

# Introduction to Linear Dynamical Systems

## Engineering Examples and State-Space Conversion

Dr. Yu Zhang

Department of Electrical and Computer Engineering  
University of California, Santa Cruz

Fall 2025

# Engineering Examples of Linear Dynamical Systems

Linear dynamical systems arise in many fields of engineering and science:

- Electrical circuits (RLC networks, filters)
- Mechanical systems (mass-spring-damper, suspension)
- Aerospace (aircraft stability, flight control)
- Robotics (joint and manipulator dynamics)
- Communications (filters, channel models)
- Power systems (generator swing dynamics, grid stability)
- Economics (linearized market and supply-demand models)

# General State-Space Model

## Continuous-Time State-Space Equations:

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

## Definitions:

- $x(t) \in \mathbb{R}^n$ : State vector (internal variables that capture system dynamics)
- $u(t) \in \mathbb{R}^m$ : Input vector (external signals that drive the system)
- $y(t) \in \mathbb{R}^p$ : Output vector (measured or controlled variables)

# The Matrices in State-Space Form

## System Matrix $A \in \mathbb{R}^{n \times n}$

- Governs how the state evolves on its own (without input)
- Encodes natural dynamics, eigenvalues, and stability
- Example: In a mass–spring–damper system,  $A$  determines natural frequency and damping
- *Intuition:*  $A$  tells you “if I leave the system alone, how do states evolve?”

## Input Matrix $B \in \mathbb{R}^{n \times m}$

- Maps the external input  $u(t)$  into the state equations
- Determines how each input affects each state
- *Intuition:*  $B$  tells you “when I push on the system, which states move and how strongly?”

# The Matrices in State-Space Form (cont.)

## Output Matrix $C \in \mathbb{R}^{p \times n}$

- Maps the internal states to the measurable outputs
- Determines what part of the state vector is visible at the output
- *Intuition:*  $C$  tells you “which states am I actually measuring or observing?”

## Feedthrough/Direct Matrix $D \in \mathbb{R}^{p \times m}$

- Maps the inputs directly to the outputs (without going through the state)
- $D = 0$  in many physical systems (no instantaneous effect), but not always
- Example: In resistor circuits, input voltage can instantly affect output current
- *Intuition:*  $D$  tells you “does the input show up instantly in the output, bypassing states?”

# Example: Electrical Circuit (RLC)

**System:** Series RLC circuit driven by voltage  $u(t)$ .

Governing equations:

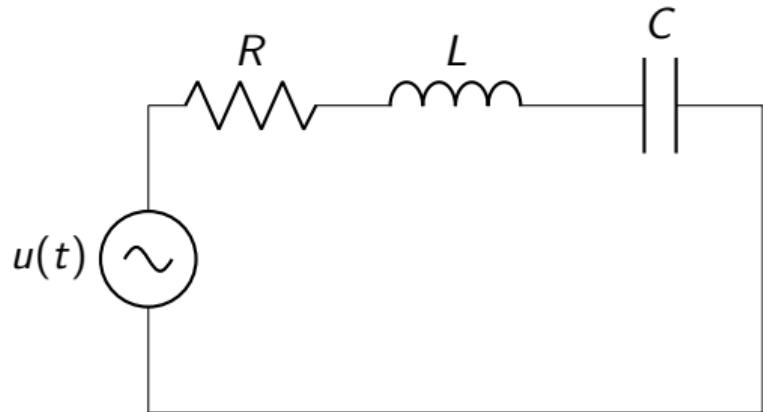
$$Ri(t) + L\frac{di(t)}{dt} + u_c(t) = u(t)$$

state-space form:

$$\dot{x}_1 = x_2,$$

$$\dot{x}_2 = -\frac{R}{L}x_2 - \frac{1}{LC}x_1 + \frac{1}{L}u(t),$$

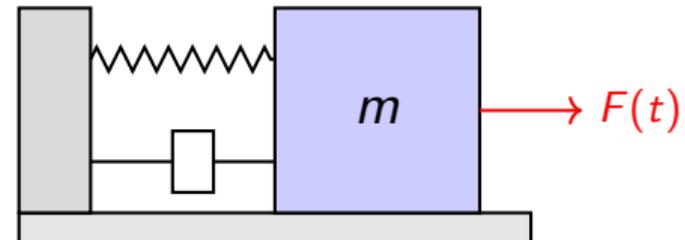
where  $x_1 = Cu_c(t)$ ,  $x_2 = i(t)$ .



