

General Description

The MAX4450 single and MAX4451 dual op amps are unity-gain-stable devices that combine high-speed performance with rail-to-rail outputs. Both devices operate from a +4.5V to +11V single supply or from ±2.25V to ±5.5V dual supplies. The common-mode input voltage range extends beyond the negative power-supply rail (ground in single-supply applications).

The MAX4450/MAX4451 require only 6.5mA of quiescent supply current per op amp while achieving a 210MHz -3dB bandwidth and a 485V/µs slew rate. Both devices are an excellent solution in low-power/lowvoltage systems that require wide bandwidth, such as video, communications, and instrumentation.

The MAX4450 is available in the ultra-small 5-pin SC70 package, while the MAX4451 is available in spacesaving 8-pin SOT23 and SO packages.

Features

- ♦ Ultra-Small SC705 and SOT23 Packages
- **♦ Low Cost**
- ♦ High Speed 210MHz -3dB Bandwidth 55MHz 0.1dB Gain Flatness 485V/µs Slew Rate
- ♦ Single +4.5V to +11V Operation
- ♦ Rail-to-Rail Outputs
- ♦ Input Common-Mode Range Extends Beyond VEE
- ♦ Low Differential Gain/Phase: 0.02%/0.08°
- ♦ Low Distortion at 5MHz
 - -65dBc SFDR
 - -63dB Total Harmonic Distortion

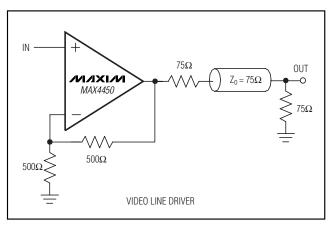
Applications

Set-Top Boxes Surveillance Video Systems **Battery-Powered Instruments** Video Line Driver Analog-to-Digital Converter Interface CCD Imaging Systems Video Routing and Switching Systems Digital Cameras

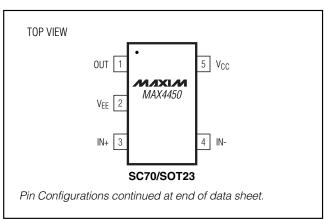
Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4450EXK-T	-40°C to +85°C	5 SC70	AAA
MAX4450EUK-T	-40°C to +85°C	5 SOT23	ADKP
MAX4451EKA-T	-40°C to +85°C	8 SOT23	AAAA
MAX4451ESA	-40°C to +85°C	8 SO	_

Typical Operating Circuit



Pin Configurations



NIXIN

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V _{CC} to V _{EE})+12V
IN, IN_+, OUT(VEE - 0.3V) to (VCC + 0.3V)
Output Short-Circuit Current to VCC or VEE150mA
Continuous Power Dissipation ($T_A = +70^{\circ}C$)
5-Pin SC70-5 (derate 2.5mW/°C above +70°C)200mW
5-Pin SOT23-5 (derate 7.1mW/°C above +70°C)571mW

8-Pin SOT23-8 (derate 5.26mW/°C abov	e +70°C)421mW
8-Pin SO (derate 5.9mW/°C above +70°	,
Operating Temperature Range	
Storage Temperature Range	
Lead Temperature (soldering, 10s)	
3, 11,	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or at any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = 0V, R_L = \infty \text{ to } V_{CC}/2, V_{OUT} = V_{CC}/2, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ (Note 1)

