

Stability Analysis of Linear Neutral Delay Systems With Two Delays via Augmented Lyapunov–Krasovskii Functionals

Yunxia Song^{ID}, Zhao-Yan Li^{ID}, and Bin Zhou^{ID}, *Senior Member, IEEE*

Abstract—Stability analysis is considered in this paper for linear neutral delay systems subject to two different delays in both the state variables and the retarded derivatives of state variables. By choosing a suitable state vector indexed by an integer k , a new augmented Lyapunov–Krasovskii functional (LKF) is constructed, and a stability criterion based on linear matrix inequalities is developed accordingly. It is shown that the proposed condition is less conservative than the existing methods due to the introduction of the delay-product-type integral terms in the LKF. The resulting stability criterion is then applied to the robust stability analysis of neutral delay systems with norm-bounded uncertainty. Moreover, a delay-independent stability criterion is developed based on the proposed LKF, and its frequency-domain interpretation is also given. These developed stability criteria indexed by an integer k exhibit a hierarchical character: the larger the integer k , the less conservatism of the resulting stability criterion. Finally, two numerical examples are carried out to illustrate the effectiveness of the proposed method.

Index Terms—Neutral delay systems, augmented Lyapunov–Krasovskii functionals, hierarchical stability criteria, robust stability.

I. INTRODUCTION

NEUTRAL delay systems, which contain delays both in system states and derivatives of states [13], can be used to model numerous engineering systems, including partial element equivalent circuits [1], [2], [8], controlled constrained manipulators [24], complex dynamical networks [22], etc. The stability is the basic requirement of systems, thus the stability analysis problem of neutral delay systems has important theoretical and practical significance [28], [29]. As we know, the LKF method is an effective tool for the stability analysis

of time-delay systems [7], [11], [16], [17], [18], [25], [31], [32]. By using this method, neutral delay systems with only a single delay have been extensively studied in [12], [14], and [15]. For neutral delay systems with multiple delays, a simple extension of the ideas for a single time delay problem has been adopted in [9]. This extension is simple and thus might lead to certain conservatism. Thus, for neutral delay systems with multiple delays, an important issue is how to find some proper LKFs, by which one can obtain less conservative results. In the current paper, we mainly focus on linear neutral delay systems with two different delays.

Over the past few decades, efforts have been made to derive stability criteria with less conservatism for linear neutral delay systems with two delays. In [27], a stability criterion was established by choosing a proper LKF and adding some appropriate zero terms to the deviation of LKF. In [5], by choosing an LKF whose derivative considers the relationships among delays, a stability criterion which is less conservative than the result in [27] was obtained. Recently, an augmented LKF was constructed in [19]. Since some delayed states and time delays information were introduced in the constructed LKF, the resulting stability criterion greatly improves the results in [27] and [5]. It is noted that the constructed LKF in [19] only involve a small number of delayed states, which leads to a requirement for further improvement of the constructed augmented LKF. Besides, it should be mentioned that augmented LKFs have also been used in the stability analysis of systems with a time-varying delay to provide some less conservative results [6], [20], [21], [23].

In [4], a nonconservative linear matrix inequalities (LMIs) condition was established by using a frequency-domain approach for neutral delay system. It is worthy mention that such an LMI condition can also be derived by an augmented LKF with a state variable $[x^T(t), x^T(t - h_1), \dots, x^T(t - (k - 1)h_1))]^T$, in which k is a positive integer. Such kind of augmented LKFs have been extended to studying the delay-independent stability analysis problem in [3], [26], and [30]. However, to the best of our knowledge, there is little related discussions regarding the delay-dependent stability analysis by using such kind of LKFs.

Manuscript received 17 September 2022; revised 14 October 2022; accepted 17 October 2022. Date of publication 9 November 2022; date of current version 25 January 2023. This work was supported in part by the NSFC for Distinguished Young Scholars under Grant 62125303, in part by the NSFC under Grant 62173111 and Grant 62188101, and in part by the Fundamental Research Funds for the Central Universities under Grant HIT.BRET.2021008. This article was recommended by Associate Editor Y. Tang. (Corresponding author: Zhao-Yan Li.)

Yunxia Song and Bin Zhou are with the Center for Control Theory and Guidance Technology, Harbin Institute of Technology, Harbin 150001, China. Zhao-Yan Li is with the School of Mathematics, Harbin Institute of Technology, Harbin 150001, China (e-mail: lizhaoyan@hit.edu.cn).

Color versions of one or more figures in this article are available at <https://doi.org/10.1109/TCSI.2022.3216576>.

Digital Object Identifier 10.1109/TCSI.2022.3216576

Motivated by the above analysis, we aim to find a new LKF with an augmented state variable associated with an integer k , to solve the stability analysis problem of linear neutral systems with two delays. The main feature of the considered system is that there are two different delays in both the left-hand (the time-derivative term) and the right-hand of the system. Such a feature makes the problem challenging in two aspects. On the one hand, for a chosen augmented state variable indexed by an integer k , all possible quadratic integral functions may contain redundant elements that must be eliminated since, otherwise, these redundant elements will greatly increase the computational burden. On the other hand, the constructed LKF will contain absolute value terms, which are not easy to handle. In this paper, by solving these two challenging problems mentioned above, we will construct a proper LKF and then derive some new stability criteria for the considered system. To summarize, the main contributions are listed as follows.

- By choosing a suitable state vector indexed by a positive integer k , a new augmented LKF is constructed. Compared with the augmented LKF in [19], the proposed one utilizes more information of the system states and time delays, which helps to reduce conservatism. Moreover, the constructed LKF contains all possible yet the minimal number of delay-product-type quadratic integral functionals (QIFs), which contributes to realizing the reduction of both the conservatism and computational complexity of the resulting stability criterion.
- By using the proposed LKF, a hierarchical delay-dependent stability criterion is established. Since more delay-product-type integral terms are involved in the constructed LKF, the obtained stability criterion is less conservative than the existing ones [5], [9], [19], and [27].
- The established stability criterion is used to obtain a robust stability criterion for neutral delay systems with norm-bounded uncertainty.
- Considering a special case of the constructed LKF, a delay-independent stability criterion is derived for neutral delay systems with two delays. Also, a frequency-domain interpretation of this criterion is given.

The remainder of this article is arranged as follows. The problem to be solved is formulated in Section II. In Section III, a novel LKF is constructed and stability criteria are proposed. In Section IV, the robust stability analysis problem is discussed. Numerical simulations are posted in Section V to verify the correctness of the given scheme and conclusions are given in Section VI.

Notation: We denote by \mathbb{C} and \mathbb{R} , \mathbb{N}^+ , \mathbb{S}^n sets of complex and real numbers, positive integers, $n \times n$ symmetric real matrices, respectively. Let $\overline{\mathbb{C}}_+$ denote the closed right half plane of the complex plane, and \mathbb{D} denote the closed unit disc on the complex plane, respectively. Let P^T and P^H represent the transpose and conjugate-transpose of P respectively. For $A \in \mathbb{R}^{n \times n}$, let $\rho(A)$ denote the spectral radius of A . Denote $\mathbb{I}[p_1, p_2] = \{p_1, p_1 + 1, \dots, p_2\}$ for two integers p_1 and p_2 with $p_2 \geq p_1$. The notation $[b]$, $b \in \mathbb{R}$ signifies the minimum

integer greater than b . Define an operator ∇ on the set of functions $x(t)$, $t \in \mathbb{R}$ by $(\nabla_h x)(t) = x(t - h)$ for time delay h . For $u \in \mathbb{C}$, let

$$u^{[k]} \triangleq [1 \ u \ \dots \ u^{k-1}]^T. \quad (1)$$

II. PROBLEM FORMULATION

Consider the neutral system formulated as

$$\dot{x}(t) - \sum_{i=1}^2 B_i \dot{x}(t - h_i) = \sum_{i=0}^2 A_i x(t - h_i), \quad t \geq 0, \quad (2)$$

where $0 = h_0 < h_1 < h_2$ are the constant delays; A_0, A_1, A_2, B_1, B_2 are $n \times n$ constant matrices. Let $x(t) = [x_1, x_2, \dots, x_n]^T \in \mathbb{R}^n$ be the unique solution to system (2) under the initial value $x(\theta) = \phi(\theta)$, $\theta \in [-h_2, 0]$. Define $x_t = x(t + \theta)$, $\forall \theta \in [-h_2, 0]$, and the operator $\mathcal{D}x_t = x(t) - B_1 x(t - h_1) - B_2 x(t - h_2)$. Throughout this paper, we impose the following assumption.

Assumption 1 ([10], [13]): $\sup_{\theta_1 \in [0, 2\pi]} \rho(B_1 + B_2 e^{j\theta_1}) < 1$.

We know from Theorem 6.1 (Chapter 9, page 286) in [13] that Assumption 1 guarantees that the operator \mathcal{D} is strongly stable, namely, the delay difference equation

$$x(t) = B_1 x(t - h_1) + B_2 x(t - h_2),$$

is stable independent of delays h_1, h_2 .

In this paper, we aim to find a novel quadratic integral LKF to investigate the stability analysis problem of system (2). In order to greatly reduce the conservatism of some existing results, we expect to construct an LKF which can make fully usage of the information of the system states and time delays. To achieve this expectation, different from the existing methods, we choose an augmented state variable indexed by an integer k for the construction of LKFs. With the augmented state variable, we will construct a proper LKF and then derive a novel stability criterion guaranteeing the stability of system (2). Besides, we will provide a stability criterion to solve the robust stability analysis problem of system (2) with norm-bounded uncertainty.

III. STABILITY CRITERIA

A. Construction of the LKF

The main objective of this subsection is to construct a new augmented LKF for system (2). To begin with, we need to choose a proper state vector $\xi(t)$. As stated in [19], a possible state vector $\xi(t)$ should consist of two parts, one part is point delays (denoted as $\xi_1(t)$) and the other part is distributed delays (denoted as $\xi_2(t)$). The main idea of choosing $\xi_1(t)$ in the present paper is to replace the usual state variable, say $\{x(t + \theta), -h_2 \leq \theta \leq 0\}$, with the augmented state $\{x(t + \theta), -2kh_2 \leq \theta \leq 0\}$ for some positive integer k . Following this idea, we define an augmented state variable

$X_{k,k}(t)$ as

$$X_{k,k}(t) = \left(\nabla_{h_1}^{[k]} \otimes \nabla_{h_2}^{[k]} \right) x(t) = \begin{bmatrix} x(t) \\ x(t-h_2) \\ \vdots \\ x(t-(k-1)h_2) \\ \hline x(t-h_1) \\ x(t-h_1-h_2) \\ \vdots \\ x(t-h_1-(k-1)h_2) \\ \hline \vdots \\ x(t-(k-1)h_1) \\ x(t-(k-1)h_1-h_2) \\ \vdots \\ x(t-(k-1)h_1-(k-1)h_2) \end{bmatrix}, \quad (3)$$

where we have used (1). From (2) and (3) we obtain the following augmented system

$$\dot{\phi}_{k,k}(t) = \sum_{i=0}^2 (I_{k^2} \otimes A_i) X_{k,k}(t-h_i), \quad (4)$$

where

$$\phi_{k,k}(t) = X_{k,k}(t) - \sum_{i=1}^2 (I_{k^2} \otimes B_i) X_{k,k}(t-h_i).$$

Consider the structure of system (4), we hope that the vectors $X_{k,k}(t)$, $X_{k,k}(t-h_1)$ and $X_{k,k}(t-h_2)$ can be expressed as linear combinations of $\xi_1(t)$. Thus a possible state vector $\xi_1(t)$ should be

$$\xi_1(t) = X_{k+1,k+1}(t). \quad (5)$$

In order to ensure that the derivative of $\xi_2(t)$ can be expressed as a linear function of $\xi_1(t)$, a possible state vector $\xi_2(t)$ should be

$$\xi_2(t) = \begin{bmatrix} \int_{t-h_1}^t X_{k,k+1}(s) ds \\ \int_{t-h_2}^t X_{1,k}(s) ds \end{bmatrix}. \quad (6)$$

By (5) and (6), we conclude that a suitable $\xi(t)$ can be

$$\xi(t) = \begin{bmatrix} \xi_1(t) \\ \xi_2(t) \end{bmatrix} = \begin{bmatrix} X_{k+1,k+1}(t) \\ \int_{t-h_1}^t X_{k,k+1}(s) ds \\ \int_{t-h_2}^t X_{1,k}(s) ds \end{bmatrix}. \quad (7)$$

Remark 1: Noticed that, for a specified augmented state vector $\xi_1(t)$, there are many vectors with distributed delays whose derivative can be expressed as a linear function of $\xi_1(t)$. It should be pointed out here that the selected $\xi_2(t)$ in (6) has the smallest dimension among those variables whose derivative can be expressed as a linear function of $\xi_1(t)$ and can cover all the state variables $\xi_1(t)$. This can help us reduce the complexity of computation.

