacitor value, C<sub>VCORE</sub>.



TERMINAL					,
		I/O <sup>(1)</sup>	DESCRIPTION		
NAME	PZ	PN	zqw		DECOM HON
P9.3/UCB2CLK/UCA2STE	71	N/A	E9	I/O	General-purpose digital I/O Clock signal input – USCI_B2 SPI slave mode Clock signal output – USCI_B2 SPI master mode Slave transmit enable – USCI_A2 SPI mode
P9.4/UCA2TXD/UCA2SIMO	72	N/A	C11	I/O	General-purpose digital I/O Transmit data – USCI_A2 UART mode Slave in, master out – USCI_A2 SPI mode
P9.5/UCA2RXD/UCA2SOMI	73	N/A	B12	I/O	General-purpose digital I/O Receive data – USCI_A2 UART mode Slave out, master in – USCI_A2 SPI mode
P9.6	74	N/A	B11	I/O	General-purpose digital I/O
P9.7	75	N/A	A12	I/O	General-purpose digital I/O
P10.0/UCB3STE/UCA3CLK	76	N/A	D9	I/O	General-purpose digital I/O Slave transmit enable – USCI_B3 SPI mode Clock signal input – USCI_A3 SPI slave mode Clock signal output – USCI_A3 SPI master mode
P10.1/UCB3SIMO/UCB3SDA	77	N/A	A11	I/O	General-purpose digital I/O Slave in, master out – USCI_B3 SPI mode I2C data – USCI_B3 I2C mode
P10.2/UCB3SOMI/UCB3SCL	78	N/A	D8	I/O	General-purpose digital I/O Slave out, master in – USCI_B3 SPI mode I2C clock – USCI_B3 I2C mode
P10.3/UCB3CLK/UCA3STE	79	N/A	B10	I/O	General-purpose digital I/O Clock signal input – USCI_B3 SPI slave mode Clock signal output – USCI_B3 SPI master mode Slave transmit enable – USCI_A3 SPI mode
P10.4/UCA3TXD/UCA3SIMO	80	N/A	A10	I/O	General-purpose digital I/O Transmit data – USCI_A3 UART mode Slave in, master out – USCI_A3 SPI mode
P10.5/UCA3RXD/UCA3SOMI	81	N/A	В9	I/O	General-purpose digital I/O Receive data – USCI_A3 UART mode Slave out, master in – USCI_A3 SPI mode
P10.6	82	N/A	A9	I/O	General-purpose digital I/O
P10.7	83	N/A	B8	I/O	General-purpose digital I/O
P11.0/ACLK	84	N/A	A8	I/O	General-purpose digital I/O ACLK output (divided by 1, 2, 4, 8, 16, or 32)
P11.1/MCLK	85	N/A	D7	I/O	General-purpose digital I/O MCLK output
P11.2/SMCLK	86	N/A	A7	I/O	General-purpose digital I/O SMCLK output
DVCC4	87	67	B7		Digital power supply
DVSS4	88	68	B6		Digital ground supply
P5.2/XT2IN	89	69	A6	I/O	General-purpose digital I/O Input terminal for crystal oscillator XT2
P5.3/XT2OUT	90	70	A5	I/O	General-purpose digital I/O Output terminal of crystal oscillator XT2
TEST/SBWTCK <sup>(3)</sup>	91	71	D6	I	Test mode pin – Selects four wire JTAG operation. Spy-Bi-Wire input clock when Spy-Bi-Wire operation activated
PJ.0/TDO <sup>(4)</sup>	92	72	B5	I/O	General-purpose digital I/O JTAG test data output port
PJ.1/TDI/TCLK <sup>(4)</sup>	93	73	A4	I/O	General-purpose digital I/O JTAG test data input or test clock input

See Bootstrap Loader (BSL) and JTAG Operation for use with BSL and JTAG functions, respectively. See JTAG Operation for use with JTAG function.

<sup>(4)</sup> 



TERMINAL						
NAME		NO.		I/O <sup>(1)</sup>	DESCRIPTION	
NAME	PZ	PN	ZQW			
PJ.2/TMS <sup>(4)</sup>	94	74	D5	I/O	General-purpose digital I/O JTAG test mode select	
PJ.3/TCK <sup>(4)</sup>	95	75	B4	I/O	General-purpose digital I/O JTAG test clock	
RST/NMI/SBWTDIO <sup>(3)</sup>	96	76	A3	I/O	Reset input active low Non-maskable interrupt input Spy-Bi-Wire data input/output when Spy-Bi-Wire operation activated.	
P6.0/A0	97	77	D4	I/O	General-purpose digital I/O Analog input A0 – ADC	
P6.1/A1	98	78	В3	I/O	General-purpose digital I/O Analog input A1 – ADC	
P6.2/A2	99	79	A2	I/O	General-purpose digital I/O Analog input A2 – ADC	
P6.3/A3	100	80	B2	I/O	General-purpose digital I/O Analog input A3 – ADC	
Reserved	N/A	N/A	(5)			

<sup>(5)</sup> C3, E5, E6, E7, E8, F5, F8, G5, G8, H5, H6, H7, H8 are reserved and should be connected to ground.



#### SHORT-FORM DESCRIPTION

#### **CPU (Link to User's Guide)**

The MSP430 CPU has a 16-bit RISC architecture that is highly transparent to the application. All operations, other than program-flow instructions, are performed as register operations in conjunction with seven addressing modes for source operand and four addressing modes for destination operand.

The CPU is integrated with 16 registers that provide reduced instruction execution time. The register-to-register operation execution time is one cycle of the CPU clock.

Four of the registers, R0 to R3, are dedicated as program counter, stack pointer, status register, and constant generator, respectively. The remaining registers are general-purpose registers.

Peripherals are connected to the CPU using data, address, and control buses, and can be handled with all instructions.

The instruction set consists of the original 51 instructions with three formats and seven address modes and additional instructions for the expanded address range. Each instruction can operate on word and byte data.

Program Counter	PC/R0
Stack Pointer	SP/R1
Status Register	SR/CG1/R2
Constant Generator	CG2/R3
General-Purpose Register	R4
General-Purpose Register	R5
General-Purpose Register	R6
General-Purpose Register	R7
General-Purpose Register	R8
General-Purpose Register	R9
General-Purpose Register	R10
General-Purpose Register	R11
General-Purpose Register	R12
General-Purpose Register	R13
General-Purpose Register	R14
General-Purpose Register	R15



#### **Operating Modes**

The MSP430 has one active mode and six software selectable low-power modes of operation. An interrupt event can wake up the device from any of the low-power modes, service the request, and restore back to the low-power mode on return from the interrupt program.

The following seven operating modes can be configured by software:

- Active mode (AM)
  - All clocks are active
- Low-power mode 0 (LPM0)
  - CPU is disabled
  - ACLK and SMCLK remain active, MCLK is disabled
  - FLL loop control remains active
- Low-power mode 1 (LPM1)
  - CPU is disabled
  - FLL loop control is disabled
  - ACLK and SMCLK remain active, MCLK is disabled
- Low-power mode 2 (LPM2)
  - CPU is disabled
  - MCLK and FLL loop control and DCOCLK are disabled
  - DCO's dc-generator remains enabled
  - ACLK remains active
- Low-power mode 3 (LPM3)
  - CPU is disabled
  - MCLK, FLL loop control, and DCOCLK are disabled
  - DCO's dc generator is disabled
  - ACLK remains active
- Low-power mode 4 (LPM4)
  - CPU is disabled
  - ACLK is disabled
  - MCLK, FLL loop control, and DCOCLK are disabled
  - DCO's dc generator is disabled
  - Crystal oscillator is stopped
  - Complete data retention
- Low-power mode 4.5 (LPM4.5)
  - Internal regulator disabled
  - No data retention
  - Wakeup from RST, digital I/O



#### **Interrupt Vector Addresses**

The interrupt vectors and the power-up start address are located in the address range 0FFFFh to 0FF80h. The vector contains the 16-bit address of the appropriate interrupt-handler instruction sequence.

Table 4. Interrupt Sources, Flags, and Vectors

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY
System Reset Power-Up External Reset Watchdog Timeout, Password Violation Flash Memory Password Violation PMM Password Violation	WDTIFG, KEYV (SYSRSTIV) <sup>(1) (2)</sup>	Reset	OFFFEh	63, highest
System NMI PMM Vacant Memory Access JTAG Mailbox	SVMLIFG, SVMHIFG, DLYLIFG, DLYHIFG, VLRLIFG, VLRHIFG, VMAIFG, JMBNIFG, JMBOUTIFG (SYSSNIV) <sup>(1)</sup>	(Non)maskable	0FFFCh	62
User NMI NMI Oscillator Fault Flash Memory Access Violation	NMIIFG, OFIFG, ACCVIFG (SYSUNIV) <sup>(1)</sup> (2)	(Non)maskable	0FFFAh	61
TB0	TBCCR0 CCIFG0 (3)	Maskable	0FFF8h	60
TB0	TBCCR1 CCIFG1 to TBCCR6 CCIFG6, TBIFG (TBIV) <sup>(1) (3)</sup>	Maskable	0FFF6h	59
Watchdog Timer_A Interval Timer Mode	WDTIFG	Maskable	0FFF4h	58
USCI_A0 Receive and Transmit	UCA0RXIFG, UCA0TXIFG (UCA0IV) (1) (3)	Maskable	0FFF2h	57
USCI_B0 Receive and Transmit	UCB0RXIFG, UCB0TXIFG (UCB0IV) (1) (3)	Maskable	0FFF0h	56
ADC12_A	ADC12IFG0 to ADC12IFG15 (ADC12IV) <sup>(1) (3)</sup>	Maskable	0FFEEh	55
TA0	TA0CCR0 CCIFG0 <sup>(3)</sup>	Maskable	0FFECh	54
TA0	TA0CCR1 CCIFG1 to TA0CCR4 CCIFG4, TA0IFG (TA0IV) <sup>(1)</sup> (3)	Maskable	0FFEAh	53
USCI_A2 Receive and Transmit	UCA2RXIFG, UCA2TXIFG (UCA2IV) (1) (3)	Maskable	0FFE8h	52
USCI_B2 Receive and Transmit	UCB2RXIFG, UCB2TXIFG (UCB2IV) (1) (3)	Maskable	0FFE6h	51
DMA	DMA0IFG, DMA1IFG, DMA2IFG (DMAIV)(1) (3)	Maskable	0FFE4h	50
TA1	TA1CCR0 CCIFG0 <sup>(3)</sup>	Maskable	0FFE2h	49
TA1	TA1CCR1 CCIFG1 to TA1CCR2 CCIFG2, TA1IFG (TA1IV) <sup>(1)</sup> (3)	Maskable	0FFE0h	48
I/O Port P1	P1IFG.0 to P1IFG.7 (P1IV) <sup>(1) (3)</sup>	Maskable	0FFDEh	47
USCI_A1 Receive and Transmit	UCA1RXIFG, UCA1TXIFG (UCA1IV) (1) (3)	Maskable	0FFDCh	46
USCI_B1 Receive and Transmit	UCB1RXIFG, UCB1TXIFG (UCB1IV)(1) (3)	Maskable	0FFDAh	45
USCI_A3 Receive and Transmit	UCA3RXIFG, UCA3TXIFG (UCA3IV) <sup>(1) (3)</sup>	Maskable	0FFD8h	44
USCI_B3 Receive and Transmit	UCB3RXIFG, UCB3TXIFG (UCB3IV) (1) (3)	Maskable	0FFD6h	43
I/O Port P2	P2IFG.0 to P2IFG.7 (P2IV) <sup>(1) (3)</sup>	Maskable	0FFD4h	42
RTC_A	RTCRDYIFG, RTCTEVIFG, RTCAIFG, RT0PSIFG, RT1PSIFG (RTCIV) <sup>(1) (3)</sup>	Maskable	0FFD2h	41
			0FFD0h	40
Reserved	Reserved <sup>(4)</sup>		:	:
			0FF80h	0, lowest

Multiple source flags

<sup>2)</sup> A reset is generated if the CPU tries to fetch instructions from within peripheral space or vacant memory space.

<sup>(</sup>Non)maskable: the individual interrupt-enable bit can disable an interrupt event, but the general-interrupt enable cannot disable it.

<sup>(3)</sup> Interrupt flags are located in the module.

<sup>(4)</sup> Reserved interrupt vectors at addresses are not used in this device and can be used for regular program code if necessary. To maintain compatibility with other devices, it is recommended to reserve these locations.



## **Memory Organization**

		MSP430F5419A MSP430F5418A	MSP430F5436A MSP430F5435A	MSP430F5438A MSP430F5437A
Memory (flash) Main: interrupt vector Main: code memory	Total Size Flash Flash	128 KB 00FFFFh-00FF80h 025BFFh-005C00h	192 KB 00FFFFh-00FF80h 035BFFh-005C00h	256 KB 00FFFFh-00FF80h 045BFFh-005C00h
	Bank D	N/A	23 KB 035BFFh-030000h	64 KB 03FFFFh-030000h
	Bank C	23 KB 025BFFh-020000h	64 KB 02FFFFh-020000h	64 KB 02FFFFh-020000h
Main: code memory	Bank B	64 KB 01FFFFh-010000h	64 KB 01FFFFh-010000h	64 KB 01FFFFh-010000h
	Bank A	41 KB 00FFFFh-005C00h	41 KB 00FFFFh-005C00h	64 KB 045BFFh-040000h 00FFFFh-005C00h
	Size	16 KB	16 KB	16 KB
	Sector 3	4 KB 005BFFh-004C00h	4 KB 005BFFh-004C00h	4 KB 005BFFh-004C00h
RAM	Sector 2	4 KB 004BFFh–003C00h	4 KB 004BFFh-003C00h	4 KB 004BFFh-003C00h
	Sector 1	4 KB 003BFFh-002C00h	4 KB 003BFFh-002C00h	4 KB 003BFFh-002C00h
	Sector 0	4 KB 002BFFh-001C00h	4 KB 002BFFh-001C00h	4 KB 002BFFh-001C00h
	Info A	128 B 0019FFh–001980h	128 B 0019FFh-001980h	128 B 0019FFh–001980h
Information memory	Info B	128 B 00197Fh–001900h	128 B 00197Fh–001900h	128 B 00197Fh–001900h
(flash)	Info C	128 B 0018FFh–001880h	128 B 0018FFh–001880h	128 B 0018FFh–001880h
	Info D	128 B 00187Fh–001800h	128 B 00187Fh–001800h	128 B 00187Fh–001800h
	BSL 3	512 B 0017FFh–001600h	512 B 0017FFh-001600h	512 B 0017FFh–001600h
Bootstrap loader (BSL)	BSL 2	512 B 0015FFh–001400h	512 B 0015FFh-001400h	512 B 0015FFh–001400h
memory (Flash)	BSL 1	512 B 0013FFh–001200h	512 B 0013FFh-001200h	512 B 0013FFh–001200h
	BSL 0	512 B 0011FFh–001000h	512 B 0011FFh–001000h	512 B 0011FFh–001000h
Peripherals	Size	4KB 000FFFh–000000h	4KB 000FFFh-000000h	4KB 000FFFh-000000h



#### **Bootstrap Loader (BSL)**

The BSL enables users to program the flash memory or RAM using a UART serial interface. Access to the device memory via the BSL is protected by an user-defined password. <u>Usage</u> of the BSL requires four pins as shown in Table 5. BSL entry requires a specific entry sequence on the RST/NMI/SBWTDIO and TEST/SBWTCK pins. For complete description of the features of the BSL and its implementation, see the *MSP430 Memory Programming via the Bootstrap Loader User's Guide* (SLAU319).

Table 5. BSL Pin Requirements and Functions

DEVICE SIGNAL	BSL FUNCTION
RST/NMI/SBWTDIO	Entry sequence signal
TEST/SBWTCK	Entry sequence signal
P1.1	Data transmit
P1.2	Data receive
VCC	Power supply
VSS	Ground supply

#### **JTAG Operation**

#### JTAG Standard Interface

The MSP430 family supports the standard JTAG interface which requires four signals for sending and receiving data. The JTAG signals are shared with general-purpose I/O. The TEST/SBWTCK pin is used to enable the JTAG signals. In addition to these signals, the RST/NMI/SBWTDIO is required to interface with MSP430 development tools and device programmers. The JTAG pin requirements are shown in Table 6. For further details on interfacing to development tools and device programmers, see the MSP430 Hardware Tools User's Guide (SLAU278). For complete description of the features of the JTAG interface and its implementation, see the MSP430 Memory Programming via the JTAG Interface User's Guide (SLAU320).

Table 6. JTAG Pin Requirements and Functions

DEVICE SIGNAL	DIRECTION	FUNCTION	
PJ.3/TCK	IN	JTAG clock input	
PJ.2/TMS	IN	JTAG state control	
PJ.1/TDI/TCLK	IN	JTAG data input/TCLK input	
PJ.0/TDO	OUT	JTAG data output	
TEST/SBWTCK	IN	Enable JTAG pins	
RST/NMI/SBWTDIO	IN	External reset	
VCC		Power supply	
VSS		Ground supply	

#### Spy-Bi-Wire Interface

In addition to the standard JTAG interface, the MSP430 family supports the two wire Spy-Bi-Wire interface. Spy-Bi-Wire can be used to interface with MSP430 development tools and device programmers. The Spy-Bi-Wire interface pin requirements are shown in Table 7. For further details on interfacing to development tools and device programmers, see the MSP430 Hardware Tools User's Guide (SLAU278). For the description of the Spy-Bi-Wire interface and its implementation, see the MSP430 Memory Programming via the JTAG Interface User's Guide (SLAU320).

Table 7. Spy-Bi-Wire Pin Requirements and Functions

DEVICE SIGNAL	DIRECTION	FUNCTION
TEST/SBWTCK	IN	Spy-Bi-Wire clock input
RST/NMI/SBWTDIO	IN, OUT	Spy-Bi-Wire data input/output
VCC		Power supply
VSS		Ground supply



#### Flash Memory (Link to User's Guide)

The flash memory can be programmed via the JTAG port, Spy-Bi-Wire (SBW), the BSL, or in-system by the CPU. The CPU can perform single-byte, single-word, and long-word writes to the flash memory. Features of the flash memory include:

- Flash memory has n segments of main memory and four segments of information memory (A to D) of 128 bytes each. Each segment in main memory is 512 bytes in size.
- Segments 0 to n may be erased in one step, or each segment may be individually erased.
- Segments A to D can be erased individually. Segments A to D are also called *information memory*.
- Segment A can be locked separately.

#### RAM Memory (Link to User's Guide)

The RAM memory is made up of n sectors. Each sector can be completely powered down to save leakage, however all data is lost. Features of the RAM memory include:

- RAM memory has n sectors. The size of a sector can be found in Memory Organization.
- Each sector 0 to n can be complete disabled; however, data retention is lost.
- Each sector 0 to n automatically enters low-power retention mode when possible.
- For devices that contain USB memory, the USB memory can be used as normal RAM if USB is not required.



#### **Peripherals**

Peripherals are connected to the CPU through data, address, and control buses and can be handled using all instructions. For complete module descriptions, see the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208).

#### Digital I/O (Link to User's Guide)

There are up to ten 8-bit I/O ports implemented: For 100-pin options, P1 through P10 are complete. P11 contains three individual I/O ports. For 80-pin options, P1 through P7 are complete. P8 contains seven individual I/O ports. P9 through P11 do not exist. Port PJ contains four individual I/O ports, common to all devices.

- All individual I/O bits are independently programmable.
- Any combination of input, output, and interrupt conditions is possible.
- Pullup or pulldown on all ports is programmable.
- Drive strength on all ports is programmable.
- Edge-selectable interrupt and LPM4.5 wakeup input capability is available for all bits of ports P1 and P2.
- · Read/write access to port-control registers is supported by all instructions.
- Ports can be accessed byte-wise (P1 through P11) or word-wise in pairs (PA through PF).

#### Oscillator and System Clock (Link to User's Guide)

The clock system in the MSP430x5xx family of devices is supported by the Unified Clock System (UCS) module that includes support for a 32-kHz watch crystal oscillator (XT1 LF mode), an internal very-low-power low-frequency oscillator (VLO), an internal trimmed low-frequency oscillator (REFO), an integrated internal digitally controlled oscillator (DCO), and a high-frequency crystal oscillator (XT1 HF mode or XT2). The UCS module is designed to meet the requirements of both low system cost and low power consumption. The UCS module features digital frequency locked loop (FLL) hardware that, in conjunction with a digital modulator, stabilizes the DCO frequency to a programmable multiple of the selected FLL reference frequency. The internal DCO provides a fast turn-on clock source and stabilizes in less than 5 µs. The UCS module provides the following clock signals:

- Auxiliary clock (ACLK), sourced from a 32-kHz watch crystal, a high-frequency crystal, the internal low-frequency oscillator (VLO), the trimmed low-frequency oscillator (REFO), or the internal digitally controlled oscillator DCO.
- Main clock (MCLK), the system clock used by the CPU. MCLK can be sourced by same sources made available to ACLK.
- Sub-Main clock (SMCLK), the subsystem clock used by the peripheral modules. SMCLK can be sourced by same sources made available to ACLK.
- ACLK/n, the buffered output of ACLK, ACLK/2, ACLK/4, ACLK/8, ACLK/16, ACLK/32.

#### Power Management Module (PMM) (Link to User's Guide)

The PMM includes an integrated voltage regulator that supplies the core voltage to the device and contains programmable output levels to provide for power optimization. The PMM also includes supply voltage supervisor (SVS) and supply voltage monitoring (SVM) circuitry, as well as brownout protection. The brownout circuit is implemented to provide the proper internal reset signal to the device during power-on and power-off. The SVS/SVM circuitry detects if the supply voltage drops below a user-selectable level and supports both supply voltage supervision (the device is automatically reset) and supply voltage monitoring (SVM, the device is not automatically reset). SVS and SVM circuitry is available on the primary supply and core supply.

### Hardware Multiplier (MPY) (Link to User's Guide)

The multiplication operation is supported by a dedicated peripheral module. The module performs operations with 32-bit, 24-bit, 16-bit, and 8-bit operands. The module is capable of supporting signed and unsigned multiplication as well as signed and unsigned multiply and accumulate operations.



#### Real-Time Clock (RTC\_A) (Link to User's Guide)

The RTC\_A module can be used as a general-purpose 32-bit counter (counter mode) or as an integrated real-time clock (RTC) (calendar mode). In counter mode, the RTC\_A also includes two independent 8-bit timers that can be cascaded to form a 16-bit timer/counter. Both timers can be read and written by software. Calendar mode integrates an internal calendar which compensates for months with less than 31 days and includes leap year correction. The RTC\_A also supports flexible alarm functions and offset-calibration hardware.

#### Watchdog Timer (WDT A) (Link to User's Guide)

The primary function of the watchdog timer (WDT\_A) module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be configured as an interval timer and can generate interrupts at selected time intervals.



#### System Module (SYS) (Link to User's Guide)

The SYS module handles many of the system functions within the device. These include power on reset and power up clear handling, NMI source selection and management, reset interrupt vector generators, boot strap loader entry mechanisms, as well as, configuration management (device descriptors). It also includes a data exchange mechanism via JTAG called a JTAG mailbox that can be used in the application.

**Table 8. System Module Interrupt Vector Registers** 

NTERRUPT VECTOR REGISTER	ADDRESS	INTERRUPT EVENT	VALUE	PRIORITY
SYSRSTIV, System Reset	019Eh	No interrupt pending	00h	
		Brownout (BOR)	02h	Highest
		RST/NMI (POR)	04h	
		PMMSWBOR (BOR)	06h	
		Wakeup from LPMx.5	08h	
		Security violation (BOR)	0Ah	
		SVSL (POR)	0Ch	
		SVSH (POR)	0Eh	
		SVML_OVP (POR)	10h	
		SVMH_OVP (POR)	12h	
		PMMSWPOR (POR)	14h	
		WDT timeout (PUC)	16h	
		WDT password violation (PUC)	18h	
		KEYV flash password violation (PUC)	1Ah	
		Reserved	1Ch	
		Peripheral area fetch (PUC)	1Eh	
		PMM password violation (PUC)	20h	
		Reserved	22h to 3Eh	Lowest
SYSSNIV, System NMI	019Ch	No interrupt pending	00h	
		SVMLIFG	02h	Highest
		SVMHIFG	04h	
		SVSMLDLYIFG	06h	
		SVSMHDLYIFG	08h	
		VMAIFG	0Ah	
		JMBINIFG	0Ch	
		JMBOUTIFG	0Eh	
		SVMLVLRIFG	10h	
		SVMHVLRIFG	12h	
		Reserved	14h to 1Eh	Lowest
SYSUNIV, User NMI	019Ah	No interrupt pending	00h	
		NMIFG	02h	Highest
		OFIFG	04h	
		ACCVIFG	06h	
		Reserved	08h	
		Reserved	0Ah to 1Eh	Lowest



#### **DMA Controller (Link to User's Guide)**

The DMA controller allows movement of data from one memory address to another without CPU intervention. For example, the DMA controller can be used to move data from the ADC12\_A conversion memory to RAM. Using the DMA controller can increase the throughput of peripheral modules. The DMA controller reduces system power consumption by allowing the CPU to remain in sleep mode, without having to awaken to move data to or from a peripheral.

Table 9. DMA Trigger Assignments (1)

TDIOCED	CHANNEL				
TRIGGER	0	1	2		
0	DMAREQ	DMAREQ	DMAREQ		
1	TA0CCR0 CCIFG	TA0CCR0 CCIFG	TA0CCR0 CCIFG		
2	TA0CCR2 CCIFG	TA0CCR2 CCIFG	TA0CCR2 CCIFG		
3	TA1CCR0 CCIFG	TA1CCR0 CCIFG	TA1CCR0 CCIFG		
4	TA1CCR2 CCIFG	TA1CCR2 CCIFG	TA1CCR2 CCIFG		
5	TB0CCR0 CCIFG	TB0CCR0 CCIFG	TB0CCR0 CCIFG		
6	TB0CCR2 CCIFG	TB0CCR2 CCIFG	TB0CCR2 CCIFG		
7	Reserved	Reserved	Reserved		
8	Reserved	Reserved	Reserved		
9	Reserved	Reserved	Reserved		
10	Reserved	Reserved	Reserved		
11	Reserved	Reserved	Reserved		
12	Reserved	Reserved	Reserved		
13	Reserved	Reserved	Reserved		
14	Reserved	Reserved	Reserved		
15	Reserved	Reserved	Reserved		
16	UCA0RXIFG	UCA0RXIFG	UCA0RXIFG		
17	UCA0TXIFG	UCA0TXIFG	UCA0TXIFG		
18	UCB0RXIFG	UCB0RXIFG	UCB0RXIFG		
19	UCB0TXIFG	UCB0TXIFG	UCB0TXIFG		
20	UCA1RXIFG	UCA1RXIFG	UCA1RXIFG		
21	UCA1TXIFG	UCA1TXIFG	UCA1TXIFG		
22	UCB1RXIFG	UCB1RXIFG	UCB1RXIFG		
23	UCB1TXIFG	UCB1TXIFG	UCB1TXIFG		
24	ADC12IFGx	ADC12IFGx	ADC12IFGx		
25	Reserved	Reserved	Reserved		
26	Reserved	Reserved	Reserved		
27	Reserved	Reserved	Reserved		
28	Reserved	Reserved	Reserved		
29	MPY ready	MPY ready	MPY ready		
30	DMA2IFG	DMA0IFG	DMA1IFG		
31	DMAE0	DMAE0	DMAE0		

Reserved DMA triggers may be used by other devices in the family. Reserved DMA triggers do not cause any DMA trigger event when selected.