PARAMETER	SYMBOL	C	MIN	TYP	MAX	UNITS			
Input Common-Mode Voltage Range	V _{CM}	Guaranteed by CMRR test		V _{EE} - 0.20		V _{CC} 2.25	V		
Input Offset Voltage (Note 2)	Vos				4	26	mV		
Input Offset Voltage Matching					1.0		mV		
Input Offset Voltage Temperature Coefficient	TCvos				8		μV/°C		
Input Bias Current	IB	(Note 2)			6.5	20	μΑ		
Input Offset Current	los	(Note 2)			0.5	4	μΑ		
Input Desistance	Duv	Differential mode ($(-1V \le V_{IN} \le +1V)$		70		kΩ		
Input Resistance	RIN	Common mode (-0	$0.2V \le V_{CM} \le +2.75V$		3		МΩ		
Common-Mode Rejection Ratio	CMRR	(VEE - 0.2V) ≤ VCM	ı≤(Vcc - 2.25V)	70	95		dB		
		0.25V ≤ V _{OUT} ≤ 4.	75V, $R_L = 2k\Omega$	50	60				
Open-Loop Gain (Note 2)	Avol	$0.5V \le V_{OUT} \le 4.5V$, $R_L = 150\Omega$		48	58		dB		
		$1V \le V_{OUT} \le 4V$, $R_L = 50\Omega$			57				
		$R_L = 2k\Omega$	Vcc - Voh		0.05	0.20			
		nL = 2K22	V _{OL} - V _{EE}		0.05	0.15			
0				$R_L = 150\Omega$	Vcc - Voh		0.30	0.50	1
Output Voltage Swing (Note 2)	Vout	N[= 13022	Vol - VEE		0.25	0.80	\rightarrow V		
(14010-2)			Vcc - Voh		0.5	0.80			
		$R_L = 75\Omega$	V _{OL} - V _{EE}		0.5	1.75			
Outrout Coursest	1	D. 500	Sourcing	45	70		A		
Output Current	lout	$R_L = 50\Omega$	Sinking	25	50		mA		
Output Short-Circuit Current	Isc	Sinking or sourcin	g		±120		mA		
Open-Loop Output Resistance	Rout				8		Ω		
Power-Supply Rejection Ratio	PSRR	\/ [\/	V _{EE} = 0V, V _{CM} = 2V	46	62		٩D		
(Note 3)	PSRR	V _{CC} = 5V	VEE = -5V, VCM = 0V	54	69		dB		
Operating Supply-Voltage Range	Vs	VCC to VEE		4.5		11.0	V		
Quiescent Supply Current (per amplifier)	Is				6.5	9.0	mA		

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AC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = 0V, V_{CM} = +2.5V, R_F = 24\Omega, R_L = 100\Omega$ to $V_{CC}/2, V_{OUT} = V_{CC}/2, A_{VCL} = +1V/V, T_A = +25^{\circ}C$, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Small-Signal -3dB Bandwidth	BWss	$V_{OUT} = 100 \text{mV}_{P-P}$			210		MHz
Large-Signal -3dB Bandwidth	BWLS	V _{OUT} = 2V _{P-P}	V _{OUT} = 2V _{P-P}		175		MHz
Bandwidth for 0.1dB Gain Flatness	BW _{0.1dB}	V _{OUT} = 100mV _{P-P}	V _{OUT} = 100mV _{P-P}		55		MHz
Slew Rate	SR	V _{OUT} = 2V step			485		V/µs
Settling Time to 0.1%	ts	Vout = 2V step			16		ns
Rise/Fall Time	t _R , t _F	$V_{OUT} = 100 \text{mV}_{P-P}$			4		ns
Spurious-Free Dynamic Range	SFDR	fc = 5MHz, Vout = 2\	/P-P		-65		dBc
Harmonic Distortion			2nd harmonic		-65		
	HD	$f_C = 5MHz$,	3rd harmonic		-58		dBc
	טוו	Vout = 2V _{P-P}	Total harmonic distortion		-63		
Two-Tone, Third-Order Intermodulation Distortion	IP3	f1 = 4.7MHz, f2 = 4.8N	f1 = 4.7MHz, f2 = 4.8MHz, V _{OUT} = 1V _{P-P}		66		dBc
Channel-to-Channel Isolation	CH _{ISO}	Specified at DC			102		dB
Input 1dB Compression Point		f _C = 10MHz, A _{VCL} = +	-2V/V		14		dBm
Differential Phase Error	DP	NTSC, $R_L = 150\Omega$	NTSC, $R_L = 150\Omega$		0.08		degrees
Differential Gain Error	DG	NTSC, $R_L = 150\Omega$			0.02		%
Input Noise-Voltage Density	en	f = 10kHz			10		nV/√Hz
Input Noise-Current Density	in	f = 10kHz			1.8		pA/√Hz
Input Capacitance	CIN				1		pF
Output Impedance	Zout	f = 10MHz			1.5		Ω