Based on the augmented state vector in (7), we will construct a suitable augmented LKF for system (2). Generally, the LKF for time-delay systems is the sum of a quadratic

functional $\tilde{V}(\xi(t))$ and some nonnegative QIFs. With the state vector $\xi(t)$ in (7), the quadratic functional $\tilde{V}(\xi(t))$ can be naturally expressed as

$$\tilde{V}(\xi(t)) = \begin{bmatrix} \phi_{k,k}(t) \\ \xi_2(t) \end{bmatrix}^T P \begin{bmatrix} \phi_{k,k}(t) \\ \xi_2(t) \end{bmatrix}, \quad (8)$$

where $P > 0$. In what follows, we need to determine the nonnegative QIFs. We know that the general form of a nonnegative QIF is

$$\int_{t-h_2}^{t-h_1} x^T(s) U x(s) ds, \quad h_1 < h_2, \quad (9)$$

where $U > 0$. The derivative of (9) can be expressed as a quadratic function of $[x^T(t-h_1), x^T(t-h_2)]^T$. Following this process, to ensure that the derivative of a QIF can be written as a quadratic function of $\xi_1(t)$, the QIF should be in the form of

$$\mathcal{V}_{ij}(x_t) = \left| \int_{t-jh_2}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \right|, \quad (10)$$

where $i, j \in \mathbb{I}[0, k]$, and U_{ij} are some positive definite matrices with appropriate dimensions. Since $i, j \in \mathbb{I}[0, k]$ in (10), all possible nonnegative QIFs contains $(k+1)^2$ different integral lengths. If the constructed LKF involves all these functions with $(k+1)^2$ different integral lengths, the computation burden is heavy, especially, when k is large. Fortunately, we notice that not all possible \mathcal{V}_{ij} are needed to be included. We take the $\mathcal{V}_{20}(x_t)$ (namely, $i=2$ and $j=0$) for example. By a simple computation, we get

$$\begin{aligned} \mathcal{V}_{20}(x_t) &= \int_{t-2h_1}^t X_{k-1,k+1}^T(s) U_{20} X_{k-1,k+1}(s) ds \\ &= \int_{t-h_1}^t X_{k,k+1}^T(s) \left(\lambda_1^T U_{20} \lambda_1 + \lambda_2^T U_{20} \lambda_2 \right) X_{k,k+1}(s) ds, \end{aligned}$$

where $U_{20} > 0$, and $\lambda_1 = [I_{k-1} \ 0_{(k-1) \times 1}] \otimes I_{(k+1)n}$, $\lambda_2 = [0_{(k-1) \times 1} \ I_{k-1}] \otimes I_{(k+1)n}$. It is clear that $\lambda_1^T U_{20} \lambda_1$ and $\lambda_2^T U_{20} \lambda_2$ are some positive semi-definite matrices. Thus $\mathcal{V}_{20}(x_t)$ can be absorbed by $\mathcal{V}_{10}(x_t)$ ($i=1$ and $j=0$) with an integral length h_1 .

In what follows, inspired by the observation above, we will make a significant effort in finding all redundant QIFs and eliminating them. In order to remove the absolute value sign of functions in (10), we discuss this problem by cases.

Case {1}: $i \leq j$. In this case, removing the absolute value sign of $\mathcal{V}_{ij}(x_t)$ in (10) gives

$$\mathcal{V}_{ij}(x_t) = \int_{t-jh_2}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds.$$

Then we can calculate

$$\begin{aligned}\mathcal{V}_{ij}(x_t) &= \int_{t-ih_2}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &\quad + \int_{t-jh_2}^{t-ih_2} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &= \sum_{m=1}^i \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) \lambda_{3m}^T U_{ij} \lambda_{3m} X_{k,k}(s) ds \\ &\quad + \sum_{m=i+1}^j \int_{t-h_2}^t X_{k+1,k}^T(s) \lambda_{4m}^T U_{ij} \lambda_{4m} X_{k+1,k}(s) ds,\end{aligned}$$

where

$$\begin{aligned}\lambda_{3m} &= [0_{(k+1-i) \times (m-1)} \quad I_{k+1-i} \quad 0_{(k+1-i) \times (i-m)}] \\ &\quad \otimes [0_{(k+1-j) \times (i-m)} \quad I_{k+1-j} \quad 0_{(k+1-j) \times (m-1+j-i)}] \otimes I_n, \\ \lambda_{4m} &= [I_{k+1-i} \quad 0_{(k+1-i) \times i}] \\ &\quad \otimes [0_{(k+1-j) \times (m-1)} \quad I_{k+1-j} \quad 0_{(k+1-j) \times (j-m)}] \otimes I_n.\end{aligned}$$

Notice that $\lambda_{3m}^T U_{ij} \lambda_{3m}$ and $\lambda_{4m}^T U_{ij} \lambda_{4m}$ are some positive semi-definite matrices. Thus the functionals $\mathcal{V}_{ij}(x_t)$ in Case {1} can be absorbed by the following two nonnegative QIFs

$$\mathcal{V}_{01}(x_t) = \int_{t-h_2}^t X_{k+1,k}^T(s) U_{01} X_{k+1,k}(s) ds, \quad (11)$$

$$\mathcal{V}_{11}(x_t) = \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) U_{11} X_{k,k}(s) ds. \quad (12)$$

Case {2}: $i > j$.

In this case, in order to remove the absolute value sign of $\mathcal{V}_{ij}(x_t)$, we introduce an integer p . For further use, we define $\check{\mu}_p = \lceil \frac{ph_2}{h_1} \rceil$ and $\check{\mu}_{p-1} = \lceil \frac{(p-1)h_2}{h_1} \rceil$ for $p \in \mathbb{N}^+$. Obviously, for a given integer k there must exist a $p \in \mathbb{N}^+$ satisfying

$$k \in \mathbb{I}[\check{\mu}_{p-1} + 1, \check{\mu}_p]. \quad (13)$$

From (13), we have $k \geq p$. With the help of (13), we take off the absolute value sign of some integral functionals $\mathcal{V}_{ij}(x_t)$, which will be discussed in three cases.

Case {2.1}: For $j \in \mathbb{I}[0, p-1]$, $i \in \mathbb{I}[\check{\mu}_j + 1, k]$ with $k \geq 1$, $p \geq 1$.

In Case {2.1}, we have $jh_2 < ih_1$ since $\check{\mu}_j \geq \frac{jh_2}{h_1}$ and $i \in \mathbb{I}[\check{\mu}_j + 1, k]$. Taking off the absolute value sign of $\mathcal{V}_{ij}(x_t)$ gives

$$\mathcal{V}_{ij}(x_t) = \int_{t-ih_1}^{t-jh_2} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds.$$

Then we can compute

$$\begin{aligned}\mathcal{V}_{ij}(x_t) &= \int_{t-\check{\mu}_j h_1}^{t-jh_2} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &\quad + \int_{t-ih_1}^{t-\check{\mu}_j h_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &= \sum_{m=\check{\mu}_j}^{i-1} \int_{t-h_1}^t X_{k,k+1}^T(s) \lambda_{5m}^T U_{ij} \lambda_{5m} X_{k,k+1}(s) ds \\ &\quad + \int_{t-\check{\mu}_j h_1}^{t-jh_2} X_{k+1-\check{\mu}_j,k+1-j}^T(s) \lambda_6^T U_{ij} \lambda_6 \\ &\quad \times X_{k+1-\check{\mu}_j,k+1-j}(s) ds,\end{aligned} \quad (14)$$

where

$$\begin{aligned}\lambda_{5m} &= [0_{(k+1-i) \times m} \quad I_{k+1-i} \quad 0_{(k+1-i) \times (i-m-1)}] \\ &\quad \otimes [I_{k+1-j} \quad 0_{(k+1-j) \times j}] \otimes I_n, \\ \lambda_6 &= [I_{k+1-i} \quad 0_{(k+1-i) \times (i-\check{\mu}_j)}] \otimes I_{(k+1-j)n}.\end{aligned}$$

It follows from (14) that $\mathcal{V}_{ij}(x_t)$ in Case {2.1} can be absorbed by

$$\begin{aligned}\mathcal{V}_{10}(x_t) &= \int_{t-h_1}^t X_{k,k+1}^T(s) U_{10} X_{k,k+1}(s) ds, \\ \mathcal{V}_{\check{\mu}_j j}(x_t) &= \int_{t-\check{\mu}_j h_1}^{t-jh_2} X_{k+1-\check{\mu}_j,k+1-j}^T(s) U_{\check{\mu}_j j} \\ &\quad \times X_{k+1-\check{\mu}_j,k+1-j}(s) ds, \quad j \in \mathbb{I}[0, p-1].\end{aligned}$$

Case {2.2}: For $j \in \mathbb{I}[2, p]$, $i \in \omega_l$ with $l \in \mathbb{I}[1, j-1]$, $p \geq 2$, $k \geq 4$, where if $j = 2$,

$$\omega_1 = \mathbb{I}[j+1, \check{\mu}_1],$$

and if $j \geq 3$,

$$\begin{aligned}\omega_1 &= \mathbb{I}[j+1, \check{\mu}_1 + j - 2], \\ \omega_2 &= \mathbb{I}[\check{\mu}_1 + j - 1, \check{\mu}_2 + j - 3], \\ \omega_3 &= \mathbb{I}[\check{\mu}_2 + j - 2, \check{\mu}_3 + j - 4], \\ &\vdots \\ \omega_{j-1} &= \mathbb{I}[\check{\mu}_{j-2} + 2, \check{\mu}_{j-1}].\end{aligned}$$