Linear ODE system from KVL/KCL.

# Mechanical System: Mass–Spring–Damper

- Block of mass  $m$  attached to wall with spring (stiffness  $k$ ) and damper (damping  $c$ ).
- The spring (like a rubber band) wants to restore position. The damper (shock absorber) resists motion and kills vibrations, which **improves stability and comfort**
- External force  $F(t)$  applied to the block.  
Displacement from equilibrium:  $x(t)$ .



*Models vibrations and suspension dynamics.*

The damper provides a force proportional to velocity that opposes motion:

- **Dissipates energy** → prevents endless oscillations.
- **Controls response type:**
  - Underdamped ( $c < c_{\text{crit}}$ ): oscillatory decay.
  - Critically damped ( $c = c_{\text{crit}}$ ): fastest return to equilibrium without oscillation.
  - Overdamped ( $c > c_{\text{crit}}$ ): no oscillation, but sluggish return.

# Governing Equation

## Newton's Second Law:

$$m\ddot{x}(t) = F_{\text{spring}} + F_{\text{damper}} + F_{\text{external}}.$$

- Spring force:  $F_{\text{spring}} = -kx(t)$  (Hooke's Law).
- Damper force:  $F_{\text{damper}} = -c\dot{x}(t)$ .
- External force:  $F_{\text{external}} = F(t)$ .

Final governing ODE:

$$m\ddot{x}(t) + c\dot{x}(t) + kx(t) = F(t).$$

# Why a Linear Dynamical System?

- The equation is **linear** in  $x(t)$ ,  $\dot{x}(t)$ , and  $\ddot{x}(t)$ .
- It is **dynamical**: the future state depends on displacement, velocity, and the input force.
- Compact representation possible using state-space form.

$$m\ddot{x}(t) + c\dot{x}(t) + kx(t) = F(t)$$

# State-Space Representation

Define states:

$$x_1 = x(t), \quad x_2 = \dot{x}(t).$$

Then:

$$\dot{x}_1 = x_2, \quad \dot{x}_2 = -\frac{k}{m}x_1 - \frac{c}{m}x_2 + \frac{1}{m}F(t).$$

state-space matrices:

$$A = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{c}{m} \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}, \quad C = [1 \quad 0], \quad D = [0].$$

# Physical Interpretation

- **Matrix A:** Intrinsic system dynamics (interaction of displacement and velocity).
- **Matrix B:** How external input force  $F(t)$  affects the system.
- **Matrix C:** Chooses displacement  $x(t)$  as output.
- **Matrix D:** Zero (input does not appear directly in output).

$$\dot{x} = Ax + Bu, \quad y = Cx + Du$$

# Engineering Applications

The mass–spring–damper model underlies many real systems:

- **Automotive suspensions:** ride comfort and vibration isolation.
- **Civil engineering:** building dynamics under wind or earthquake loads.
- **Robotics/aerospace:** vibration damping in structures and arms.
- **Control theory:** canonical test system for PID and state-feedback design.

