

EE3025 Assignment-1

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Download all python codes from

<https://github.com/yuvateja-ctrl/EE3025/tree/main/codes>

and latex-tikz codes from

<https://github.com/yuvateja-ctrl/EE3025>

```
#write the output signal into .wav file
sf.write('Sound_With_ReducedNoise.wav',
        output_signal, fs)
```

```
print(a)
print(b)
```

1 DIGITAL FILTER

1.1 Download the sound file from

`wget https://raw.githubusercontent.com/gadepall/EE1310/master/filter/codes/Sound_Noise.wav`

1.2 Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal

#read .wav file
input_signal,fs = sf.read('Sound_Noise.wav')

#sampling frequency of Input signal
sampl_freq=fs

#order of the filter
order=4

#cutoff frquency 4kHz
cutoff_freq=4000.0

#digital frequency
Wn=2*cutoff_freq/sampl_freq

# b and a are numerator and denominator
  polynomials respectively
b, a = signal.butter(order,Wn, 'low')

#filter the input signal with butterworth filter
output_signal = signal.filtfilt(b, a, input_signal)
#output_signal = signal.lfilter(b, a, input_signal)
```

2 DIFFERENCE EQUATION

2.1 Write the difference equation of above Digital filter obtained in problem 1.2.

Solution:

$$\sum_{m=0}^M a(m) y(n-m) = \sum_{k=0}^N b(k) x(n-k) \quad (2.0.1)$$

$$y(n) - 2.52y(n-1) + 2.56y(n-2) - 1.206y(n-3) + 0.22013y(n-4) = 0.00345x(n) + 0.0138x(n-1) + 0.020725x(n-2) + 0.0138x(n-3) + 0.00345x(n-4) \quad (2.0.2)$$

2.2 Sketch $x(n]$ and $y(n]$.

Solution: The following code yields Fig 2.2

`codes/inp_out.py`

The filtered sound signal obtained through difference equation is found in

`codes/Sound_de.wav`

3 Z-TRANSFORM

3.1

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (3.0.1)$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \quad (3.0.2)$$

and find

$$\mathcal{Z}\{x(n-k)\} \quad (3.0.3)$$

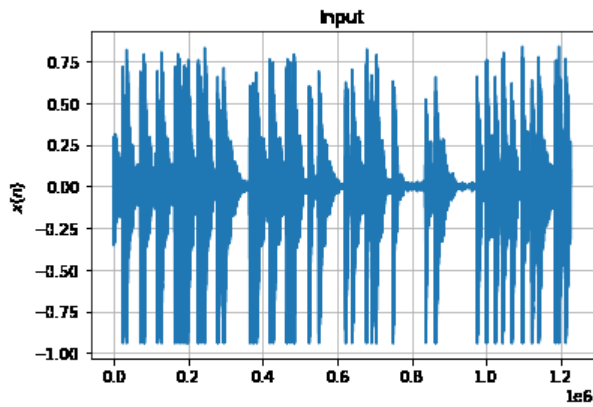


Fig. 0

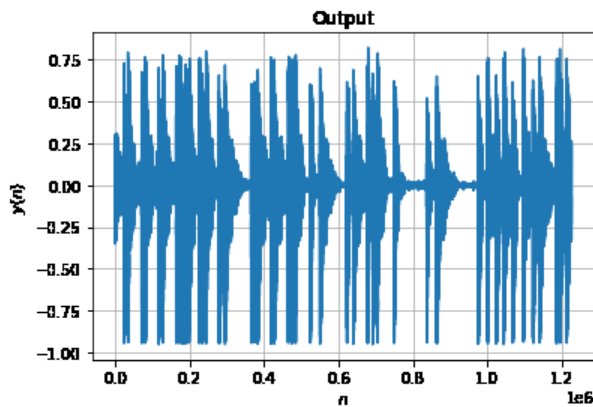


Fig. 0

Solution: Applying 3.0.6 in 2.0.2 we get,

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b[0] + b[1]z^{-1} + b[2]z^{-2} + b[3]z^{-3} + b[4]z^{-4}}{a[0] + a[1]z^{-1} + a[2]z^{-2} + a[3]z^{-3} + a[4]z^{-4}} \quad (3.0.8)$$

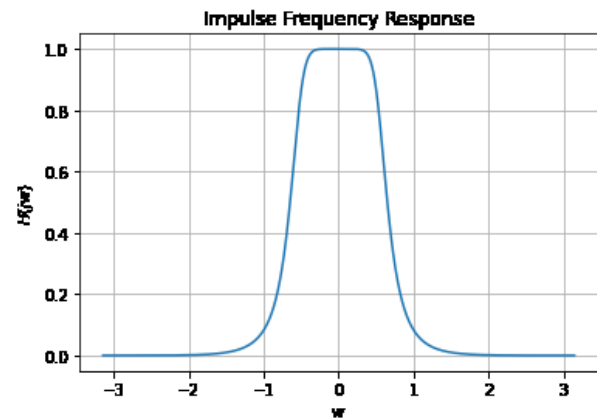
Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \quad (3.0.9)$$

Plot $|H(e^{j\omega})|$.

Solution: The following code plots Fig 3.3.

codes/dtft.py

Fig. 0: $|H(e^{j\omega})|$

4 IMPULSE RESPONSE

4.1 From the difference equation eq. 2.0.2. Sketch $h(n)$

Solution: we know that output would be impulse response when the given input is impulse. From 2.0.1,

