

# ECE240 Introduction to Linear Dynamical Systems

## Lecture 10: Laplace Transform

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# Outline

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- 2 Laplace Transform of Matrix-Valued Function
- 3 Resolvent and State Transition Matrix
- 4 Qualitative Behavior of Trajectory  $\mathbf{x}(t)$
- 5 Transfer Function and Impulse Response

## Laplace Transform: Definition

For a signal  $f(t)$  defined for  $t \geq 0$ , the Laplace transform is

$$F(s) = \mathcal{L}\{f(t)\} = \int_0^{\infty} e^{-st} f(t) dt, \quad s \in \mathbb{C}.$$

- Converts a time-domain function to a complex-frequency representation.
- Converges for  $\text{Re}(s)$  in the *region of convergence* (ROC).

# Common Laplace Transform Properties

- Linearity:

$$\mathcal{L}\{af(t) + bg(t)\} = aF(s) + bG(s).$$

- Time scaling:

$$\mathcal{L}\{f(at)\} = \frac{1}{a} F\left(\frac{s}{a}\right), \quad a > 0.$$

- Exponential shift:

$$\mathcal{L}\{e^{ct}f(t)\} = F(s - c).$$

- Derivative:

$$\mathcal{L}\{f'(t)\} = sF(s) - f(0).$$

# Useful Laplace Transform Table (I)

$f(t)$	$F(s) = \mathcal{L}\{f(t)\}$
1	$\frac{1}{s}, \quad \text{Re}(s) > 0$
$t^n, \quad n = 0, 1, 2, \dots$	$\frac{n!}{s^{n+1}}, \quad \text{Re}(s) > 0$
$e^{at}$	$\frac{1}{s - a}, \quad \text{Re}(s) > \text{Re}(a)$
$t^n e^{at}$	$\frac{n!}{(s - a)^{n+1}}, \quad \text{Re}(s) > \text{Re}(a)$
$\sin(\omega t)$	$\frac{\omega}{s^2 + \omega^2}, \quad \text{Re}(s) > 0$
$\cos(\omega t)$	$\frac{s}{s^2 + \omega^2}, \quad \text{Re}(s) > 0$
$e^{at} \sin(\omega t)$	$\frac{\omega}{(s - a)^2 + \omega^2}, \quad \text{Re}(s) > \text{Re}(a)$
$e^{at} \cos(\omega t)$	$\frac{s - a}{(s - a)^2 + \omega^2}, \quad \text{Re}(s) > \text{Re}(a)$

## Useful Laplace Transform Table (II)

$f(t)$	$F(s) = \mathcal{L}\{f(t)\}$
$u(t)$ (unit step)	$\frac{1}{s}, \quad \text{Re}(s) > 0$
$u(t - T)$ (delayed step)	$\frac{e^{-Ts}}{s}, \quad \text{Re}(s) > 0$
$t u(t)$	$\frac{1}{s^2}, \quad \text{Re}(s) > 0$
$(t - T)u(t - T)$	$\frac{e^{-Ts}}{s^2}, \quad \text{Re}(s) > 0$
$\delta(t)$ (impulse)	1
$\delta(t - T)$	$e^{-Ts}$
$f(t - T)u(t - T)$	$e^{-Ts}F(s) \quad (\text{time shift})$

## Inverse Laplace Transform

The *inverse Laplace transform* of a function  $F(s)$  is defined by the **Bromwich integral** (a.k.a. inverse Laplace integral):

$$f(t) = \mathcal{L}^{-1}\{F(s)\} = \frac{1}{2\pi j} \int_{\gamma-j\infty}^{\gamma+j\infty} F(s) e^{st} ds,$$

where:

- $\gamma$  is a real constant chosen so that the vertical line  $\text{Re}(s) = \gamma$  lies to the *right* of all singularities (poles) of  $F(s)$ ;
- the integration path is a vertical line in the complex plane (parallel to the imaginary axis), called the *Bromwich contour*;
- the integral converges for all  $t \geq 0$  when  $F(s)$  is of exponential order.

In practice, we use: 1) Tables of standard transforms; 2) Partial fraction expansion; and 3) Polynomial / residue methods for rational  $F(s)$ .

# Partial Fraction Expansion

Suppose

$$F(s) = \frac{P(s)}{Q(s)},$$

with  $Q(s)$  factorizable into linear or quadratic terms.

