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1

## Derivation of Serdes Channel Simulation (Simplified)

References [1] 02-dec-2014 derivation

Objective: In 02-dec-2014 derivation, ~~at~~ there were no assumptions on excitations as well as impedance. The derivation is applicable to most general cases.

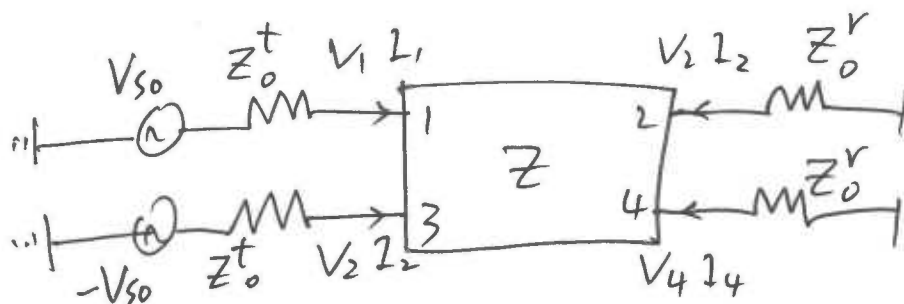
In this derivation, assumptions are made to

Simplify the process:

assumptions:

(1) there is no common-mode excitation

(2) p/n impedances are the same



- ① the excitations are purely differential  
 ② the p/n impedances are equal

System equations:

$$V = [V_1 \ V_2 \ V_3 \ V_4]^T$$

$$I = [I_1 \ I_2 \ I_3 \ I_4]^T$$

$$V = Z I, \quad I = Y V$$

1a

1b

1c, 1d

Terminal eqns:

$$V_2 = -Z_0^r I_2$$

$$V_4 = -Z_0^r I_4$$

$$V_{s0} - V_1 = Z_0^t I_1$$

$$-V_{s0} - V_3 = Z_0^t I_3$$

2a

2b

3a

3b

mixed mode voltages and currents

$$V_{d1} = V_1 - V_3$$

$$V_{c1} = (V_1 + V_3)/2$$

(4a, 4b)

$$V_{d2} = V_2 - V_4$$

$$V_{c2} = (V_2 + V_4)/2$$

(5a, 5b)

$$I_{d1} = (I_1 - I_3)/2, \quad I_{c1} = I_1 + I_3$$

(6a, 6b)

$$I_{d2} = (I_2 - I_4)/2, \quad I_{c2} = I_2 + I_4$$

(7a, 7b)

single-ended to mixed mode mapping

$$V_m = \begin{pmatrix} V_{d1} \\ V_{d2} \\ V_{c1} \\ V_{c2} \end{pmatrix} = \begin{pmatrix} 1 & 0 & -1 & 0 \\ 0 & 1 & 0 & -1 \\ \frac{1}{2} & 0 & \frac{1}{2} & 0 \\ 0 & \frac{1}{2} & 0 & \frac{1}{2} \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \\ V_3 \\ V_4 \end{pmatrix}$$

eqn #  
(mv)

$$I_m = \begin{pmatrix} I_{d1} \\ I_{d2} \\ I_{c1} \\ I_{c2} \end{pmatrix} = \begin{pmatrix} \frac{1}{2} & 0 & -\frac{1}{2} & 0 \\ 0 & \frac{1}{2} & 0 & -\frac{1}{2} \\ 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \\ I_3 \\ I_4 \end{pmatrix}$$

eqn #  
(mI)

$$V_m = M_v V$$

$$I_m = M_I I$$

eqn # (mrb)  
(m Ib)

$$V_m = M_v Z M_I^{-1} I_m$$

eqn # (V<sub>m</sub>)

$$Z_m = M_v Z M_I^{-1}$$

# (Z<sub>ma</sub>)

$$= \begin{bmatrix} Z_{dd} & Z_{dc} \\ Z_{cd} & Z_{cc} \end{bmatrix}$$

# (Z<sub>mb</sub>)

$$\begin{pmatrix} V_{d1} \\ V_{d2} \\ V_{c1} \\ V_{c2} \end{pmatrix} = \begin{bmatrix} Z_{dd} & Z_{dc} \\ Z_{cd} & Z_{cc} \end{bmatrix} \begin{pmatrix} I_{d1} \\ I_{d2} \\ I_{c1} \\ I_{c2} \end{pmatrix}$$