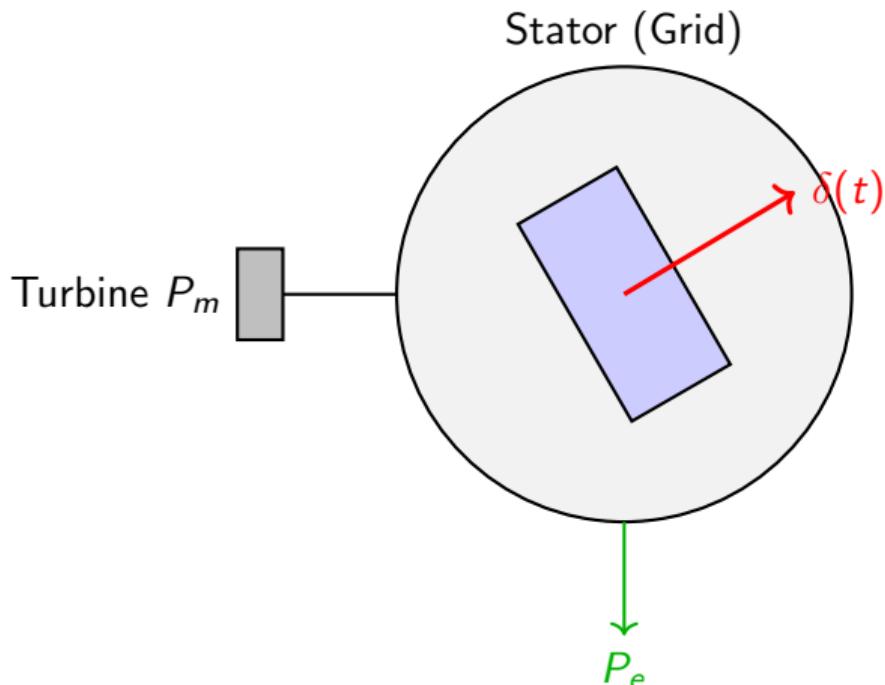
# Power Grid Example: Swing Equation (Generator Rotor Dynamics)

- Rotor angle  $\delta(t)$  is measured relative to the synchronous reference.
- First derivative:  $\dot{\delta}(t) = \Delta\omega(t)$  is speed deviation.
- Second derivative follows from the **Swing Equation**:

$$M \ddot{\delta}(t) + D \dot{\delta}(t) = P_m(t) - P_e(\delta(t)),$$

where:

- $M$  inertia, resists acceleration.
- $D$  damping, resists oscillations.
- $P_m$  mechanical input power,  $P_e$  electrical output power.



# Operating Point and Small Deviations

- Choose an operating point ( $\delta_0$ ,  $\dot{\delta} = 0$ ) with power balance

$$P_{m0} = P_e(\delta_0).$$

- Define small deviations:

$$\Delta\delta = \delta - \delta_0, \quad \Delta\omega = \dot{\delta}, \quad \Delta P_m = P_m - P_{m0}.$$

- Linearize electrical power around  $\delta_0$ :

$$P_e(\delta) \approx P_e(\delta_0) + K_s \Delta\delta, \quad K_s = \left. \frac{dP_e}{d\delta} \right|_{\delta_0}.$$

# Linearization around the Operating Point

Start from

$$M \ddot{\delta} + D \dot{\delta} = P_m - P_e(\delta).$$

Subtract the equilibrium and use the linearization:

$$M \Delta \ddot{\delta} + D \Delta \dot{\delta} = \Delta P_m - K_s \Delta \delta.$$

**State-space form** with  $x = \begin{bmatrix} \Delta \delta \\ \Delta \omega \end{bmatrix}$ ,  $\Delta \omega = \Delta \dot{\delta}$ ,  $u = \Delta P_m$ :

$$\begin{aligned} \Delta \dot{\delta} &= \Delta \omega, \\ M \Delta \dot{\omega} &= \Delta P_m - K_s \Delta \delta - D \Delta \omega, \end{aligned} \quad \Rightarrow \quad \dot{x} = \underbrace{\begin{bmatrix} 0 & 1 \\ -\frac{K_s}{M} & -\frac{D}{M} \end{bmatrix}}_A x + \underbrace{\begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix}}_B u.$$

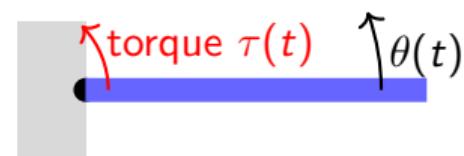
# Robotics Example: Single-Joint Position Control

## Joint Dynamics:

$$J\ddot{\theta}(t) + b\dot{\theta}(t) = \tau(t).$$

State-space:

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{b}{J} \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{1}{J} \end{bmatrix} \tau(t), \quad x = \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix}$$



## Interpretation:

- Models one rotary robotic joint.
- $J$ : inertia  $\rightarrow$  resists acceleration.
- $b$ : damping  $\rightarrow$  frictional losses.
- $\tau$ : motor torque (control input).