## Universal Serial Communication Interface (USCI) (Links to User's Guide: UART Mode, SPI Mode, I2C Mode)

The USCI modules are used for serial data communication. The USCI module supports synchronous communication protocols such as SPI (3 or 4 pin) and  $I^2C$ , and asynchronous communication protocols such as UART, enhanced UART with automatic baudrate detection, and IrDA. Each USCI module contains two portions, A and B.

The USCI An module provides support for SPI (3 pin or 4 pin), UART, enhanced UART, or IrDA.

The USCI\_Bn module provides support for SPI (3 pin or 4 pin) or I2C.

The MSP430F5438A, MSP430F5436A, and MSP430F5419A include four complete USCI modules (n = 0 to 3). The MSP430F5437A, MSP430F5435A, and MSP430F5418A include two complete USCI modules (n = 0 to 1).

#### TA0 (Link to User's Guide)

TA0 is a 16-bit timer/counter (Timer\_A type) with five capture/compare registers. It can support multiple capture/compares, PWM outputs, and interval timing. It also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

**Table 10. TA0 Signal Connections** 

INPUT PIN	NUMBER	DEVICE	MODULE	MODULE	MODULE	DEVICE	OUTPUT PI	N NUMBER
PZ, ZQW	PN	INPUT SIGNAL	INPUT SIGNAL	BLOCK	OUTPUT SIGNAL	OUTPUT SIGNAL	PZ, ZQW	PN
17, H1-P1.0	17-P1.0	TA0CLK	TACLK					
		ACLK	ACLK	<b>T</b> '	NIA	N.I.A		
		SMCLK	SMCLK	Timer	NA	NA		
17, H1-P1.0	17-P1.0	TA0CLK	TACLK					
18, H4-P1.1	18-P1.1	TA0.0	CCI0A				18, H4-P1.1	18-P1.1
57, H9-P8.0	60-P8.0	TA0.0	CCI0B				57, H9-P8.0	60-P8.0
		DV <sub>SS</sub>	GND	CCR0	TA0	TA0.0	ADC12 (internal) ADC12SHSx = {1}	ADC12 (internal) ADC12SHSx = {1}
		DV <sub>CC</sub>	V <sub>CC</sub>					
19, J4-P1.2	19-P1.2	TA0.1	CCI1A			TA0.1	19, J4-P1.2	19-P1.2
58, H11-P8.1	61-P8.1	TA0.1	CCI1B	CCD4	T 4 4		58, H11-P8.1	61-P8.1
		DV <sub>SS</sub>	GND	CCR1	TA1			
		DV <sub>CC</sub>	V <sub>CC</sub>					
20, J1-P1.3	20-P1.3	TA0.2	CCI2A				20, J1-P1.3	20-P1.3
59, H12-P8.2	62-P8.2	TA0.2	CCI2B	CCDO	TA2	TA0.2	59, H12-P8.2	62-P8.2
		DV <sub>SS</sub>	GND	CCR2	IA2	1AU.2		
		DV <sub>CC</sub>	V <sub>CC</sub>					
21, J2-P1.4	21-P1.4	TA0.3	CCI3A				21, J2-P1.4	21-P1.4
60, G9-P8.3	63-P8.3	TA0.3	CCI3B	CCDa	TA3	TA0.3	60, G9-P8.3	63-P8.3
		DV <sub>SS</sub>	GND	CCR3	TA3	1A0.3		
		DV <sub>CC</sub>	V <sub>CC</sub>					
22, K1-P1.5	22-P1.5	TA0.4	CCI4A				22, K1-P1.5	22-P1.5
61, G11-P8.4	64-P8.4	TA0.4	CCI4B	CCD4	T		61, G11-P8.4	64-P8.4
		DV <sub>SS</sub>	GND	CCR4	TA4	TA0.4		
		DV <sub>CC</sub>	V <sub>CC</sub>					



#### TA1 (Link to User's Guide)

TA1 is a 16-bit timer/counter (Timer\_A type) with three capture/compare registers. It can support multiple capture/compares, PWM outputs, and interval timing. It also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

**Table 11. TA1 Signal Connections** 

INPUT PIN	NUMBER	DEVICE	MODULE	MODULE	MODULE	DEVICE	OUTPUT PI	N NUMBER	
PZ, ZQW	PN	INPUT SIGNAL	INPUT SIGNAL	BLOCK	OUTPUT SIGNAL	OUTPUT SIGNAL	PZ, ZQW	PN	
25, M1-P2.0	25-P2.0	TA1CLK	TACLK						
		ACLK	ACLK	T:	NIA	NIA			
		SMCLK	SMCLK	Timer	NA	NA			
25, M1-P2.0	25-P2.0	TA1CLK	TACLK						
26, L2-P2.1	26-P2.1	TA1.0	CCI0A		R0 TA0			26, L2-P2.1	26-P2.1
65, F11-P8.5	65-P8.5	TA1.0	CCI0B	CCR0		TA4.0	65, F11-P8.5	65-P8.5	
		DV <sub>SS</sub>	GND			TA1.0			
		DV <sub>CC</sub>	V <sub>CC</sub>						
27, M2-P2.2	27-P2.2	TA1.1	CCI1A	CCR1	CCR1			27, M2-P2.2	27-P2.2
66, E11-P8.6	66-P8.6	TA1.1	CCI1B			CCR1 TA1	T.4.4	66, E11-P8.6	66-P8.6
		DV <sub>SS</sub>	GND				TA1.1		
		DV <sub>CC</sub>	V <sub>CC</sub>						
28, L3-P2.3	28-P2.3	TA1.2	CCI2A				28, L3-P2.3	28-P2.3	
56, J12-P7.3	59-P7.3	TA1.2	CCI2B	CCBa	TA 2	TA4.0	56, J12-P7.3	59-P7.3	
		DV <sub>SS</sub>	GND	CCR2	TA2	TA1.2			
		DV <sub>CC</sub>	V <sub>CC</sub>						



#### TB0 (Link to User's Guide)

TB0 is a 16-bit timer/counter (Timer\_B type) with seven capture/compare registers. It can support multiple capture/compares, PWM outputs, and interval timing. It also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

**Table 12. TB0 Signal Connections** 

INPUT PIN	NUMBER	DEVICE	MODULE	MODULE	MODULE	DEVICE	OUTPUT PI	N NUMBER
PZ, ZQW	PN	INPUT SIGNAL	INPUT SIGNAL	BLOCK	OUTPUT SIGNAL	OUTPUT SIGNAL	PZ, ZQW	PN
50, M12-P4.7	53-P4.7	TB0CLK	TBCLK					
		ACLK	ACLK	Timer	NA	NA		
		SMCLK	SMCLK	Tilliel	INA	INA		
50, M12-P4.7	53-P4.7	TB0CLK	TBCLK					
43, J8-P4.0	43-P4.0	TB0.0	CCI0A				43, J8-P4.0	43-P4.0
43, J8-P4.0	43-P4.0	TB0.0	CCI0B	CCR0	TB0	TB0.0	ADC12 (internal) ADC12SHSx = {2}	ADC12 (internal) ADC12SHSx = {2}
		$DV_SS$	GND					
		DV <sub>CC</sub>	V <sub>CC</sub>					
44, M9-P4.1	44-P4.1	TB0.1	CCI1A				44, M9-P4.1	44-P4.1
44, M9-P4.1	44-P4.1	TB0.1	CCI1B	CCR1	TB1	TB0.1	ADC12 (internal) ADC12SHSx = {3}	ADC12 (internal) ADC12SHSx = {3}
		DV <sub>SS</sub>	GND					
		DV <sub>CC</sub>	V <sub>CC</sub>					
45, L9-P4.2	45-P4.2	TB0.2	CCI2A			TB0.2	45, L9-P4.2	45-P4.2
45, L9-P4.2	45-P4.2	TB0.2	CCI2B	CCR2	TB2			
		$DV_SS$	GND	CCR2	I DZ			
		DV <sub>CC</sub>	V <sub>CC</sub>					
46, L10-P4.3	46-P4.3	TB0.3	CCI3A				46, L10-P4.3	46-P4.3
46, L10-P4.3	46-P4.3	TB0.3	CCI3B	CCR3	TB3	TB0.3		
		DV <sub>SS</sub>	GND	CCR3	103	100.3		
		DV <sub>CC</sub>	V <sub>CC</sub>					
47, M10-P4.4	47-P4.4	TB0.4	CCI4A				47, M10-P4.4	47-P4.4
47, M10-P4.4	47-P4.4	TB0.4	CCI4B	CCD4	TB4	TD0 4		
		DV <sub>SS</sub>	GND	CCR4	104	TB0.4		
		$DV_CC$	V <sub>CC</sub>					
48, L11-P4.5	48-P4.5	TB0.5	CCI5A				48, L11-P4.5	48-P4.5
48, L11-P4.5	48-P4.5	TB0.5	CCI5B	CODE	TDE	TD0 F		
		$DV_SS$	GND	CCR5	TB5	TB0.5		
		DV <sub>CC</sub>	V <sub>CC</sub>					
49, M11-P4.6	52-P4.6	TB0.6	CCI6A			TB0.6	49, M11-P4.6	52-P4.6
		ACLK (internal)	CCI6B	CCR6	TB6			
		DV <sub>SS</sub>	GND					
		DV <sub>CC</sub>	V <sub>CC</sub>					



#### ADC12 A (Link to User's Guide)

The ADC12\_A module supports fast 12-bit analog-to-digital conversions. The module implements a 12-bit SAR core, sample select control, reference generator, and a 16-word conversion-and-control buffer. The conversion-and-control buffer allows up to 16 independent ADC samples to be converted and stored without any CPU intervention.

#### CRC16 (Link to User's Guide)

The CRC16 module produces a signature based on a sequence of entered data values and can be used for data checking purposes. The CRC16 module signature is based on the CRC-CCITT standard.

#### **REF Voltage Reference (Link to User's Guide)**

The reference module (REF) is responsible for generation of all critical reference voltages that can be used by the various analog peripherals in the device.

#### Embedded Emulation Module (EEM) (L Version) (Link to User's Guide)

The EEM supports real-time in-system debugging. The L version of the EEM implemented on all devices has the following features:

- Eight hardware triggers or breakpoints on memory access
- Two hardware trigger or breakpoint on CPU register write access
- Up to ten hardware triggers can be combined to form complex triggers or breakpoints
- · Two cycle counters
- Sequencer
- State storage
- Clock control on module level

## TEXAS INSTRUMENTS

## **Peripheral File Map**

Table 13. Peripherals

MODULE NAME	BASE ADDRESS	OFFSET ADDRESS RANGE
Special Functions (see Table 14)	0100h	000h - 01Fh
PMM (see Table 15)	0120h	000h - 010h
Flash Control (see Table 16)	0140h	000h - 00Fh
CRC16 (see Table 17)	0150h	000h - 007h
RAM Control (see Table 18)	0158h	000h - 001h
Watchdog (see Table 19)	015Ch	000h - 001h
UCS (see Table 20)	0160h	000h - 01Fh
SYS (see Table 21)	0180h	000h - 01Fh
Shared Reference (see Table 22)	01B0h	000h - 001h
Port P1, P2 (see Table 23)	0200h	000h - 01Fh
Port P3, P4 (see Table 24)	0220h	000h - 00Bh
Port P5, P6 (see Table 25)	0240h	000h - 00Bh
Port P7, P8 (see Table 26)	0260h	000h - 00Bh
Port P9, P10 (see Table 27)	0280h	000h - 00Bh
Port P11 (see Table 28)	02A0h	000h - 00Ah
Port PJ (see Table 29)	0320h	000h - 01Fh
TA0 (see Table 30)	0340h	000h - 02Eh
TA1 (see Table 31)	0380h	000h - 02Eh
TB0 (see Table 32)	03C0h	000h - 02Eh
Real Timer Clock (RTC_A) (see Table 33)	04A0h	000h - 01Bh
32-bit Hardware Multiplier (see Table 34)	04C0h	000h - 02Fh
DMA General Control (see Table 35)	0500h	000h - 00Fh
DMA Channel 0 (see Table 35)	0510h	000h - 00Ah
DMA Channel 1 (see Table 35)	0520h	000h - 00Ah
DMA Channel 2 (see Table 35)	0530h	000h - 00Ah
USCI_A0 (see Table 36)	05C0h	000h - 01Fh
USCI_B0 (see Table 37)	05E0h	000h - 01Fh
USCI_A1 (see Table 38)	0600h	000h - 01Fh
USCI_B1 (see Table 39)	0620h	000h - 01Fh
USCI_A2 (see Table 40)	0640h	000h - 01Fh
USCI_B2 (see Table 41)	0660h	000h - 01Fh
USCI_A3 (see Table 42)	0680h	000h - 01Fh
USCI_B3 (see Table 43)	06A0h	000h - 01Fh
ADC12_A (see Table 44)	0700h	000h - 03Eh



#### Table 14. Special Function Registers (Base Address: 0100h)

REGISTER DESCRIPTION	REGISTER	OFFSET
SFR interrupt enable	SFRIE1	00h
SFR interrupt flag	SFRIFG1	02h
SFR reset pin control	SFRRPCR	04h

## Table 15. PMM Registers (Base Address: 0120h)

REGISTER DESCRIPTION	REGISTER	OFFSET
PMM Control 0	PMMCTL0	00h
PMM control 1	PMMCTL1	02h
SVS high side control	SVSMHCTL	04h
SVS low side control	SVSMLCTL	06h
PMM interrupt flags	PMMIFG	0Ch
PMM interrupt enable	PMMIE	0Eh
PMM power mode 5 control	PM5CTL0	10h

#### Table 16. Flash Control Registers (Base Address: 0140h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Flash control 1	FCTL1	00h
Flash control 3	FCTL3	04h
Flash control 4	FCTL4	06h

## Table 17. CRC16 Registers (Base Address: 0150h)

REGISTER DESCRIPTION	REGISTER	OFFSET
CRC data input	CRC16DI	00h
CRC data input reverse byte	CRCDIRB	02h
CRC initialization and result	CRCINIRES	04h
CRC result reverse byte	CRCRESR	06h

## Table 18. RAM Control Registers (Base Address: 0158h)

REGISTER DESCRIPTION	REGISTER	OFFSET
RAM control 0	RCCTL0	00h

## Table 19. Watchdog Registers (Base Address: 015Ch)

REGISTER DESCRIPTION	REGISTER	OFFSET
Watchdog timer control	WDTCTL	00h

#### Table 20. UCS Registers (Base Address: 0160h)

REGISTER DESCRIPTION	REGISTER	OFFSET
UCS control 0	UCSCTL0	00h
UCS control 1	UCSCTL1	02h
UCS control 2	UCSCTL2	04h
UCS control 3	UCSCTL3	06h
UCS control 4	UCSCTL4	08h
UCS control 5	UCSCTL5	0Ah
UCS control 6	UCSCTL6	0Ch
UCS control 7	UCSCTL7	0Eh
UCS control 8	UCSCTL8	10h



## Table 21. SYS Registers (Base Address: 0180h)

REGISTER DESCRIPTION	REGISTER	OFFSET
System control	SYSCTL	00h
Bootstrap loader configuration area	SYSBSLC	02h
JTAG mailbox control	SYSJMBC	06h
JTAG mailbox input 0	SYSJMBI0	08h
JTAG mailbox input 1	SYSJMBI1	0Ah
JTAG mailbox output 0	SYSJMBO0	0Ch
JTAG mailbox output 1	SYSJMBO1	0Eh
Bus Error vector generator	SYSBERRIV	18h
User NMI vector generator	SYSUNIV	1Ah
System NMI vector generator	SYSSNIV	1Ch
Reset vector generator	SYSRSTIV	1Eh

## Table 22. Shared Reference Registers (Base Address: 01B0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Shared reference control	REFCTL	00h

## Table 23. Port P1, P2 Registers (Base Address: 0200h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P1 input	P1IN	00h
Port P1 output	P1OUT	02h
Port P1 direction	P1DIR	04h
Port P1 pullup/pulldown enable	P1REN	06h
Port P1 drive strength	P1DS	08h
Port P1 selection	P1SEL	0Ah
Port P1 interrupt vector word	P1IV	0Eh
Port P1 interrupt edge select	P1IES	18h
Port P1 interrupt enable	P1IE	1Ah
Port P1 interrupt flag	P1IFG	1Ch
Port P2 input	P2IN	01h
Port P2 output	P2OUT	03h
Port P2 direction	P2DIR	05h
Port P2 pullup/pulldown enable	P2REN	07h
Port P2 drive strength	P2DS	09h
Port P2 selection	P2SEL	0Bh
Port P2 interrupt vector word	P2IV	1Eh
Port P2 interrupt edge select	P2IES	19h
Port P2 interrupt enable	P2IE	1Bh
Port P2 interrupt flag	P2IFG	1Dh



## Table 24. Port P3, P4 Registers (Base Address: 0220h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P3 input	P3IN	00h
Port P3 output	P3OUT	02h
Port P3 direction	P3DIR	04h
Port P3 pullup/pulldown enable	P3REN	06h
Port P3 drive strength	P3DS	08h
Port P3 selection	P3SEL	0Ah
Port P4 input	P4IN	01h
Port P4 output	P4OUT	03h
Port P4 direction	P4DIR	05h
Port P4 pullup/pulldown enable	P4REN	07h
Port P4 drive strength	P4DS	09h
Port P4 selection	P4SEL	0Bh

## Table 25. Port P5, P6 Registers (Base Address: 0240h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P5 input	P5IN	00h
Port P5 output	P5OUT	02h
Port P5 direction	P5DIR	04h
Port P5 pullup/pulldown enable	P5REN	06h
Port P5 drive strength	P5DS	08h
Port P5 selection	P5SEL	0Ah
Port P6 input	P6IN	01h
Port P6 output	P6OUT	03h
Port P6 direction	P6DIR	05h
Port P6 pullup/pulldown enable	P6REN	07h
Port P6 drive strength	P6DS	09h
Port P6 selection	P6SEL	0Bh

## Table 26. Port P7, P8 Registers (Base Address: 0260h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P7 input	P7IN	00h
Port P7 output	P7OUT	02h
Port P7 direction	P7DIR	04h
Port P7 pullup/pulldown enable	P7REN	06h
Port P7 drive strength	P7DS	08h
Port P7 selection	P7SEL	0Ah
Port P8 input	P8IN	01h
Port P8 output	P8OUT	03h
Port P8 direction	P8DIR	05h
Port P8 pullup/pulldown enable	P8REN	07h
Port P8 drive strength	P8DS	09h
Port P8 selection	P8SEL	0Bh



## Table 27. Port P9, P10 Registers (Base Address: 0280h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P9 input	P9IN	00h
Port P9 output	P9OUT	02h
Port P9 direction	P9DIR	04h
Port P9 pullup/pulldown enable	P9REN	06h
Port P9 drive strength	P9DS	08h
Port P9 selection	P9SEL	0Ah
Port P10 input	P10IN	01h
Port P10 output	P10OUT	03h
Port P10 direction	P10DIR	05h
Port P10 pullup/pulldown enable	P10REN	07h
Port P10 drive strength	P10DS	09h
Port P10 selection	P10SEL	0Bh

## Table 28. Port P11 Registers (Base Address: 02A0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port P11 input	P11IN	00h
Port P11 output	P11OUT	02h
Port P11 direction	P11DIR	04h
Port P11 pullup/pulldown enable	P11REN	06h
Port P11 drive strength	P11DS	08h
Port P11 selection	P11SEL	0Ah

## Table 29. Port J Registers (Base Address: 0320h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Port PJ input	PJIN	00h
Port PJ output	PJOUT	02h
Port PJ direction	PJDIR	04h
Port PJ pullup/pulldown enable	PJREN	06h
Port PJ drive strength	PJDS	08h



## Table 30. TA0 Registers (Base Address: 0340h)

REGISTER DESCRIPTION	REGISTER	OFFSET
TA0 control	TA0CTL	00h
Capture/compare control 0	TA0CCTL0	02h
Capture/compare control 1	TA0CCTL1	04h
Capture/compare control 2	TA0CCTL2	06h
Capture/compare control 3	TA0CCTL3	08h
Capture/compare control 4	TA0CCTL4	0Ah
TA0 counter register	TA0R	10h
Capture/compare register 0	TA0CCR0	12h
Capture/compare register 1	TA0CCR1	14h
Capture/compare register 2	TA0CCR2	16h
Capture/compare register 3	TA0CCR3	18h
Capture/compare register 4	TA0CCR4	1Ah
TA0 expansion register 0	TA0EX0	20h
TA0 interrupt vector	TAOIV	2Eh

## Table 31. TA1 Registers (Base Address: 0380h)

REGISTER DESCRIPTION	REGISTER	OFFSET
TA1 control	TA1CTL	00h
Capture/compare control 0	TA1CCTL0	02h
Capture/compare control 1	TA1CCTL1	04h
Capture/compare control 2	TA1CCTL2	06h
TA1 counter register	TA1R	10h
Capture/compare register 0	TA1CCR0	12h
Capture/compare register 1	TA1CCR1	14h
Capture/compare register 2	TA1CCR2	16h
TA1 expansion register 0	TA1EX0	20h
TA1 interrupt vector	TA1IV	2Eh



## Table 32. TB0 Registers (Base Address: 03C0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
TB0 control	TB0CTL	00h
Capture/compare control 0	TB0CCTL0	02h
Capture/compare control 1	TB0CCTL1	04h
Capture/compare control 2	TB0CCTL2	06h
Capture/compare control 3	TB0CCTL3	08h
Capture/compare control 4	TB0CCTL4	0Ah
Capture/compare control 5	TB0CCTL5	0Ch
Capture/compare control 6	TB0CCTL6	0Eh
TB0 register	TBOR	10h
Capture/compare register 0	TB0CCR0	12h
Capture/compare register 1	TB0CCR1	14h
Capture/compare register 2	TB0CCR2	16h
Capture/compare register 3	TB0CCR3	18h
Capture/compare register 4	TB0CCR4	1Ah
Capture/compare register 5	TB0CCR5	1Ch
Capture/compare register 6	TB0CCR6	1Eh
TB0 expansion register 0	TB0EX0	20h
TB0 interrupt vector	TB0IV	2Eh

## Table 33. Real Time Clock Registers (Base Address: 04A0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
RTC control 0	RTCCTL0	00h
RTC control 1	RTCCTL1	01h
RTC control 2	RTCCTL2	02h
RTC control 3	RTCCTL3	03h
RTC prescaler 0 control	RTCPS0CTL	08h
RTC prescaler 1 control	RTCPS1CTL	0Ah
RTC prescaler 0	RTCPS0	0Ch
RTC prescaler 1	RTCPS1	0Dh
RTC interrupt vector word	RTCIV	0Eh
RTC seconds/counter register 1	RTCSEC/RTCNT1	10h
RTC minutes/counter register 2	RTCMIN/RTCNT2	11h
RTC hours/counter register 3	RTCHOUR/RTCNT3	12h
RTC day of week/counter register 4	RTCDOW/RTCNT4	13h
RTC days	RTCDAY	14h
RTC month	RTCMON	15h
RTC year low	RTCYEARL	16h
RTC year high	RTCYEARH	17h
RTC alarm minutes	RTCAMIN	18h
RTC alarm hours	RTCAHOUR	19h
RTC alarm day of week	RTCADOW	1Ah
RTC alarm days	RTCADAY	1Bh



## Table 34. 32-bit Hardware Multiplier Registers (Base Address: 04C0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
16-bit operand 1 – multiply	MPY	00h
16-bit operand 1 – signed multiply	MPYS	02h
16-bit operand 1 – multiply accumulate	MAC	04h
16-bit operand 1 – signed multiply accumulate	MACS	06h
16-bit operand 2	OP2	08h
16 x 16 result low word	RESLO	0Ah
16 x 16 result high word	RESHI	0Ch
16 x 16 sum extension register	SUMEXT	0Eh
32-bit operand 1 – multiply low word	MPY32L	10h
32-bit operand 1 – multiply high word	MPY32H	12h
32-bit operand 1 – signed multiply low word	MPYS32L	14h
32-bit operand 1 – signed multiply high word	MPYS32H	16h
32-bit operand 1 – multiply accumulate low word	MAC32L	18h
32-bit operand 1 – multiply accumulate high word	MAC32H	1Ah
32-bit operand 1 – signed multiply accumulate low word	MACS32L	1Ch
32-bit operand 1 – signed multiply accumulate high word	MACS32H	1Eh
32-bit operand 2 – low word	OP2L	20h
32-bit operand 2 – high word	OP2H	22h
32 × 32 result 0 – least significant word	RES0	24h
32 x 32 result 1	RES1	26h
32 x 32 result 2	RES2	28h
32 × 32 result 3 – most significant word	RES3	2Ah
MPY32 control register 0	MPY32CTL0	2Ch



## Table 35. DMA Registers (Base Address DMA General Control: 0500h, DMA Channel 0: 0510h, DMA Channel 1: 0520h, DMA Channel 2: 0530h)

REGISTER DESCRIPTION	REGISTER	OFFSET
DMA channel 0 control	DMA0CTL	00h
DMA channel 0 source address low	DMA0SAL	02h
DMA channel 0 source address high	DMA0SAH	04h
DMA channel 0 destination address low	DMA0DAL	06h
DMA channel 0 destination address high	DMA0DAH	08h
DMA channel 0 transfer size	DMA0SZ	0Ah
DMA channel 1 control	DMA1CTL	00h
DMA channel 1 source address low	DMA1SAL	02h
DMA channel 1 source address high	DMA1SAH	04h
DMA channel 1 destination address low	DMA1DAL	06h
DMA channel 1 destination address high	DMA1DAH	08h
DMA channel 1 transfer size	DMA1SZ	0Ah
DMA channel 2 control	DMA2CTL	00h
DMA channel 2 source address low	DMA2SAL	02h
DMA channel 2 source address high	DMA2SAH	04h
DMA channel 2 destination address low	DMA2DAL	06h
DMA channel 2 destination address high	DMA2DAH	08h
DMA channel 2 transfer size	DMA2SZ	0Ah
DMA module control 0	DMACTL0	00h
DMA module control 1	DMACTL1	02h
DMA module control 2	DMACTL2	04h
DMA module control 3	DMACTL3	06h
DMA module control 4	DMACTL4	08h
DMA interrupt vector	DMAIV	0Eh

## Table 36. USCI\_A0 Registers (Base Address: 05C0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI control 1	UCA0CTL1	00h
USCI control 0	UCA0CTL0	01h
USCI baud rate 0	UCA0BR0	06h
USCI baud rate 1	UCA0BR1	07h
USCI modulation control	UCA0MCTL	08h
USCI status	UCA0STAT	0Ah
USCI receive buffer	UCA0RXBUF	0Ch
USCI transmit buffer	UCA0TXBUF	0Eh
USCI LIN control	UCA0ABCTL	10h
USCI IrDA transmit control	UCA0IRTCTL	12h
USCI IrDA receive control	UCA0IRRCTL	13h
USCI interrupt enable	UCA0IE	1Ch
USCI interrupt flags	UCA0IFG	1Dh
USCI interrupt vector word	UCA0IV	1Eh



## Table 37. USCI\_B0 Registers (Base Address: 05E0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI synchronous control 1	UCB0CTL1	00h
USCI synchronous control 0	UCB0CTL0	01h
USCI synchronous bit rate 0	UCB0BR0	06h
USCI synchronous bit rate 1	UCB0BR1	07h
USCI synchronous status	UCB0STAT	0Ah
USCI synchronous receive buffer	UCB0RXBUF	0Ch
USCI synchronous transmit buffer	UCB0TXBUF	0Eh
USCI I2C own address	UCB0I2COA	10h
USCI I2C slave address	UCB0I2CSA	12h
USCI interrupt enable	UCB0IE	1Ch
USCI interrupt flags	UCB0IFG	1Dh
USCI interrupt vector word	UCB0IV	1Eh