In this case, it is clear that $jh_2 > ih_1$. Then one can get

$$\begin{aligned}\mathcal{V}_{ij}(x_t) &= \int_{t-jh_2}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &= \int_{t-lh_2}^{t-(i-j+l)h_1} \hat{\mathcal{X}}_{ij}^T U_{ij} \hat{\mathcal{X}}_{ij} ds + \int_{t-(j-l)h_2}^{t-(j-l)h_1} \check{\mathcal{X}}_{ij}^T U_{ij} \check{\mathcal{X}}_{ij} ds \\ &= \int_{t-lh_2}^{t-(i-j+l)h_1} \hat{\mathcal{X}}_{ij}^T U_{ij} \hat{\mathcal{X}}_{ij} ds \\ &\quad + \sum_{m=1}^{j-l} \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) \lambda_{7m}^T U_{ij} \lambda_{7m} X_{k,k}(s) ds,\end{aligned} \quad (15)$$

in which $\hat{\mathcal{X}}_{ij} = X_{k+1-i,k+1-j}(s - (j-l)h_1)$, $\check{\mathcal{X}}_{ij} = X_{k+1-i,k+1-j}(s - lh_2)$, and

$$\begin{aligned}\lambda_{7m} &= [0_{(k+1-i) \times (m-1)} \quad I_{k+1-i} \quad 0_{(k+1-i) \times (i-m)}] \\ &\quad \otimes [0_{(k+1-j)n \times (j-m)n} \quad I_{(k+1-j)n} \quad 0_{(k+1-j)n \times (m-1)n}].\end{aligned} \quad (16)$$

Let $i' = i - j + l$, $l \in \mathbb{I}[1, j-1]$. Then we have $i' \in \tilde{\omega}_l$, $l \in \mathbb{I}[1, j-1]$, in which $j = 2$,

$$\tilde{\omega}_1 = \mathbb{I}[2, \check{\mu}_1 - 1], \quad (17)$$

and if $j \geq 3$,

$$\begin{cases} \tilde{\omega}_1 = \mathbb{I}[2, \check{\mu}_1 - 1], \\ \tilde{\omega}_2 = \mathbb{I}[\check{\mu}_1 + 1, \check{\mu}_2 - 1], \\ \tilde{\omega}_3 = \mathbb{I}[\check{\mu}_2 + 1, \check{\mu}_3 - 1], \\ \vdots \\ \tilde{\omega}_{j-1} = \mathbb{I}[\check{\mu}_{j-2} + 1, \check{\mu}_{j-1} - 1]. \end{cases} \quad (18)$$

Therefore (15) can be expressed as

$$\begin{aligned} \mathcal{V}_{ij}(x_t) &= \sum_{m=1}^{j-l} \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) \lambda_{7m}^T U_{ij} \lambda_{7m} X_{k,k}(s) ds \\ &\quad + \int_{t-lh_2}^{t-i'h_1} X_{k+1-i',k+1-l}^T(s) \lambda_8^T U_{ij} \lambda_8 X_{k+1-i',k+1-l}(s) ds, \end{aligned} \quad (19)$$

where

$$\begin{aligned} \lambda_8 &= [0_{(k+1-i) \times (i-i')} \quad I_{(k+1-i)}] \\ &\quad \otimes [I_{k+1-j} \quad 0_{(k+1-j) \times (j-l)}] \otimes I_n. \end{aligned}$$

From (19), we know that $\mathcal{V}_{ij}(x_t)$ in Case {2.2} can be absorbed by

$$\begin{aligned} \mathcal{V}_{11}(x_t) &= \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) U_{11} X_{k,k}(s) ds, \\ \mathcal{V}_{i'l}(x_t) &= \int_{t-lh_2}^{t-i'h_1} X_{k+1-i',k+1-l}^T(s) U_{i'l} X_{k+1-i',k+1-l}(s) ds, \end{aligned} \quad (20)$$

where $i' \in \tilde{\omega}_l$, and $\tilde{\omega}_l$ is defined in (17)/(18).

Case {2.3}: For $j \in \mathbb{I}[p+1, k-1]$, $i \in \mathbb{I}[j+1, k]$ with $k \geq 3$, $1 \leq p \leq k-2$.

Since $k \leq \left\lceil \frac{ph_2}{h_1} \right\rceil < \frac{ph_2}{h_1} + 1$, we have $kh_1 < (p+1)h_2$, which yields $jh_2 > ih_1$ in Case {2.3}. Taking off the absolute value sign of $\mathcal{V}_{ij}(x_t)$ gives

$$\mathcal{V}_{ij}(x_t) = \int_{t-jh_2}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds.$$

Then we can compute

$$\begin{aligned} \mathcal{V}_{ij}(x_t) &= v_{ij} + \int_{t-ph_2-(j-p)h_1}^{t-ph_2-(j-p)h_2} X_{k+1-i,k+1-j}^T(s) \\ &\quad \times U_{ij} X_{k+1-i,k+1-j}(s) ds \\ &= v_{ij} + \sum_{m=1}^{j-p} \int_{t-h_2}^{t-h_1} X_{k,k}^T(s) \lambda_{7m}^T U_{ij} \lambda_{7m} X_{k,k}(s) ds, \end{aligned}$$

in which λ_{7m} is defined in (16) and

$$v_{ij} = \int_{t-ph_2-(j-p)h_1}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds.$$

Notice that $i-j+p \in \mathbb{I}[p+1, k-1]$. When $p=1$, we have $i-j+1 \in \mathbb{I}[2, k-1]$. Obviously, the functional

$$v_{ij} = \int_{t-h_2-(j-1)h_1}^{t-ih_1} X_{k+1-i,k+1-j}^T(s) U_{ij} X_{k+1-i,k+1-j}(s) ds$$

can not be removed. When $p \geq 2$, we divide the region $\mathbb{I}[p+1, k-1]$ into two parts for analysis. For $q_p \triangleq i-j+p \in \mathbb{I}[\check{\mu}_{p-1}+1, k-1]$, the functional

$$v_{ij} = \int_{t-ph_2}^{t-q_ph_1} X_{k+1-q_p,k+1-p}^T(s) \lambda_9^T U_{ij} \lambda_9 X_{k+1-q_p,k+1-p}(s) ds,$$

in which

$$\begin{aligned} \lambda_9 &= [0_{(k+1-i) \times (j-p)} \quad I_{k+1-i}] \\ &\quad \otimes [I_{k+1-j} \quad 0_{(k+1-j) \times (j-p)}] \otimes I_n. \end{aligned}$$

For $i-j+p \in \mathbb{I}[p+1, \check{\mu}_{p-1}]$, it follows from the Case {2.2} that the functional v_{ij} can be eliminated by $\mathcal{V}_{11}(x_t)$ and (20).

Now, by removing the redundant functionals in Cases {1} {2.1}, {2.2}, {2.3}, we get all nonnegative QIFs $\mathcal{V}_{ij}(x_t)$ which should be included in the constructed LKF. For clarity, we write them below

$$\mathcal{V}_{10}(x_t) = \int_{t-h_1}^t X_{k,k+1}^T(s) U_{10} X_{k,k+1}(s) ds, \quad (21)$$

$$\mathcal{V}_{01}(x_t) = \int_{t-h_2}^t X_{k+1,k}^T(s) U_{01} X_{k+1,k}(s) ds, \quad (22)$$

$$\mathcal{V}_{il}(x_t) = \left| \int_{t-lh_2}^{t-ih_1} X_{k+1-i,k+1-l}^T(s) U_{il} X_{k+1-i,k+1-l}(s) ds \right|, \quad (23)$$

where U_{10}, U_{01}, U_{il} are some positive matrices with appropriate dimensions, $l \in \mathbb{I}[1, p]$, $i \in q_l$, and

- if $1 \leq l \leq p-1$,

$$q_l = \mathbb{I}[\check{\mu}_{l-1}+1, \check{\mu}_l]; \quad (24)$$

- if $l = p$,

$$q_p = \mathbb{I}[\check{\mu}_{p-1}+1, k]. \quad (25)$$

On the other hand, in order to ensure that the time-derivative of $\mathcal{V}_{10}, \mathcal{V}_{01}, \mathcal{V}_{il}$ still contains integral functions [19], the sum of nonnegative QIFs can be expressed as

$$\begin{aligned} \check{V}(x_t) &= \int_{t-h_1}^t X_{k,k+1}^T(s) \left(Q_1 + \frac{s-t+h_1}{h_1} W_1 \right) X_{k,k+1}(s) ds \\ &\quad + \int_{t-h_2}^t X_{k+1,k}^T(s) \left(Q_2 + \frac{s-t+h_2}{h_2} W_2 \right) X_{k+1,k}(s) ds \\ &\quad + \sum_{l=1}^p V_l(x_t), \end{aligned} \quad (26)$$

where $Q_1, Q_2, W_1, W_2 \in \mathbb{S}^{k(1+k)n}$ are four positive definite matrices and

- if $1 \leq l \leq p-1$,

$$V_l(x_t) = \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l-1} \hat{v}_{k,k}(i, l) + \check{v}_{k,k}(\check{\mu}_l, l),$$

- if $l = p$,

$$V_p(x_t) = \begin{cases} \sum_{i=\check{\mu}_{p-1}+1}^k \hat{v}_{k,k}(i, p), & k < \check{\mu}_p, \\ \sum_{i=\check{\mu}_{p-1}+1}^{k-1} \hat{v}_{k,k}(i, p) + \check{v}_{k,k}(k, p), & k = \check{\mu}_p, \end{cases}$$

in which

$$\begin{aligned} \hat{v}_{\varsigma,\varepsilon}(i, l) &= \int_{t-lh_2}^{t-ih_1} X_{\varsigma+1-i,\varepsilon+1-l}^T(s) \\ &\quad \times \frac{s-t+lh_2}{lh_2-ih_1} W_{i+2} X_{\varsigma+1-i,\varepsilon+1-l}(s) ds, \end{aligned}$$

$$\begin{aligned}\check{\vartheta}_{\varsigma,\varepsilon}(i,l) &= \int_{t-lh_2}^{t-ih_1} X_{\varsigma+1-i,\varepsilon+1-l}^T(s) \\ &\quad \times \frac{s-t+ih_1}{lh_2-ih_1} W_{i+2} X_{\varsigma+1-i,\varepsilon+1-l}(s) ds,\end{aligned}$$

with W_3, W_4, \dots, W_{k+2} being some positive definite matrices.

Finally, it follows from (8) and (26) that the new augmented LKF for system (2) can be written as

$$V(x_t) = \tilde{V}(x_t) + \check{V}(x_t), \quad (27)$$

in which $\tilde{V}(x_t) = \tilde{V}(\xi(t))$ is defined in (8) and $\check{V}(x_t)$ is defined in (26). Noticed that, in $V_l(x_t)$ of (26), we have taken the case of $lh_2 \neq ih_1$, $i \in q_l$, $l \in \mathbb{I}[1, p]$ into consideration.

Remark 2: In $V_l(x_t)$, if $lh_2 = ih_1$ for some $i \in q_l$, $l \in \mathbb{I}[1, p]$, the quadratic integral term $\hat{\vartheta}_{k,k}(i, l)$ or $\check{\vartheta}_{k,k}(i, l)$ in (26) is absent, namely, $W_{i+2} = 0$.

Remark 3: Compared with the existing LKFs in [5], [9], [19], and [27], we see from (26) that the proposed LKF (27) includes more information about the delayed states and time delays since the introduction of the augmented state variable $X_{k,k}(t)$ and the delay-product-type integral terms $V_l(x_t)$, $l \in \mathbb{I}[1, p]$. They are essential for deriving less conservative results.

Remark 4: It can be seen from Section III(A) that the proper LKF (27) is constructed by taking 4 different cases into account. However, for the case of more delays, it is obvious that much more than 4 different cases should be taken into account to construct a suitable LKF, which will result in a great challenge in the stability analysis. Such a problem deserves a further study.