If  $Q(s) = \prod_k (s - \lambda_k)$ , then

$$F(s) = \sum_k \frac{A_k}{s - \lambda_k}.$$

Inverse transform:

$$\mathcal{L}^{-1} \left\{ \frac{A_k}{s - \lambda_k} \right\} = A_k e^{\lambda_k t}.$$

## Example: Simple Inverse Transform

Given

$$F(s) = \frac{3}{s+2} + \frac{1}{s^2+4},$$

apply known transforms:

$$\mathcal{L}^{-1}\left(\frac{3}{s+2}\right) = 3e^{-2t}, \quad \mathcal{L}^{-1}\left(\frac{1}{s^2+4}\right) = \frac{1}{2}\sin(2t).$$

Thus

$$f(t) = 3e^{-2t} + \frac{1}{2}\sin(2t).$$

# Laplace Transform of Matrix-Valued Function

Given a matrix-valued function  $\mathbf{z}(t) : \mathbb{R}_+ \rightarrow \mathbb{R}^{p \times q}$ , the Laplace transform is defined as

$$\mathbf{Z}(s) = \mathcal{L}\{\mathbf{z}(t)\} = \int_0^\infty e^{-st} \mathbf{z}(t) dt$$

with  $\mathbf{Z} : D \subseteq \mathbb{C} \rightarrow \mathbb{C}^{p \times q}$ , where  $D$  is the *domain or region of convergence* of  $\mathbf{Z}$ .

- $D$  includes at least  $\{ s \mid \operatorname{Re}(s) > a \}$ , where  $a$  satisfies  
 $|z_{ij}(t)| \leq \alpha e^{at}, \quad t \geq 0, \quad i = 1, \dots, p, \quad j = 1, \dots, q.$
- **Integral applied entrywise.**

# Derivative Property

$$\mathcal{L}(\dot{\mathbf{z}}) = s \mathbf{Z}(s) - \mathbf{z}(0)$$

Proof (via integrate by parts):

$$\begin{aligned}\mathcal{L}(\dot{\mathbf{z}})(s) &= \int_0^\infty e^{-st} \dot{\mathbf{z}}(t) dt \\ &= e^{-st} \mathbf{z}(t) \Big|_{t=0}^{t \rightarrow \infty} + s \int_0^\infty e^{-st} \mathbf{z}(t) dt \\ &= s \mathbf{Z}(s) - \mathbf{z}(0).\end{aligned}$$

## Resolvent and State Transition Matrix (1/2)

Take Laplace transform of both sides of the system  $\dot{\mathbf{x}} = \mathbf{Ax}$ :

$$s\mathbf{X}(s) - \mathbf{x}(0) = \mathbf{AX}(s) \implies \mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1}\mathbf{x}(0).$$

Inverse transform:

$$\mathbf{x}(t) = \mathcal{L}^{-1}[(s\mathbf{I} - \mathbf{A})^{-1}] \mathbf{x}(0).$$

- $(s\mathbf{I} - \mathbf{A})^{-1}$  is called the *resolvent* of  $\mathbf{A}$ .
- Poles occur at eigenvalues of  $\mathbf{A}$ .
- State-transition matrix:

$$\Phi(t) = \mathcal{L}^{-1}[(s\mathbf{I} - \mathbf{A})^{-1}].$$

- Solution:

$$\mathbf{x}(t) = \Phi(t)\mathbf{x}(0).$$

## Resolvent and State Transition Matrix (2/2)

- Series expansion:

$$(s\mathbf{I} - \mathbf{A})^{-1} = \frac{1}{s} \left( \mathbf{I} - \frac{\mathbf{A}}{s} \right)^{-1} = \frac{1}{s} \left( \mathbf{I} + \frac{\mathbf{A}}{s} + \frac{\mathbf{A}^2}{s^2} + \dots \right).$$

- Inverse Laplace transform gives

$$\Phi(t) = \mathbf{I} + t\mathbf{A} + \frac{(t\mathbf{A})^2}{2!} + \frac{(t\mathbf{A})^3}{3!} + \dots = e^{t\mathbf{A}}.$$

- Thus, the resolvent  $(s\mathbf{I} - \mathbf{A})^{-1}$  and the state-transition matrix  $\Phi(t) = e^{t\mathbf{A}}$  form a Laplace transform pair:

$$e^{t\mathbf{A}} = \mathcal{L}^{-1}[(s\mathbf{I} - \mathbf{A})^{-1}]$$

## Example 1: Computing $e^{t\mathbf{A}}$ for a Skew-symmetric Matrix

For

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}, \quad s\mathbf{I} - \mathbf{A} = \begin{bmatrix} s & -1 \\ 1 & s \end{bmatrix},$$

the resolvent is

$$(s\mathbf{I} - \mathbf{A})^{-1} = \begin{bmatrix} \frac{s}{s^2 + 1} & \frac{1}{s^2 + 1} \\ \frac{1}{s^2 + 1} & \frac{s}{s^2 + 1} \end{bmatrix}.$$