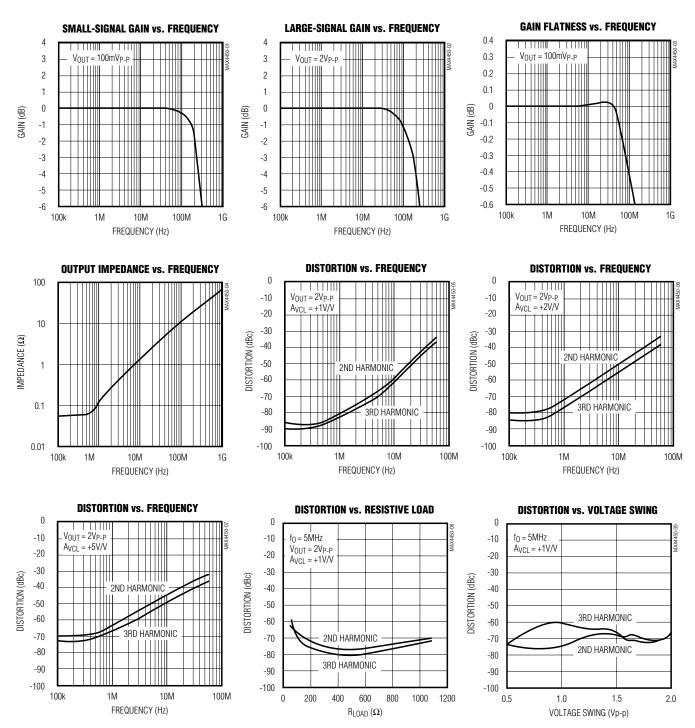
Note 1: All devices are 100% production tested at $T_A = +25$ °C. Specifications over temperature limits are guaranteed by design.

Note 2: Tested with $V_{CM} = +2.5V$.

Note 3: PSR for single +5V supply tested with $V_{EE} = 0V$, $V_{CC} = +4.5V$ to +5.5V; PSR for dual $\pm 5V$ supply tested with $V_{EE} = -4.5V$ to -5.5V, $V_{CC} = +4.5V$ to +5.5V.

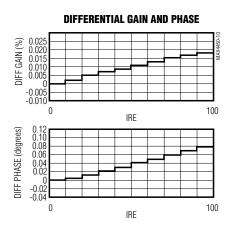
Typical Operating Characteristics

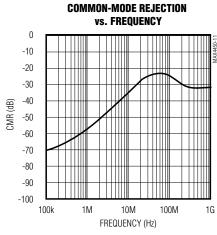
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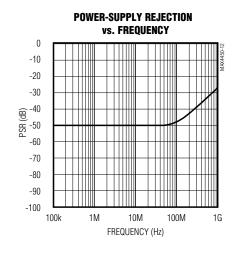


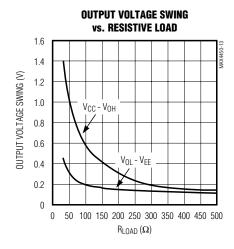
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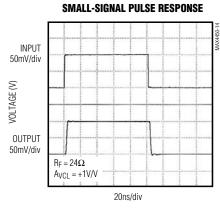
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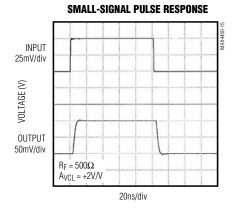