B. Stability Criteria

Based on the LKF (27), we propose a stability criterion for system (2), which exhibits a hierarchical character: the larger the integer k , the less conservatism of the resulting stability criterion, since more decision variables are involved in (27). To keep the representation simple, we define

$$\mathcal{A}_k = \begin{bmatrix} \mathcal{A}_{k11} & 0_{k^2n \times gn} \\ \mathcal{A}_{k21} & 0_{gn} \end{bmatrix}, \mathcal{B}_k = \begin{bmatrix} \mathcal{B}_{k11} & 0_{k^2n \times gn} \\ 0_{gn \times (k+1)^2n} & I_{gn} \end{bmatrix}, \quad (28)$$

in which, for $k \in \mathbb{N}^+$, $g = k(k+2)$ and

$$\begin{aligned}\mathcal{A}_{k11} &= L_k \otimes L_k \otimes A_0 + R_k \otimes L_k \otimes A_1 + L_k \otimes R_k \otimes A_2, \\ \mathcal{B}_{k11} &= L_k \otimes L_k \otimes I_n - R_k \otimes L_k \otimes B_1 - L_k \otimes R_k \otimes B_2, \\ \mathcal{A}_{k21} &= \begin{bmatrix} L_k \otimes I_{(k+1)n} - R_k \otimes I_{(k+1)n} \\ [1 \ 0_{1 \times k}] \otimes L_k \otimes I_n - [1 \ 0_{1 \times k}] \otimes R_k \otimes I_n \end{bmatrix},\end{aligned}$$

with

$$L_k = [I_k \ 0_{k \times 1}], \quad R_k = [0_{k \times 1} \ I_k].$$

Define

$$\begin{aligned}\mathcal{C}_{\varsigma,\varepsilon}(i,l) &= [0_{(\varsigma+1-i) \times i} \ I_{\varsigma+1-i}] \\ &\quad \otimes [I_{(\varepsilon+1-l)} \ 0_{(\varepsilon+1-l) \times l}] \otimes I_n,\end{aligned} \quad (29)$$

$$\begin{aligned}\mathcal{D}_{\varsigma,\varepsilon}(i,l) &= [I_{\varsigma+1-i} \ 0_{(\varsigma+1-i) \times i}] \\ &\quad \otimes [0_{(\varepsilon+1-l) \times l} \ I_{(\varepsilon+1-l)}] \otimes I_n,\end{aligned} \quad (30)$$

for any $l \in \mathbb{I}[1, \varepsilon]$ and $i \in \mathbb{I}[1, \varsigma]$ with $\varepsilon, \varsigma \in \mathbb{N}^+$. With the help of (29) and (30), we define

$$\begin{aligned}\Psi_{\varsigma,\varepsilon}^* &= \mathcal{D}_{\varsigma,\varepsilon}^T(1,0) \left(Q_1 + h_1^2 W_1 \right) \mathcal{D}_{\varsigma,\varepsilon}(1,0) \\ &\quad + \mathcal{C}_{\varsigma,\varepsilon}^T(0,1) (Q_2 + h_2^2 W_2) \mathcal{C}_{\varsigma,\varepsilon}(0,1) \\ &\quad - \mathcal{C}_{\varsigma,\varepsilon}^T(1,0) Q_1 \mathcal{C}_{\varsigma,\varepsilon}(1,0) - \mathcal{D}_{\varsigma,\varepsilon}^T(0,1) Q_2 \mathcal{D}_{\varsigma,\varepsilon}(0,1),\end{aligned} \quad (31)$$

and, if $1 \leq l \leq p-1$,

$$\begin{aligned}\Psi_{\varsigma,\varepsilon}(i,l) &= \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l-1} \zeta_{il}^2 \mathcal{C}_{\varsigma,\varepsilon}^T(i,l) W_{i+2} \mathcal{C}_{\varsigma,\varepsilon}(i,l) \\ &\quad + \zeta_{\check{\mu}_l}^2 \mathcal{D}_{\varsigma,\varepsilon}^T(\check{\mu}_l, l) W_{\check{\mu}_l+2} \mathcal{D}_{\varsigma,\varepsilon}(\check{\mu}_l, l),\end{aligned} \quad (32)$$

if $l = p$,

$$\Psi_{\varsigma,\varepsilon}(i,p) = \begin{cases} \sum_{i=\check{\mu}_{p-1}+1}^k \zeta_{ip}^2 \mathcal{C}_{\varsigma,\varepsilon}^T(i,p) W_{i+2} \mathcal{C}_{\varsigma,\varepsilon}(i,p), & k < \check{\mu}_p, \\ \sum_{i=\check{\mu}_{p-1}+1}^{\check{\mu}_p-1} \zeta_{ip}^2 \mathcal{C}_{\varsigma,\varepsilon}^T(i,p) W_{i+2} \mathcal{C}_{\varsigma,\varepsilon}(i,p) \\ \quad + \zeta_{\check{\mu}_p}^2 \mathcal{D}_{\varsigma,\varepsilon}^T(\check{\mu}_p, p) W_{\check{\mu}_p+2} \mathcal{D}_{\varsigma,\varepsilon}(\check{\mu}_p, p), & k = \check{\mu}_p, \end{cases} \quad (33)$$

in which $\zeta_{il} = lh_2 - ih_1$, $l \in \mathbb{I}[1, p]$, $i \in \mathbb{I}[1, k]$.

For $l \in \mathbb{I}[1, \varepsilon]$ and $i \in \mathbb{I}[1, \varsigma]$, we define

$$\begin{aligned}\Xi_{\varsigma,\varepsilon}(i,l) &= \sum_{m=1}^i [0_{(\varsigma+1-i) \times (m-1)} \ I_{\varsigma+1-i} \ 0_{(\varsigma+1-i) \times (i-m)}] \\ &\quad \otimes [I_{\varepsilon+1-l} \ 0_{(\varepsilon+1-l) \times l}] \otimes I_n, \\ \pi_{\varsigma,\varepsilon}(i,l) &= \sum_{d=1}^l [I_{\varsigma+1-i} \ 0_{(\varsigma+1-i) \times i}] \\ &\quad \otimes [0_{(\varepsilon+1-l) \times (d-1)} \ I_{\varepsilon+1-l} \ 0_{(\varepsilon+1-l) \times (l-d)}] \otimes I_n,\end{aligned}$$

and $\alpha_\varepsilon = -L_\varepsilon \otimes I_n + R_\varepsilon \otimes I_n$,

$$\gamma_\varsigma = \begin{bmatrix} 0_{1 \times \varsigma} \\ 1 \ 0_{1 \times (\varsigma-1)} \\ 1_{1 \times 2} \ 0_{1 \times (\varsigma-2)} \\ \vdots \\ 1_{1 \times \varsigma} \end{bmatrix}, \beta_{\varsigma,\varepsilon} = \begin{bmatrix} I_{\varepsilon n} \\ I_{\varepsilon n} \\ \vdots \\ I_{\varepsilon n} \end{bmatrix} \in \mathbb{R}^{(\varsigma+1)\varepsilon n \times \varepsilon n}.$$

Then we define

$$\begin{aligned}\Phi_{\varsigma,\varepsilon}^* &= -[\gamma_\varsigma \otimes \alpha_\varepsilon \ \beta_{\varsigma,\varepsilon}]^T W_2 [\gamma_\varsigma \otimes \alpha_\varepsilon \ \beta_{\varsigma,\varepsilon}] \\ &\quad - [I_{\varsigma(\varepsilon+1)n} \ 0_{\varsigma(\varepsilon+1)n \times \varepsilon n}]^T W_1 [I_{\varsigma(\varepsilon+1)n} \ 0_{\varsigma(\varepsilon+1)n \times \varepsilon n}],\end{aligned} \quad (34)$$

and, if $1 \leq l \leq p-1$,

$$\Phi_{\varsigma,\varepsilon}(i,l) = \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l} \mathcal{H}_{\varsigma,\varepsilon}^T(i,l) W_{i+2} \mathcal{H}_{\varsigma,\varepsilon}(i,l), \quad (35)$$

and, if $l = p$,

$$\Phi_{\varsigma,\varepsilon}(i,p) = \sum_{i=\check{\mu}_{p-1}+1}^k \mathcal{H}_{\varsigma,\varepsilon}^T(i,p) W_{i+2} \mathcal{H}_{\varsigma,\varepsilon}(i,p), \quad (36)$$

where

$$\mathcal{H}_{\zeta,\varepsilon}(i,l) = [\pi_{\zeta,\varepsilon}(i,l) (\gamma_{\zeta} \otimes \alpha_{\varepsilon}) - \Xi_{\zeta,\varepsilon}(i,l), \pi_{\zeta,\varepsilon}(i,l) \beta_{\zeta,\varepsilon}]. \quad (37)$$

We now state the following hierarchical stability criterion.

Theorem 1: Let Assumption 1 be satisfied and $lh_2 \neq ih_1$ for $l \in \mathbb{I}[1, p]$ with p satisfying (13) and $i \in q_l$ where q_l is defined in (24) and (25). Then system (2) is asymptotically stable, if there exists a positive definite matrix $P \in \mathbb{S}^{(g+k^2)n}$, four positive definite matrices $Q_1, Q_2, W_1, W_2 \in \mathbb{S}^{k(1+k)n}$, and k positive definite matrices $W_{i+2} \in \mathbb{S}^{(k+1-i)(k+1-l)n}$, $i \in q_l$, $l \in \mathbb{I}[1, p]$, such that

$$\mathcal{R}_k = \mathcal{A}_k^T P \mathcal{B}_k + \mathcal{B}_k^T P \mathcal{A}_k + \Gamma_k < 0, \quad (38)$$

where \mathcal{A}_k and \mathcal{B}_k are defined in (28) and

$$\Gamma_k = \begin{bmatrix} \Psi_{k,k}^* + \sum_{l=1}^p \Psi_{k,k}(i,l) & 0_{(k+1)^2 n \times gn} \\ 0_{gn \times (k+1)^2 n} & \Phi_{k,k}^* - \sum_{l=1}^p \Phi_{k,k}(i,l) \end{bmatrix}$$

with $\Psi_{k,k}^*, \Psi_{k,k}(i,l)$ and $\Phi_{k,k}^*, \Phi_{k,k}(i,l)$ being defined by (31)-(33) and (34)-(36), respectively.