The state transition matrix is

$$e^{t\mathbf{A}} = \mathcal{L}^{-1}[(s\mathbf{I} - \mathbf{A})^{-1}] = \begin{bmatrix} \cos t & \sin t \\ -\sin t & \cos t \end{bmatrix},$$

a rotation matrix by angle  $-t$ .

## Example 2: Computing $e^{\mathbf{A}}$ for a Nilpotent Matrix

Let

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}.$$

We have

$$e^{t\mathbf{A}} = \mathcal{L}^{-1}((s\mathbf{I} - \mathbf{A})^{-1}) = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} \implies e^{\mathbf{A}} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \text{ (setting } t = 1\text{)}$$

Check using the power series:

$$e^{\mathbf{A}} = \mathbf{I} + \mathbf{A} + \frac{\mathbf{A}^2}{2!} + \cdots = \mathbf{I} + \mathbf{A},$$

since  $\mathbf{A}^k = \mathbf{0}$ ,  $\forall k \geq 2$ .

# Matrix Exponential Solution

The exact solution of  $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}$  is  $\mathbf{x}(t) = e^{t\mathbf{A}}\mathbf{x}(0)$ .

- Generalization of scalar  $x(t) = e^{at}x(0)$ .
- $e^{t\mathbf{A}}$  is always nonsingular, with  $(e^{t\mathbf{A}})^{-1} = e^{-t\mathbf{A}}$ .
- $e^{\mathbf{A}+\mathbf{B}} = e^{\mathbf{A}} e^{\mathbf{B}}$  if  $\mathbf{AB} = \mathbf{BA}$ , i.e.,  $\mathbf{A}$  and  $\mathbf{B}$  commute.

## Time Transfer Property

- For any  $\tau$  and  $t$ , the exact update is

$$\mathbf{x}(\tau + t) = e^{t\mathbf{A}} \mathbf{x}(\tau) = \left( \mathbf{I} + t\mathbf{A} + \frac{(t\mathbf{A})^2}{2!} + \frac{(t\mathbf{A})^3}{3!} + \dots \right) \mathbf{x}(\tau).$$

*Interpretation:*  $e^{t\mathbf{A}}$  propagates state  $t$  seconds forward in time (backward if  $t < 0$ ).

- Forward Euler approximation is a first-order numerical procedure for solving ODEs.

For small  $t$ :

$$\mathbf{x}(\tau + t) \approx (\mathbf{I} + t\mathbf{A})\mathbf{x}(\tau).$$

## Eigenvalues of $\mathbf{A}$ and Poles of the Resolvent

The  $(i,j)$  entry of the resolvent  $(s\mathbf{I} - \mathbf{A})^{-1}$  can be written via Cramer's rule as

$$\frac{(-1)^{i+j} \det \Delta_{ij}}{\det(s\mathbf{I} - \mathbf{A})},$$

where  $\Delta_{ij}$  is  $s\mathbf{I} - \mathbf{A}$  with the  $j$ th row and  $i$ th column removed.

- $\det \Delta_{ij}$  is a polynomial of degree less than  $n$ . Therefore each entry of the resolvent has the form

$$\frac{f_{ij}(s)}{\mathcal{X}(s)},$$

where  $f_{ij}(s)$  has degree  $< n$  and  $\mathcal{X} = \det(s\mathbf{I} - \mathbf{A})$  is the characteristic polynomial.

- Poles of the resolvent entries must be eigenvalues of  $\mathbf{A}$ .
- But not every eigenvalue appears in every entry (cancellations between  $\det \Delta_{ij}$  and  $\mathcal{X}(s)$  may occur).