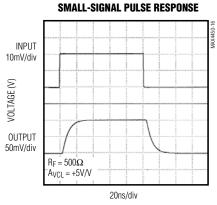


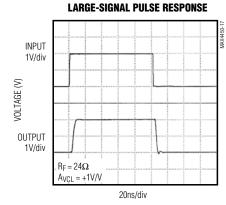


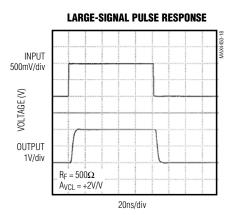






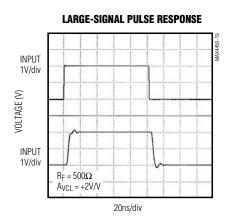


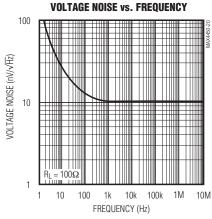


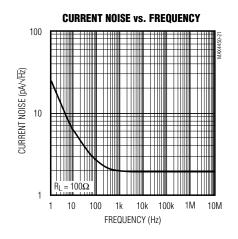


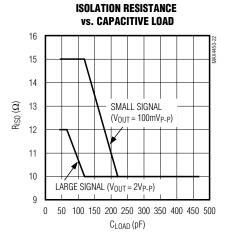
Typical Operating Characteristics (continued)

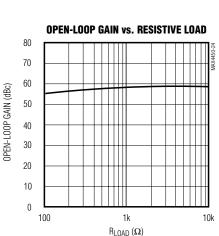
 $(V_{CC} = +5V, V_{EE} = 0, V_{CM} = +2.5V, A_{VCL} = +1V/V, R_F = 24\Omega, R_L = 100\Omega$ to $V_{CC}/2$, $T_A = +25^{\circ}C$, unless otherwise noted.)

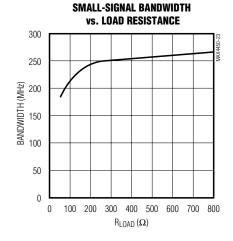


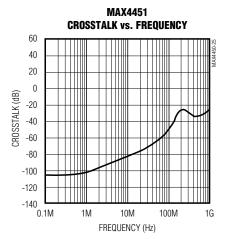












Pin Description

P	PIN	NAME	FUNCTION
MAX4450	MAX4451	NAME	FUNCTION
1	_	OUT	Amplifier Output
2	4	V _{EE}	Negative Power Supply or Ground (in single- supply operation)
3	_	IN+	Noninverting Input
4	_	IN-	Inverting Input
5	8	Vcc	Positive Power Supply
_	1	OUTA	Amplifier A Output
_	2	INA-	Amplifier A Inverting Input
_	3	INA+	Amplifier A Noninverting Input
_	7	OUTB	Amplifier B Output
_	6	INB-	Amplifier B Inverting Input
_	5	INB+	Amplifier B Noninverting Input

Detailed Description

The MAX4450/MAX4451 are single-supply, rail-to-rail, voltage-feedback amplifiers that employ current-feedback techniques to achieve 485V/µs slew rates and 210MHz bandwidths. Excellent harmonic distortion and differential gain/phase performance make these amplifiers an ideal choice for a wide variety of video and RF signal-processing applications.

The output voltage swings to within 55mV of each supply rail. Local feedback around the output stage ensures low open-loop output impedance to reduce gain sensitivity to load variations. The input stage permits common-mode voltages beyond the negative supply and to within 2.25V of the positive supply rail.

Applications Information

Choosing Resistor Values

Unity-Gain Configuration

The MAX4450/MAX4451 are internally compensated for unity gain. When configured for unity gain, the devices require a 24Ω resistor (RF) in series with the feedback path. This resistor improves AC response by reducing the Q of the parallel LC circuit formed by the parasitic feedback capacitance and inductance.