## Table 38. USCI\_A1 Registers (Base Address: 0600h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI control 1	UCA1CTL1	00h
USCI control 0	UCA1CTL0	01h
USCI baud rate 0	UCA1BR0	06h
USCI baud rate 1	UCA1BR1	07h
USCI modulation control	UCA1MCTL	08h
USCI status	UCA1STAT	0Ah
USCI receive buffer	UCA1RXBUF	0Ch
USCI transmit buffer	UCA1TXBUF	0Eh
USCI LIN control	UCA1ABCTL	10h
USCI IrDA transmit control	UCA1IRTCTL	12h
USCI IrDA receive control	UCA1IRRCTL	13h
USCI interrupt enable	UCA1IE	1Ch
USCI interrupt flags	UCA1IFG	1Dh
USCI interrupt vector word	UCA1IV	1Eh



## Table 39. USCI\_B1 Registers (Base Address: 0620h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI synchronous control 1	UCB1CTL1	00h
USCI synchronous control 0	UCB1CTL0	01h
USCI synchronous bit rate 0	UCB1BR0	06h
USCI synchronous bit rate 1	UCB1BR1	07h
USCI synchronous status	UCB1STAT	0Ah
USCI synchronous receive buffer	UCB1RXBUF	0Ch
USCI synchronous transmit buffer	UCB1TXBUF	0Eh
USCI I2C own address	UCB1I2COA	10h
USCI I2C slave address	UCB1I2CSA	12h
USCI interrupt enable	UCB1IE	1Ch
USCI interrupt flags	UCB1IFG	1Dh
USCI interrupt vector word	UCB1IV	1Eh

## Table 40. USCI\_A2 Registers (Base Address: 0640h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI control 1	UCA2CTL1	00h
USCI control 0	UCA2CTL0	01h
USCI baud rate 0	UCA2BR0	06h
USCI baud rate 1	UCA2BR1	07h
USCI modulation control	UCA2MCTL	08h
USCI status	UCA2STAT	0Ah
USCI receive buffer	UCA2RXBUF	0Ch
USCI transmit buffer	UCA2TXBUF	0Eh
USCI LIN control	UCA2ABCTL	10h
USCI IrDA transmit control	UCA2IRTCTL	12h
USCI IrDA receive control	UCA2IRRCTL	13h
USCI interrupt enable	UCA2IE	1Ch
USCI interrupt flags	UCA2IFG	1Dh
USCI interrupt vector word	UCA2IV	1Eh

## Table 41. USCI\_B2 Registers (Base Address: 0660h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI synchronous control 1	UCB2CTL1	00h
USCI synchronous control 0	UCB2CTL0	01h
USCI synchronous bit rate 0	UCB2BR0	06h
USCI synchronous bit rate 1	UCB2BR1	07h
USCI synchronous status	UCB2STAT	0Ah
USCI synchronous receive buffer	UCB2RXBUF	0Ch
USCI synchronous transmit buffer	UCB2TXBUF	0Eh
USCI I2C own address	UCB2I2COA	10h
USCI I2C slave address	UCB2I2CSA	12h
USCI interrupt enable	UCB2IE	1Ch
USCI interrupt flags	UCB2IFG	1Dh
USCI interrupt vector word	UCB2IV	1Eh



## Table 42. USCI\_A3 Registers (Base Address: 0680h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI control 1	UCA3CTL1	00h
USCI control 0	UCA3CTL0	01h
USCI baud rate 0	UCA3BR0	06h
USCI baud rate 1	UCA3BR1	07h
USCI modulation control	UCA3MCTL	08h
USCI status	UCA3STAT	0Ah
USCI receive buffer	UCA3RXBUF	0Ch
USCI transmit buffer	UCA3TXBUF	0Eh
USCI LIN control	UCA3ABCTL	10h
USCI IrDA transmit control	UCA3IRTCTL	12h
USCI IrDA receive control	UCA3IRRCTL	13h
USCI interrupt enable	UCA3IE	1Ch
USCI interrupt flags	UCA3IFG	1Dh
USCI interrupt vector word	UCA3IV	1Eh

## Table 43. USCI\_B3 Registers (Base Address: 06A0h)

REGISTER DESCRIPTION	REGISTER	OFFSET
USCI synchronous control 1	UCB3CTL1	00h
USCI synchronous control 0	UCB3CTL0	01h
USCI synchronous bit rate 0	UCB3BR0	06h
USCI synchronous bit rate 1	UCB3BR1	07h
USCI synchronous status	UCB3STAT	0Ah
USCI synchronous receive buffer	UCB3RXBUF	0Ch
USCI synchronous transmit buffer	UCB3TXBUF	0Eh
USCI I2C own address	UCB3I2COA	10h
USCI I2C slave address	UCB3I2CSA	12h
USCI interrupt enable	UCB3IE	1Ch
USCI interrupt flags	UCB3IFG	1Dh
USCI interrupt vector word	UCB3IV	1Eh



# Table 44. ADC12\_A Registers (Base Address: 0700h)

REGISTER DESCRIPTION	REGISTER	OFFSET
Control register 0	ADC12CTL0	00h
Control register 1	ADC12CTL1	02h
Control register 2	ADC12CTL2	04h
Interrupt-flag register	ADC12IFG	0Ah
Interrupt-enable register	ADC12IE	0Ch
Interrupt-vector-word register	ADC12IV	0Eh
ADC memory-control register 0	ADC12MCTL0	10h
ADC memory-control register 1	ADC12MCTL1	11h
ADC memory-control register 2	ADC12MCTL2	12h
ADC memory-control register 3	ADC12MCTL3	13h
ADC memory-control register 4	ADC12MCTL4	14h
ADC memory-control register 5	ADC12MCTL5	15h
ADC memory-control register 6	ADC12MCTL6	16h
ADC memory-control register 7	ADC12MCTL7	17h
ADC memory-control register 8	ADC12MCTL8	18h
ADC memory-control register 9	ADC12MCTL9	19h
ADC memory-control register 10	ADC12MCTL10	1Ah
ADC memory-control register 11	ADC12MCTL11	1Bh
ADC memory-control register 12	ADC12MCTL12	1Ch
ADC memory-control register 13	ADC12MCTL13	1Dh
ADC memory-control register 14	ADC12MCTL14	1Eh
ADC memory-control register 15	ADC12MCTL15	1Fh
Conversion memory 0	ADC12MEM0	20h
Conversion memory 1	ADC12MEM1	22h
Conversion memory 2	ADC12MEM2	24h
Conversion memory 3	ADC12MEM3	26h
Conversion memory 4	ADC12MEM4	28h
Conversion memory 5	ADC12MEM5	2Ah
Conversion memory 6	ADC12MEM6	2Ch
Conversion memory 7	ADC12MEM7	2Eh
Conversion memory 8	ADC12MEM8	30h
Conversion memory 9	ADC12MEM9	32h
Conversion memory 10	ADC12MEM10	34h
Conversion memory 11	ADC12MEM11	36h
Conversion memory 12	ADC12MEM12	38h
Conversion memory 13	ADC12MEM13	3Ah
Conversion memory 14	ADC12MEM14	3Ch
Conversion memory 15	ADC12MEM15	3Eh



## Absolute Maximum Ratings(1)

over operating free-air temperature range (unless otherwise noted)

Voltage applied at $V_{CC}$ to $V_{SS}$	–0.3 V to 4.1 V
Voltage applied to any pin (excluding VCORE) <sup>(2)</sup>	-0.3 V to V <sub>CC</sub> + 0.3 V
Diode current at any device pin	±2 mA
Storage temperature range, T <sub>stg</sub> <sup>(3)</sup>	−55°C to 105°C
Maximum junction temperature, T <sub>J</sub>	95°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages referenced to V<sub>SS</sub>. VCORE is for internal device use only. No external DC loading or voltage should be applied.
- (3) Higher temperature may be applied during board soldering according to the current JEDEC J-STD-020 specification with peak reflow temperatures not higher than classified on the device label on the shipping boxes or reels.

#### **Thermal Packaging Characteristics**

				VALUE	UNIT
			QFP (PZ)	50.1	
		Low-K board (JESD51-3)	QFP (PN)	57.9	
$\theta_{JA}$			BGA (ZQW)	60	0000
	Junction-to-ambient thermal resistance, still air		QFP (PZ)	40.8	°C/W
		High-K board (JESD51-7)	QFP (PN)	37.9	
			BGA (ZQW)	42	
	Junction-to-case thermal resistance		QFP (PZ)	8.9	
$\theta_{JC}$			QFP (PN)	10.3	°C/W
			BGA (ZQW)	8	

## **Recommended Operating Conditions**

Typical values are specified at  $V_{CC} = 3.3 \text{ V}$  and  $T_A = 25^{\circ}\text{C}$  (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V <sub>CC</sub>	Supply voltage during program execution $(AV_{CC} = DV_{CC1/2/3/4} = DV_{CC})^{(1)(2)}$	and flash programming	1.8		3.6	٧
$V_{SS}$	Supply voltage (AV <sub>SS</sub> = DV <sub>SS1/2/3/4</sub> = DV <sub>S</sub>	s)		0		٧
T <sub>A</sub>	Operating free-air temperature	-40		85	°C	
$T_{J}$	Operating junction temperature	-40		85	°C	
C <sub>VCORE</sub>	Recommended capacitor at VCORE		470		nF	
C <sub>DVCC</sub> / C <sub>VCORE</sub>	Capacitor ratio of DVCC to VCORE		10			
		PMMCOREVx = 0, 1.8 V ≤ V <sub>CC</sub> ≤ 3.6 V	0		8.0	
	Processor frequency (maximum MCLK	PMMCOREVx = 1, 2.0 V ≤ V <sub>CC</sub> ≤ 3.6 V	0		12.0	N 41 1-
f <sub>SYSTEM</sub>	Processor frequency (maximum MCLK frequency) <sup>(3)</sup> (see Figure 1)	PMMCOREVx = 2, 2.2 V ≤ V <sub>CC</sub> ≤ 3.6 V	N	MHz		
		PMMCOREVx = 3, 2.4 V ≤ V <sub>CC</sub> ≤ 3.6 V	0		25.0	

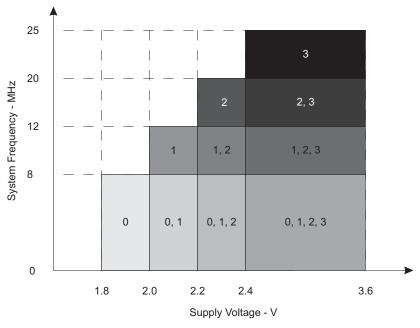
<sup>(1)</sup> It is recommended to power AV<sub>CC</sub> and DV<sub>CC</sub> from the same source. A maximum difference of 0.3 V between AV<sub>CC</sub> and DV<sub>CC</sub> can be tolerated during power up and operation.

<sup>(2)</sup> The minimum supply voltage is defined by the supervisor SVS levels when it is enabled. See the PMM, SVS High Side threshold parameters for the exact values and further details.

<sup>(3)</sup> The MSP430 CPU is clocked directly with MCLK. Both the high and low phase of MCLK must not exceed the pulse duration of the specified maximum frequency.

<sup>(4)</sup> Modules may have a different maximum input clock specification. See the specification of the respective module in this data sheet.





The numbers within the fields denote the supported PMMCOREVx settings.

Figure 1. Frequency vs Supply Voltage



#### **Electrical Characteristics**

# Active Mode Supply Current Into $V_{\text{CC}}$ Excluding External Current

over recommended operating free-air temperature (unless otherwise noted) (1)(2)(3)

				FREQUENCY $(f_{DCO} = f_{MCLK} = f_{SMCLK})$										
PARAMETER	EXECUTION MEMORY	V <sub>CC</sub>	<b>PMMCOREV</b> x	1 N	lHz	8 MHz		12 MHz		20 MHz		25 MHz		UNIT
	III EIII OKT			TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
			0	0.29	0.33	1.84	2.08							
	Flash 3.0 V	201/	1	0.32		2.08		3.10						
I <sub>AM</sub> , Flash		3.0 V	2	0.33		2.24		3.50		6.37				mA
			3	0.35		2.36		3.70		6.75		8.90	9.60	
			0	0.17	0.19	0.88	0.99							
	RAM 3.0 V	201/	1	0.18		1.00		1.47						
I <sub>AM, RAM</sub>		3.0 V	2	0.19		1.13		1.68		2.82				mA
			3	0.20		1.20		1.78		3.00		4.50	4.90	

All inputs are tied to 0 V or to V<sub>CC</sub>. Outputs do not source or sink any current. The currents are characterized with a Micro Crystal MS1V-T1K crystal with a load capacitance of 12.5 pF. The internal and external load capacitance are chosen to closely match the required 12.5 pF.

Characterized with program executing typical data processing.  $f_{ACLK} = 32768$  Hz,  $f_{DCO} = f_{MCLK} = f_{SMCLK}$  at specified frequency. XTS = CPUOFF = SCG0 = SCG1 = OSCOFF= SMCLKOFF = 0.



## Low-Power Mode Supply Currents (Into V<sub>cc</sub>) Excluding External Current

	PARAMETER	V	PMMCOREVx	-40	-40°C 25°C		°C	60°C		85°C		UNIT
	PARAMETER	V <sub>CC</sub>	PIVIIVICOREVX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	UNIT
	Low-power mode 0 <sup>(3)</sup>	2.2 V	0	69	93	69	93	69	93	69	93	
I <sub>LPM0,1MHz</sub>	(4)	3.0 V	3	73	100	73	100	73	100	73	100	μA
	Low-power mode 2 <sup>(5)</sup>	2.2 V	0	11	15.5	11	15.5	11	15.5	11	15.5	
I <sub>LPM2</sub>	(4)	3.0 V	3	11.7	17.5	11.7	17.5	11.7	17.5	11.7	17.5	μΑ
			0	1.4		1.7		2.6		6.6		
		2.2 V	1	1.5		1.8		2.9		9.9		
			2	1.5		2.0		3.3		10.1		
I <sub>LPM3,XT1LF</sub>	Low-power mode 3, crystal mode (6) (4)	3.0 V	0	1.8		2.1	2.4	2.8		7.1	13.6	μΑ
			1	1.8		2.3		3.1		10.5		
			2	1.9		2.4		3.5		10.6		
			3	2.0		2.3	2.6	3.9		11.8	14.8	
			0	1.0		1.2	1.42	2.0		5.8	12.9	
	Low-power mode 3,	3.0 V	1	1.0		1.3		2.3		6.0		
I <sub>LPM3,VLO</sub>	VLO mode <sup>(7)(4)</sup>	3.0 V	2	1.1		1.4		2.8		6.2		μA
			3	1.2		1.4	1.62	3.0		6.2	13.9	
			0	1.1		1.2	1.35	1.9		5.7	12.9	
	Low power mode 4 (8) (4)	201/	1	1.2		1.2		2.2		5.9		μA
I <sub>LPM4</sub>	Low-power mode 4 <sup>(8)(4)</sup>	3.0 V	2	1.3		1.3		2.6		6.1		
			3	1.3		1.3	1.52	2.9		6.2	13.9	
I <sub>LPM4.5</sub>	Low-power mode 4.5 <sup>(9)</sup>	3.0 V		0.10		0.10	0.13	0.20		0.50	1.14	μΑ

- (1) All inputs are tied to 0 V or to V<sub>CC</sub>. Outputs do not source or sink any current.
- (2) The currents are characterized with a Micro Crystal MS1V-T1K crystal with a load capacitance of 12.5 pF. The internal and external load capacitance are chosen to closely match the required 12.5 pF.
- (3) Current for watchdog timer clocked by SMCLK included. ACLK = low frequency crystal operation (XTS = 0, XT1DRIVEx = 0). CPUOFF = 1, SCG0 = 0, SCG1 = 0, OSCOFF = 0 (LPM0); f<sub>ACLK</sub> = 32768 Hz, f<sub>MCLK</sub> = 0 MHz, f<sub>SMCLK</sub> = f<sub>DCO</sub> = 1 MHz
- (4) Current for brownout, high side supervisor (SVS<sub>H</sub>) normal mode included. Low side supervisor and monitors disabled (SVS<sub>L</sub>, SVM<sub>L</sub>). High side monitor disabled (SVM<sub>H</sub>). RAM retention enabled.
- (5) Current for watchdog timer and RTC clocked by ACLK included. ACLK = low frequency crystal operation (XTS = 0, XT1DRIVEx = 0). CPUOFF = 1, SCG0 = 0, SCG1 = 1, OSCOFF = 0 (LPM2); f<sub>ACLK</sub> = 32768 Hz, f<sub>MCLK</sub> = 0 MHz, f<sub>SMCLK</sub> = f<sub>DCO</sub> = 0 MHz; DCO setting = 1 MHz operation, DCO bias generator enabled.
- (6) Current for watchdog timer and RTC clocked by ACLK included. ACLK = low frequency crystal operation (XTS = 0, XT1DRIVEx = 0). CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 0 (LPM3); f<sub>ACLK</sub> = 32768 Hz, f<sub>MCLK</sub> = f<sub>SMCLK</sub> = f<sub>DCO</sub> = 0 MHz
- (7) Current for watchdog timer and RTC clocked by ACLK included. ACLK = VLO.
  - CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 0 (LPM3); f<sub>ACLK</sub> = f<sub>VLO</sub>, f<sub>MCLK</sub> = f<sub>SMCLK</sub> = f<sub>DCO</sub> = 0 MHz
- (8) CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 1 (LPM4);  $f_{DCO} = f_{ACLK} = f_{MCLK} = f_{SMCLK} = 0$  MHz
- (9) Internal regulator disabled. No data retention.
  - CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 1, PMMREGOFF = 1 (LPM4.5); fDCO = fACLK = fMCLK = 0 MHz



## Schmitt-Trigger Inputs – General Purpose I/O(1)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V	Desitive gains input threehold valtege		1.8 V	0.80		1.40	V
V <sub>IT+</sub>	Positive-going input threshold voltage		3 V	1.50		2.10	V
.,	Negative going input threehold valtage		1.8 V	0.45		1.00	V
V <sub>IT</sub>	Negative-going input threshold voltage		3 V	0.75		1.65	V
V	Leavet well-are level-area's (M		1.8 V	0.3		0.85	V
V <sub>hys</sub>	Input voltage hysteresis (V <sub>IT+</sub> – V <sub>IT-</sub> )		3 V	0.4		1.0	V
R <sub>Pull</sub>	Pullup or pulldown resistor <sup>(2)</sup>	For pullup: $V_{IN} = V_{SS}$ For pulldown: $V_{IN} = V_{CC}$		20	35	50	kΩ
$C_{I}$	Input capacitance	$V_{IN} = V_{SS}$ or $V_{CC}$			5		рF

Same parametrics apply to clock input pin when crystal bypass mode is used on XT1 (XIN) or XT2 (XT2IN). Also applies to the RST pin when the pullup or pulldown resistor is enabled.

## Inputs – Ports P1 and P2(1)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	MAX	UNIT
t <sub>(i</sub>	External interrupt timing (2)	Port P1, P2: P1.x to P2.x, External trigger pulse duration to set interrupt flag	2.2 V, 3 V	20		ns

Some devices may contain additional ports with interrupts. See the block diagram and terminal function descriptions.

#### Leakage Current – General Purpose I/O

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	MAX	UNIT
I <sub>lkg(Px.y)</sub>	High-impedance leakage current	(1) (2)	1.8 V, 3 V		±50	nA

The leakage current is measured with  $V_{SS}$  or  $V_{CC}$  applied to the corresponding pin(s), unless otherwise noted.

An external signal sets the interrupt flag every time the minimum interrupt pulse duration t(int) is met. It may be set by trigger signals shorter than t<sub>(int)</sub>.

The leakage of the digital port pins is measured individually. The port pin is selected for input and the pullup/pulldown resistor is



## Outputs - General Purpose I/O (Full Drive Strength)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	MAX	UNIT
		$I_{(OHmax)} = -3 \text{ mA}^{(1)}$	1.8 V	V <sub>CC</sub> - 0.25	$V_{CC}$	
V <sub>OH</sub> High-level output vo		$I_{(OHmax)} = -10 \text{ mA}^{(2)}$	1.0 V	V <sub>CC</sub> - 0.60	$V_{CC}$	]
	High-level output voltage	$I_{(OHmax)} = -5 \text{ mA}^{(1)}$	2.1/	V <sub>CC</sub> - 0.25	$V_{CC}$	
		$I_{(OHmax)} = -15 \text{ mA}^{(2)}$	3 V	V <sub>CC</sub> - 0.60	V <sub>CC</sub>	
		$I_{(OLmax)} = 3 \text{ mA}^{(1)}$	1.8 V	V <sub>SS</sub>	V <sub>SS</sub> + 0.25	V
V		$I_{(OLmax)} = 10 \text{ mA}^{(2)}$	1.6 V	V <sub>SS</sub>	$V_{SS} + 0.60$	
V <sub>OL</sub>	Low-level output voltage	$I_{(OLmax)} = 5 \text{ mA}^{(1)}$	3 V	V <sub>SS</sub>	V <sub>SS</sub> + 0.25	
		I <sub>(OLmax)</sub> = 15 mA <sup>(2)</sup>	3 V	V <sub>SS</sub>	V <sub>SS</sub> + 0.60	

<sup>(1)</sup> The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined should not exceed ±48 mA to hold the maximum voltage drop specified.

#### Outputs - General Purpose I/O (Reduced Drive Strength)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)(1)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	MAX	UNIT
		$I_{(OHmax)} = -1 \text{ mA}^{(2)}$	1.8 V	V <sub>CC</sub> - 0.25	$V_{CC}$	
V <sub>OH</sub> High-level output voltage	$I_{(OHmax)} = -3 \text{ mA}^{(3)}$	1.6 V	V <sub>CC</sub> - 0.60	$V_{CC}$	1 V I	
	$I_{(OHmax)} = -2 \text{ mA}^{(2)}$	201/	V <sub>CC</sub> - 0.25	$V^{CC}$		
		$I_{\text{(OHmax)}} = -6 \text{ mA}^{(3)}$	3.0 V	V <sub>CC</sub> - 0.60	$V_{CC}$	
		$I_{(OLmax)} = 1 \text{ mA}^{(2)}$	4.0.1/	V <sub>SS</sub>	V <sub>SS</sub> + 0.25	V
.,		$I_{(OLmax)} = 3 \text{ mA}^{(3)}$	1.8 V	V <sub>SS</sub>	$V_{SS} + 0.60$	
V <sub>OL</sub>	Low-level output voltage	I <sub>(OLmax)</sub> = 2 mA <sup>(2)</sup>	201/	V <sub>SS</sub>	V <sub>SS</sub> + 0.25	
		I <sub>(OLmax)</sub> = 6 mA <sup>(3)</sup>	3.0 V	V <sub>SS</sub>	V <sub>SS</sub> + 0.60	

<sup>(1)</sup> Selecting reduced drive strength may reduce EMI.

#### Output Frequency – General Purpose I/O

	PARAMETER	TEST CO	ONDITIONS	MIN MAX	UNIT	
	Port output frequency (with load)	P1.6/SMCLK <sup>(1)</sup> <sup>(2)</sup>	V <sub>CC</sub> = 1.8 V, PMMCOREVx = 0	16	MHz	
		P1.6/SWICER (*/ C)	V <sub>CC</sub> = 3 V, PMMCOREVx = 3	25		
( Obel estet (common		P1.0/TA0CLK/ACLK P1.6/SMCLK	V <sub>CC</sub> = 1.8 V, PMMCOREVx = 0	16		
f <sub>Port_CLK</sub>	Clock output frequency	$P2.0/TA1CLK/MCLK$ $C_{L} = 20 \text{ pF}^{(2)}$	V <sub>CC</sub> = 3 V, PMMCOREVx = 3	25	MHz	

<sup>(1)</sup> A resistive divider with 2 x R1 between  $V_{CC}$  and  $V_{SS}$  is used as load. The output is connected to the center tap of the divider. For full drive strength, R1 = 550  $\Omega$ . For reduced drive strength, R1 = 1.6 k $\Omega$ .  $C_L$  = 20 pF is connected to the output to  $V_{SS}$ .

<sup>(2)</sup> The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined should not exceed ±100 mA to hold the maximum voltage drop specified.

<sup>(2)</sup> The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined, should not exceed ±48 mA to hold the maximum voltage drop specified.

<sup>(3)</sup> The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined, should not exceed ±100 mA to hold the maximum voltage drop specified.

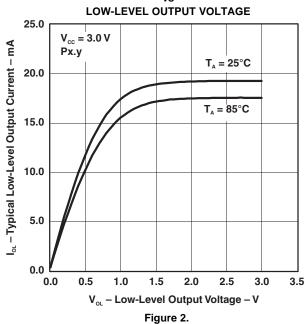
<sup>(2)</sup> The output voltage reaches at least 10% and 90% V<sub>CC</sub> at the specified toggle frequency.



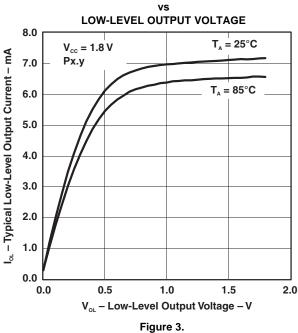
## Typical Characteristics – Outputs, Reduced Drive Strength (PxDS.y = 0)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

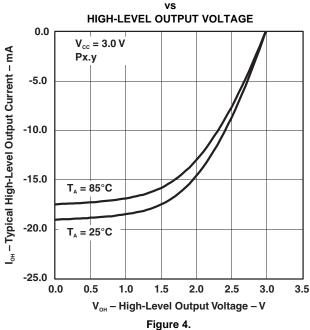




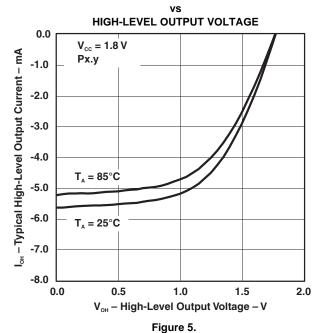
# TYPICAL LOW-LEVEL OUTPUT CURRENT



## TYPICAL HIGH-LEVEL OUTPUT CURRENT



#### TYPICAL HIGH-LEVEL OUTPUT CURRENT

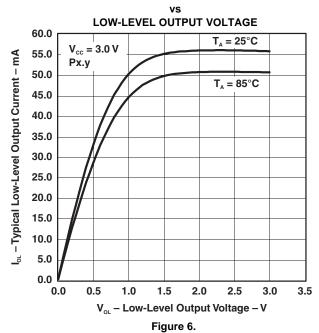




## Typical Characteristics – Outputs, Full Drive Strength (PxDS.y = 1)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

#### TYPICAL LOW-LEVEL OUTPUT CURRENT



#### TYPICAL LOW-LEVEL OUTPUT CURRENT

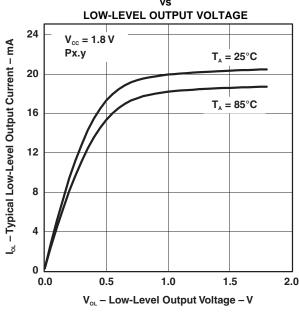
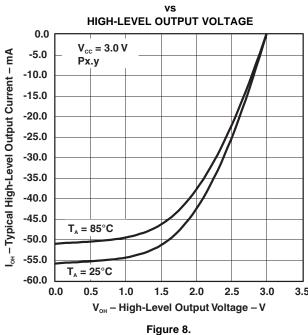


Figure 7.