By substituting $x(n-k) = \delta(n-k)$, then $y(n-k)$ becomes $h(n-k)$ for all $k=0,1,2,3,4$. Now, the following code plots Fig 4.1

codes/impulseRes_def.py

4.2 Check whether $h(n)$ obtained is stable.

Solution: The system is defined by the equation 2.0.2

By BIBO(Bounded Input Bounded Output) we know that if the input is bounded then the output should also be bounded.

Since $x(n)$ is bounded, let B_x be some finite value

$$|y(n)| \leq B_x \sum_{k=-\infty}^{\infty} |x(n-k)| \quad (4.0.1)$$

Solution: From 3.0.1,

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \quad (3.0.4)$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-(n+k)} = z^{-k} \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (3.0.5)$$

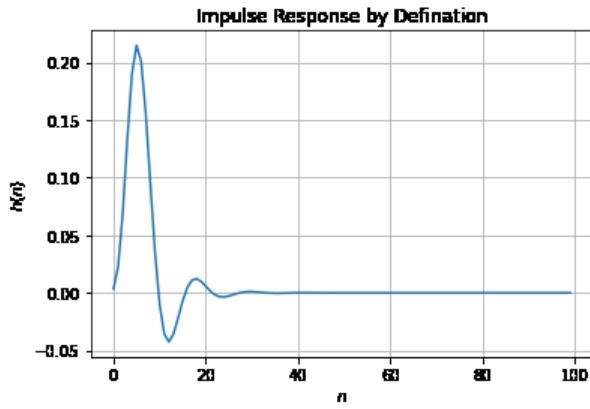
resulting in 3.0.2. Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \quad (3.0.6)$$

Find

$$H(z) = \frac{Y(z)}{X(z)} \quad (3.0.7)$$

from 2.0.2 assuming that the Z-transform is a linear operation.

Fig. 0: $h(n)$

From convolution formula,

$$|y(n)| = \left| \sum_{-\infty}^{\infty} h(k)x(n-k) \right| \quad (4.0.2)$$

$$|y(n)| \leq \sum_{-\infty}^{\infty} |h(k)| |x(n-k)| \quad (4.0.3)$$

Let B_x be the maximum value $x(n-k)$ can take, then

$$|y(n)| \leq B_x \sum_{-\infty}^{\infty} |h(k)| \quad (4.0.4)$$

If

$$\sum_{-\infty}^{\infty} |h(k)| < \infty \quad (4.0.5)$$

Then

$$|y(n)| \leq B_y < \infty \quad (4.0.6)$$

Hence $y(n)$ is bounded if both input $x(n)$ and system transfer function $h(n)$ are bounded.

Here as audio input is also bounded we can say the system is said to be stable.

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty \quad (4.0.7)$$

The above equation can be rewritten as,

$$\sum_{n=-\infty}^{\infty} |h(n)z^{-n}|_{|z|=1} < \infty \quad (4.0.8)$$

$$\sum_{n=-\infty}^{\infty} |h(n)| |z^{-n}|_{|z|=1} < \infty \quad (4.0.9)$$

From Triangle inequality,

$$\left| \sum_{n=-\infty}^{\infty} h(n)z^{-n} \right|_{|z|=1} < \infty \quad (4.0.10)$$

$$\Rightarrow |H(n)|_{|z|=1} < \infty \quad (4.0.11)$$

Therefore, the Region of Convergence (ROC) should include the unit circle for the system to be stable. Since, $h(n)$ is right sided the ROC is outside the outer most pole. From the equation 3.0.8 Poles of the given transfer equation is:

$$z(\text{approx}) = 0.69382 \pm 0.41i, \quad 0.56617835 \pm 0.134423 \quad (4.0.12)$$

From the above poles, we can see that the ROC of the system is $|z| > \sqrt{0.69382^2 + 0.41^2}$. From

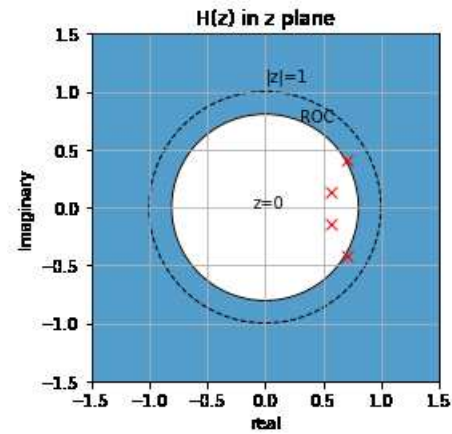


Fig. 0: ROC plot

the figure we can observe that ROC of the system includes unit circle $|z| = 1$. The code for plotting the figure is:

```
codes/roc.py
```

Which implies that the given IIR filter is stable, because $h(n)$ is absolutely summable. And the given system is stable.

4.3 Compute Filtered output using convolution formula using $h(n)$ obtained in 4.1

$$y(n) = x(n) * h(n) = \sum_{n=-\infty}^{\infty} x(k)h(n-k) \quad (4.0.13)$$

Solution: The following code plots Fig 4.3

```
/codes/out_conv.py
```

The filtered sound signal through convolution from this method is found in

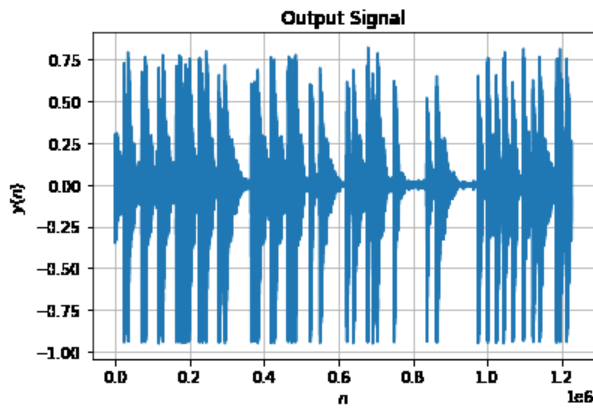


Fig. 0: $y(n)$ from the definition of convolution

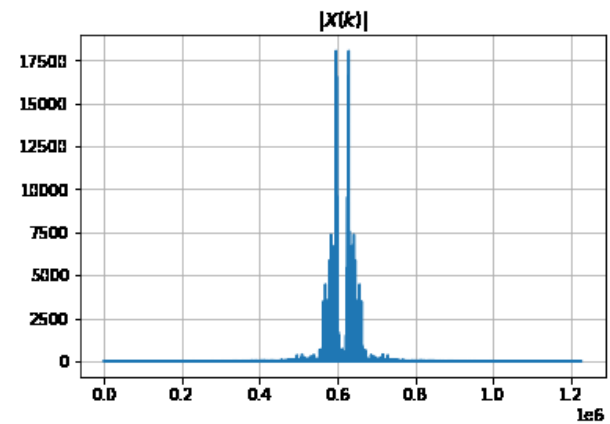


Fig. 0: Magnitude of $X(k)$

codes/Sound_conv.wav

We can observe that the output obtained is same as $y(n)$ obtained in Fig 2.2

5 FFT AND IFFT

5.1 compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.1)$$

and $H(k)$ using $h(n)$.

Solution: For this given IIR system with audio sample as $x(n)$ and $h(n)$ as impulse response $h(n)$ obtained in 4.1 DFT of a Input Signal $x(n)$ is