Inverting and Noninverting Configurations

Select the gain-setting feedback (RF) and input (RG) resistor values to fit your application. Large resistor values increase voltage noise and interact with the amplifier's input and PC board capacitance. This can generate undesirable poles and zeros and decrease bandwidth or cause oscillations. For example, a noninverting gain-of-two configuration (RF = RG) using $1k\Omega$ resistors, combined with 1pF of amplifier input capacitance and 1pF of PC board capacitance, causes a pole at 159MHz. Since this pole is within the amplifier bandwidth, it jeopardizes stability. Reducing the $1k\Omega$ resistors to 100Ω extends the pole frequency to 1.59GHz, but could limit output swing by adding 200Ω in parallel with the amplifier's load resistor. Table 1 lists suggested feedback and gain resistors, and bandwidths for several gain values in the configurations shown in Figures 1a and 1b.

Layout and Power-Supply Bypassing

These amplifiers operate from a single ± 4.5 V to ± 11 V power supply or from dual ± 2.25 V to ± 5.5 V supplies. For single-supply operation, bypass VCC to ground with a

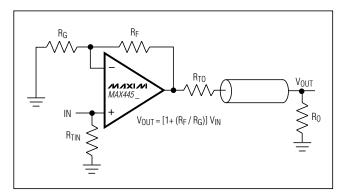


Figure 1a. Noninverting Gain Configuration

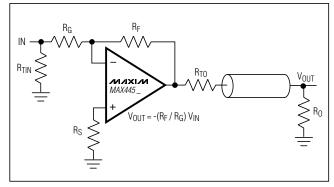


Figure 1b. Inverting Gain Configuration

Table 1. Recommended Component Values

COMPONENT		GAIN (V/V)								
	+1	-1	+2	-2	+5	-5	+10	-10	+25	-25
$R_F\left(\Omega ight)$	24	500	500	500	500	500	500	500	500	1200
R _G (Ω)	∞	500	500	250	124	100	56	50	20	50
$R_S\left(\Omega ight)$	_	0	_	0	_	0	_	0	_	0
R _{TIN} (Ω)	49.9	56	49.9	62	49.9	100	49.9	∞	49.9	∞
$R_TO\left(\Omega ight)$	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9
Small-Signal -3dB Bandwidth (MHz)	210	100	95	50	25	25	11	15	5	10

 $R_{TIN} = \frac{75}{1 - \frac{75}{R_G}} \Omega$

 $0.1\mu F$ capacitor as close to the pin as possible. If operating with dual supplies, bypass each supply with a $0.1\mu F$ capacitor.

Maxim recommends using microstrip and stripline techniques to obtain full bandwidth. To ensure that the PC board does not degrade the amplifier's performance, design it for a frequency greater than 1GHz. Pay careful attention to inputs and outputs to avoid large parasitic capacitance. Whether or not you use a constant-impedance board, observe the following design guidelines:

- Don't use wire-wrap boards; they are too inductive.
- Don't use IC sockets; they increase parasitic capacitance and inductance.
- Use surface-mount instead of through-hole components for better high-frequency performance.
- Use a PC board with at least two layers; it should be as free from voids as possible.
- Keep signal lines as short and as straight as possible. Do not make 90° turns; round all corners.

Rail-to-Rail Outputs, Ground-Sensing Input

The input common-mode range extends from (VEE - 200mV) to (VCC - 2.25V) with excellent common-mode rejection. Beyond this range, the amplifier output is a nonlinear function of the input, but does not undergo phase reversal or latchup.

The output swings to within 55mV of either power-supply rail with a $2k\Omega$ load. The input ground sensing

and the rail-to-rail output substantially increase the dynamic range. With a symmetric input in a single +5V application, the input can swing 2.95V_{P-P} and the output can swing 4.9V_{P-P} with minimal distortion.