Proof: Simple computations give that

$$\begin{aligned} \dot{V}(x_t) &= \begin{bmatrix} \dot{\varphi}_{k,k}(t) \\ \dot{\zeta}_2(t) \end{bmatrix}^T P \begin{bmatrix} \varphi_{k,k}(t) \\ \zeta_2(t) \end{bmatrix} + \begin{bmatrix} \varphi_{k,k}(t) \\ \zeta_2(t) \end{bmatrix}^T P \begin{bmatrix} \dot{\varphi}_{k,k}(t) \\ \dot{\zeta}_2(t) \end{bmatrix} \\ &= \zeta^T(t) \left(\mathcal{A}_k^T P \mathcal{B}_k + \mathcal{B}_k^T P \mathcal{A}_k \right) \zeta(t). \end{aligned} \quad (39)$$

Notice that

$$\begin{aligned} \int_{t-h_1}^t X_{k,k+1}(s) ds &= [I_{k(k+1)n} \ 0_{k(k+1)n \times kn}] \zeta_2(t) \triangleq \Omega_1 \zeta_2(t), \\ \int_{t-h_2}^t X_{k+1,k}(s) ds &= [\gamma_k \otimes \alpha_k \ \beta_{k,k}] \zeta_2(t) \triangleq \Omega_2 \zeta_2(t). \end{aligned}$$

Then the time derivative of the first two terms in $\dot{V}(x_t)$ can be evaluated as

$$\begin{aligned} &X_{k+1,k+1}^T(t) \tilde{\Psi}_{k,k}^* X_{k+1,k+1}(t) - \int_{t-h_1}^t X_{k,k+1}^T(s) \\ &\quad \times \frac{W_1}{h_1} X_{k,k+1}(s) ds - \int_{t-h_2}^t X_{k+1,k}^T(s) \frac{W_2}{h_2} X_{k+1,k}(s) ds \\ &\leq X_{k+1,k+1}^T(t) \tilde{\Psi}_{k,k}^* X_{k+1,k+1}(t) \\ &\quad - \left(\int_{t-h_1}^t X_{k,k+1}^T(s) ds \right) \frac{W_1}{h_1^2} \int_{t-h_1}^t X_{k,k+1}(s) ds \\ &\quad - \left(\int_{t-h_2}^t X_{k+1,k}^T(s) ds \right) \frac{W_2}{h_2^2} \int_{t-h_2}^t X_{k+1,k}(s) ds \\ &\leq X_{k+1,k+1}^T(t) \tilde{\Psi}_{k,k}^* X_{k+1,k+1}(t) + \zeta_2^T(t) \tilde{\Phi}_{k,k}^* \zeta_2(t), \end{aligned} \quad (40)$$

where we have used the well known Jensen inequality and

$$\begin{aligned} \tilde{\Psi}_{k,k}^* &= \mathcal{D}_{k,k}^T(1,0) (Q_1 + W_1) \mathcal{D}_{k,k}(1,0) \\ &\quad + \mathcal{C}_{k,k}^T(0,1) (Q_2 + W_2) \mathcal{C}_{k,k}(0,1) \\ &\quad - \mathcal{C}_{k,k}^T(1,0) Q_1 \mathcal{C}_{k,k}(1,0) - \mathcal{D}_{k,k}^T(0,1) Q_2 \mathcal{D}_{k,k}(0,1), \\ \tilde{\Phi}_{k,k}^* &= -\frac{1}{h_2^2} \Omega_2^T W_2 \Omega_2 - \frac{1}{h_1^2} \Omega_1^T W_1 \Omega_1. \end{aligned}$$

The time derivative of $\sum_{l=1}^p V_l(x_t)$ is estimated as

$$\begin{aligned} \sum_{l=1}^p \dot{V}_l(x_t) &= \sum_{l=1}^p X_{k+1,k+1}^T(t) \tilde{\Psi}_{k,k}(i,l) X_{k+1,k+1}(t) \\ &\quad - \sum_{l=1}^{p-1} \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l} \int_{t-lh_2}^{t-lh_1} X_{k+1-i,k+1-l}^T(s) \\ &\quad \times \frac{W_{i+2}}{\zeta_{il}} X_{k+1-i,k+1-l}(s) ds \\ &\quad - \sum_{i=\check{\mu}_{p-1}+1}^k \int_{t-ph_2}^{t-lh_1} X_{k+1-i,k+1-p}^T(s) \\ &\quad \times \frac{W_{i+2}}{\zeta_{ip}} X_{k+1-i,k+1-p}(s) ds \\ &\leq \sum_{l=1}^p X_{k+1,k+1}^T(t) \tilde{\Psi}_{k,k}(i,l) X_{k+1,k+1}(t) \\ &\quad - \sum_{l=1}^{p-1} \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l} \left(\int_{t-lh_2}^{t-lh_1} X_{k+1-i,k+1-l}^T(s) ds \right) \\ &\quad \times \frac{W_{i+2}}{\zeta_{il}^2} \left(\int_{t-lh_2}^{t-lh_1} X_{k+1-i,k+1-l}(s) ds \right) \\ &\quad - \sum_{i=\check{\mu}_{p-1}+1}^k \left(\int_{t-ph_2}^{t-lh_1} X_{k+1-i,k+1-p}^T(s) ds \right) \\ &\quad \times \frac{W_{i+2}}{\zeta_{ip}^2} \left(\int_{t-ph_2}^{t-lh_1} X_{k+1-i,k+1-p}(s) ds \right), \end{aligned} \quad (41)$$

where $\zeta_{il} = lh_2 - ih_1$, and if $1 \leq l \leq p-1$,

$$\begin{aligned} \tilde{\Psi}_{k,k}(i,l) &= \sum_{i=\check{\mu}_{l-1}+1}^{\check{\mu}_l-1} \mathcal{C}_{k,k}^T(i,l) W_{i+2} \mathcal{C}_{k,k}(i,l) \\ &\quad + \mathcal{D}_{k,k}^T(\check{\mu}_l, l) W_{\check{\mu}_l+2} \mathcal{D}_{k,k}(\check{\mu}_l, l), \end{aligned}$$

if $l = p$,

$$\tilde{\Psi}_{k,k}(i,p) = \begin{cases} \sum_{i=\check{\mu}_{l-1}+1}^k \mathcal{C}_{k,k}^T(i,p) W_{i+2} \mathcal{C}_{k,k}(i,p), & k < \check{\mu}_p, \\ \sum_{i=\check{\mu}_{p-1}+1}^{\check{\mu}_p-1} \mathcal{C}_{k,k}^T(i,p) W_{i+2} \mathcal{C}_{k,k}(i,p) \\ \quad + \mathcal{D}_{k,k}^T(k,p) W_{k+2} \mathcal{D}_{k,k}(k,p), & k = \check{\mu}_p. \end{cases}$$

It can be verified that

$$\begin{aligned} &\int_{t-lh_2}^{t-lh_1} X_{k+1-i,k+1-l}(s) ds \\ &= - \int_{t-lh_1}^t X_{k+1-i,k+1-l}(s) ds + \int_{t-lh_2}^t X_{k+1-i,k+1-l}(s) ds \\ &= [\pi_{k,k}(i,l) (\gamma_k \otimes \alpha_k) - \Xi_{k,k}(i,l), \pi_{k,k}(i,l) \beta_{k,k}] \zeta_2(t) \\ &= \mathcal{H}_{k,k}(i,l) \zeta_2(t), \end{aligned} \quad (42)$$

where $\mathcal{H}_{k,k}(i,l)$ is defined by (37). Then it follows from (41) and (42) that

$$\begin{aligned} \sum_{l=1}^p \dot{V}_l(x_t) &\leq X_{k+1,k+1}^T(t) \sum_{l=1}^p \tilde{\Psi}_{k,k}(i,l) X_{k+1,k+1}(t) \\ &\quad - \zeta_2^T(t) \sum_{l=1}^p \tilde{\Phi}_{k,k}(i,l) \zeta_2(t), \end{aligned} \quad (43)$$

in which, if $1 \leq l \leq p-1$,

$$\tilde{\Phi}_{k,k}(i, l) = \sum_{i=\tilde{\mu}_{l-1}+1}^{\tilde{\mu}_l} \frac{1}{\zeta_{il}^2} \mathcal{H}_{k,k}^T(i, l) W_{i+2} \mathcal{H}_{k,k}(i, l),$$

if $l = p$,

$$\tilde{\Phi}_{k,k}(i, p) = \sum_{i=\tilde{\mu}_{p-1}+1}^k \frac{1}{\zeta_{ip}^2} \mathcal{H}_{k,k}^T(i, p) W_{i+2} \mathcal{H}_{k,k}(i, p).$$

Combining (39), (40), and (43), we have

$$\dot{V}(x_t) = \dot{\tilde{V}}(x_t) + \dot{\check{V}}(x_t) \leq \zeta^T(t) \tilde{\Pi} \zeta(t), \quad (44)$$

where

$$\begin{aligned} \tilde{\Pi} = & \mathcal{A}_k^T P \mathcal{B}_k + \mathcal{B}_k^T P \mathcal{A}_k \\ & + \begin{bmatrix} \tilde{\Psi}_{k,k}^* + \sum_{l=1}^p \tilde{\Psi}_{k,k}(i, l) & 0_{(k+1)^2 n \times gn} \\ 0_{gn \times (k+1)^2 n} & \tilde{\Phi}_{k,k}^* - \sum_{l=1}^p \tilde{\Phi}_{k,k}(i, l) \end{bmatrix}, \end{aligned}$$

which is equivalent to (38) by setting $\frac{1}{h_1^2} W_1 \rightarrow W_1$, $\frac{1}{h_2^2} W_2 \rightarrow W_2$, and $\frac{1}{\zeta_{il}^2} W_{i+2} \rightarrow W_{i+2}$. Thus we have from (44) and (38) that $\dot{V}(x_t) < 0$. Note that Assumption 1 guarantees that the operator \mathcal{D} is stable. According to Theorem 8.1 of [13], we conclude that system (2) is asymptotically stable. ■

Notice that the range of i and l in the augmented state variable $X_{k+1-i, k+1-l}(s)$ of $\dot{V}(x_t)$ is $i \in \mathbf{I}[0, k]$ and $l \in \mathbf{I}[1, p]$. Since $p \leq k$, in order to reduced the computational complexity, the augmented state variable $X_{k+1-i, k+1-l}(s)$ in $\dot{V}(x_t)$ can be replaced with $X_{k+1-i, p+1-l}(s)$. Then the augmented LKF for system (2) is

$$\begin{aligned} \mathcal{W}(x_t) = & \begin{bmatrix} \tilde{\varphi}_{k,p}(t) \\ \int_{t-h_1}^t X_{k,p+1}(s) ds \\ \int_{t-h_2}^t X_{1,p}(s) ds \end{bmatrix}^T P \begin{bmatrix} \tilde{\varphi}_{k,p}(t) \\ \int_{t-h_1}^t X_{k,p+1}(s) ds \\ \int_{t-h_2}^t X_{1,p}(s) ds \end{bmatrix} \\ & + \int_{t-h_1}^t X_{k,p+1}^T(s) \left(Q_1 + \frac{s-t+h_1}{h_1} W_1 \right) X_{k,p+1}(s) ds \\ & + \int_{t-h_2}^t X_{k+1,p}^T(s) \left(Q_2 + \frac{s-t+h_2}{h_2} W_2 \right) X_{k+1,p}(s) ds \\ & + \sum_{l=1}^p \mathcal{W}_l(x_t), \end{aligned} \quad (45)$$

where $\tilde{\varphi}_{k,p}(t) = X_{k,p}(t) - \sum_{i=1}^2 (I_{kp} \otimes B_i) X_{k,p}(t-h_i)$ and

- if $1 \leq l \leq p-1$,

$$\mathcal{W}_l(x_t) = \sum_{i=\tilde{\mu}_{l-1}+1}^{\tilde{\mu}_l} \hat{v}_{k,p}(i, l) + \check{v}_{k,p}(\tilde{\mu}_l, l),$$

- if $l = p$,

$$\mathcal{W}_p(x_t) = \begin{cases} \sum_{i=\tilde{\mu}_{p-1}+1}^k \hat{v}_{k,p}(i, p), & k < \tilde{\mu}_p, \\ \sum_{i=\tilde{\mu}_{p-1}+1}^{k-1} \hat{v}_{k,p}(i, p) + \check{v}_{k,p}(k, p), & k = \tilde{\mu}_p. \end{cases}$$

Based on the LKF $\mathcal{W}(x_t)$ in (45) and following a similar analysis of Theorem 1, we derive a hierarchical stability criterion with less decision variable.