#### **TYPICAL HIGH-LEVEL OUTPUT CURRENT**



#### TYPICAL HIGH-LEVEL OUTPUT CURRENT

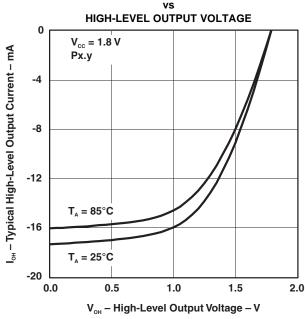


Figure 9.



## Crystal Oscillator, XT1, Low-Frequency Mode<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN TY	P MAX	UNIT	
		$ \begin{aligned} f_{OSC} &= 32768 \text{ Hz, XTS} = 0, \\ \text{XT1BYPASS} &= 0, \text{XT1DRIVEx} = 1, \\ T_A &= 25^{\circ}\text{C} \end{aligned} $		0.07	5		
ΔI <sub>DVCC.LF</sub>	Differential XT1 oscillator crystal current consumption from lowest drive setting, LF mode	$ \begin{aligned} &f_{OSC} = 32768 \text{ Hz, XTS} = 0, \\ &\text{XT1BYPASS} = 0, \text{XT1DRIVEx} = 2, \\ &T_A = 25^{\circ}\text{C} \end{aligned} $	3.0 V	0.17	0	μΑ	
		$\label{eq:fosc} \begin{array}{l} f_{OSC} = 32768 \text{ Hz, XTS} = 0, \\ \text{XT1BYPASS} = 0, \text{XT1DRIVEx} = 3, \\ T_A = 25^{\circ}\text{C} \end{array}$		0.29	0		
f <sub>XT1,LF0</sub>	XT1 oscillator crystal frequency, LF mode	XTS = 0, XT1BYPASS = 0		3276	3	Hz	
f <sub>XT1,LF,SW</sub>	XT1 oscillator logic-level square- wave input frequency, LF mode	XTS = 0, XT1BYPASS = 1 (2) (3)		10 32.76	3 50	kHz	
OA <sub>LF</sub>	Oscillation allowance for	$ \begin{aligned} &XTS = 0, \\ &XT1BYPASS = 0,  XT1DRIVEx = 0, \\ &f_{XT1,LF} = 32768 \; Hz,  C_{L,eff} = 6 \; pF \end{aligned} $		21	0	kΩ	
OALF	LF crystals <sup>(4)</sup>	$\begin{split} XTS &= 0, \\ XT1BYPASS &= 0, XT1DRIVEx = 1, \\ f_{XT1,LF} &= 32768 \text{ Hz}, C_{L,eff} = 12 \text{ pF} \end{split}$		30	0	K22	
		$XTS = 0$ , $XCAPx = 0^{(6)}$	2	2			
C	Integrated effective load	XTS = 0, $XCAPx = 1$		5.	5	pF	
$C_{L,eff}$	capacitance, LF mode <sup>(5)</sup>	XTS = 0, $XCAPx = 2$	8.5		5	ρı	
		XTS = 0, $XCAPx = 3$		12.	)		
	Duty cycle, LF mode	$XTS = 0$ , Measured at ACLK, $f_{XT1,LF} = 32768$ Hz		30	70	%	
f <sub>Fault,LF</sub>	Oscillator fault frequency, LF mode (7)	$XTS = 0^{(8)}$		10	10000	Hz	
t	Startup time I E mode	$\begin{split} f_{OSC} &= 32768 \text{ Hz, XTS} = 0, \\ \text{XT1BYPASS} &= 0, \text{XT1DRIVEx} = 0, \\ T_{A} &= 25^{\circ}\text{C, C}_{L,eff} = 6 \text{ pF} \end{split}$	3.0 V	100	0		
t <sub>START,LF</sub>	Startup time, LF mode	$\begin{aligned} &f_{OSC} = 32768 \text{ Hz, XTS} = 0,\\ &\text{XT1BYPASS} = 0, \text{XT1DRIVEx} = 3,\\ &T_{A} = 25^{\circ}\text{C, C}_{L,\text{eff}} = 12 \text{ pF} \end{aligned}$	3.0 V	50	)	ms	

- (1) To improve EMI on the XT1 oscillator, the following guidelines should be observed.
  - (a) Keep the trace between the device and the crystal as short as possible.
  - (b) Design a good ground plane around the oscillator pins.
  - (c) Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
  - (d) Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
  - (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.
  - (f) If conformal coating is used, ensure that it does not induce capacitive or resistive leakage between the oscillator pins.
- (2) When XT1BYPASS is set, XT1 circuits are automatically powered down. Input signal is a digital square wave with parametrics defined in the Schmitt-trigger Inputs section of this datasheet.
- (3) Maximum frequency of operation of the entire device cannot be exceeded.
- (4) Oscillation allowance is based on a safety factor of 5 for recommended crystals. The oscillation allowance is a function of the XT1DRIVEx settings and the effective load. In general, comparable oscillator allowance can be achieved based on the following guidelines, but should be evaluated based on the actual crystal selected for the application:
  - (a) For XT1DRIVEx = 0,  $C_{L.eff} \le 6 \text{ pF}$ .
  - (b) For XT1DRIVEx = 1, 6 pF  $\leq$  C<sub>L,eff</sub>  $\leq$  9 pF.
  - (c) For XT1DRIVEx = 2, 6 pF  $\leq$  C<sub>L,eff</sub>  $\leq$  10 pF.
  - (d) For XT1DRIVEx = 3,  $C_{L,eff} \ge 6 \text{ pF}$ .
- (5) Includes parasitic bond and package capacitance (approximately 2 pF per pin).
  - Since the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.
- (6) Requires external capacitors at both terminals. Values are specified by crystal manufacturers.
- (7) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- (8) Measured with logic-level input frequency but also applies to operation with crystals.



## Crystal Oscillator, XT1, High-Frequency Mode<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT	
	XT1 oscillator crystal current, HF mode	$ \begin{aligned} &f_{OSC} = 4 \text{ MHz}, \\ &XTS = 1, XOSCOFF = 0, \\ &XT1BYPASS = 0, XT1DRIVEx = 0, \\ &T_A = 25^{\circ}C \end{aligned} $			200			
1		$ \begin{cases} f_{OSC} = 12 \text{ MHz,} \\ \text{XTS} = 1, \text{XOSCOFF} = 0, \\ \text{XT1BYPASS} = 0, \text{XT1DRIVEx} = 1, \\ T_A = 25^{\circ}\text{C} \end{cases} $	- 3.0 V		260			
IDVCC.HF		$ \begin{aligned} &f_{OSC} = 20 \text{ MHz,} \\ &\text{XTS} = 1, \text{ XOSCOFF} = 0, \\ &\text{XT1BYPASS} = 0, \text{ XT1DRIVEx} = 2, \\ &T_A = 25^{\circ}\text{C} \end{aligned} $	3.0 V		325		μА	
		$ \begin{aligned} &f_{OSC} = 32 \text{ MHz,} \\ &\text{XTS} = 1, \text{ XOSCOFF} = 0, \\ &\text{XT1BYPASS} = 0, \text{ XT1DRIVEx} = 3, \\ &T_A = 25^{\circ}\text{C} \end{aligned} $			450			
f <sub>XT1,HF0</sub>	XT1 oscillator crystal frequency, HF mode 0	XTS = 1, XT1BYPASS = 0, $XT1DRIVEx = 0$ <sup>(2)</sup>		4		8	MHz	
f <sub>XT1,HF1</sub>	XT1 oscillator crystal frequency, HF mode 1	XTS = 1, XT1BYPASS = 0, XT1DRIVEx = 1 <sup>(2)</sup>		8		16	MHz	
f <sub>XT1,HF2</sub>	XT1 oscillator crystal frequency, HF mode 2	XTS = 1, XT1BYPASS = 0, XT1DRIVEx = 2 <sup>(2)</sup>		16		24	MHz	
f <sub>XT1,HF3</sub>	XT1 oscillator crystal frequency, HF mode 3	XTS = 1, XT1BYPASS = 0, XT1DRIVEx = 3 <sup>(2)</sup>		24		32	MHz	
f <sub>XT1,HF,SW</sub>	XT1 oscillator logic-level square- wave input frequency, HF mode, bypass mode	XTS = 1, XT1BYPASS = 1 <sup>(3)(2)</sup>		0.7		32	MHz	
		$\begin{split} &XTS = 1,\\ &XT1BYPASS = 0, XT1DRIVEx = 0,\\ &f_{XT1,HF} = 6\;MHz, C_{L,eff} = 15\;pF \end{split}$			450			
OA <sub>HF</sub>	Oscillation allowance for	$\begin{split} XTS &= 1, \\ XT1BYPASS &= 0, XT1DRIVEx = 1, \\ f_{XT1,HF} &= 12 \text{ MHz}, C_{L,eff} = 15 \text{ pF} \end{split}$			320		Ω	
OAHF	HF crystals <sup>(4)</sup>	$\begin{split} &\text{XTS} = 1, \\ &\text{XT1BYPASS} = 0, \text{XT1DRIVEx} = 2, \\ &\text{f}_{\text{XT1,HF}} = 20 \text{ MHz}, \text{C}_{\text{L,eff}} = 15 \text{ pF} \end{split}$			200		12	
		$\begin{tabular}{ll} XTS = 1, \\ XT1BYPASS = 0, XT1DRIVEx = 3, \\ f_{XT1,HF} = 32 \ MHz, C_{L,eff} = 15 \ pF \end{tabular}$			200			
	Orași de LIE	$\begin{split} f_{OSC} &= 6 \text{ MHz, XTS} = 1, \\ \text{XT1BYPASS} &= 0, \text{XT1DRIVEx} = 0, \\ T_{A} &= 25^{\circ}\text{C, C}_{L,\text{eff}} = 15 \text{ pF} \end{split}$	3.0 V		0.5			
t <sub>START,HF</sub>	Startup time, HF mode		3.0 V		0.3		ms	

- (1) To improve EMI on the XT1 oscillator the following guidelines should be observed.
  - (a) Keep the traces between the device and the crystal as short as possible.
  - (b) Design a good ground plane around the oscillator pins.
  - (c) Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
  - (d) Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
  - (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.

    (f) If conformal coating is used, ensure that it does not induce capacitive or resistive leakage between the oscillator pins.
- (2) This represents the maximum frequency that can be input to the device externally. Maximum frequency achievable on the device operation is based on the frequencies present on ACLK, MCLK, and SMCLK cannot be exceed for a given range of operation.
- (3) When XT1BYPASS is set, XT1 circuits are automatically powered down. Input signal is a digital square wave with parametrics defined in the Schmitt-trigger Inputs section of this datasheet.
- (4) Oscillation allowance is based on a safety factor of 5 for recommended crystals.



# Crystal Oscillator, XT1, High-Frequency Mode<sup>(1)</sup> (continued)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
$C_{L,eff}$	Integrated effective load capacitance, HF mode (5) (6)	XTS = 1			1		pF
	Duty cycle, HF mode	XTS = 1, Measured at ACLK, f <sub>XT1,HF2</sub> = 20 MHz		40	50	60	%
f <sub>Fault,HF</sub>	Oscillator fault frequency, HF mode <sup>(7)</sup>	XTS = 1 <sup>(8)</sup>		30		300	kHz

- (5) Includes parasitic bond and package capacitance (approximately 2 pF per pin).
  - Since the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.
- (6) Requires external capacitors at both terminals. Values are specified by crystal manufacturers. In general, an effective load capacitance of up to 18 pF can be supported.
- (7) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- (8) Measured with logic-level input frequency but also applies to operation with crystals.

#### Crystal Oscillator, XT2

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (1) (2)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT	
		$\begin{aligned} f_{OSC} &= 4 \text{ MHz, XT2OFF} = 0, \\ \text{XT2BYPASS} &= 0, \text{XT2DRIVEx} = 0, \\ T_{A} &= 25^{\circ}\text{C} \end{aligned}$			200			
	XT2 oscillator crystal current consumption	$ \begin{aligned} &f_{OSC} = 12 \text{ MHz, } XT2OFF = 0, \\ &XT2BYPASS = 0, XT2DRIVEx = 1, \\ &T_A = 25^{\circ}C \end{aligned} $	3.0 V		260		μΑ	
IDVCC.XT2		$ \begin{aligned} &f_{OSC} = 20 \text{ MHz, } XT2OFF = 0, \\ &XT2BYPASS = 0, XT2DRIVEx = 2, \\ &T_A = 25^{\circ}C \end{aligned} $			325			
		$ \begin{aligned} &f_{OSC} = 32 \text{ MHz, } XT2OFF = 0, \\ &XT2BYPASS = 0, XT2DRIVEx = 3, \\ &T_A = 25^{\circ}C \end{aligned} $			450			
f <sub>XT2,HF0</sub>	XT2 oscillator crystal frequency, mode 0	XT2DRIVEx = 0, XT2BYPASS = 0 <sup>(3)</sup>		4		8	MHz	
f <sub>XT2,HF1</sub>	XT2 oscillator crystal frequency, mode 1	XT2DRIVEx = 1, XT2BYPASS = 0 <sup>(3)</sup>		8		16	MHz	
f <sub>XT2,HF2</sub>	XT2 oscillator crystal frequency, mode 2	XT2DRIVEx = 2, XT2BYPASS = 0 <sup>(3)</sup>		16		24	MHz	
f <sub>XT2,HF3</sub>	XT2 oscillator crystal frequency, mode 3	XT2DRIVEx = 3, XT2BYPASS = 0 <sup>(3)</sup>		24		32	MHz	
f <sub>XT2,HF,SW</sub>	XT2 oscillator logic-level square- wave input frequency, bypass mode	XT2BYPASS = 1 (4) (3)		0.7		32	MHz	

(a) Keep the traces between the device and the crystal as short as possible.

(b) Design a good ground plane around the oscillator pins.

- (d) Avoid running PCB traces underneath or adjacent to the XT2IN and XT2OUT pins.
- (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XT2IN and XT2OUT pins.
- (f) If conformal coating is used, ensure that it does not induce capacitive or resistive leakage between the oscillator pins.
- (3) This represents the maximum frequency that can be input to the device externally. Maximum frequency achievable on the device operation is based on the frequencies present on ACLK, MCLK, and SMCLK cannot be exceed for a given range of operation.
- (4) When XT2BYPASS is set, the XT2 circuit is automatically powered down. Input signal is a digital square wave with parametrics defined in the Schmitt-trigger Inputs section of this datasheet.

<sup>(1)</sup> Requires external capacitors at both terminals. Values are specified by crystal manufacturers. In general, an effective load capacitance of up to 18 pF can be supported.

<sup>(2)</sup> To improve EMI on the XT2 oscillator the following guidelines should be observed.

<sup>(</sup>c) Prevent crosstalk from other clock or data lines into oscillator pins XT2IN and XT2OUT.



## Crystal Oscillator, XT2 (continued)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)<sup>(1)</sup> (2)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		$XT2DRIVEx = 0$ , $XT2BYPASS = 0$ , $f_{XT2,HF0} = 6$ MHz, $C_{L,eff} = 15$ pF			450		
OA <sub>HF</sub>	Oscillation allowance for	$XT2DRIVEx = 1$ , $XT2BYPASS = 0$ , $f_{XT2,HF1} = 12$ MHz, $C_{L,eff} = 15$ pF			320		Ω
	HF crystals <sup>(5)</sup>	$XT2DRIVEx = 2$ , $XT2BYPASS = 0$ , $f_{XT2,HF2} = 20$ MHz, $C_{L,eff} = 15$ pF			200		77
		$XT2DRIVEx = 3$ , $XT2BYPASS = 0$ , $f_{XT2,HF3} = 32$ MHz, $C_{L,eff} = 15$ pF			200		
	Startup time	$\begin{split} f_{OSC} &= 6 \text{ MHz} \\ \text{XT2BYPASS} &= 0, \text{XT2DRIVEx} = 0, \\ T_{A} &= 25^{\circ}\text{C}, \text{ C}_{\text{L,eff}} = 15 \text{ pF} \end{split}$	3.0 V		0.5		ms
<sup>t</sup> START,HF		$f_{OSC}$ = 20 MHz XT2BYPASS = 0, XT2DRIVEx = 2, $T_A$ = 25°C, $C_{L,eff}$ = 15 pF	3.0 V		0.3		1115
$C_{L,eff}$	Integrated effective load capacitance, HF mode (6)(1)				1		pF
	Duty cycle	Measured at ACLK, f <sub>XT2,HF2</sub> = 20 MHz		40	50	60	%
f <sub>Fault,HF</sub>	Oscillator fault frequency <sup>(7)</sup>	XT2BYPASS = 1 <sup>(8)</sup>		30		300	kHz

- Oscillation allowance is based on a safety factor of 5 for recommended crystals.
- Includes parasitic bond and package capacitance (approximately 2 pF per pin). Since the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.
- (7) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- Measured with logic-level input frequency but also applies to operation with crystals.

## Internal Very-Low-Power Low-Frequency Oscillator (VLO)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
$f_{VLO}$	VLO frequency	Measured at ACLK	1.8 V to 3.6 V	6	9.4	14	kHz
$df_{VLO}/d_{T}$	VLO frequency temperature drift	Measured at ACLK <sup>(1)</sup>	1.8 V to 3.6 V		0.5		%/°C
$df_{VLO}/dV_{CC}$	VLO frequency supply voltage drift	Measured at ACLK <sup>(2)</sup>	1.8 V to 3.6 V		4		%/V
	Duty cycle	Measured at ACLK	1.8 V to 3.6 V	40	50	60	%

- $(1) \quad \text{Calculated using the box method: } (\text{MAX}(\text{-40 to 85}^{\circ}\text{C}) \text{MIN}(\text{-40 to 85}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ \text{MIN}(\text{-40 to 85}^{\circ}\text{C}) \ / \ (85^{\circ}\text{C} (\text{-40}^{\circ}\text{C})) \ / \ (85^{\circ}\text{C} (\text{-4$
- Calculated using the box method: (MAX(1.8 to 3.6 V) MIN(1.8 to 3.6 V)) / MIN(1.8 to 3.6 V) / (3.6 V 1.8 V)

#### Internal Reference, Low-Frequency Oscillator (REFO)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN TY	P MAX	UNIT
I <sub>REFO</sub>	REFO oscillator current consumption	T <sub>A</sub> = 25°C	1.8 V to 3.6 V		3	μΑ
f <sub>REFO</sub>	REFO frequency calibrated	Measured at ACLK	1.8 V to 3.6 V	3276	88	Hz
	REFO absolute tolerance calibrated	Full temperature range	1.8 V to 3.6 V		±3.5	%
		T <sub>A</sub> = 25°C	3 V		±1.5	%
$df_{REFO}/d_{T}$	REFO frequency temperature drift	Measured at ACLK <sup>(1)</sup>	1.8 V to 3.6 V	0.0	)1	%/°C
$df_{REFO}/dV_{CC}$	REFO frequency supply voltage drift	Measured at ACLK (2)	1.8 V to 3.6 V	1	.0	%/V
	Duty cycle	Measured at ACLK	1.8 V to 3.6 V	40 !	50 60	%
t <sub>START</sub>	REFO startup time	40%/60% duty cycle	1.8 V to 3.6 V	2	25	μs

- Calculated using the box method: (MAX(-40 to 85°C) MIN(-40 to 85°C)) / MIN(-40 to 85°C) / (85°C (-40°C)) Calculated using the box method: (MAX(1.8 to 3.6 V) MIN(1.8 to 3.6 V)) / MIN(1.8 to 3.6 V) / (3.6 V 1.8 V)



#### DCO Frequency

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
f <sub>DCO(0,0)</sub>	DCO frequency (0, 0) <sup>(1)</sup>	DCORSELx = 0, DCOx = 0, MODx = 0	0.07		0.20	MHz
f <sub>DCO(0,31)</sub>	DCO frequency (0, 31) <sup>(1)</sup>	DCORSELx = 0, $DCOx = 31$ , $MODx = 0$	0.70		1.70	MHz
f <sub>DCO(1,0)</sub>	DCO frequency (1, 0) <sup>(1)</sup>	DCORSELx = 1, $DCOx = 0$ , $MODx = 0$	0.15		0.36	MHz
f <sub>DCO(1,31)</sub>	DCO frequency (1, 31) <sup>(1)</sup>	DCORSELx = 1, DCOx = 31, MODx = 0	1.47		3.45	MHz
f <sub>DCO(2,0)</sub>	DCO frequency (2, 0) <sup>(1)</sup>	DCORSELx = 2, $DCOx = 0$ , $MODx = 0$	0.32		0.75	MHz
f <sub>DCO(2,31)</sub>	DCO frequency (2, 31) <sup>(1)</sup>	DCORSELx = 2, DCOx = 31, MODx = 0	3.17		7.38	MHz
f <sub>DCO(3,0)</sub>	DCO frequency (3, 0) <sup>(1)</sup>	DCORSELx = 3, $DCOx = 0$ , $MODx = 0$	0.64		1.51	MHz
f <sub>DCO(3,31)</sub>	DCO frequency (3, 31) <sup>(1)</sup>	DCORSELx = 3, DCOx = 31, MODx = 0	6.07		14.0	MHz
f <sub>DCO(4,0)</sub>	DCO frequency (4, 0) <sup>(1)</sup>	DCORSELx = 4, DCOx = 0, MODx = 0	1.3		3.2	MHz
f <sub>DCO(4,31)</sub>	DCO frequency (4, 31) <sup>(1)</sup>	DCORSELx = 4, DCOx = 31, MODx = 0	12.3		28.2	MHz
f <sub>DCO(5,0)</sub>	DCO frequency (5, 0) <sup>(1)</sup>	DCORSELx = 5, DCOx = 0, MODx = 0	2.5		6.0	MHz
f <sub>DCO(5,31)</sub>	DCO frequency (5, 31) <sup>(1)</sup>	DCORSELx = 5, DCOx = 31, MODx = 0	23.7		54.1	MHz
f <sub>DCO(6,0)</sub>	DCO frequency (6, 0) <sup>(1)</sup>	DCORSELx = 6, DCOx = 0, MODx = 0	4.6		10.7	MHz
f <sub>DCO(6,31)</sub>	DCO frequency (6, 31) <sup>(1)</sup>	DCORSELx = 6, DCOx = 31, MODx = 0	39.0		88.0	MHz
f <sub>DCO(7,0)</sub>	DCO frequency (7, 0) <sup>(1)</sup>	DCORSELx = 7, DCOx = 0, MODx = 0	8.5		19.6	MHz
f <sub>DCO(7,31)</sub>	DCO frequency (7, 31) <sup>(1)</sup>	DCORSELx = 7, DCOx = 31, MODx = 0	60		135	MHz
S <sub>DCORSEL</sub>	Frequency step between range DCORSEL and DCORSEL + 1	$S_{RSEL} = f_{DCO(DCORSEL+1,DCO)}/f_{DCO(DCORSEL,DCO)}$	1.2		2.3	ratio
S <sub>DCO</sub>	Frequency step between tap DCO and DCO + 1	$S_{DCO} = f_{DCO(DCORSEL,DCO+1)}/f_{DCO(DCORSEL,DCO)}$	1.02		1.12	ratio
	Duty cycle	Measured at SMCLK	40	50	60	%
df <sub>DCO</sub> /dT	DCO frequency temperature drift <sup>(2)</sup>	f <sub>DCO</sub> = 1 MHz		0.1		%/°C
df <sub>DCO</sub> /dV <sub>CC</sub>	DCO frequency voltage drift <sup>(3)</sup>	f <sub>DCO</sub> = 1 MHz		1.9		%/V

- (1) When selecting the proper DCO frequency range (DCORSELx), the target DCO frequency,  $f_{DCO}$ , should be set to reside within the range of  $f_{DCO(n, 0),MAX} \le f_{DCO} \le f_{DCO(n, 31),MIN}$ , where  $f_{DCO(n, 0),MAX}$  represents the maximum frequency specified for the DCO frequency, range n, tap 0 (DCOx = 0) and  $f_{DCO(n, 31),MIN}$  represents the minimum frequency specified for the DCO frequency, range n, tap 31 (DCOx = 31). This ensures that the target DCO frequency resides within the range selected. It should also be noted that if the actual  $f_{DCO}$ frequency for the selected range causes the FLL or the application to select tap 0 or 31, the DCO fault flag is set to report that the selected range is at its minimum or maximum tap setting.
- Calculated using the box method: (MAX(-40 to  $85^{\circ}$ C) MIN(-40 to  $85^{\circ}$ C)) / MIN(-40 to  $85^{\circ}$ C) / ( $85^{\circ}$ C ( $-40^{\circ}$ C)) Calculated using the box method: (MAX(1.8 to 3.6 V) MIN(1.8 to 3.6 V)) / MIN(1.8 to 3.6 V) / (3.6 V 1.8 V)

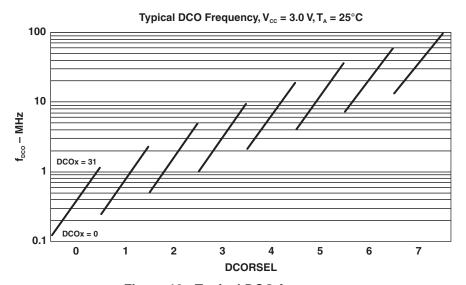


Figure 10. Typical DCO frequency



## PMM, Brown-Out Reset (BOR)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V(DV <sub>CC</sub> _BOR_IT-)	BOR <sub>H</sub> on voltage, DV <sub>CC</sub> falling level	$\mid dDV_{CC}/d_t \mid < 3 \text{ V/s}$			1.45	٧
V(DV <sub>CC</sub> _BOR_IT+)	BOR <sub>H</sub> off voltage, DV <sub>CC</sub> rising level	dDV <sub>CC</sub> /d <sub>t</sub>   < 3 V/s	0.80	1.30	1.50	٧
V(DV <sub>CC</sub> _BOR_hys)	BOR <sub>H</sub> hysteresis		60		250	mV
t <sub>RESET</sub>	Pulse length required at RST/NMI pin to accept a reset		2			μs

# **PMM**, Core Voltage

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>CORE3</sub> (AM)	Core voltage, active mode, PMMCOREV = 3	2.4 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.90		V
V <sub>CORE2</sub> (AM)	Core voltage, active mode, PMMCOREV = 2	2.2 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.80		V
V <sub>CORE1</sub> (AM)	Core voltage, active mode, PMMCOREV = 1	2.0 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.60		V
V <sub>CORE0</sub> (AM)	Core voltage, active mode, PMMCOREV = 0	1.8 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.40		V
V <sub>CORE3</sub> (LPM)	Core voltage, low-current mode, PMMCOREV = 3	2.4 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.94		V
V <sub>CORE2</sub> (LPM)	Core voltage, low-current mode, PMMCOREV = 2	2.2 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.84		V
V <sub>CORE1</sub> (LPM)	Core voltage, low-current mode, PMMCOREV = 1	2.0 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.64		V
V <sub>CORE0</sub> (LPM)	Core voltage, low-current mode, PMMCOREV = 0	1.8 V ≤ DV <sub>CC</sub> ≤ 3.6 V		1.44		V



## PMM, SVS High Side

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		SVSHE = 0, DV <sub>CC</sub> = 3.6 V		0		nA
I <sub>(SVSH)</sub>	SVS current consumption	SVSHE = 1, DV <sub>CC</sub> = 3.6 V, SVSHFP = 0		200		nA
		SVSHE = 1, DV <sub>CC</sub> = 3.6 V, SVSHFP = 1		1.5		μA
		SVSHE = 1, SVSHRVL = 0	1.57	1.68	1.78	
V <sub>(SVSH_IT-)</sub>	CVC on voltage level(1)	SVSHE = 1, SVSHRVL = 1	1.79	1.88	1.98	V
	SVS <sub>H</sub> on voltage level <sup>(1)</sup>	SVSHE = 1, SVSHRVL = 2	1.98	2.08	2.21	\ \ \
		SVSHE = 1, SVSHRVL = 3	2.10	2.18	2.31	
		SVSHE = 1, SVSMHRRL = 0	1.62	1.74	1.85	
	SVS <sub>H</sub> off voltage level <sup>(1)</sup>	SVSHE = 1, SVSMHRRL = 1	1.88	1.94	2.07	V
		SVSHE = 1, SVSMHRRL = 2	2.07	2.14	2.28	
\ /		SVSHE = 1, SVSMHRRL = 3	2.20	2.30	2.42	
V <sub>(SVSH_IT+)</sub>	SVS <sub>H</sub> on voltage level(1)	SVSHE = 1, SVSMHRRL = 4	2.32	2.40	2.55	
		SVSHE = 1, SVSMHRRL = 5	2.52	2.70	2.88	
		SVSHE = 1, SVSMHRRL = 6	2.90	3.10	3.23	
		SVSHE = 1, SVSMHRRL = 7	2.90	3.10	3.23	
	CVC representation delevi-	SVSHE = 1, dV <sub>DVCC</sub> /dt = 10 mV/µs, SVSHFP = 1		2.5		
t <sub>pd(SVSH)</sub>	SVS <sub>H</sub> propagation delay	SVSHE = 1, $dV_{DVCC}/dt = 1 \text{ mV/}\mu\text{s}$ , SVSHFP = 0		20		μs
	CVC an an aff dalay time	SVSHE = 0 → 1, SVSHFP = 1		12.5		
t <sub>(SVSH)</sub>	SVS <sub>□</sub> on or off delay time	SVSHE = $0 \rightarrow 1$ , SVSHFP = $0$		100		μs
dV <sub>DVCC</sub> /dt	DV <sub>CC</sub> rise time		0		1000	V/s

<sup>(1)</sup> The SVS<sub>H</sub> settings available depend on the VCORE (PMMCOREVx) setting. See the *Power Management Module and Supply Voltage Supervisor* chapter in the *MSP430x5xx and MSP430x6xx Family User's Guide* (SLAU208) on recommended settings and use.