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.2)$$

DFT of a Impulse Response $h(n)$ is

$$H(k) \triangleq \sum_{n=0}^{N-1} h(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.3)$$

The following code plots FFT of $x(n)$ and $h(n)$.

codes/xn_hn_fft.py

Magnitude and Phase plots obtained through above code is

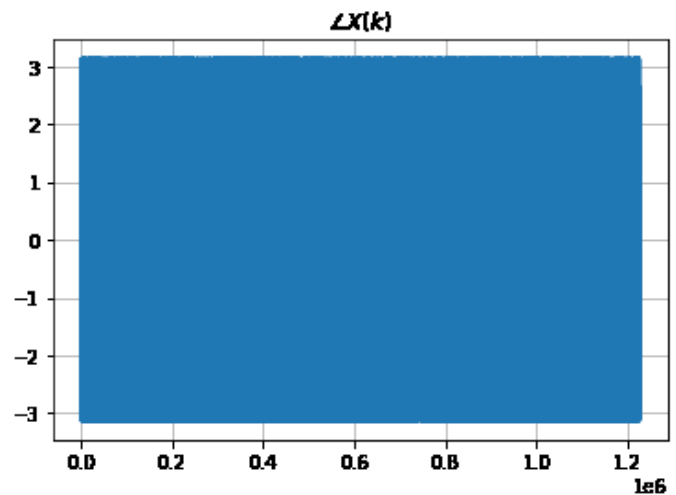


Fig. 0: Phase of $X(k)$

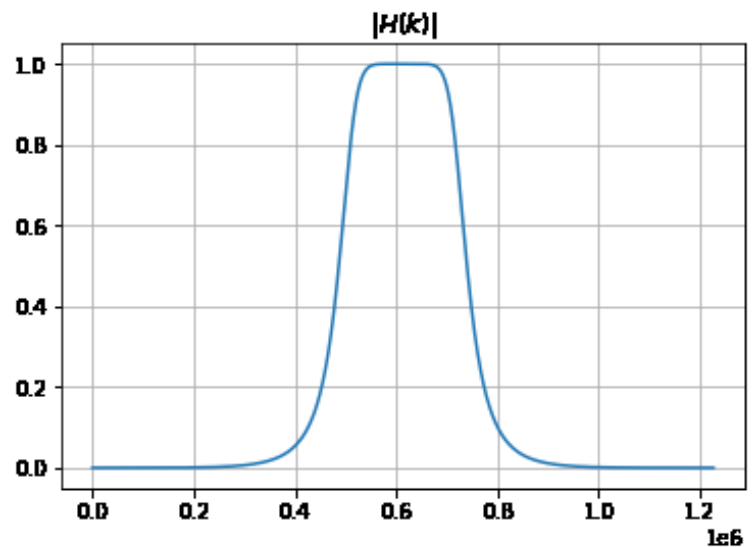


Fig. 0: Magnitude of $H(k)$

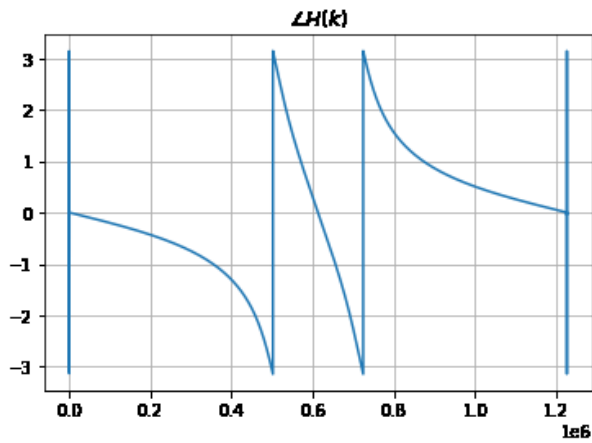


Fig. 0: Phase of $H(k)$

5.2 From

$$Y(k) = X(k)H(k) \quad (5.0.4)$$

Compute

$$y(n) \triangleq \sum_{k=0}^{N-1} Y(k)e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1 \quad (5.0.5)$$

Solution:

The following code plots Fig 5.2

```
codes/out_fft.py
```

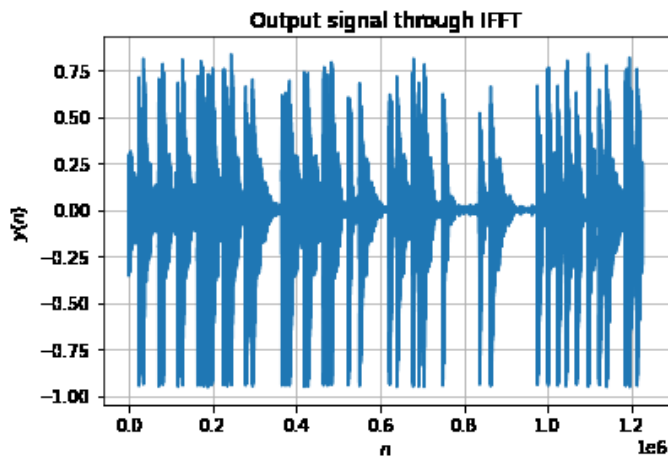


Fig. 0: $y(n)$

The filtered sound signal from this method is found in

```
codes/Sound_fft.wav
```

We can observe from the above plot that it is same as the $y(n)$ observed in Fig 2.2