## Example: Eigenvalues vs. Poles of Resolvent Entries

Consider

$$\mathbf{A} = \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix}, \quad \mathcal{X}(s) = \det(s\mathbf{I} - \mathbf{A}) = (s - 1)(s - 2).$$

Then

$$s\mathbf{I} - \mathbf{A} = \begin{bmatrix} s - 1 & 0 \\ 0 & s - 2 \end{bmatrix}, \quad (s\mathbf{I} - \mathbf{A})^{-1} = \begin{bmatrix} \frac{1}{s - 1} & 0 \\ 0 & \frac{1}{s - 2} \end{bmatrix}.$$

Using Cramer's rule for the (1, 1) entry:

$$[(s\mathbf{I} - \mathbf{A})^{-1}]_{11} = \frac{\det \Delta_{11}}{\det(s\mathbf{I} - \mathbf{A})} = \frac{s - 2}{(s - 1)(s - 2)} = \frac{1}{s - 1}.$$

- Eigenvalues of  $\mathbf{A}$  are 1 and 2.
- The (1, 1) entry has only a pole at  $s = 1$ ; the factor  $(s - 2)$  cancels out.
- Similarly, the (2, 2) entry has only a pole at  $s = 2$ .

## Qualitative Behavior of $\mathbf{x}(t)$

Suppose  $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}$ , with  $\mathbf{x}(t) \in \mathbb{R}^n$ .

$$\mathbf{x}(t) = e^{t\mathbf{A}}\mathbf{x}(0), \quad \mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1}\mathbf{x}(0).$$

The  $i$ th component has form

$$X_i(s) = \frac{a_i(s)}{\mathcal{X}(s)},$$

where  $a_i(s)$  is a polynomial with degree  $< n$  and  $\mathcal{X}(s) = \det(s\mathbf{I} - \mathbf{A})$ .

Thus, all poles of  $X_i(s)$  are eigenvalues of  $\mathbf{A}$  (but not necessarily the other way around, as shown before).

## Case 1: Distinct Eigenvalues

Assume eigenvalues  $\lambda_j$  are distinct, so  $X_i(s)$  has no repeated poles.

Then

$$x_i(t) = \sum_{j=1}^n \beta_{ij} e^{\lambda_j t},$$

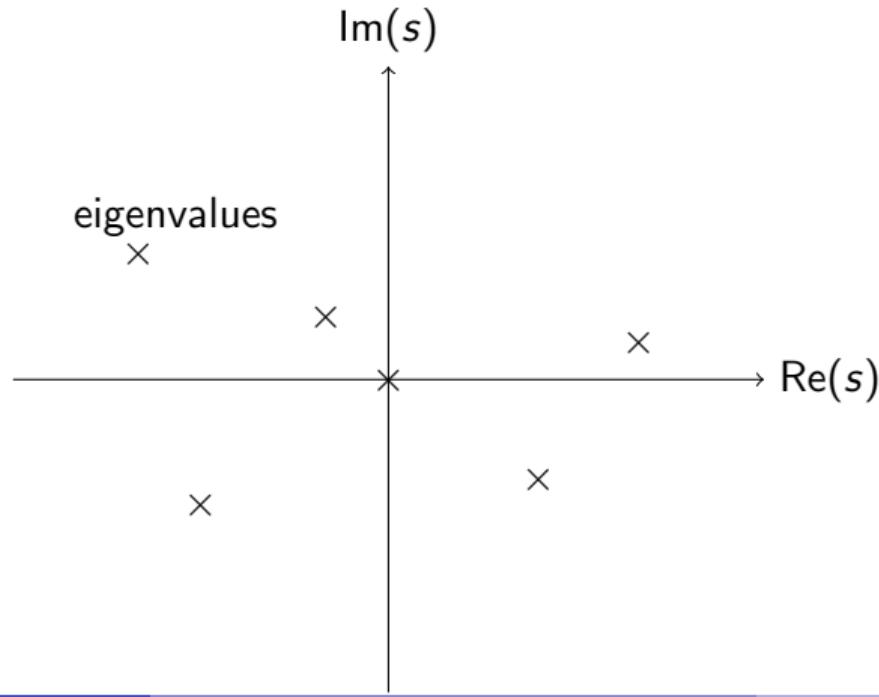
where coefficients  $\beta_{ij}$  depend linearly on  $\mathbf{x}(0)$ .

Eigenvalues determine the possible qualitative behavior of  $\mathbf{x}(t)$ :

- Real  $\lambda$  gives exponential decay or growth:  $e^{\lambda t}$ .
- Complex  $\lambda = \sigma + j\omega$  gives decaying/growing sinusoids:  $e^{\sigma t} \cos(\omega t + \phi)$ .

## Growth/Oscillation Rates from Eigenvalues

- $\text{Re}(\lambda_j)$  gives exponential growth rate (if  $> 0$ ) or decay rate (if  $< 0$ ).
- $\text{Im}(\lambda_j)$  gives frequency of oscillation (if  $\neq 0$ ).