## PMM, SVM High Side

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		SVMHE = 0, DV <sub>CC</sub> = 3.6 V		0		nA
I <sub>(SVMH)</sub>	SVM <sub>H</sub> current consumption	SVMHE= 1, DV <sub>CC</sub> = 3.6 V, SVMHFP = 0		200		nA
		SVMHE = 1, DV <sub>CC</sub> = 3.6 V, SVMHFP = 1		1.5		μΑ
 		SVMHE = 1, SVSMHRRL = 0	1.62	1.74	1.85	
		SVMHE = 1, SVSMHRRL = 1	1.88	1.94	2.07	
		SVMHE = 1, SVSMHRRL = 2	2.07	2.14	2.28	
		SVMHE = 1, SVSMHRRL = 3	2.20	2.30	2.42	
V <sub>(SVMH)</sub>	SVM <sub>H</sub> on or off voltage level <sup>(1)</sup>	SVMHE = 1, SVSMHRRL = 4	2.32	2.40	2.55	V
		SVMHE = 1, SVSMHRRL = 5	2.52	2.70	2.88	
		SVMHE = 1, SVSMHRRL = 6	2.90	3.10	3.23	
		SVMHE = 1, SVSMHRRL = 7	2.90	3.10	3.23	
		SVMHE = 1, SVMHOVPE = 1		3.75		
	CVM propagation dolor	SVMHE = 1, dV <sub>DVCC</sub> /dt = 10 mV/µs, SVMHFP = 1		2.5		
t <sub>pd(SVMH)</sub>	SVM <sub>H</sub> propagation delay	SVMHE = 1, dV <sub>DVCC</sub> /dt = 1 mV/µs, SVMHFP = 0		20		μs
	C)/M are an eff delevations	SVMHE = 0 → 1, SVMHFP = 1		12.5		
t <sub>(SVMH)</sub>	SVM <sub>H</sub> on or off delay time	SVMHE = $0 \rightarrow 1$ , SVMHFP = $0$		100		μs

<sup>(1)</sup> The SVM<sub>H</sub> settings available depend on the VCORE (PMMCOREVx) setting. See the Power Management Module and Supply Voltage Supervisor chapter in the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208) on recommended settings and use.



#### PMM, SVS Low Side

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>(SVSL)</sub>		SVSLE = 0, PMMCOREV = 2		0		nA
	SVS <sub>L</sub> current consumption	SVSLE = 1, PMMCOREV = 2, SVSLFP = 0		200		nA
		SVSLE = 1, PMMCOREV = 2, SVSLFP = 1		1.5		μΑ
t <sub>pd(SVSL)</sub>	SVS <sub>L</sub> propagation delay	SVSLE = 1, dV <sub>CORE</sub> /dt = 10 mV/µs, SVSLFP = 1		2.5		
		SVSLE = 1, dV <sub>CORE</sub> /dt = 1 mV/µs, SVSLFP = 0		20		μs
t <sub>(SVSL)</sub>	SVS <sub>1</sub> on or off delay time	SVSLE = $0 \rightarrow 1$ , $dV_{CORE}/dt = 10$ mV/ $\mu$ s, SVSLFP = 1		12.5		
		SVSLE = $0 \rightarrow 1$ , $dV_{CORE}/dt = 1$ mV/ $\mu$ s, SVSLFP = $0$		100		μs

#### PMM, SVM Low Side

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		SVMLE = 0, PMMCOREV = 2		0		nA
I <sub>(SVML)</sub>	SVM <sub>L</sub> current consumption	SVMLE= 1, PMMCOREV = 2, SVMLFP = 0		200		nA
		SVMLE= 1, PMMCOREV = 2, SVMLFP = 1		1.5		μΑ
t <sub>pd(SVML)</sub>	SVM <sub>L</sub> propagation delay	SVMLE = 1, dV <sub>CORE</sub> /dt = 10 mV/µs, SVMLFP = 1		2.5		
		SVMLE = 1, $dV_{CORE}/dt = 1 \text{ mV/}\mu s$ , SVMLFP = 0		20		μs
	SVM <sub>i</sub> on or off delay time	SVMLE = $0 \rightarrow 1$ , $dV_{CORE}/dt = 10 \text{ mV/}\mu\text{s}$ , $SVMLFP = 1$		12.5		
t <sub>(SVML)</sub>		SVMLE = $0 \rightarrow 1$ , $dV_{CORE}/dt = 1 \text{ mV/}\mu\text{s}$ , SVMLFP = $0$		100		μs

#### Wake-Up From Low-Power Modes and Reset

	PARAMETER	TEST CONDITIO	NS	MIN	TYP	MAX	UNIT
	Wake-up time from LPM2,	PMMCOREV = SVSMLRRL = n	f <sub>MCLK</sub> ≥ 4.0 MHz		3.5	7.5	
WAKE-UP-FAST LPM3, or LPM4 to active mode (1) (where n = 0, 1, 2, or 3), SVSLFP = 1	1.0 MHz < f <sub>MCLK</sub> < 4.0 MHz		4.5	9	μs		
twake-up-slow	Wake-up time from LPM2, LPM3 or LPM4 to active mode <sup>(2)</sup>	PMMCOREV = SVSMLRRL = n (where n = 0, 1, 2, or 3), SVSLFP = 0			150	165	μs
twake-up-lpm5	Wake-up time from LPM4.5 to active mode (3)				2	3	ms
t <sub>WAKE-UP-RESET</sub>	Wake-up time from RST or BOR event to active mode (3)				2	3	ms

- (1) This value represents the time from the wakeup event to the first active edge of MCLK. The wakeup time depends on the performance mode of the low side supervisor (SVS<sub>L</sub>) and low side monitor (SVM<sub>L</sub>). Fastest wakeup times are possible with SVS<sub>L</sub> and SVM<sub>L</sub> in full performance mode or disabled when operating in AM, LPM0, and LPM1. Various options are available for SVS<sub>L</sub> and SVM<sub>L</sub> while operating in LPM2, LPM3, and LPM4. See the *Power Management Module and Supply Voltage Supervisor* chapter in the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208).
- (2) This value represents the time from the wakeup event to the first active edge of MCLK. The wakeup time depends on the performance mode of the low side supervisor (SVS<sub>L</sub>) and low side monitor (SVM<sub>L</sub>). In this case, the SVS<sub>L</sub> and SVM<sub>L</sub> are in normal mode (low current) mode when operating in AM, LPM0, and LPM1. Various options are available for SVS<sub>L</sub> and SVM<sub>L</sub> while operating in LPM2, LPM3, and LPM4. See the *Power Management Module and Supply Voltage Supervisor* chapter in the *MSP430x5xx and MSP430x6xx Family User's Guide* (SLAU208).
- (3) This value represents the time from the wakeup event to the reset vector execution.



#### Timer\_A

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
$f_{TA}$	Timer_A input clock frequency	Internal: SMCLK, ACLK, External: TACLK, Duty cycle = 50% ± 10%	1.8 V/ 3.0 V			25	MHz
t <sub>TA,cap</sub>	Timer_A capture timing	All capture inputs, Minimum pulse duration required for capture	1.8 V/ 3.0 V	20			ns

#### Timer\_B

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
f <sub>TB</sub>	Timer_B input clock frequency	Internal: SMCLK, ACLK, External: TBCLK, Duty cycle = 50% ± 10%	1.8 V/ 3.0 V			25	MHz
t <sub>TB,cap</sub>	Timer_B capture timing	All capture inputs, Minimum pulse duration required for capture	1.8 V/ 3.0 V	20			ns

## **USCI (UART Mode) Recommended Operating Conditions**

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
f <sub>USCI</sub>	USCI input clock frequency	Internal: SMCLK, ACLK, External: UCLK, Duty cycle = 50% ± 10%	- 55			f <sub>SYSTEM</sub>	MHz
f <sub>BITCLK</sub>	BITCLK clock frequency (equals baud rate in MBaud)					1	MHz

## **USCI (UART Mode)**

0 117 0 1 0 1			,	
PARAMETER	V <sub>CC</sub>	MIN	TYP MAX	UNIT
t LIART receive dealitch time (1)		50	600	20
UART receive deglitch time <sup>(1)</sup>	3 V	50	600	ns

<sup>(1)</sup> Pulses on the UART receive input (UCxRX) that are shorter than the UART receive deglitch time are suppressed. To ensure that pulses are correctly recognized, their duration should exceed the maximum specification of the deglitch time.



## **USCI (SPI Master Mode) Recommended Operating Conditions**

PARAMETER	CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
f <sub>USCI</sub> USCI input clock frequency	Internal: SMCLK, ACLK Duty cycle = 50% ± 10%				f <sub>SYSTEM</sub>	MHz

#### **USCI (SPI Master Mode)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (see Note (1), Figure 11 and Figure 12)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP M	AX UNIT
f <sub>USCI</sub>	USCI input clock frequency	SMCLK, ACLK, Duty cycle = 50% ± 10%			f <sub>SYS</sub>	<sub>EM</sub> MHz
		PMMCOREV = 0	1.8 V	55		20
	SOMI input data setup time	PIVIVICOREV = 0	3.0 V	38		ns
t <sub>SU,MI</sub>		PMMCOREV = 3	2.4 V	30		
		PIVIIVICOREV = 3	3.0 V	25		ns
	SOMI input data hold time	PMMCOREV = 0	1.8 V	0		
		PIVIIVICOREV = 0	3.0 V	0		ns
t <sub>HD,MI</sub>		PMMCOREV = 3	2.4 V	0		
			3.0 V	0		ns
		UCLK edge to SIMO valid,	1.8 V			20
	OIMO	C <sub>L</sub> = 20 pF, PMMCOREV = 0	3.0 V			18 ns
t <sub>VALID,MO</sub>	SIMO output data valid time <sup>(2)</sup>	UCLK edge to SIMO valid,	2.4 V			16
		$C_L = 20 \text{ pF}, PMMCOREV = 3$	3.0 V			15 ns
t <sub>HD,MO</sub>		0 00 = F PMMOOREV	1.8 V	-10		
	OIMO control data haddisaa (3)	$C_L = 20 \text{ pF}, PMMCOREV = 0$	3.0 V	-8		ns
	SIMO output data hold time <sup>(3)</sup>	0 00 5 5144005511 0	2.4 V	-10		
		$C_L = 20 \text{ pF}, PMMCOREV = 3$	3.0 V	-8		ns

 $f_{UCXCLK} = 1/2t_{LO/HI} \text{ with } t_{LO/HI} \ge \max(t_{VALID,MO(USCI)} + t_{SU,SI(Slave)}, t_{SU,MI(USCI)} + t_{VALID,SO(Slave)}).$  For the slave's parameters  $t_{SU,SI(Slave)}$  and  $t_{VALID,SO(Slave)}$ , see the SPI parameters of the attached slave. Specifies the time to drive the next valid data to the SIMO output after the output changing UCLK clock edge. See the timing diagrams

in Figure 11 and Figure 12.

Specifies how long data on the SIMO output is valid after the output changing UCLK clock edge. Negative values indicate that the data on the SIMO output can become invalid before the output changing clock edge observed on UCLK. See the timing diagrams in Figure 11 and Figure 12.



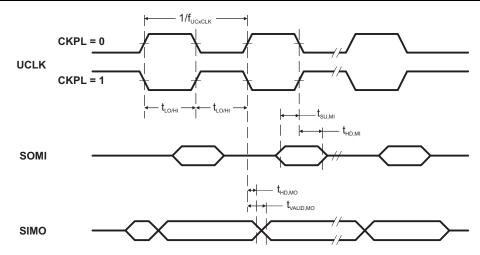


Figure 11. SPI Master Mode, CKPH = 0

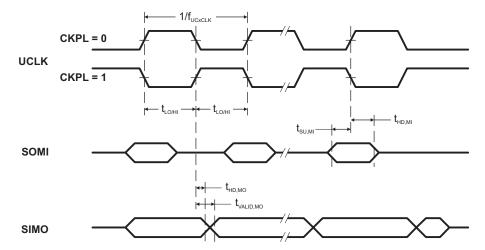


Figure 12. SPI Master Mode, CKPH = 1



## **USCI (SPI Slave Mode)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (see Note (1), Figure 13 and Figure 14)

t <sub>STE,LEAD </sub>   STE lead time, STE low to clock   PMMCOREV = 0   1.8 V   7   3.0 V   8   8   1.8 V   7   7   7   7   7   7   7   7   7		PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
t <sub>STE,LEAD</sub> STE lead time, STE low to clock  PMMCOREV = 3  3.0 ∨ 8  2.4 ∨ 7  3.0 ∨ 6  1.8 ∨ 3  3.0 ∨ 3  PMMCOREV = 0  PMMCOREV = 3  3.0 ∨ 3  2.4 ∨ 3  3.0 ∨ 3  PMMCOREV = 0  1.8 ∨ 66  3.0 ∨ 50  3.0 ∨ 30  PMMCOREV = 3  2.4 ∨ 36  3.0 ∨ 30  PMMCOREV = 3  3.0 ∨ 30  PMMCOREV = 3  3.0 ∨ 30  3.0 ∨ 30  PMMCOREV = 3  3.0 ∨ 30  3.0 ∨ 30  PMMCOREV = 3  3.0 ∨ 30  3.0 ∨ 30  1.8 ∨ 30  1.8 ∨ 30  PMMCOREV = 3  3.0 ∨ 5  1.8 ∨ 30  2.4 ∨ 16  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 5  3.0 ∨ 5  1.8 ∨ 6  1.8 ∨ 6  1.8 ∨ 6  1.8 ∨ 76			DMMCODEV 0	1.8 V	11			
PMMCOREV = 3   2.4 V   7   3.0 V   6   6   6   6   6   6   6   6   6		OTE lead the OTE levels also	PMMCOREV = 0	3.0 V	8			ns
tstellag time, Last clock to STE high  tstellag time, Last clock to STE high  tstellag time, Last clock to STE high  PMMCOREV = 3  PMMCOREV = 3  3.0 V 3  2.4 V 3  3.0 V 3  3.0 V 3  FMMCOREV = 0  PMMCOREV = 0  PMMCOREV = 3  3.0 V 50  3.0 V 50  3.0 V 30  FMMCOREV = 3  Testellag time, STE low to SOMI data out  Testellag time, STE high to SOMI high impedance  PMMCOREV = 3  PMMCOREV = 0  Testellag time, STE high to SOMI high impedance  PMMCOREV = 0  Testellag time, STE high to SOMI high impedance  PMMCOREV = 3  Testellag time, Last clock to STE high  PMMCOREV = 3  Testellag time, Last clock to STE high  PMMCOREV = 3  Testellag time, Last clock to STE high  Testellag time, Last clock to STE high to SOMI valid, Clast clock to SOMI valid, Cl	TE,LEAD	STE lead time, STE low to clock	DMM400DEV/ 0	2.4 V	7			
Name			PMMCOREV = 3	3.0 V	6			ns
t <sub>STE,LAG</sub> STE lag time, Last clock to STE high  PMMCOREV = 3  PMMCOREV = 3  3.0 ∨ 3  3.0 ∨ 3  1.8 ∨ 66  3.0 ∨ 3  PMMCOREV = 0  3.0 ∨ 3  PMMCOREV = 0  PMMCOREV = 3  3.0 ∨ 3  2.4 ∨ 36  3.0 ∨ 36			DMMCOREY 0	1.8 V	3			
PMMCOREV = 3  2.4 V 3 3.0 V 3  PMMCOREV = 0  1.8 V 66 3.0 V 50  PMMCOREV = 3  PMMCOREV = 3  2.4 V 36 3.0 V 30  PMMCOREV = 3  2.4 V 36 3.0 V 30  PMMCOREV = 3  3.0 V 30  3.0 V 30  PMMCOREV = 3  3.0 V 30  3.0 V 23  PMMCOREV = 0  PMMCOREV = 0  PMMCOREV = 3  2.4 V 16 3.0 V 23  PMMCOREV = 3  2.4 V 16 3.0 V 13  PMMCOREV = 3  PMMCOREV = 0  1.8 V 5 3.0 V 5  PMMCOREV = 3  2.4 V 2  PMMCOREV = 3  3.0 V 5  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 0  1.8 V 5 3.0 V 5  PMMCOREV = 0  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  PMMCOREV = 3  1.8 V 5 3.0 V 5  1.8 V 76  CL = 20 pF, PMMCOREV = 0  UCLK edge to SOMI valid, CL = 20 pF, PMMCOREV = 3  1.8 V 18		CTE les times I set electric CTE high	PININCOREV = 0	3.0 V	3			ns
STE access time, STE low to SOMI data out   PMMCOREV = 0   1.8 V   66   3.0 V   3   66   3.0 V   50   3.0 V   3   66   3.0 V   3   3   3   3   3   3   3   3   3	TE,LAG	STE lag time, Last clock to STE high	DMMCODEV 2	2.4 V	3			
			PININCOREV = 3	3.0 V	3			ns
$t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = t_{STE} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out t_{STE,ACC} = access time, STE low to SOMI data out access time, STE low time, STE low time out t_{STE,ACC} = access time, STE low time out time out time out time out time out time$			DMMCOREV 0	1.8 V			66	
PMMCOREV = 3	<sup>L</sup> STE,ACC out	STE access time. STE low to SOMI data	PININCOREV = 0	3.0 V			50	ns
STE disable time, STE high to SOMI high impedance   PMMCOREV = 0   1.8 V   30   3.0 V   23   2.4 V   16   3.0 V   13   1.8 V   5   3.0 V   5   1.8 V   1.8 V   1.8 V   1.8 V		DMMCODEV 2	2.4 V			36		
$t_{STE,DIS}  \begin{array}{c} \text{STE disable time, STE high to SOMI high impedance} \\ \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & $			PMMCOREV = 3	3.0 V			30	ns
STE disable time, STE high to SOMI high impedance   PMMCOREV = 3   2.4 V   16   3.0 V   13   13   18 V   5     18 V			DMMCOREV - 0	1.8 V			30	ne
			PMMCOREV = 0	3.0 V			23	ns
$t_{SU,SI}  SIMO \text{ input data setup time} \\ \hline t_{SU,SI}  SIMO \text{ input data setup time} \\ \hline \\ t_{HD,SI}  SIMO \text{ input data hold time} \\ \hline \\ t_{VALID,SO}  SOMI \text{ output data valid time}^{(2)} \\ \hline \\ \hline \\ t_{VALID,SO}  SOMI \text{ output data valid time}^{(2)} \\ \hline \\ $	TE,DIS		DMMCODEV 2	2.4 V			16	
$t_{SU,SI}  SIMO \text{ input data setup time} \\ \hline \\ t_{HD,SI}  SIMO \text{ input data hold time} \\ \hline \\ t_{VALID,SO}  SOMI \text{ output data valid time} \\ \hline \\ t_{VALID,SO}  SOMI  output data vali$			PMMCOREV = 3	3.0 V			13	ns
$t_{\text{SU,SI}}  \text{SIMO input data setup time} \\ \hline \\ PMMCOREV = 3 \\ \hline \\ PMMCOREV = 3 \\ \hline \\ PMMCOREV = 0 \\ \hline \\ PMMCOREV = 0 \\ \hline \\ RATE  \\ \hline$		SIMO input data setup time	DIMINOCESTY O	1.8 V	5			
PMMCOREV = 3			PMMCOREV = 0	3.0 V	5			ns
$t_{\text{HD,SI}}  \text{SIMO input data hold time} \\ t_{\text{HD,SI}}  \text{SIMO input data hold time} \\ \\ t_{\text{VALID,SO}}  \text{SOMI output data valid time}^{(2)} \\ \\ t_{$	U,SI		PMMCOREV = 3	2.4 V	2			no
$t_{\text{HD,SI}}  \text{SIMO input data hold time} \\ \hline \\ PMMCOREV = 0 \\ \hline \\ PMMCOREV = 3 \\ \hline \\ UCLK \ \text{edge to SOMI valid,} \\ C_L = 20 \ \text{pF, PMMCOREV} = 0 \\ \hline \\ UCLK \ \text{edge to SOMI valid,} \\ C_L = 20 \ \text{pF, PMMCOREV} = 0 \\ \hline \\ UCLK \ \text{edge to SOMI valid,} \\ C_L = 20 \ \text{pF, PMMCOREV} = 3 \\ \hline \\ UCLK \ \text{edge to SOMI valid,} \\ C_L = 20 \ \text{pF, PMMCOREV} = 3 \\ \hline \\ UCLK \ \text{edge to SOMI valid,} \\ UCLK \ edge to$				3.0 V	2			ns
$t_{\text{HD,SI}}  \text{SIMO input data hold time} \\ \hline \\ PMMCOREV = 3 \\ \hline \\ t_{\text{VALID,SO}}  \text{SOMI output data valid time}^{(2)} \\ \hline \\ \\ t_{\text{VALID,SO}}  \text{SOMI output data valid time}^{(2)} \\ \hline \\ \\ \\ t_{\text{VALID,SO}}  \text{SOMI output data valid time}^{(2)} \\ \hline \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ $				1.8 V	5			
PMMCOREV = 3		0.00	PMMCOREV = 0	3.0 V	5			ns
$t_{VALID,SO}  \text{SOMI output data valid time}^{(2)}  \begin{aligned} & & & & & & & & & & & & & & & & & &$	D,SI	SIMO input data hold time	DMM400DEV/ 0	2.4 V	5			
t <sub>VALID,SO</sub> SOMI output data valid time (2)			PMMCOREV = 3	3.0 V	5			ns
			UCLK edge to SOMI valid,	1.8 V			76	
UCLK edge to SOMI valid, C <sub>L</sub> = 20 pF, PMMCOREV = 3 3.0 V 40		2014	C <sub>L</sub> = 20 pF, PMMCOREV = 0	3.0 V			60	ns
$C_L = 20 \text{ pF}, \text{ PMMCOREV} = 3$ 3.0 V 40	t <sub>VALID,SO</sub>	Solvii output data valid time	UCLK edge to SOMI valid,	2.4 V			44	
C 20 pF PMMCORFV 0 1.8 V 18			C <sub>L</sub> = 20 pF, PMMCOREV = 3	3.0 V			40	ns
	t <sub>HD,SO</sub> S		C 20 % F DMMCOREV C	1.8 V	18			
3.0 V   12		SOMI output data hold time <sup>(3)</sup>	$C_L = 20 \text{ pF}, PMMCOREV = 0$	3.0 V	12			ns
2.4 V   10			C = 20 pE DMMCOREV 2	2.4 V	10			
$C_L = 20 \text{ pF}, \text{ PMMCOREV} = 3$ 3.0 V 8			$C_L = 20 \text{ pr}, \text{PIMINICOREV} = 3$	3.0 V	8			ns

 <sup>(1)</sup> f<sub>UCXCLK</sub> = 1/2t<sub>LO/HI</sub> with t<sub>LO/HI</sub> ≥ max(t<sub>VALID,MO(Master)</sub> + t<sub>SU,SI(USCI)</sub>, t<sub>SU,MI(Master)</sub> + t<sub>VALID,SO(USCI)</sub>).
 For the master's parameters t<sub>SU,MI(Master)</sub> and t<sub>VALID,MO(Master)</sub> refer to the SPI parameters of the attached slave.
 (2) Specifies the time to drive the next valid data to the SOMI output after the output changing UCLK clock edge. See the timing diagrams

in Figure 11 and Figure 12.

Specifies how long data on the SOMI output is valid after the output changing UCLK clock edge. See the timing diagrams in Figure 11 and Figure 12.