Output Capacitive Loading and Stability

The MAX4450/MAX4451 are optimized for AC performance. They are not designed to drive highly reactive loads, which decrease phase margin and may produce excessive ringing and oscillation. Figure 2 shows a circuit that eliminates this problem. Figure 3 is a graph of the optimal isolation resistor (Rs) vs. capacitive load. Figure 4 shows how a capacitive load causes excessive peaking of the amplifier's frequency response if the capacitor is not isolated from the amplifier by a resistor. A small isolation resistor (usually 20Ω to 30Ω) placed before the reactive load prevents ringing and oscillation. At higher capacitive loads, AC performance is controlled by the interaction of the load capacitance and the isolation resistor. Figure 5 shows the effect of a 27Ω isolation resistor on closed-loop response.

Coaxial cable and other transmission lines are easily driven when properly terminated at both ends with their characteristic impedance. Driving back-terminated transmission lines essentially eliminates the line's capacitance.

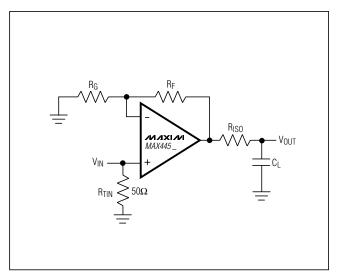


Figure 2. Driving a Capacitive Load Through an Isolation Resistor

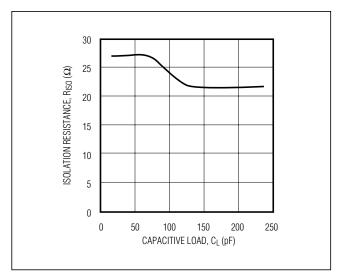


Figure 3. Capacitive Load vs. Isolation Resistance

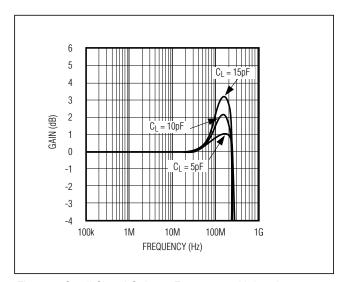


Figure 4. Small-Signal Gain vs. Frequency with Load Capacitance and No Isolation Resistor

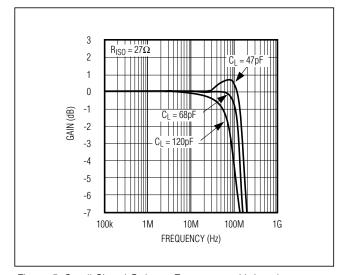
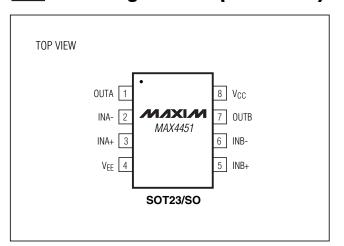


Figure 5. Small-Signal Gain vs. Frequency with Load Capacitance and 27Ω Isolation Resistor

_Pin Configurations (continued)