Corollary 1: Let Assumption 1 be satisfied and $lh_2 \neq ih_1$ for $l \in \mathbf{I}[1, p]$ with p satisfying (13) and $i \in q_l$ where q_l is defined in (24) and (25). Then system (2) is asymptotically stable, if there exists a positive definite matrix $P \in \mathbb{S}^{(2kp+k+p)n}$, two positive definite matrices $Q_1, W_1 \in \mathbb{S}^{k(1+p)n}$, two positive definite matrices $Q_2, W_2 \in \mathbb{S}^{p(1+k)n}$, and k positive definite matrices $W_{i+2} \in \mathbb{S}^{(k+1-i)(p+1-l)n}$, $i \in q_l$, $l \in \mathbf{I}[1, p]$, such that

$$\hat{\mathcal{A}}_k^T P \hat{\mathcal{B}}_k + \hat{\mathcal{B}}_k^T P \hat{\mathcal{A}}_k + \hat{\Gamma}_k < 0,$$

where

$$\begin{aligned} \hat{\mathcal{A}}_k = & \begin{bmatrix} \hat{\mathcal{A}}_{k11} & 0_{kp n \times \hat{g} n} \\ \hat{\mathcal{A}}_{k21} & 0_{\hat{g} n} \end{bmatrix}, \quad \hat{\mathcal{B}}_k = \begin{bmatrix} \hat{\mathcal{B}}_{k11} & 0_{kp n \times \hat{g} n} \\ 0_{\hat{g} n \times \hat{g} n} & I_{\hat{g} n} \end{bmatrix}, \\ \hat{\Gamma}_k = & \begin{bmatrix} \Psi_{k,p}^* + \sum_{l=1}^p \Psi_{k,p}(i, l) & 0_{\hat{g} n \times \hat{g} n} \\ 0_{\hat{g} n \times \hat{g} n} & \Phi_{k,p}^* - \sum_{l=1}^p \Phi_{k,p}(i, l) \end{bmatrix}, \end{aligned}$$

in which $\Psi_{k,p}^*$, $\Psi_{k,p}(i, l)$ and $\Phi_{k,p}^*$, $\Phi_{k,p}(i, l)$ are defined by (31)-(33) and (34)-(36), and $\hat{g} = k(p+1) + p$, $\check{g} = (k+1)(p+1)$,

$$\begin{aligned} \hat{\mathcal{A}}_{k11} = & L_k \otimes L_p \otimes A_0 + R_k \otimes L_p \otimes A_1 + L_k \otimes R_p \otimes A_2, \\ \hat{\mathcal{B}}_{k11} = & L_k \otimes L_p \otimes I_n - R_k \otimes L_p \otimes B_1 - L_k \otimes R_p \otimes B_2, \\ \hat{\mathcal{A}}_{k21} = & \begin{bmatrix} L_k \otimes I_{(p+1)n} - R_k \otimes I_{(p+1)n} \\ [1 \ 0_{1 \times k}] \otimes L_p \otimes I_n - [1 \ 0_{1 \times k}] \otimes R_p \otimes I_n \end{bmatrix}. \end{aligned}$$

If we set $W_1 = W_2 = \dots = W_{k+2} = 0$ in (27), then we have

$$\begin{aligned} V(x_t) = & \varphi_{k,k}^T(t) P \varphi_{k,k}(t) + \int_{t-h_1}^t X_{k,k+1}^T(s) Q_1 X_{k,k+1}(s) ds \\ & + \int_{t-h_2}^t X_{k+1,k}^T(s) Q_2 X_{k+1,k}(s) ds. \end{aligned}$$

By computing $\dot{V}(x_t)$, we get a delay-independent stability condition.

Proposition 1: Let Assumption 1 be satisfied. Then system (2) is asymptotically stable for all delays, if there exists a positive definite matrix $P \in \mathbb{S}^{k^2 n}$, two positive definite matrices $Q_1, Q_2 \in \mathbb{S}^{k(k+1)n}$ such that

$$\tilde{\mathcal{R}}_k = \mathcal{A}_{k11}^T P \mathcal{B}_{k11} + \mathcal{B}_{k11}^T P \mathcal{A}_{k11} + O_k < 0, \quad (46)$$

where

$$O_k = \eta_L^T Q_1 \eta_L - \eta_R^T Q_1 \eta_R + \tilde{\eta}_L^T Q_2 \tilde{\eta}_L - \tilde{\eta}_R^T Q_2 \tilde{\eta}_R,$$

with $\eta_L = L_k \otimes I_{(k+1)n}$, $\eta_R = R_k \otimes I_{(k+1)n}$, $\tilde{\eta}_L = I_{(k+1)} \otimes L_k \otimes I_n$, $\tilde{\eta}_R = I_{(k+1)} \otimes R_k \otimes I_n$.

In fact, Proposition 1 can also be obtained by a frequency-domain method. The details can be found in Appendix.

Remark 5: For the case of $lh_2 = ih_1$, $i \in q_l$, $l \in \mathbf{I}[1, p]$, according to the discussion in [19] and Remark 2, Theorem 1 is still true when $\mathcal{R}_k < 0$ is replaced by $K^T \mathcal{R}_k K < 0$, where $W_{i+2} = 0$ and K can be obtained by the method in Remark 3 of [19].

Remark 6: It can be seen from (38) in Theorem 1 that for a given k , the computation complexity of (38) depends polynomially on the state dimension n . Obviously, the computation burden grows rapidly with the number of state dimension. So, when applying this approach to large-scale networks, it may

induce a huge computational burden and lead to significant memory management problems. Thus, in authors' opinion, this approach is more suitable for system with small state dimension. Besides, it should be mentioned that the stability criterion is less conservative while the number of decision variables also increases. So, in the future, it is a worthwhile work to find a k which gives the best trade-off between maximum allowable delay and number of decision variables.

IV. ROBUST STABILITY CRITERIA

In this section, based on the results in Theorem 1, we will discuss the stability analysis problem of systems described by

$$\dot{x}(t) - \sum_{i=1}^2 B_i \dot{x}(t - h_i) = \sum_{i=0}^2 (A_i + \Delta A_i) x(t - h_i), \quad (47)$$

where $A_0, B_i, A_i, i = 1, 2$, are the same as that in (2), and

$$[\Delta A_0 \ \Delta A_1 \ \Delta A_2] = E_0 F [\check{A}_0 \ \check{A}_1 \ \check{A}_2], \quad (48)$$

in which $E_0 \in \mathbb{R}^{n \times u}$, $\check{A}_i \in \mathbb{R}^{v \times n}, i = 0, 1, 2$, are constant matrices, and $F \in \mathbb{R}^{u \times v}$ represents the norm bounded uncertainty satisfying

$$F^T F \leq I_v. \quad (49)$$

Now, we give a stability criterion guaranteeing system (47) to be asymptotically stable. For convenience presentation, we denote, as shown in the equation at the bottom of the next page, and

$$\Delta A_{k11} = \begin{bmatrix} \Delta \check{A}_{k11} & 0_{k^2 n \times g n} \\ 0_{g n \times (k+1)^2 n} & 0_{g n} \end{bmatrix} \in \mathbb{R}^{r n \times c n},$$

in which $g = k(k+2), r = k^2 + g, c = (k+1)^2 + g$ and

$$\begin{aligned} \Delta \check{A}_{k11} &= L_k \otimes L_k \otimes \Delta A_0 + R_k \otimes L_k \otimes \Delta A_1 \\ &\quad + L_k \otimes R_k \otimes \Delta A_2. \end{aligned}$$

Then a hierarchical stability criterion can be stated as follows.

Theorem 2: Let Assumption 1 be satisfied and $lh_2 \neq ih_1$ for $l \in \mathbb{I}[1, p]$ with p satisfying (13) and $i \in q_l$ where q_l is defined in (24) and (25). Then the perturbed neutral delay system (47), with any F satisfying (49), is asymptotically stable, if there exist two positive definite matrices $P, M \in \mathbb{S}^{(g+k^2)n}$, four positive definite matrices $Q_1, Q_2, W_1, W_2 \in \mathbb{S}^{k(1+k)n}$, a positive definite matrix $J \in \mathbb{S}^r$, and k positive definite matrices $W_{i+2}, i \in q_l, l \in \mathbb{I}[1, p]$ such that

$$\begin{bmatrix} \mathcal{R}_k + \Upsilon_k & 0_{cn \times rn} & 0_{cn \times ru} \\ 0_{rn \times cn} & -M & P(I_r \otimes E_0) \\ 0_{ru \times cn} & (I_r \otimes E_0^T) P & -J \otimes I_u \end{bmatrix} < 0, \quad (50)$$

where $\Upsilon_k = T_k^T (J \otimes I_v) T_k + \mathcal{B}_k^T M \mathcal{B}_k$.

Proof: Let

$$Y_1 = \begin{bmatrix} 0_{cn \times ru} \\ P(I_r \otimes E_0) \end{bmatrix}, \quad Y_2 = \begin{bmatrix} T_k^T \\ 0_{rn \times rv} \end{bmatrix}.$$

By Schur complements, we have from (50) that

$$\begin{aligned} 0 &> \begin{bmatrix} \mathcal{R}_k + \mathcal{B}_k^T M \mathcal{B}_k & 0_{cn \times rn} \\ 0_{rn \times cn} & -M \end{bmatrix} \\ &\quad + Y_1 (J^{-1} \otimes I_u) Y_1^T + Y_2 (J \otimes I_v) Y_2^T, \end{aligned} \quad (51)$$

where \mathcal{R}_k is defined in (38). By applying the inequality

$$XY + Y^T X^T \leq XZX^T + Y^T Z^{-1} Y, \quad (52)$$

with X, Y being real matrices and $Z > 0$, it follows from (49) and (51) that

$$\begin{aligned} &\begin{bmatrix} \mathcal{R}_k + \mathcal{B}_k^T M \mathcal{B}_k & \Delta A_{k11}^T P \\ P \Delta A_{k11} & -M \end{bmatrix} \\ &= \Theta + \begin{bmatrix} 0_{cn} & \Delta A_{k11}^T P \\ P \Delta A_{k11} & 0_{rn} \end{bmatrix} \\ &= \Theta + Y_1 (I_r \otimes F) Y_2^T + Y_2 (I_r \otimes F^T) Y_1^T \\ &\leq \Theta + Y_1 (J^{-1} \otimes I_u) Y_1^T \\ &\quad + Y_2 (I_r \otimes F^T) (J \otimes I_u) (I_r \otimes F) Y_2^T \\ &\leq \Theta + Y_1 (J^{-1} \otimes I_u) Y_1^T + Y_2 (J \otimes I_v) Y_2^T \\ &< 0, \end{aligned}$$

where

$$\Theta = \begin{bmatrix} \mathcal{R}_k + \mathcal{B}_k^T M \mathcal{B}_k & 0_{cn \times rn} \\ 0_{rn \times cn} & -M \end{bmatrix}.$$

By a Schur complement again, we have

$$\Delta A_{k11}^T P M^{-1} P \Delta A_{k11} + \mathcal{B}_k^T M \mathcal{B}_k + \mathcal{R}_k < 0.$$

By using the inequality (52) again, it follows from the above inequality that

$$\begin{aligned} &\Delta A_{k11}^T P \mathcal{B}_k + \mathcal{B}_k^T P \Delta A_{k11} + \mathcal{R}_k \\ &\leq \Delta A_{k11}^T P M^{-1} P \Delta A_{k11} + \mathcal{B}_k^T M \mathcal{B}_k + \mathcal{R}_k \\ &< 0. \end{aligned} \quad (53)$$

By using Theorem 1, we have from (53) that system (47) is asymptotically stable. ■

V. EXAMPLES

A. Example 1

Consider system (2) with matrices (borrowed from [19])

$$\begin{aligned} A_0 &= 0.2 \begin{bmatrix} -2 & 1 \\ 0 & -1 \end{bmatrix}, \quad A_1 = 0.3 \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}, \\ A_2 &= \begin{bmatrix} 0.1 & -0.05 \\ 0.05 & 0.1 \end{bmatrix}, \quad B_1 = -0.1 \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, \quad B_2 = 0.25 I_2. \end{aligned}$$

Let $h_1 = 0.3h_2$, then we have $p = 1$ for $k = 1, 2, 3$. We use different methods to obtain the maximum value of h_2 (labeled as h_2^*) such that system (2) is asymptotically stable. From Table I, it is observed that for $k = 1$, both Theorem 1 and Corollary 1 provide much less conservative results than the stability criteria in [5], [9], and [27]; for $k = 2$, the obtained h_2^* coincides with the results in [19], and for $k = 3$, Theorem 1 and Corollary 1 are much less conservative than those results in [5], [9], [19], and [27]. It should be mentioned that the maximum allowed value of h_2 by simulation is 20.3. The obtained h_2^* by Theorem 1 with $k = 3$ is close to the simulation result.