# From ODE to State-Space (1/2)

**General idea:** Any higher-order ODE can be rewritten as a first-order system.

**General  $n$ th-order ODE:**

$$a_n y^{(n)} + a_{n-1} y^{(n-1)} + \cdots + a_1 \dot{y} + a_0 y = b_0 u(t).$$

**Step 1. Define the states**

$$x_1(t) = y(t), \quad x_2(t) = \dot{y}(t), \quad \dots, \quad x_n(t) = y^{(n-1)}(t).$$

**Step 2. Write the state equations**

$$\dot{x}_1(t) = x_2(t), \quad \dot{x}_2(t) = x_3(t), \quad \dots, \quad \dot{x}_{n-1}(t) = x_n(t),$$

$$\dot{x}_n(t) = -\frac{a_0}{a_n}x_1(t) - \frac{a_1}{a_n}x_2(t) - \cdots - \frac{a_{n-1}}{a_n}x_n(t) + \frac{b_0}{a_n}u(t).$$

## From ODE to State-Space (2/2)

### Step 3. Express in matrix form

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t),$$

with

$$A = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & 0 & \cdots & 0 & 1 \\ -\frac{a_0}{a_n} & -\frac{a_1}{a_n} & \cdots & -\frac{a_{n-2}}{a_n} & -\frac{a_{n-1}}{a_n} \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ \frac{b_0}{a_n} \end{bmatrix},$$
$$C = [1 \ 0 \ \cdots \ 0], \quad D = [0].$$

This representation is the *companion form* (or canonical form), which is useful for analysis (stability, modes) and controller design. This works for any linear constant-coefficient ODE.

## Example: 2nd-Order ODE to state-space

Mass-spring-damper ODE:

$$m\ddot{x}(t) + c\dot{x}(t) + kx(t) = F(t).$$

Define states:

$$x_1 = x(t), \quad x_2 = \dot{x}(t).$$

State equations:

$$\dot{x}_1 = x_2, \quad \dot{x}_2 = -\frac{k}{m}x_1 - \frac{c}{m}x_2 + \frac{1}{m}F(t).$$

Matrix form:

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{c}{m} \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix} u, \quad y = [1 \ 0] x.$$

## Example: 3rd-Order ODE to state-space

Consider:

$$y^{(3)} + a_2 y^{(2)} + a_1 y^{(1)} + a_0 y = b_0 u(t).$$

Define states:

$$x_1 = y, \quad x_2 = \dot{y}, \quad x_3 = \ddot{y}.$$

State equations:

$$\dot{x}_1 = x_2, \quad \dot{x}_2 = x_3, \quad \dot{x}_3 = -a_0 x_1 - a_1 x_2 - a_2 x_3 + b_0 u.$$

Matrix form:

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ b_0 \end{bmatrix} u, \quad y = [1 \ 0 \ 0] x.$$

# Linearity in ODEs

## Definition

A system is called *linear* if its equations are linear in the unknown variables and their derivatives. This means the **superposition principle** holds.

## High-order Linear ODEs

An  $n$ th-order ODE of the form

$$a_n \frac{d^n y}{dt^n} + a_{n-1} \frac{d^{n-1} y}{dt^{n-1}} + \cdots + a_1 \frac{dy}{dt} + a_0 y(t) = b_0 u(t) + b_1 \frac{du}{dt} + \cdots + b_m \frac{d^m u}{dt^m}$$

is linear because

- $y(t)$  and its derivatives appear only to the first power.
- No products like  $y^2$ ,  $y\dot{y}$ , or  $\sin(y)$ .
- Coefficients  $\{a_i, b_i\}$  are constants (or functions of  $t$ , but not of  $y$ ).

# Linearity in State-Space Models

## State-Space Models

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

This is linear because both equations are linear functions of  $x(t)$  and  $u(t)$ . There are no products (e.g.,  $x_1x_2$ ) or nonlinear functions (e.g.,  $\sin(x)$ ).

## Superposition Principle

If  $(x_1(t), y_1(t))$  is the solution for input  $u_1(t)$ , and  $(x_2(t), y_2(t))$  is the solution for input  $u_2(t)$ , then for any scalars  $\alpha, \beta$ :

$$u(t) = \alpha u_1(t) + \beta u_2(t)$$

produces

$$x(t) = \alpha x_1(t) + \beta x_2(t), \quad y(t) = \alpha y_1(t) + \beta y_2(t).$$