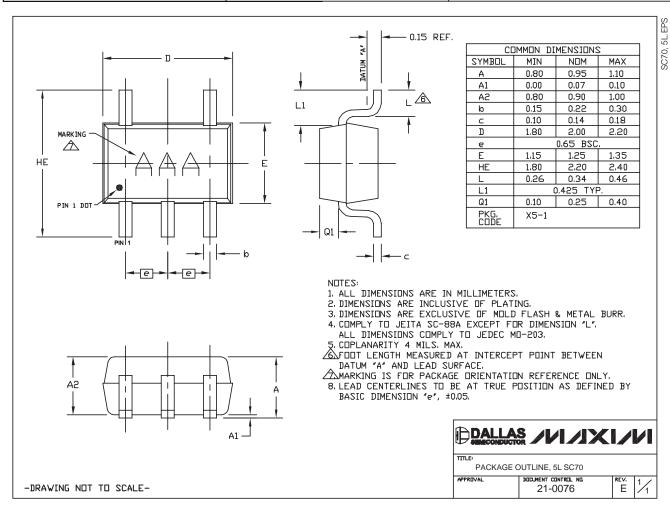
__Chip Information

MAX4450 TRANSISTOR COUNT: 86 MAX4451 TRANSISTOR COUNT: 170

Package Information

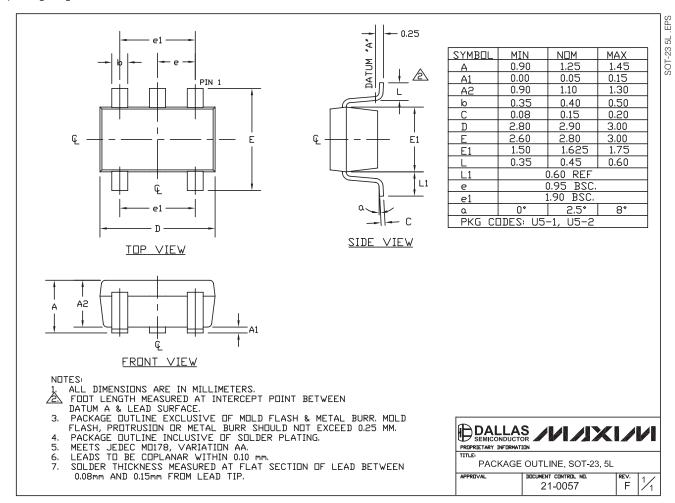
For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
5 SC70	X5-1	<u>21-0076</u>
5 SOT23	U5-2	<u>21-0057</u>
8 SOT23	K8-2	<u>21-0078</u>
8 SO	S8-5	<u>21-0041</u>



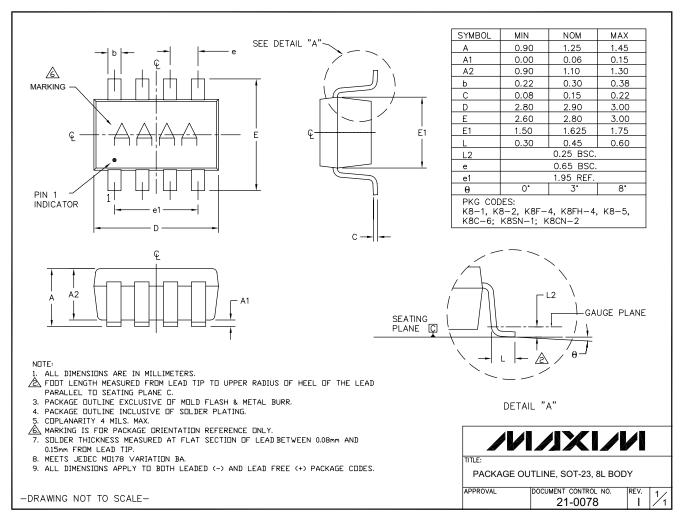
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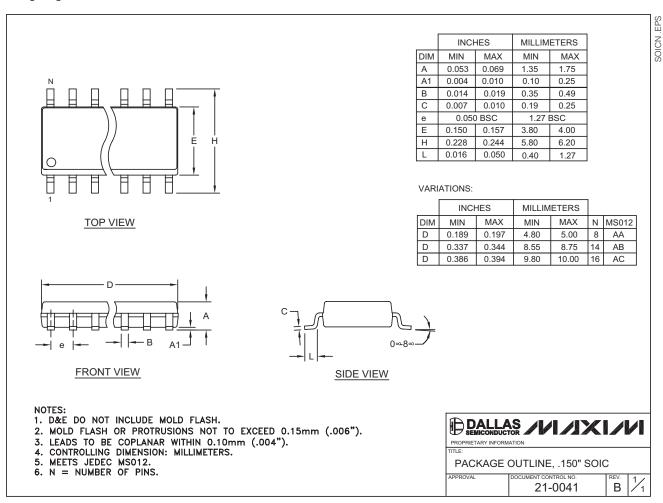
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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
4	11/09	Corrected TOC 20	6

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