# (V<sub>m</sub> & I<sub>m</sub>)

to solve for  $V_m, I_m$ , from 3a, 3b. 4  
we have

$$V_1 - V_3 = 2V_{s0} - Z_0^t (I_1 - I_3) \quad (10)$$

$$V_1 + V_3 = -Z_0^t (I_1 + I_3) \quad (11)$$

$$\text{define } V_s^d = 2V_{s0} \quad (12a)$$

$$V_s^c = V_{s0} - V_{s0} = 0 \quad (12b)$$

from (4a, 4b) we have

$$V_{d1} = V_s^d - 2Z_0^t I_{d1} \quad (13a)$$

$$V_{c1} = -\frac{Z_0^t}{2} I_{c1} \quad (13b)$$

similarly

$$V_{d2} = -Z_0^r (I_2 - I_4) \quad (14a)$$

$$= -2Z_0^r I_{d2}$$

$$V_{c2} = -Z_0^r \frac{(I_2 + I_4)}{2} = -Z_0^r \frac{I_{c2}}{2} \quad (14b)$$

$$\begin{pmatrix} V_{d1} \\ V_{d2} \\ V_{c1} \\ V_{c2} \end{pmatrix} = \begin{pmatrix} -Z_0^t & 0 & Z_0^t & 0 \\ 0 & -Z_0^r & 0 & Z_0^r \\ -\frac{Z_0^t}{2} & 0 & -\frac{Z_0^t}{2} & 0 \\ 0 & -\frac{Z_0^r}{2} & 0 & -\frac{Z_0^r}{2} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \\ I_3 \\ I_4 \end{pmatrix} + \begin{pmatrix} V_s^d \\ 0 \\ 0 \\ 0 \end{pmatrix}$$

there is  
no common  
mode  
excitation

eqn #15

$$I = YV = Y M_v^{-1} V_m \quad (16)$$

$$V_m = Z^T Y M_v^{-1} V_m + V_s \quad (17)$$

$$V_m = (U - Z^T Y M_v^{-1})^{-1} V_s \quad (18)$$

$$I_m = M_2 Y M_v^{-1} V_m \quad (19)$$

$$V_m(j\omega) = H_c(j\omega) V_s(j\omega)$$

where

$$H_c(j\omega) = (U - Z^T Y M_v^{-1})^{-1}$$

(6)

alternatively, from 12, 13 14

$$\begin{pmatrix} V_{d1} \\ V_{d2} \\ V_{c1} \\ V_{c2} \end{pmatrix} = \overbrace{\begin{pmatrix} -2Z_0^t & 0 & 0 & 0 \\ 0 & -2Z_0^r & 0 & 0 \\ 0 & 0 & -\frac{Z_0^t}{2} & 0 \\ 0 & 0 & 0 & -\frac{Z_0^r}{2} \end{pmatrix}}^{Z^{tm}} \begin{pmatrix} I_{d1} \\ I_{d2} \\ I_{c1} \\ I_{c2} \end{pmatrix} + \begin{pmatrix} V_s^d \\ 0 \\ 0 \\ 0 \end{pmatrix}$$

$$V_m = Z^{tm} I_m + V_s$$

$$= Z^{tm} M_1 Y M_V^{-1} V_m + V_s$$

$$V_m = (I - Z^{tm} M_1 Y M_V^{-1})^{-1} V_s$$

this is equivalent to (18) provided that

$$Z^{tm} M_1 = Z^T$$

at this point, it is obvious that by assuming

- ① the lack of common mode excitation
- ② ~~the~~ p/n impedances are equal

have not resulted in any real savings in the solution process, the main computation cost of this process is the inversion of eqn (18) of the  $4 \times 4$  matrix. at all frequency points.

If one really needs to <sup>further</sup> reduce the cost of computation, ~~one must~~ more aggressive assumptions must be made:

in eqn # (18), let  
on page 3

$$Z_{dc} = Z_{cd} = Z_{cc} = 0 \quad \text{we have}$$

$$\begin{pmatrix} V_{d1} \\ V_{d2} \end{pmatrix} = Z_{dd} \begin{pmatrix} I_{d1} \\ I_{d2} \end{pmatrix} \quad \dots \quad (101)$$

(8)

$$\begin{pmatrix} V_{d1} \\ V_{d2} \end{pmatrix} = \begin{pmatrix} -2\tilde{z}_0^t & 0 \\ 0 & -2\tilde{z}_0^r \end{pmatrix} \begin{pmatrix} I_{d1} \\ I_{d2} \end{pmatrix} + \begin{pmatrix} V_s^d \\ 0 \end{pmatrix}$$

$$\begin{pmatrix} V_{d1} \\ V_{d2} \end{pmatrix} = \begin{bmatrix} \tilde{z}^t \\ \tilde{z} \end{bmatrix} \tilde{z}_{dd}^{-1} \begin{pmatrix} V_{d1} \\ V_{d2} \end{pmatrix} + \begin{pmatrix} V_s^d \\ 0 \end{pmatrix}$$

$$\begin{pmatrix} V_{d1} \\ V_{d2} \end{pmatrix} = \left( I - \begin{bmatrix} \tilde{z}^t \\ \tilde{z} \end{bmatrix} \tilde{z}_{dd}^{-1} \right)^{-1} \begin{pmatrix} V_s^d \\ 0 \end{pmatrix}$$

now, this inversion is  $2 \times 2$  matrix  
which can be solved analytically.

I doubt the time saving from  $4 \times 4$  inversion  
to  $2 \times 2$  inversion is significant.