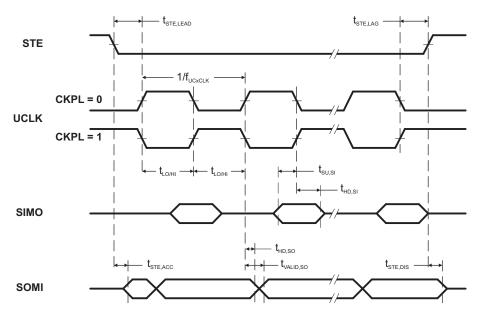


Figure 13. SPI Slave Mode, CKPH = 0

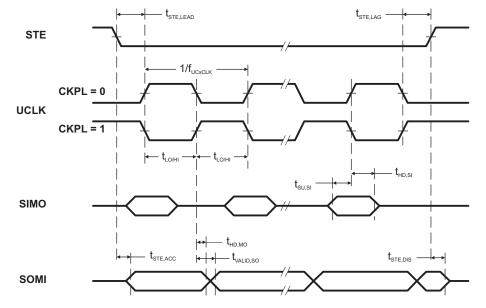


Figure 14. SPI Slave Mode, CKPH = 1



#### **USCI (I2C Mode)**

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>USCI</sub>	USCI input clock frequency	Internal: SMCLK, ACLK, External: UCLK, Duty cycle = 50% ± 10%			f <sub>S</sub>	YSTEM	MHz
f <sub>SCL</sub>	SCL clock frequency		2.2 V, 3 V	0		400	kHz
	Lightime (repeated) START	f <sub>SCL</sub> ≤ 100 kHz	0.01/.01/	4.0			
t <sub>HD,STA</sub>	Hold time (repeated) START	f <sub>SCL</sub> > 100 kHz	2.2 V, 3 V	0.6			μs
	Coturn time for a repeated CTART	f <sub>SCL</sub> ≤ 100 kHz	2.2 V, 3 V	4.7			
t <sub>SU,STA</sub>	Setup time for a repeated START	f <sub>SCL</sub> > 100 kHz	2.2 V, 3 V	0.6			μs
t <sub>HD,DAT</sub>	Data hold time		2.2 V, 3 V	0			ns
t <sub>SU,DAT</sub>	Data setup time		2.2 V, 3 V	250			ns
	Cotion times for CTOD	f <sub>SCL</sub> ≤ 100 kHz	227/27/	4.0			
t <sub>SU,STO</sub>	Setup time for STOP	f <sub>SCL</sub> > 100 kHz	2.2 V, 3 V	0.6			μs
	Pulse duration of spikes suppressed by input		2.2 V	50		600	
t <sub>SP</sub>	filter		3 V	50		600	ns

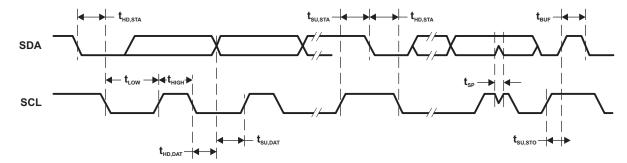


Figure 15. I2C Mode Timing



## 12-Bit ADC, Power Supply and Input Range Conditions

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)(1)

	PARAMETER	TEST CONDITIONS V <sub>CC</sub>		MIN	TYP	MAX	UNIT
AV <sub>CC</sub>	Analog supply voltage	AVCC and DVCC are connected together, AVSS and DVSS are connected together, $V_{(AVSS)} = V_{(DVSS)} = 0 \text{ V}$		2.2		3.6	٧
$V_{(Ax)}$	Analog input voltage range (2)	All ADC12 analog input pins Ax		0		$AV_{CC}$	V
	Operating supply current into	5 O MI I-(4)	2.2 V		125	155	
I <sub>ADC12_A</sub>	Operating supply current into AVCC terminal (3)	$f_{ADC12CLK} = 5.0 \text{ MHz}^{(4)}$	3 V		150	220	μA
Cı	Input capacitance	Only one terminal Ax can be selected at one time	2.2 V		20	25	pF
R <sub>I</sub>	Input MUX ON resistance	0 V ≤ V <sub>Ax</sub> ≤ AVCC		10	200	1900	Ω

- The leakage current is specified by the digital I/O input leakage.
- The analog input voltage range must be within the selected reference voltage range V<sub>R+</sub> to V<sub>R</sub> for valid conversion results. If the reference voltage is supplied by an external source or if the internal reference voltage is used and REFOUT = 1, then decoupling capacitors are required. See REF, External Reference and REF, Built-In Reference.
- The internal reference supply current is not included in current consumption parameter I<sub>ADC12</sub> A.
- ADC12ON = 1, REFON = 0, SHT0 = 0, SHT1 = 0, ADC12DIV = 0.

#### 12-Bit ADC, Timing Parameters

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		For specified performance of ADC12 linearity parameters using an external reference voltage or AVCC as reference. (1)		0.45	4.8	5.0	
f <sub>ADC12CLK</sub>	ADC conversion clock  For specified performance of ADC12 linearity parameters using the internal reference. (2)  For specified performance of ADC12 linearity parameters using the internal reference. (3)	2.2 V, 3 V	0.45	2.4	4.0	MHz	
		For specified performance of ADC12 linearity parameters using the internal reference. (3)		0.45	2.4	2.7	
f <sub>ADC12OSC</sub>	Internal ADC12 oscillator (4)	ADC12DIV = 0, f <sub>ADC12CLK</sub> = f <sub>ADC12OSC</sub>	2.2 V, 3 V	4.2	4.8	5.4	MHz
	Communication time	REFON = 0, Internal oscillator, ADC12OSC used for ADC conversion clock	2.2 V, 3 V	2.4		3.1	μs
<sup>t</sup> CONVERT	Conversion time	External $f_{ADC12CLK}$ from ACLK, MCLK, or SMCLK, ADC12SSEL $\neq 0$			(5)		
t <sub>Sample</sub>	Sampling time	$R_S = 400 \Omega$ , $R_I = 1000 \Omega$ , $C_I = 20 pF$ , $T = [R_S + R_I] \times C_I$ (6)	2.2 V, 3 V	1000			ns

<sup>(1)</sup> REFOUT = 0, external reference voltage: SREF2 = 0, SREF1 = 1, SREF0 = 0. AVCC as reference voltage: SREF2 = 0, SREF1 = 0, SREF0 = 0. The specified performance of the ADC12 linearity is ensured when using the ADC12OSC. For other clock sources, the specified performance of the ADC12 linearity is ensured with f<sub>ADC12CLK</sub> maximum of 5.0 MHz.

SREF2 = 0, SREF1 = 1, SREF0 = 0, ADC12SR = 0, REFOUT = 1
SREF2 = 0, SREF1 = 1, SREF0 = 0, ADC12SR = 0, REFOUT = 0. The specified performance of the ADC12 linearity is ensured when using the ADC12OSC divided by 2.

The ADC12OSC is sourced directly from MODOSC inside the UCS.

 $<sup>13 \</sup>times ADC12DIV \times 1/f_{ADC12CLK}$ 

Approximately ten Tau (t) are needed to get an error of less than ±0.5 LSB:  $t_{Sample} = \ln(2^{n+1}) \times (R_S + R_I) \times C_I + 800 \text{ ns}$ , where n = ADC resolution = 12,  $R_S$  = external source resistance



# 12-Bit ADC, Linearity Parameters Using an External Reference Voltage or AVCC as Reference Voltage

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN TYP	MAX	UNIT
_	Into much line a miter a much (1)	1.4 V ≤ dVREF ≤ 1.6 V <sup>(2)</sup>	227 27		±2.0	- CD
E <sub>I</sub>	Integral linearity error <sup>(1)</sup>	1.6 V < dVREF <sup>(2)</sup>	2.2 V, 3 V		±1.7	LSB
E <sub>D</sub>	Differential linearity error <sup>(1)</sup>	(2)	2.2 V, 3 V		±1.0	LSB
г	Offset error <sup>(3)</sup>	dVREF ≤ 2.2 V <sup>(2)</sup>	2.2 V, 3 V	±1.0	±2.0	LSB
Eo	Oliset ellor	dVREF > 2.2 V <sup>(2)</sup>	2.2 V, 3 V	±1.0	±2.0	LOD
E <sub>G</sub>	Gain error <sup>(3)</sup>	(2)	2.2 V, 3 V	±1.0	±2.0	LSB
_	Total unadjusted arror	dVREF ≤ 2.2 V <sup>(2)</sup>	2.2 V, 3 V	±1.4	±3.5	LCD
E <sub>T</sub>	Total unadjusted error	dVREF > 2.2 V <sup>(2)</sup>	2.2 V, 3 V	±1.4	±3.5	LSB

(1) Parameters are derived using the histogram method.

#### 12-Bit ADC, Linearity Parameters Using the Internal Reference Voltage

	PARAMETER	TEST COND	ITIONS <sup>(1)</sup>	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
Е	Integral linearity	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 4.0 MHz	2.2 V, 3 V			±1.7	LSB
Eı	error <sup>(2)</sup>	ADC12SR = 0, REFOUT = 0	f <sub>ADC12CLK</sub> ≤ 2.7 MHz	2.2 V, 3 V			±2.5	LOD
5.4	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 4.0 MHz		-1.0		+1.5		
$E_D$	Differential linearity error <sup>(2)</sup>	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 2.7 MHz	2.2 V, 3 V	-1.0		+1.0	LSB
	inidanty direi	ADC12SR = 0, REFOUT = 0	f <sub>ADC12CLK</sub> ≤ 2.7 MHz		-1.0		+2.5	
_	Offset error <sup>(3)</sup>	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 4.0 MHz	2.2 V, 3 V		±2.0	±4.0	LSB
Eo	Offset efforts	ADC12SR = 0, REFOUT = 0	f <sub>ADC12CLK</sub> ≤ 2.7 MHz	2.2 V, 3 V		±2.0	±4.0	LOD
_	Gain error <sup>(3)</sup>	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 4.0 MHz	2.2 V, 3 V		±1.0	±2.5	LSB
E <sub>G</sub>	Gain error	ADC12SR = 0, REFOUT = 0	f <sub>ADC12CLK</sub> ≤ 2.7 MHz	2.2 V, 3 V			±1.5% <sup>(4)</sup>	VREF
г	Total unadjusted	ADC12SR = 0, REFOUT = 1	f <sub>ADC12CLK</sub> ≤ 4.0 MHz	2.2 V, 3 V		±2	±5	LSB
E <sub>T</sub>	error	ADC12SR = 0, REFOUT = 0	f <sub>ADC12CLK</sub> ≤ 2.7 MHz	2.2 V, 3 V			±1.5% <sup>(4)</sup>	VREF

<sup>(1)</sup> The internal reference voltage is selected by: SREF2 = 0 or 1, SREF1 = 1, SREF0 = 1. dVREF = V<sub>R+</sub> - V<sub>R-</sub>.

<sup>(2)</sup> The external reference voltage is selected by: SREF2 = 0 or 1, SREF1 = 1, SREF0 = 0. dVREF = V<sub>R+</sub> - V<sub>R+</sub>, V<sub>R+</sub> < AVCC, V<sub>R-</sub>> AVSS. Unless otherwise mentioned, dVREF > 1.5 V. Impedance of the external reference voltage R < 100 Ω and two decoupling capacitors, 10 μF and 100 nF, should be connected to VREF to decouple the dynamic current. See also the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208).</p>

<sup>(3)</sup> Parameters are derived using a best fit curve.

<sup>(2)</sup> Parameters are derived using the histogram method.

<sup>(3)</sup> Parameters are derived using a best fit curve.

<sup>(4)</sup> The gain error and total unadjusted error are dominated by the accuracy of the integrated reference module absolute accuracy. In this mode the reference voltage used by the ADC12\_A is not available on a pin.



# 12-Bit ADC, Temperature Sensor and Built-In V<sub>MID</sub><sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
M	See (2)	ADC12ON = 1, INCH = 0Ah,	2.2 V		680		m)/
V <sub>SENSOR</sub>	See (-)	$T_A = 0$ °C	3 V		680		mV
T0		ADC420N 4 INCH 0Ab	2.2 V		2.25		\//00
TC <sub>SENSOR</sub>		ADC12ON = 1, INCH = 0Ah	3 V		2.25		mV/°C
	Sample time required if		2.2 V	100			
<sup>t</sup> SENSOR(sample)	channel 10 is selected (3)		3 V	100			μs
	AV <sub>CC</sub> divider at channel 11, V <sub>AVCC</sub> factor	ADC12ON = 1, INCH = 0Bh		0.48	0.5	0.52	V <sub>AVCC</sub>
$V_{MID}$	AN/ divides at absence 44	ADC42ON 4 INCLL ODE	2.2 V	1.06	1.1	1.14	V
	AV <sub>CC</sub> divider at channel 11	ADC12ON = 1, INCH = 0Bh	3 V	1.44	1.5	1.56	
t <sub>VMID</sub> (sample)	Sample time required if channel 11 is selected (4)	ADC12ON = 1, INCH = 0Bh, Error of conversion result ≤ 1 LSB	2.2 V, 3 V	1000			ns

- (1) The temperature sensor is provided by the REF module. See the REF module parametric I<sub>REF+</sub> regarding the current consumption of the temperature sensor.
- (2) The temperature sensor offset can be significant. A single-point calibration is recommended to minimize the offset error of the built-in temperature sensor. The TLV structure contains calibration values for 30°C ± 3°C and 85°C ± 3°C for each of the available reference voltage levels. The sensor voltage can be computed as V<sub>SENSOR</sub> \* (Temperature, °C) + V<sub>SENSOR</sub>, where TC<sub>SENSOR</sub> and V<sub>SENSOR</sub> can be computed from the calibration values for higher accuracy. See also the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208).
- (3) The typical equivalent impedance of the sensor is 51 kΩ. The sample time required includes the sensor-on time t<sub>SENSOR(on)</sub>.
- (4) The on-time t<sub>VMID(on)</sub> is included in the sampling time t<sub>VMID(sample)</sub>; no additional on time is needed.

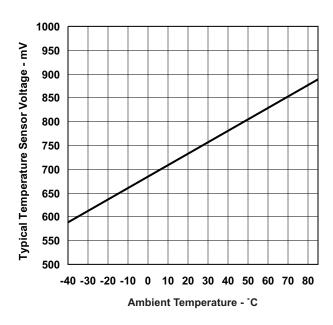


Figure 16. Typical Temperature Sensor Voltage



#### **REF, External Reference**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)(1)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP MAX	UNIT
V <sub>eREF+</sub>	Positive external reference voltage input	$V_{eREF+} > V_{REF-}/V_{eREF-}$ (2)		1.4	$AV_CC$	>
V <sub>REF</sub> _/V <sub>eREF</sub> _	Negative external reference voltage input	V <sub>eREF+</sub> > V <sub>REF</sub> _/V <sub>eREF</sub> _ <sup>(3)</sup>		0	1.2	٧
(V <sub>eREF+</sub> – V <sub>REF</sub> _/V <sub>eREF</sub> _)	Differential external reference voltage input	V <sub>eREF+</sub> > V <sub>REF</sub> _/V <sub>eREF</sub> _ <sup>(4)</sup>		1.4	$AV_CC$	٧
lveref+, lvref-/veref-	Static input current	$\begin{array}{l} 1.4~\text{V} \leq \text{V}_{\text{eREF+}} \leq \text{V}_{\text{AVCC}},\\ \text{V}_{\text{eREF-}} = 0~\text{V},~\text{f}_{\text{ADC12CLK}} = 5~\text{MHz},\\ \text{ADC12SHTx} = 1\text{h},\\ \text{Conversion rate 200 ksps} \end{array}$	2.2 V, 3 V	-26	26	μΑ
		$ \begin{array}{l} 1.4~\text{V} \leq \text{V}_{\text{eREF+}} \leq \text{V}_{\text{AVCC}}, \\ \text{V}_{\text{eREF-}} = 0~\text{V}, \text{f}_{\text{ADC12CLK}} = 5~\text{MHz}, \\ \text{ADC12SHTx} = 8\text{h}, \\ \text{Conversion rate 20 ksps} \\ \end{array} $	2.2 V, 3 V	-1	1	μA
C <sub>VREF+/-</sub>	Capacitance at $V_{\text{REF+}}$ and $V_{\text{REF-}}$ terminals			<sup>(5)</sup> 10		μF

- (1) The external reference is used during ADC conversion to charge and discharge the capacitance array. The input capacitance, C<sub>i</sub>, is also the dynamic load for an external reference during conversion. The dynamic impedance of the reference supply should follow the recommendations on analog-source impedance to allow the charge to settle for 12-bit accuracy.
- (2) The accuracy limits the minimum positive external reference voltage. Lower reference voltage levels may be applied with reduced accuracy requirements.
- (3) The accuracy limits the maximum negative external reference voltage. Higher reference voltage levels may be applied with reduced accuracy requirements.
- (4) The accuracy limits minimum external differential reference voltage. Lower differential reference voltage levels may be applied with reduced accuracy requirements.
- (5) Two decoupling capacitors, 10 μF and 100 nF, should be connected to VREF to decouple the dynamic current required for an external reference source if it is used for the ADC12\_A. See also the MSP430x5xx and MSP430x6xx Family User's Guide (SLAU208).

#### REF, Built-In Reference

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	Positive built-in reference voltage output	REFVSEL = $\{2\}$ for 2.5 V, REFON = REFOUT = 1, $I_{VREF+}$ = 0 A	3 V		2.50	±1.5%	
V <sub>REF+</sub>		REFVSEL = {1} for 2.0 V, REFON = REFOUT = 1, $I_{VREF+}$ = 0 A	3 V		1.98	±1.5%	V
		REFVSEL = $\{0\}$ for 1.5 V, REFON = REFOUT = 1, $I_{VREF+}$ = 0 A	2.2 V, 3 V		1.49	±1.5%	
	AV <sub>CC</sub> minimum voltage, Positive built-in reference active	REFVSEL = {0} for 1.5 V		2.2			
AV <sub>CC(min)</sub>		REFVSEL = {1} for 2.0 V		2.3			V
		REFVSEL = {2} for 2.5 V		2.8			
		ADC12SR = 1, REFON = 1, REFOUT = 0, REFBURST = 0	3 V		70	100	μΑ
	Operating supply current into	ADC12SR = 1, REFON = 1, REFOUT = 1, REFBURST = 0	3 V		0.45	0.75	mA
IREF+	Operating supply current into AV <sub>CC</sub> terminal (2) (3)	ADC12SR = 0, REFON = 1, REFOUT = 0, REFBURST = 0	3 V		210	310	μΑ
		ADC12SR = 0, REFON = 1, REFOUT = 1, REFBURST = 0	3 V		0.95	1.7	mA

- (1) The reference is supplied to the ADC by the REF module and is buffered locally inside the ADC. The ADC uses two internal buffers, one smaller and one larger for driving the V<sub>REF+</sub> terminal. When REFOUT = 1, the reference is available at the V<sub>REF+</sub> terminal, as well as, used as the reference for the conversion and utilizes the larger buffer. When REFOUT = 0, the reference is only used as the reference for the conversion and utilizes the smaller buffer.
- (2) The internal reference current is supplied via terminal AV<sub>CC</sub>. Consumption is independent of the ADC12ON control bit, unless a conversion is active. REFOUT = 0 represents the current contribution of the smaller buffer. REFOUT = 1 represents the current contribution of the larger buffer without external load.
- (3) The temperature sensor is provided by the REF module. Its current is supplied via terminal AV<sub>CC</sub> and is equivalent to I<sub>REF+</sub> with REFON =1 and REFOUT = 0.



#### **REF, Built-In Reference (continued)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)(1)

	PARAMETER	TEST CONDITIONS	$v_{cc}$	MIN	TYP	MAX	UNIT
I <sub>L(VREF+)</sub>	Load-current regulation, V <sub>REF+</sub> terminal <sup>(4)</sup>	REFVSEL = $(0, 1, 2)$ $I_{VREF+}$ = +10 $\mu$ A/-1000 $\mu$ A $AV_{CC}$ = $AV_{CC}$ (min) for each reference level, REFVSEL = $(0, 1, 2)$ , REFON = REFOUT = 1				2500	μV/mA
C <sub>VREF+</sub>	Capacitance at VREF+ terminals	REFON = REFOUT = 1		20		100	pF
TC <sub>REF+</sub>	Temperature coefficient of built-in reference (5)	I <sub>VREF+</sub> = 0 A, REFVSEL = (0, 1, 2), REFON = 1, REFOUT = 0 or 1			30	50	ppm/ °C
PSRR_DC	Power supply rejection ratio (DC)	$\begin{array}{l} {\sf AV}_{\sf CC} = {\sf AV}_{\sf CC~(min)} \cdot {\sf AV}_{\sf CC(max)},  {\sf T_A} = 25^\circ {\sf C}, \\ {\sf REFVSEL} = (0,1,2\},  {\sf REFON} = 1, \\ {\sf REFOUT} = 0  {\sf or}  1 \end{array}$			120	300	μV/V
PSRR_AC	Power supply rejection ratio (AC)	$\begin{array}{l} AV_{CC} = AV_{CC~(min)} \cdot AV_{CC(max)},  T_A = 25^{\circ}C, \\ f = 1~kHz,  \Delta Vpp = 100~mV, \\ REFVSEL = (0,  1,  2\},  REFON = 1, \\ REFOUT = 0~or~1 \end{array}$			6.4		mV/V
	Sottling time of reference	$\begin{aligned} &AV_{CC} = AV_{CC}_{(min)} \cdot AV_{CC(max)}, \\ &REFVSEL = \{0, \ 1, \ 2\}, \ REFOUT = 0, \\ &REFON = 0 \to 1 \end{aligned}$			75		
t <sub>SETTLE</sub>	Settling time of reference voltage <sup>(6)</sup>	$\begin{array}{l} AV_{CC} = AV_{CC~(min)} \cdot AV_{CC(max)}, \\ C_{VREF} = C_{VREF}(max), \\ REFVSEL = (0, 1, 2\}, REFOUT = 1, \\ REFON = 0 \rightarrow 1 \end{array}$			75		μs

<sup>(4)</sup> Contribution only due to the reference and buffer including package. This does not include resistance due to PCB trace, etc.

## Flash Memory

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DV <sub>CC(PGM/ERASE)</sub>	Program and erase supply voltage		1.8		3.6	V
I <sub>PGM</sub>	Average supply current from DVCC during program			3	5	mA
I <sub>ERASE</sub>	Average supply current from DVCC during erase				2	mA
I <sub>MERASE</sub> , I <sub>BANK</sub>	Average supply current from DVCC during mass erase or bank erase				2	mA
t <sub>CPT</sub>	Cumulative program time	See (1)			16	ms
	Program and erase endurance		10 <sup>4</sup>	10 <sup>5</sup>		cycles
t <sub>Retention</sub>	Data retention duration	T <sub>J</sub> = 25°C	100			years
t <sub>Word</sub>	Word or byte program time	See (2)	64		85	μs
t <sub>Block, 0</sub>	Block program time for first byte or word	See (2)	49		65	μs
t <sub>Block, 1-(N-1)</sub>	Block program time for each additional byte or word, except for last byte or word	See (2)	37		49	μs
t <sub>Block, N</sub>	Block program time for last byte or word	See (2)	55		73	μs
t <sub>Erase</sub>	Erase time for segment, mass erase, and bank erase when available.	See (2)	23		32	ms
f <sub>MCLK,MGR</sub>	MCLK frequency in marginal read mode (FCTL4.MGR0 = 1 or FCTL4. MGR1 = 1)		0		1	MHz

<sup>(1)</sup> The cumulative program time must not be exceeded when writing to a 128-byte flash block. This parameter applies to all programming methods: individual word/byte write and block write modes.

<sup>(5)</sup> Calculated using the box method: (MAX(-40 to 85°C) – MĬN(-40 to 85°C)) / MIN(-40 to 85°C)/(85°C – (–40°C)).

<sup>(6)</sup> The condition is that the error in a conversion started after t<sub>REFON</sub> is less than ±0.5 LSB. The settling time depends on the external capacitive load when REFOUT = 1.

<sup>(2)</sup> These values are hardwired into the flash controller's state machine.



## JTAG and Spy-Bi-Wire Interface

	PARAMETER	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
f <sub>SBW</sub>	Spy-Bi-Wire input frequency	2.2 V, 3 V	0		20	MHz
t <sub>SBW,Low</sub>	Spy-Bi-Wire low clock pulse duration	2.2 V, 3 V	0.025		15	μs
t <sub>SBW, En</sub>	Spy-Bi-Wire enable time (TEST high to acceptance of first clock edge) <sup>(1)</sup>	2.2 V, 3 V			1	μs
t <sub>SBW,Rst</sub>	Spy-Bi-Wire return to normal operation time		15		100	μs
,	TO((:	2.2 V	0		5	MHz
f <sub>TCK</sub>	TCK input frequency, 4-wire JTAG <sup>(2)</sup>	3 V	0		10	MHz
R <sub>internal</sub>	Internal pulldown resistance on TEST	2.2 V, 3 V	45	60	80	kΩ

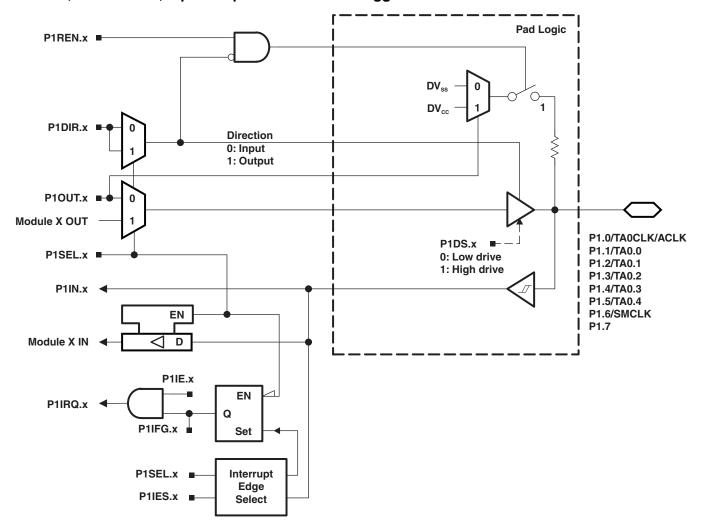
<sup>(1)</sup> Tools accessing the Spy-Bi-Wire interface must wait for the t<sub>SBW,En</sub> time after pulling the TEST/SBWTCK pin high before applying the first SBWTCK clock edge.

<sup>(2)</sup>  $f_{TCK}$  may be restricted to meet the timing requirements of the module selected.