## Case 2: Repeated Eigenvalues

If  $\mathbf{A}$  has repeated eigenvalues, then  $X_i(s)$  can have repeated poles.

Let the distinct eigenvalues be  $\lambda_1, \dots, \lambda_r$  with multiplicities  $n_1, \dots, n_r$  (so  $n_1 + \dots + n_r = n$ ).

Then

$$x_i(t) = \sum_{j=1}^r p_{ij}(t) e^{\lambda_j t},$$

where  $p_{ij}(t)$  is a polynomial of degree  $< n_j$  and depends linearly on  $\mathbf{x}(0)$ .

## Example: Repeated Eigenvalues and Polynomial Terms (1/2)

Consider

$$\mathbf{A} = \begin{bmatrix} 2 & 1 \\ 0 & 2 \end{bmatrix},$$

a  $2 \times 2$  Jordan block with repeated eigenvalue  $\lambda = 2$ .

Compute the resolvent:

$$(s\mathbf{I} - \mathbf{A}) = \begin{bmatrix} s-2 & -1 \\ 0 & s-2 \end{bmatrix}, \quad (s\mathbf{I} - \mathbf{A})^{-1} = \begin{bmatrix} \frac{1}{s-2} & \frac{1}{(s-2)^2} \\ 0 & \frac{1}{s-2} \end{bmatrix}.$$

Taking inverse Laplace transform:

$$\Phi(t) = e^{t\mathbf{A}} = e^{2t} \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix}.$$

## Example: Repeated Eigenvalues and Polynomial Terms (2/2)

Thus, for any initial state  $\mathbf{x}(0)$ ,

$$\mathbf{x}(t) = e^{2t} \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} \mathbf{x}(0).$$

**Key observation:** The repeated eigenvalue  $\lambda = 2$  produces a term of the form  $te^{2t}$ , which is the signature of a repeated pole and a nontrivial Jordan block.

## Region of Convergence for $(s\mathbf{I} - \mathbf{A})^{-1}$

For the homogeneous LDS  $\dot{\mathbf{x}} = \mathbf{Ax}$ , we have  $\mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1}\mathbf{x}(0)$ . The characteristic polynomial is  $\mathcal{X}(s) = \det(s\mathbf{I} - \mathbf{A})$ , with roots  $\lambda_1, \dots, \lambda_n$ .

The resolvent  $(s\mathbf{I} - \mathbf{A})^{-1}$  converges for

$$\operatorname{Re}(s) > \max_j \operatorname{Re}(\lambda_j).$$

- If all eigenvalues satisfy  $\operatorname{Re}(\lambda_j) < 0$ , the ROC includes the entire right-half-plane and the system is stable.
- If any eigenvalue satisfies  $\operatorname{Re}(\lambda_j) > 0$ , the ROC shifts right and the system is unstable.

ROC directly reflects growth/decay rates of  $e^{t\mathbf{A}}$ .

# Transfer Function and Impulse Response

For the LDS

$$\dot{\mathbf{x}} = \mathbf{Ax} + \mathbf{Bu}, \quad \mathbf{y} = \mathbf{Cx} + \mathbf{Du},$$

taking Laplace transforms yields

$$\mathbf{Y}(s) = (\mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B} + \mathbf{D}) \mathbf{U}(s).$$

The transfer function is

$$\mathbf{G}(s) := \frac{Y(s)}{U(s)} = \mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B} + \mathbf{D}.$$

Impulse response is the inverse Laplace transform:

$$\mathbf{h}(t) = \mathcal{L}^{-1}\{\mathbf{G}(s)\} = \mathbf{C}e^{t\mathbf{A}}\mathbf{B} + \mathbf{D}\delta(t).$$

**G(s)** and **h(t)** are a Laplace transform pair.

## Forced LDS and Convolution Representation

Consider the input–output LDS

$$\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t).$$

Taking the Laplace transform (zero initial condition shown for clarity):

$$\mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1} \mathbf{B} \mathbf{U}(s).$$

Inverse transform gives

$$\mathbf{x}(t) = \int_0^t e^{(t-\tau)\mathbf{A}} \mathbf{B} \mathbf{u}(\tau) d\tau.$$

Using  $\mathcal{L}(f * g) = F(s)G(s)$ :

$$\mathbf{x}(t) = e^{t\mathbf{A}} * \mathbf{B}\mathbf{u}(t),$$

so the state is a **convolution** of the matrix exponential with the input.