The number of decision variables of different methods, which is associated with the computational complexity, is recorded in Table I. It is clear that although h_2^* obtained by

TABLE I
 h_2^* BY DIFFERENT METHODS

Methods	h_2^*	Number of Decision Variables
Theorem 1 [9]	0.8279	$4.5n^2 + 4.5n$
Theorem 1 [27]	0.8279	$6.5n^2 + 4.5n$
Theorem 1 [5]	1.0571	$10n^2 + 6n$
Theorem 1 [19]	8.0665	$37n^2 + 10n$
Theorem 1	$6.5345 (k=1)$	$16.5n^2 + 6.5n$
	$8.0665 (k=2)$	$154n^2 + 21n$
	$18.0524 (k=3)$	$639n^2 + 45n$
Corollary 1	$6.5345 (k=1)$	$16.5n^2 + 6.5n$
	$8.0665 (k=2)$	$52n^2 + 12n$
	$18.0524 (k=3)$	$109n^2 + 18n$

Theorem 1 and Corollary 1 is identical, less decision variables is needed in Corollary 1, which implies that Corollary 1 is more effective in this example. Besides, Theorem 1 and Corollary 1 with $k=3$ is much less conservative than those results in [5], [9], [19], and [27], while more decision variables are required. Clearly, they achieve the goal of reducing conservatism at the cost of increasing the computational complexity. To verify the derived result, for the initial condition $x_0 = [3, -10]^T$, the state responses of system (2) with different h_2 and $h_1 = 0.3h_2$ are presented in Figure 1. From Figure 1, we can see that the larger the time delay, the slower the convergence rate of the system.

B. Example 2

Assume that the system in Example 1 is subject to uncertainties, namely

$$\dot{x}(t) - \sum_{i=1}^2 B_i \dot{x}(t - h_i) = \sum_{i=0}^2 (A_i + \Delta A_i) x(t - h_i), \quad (54)$$

in which A_0, A_1, A_2, B_1, B_2 are given in Example 1, and

$$\Delta A_0 = \begin{bmatrix} a_0 & 0 \\ 0 & 0 \end{bmatrix}, \quad \Delta A_1 = \begin{bmatrix} 0 & 0 \\ a_1 & 0 \end{bmatrix}, \quad \Delta A_2 = \begin{bmatrix} 0 & a_2 \\ 0 & 0 \end{bmatrix},$$

in which $a_0 \in [-a, a], a_1 \in [-a, a], a_2 \in [-a, a]$ with a_0, a_1, a_2 being uncertainties. We choose

$$E_0 = \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \end{bmatrix}, \quad \check{A}_0 = \begin{bmatrix} a & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix},$$

$$\check{A}_1 = \begin{bmatrix} 0 & 0 \\ a & 0 \\ 0 & 0 \end{bmatrix}, \quad \check{A}_2 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0 & a \end{bmatrix},$$

which satisfy (48) with $F = \frac{1}{a} \text{diag}\{a_0, a_1, a_2\}$. Let $h_2 = 5$, $h_1 = 0.3h_2 = 1.5$, then we use Theorem 2 for different k to

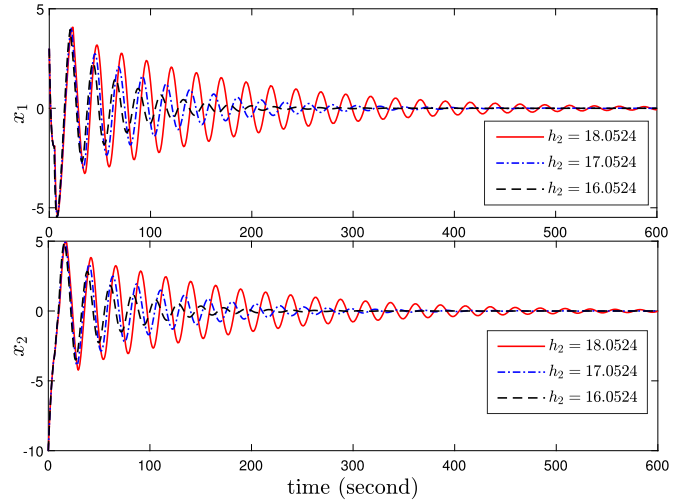


Fig. 1. State responses of system (2) with different h_2 and $h_1 = 0.3h_2$.

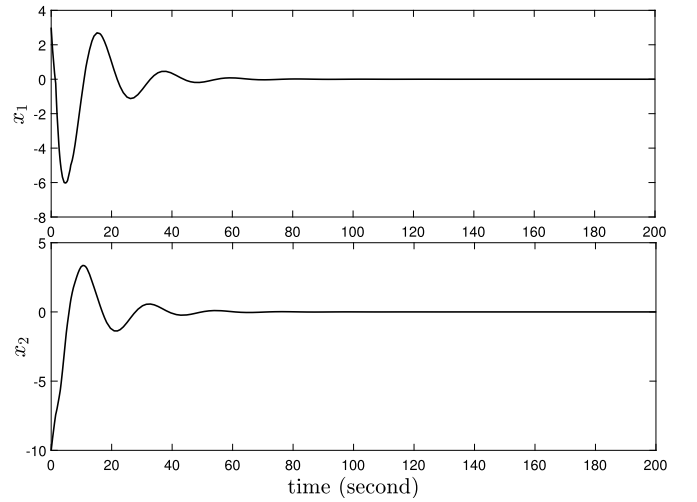


Fig. 2. State responses of system (54) with $h_2 = 5$, $h_1 = 1.5$, $a_0 = 0.0267$, $a_1 = -0.02$, and $a_2 = -0.024$.

determine the maximal value of a (denoted as $a^*(k)$) which is such that (54) is asymptotically stable. By a linear search technique, it can be found that $a^*(2) = 0.0267$, $a^*(3) = 0.0745$. Thus, the computed maximum allowable value of a is 0.0745. A simulation result is performed under the initial condition $x_0 = [3, -10]^T$ and $a_0 = 0.0267$, $a_1 = -0.02$, $a_2 = -0.024$. The state responses of system (54) are shown in Figure 2, which demonstrates the correctness of our results.

VI. CONCLUSION

The stability analysis issues of linear neutral systems with two delays have been investigated. By choosing a suitable

$$T_k = \begin{bmatrix} L_k \otimes L_k \otimes \check{A}_0 + R_k \otimes L_k \otimes \check{A}_1 + L_k \otimes R_k \otimes \check{A}_2 & 0_{k^2 v \times g n} \\ 0_{g v \times (k+1)^2 n} & 0_{g v \times g n} \end{bmatrix}$$

augmented state vector, new augmented LKFs with some delay-product-type terms have been constructed and two hierarchical stability criteria based on LMIs have been derived accordingly. Applying the resulting stability criterion to the neutral delay systems with uncertainties, a hierarchical robust stability criterion has been proposed. Examples have been provided to show that the proposed approach is very effective in reducing conservatism.

APPENDIX: THE PROOF OF PROPOSITION 1

The equality (46) can be written as

$$\begin{bmatrix} \mathcal{B}_{k11} \\ R_k \otimes L_k \otimes I_n \\ L_k \otimes R_k \otimes I_n \end{bmatrix}^T \Lambda_k \begin{bmatrix} \mathcal{B}_{k11} \\ R_k \otimes L_k \otimes I_n \\ L_k \otimes R_k \otimes I_n \end{bmatrix} + O_k < 0,$$

where

$$\begin{aligned} \Lambda_k = & \begin{bmatrix} I_{k^2n} \\ 0_{k^2n} \\ 0_{k^2n} \end{bmatrix} P \begin{bmatrix} I_{k^2} \otimes A_0^T \\ I_{k^2} \otimes (A_0 B_1 + A_1)^T \\ I_{k^2} \otimes (A_0 B_2 + A_2)^T \end{bmatrix}^T \\ & + \begin{bmatrix} I_{k^2} \otimes A_0^T \\ I_{k^2} \otimes (A_0 B_1 + A_1)^T \\ I_{k^2} \otimes (A_0 B_2 + A_2)^T \end{bmatrix} P \begin{bmatrix} I_{k^2n} \\ 0_{k^2n} \\ 0_{k^2n} \end{bmatrix}^T. \end{aligned}$$

Let $\tilde{z}_k = z_1^{[k]} \otimes z_2^{[k]} \otimes I_n$ for any $z_1, z_2 \in \mathbb{D}$. Then we have

$$\begin{bmatrix} \mathcal{B}_{k11} \\ R_k \otimes L_k \otimes I_n \\ L_k \otimes R_k \otimes I_n \end{bmatrix} \tilde{z}_{k+1} = \begin{bmatrix} I_{k^2} \otimes B(z) \\ z_1 I_{k^2n} \\ z_2 I_{k^2n} \end{bmatrix} \tilde{z}_k, \quad (55)$$

where $B(z) = I_n - z_1 B_1 - z_2 B_2$. Let

$$\tilde{B}(z) = \begin{bmatrix} I_{k^2} \otimes B(z) \\ z_1 I_{k^2n} \\ z_2 I_{k^2n} \end{bmatrix}.$$

By multiplying both side of $\tilde{\mathcal{H}}_k$ in (46) on the right by $\tilde{z}_{k+1} = z_1^{[k+1]} \otimes z_2^{[k+1]} \otimes I_n$ and on the left side by \tilde{z}_{k+1}^H , it follows from (55) that

$$\begin{aligned} \tilde{z}_{k+1}^H \tilde{\mathcal{H}}_k \tilde{z}_{k+1} &= \tilde{z}_k^H \tilde{B}^H(z) \Lambda_k \tilde{B}(z) \tilde{z}_k + (1 - |z_1|^2) \varrho_1^H Q_1 \varrho_1 \\ &\quad + (1 - |z_2|^2) \varrho_2^H Q_2 \varrho_2 \\ &< 0, \end{aligned}$$

where $\varrho_1 = z_1^{[k]} \otimes z_2^{[k+1]} \otimes I_n$, $\varrho_2 = z_1^{[k+1]} \otimes z_2^{[k]} \otimes I_n$. Since $1 - |z_i|^2 \geq 0, i = 1, 2$, it yields

$$\begin{aligned} \tilde{z}_k^H \tilde{B}^H(z) \Lambda_k \tilde{B}(z) \tilde{z}_k &= B^H(z) \tilde{z}_k^H ((I_{k^2} \otimes A(z))^H P + P (I_{k^2} \otimes A(z))) \tilde{z}_k B(z) \\ &= B^H(z) \left(A^H(z) \tilde{z}_k^H P \tilde{z}_k + \tilde{z}_k^H P \tilde{z}_k A(z) \right) B(z) \\ &< 0, \end{aligned} \quad (56)$$

where

$$\begin{aligned} A(z) &= A_0 + z_1 (A_1 + A_0 B_1) B^{-1}(z) \\ &\quad + z_2 (A_2 + A_0 B_2) B^{-1}(z). \end{aligned}$$

According to Assumption 1, it follows from (56) that

$$A^H(z) \tilde{z}_k^H P \tilde{z}_k + \tilde{z}_k^H P \tilde{z}_k A(z) < 0.$$

which further implies that

$$\alpha(A(z)) < 0, \quad (57)$$

since $\tilde{z}_k^H P \tilde{z}_k > 0$. Here, $\alpha(A(z))$ denotes the spectral abscissa of matrix $A(z)$.