#### **INPUT/OUTPUT SCHEMATICS**

## Port P1, P1.0 to P1.7, Input/Output With Schmitt Trigger



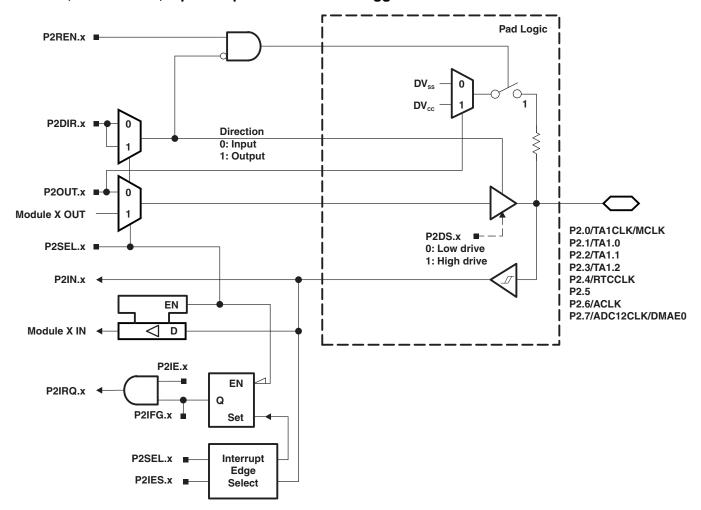


# Table 45. Port P1 (P1.0 to P1.7) Pin Functions

DINI NAME (D4)		FUNCTION	CONTROL BI	TS/SIGNALS
PIN NAME (P1.x)	x	FUNCTION	P1DIR.x	P1SEL.x
P1.0/TA0CLK/ACLK	0	P1.0 (I/O)	I: 0; O: 1	0
		TA0.TA0CLK	0	1
		ACLK	1	1
P1.1/TA0.0	1	P1.1 (I/O)	I: 0; O: 1	0
		TA0.CCI0A	0	1
		TA0.0	1	1
P1.2/TA0.1	2	P1.2 (I/O)	I: 0; O: 1	0
	-	TA0.CCI1A	0	1
		TA0.1	1	1
P1.3/TA0.2	3	P1.3 (I/O)	I: 0; O: 1	0
		TA0.CCI2A	0	1
		TA0.2	1	1
P1.4/TA0.3	4	P1.4 (I/O)	I: 0; O: 1	0
		TA0.CCI3A	0	1
		TA0.3	1	1
P1.5/TA0.4	5	P1.5 (I/O)	I: 0; O: 1	0
		TA0.CCI4A	0	1
		TA0.4	1	1
P1.6/SMCLK	6	P1.6 (I/O)	I: 0; O: 1	0
		SMCLK	1	1
P1.7	7	P1.7 (I/O)	I: 0; O: 1	0



# Port P2, P2.0 to P2.7, Input/Output With Schmitt Trigger





# Table 46. Port P2 (P2.0 to P2.7) Pin Functions

PIN NAME (P2.x)	x	FUNCTION	CONTROL BI	CONTROL BITS/SIGNALS	
			P2DIR.x	P2SEL.x	
P2.0/TA1CLK/MCLK	0	P2.0 (I/O)	I: 0; O: 1	0	
		TA1CLK	0	1	
		MCLK	1	1	
P2.1/TA1.0	1	P2.1 (I/O)	I: 0; O: 1	0	
		TA1.CCI0A	0	1	
		TA1.0	1	1	
P2.2/TA1.1	2	P2.2 (I/O)	I: 0; O: 1	0	
		TA1.CCI1A	0	1	
		TA1.1	1	1	
P2.3/TA1.2	3	P2.3 (I/O)	I: 0; O: 1	0	
		TA1.CCI2A	0	1	
		TA1.2	1	1	
P2.4/RTCCLK	4	P2.4 (I/O)	I: 0; O: 1	0	
		RTCCLK	1	1	
P2.5	5	P2.5 (I/O)	I: 0; O: 1	0	
P2.6/ACLK	6	P2.6 (I/O)	I: 0; O: 1	0	
		ACLK	1	1	
P2.7/ADC12CLK/DMAE0	7	P2.7 (I/O)	I: 0; O: 1	0	
		DMAE0	0	1	
		ADC12CLK	1	1	



## Port P3, P3.0 to P3.7, Input/Output With Schmitt Trigger

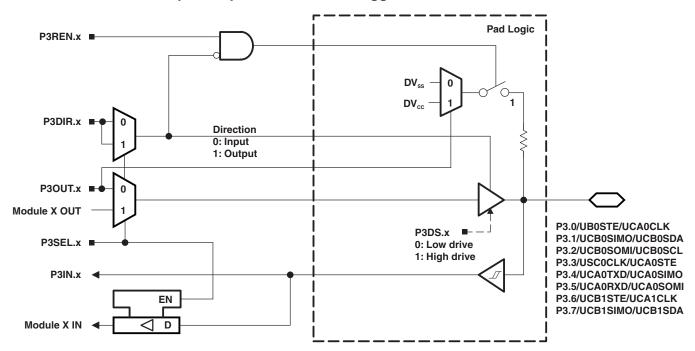


Table 47. Port P3 (P3.0 to P3.7) Pin Functions

PIN NAME (P3.x)	x	FUNCTION	CONTROL BIT	CONTROL BITS/SIGNALS <sup>(1)</sup>	
			P3DIR.x	P3SEL.x	
P3.0/UCB0STE/UCA0CLK	0	P3.0 (I/O)	I: 0; O: 1	0	
		UCB0STE/UCA0CLK(2) (3)	X	1	
P3.1/UCB0SIMO/UCB0SDA	1	P3.1 (I/O)	I: 0; O: 1	0	
		UCB0SIMO/UCB0SDA(2)(4)	X	1	
P3.2/UCB0SOMI/UCB0SCL	2	P3.2 (I/O)	I: 0; O: 1	0	
		UCB0SOMI/UCB0SCL <sup>(2)</sup> (4)	X	1	
P3.3/UCB0CLK/UCA0STE	3	P3.3 (I/O)	I: 0; O: 1	0	
		UCB0CLK/UCA0STE(2)(5)	X	1	
P3.4/UCA0TXD/UCA0SIMO	4	P3.4 (I/O)	I: 0; O: 1	0	
		UCA0TXD/UCA0SIMO(2)	X	1	
P3.5/UCA0RXD/UCA0SOMI	5	P3.5 (I/O)	I: 0; O: 1	0	
		UCA0RXD/UCA0SOMI(2)	X	1	
P3.6/UCB1STE/UCA1CLK	6	P3.6 (I/O)	I: 0; O: 1	0	
		UCB1STE/UCA1CLK <sup>(2)</sup> (6)	X	1	
P3.7/UCB1SIMO/UCB1SDA	7	P3.7 (I/O)	l: 0; O: 1	0	
		UCB1SIMO/UCB1SDA(2) (4)	X	1	

<sup>(1)</sup> X = Don't care

<sup>(2)</sup> The pin direction is controlled by the USCI module.

<sup>(3)</sup> UCAOCLK function takes precedence over UCBOSTE function. If the pin is required as UCAOCLK input or output, USCI B0 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

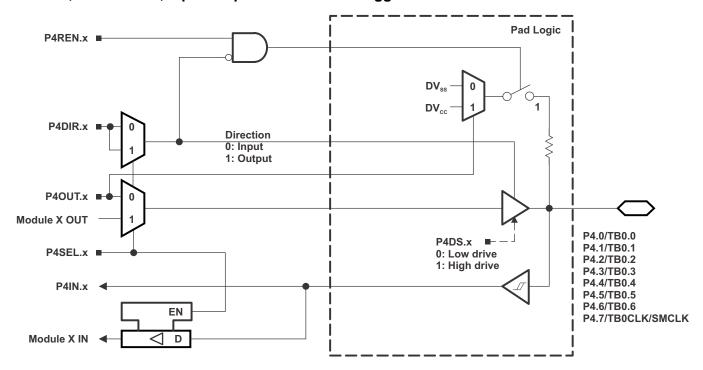
<sup>(4)</sup> If the I2C functionality is selected, the output drives only the logical 0 to V<sub>SS</sub> level.

<sup>(5)</sup> UCB0CLK function takes precedence over UCA0STE function. If the pin is required as UCB0CLK input or output, USCI A0 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

<sup>(6)</sup> UCA1CLK function takes precedence over UCB1STE function. If the pin is required as UCA1CLK input or output, USCI B1 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.



# Port P4, P4.0 to P4.7, Input/Output With Schmitt Trigger





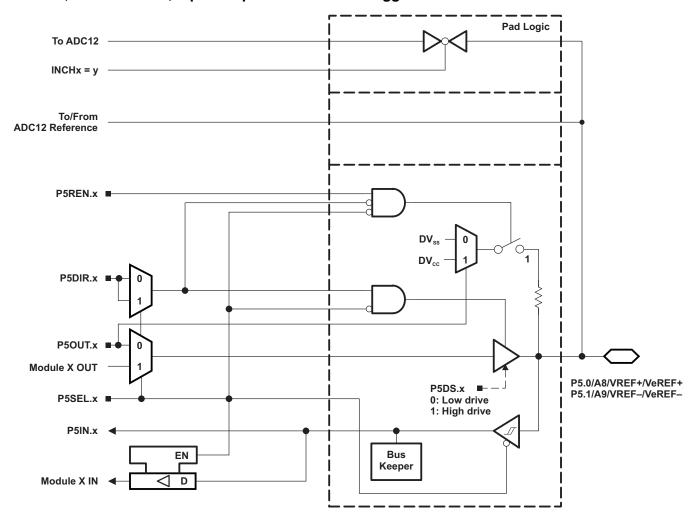
## Table 48. Port P4 (P4.0 to P4.7) Pin Functions

PIN NAME (P4.x)		FUNCTION	CONTROL B	CONTROL BITS/SIGNALS		
	X		P4DIR.x	P4SEL.x		
P4.0/TB0.0	0	4.0 (I/O)	I: 0; O: 1	0		
		TB0.CCI0A and TB0.CCI0B	0	1		
		TB0.0 <sup>(1)</sup>	1	1		
P4.1/TB0.1	1	4.1 (I/O)	I: 0; O: 1	0		
		TB0.CCI1A and TB0.CCI1B	0	1		
		TB0.1 <sup>(1)</sup>	1	1		
P4.2/TB0.2	2	4.2 (I/O)	I: 0; O: 1	0		
		TB0.CCI2A and TB0.CCI2B	0	1		
		TB0.2 <sup>(1)</sup>	1	1		
P4.3/TB0.3	3	4.3 (I/O)	I: 0; O: 1	0		
		TB0.CCI3A and TB0.CCI3B	0	1		
		TB0.3 <sup>(1)</sup>	1	1		
P4.4/TB0.5	4	4.4 (I/O)	I: 0; O: 1	0		
		TB0.CCI4A and TB0.CCI4B	0	1		
		TB0.4 <sup>(1)</sup>	1	1		
P4.5/TB0.5	5	4.5 (I/O)	I: 0; O: 1	0		
		TB0.CCI5A and TB0.CCI5B	0	1		
		TB0.5 <sup>(1)</sup>	1	1		
P4.6/TB0.6	6	4.6 (I/O)	I: 0; O: 1	0		
		TB0.CCI6A and TB0.CCI6B	0	1		
		TB0.6 <sup>(1)</sup>	1	1		
P4.7/TB0CLK/SMCLK	7	4.7 (I/O)	I: 0; O: 1	0		
		TB0CLK	0	1		
		SMCLK	1	1		

<sup>(1)</sup> Setting TBOUTH causes all Timer\_B configured outputs to be set to high impedance.



## Port P5, P5.0 and P5.1, Input/Output With Schmitt Trigger





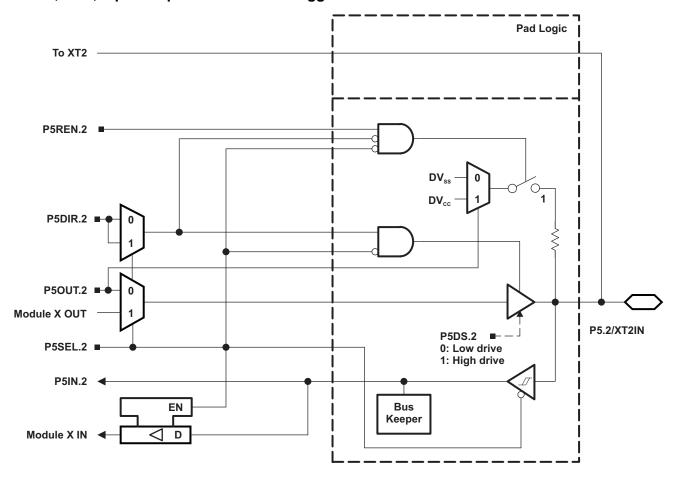
### Table 49. Port P5 (P5.0 and P5.1) Pin Functions

PIN NAME (P5.x)		FUNCTION	CONTROL BITS/SIGNALS <sup>(1)</sup>			
	Х	FUNCTION	P5DIR.x	P5SEL.x	REFOUT	
P5.0/A8/VREF+/VeREF+	0	P5.0 (I/O) <sup>(2)</sup>	I: 0; O: 1	0	Х	
		A8/VeREF+ <sup>(3)</sup>	Х	1	0	
		A8/VREF+ <sup>(4)</sup>	Х	1	1	
P5.1/A9/VREF-/VeREF-	1	P5.1 (I/O) <sup>(2)</sup>	I: 0; O: 1	0	Х	
		A9/VeREF-(5)	Х	1	0	
		A9/VREF-(6)	Х	1	1	

- (1) X = Don't care
- (2) Default condition
- (3) Setting the P5SEL.0 bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals. An external voltage can be applied to VeREF+ and used as the reference for the ADC12\_A. Channel A8, when selected with the INCHx bits, is connected to the VREF+/VeREF+ pin.
- (4) Setting the P5SEL.0 bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals. The ADC12\_A, VREF+ reference is available at the pin. Channel A8, when selected with the INCHx bits, is connected to the VREF+/VeREF+ pin.
- (5) Setting the P5SEL.1 bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals. An external voltage can be applied to VeREF- and used as the reference for the ADC12\_A. Channel A9, when selected with the INCHx bits, is connected to the VREF-/VeREF- pin.
- (6) Setting the P5SEL.1 bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals. The ADC12\_A, VREF- reference is available at the pin. Channel A9, when selected with the INCHx bits, is connected to the VREF-/VeREF- pin.



## Port P5, P5.2, Input/Output With Schmitt Trigger





## Port P5, P5.3, Input/Output With Schmitt Trigger

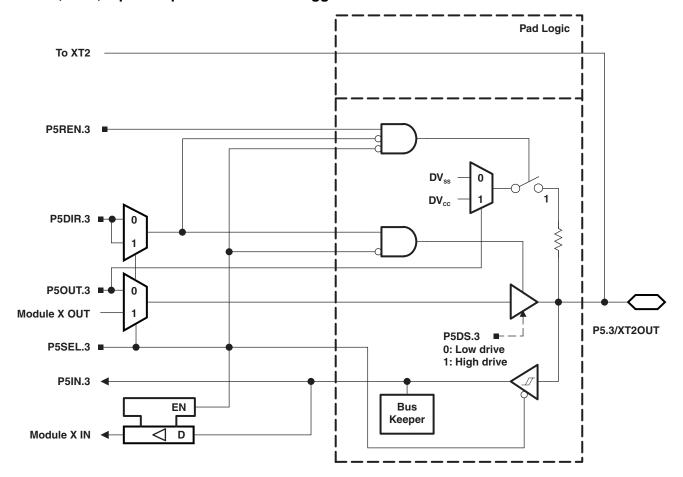


Table 50. Port P5 (P5.2) Pin Functions

PIN NAME (P5.x)		FUNCTION	CONTROL BITS/SIGNALS <sup>(1)</sup>				
	х	FUNCTION	P5DIR.x	P5SEL.2	P5SEL.3	XT2BYPASS	
P5.2/XT2IN	2	P5.2 (I/O)	I: 0; O: 1	0	Х	Х	
		XT2IN crystal mode <sup>(2)</sup>	Х	1	Х	0	
		XT2IN bypass mode <sup>(2)</sup>	X	1	Х	1	
P5.3/XT2OUT	3	P5.3 (I/O)	I: 0; O: 1	0	Х	Х	
		XT2OUT crystal mode <sup>(3)</sup>	Х	1	Х	0	
		P5.3 (I/O) <sup>(3)</sup>	X	1	Х	1	

<sup>(1)</sup> X = Don't care

<sup>(2)</sup> Setting P5SEL.2 causes the general-purpose I/O to be disabled. Pending the setting of XT2BYPASS, P5.2 is configured for crystal mode or bypass mode.

<sup>(3)</sup> Setting P5SEL.2 causes the general-purpose I/O to be disabled in crystal mode. When using bypass mode, P5.3 can be used as general-purpose I/O.



## Port P5, P5.4 to P5.7, Input/Output With Schmitt Trigger

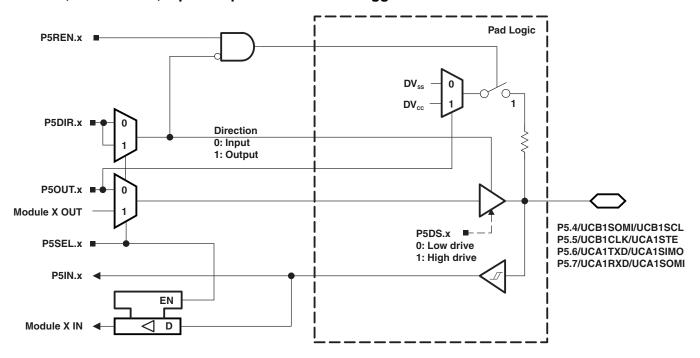


Table 51. Port P5 (P5.4 to P5.7) Pin Functions

DINI NIAME (DE)		FUNCTION	CONTROL BITS/SIGNALS <sup>(1)</sup>	
PIN NAME (P5.x)	X	FUNCTION	P5DIR.x	P5SEL.x
P5.4/UCB1SOMI/UCB1SCL	4	P5.4 (I/O)	I: 0; O: 1	0
		UCB1SOMI/UCB1SCL (2) (3)	Х	1
P5.5/UCB1CLK/UCA1STE	5	P5.5 (I/O)	I: 0; O: 1	0
		UCB1CLK/UCA1STE <sup>(2)</sup> (4)	Х	1
P5.6/UCA1TXD/UCA1SIMO	6	P5.6 (I/O)	I: 0; O: 1	0
		UCA1TXD/UCA1SIMO <sup>(2)</sup>	Х	1
P5.7/UCA1RXD/UCA1SOMI	7	P5.7 (I/O)	I: 0; O: 1	0
		UCA1RXD/UCA1SOMI(2)	Х	1

<sup>(1)</sup> X = Don't care

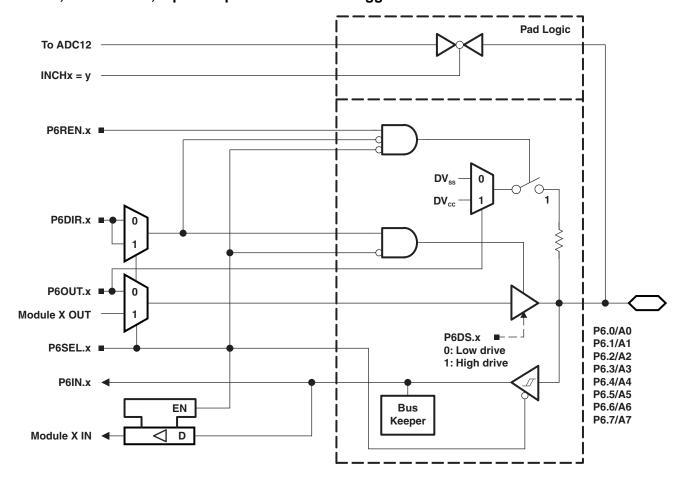
<sup>(2)</sup> The pin direction is controlled by the USCI module.

<sup>(3)</sup> If the I2C functionality is selected, the output drives only the logical 0 to V<sub>SS</sub> level.

<sup>(4)</sup> UCB1CLK function takes precedence over UCA1STE function. If the pin is required as UCB1CLK input or output, USCI A1 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.



# Port P6, P6.0 to P6.7, Input/Output With Schmitt Trigger





## Table 52. Port P6 (P6.0 to P6.7) Pin Functions

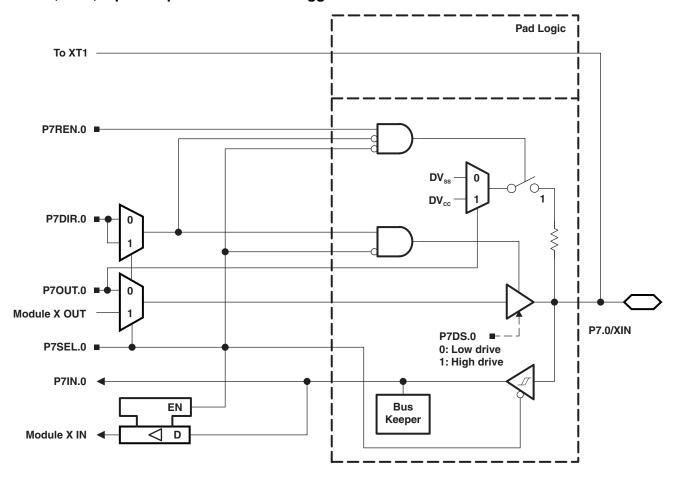
PIN NAME (P6.x)		FUNCTION	CONT	ROL BITS/SIGNA	LS <sup>(1)</sup>
	X	FUNCTION	P6DIR.x	P6SEL.x	INCHx
P6.0/A0	0	P6.0 (I/O)	I: 0; O: 1	0	Х
		A0 <sup>(2)</sup> (3)	X	Х	0
P6.1/A1	1	P6.1 (I/O)	I: 0; O: 1	0	Х
		A1 <sup>(2)</sup> (3)	Х	Х	1
P6.2/A2	2	P6.2 (I/O)	I: 0; O: 1	0	Х
		A2 <sup>(2)</sup> (3)	X	Х	2
P6.3/A3	3	P6.3 (I/O)	I: 0; O: 1	0	Х
		A3 <sup>(2)</sup> (3)	Х	Х	3
P6.4/A4	4	P6.4 (I/O)	I: 0; O: 1	0	Х
		A4 <sup>(2)</sup> (3)	X	Х	4
P6.5/A5	5	P6.5 (I/O)	I: 0; O: 1	0	Х
		A5 <sup>(1)</sup> (2) (3)	X	Х	5
P6.6/A6	6	P6.6 (I/O)	I: 0; O: 1	0	Х
		A6 <sup>(2)</sup> (3)	Х	Х	6
P6.7/A7	7	P6.7 (I/O)	I: 0; O: 1	0	Х
		A7 <sup>(2)</sup> (3)	X	Х	7

X = Don't care Setting the P6SEL.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals.

The ADC12\_A channel Ax is connected internally to AV<sub>SS</sub> if not selected via the respective INCHx bits.



# Port P7, P7.0, Input/Output With Schmitt Trigger





## Port P7, P7.1, Input/Output With Schmitt Trigger

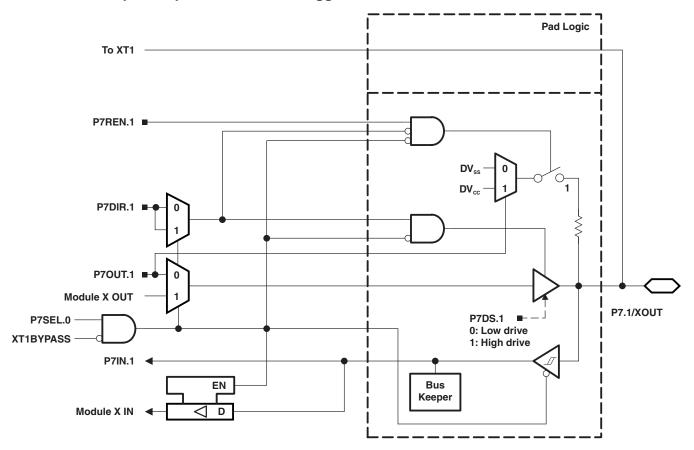


Table 53. Port P7 (P7.0 and P7.1) Pin Functions

PIN NAME (P7.x)		FUNCTION	CONTROL BITS/SIGNALS <sup>(1)</sup>				
	X	FUNCTION	P7DIR.x	P7SEL.0	P7SEL.1	XT1BYPASS	
P7.0/XIN	0	P7.0 (I/O)	I: 0; O: 1	0	Х	Х	
		XIN crystal mode <sup>(2)</sup>	X	1	Х	0	
		XIN bypass mode <sup>(2)</sup>	Х	1	Х	1	
P7.1/XOUT	1	P7.1 (I/O)	I: 0; O: 1	0	Х	Х	
		XOUT crystal mode (3)	Х	1	Х	0	
		P7.1 (I/O) <sup>(3)</sup>	Х	1	X	1	

<sup>(1)</sup> X = Don't care

<sup>(2)</sup> Setting P7SEL.0 causes the general-purpose I/O to be disabled. Pending the setting of XT1BYPASS, P7.0 is configured for crystal mode or bypass mode.

<sup>(3)</sup> Setting PTSEL.0 causes the general-purpose I/O to be disabled in crystal mode. When using bypass mode, P7.1 can be used as general-purpose I/O.



# Port P7, P7.2 and P7.3, Input/Output With Schmitt Trigger

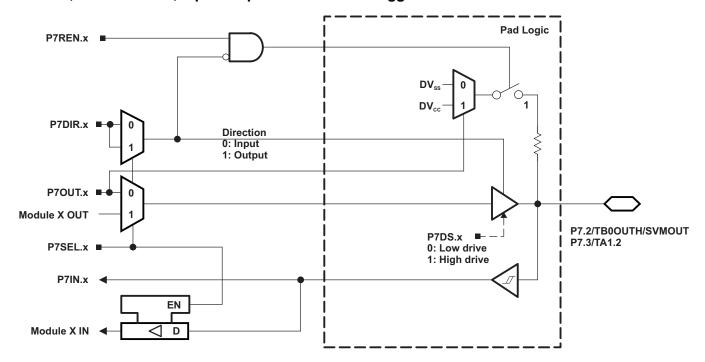


Table 54. Port P7 (P7.2 and P7.3) Pin Functions

DIN NAME (DZ v)	x	FUNCTION	CONTROL BITS/SIGNALS	
PIN NAME (P7.x)			P7DIR.x	P7SEL.x
P7.2/TB0OUTH/SVMOUT	2	P7.2 (I/O)	I: 0; O: 1	0
		TB0OUTH	0	1
		SVMOUT	1	1
P7.3/TA1.2	3	P7.3 (I/O)	I: 0; O: 1	0
		TA1.CCI2B	0	1
		TA1.2	1	1



## Port P7, P7.4 to P7.7, Input/Output With Schmitt Trigger

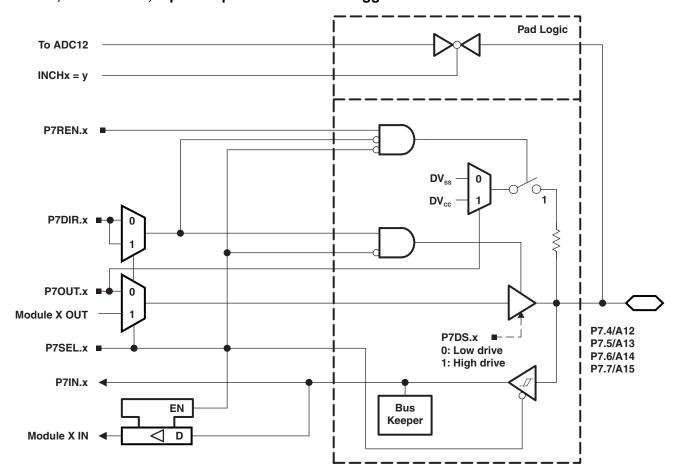


Table 55. Port P7 (P7.4 to P7.7) Pin Functions

DIN NAME (DZ)		FUNCTION	CONT	CONTROL BITS/SIGNALS <sup>(1)</sup>		
PIN NAME (P7.x)	X	FUNCTION	P7DIR.x	P7SEL.x	INCHx	
P7.4/A12	4	P7.4 (I/O)	I: 0; O: 1	0	Х	
		A12 <sup>(2) (3)</sup>	X	X	12	
P7.5/A13	5	P7.5 (I/O)	I: 0; O: 1	0	Х	
		A13 <sup>(4) (5)</sup>	X	Х	13	
P7.6/A14	6	P7.6 (I/O)	I: 0; O: 1	0	Х	
		A14 <sup>(4) (5)</sup>	Х	Х	14	
P7.7/A15	7	P7.7 (I/O)	I: 0; O: 1	0	Х	
		A15 <sup>(4) (5)</sup>	X	Х	15	

<sup>(1)</sup> X = Don't care

<sup>(2)</sup> Setting the P7SEL.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals.

<sup>(3)</sup> The ADC12\_A channel Ax is connected internally to AV<sub>SS</sub> if not selected via the respective INCHx bits.

<sup>(4)</sup> Setting the P7SEL.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals.

<sup>(5)</sup> The ADC12\_A channel Ax is connected internally to AV<sub>SS</sub> if not selected via the respective INCHx bits.



## Port P8, P8.0 to P8.7, Input/Output With Schmitt Trigger

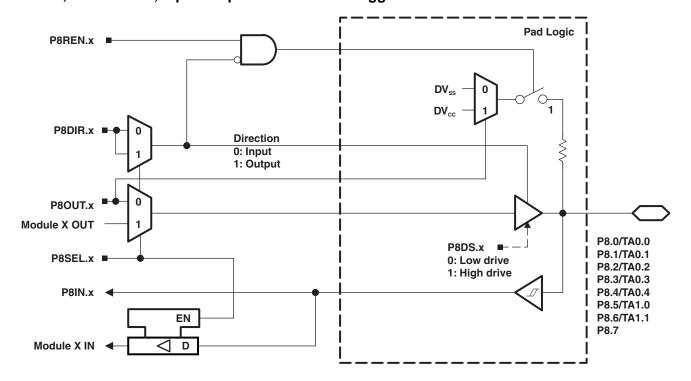


Table 56. Port P8 (P8.0 to P8.7) Pin Functions

DINI NIAME (DO)		FUNCTION	CONTROL BI	TS/SIGNALS
PIN NAME (P8.x)	х		P8DIR.x	P8SEL.x
P8.0/TA0.0	0	P8.0 (I/O)	I: 0; O: 1	0
		TA0.CCI0B	0	1
		TA0.0	1	1
P8.1/TA0.1	1	P8.1 (I/O)	I: 0; O: 1	0
		TA0.CCI1B	0	1
		TA0.1	1	1
P8.2/TA0.2	2	P8.2 (I/O)	I: 0; O: 1	0
		TA0.CCI2B	0	1
		TA0.2	1	1
P8.3/TA0.3	3	P8.3 (I/O)	I: 0; O: 1	0
		TA0.CCI3B	0	1
		TA0.3	1	1
P8.4/TA0.4	4	P8.4 (I/O)	I: 0; O: 1	0
		TA0.CCI4B	0	1
		TA0.4	1	1
P8.5/TA1.0	5	P8.5 (I/O)	I: 0; O: 1	0
		TA1.CCI0B	0	1
		TA1.0	1	1
P8.6/TA1.1	6	P8.6 (I/O)	I: 0; O: 1	0
		TA1.CCI1B	0	1
		TA1.1	1	1
P8.7	7	P8.7 (I/O)	I: 0; O: 1	0



### Port P9, P9.0 to P9.7, Input/Output With Schmitt Trigger

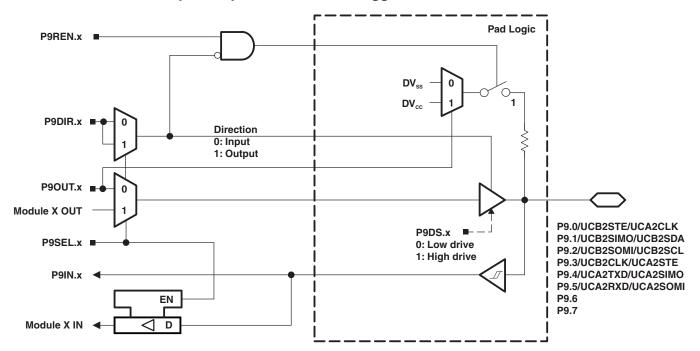


Table 57. Port P9 (P9.0 to P9.7) Pin Functions

PIN NAME (P9.x)		FUNCTION	CONTROL BIT	rs/signals <sup>(1)</sup>
	X	FUNCTION	P9DIR.x	P9SEL.x
P9.0/UCB2STE/UCA2CLK	0	P9.0 (I/O)	I: 0; O: 1	0
		UCB2STE/UCA2CLK <sup>(2)</sup> (3)	X	1
P9.1/UCB2SIMO/UCB2SDA	1	P9.1 (I/O)	I: 0; O: 1	0
		UCB2SIMO/UCB2SDA <sup>(2)</sup> (4)	Х	1
P9.2/UCB2SOMI/UCB2SCL	2	P9.2 (I/O)	I: 0; O: 1	0
		UCB2SOMI/UCB2SCL <sup>(2)</sup> (4)	Х	1
P9.3/UCB2CLK/UCA2STE	3	P9.3 (I/O)	I: 0; O: 1	0
		UCB2CLK/UCA2STE (2) (5)	Х	1
P9.4/UCA2TXD/UCA2SIMO	4	P9.4 (I/O)	I: 0; O: 1	0
		UCA2TXD/UCA2SIMO(2)	Х	1
P9.5/UCA2RXD/UCA2SOMI	5	P9.5 (I/O)	I: 0; O: 1	0
		UCA2RXD/UCA2SOMI(2)	Х	1
P9.6	6	P9.6 (I/O)	I: 0; O: 1	0
P9.7	7	P9.7 (I/O)	I: 0; O: 1	0

<sup>(1)</sup> X = Don't care

<sup>(2)</sup> The pin direction is controlled by the USCI module.

<sup>(3)</sup> UCA2CLK function takes precedence over UCB2STE function. If the pin is required as UCA2CLK input or output, USCI B2 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

<sup>(4)</sup> If the I2C functionality is selected, the output drives only the logical 0 to V<sub>SS</sub> level.

<sup>(5)</sup> UCB2CLK function takes precedence over UCA2STE function. If the pin is required as UCB2CLK input or output, USCI A2 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.



### Port P10, P10.0 to P10.7, Input/Output With Schmitt Trigger

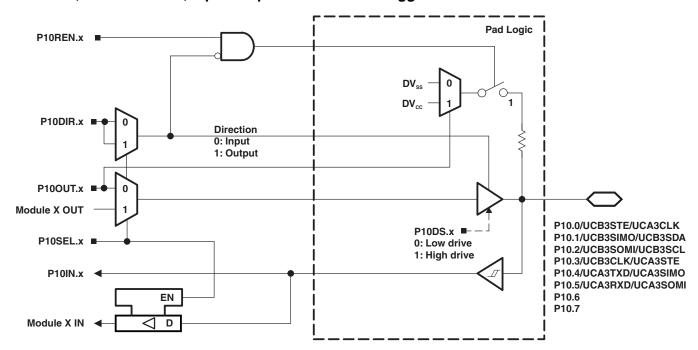


Table 58. Port P10 (P10.0 to P10.7) Pin Functions

DIN MARKE (D40)		FINATION	CONTROL BIT	CONTROL BITS/SIGNALS <sup>(1)</sup>		
PIN NAME (P10.x)	X	FUNCTION	P10DIR.x	P10SEL.x		
P10.0/UCB3STE/UCA3CLK	0	P10.0 (I/O)	I: 0; O: 1	0		
		UCB3STE/UCA3CLK <sup>(2)</sup> (3)	Х	1		
P10.1/UCB3SIMO/UCB3SDA	1	P10.1 (I/O)	I: 0; O: 1	0		
		UCB3SIMO/UCB3SDA <sup>(2)</sup> (4)	Х	1		
P10.2/UCB3SOMI/UCB3SCL	2	P10.2 (I/O)	I: 0; O: 1	0		
		UCB3SOMI/UCB3SCL <sup>(2)</sup> (4)	Х	1		
P10.3/UCB3CLK/UCA3STE	3	P10.3 (I/O)	I: 0; O: 1	0		
		UCB3CLK/UCA3STE (2) (5)	Х	1		
P10.4/UCA3TXD/UCA3SIMO	4	P10.4 (I/O)	I: 0; O: 1	0		
		UCA3TXD/UCA3SIMO(2)	Х	1		
P10.5/UCA3RXD/UCA3SOMI	5	P10.5 (I/O)	I: 0; O: 1	0		
		UCA3RXD/UCA3SOMI(2)	Х	1		
P10.6	6	P10.6 (I/O)	I: 0; O: 1	0		
		Reserved <sup>(6)</sup>	Х	1		
P10.7	7	P10.7 (I/O)	I: 0; O: 1	0		
		Reserved <sup>(6)</sup>	Х	1		

X = Don't care

The pin direction is controlled by the USCI module.

UCA3CLK function takes precedence over UCB3STE function. If the pin is required as UCA3CLK input or output, USCI B3 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

If the I2C functionality is selected, the output drives only the logical 0 to  $V_{SS}$  level. UCB3CLK function takes precedence over UCA3STE function. If the pin is required as UCB3CLK input or output, USCI A3 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

The secondary function on these pins are reserved for factory test purposes. Application should keep the P10SEL.x of these ports cleared to prevent potential conflicts with the application.



## Port P11, P11.0 to P11.2, Input/Output With Schmitt Trigger

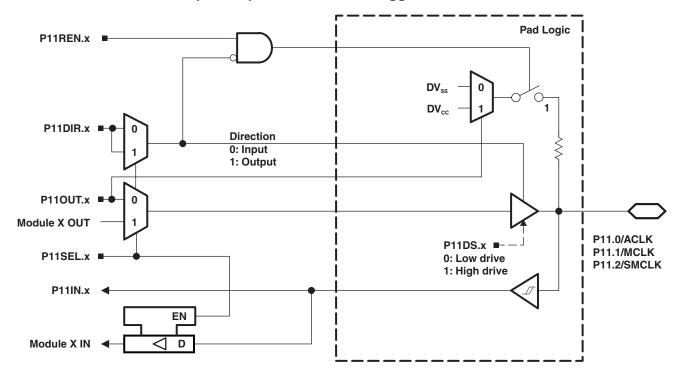
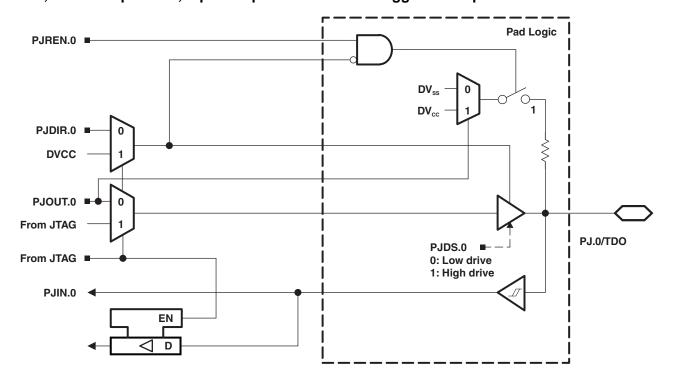


Table 59. Port P11 (P11.0 to P11.2) Pin Functions

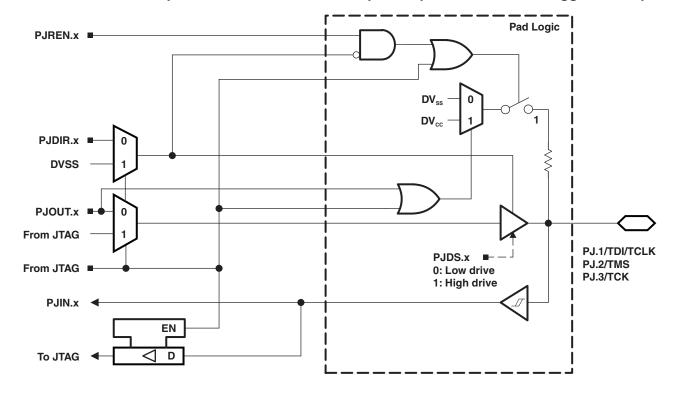
DIN NAME (D44)		FUNCTION	CONTROL BITS/SIGNALS	
PIN NAME (P11.x)	Х		P11DIR.x	P11SEL.x
P11.0/ACLK	0	P11.0 (I/O)	I: 0; O: 1	0
		ACLK	1	1
P11.1/MCLK	1	P11.1 (I/O)	I: 0; O: 1	0
		MCLK	1	1
P11.2/SMCLK	2	P11.2 (I/O)	I: 0; O: 1	0
		SMCLK	1	1



Port J, J.0 JTAG pin TDO, Input/Output With Schmitt Trigger or Output



Port J, J.1 to J.3 JTAG pins TMS, TCK, TDI/TCLK, Input/Output With Schmitt Trigger or Output





## Table 60. Port PJ (PJ.0 to PJ.3) Pin Functions

PIN NAME (PJ.x)	x	FUNCTION	CONTROL BITS/ SIGNALS <sup>(1)</sup>
, ,			PJDIR.x
PJ.0/TDO	0	PJ.0 (I/O) <sup>(2)</sup>	I: 0; O: 1
		TDO <sup>(3)</sup>	X
PJ.1/TDI/TCLK	1	PJ.1 (I/O) <sup>(2)</sup>	I: 0; O: 1
		TDI/TCLK <sup>(3)</sup> (4)	X
PJ.2/TMS	2	PJ.2 (I/O) <sup>(2)</sup>	I: 0; O: 1
		TMS <sup>(3)</sup> (4)	X
PJ.3/TCK	3	PJ.3 (I/O) <sup>(2)</sup>	I: 0; O: 1
		TCK <sup>(3)</sup> (4)	X

X = Don't care

Default condition

The pin direction is controlled by the JTAG module.
In JTAG mode, pullups are activated automatically on TMS, TCK, and TDI/TCLK. PJREN.x are do not care.



## **DEVICE DESCRIPTORS (TLV)**

Table 61 lists the complete contents of the device descriptor tag-length-value (TLV) structure for each device type.

Table 61. Device Descriptor Table (1)

	T.								
	Description	Address	Size	'F5438A	'F5437A	'F5436A	'F5435A	'F5419A	'F5418A
			bytes	Value	Value	Value	Value	Value	Value
Info Block	Info length	01A00h	1	06h	06h	06h	06h	06h	06h
	CRC length	01A01h	1	06h	06h	06h	06h	06h	06h
	CRC value	01A02h	2	per unit					
	Device ID	01A04h	1	05h	04h	03h	02h	01h	00h
	Device ID	01A05h	1	80h	80h	80h	80h	80h	80h
	Hardware revision	01A06h	1	per unit					
	Firmware revision	01A07h	1	per unit					
Die Record	Die Record Tag	01A08h	1	08h	08h	08h	08h	08h	08h
	Die Record length	01A09h	1	0Ah	0Ah	0Ah	0Ah	0Ah	0Ah
	Lot/Wafer ID	01A0Ah	4	per unit					
	Die X position	01A0Eh	2	per unit					
	Die Y position	01A10h	2	per unit					
	Test results	01A12h	2	per unit					
ADC12 Calibration	ADC12 Calibration Tag	01A14h	1	11h	11h	11h	11h	11h	11h
	ADC12 Calibration length	01A15h	1	10h	10h	10h	10h	10h	10h
	ADC Gain Factor	01A16h	2	per unit					
	ADC Offset	01A18h	2	per unit					
	ADC 1.5-V Reference Temp. Sensor 30°C	01A1Ah	2	per unit					
	ADC 1.5-V Reference Temp. Sensor 85°C	01A1Ch	2	per unit					
	ADC 2.0-V Reference Temp. Sensor 30°C	01A1Eh	2	per unit					
	ADC 2.0-V Reference Temp. Sensor 85°C	01A20h	2	per unit					
	ADC 2.5-V Reference Temp. Sensor 30°C	01A22h	2	per unit					
	ADC 2.5-V Reference Temp. Sensor 85°C	01A24h	2	per unit					
REF Calibration	REF Calibration Tag	01A26h	1	12h	12h	12h	12h	12h	12h
	REF Calibration length	01A27h	1	06h	06h	06h	06h	06h	06h



# Table 61. Device Descriptor Table<sup>(1)</sup> (continued)

			Size	'F5438A	'F5437A	'F5436A	'F5435A	'F5419A	'F5418A
	Description	Address	bytes	Value	Value	Value	Value	Value	Value
	REF 1.5-V Reference	01A28h	2	per unit					
	REF 2.0-V Reference	01A2Ah	2	per unit					
	REF 2.5-V Reference	01A2Ch	2	per unit					
Peripheral Descriptor	Peripheral Descriptor Tag	01A2Eh	1	02h	02h	02h	02h	02h	02h
	Peripheral Descriptor Length	01A2Fh	1	61h	059h	62h	5Ah	61h	59h
	Memory 1		2	08h 8Ah	08h 8Ah	08h 8Ah	08h 8Ah	08h 8Ah	08h 8Ah
	Memory 2		2	0Ch 86h	0Ch 86h	0Ch 86h	0Ch 86h	0Ch 86h	0Ch 86h
	Memory 3		2	0Eh 30h	0Eh 30h	0Eh 30h	0Eh 30h	0Eh 30h	0Eh 30h
	Memory 4		2	2Eh 98h	2Eh 98h	2Eh 97h	2Eh 97h	2Eh 96h	2Eh 96h
	Memory 5		0/1	NA	NA	94h	94h	NA	NA
	delimiter		1	00h	00h	00h	00h	00h	00h
	Peripheral count		1	21h	1Dh	21h	1Dh	21h	1Dh
	MSP430CPUXV2		2	00h 23h	00h 23h	00h 23h	00h 23h	00h 23h	00h 23h
	SBW		2	00h 0Fh	00h 0Fh	00h 0Fh	00h 0Fh	00h 0Fh	00h 0Fh
	EEM-8		2	00h 05h	00h 05h	00h 05h	00h 05h	00h 05h	00h 05h
	TI BSL		2	00h FCh	00h FCh	00h FCh	00h FCh	00h FCh	00h FCh
	Package		2	00h 1Fh	00h 1Fh	00h 1Fh	00h 1Fh	00h 1Fh	00h 1Fh
	SFR		2	10h 41h	10h 41h	10h 41h	10h 41h	10h 41h	10h 41h
	PMM		2	02h 30h	02h 30h	02h 30h	02h 30h	02h 30h	02h 30h
	FCTL		2	02h 38h	02h 38h	02h 38h	02h 38h	02h 38h	02h 38h
	CRC16-straight		2	01h 3Ch	01h 3Ch	01h 3Ch	01h 3Ch	01h 3Ch	01h 3Ch
	CRC16-bit reversed		2	00h 3Dh	00h 3Dh	00h 3Dh	00h 3Dh	00h 3Dh	00h 3Dh
	RAMCTL		2	00h 44h	00h 44h	00h 44h	00h 44h	00h 44h	00h 44h
	WDT_A		2	00h 40h	00h 40h	00h 40h	00h 40h	00h 40h	00h 40h
	UCS		2	01h 48h	01h 48h	01h 48h	01h 48h	01h 48h	01h 48h
	SYS		2	02h 42h	02h 42h	02h 42h	02h 42h	02h 42h	02h 42h
	REF		2	03h A0h	03h A0h	03h A0h	03h A0h	03h A0h	03h A0h
	Port 1/2		2	05h 51h	05h 51h	05h 51h	05h 51h	05h 51h	05h 51h



# Table 61. Device Descriptor Table<sup>(1)</sup> (continued)

		1							1554404
	Description	Address	Size bytes	'F5438A	'F5437A	'F5436A	'F5435A	'F5419A	'F5418A
			Dytes	Value	Value	Value	Value	Value	Value
	Port 3/4		2	02h 52h	02h 52h	02h 52h	02h 52h	02h 52h	02h 52h
	Port 5/6		2	02h 53h	02h 53h	02h 53h	02h 53h	02h 53h	02h 53h
	Port 7/8		2	02h 54h	02h 54h	02h 54h	02h 54h	02h 54h	02h 54h
	Port 9/10		2	02h 55h	NA	02h 55h	NA	02h 55h	NA
	Port 11/12		2	02h 56h	NA	02h 56h	NA	02h 56h	NA
	JTAG		2	08h 5Fh	0Ch 5Fh	08h 5Fh	0Ch 5Fh	08h 5Fh	0Ch 5Fh
	TA0		2	02h 62h	02h 62h	02h 62h	02h 62h	02h 62h	02h 62h
	TA1		2	04h 61h	04h 61h	04h 61h	04h 61h	04h 61h	04h 61h
	TB0		2	04h 67h	04h 67h	04h 67h	04h 67h	04h 67h	04h 67h
	RTC		2	0Eh 68h	0Eh 68h	0Eh 68h	0Eh 68h	0Eh 68h	0Eh 68h
	MPY32		2	02h 85h	02h 85h	02h 85h	02h 85h	02h 85h	02h 85h
	DMA-3		2	04h 47h	04h 47h	04h 47h	04h 47h	04h 47h	04h 47h
	USCI_A/B		2	0Ch 90h	0Ch 90h	0Ch 90h	0Ch 90h	0Ch 90h	0Ch 90h
	USCI_A/B		2	04h 90h	04h 90h	04h 90h	04h 90h	04h 90h	04h 90h
	USCI_A/B		2	04h 90h	NA	04h 90h	NA	04h 90h	NA
	USCI_A/B		2	04h 90h	NA	04h 90h	NA	04h 90h	NA
	ADC12_A		2	08h D1h	10h D1h	08h D1h	10h D1h	08h D1h	10h D1h
Interrupts	TB0.CCIFG0		1	64h	64h	64h	64h	64h	64h
•	TB0.CCIFG16		1	65h	65h	65h	65h	65h	65h
	WDTIFG		1	40h	40h	40h	40h	40h	40h
	USCI_A0		1	90h	90h	90h	90h	90h	90h
	USCI_B0		1	91h	91h	91h	91h	91h	91h
	ADC12_A		1	D0h	D0h	D0h	D0h	D0h	D0h
	TA0.CCIFG0		1	60h	60h	60h	60h	60h	60h
	TA0.CCIFG14		1	61h	61h	61h	61h	61h	61h
	USCI_A2		1	94h	01h	94h	01h	94h	01h
	USCI_B2		1	95h	01h	95h	01h	95h	01h
	DMA		1	46h	46h	46h	46h	46h	46h
	TA1.CCIFG0		1	62h	62h	62h	62h	62h	62h
	TA1.CCIFG12		1	63h	63h	63h	63h	63h	63h
	P1		1	50h	50h	50h	50h	50h	50h
	USCI_A1		1	92h	92h	92h	92h	92h	92h
	USCI_B1		1	93h	93h	93h	93h	93h	93h
	USCI_A3		1	96h	01h	96h	01h	96h	01h
	USCI_B3		1	97h	01h	97h	01h	97h	01h



# Table 61. Device Descriptor Table<sup>(1)</sup> (continued)

	Description	Adduses	Size	'F5438A	'F5437A	'F5436A	'F5435A	'F5419A	'F5418A	
		Address	bytes	Value	Value	Value	Value	Value	Value	
	P2		1	51h	51h	51h	51h	51h	51h	
	RTC_A		1	68h	68h	68h	68h	68h	68h	
	delimiter		1	00h	00h	00h	00h	00h	00h	



## **REVISION HISTORY**

REVISION	DESCRIPTION						
SLAS655	Product Preview release						
SLAS655A	Production Data release						
SLAS655B	Changed f <sub>XT1,HF,SW</sub> MIN from 1.5 MHz to 0.7 MHz, page 48						
OE (COOOD	Changed f <sub>XT2,HF,SW</sub> MIN from 1.5 MHz to 0.7 MHz, page 49						
	Features, Changed Wake-Up From Standby Mode time to 3.5 μs.						
	Table 3, Changed ACLK description.						
	Table 8, Changed SYSRSTIV interrupt event at 1Ch to Reserved.						
	Recommended Operating Conditions, Added note regarding interaction between minimum VCC and SVS. Adde typical test conditions.						
	DCO Frequency, Added note.						
SLAS655C	Renamed flash banks from numerical to alphabetical to match user's guide.						
	Updated ADC12_A, REF, and PMM electrical specifications throughout.						
	12-Bit ADC, Temperature Sensor and Built-In VMID, Changed t <sub>SENSOR(sample)</sub> to 100 µs MIN and changed note 2.						
	Wake-Up From Low-Power Modes and Reset, Updated wakeup times.						
	Input/Output Schematics, Corrected notes regarding USCI CLK functions taking precedence over USCI STE functions.						
	Port P4, P4.0 to P4.7, Input/Output With Schmitt Trigger, Corrected schematic.						



2-Oct-2012

### **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
MSP430F5418AIPN	ACTIVE	LQFP	PN	80	119	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5418AIPNR	ACTIVE	LQFP	PN	80	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5419AIPZ	ACTIVE	LQFP	PZ	100	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5419AIPZR	ACTIVE	LQFP	PZ	100	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5419AIZQW	OBSOLETE	BGA MICROSTAR JUNIOR	ZQW	113		TBD	Call TI	Call TI	
MSP430F5419AIZQWR	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	2500	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	
MSP430F5419AIZQWT	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	
MSP430F5435AIPN	ACTIVE	LQFP	PN	80	119	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5435AIPNR	ACTIVE	LQFP	PN	80	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5436AIPZ	ACTIVE	LQFP	PZ	100	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5436AIPZR	ACTIVE	LQFP	PZ	100	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5436AIZQW	OBSOLETE	BGA MICROSTAR JUNIOR	ZQW	113		TBD	Call TI	Call TI	
MSP430F5436AIZQWR	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	2500	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	
MSP430F5436AIZQWT	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	





2-Oct-2012

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
MSP430F5437AIPN	ACTIVE	LQFP	PN	80	119	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5437AIPNR	ACTIVE	LQFP	PN	80	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5438ACY	ACTIVE	DIESALE	Υ	0	64	Green (RoHS & no Sb/Br)	Call TI	N / A for Pkg Type	
MSP430F5438ACYS	ACTIVE	WAFERSALE	YS	0	1	TBD	Call TI	Call TI	
MSP430F5438AGACYS	ACTIVE	WAFERSALE	YS	0	1	TBD	Call TI	Call TI	
MSP430F5438AIPZ	ACTIVE	LQFP	PZ	100	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5438AIPZR	ACTIVE	LQFP	PZ	100	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
MSP430F5438AIZQWR	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	2500	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	
MSP430F5438AIZQWT	ACTIVE	BGA MICROSTAR JUNIOR	ZQW	113	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	
XMS430F5438AIPZ	OBSOLETE	LQFP	PZ	100		TBD	Call TI	Call TI	
XMS430F5438AIPZR	OBSOLETE	LQFP	PZ	100		TBD	Call TI	Call TI	

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



### PACKAGE OPTION ADDENDUM

2-Oct-2012

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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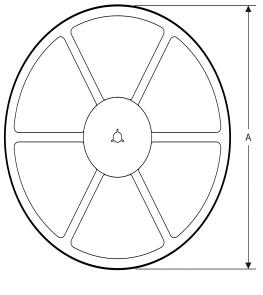
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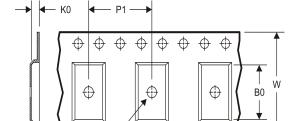
**TAPE DIMENSIONS** 

Cavity

## TAPE AND REEL INFORMATION

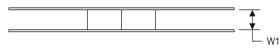
#### **REEL DIMENSIONS**





**◆** A0 **▶** 

A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

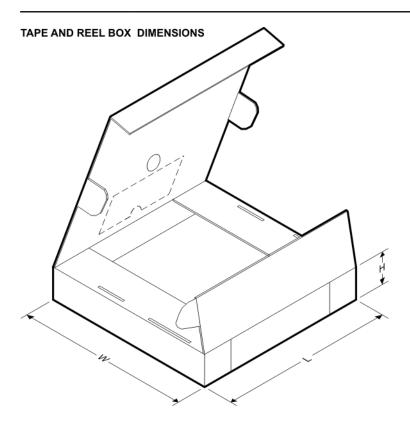


#### TAPE AND REEL INFORMATION

\*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430F5418AIPNR	LQFP	PN	80	1000	330.0	24.4	14.6	14.6	1.9	20.0	24.0	Q2
MSP430F5419AIZQWT	BGA MI CROSTA R JUNI OR	ZQW	113	250	330.0	16.4	7.3	7.3	1.5	12.0	16.0	Q1
MSP430F5436AIZQWT	BGA MI CROSTA R JUNI OR	ZQW	113	250	330.0	16.4	7.3	7.3	1.5	12.0	16.0	Q1
MSP430F5438AIZQWT	BGA MI CROSTA R JUNI OR	ZQW	113	250	330.0	16.4	7.3	7.3	1.5	12.0	16.0	Q1

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#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430F5418AIPNR	LQFP	PN	80	1000	367.0	367.0	45.0
MSP430F5419AIZQWT	BGA MICROSTAR JUNIOR	ZQW	113	250	336.6	336.6	28.6
MSP430F5436AIZQWT	BGA MICROSTAR JUNIOR	ZQW	113	250	336.6	336.6	28.6
MSP430F5438AIZQWT	BGA MICROSTAR JUNIOR	ZQW	113	250	336.6	336.6	28.6

# ZQW (S-PBGA-N113)

## PLASTIC BALL GRID ARRAY



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MO-225
- D. This is a Pb-free solder ball design.

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## PN (S-PQFP-G80)

#### PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026

## PZ (S-PQFP-G100)

#### PLASTIC QUAD FLATPACK

1



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026

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