In view of Assumption 1, it follows from (57) that

$$\begin{aligned} & \det \left(s (I_n - z_1 B_1 - z_2 B_2) - \sum_{i=0}^2 A_i z_i \right) \\ &= \det \left((s I_n - A_0) B(z) - \sum_{i=1}^2 z_i (A_0 B_i + A_i) \right) \\ &= \det \left(s I_n - A_0 - \sum_{i=1}^2 z_i (A_0 B_i + A_i) B^{-1}(z) \right) \det(B(z)) \\ &\neq 0, \forall (s, z_i) \in (\overline{\mathbb{C}}_+, \mathbb{D}), \end{aligned}$$

which shows that system (2) is asymptotically stable independent of h_1, h_2 .

REFERENCES

- [1] G. Antonini and P. Pepe, "Input-to-state stability analysis of partial-element equivalent-circuit models," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 56, no. 3, pp. 673–684, Mar. 2009.
- [2] A. Bellen, N. Guglielmi, and A. E. Ruehli, "Methods for linear systems of circuit delay differential equations of neutral type," *IEEE Trans. Circuits Syst. I, Fundam. Theory Appl.*, vol. 46, no. 1, pp. 212–215, Jan. 1999.
- [3] P.-A. Bliman, "A convex approach to robust stability for linear systems with uncertain scalar parameters," *SIAM J. Control Optim.*, vol. 42, no. 6, pp. 2016–2042, 2004.
- [4] P. A. Bliman, "Lyapunov equation for the stability of linear delay systems of retarded and neutral type," *IEEE Trans. Autom. Control*, vol. 47, no. 2, pp. 327–335, Feb. 2002.
- [5] D. Y. Chen, C. Y. Jin, N. H. Hu, and L. Luo, "Delay-dependent stability criteria for neutral systems with multiple time-delay systems," *Sci. Technol. Eng.*, vol. 8, no. 2, pp. 314–319, 2008.
- [6] Y. Chen and G. Chen, "Stability analysis of delayed neural networks based on a relaxed delay-product-type Lyapunov functional," *Neurocomputing*, vol. 439, pp. 340–347, Jun. 2021.
- [7] B. Du, J. Lam, Y. Zou, and Z. Shu, "Stability and stabilization for Markovian jump time-delay systems with partially unknown transition rates," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 60, no. 2, pp. 341–351, Feb. 2013.
- [8] L. Feng, L. Lombardi, P. Benner, D. Romano, and G. Antonini, "Model order reduction for delayed PEEC models with guaranteed accuracy and observed stability," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 69, no. 10, pp. 4177–4190, Oct. 2022.
- [9] E. Fridman, "New Lyapunov–Krasovskii functionals for stability of linear retarded and neutral type systems," *Syst. Control Lett.*, vol. 43, no. 4, pp. 309–319, Jul. 2001.
- [10] E. Fridman, *Introduction to Time-delay Systems: Analysis and Control*. Cham, Switzerland: Springer, 2014.
- [11] Z. Fu and S. Peng, "Input-to-state stability criteria of discrete-time time-varying impulsive switched delayed systems with applications to multi-agent systems," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 69, no. 7, pp. 3016–3025, Jul. 2022, doi: [10.1109/TCSI.2022.3163746](https://doi.org/10.1109/TCSI.2022.3163746).
- [12] M. A. Gomez, A. V. Egorov, and S. Mondie, "Necessary and sufficient stability condition by finite number of mathematical operations for time-delay systems of neutral type," *IEEE Trans. Autom. Control*, vol. 66, no. 6, pp. 2802–2808, Jun. 2021.
- [13] J. K. Hale and S. M. Lunel, *Introduction to Functional Differential Equations*. Cham, Switzerland: Springer, 1993.
- [14] Q.-L. Han, "Improved stability criteria and controller design for linear neutral systems," *Automatica*, vol. 45, no. 8, pp. 1948–1952, Aug. 2009.
- [15] Y. He, Q.-G. Wang, C. Lin, and M. Wu, "Augmented Lyapunov functional and delay-dependent stability criteria for neutral systems," *Int. J. Robust Nonlinear Control*, vol. 15, no. 18, pp. 923–933, Dec. 2005.

- [16] H. Huang, G. Feng, and X. Chen, "Stability and stabilization of Markovian jump systems with time delay via new Lyapunov functionals," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 59, no. 10, pp. 2413–2421, Oct. 2012.
- [17] F. Kong, Q. Zhu, and T. Huang, "Improved fixed-time stability lemma of discontinuous system and its application," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 69, no. 2, pp. 835–846, Feb. 2022.
- [18] T. Li and E. Tian, "Robust H_∞ control for ICPT process with coil misalignment and time delay: A sojourn-probability-based switching case," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 68, no. 12, pp. 5156–5167, Dec. 2021.
- [19] Z.-Y. Li, J. Lam, and Y. Wang, "Stability analysis of linear stochastic neutral-type time-delay systems with two delays," *Automatica*, vol. 91, pp. 179–189, May 2018.
- [20] H.-H. Lian, S.-P. Xiao, H. Yan, F. Yang, and H.-B. Zeng, "Dissipativity analysis for neural networks with time-varying delays via a delay-product-type Lyapunov functional approach," *IEEE Trans. Neural Netw. Learn. Syst.*, vol. 32, no. 3, pp. 975–984, Mar. 2021.
- [21] H. Lin, H. Zeng, and W. Wang, "New Lyapunov–Krasovskii functional for stability analysis of linear systems with time-varying delay," *J. Syst. Sci. Complex.*, vol. 34, no. 2, pp. 632–641, Apr. 2021.
- [22] X. Liu, X. Yu, and H. Xi, "Finite-time synchronization of neutral complex networks with Markovian switching based on pinning controller," *Neurocomputing*, vol. 153, pp. 148–158, Sep. 2015.
- [23] F. Long, C.-K. Zhang, Y. He, Q.-G. Wang, Z.-M. Gao, and M. Wu, "Hierarchical passivity criterion for delayed neural networks via a general delay-product-type Lyapunov–Krasovskii functional," *IEEE Trans. Neural Netw. Learn. Syst.*, early access, doi: 10.1109/TNNLS.2021.3095183.
- [24] S.-I. Niculescu and B. Brogliato, "Force measurement time-delays and contact instability phenomenon," *Eur. J. Control*, vol. 5, nos. 2–4, pp. 279–289, Jan. 1999.
- [25] X. Song, J. Man, S. Song, and C. K. Ahn, "Finite-time fault estimation and tolerant control for nonlinear interconnected distributed parameter systems with Markovian switching channels," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 69, no. 3, pp. 1347–1359, Mar. 2022.
- [26] Y. Song, W. Michiels, B. Zhou, and G.-R. Duan, "Strong stability analysis of linear delay-difference equations with multiple time delays," *IEEE Trans. Autom. Control*, vol. 66, no. 8, pp. 3741–3748, Aug. 2021.
- [27] X. M. Zhang, M. Wu, and Y. He, "Delay-dependent stability for linear neutral type systems with delay," *Acta Automatica Sinica*, vol. 30, no. 4, pp. 624–628, 2004.
- [28] Y. Zhang, D.-W. Gu, and S. Xu, "Global exponential adaptive synchronization of complex dynamical networks with neutral-type neural network nodes and stochastic disturbances," *IEEE Trans. Circuits Syst. I, Reg. Papers*, vol. 60, no. 10, pp. 2709–2718, Oct. 2013.
- [29] B. Zhou and Q. Liu, "Input delay compensation for neutral type time-delay systems," *Automatica*, vol. 78, pp. 309–319, Apr. 2017.
- [30] B. Zhou, "On strong stability and robust strong stability of linear difference equations with two delays," *Automatica*, vol. 110, Dec. 2019, Art. no. 108610.
- [31] L. Zhang, B. Wang, Y. Li, and Y. Tang, "Distributed stochastic model predictive control for cyber–physical systems with multiple state delays and probabilistic saturation constraints," *Automatica*, vol. 129, Jul. 2021, Art. no. 109574.
- [32] W. Zhang, Y. Tang, W. K. Wong, and Q. Miao, "Stochastic stability of delayed neural networks with local impulsive effects," *IEEE Trans. Neural Netw. Learn. Syst.*, vol. 26, no. 10, pp. 2336–2345, Oct. 2015.



Yunxia Song received the B.S. and M.S. degrees from the Qilu University of Technology, Jinan, China, in 2014 and 2018, respectively. She is currently pursuing the Ph.D. degree in control science and engineering with the Harbin Institute of Technology, Harbin, China. Her research interests include time-delay systems and stochastic systems.



Zhao-Yan Li received the B.S. degree from the Department of Information Engineering, North China University of Water Conservancy and Electric Power, Zhengzhou, China, in 2005, and the M.S. and Ph.D. degrees from the Department of Mathematics, Harbin Institute of Technology, China, in 2007 and 2010, respectively. She is currently a Professor and the Ph.D. Supervisor with the School of Mathematics, Harbin Institute of Technology. She has published over 40 articles in archival journals. Her research interests include stochastic systems theory and time-delay systems. She is a Reviewer of American *Mathematical Reviews* and an active reviewer for a number of journals and conferences.



Bin Zhou (Senior Member, IEEE) received the B.S., M.S., and Ph.D. degrees from the Department of Control Science and Engineering, Harbin Institute of Technology, Harbin, China, in 2004, 2006, and 2010, respectively. He is currently a Professor with the Department of Control Science and Engineering, Harbin Institute of Technology. His current research interests include time-delay systems, time-varying systems, nonlinear control, multiagent systems, and control applications in astronautical engineering. In these areas, he has published over 100 articles in archival journals. He is the author of the book *Truncated Predictor Feedback for Time-Delay Systems* (Springer-Verlag, Berlin Heidelberg, 2014). He received the "National Excellent Doctoral Dissertation Award" in 2012 from the Academic Degrees Committee of the State Council and the Ministry of Education, China. He is a Reviewer of American *Mathematical Reviews* and an active reviewer for many journals. He is currently an Associate Editor of *Automatica*, *IET Control Theory and Applications*, *Asian Journal of Control*, *Journal of System Science and Mathematical Science*, and *Control and Decision*. He is also an Associate Editor on the Conference Editorial Board of the IEEE Control Systems Society.