# SF2832 Systems Theory - Course Summary

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# **Basic**

Definition: (Basic properties of a system)

Let  $y(t) = f_{\Sigma}(u(t))$  denote a system.

- The system is relaxed if  $f_{\Sigma}(0) = 0$ .
- The system is linear if  $f_{\Sigma}(\alpha u_1(t) + \beta u_2(t)) = \alpha f_{\Sigma}(u_1(t)) + \beta f_{\Sigma}(u_2(t))$ .
- The system is memoryless if y(t) only depends on the current input u(t).
- The system is time-invariant if  $\forall T > 0$ ,  $y_T(t) = f_{\Sigma}(u_T(t))$ , where  $u_T(t) = u(t-T)$  if  $t-t_0 \ge T$  and 0 else.

Definition: (Input-output description)

The input-output description of a linear model is

$$y(t) = \int_{t_0}^t G(t, s)u(s) ds + D(t)u(t)$$

Here, G is called the impulse response. (achieved if  $u = \delta$ )

- The system is always relaxed and linear.
- The system is memoryless if G = 0.
- The system is time-invariant if G(t,s) = G(t-s).
- G is finite-dimensional if G(t,s) = H(t)K(s) for some matrices H and K.

Definition: (State space model)

A system is a state space model if it is on the form

$$\dot{x}(t) = A(t)x(t) + B(t)u(t)$$
  
$$y(t) = C(t)x(t) + D(t)u(t)$$

where  $x(t) \in \mathbb{R}^n$ ,  $u(t) \in \mathbb{R}^m$ ,  $y(t) \in \mathbb{R}^p$ .

Definition: (State transition matrix)

The state transition matrix  $\Phi(t,t_0)$  to a system is the unique solution to

$$\begin{cases} \dot{\Phi}(t, t_0) = A(t)\Phi(t, t_0) \\ \Phi(t_0, t_0) = I \end{cases}$$

 $\hookrightarrow$  Comment: Each column  $\phi_i(t)$  in  $\Phi(t, t_0)$  satisfies  $\dot{\phi}_i(t) = A(t)\phi_i(t)$  and  $\phi_i(t_0) = e_i$ .

Theorem:

If the system

$$\begin{cases} \dot{x}(t) = A(t)x(t) \\ x(t_0) = a \end{cases}$$

has the solution x(t), then  $x(t) = \Phi(t, t_0)a$ .

Definition: (Fundamental matrix)

A fundamental matrix  $\Psi(t)$  to a system is a matrix that contains linearly independent solutions  $\psi_i(t)$  in its columns that satisfy  $\dot{\psi}_i(t) = A(t)\psi_i(t)$ .

 $\hookrightarrow$  Comment: By linear independence,  $\Psi(t)$  is non-singular for all t.

Theorem:

 $\overline{\Phi(t,t_0)} = \Psi(t)\Psi^{-1}(t_0)$  for any t and any fundamental matrix  $\Psi(t)$ .

Theorem:

$$\overline{\Phi(t,s)^{-1}} = \Phi(s,t).$$

Theorem:

$$\overline{\Phi(t,\tau)\Phi(\tau,t_0)} = \Phi(t,t_0).$$

 $\rightarrow$  Intuition: We can partition the solution at  $\tau$ .

Theorem: (Solution to state space model)

The state space model

$$\begin{cases} \dot{x}(t) = A(t)x(t) + B(t)u(t) \\ x(t_0) = a \end{cases}$$

has the solution

$$x(t) = \Phi(t, t_0)a + \int_{t_0}^t \Phi(t, s)B(s)u(s) ds.$$

Theorem: (Input-output description  $\leftrightarrow$  State space model)

The input-output description agrees with a state space model where  $G(t,s) = C(t)\Phi(t,s)B(s)$ .

Definition: (Matrix exponential)

For a matrix A, we define its matrix exponential as

$$e^{At} := \sum_{k=0}^{\infty} \frac{(At)^k}{k!}$$

or equivalently,

$$e^{At} := \mathcal{L}^{-1} \left\{ (sI - A)^{-1} \right\}.$$

Theorem: (Matrix exponential properties)

For a matrix A:

- $\bullet \|e^{At}\| < \infty.$
- If A is diagonal, then  $e^{At}$  is given by element-wise exponentiation of At.
- $\bullet \ e^{PAP^{-1}t} = Pe^{At}P^{-1}.$
- If  $A_1$  and  $A_2$  commute, then  $e^{(A_1+A_2)t}=e^{A_1t}e^{A_2t}$ .
- $(e^{At})^{-1} = e^{-At}$ .
- $\frac{\mathrm{d}}{\mathrm{d}t}e^{At} = Ae^{At}$ .

Theorem:

For a time-invariant system, the state transition matrix is given by  $\Phi(t,s) = \Phi(t-s) = e^{A(t-s)}$ .

# Controllability & Observability

Definition: (Controllable)

A system

$$\dot{x} = A(t)x(t) + B(t)u(t), \qquad x(t_0) = x_0$$

is controllable if for all  $x_1$  we can find some continuous u(t) such that  $x(t_1) = x_1$  for some  $t_1$ .

Definition: (Reachable)

 $\overline{\text{Coincides}}$  with "Controllable", but with  $x_0 = 0$ .

Definition: (Null-controllable)

Coincides with "Controllable", but with  $x_1 = 0$ .

Theorem:

 $\overline{\text{``Controllable''}}\iff \text{``Reachable''}\iff \text{``Null-controllable''}\iff \forall d\in\mathbb{R}^n\ \exists u(t): \int_{t_0}^{t_1}\Phi(t,s)B(s)u(s)\ \mathrm{d}s=d.$ 

Definition: (Reachability Gramian)

The reachability Gramian to a system is

$$W(t_0, t_1) := \int_{t_0}^{t_1} \Phi(t_1, s) B(s) B^{\mathrm{T}}(S) \Phi^{\mathrm{T}}(t_1, s) \, \mathrm{d}s.$$

Theorem:

The reachability Gramian  $W(t_0, t_1)$  is symmetric and positive definite for all  $t_0 < t_1$ .

#### Theorem:

A system is controllable iff  $W(t_0, t_1)$  is nonsingular.

#### Theorem:

The state transfer from  $x(t_0) = x_0$  to  $x(t_1) = x_1$  is possible iff  $x_1 - \Phi(t_1, t_0)x_0 \in \operatorname{im} W(t_0, t_1)$ .

The solution u(t) that minimizes  $\int_{t_0}^{t_1} u^{\mathrm{T}}(s)u(s) \, \mathrm{d}s$  is given by  $u(t) = B^{\mathrm{T}}(t)\Phi^{\mathrm{T}}(t_1,t)a$  where a is a solution to  $W(t_0,t_1)a = x_1 - \Phi(t_1,t_0)x_0$ .

#### Theorem:

The rows of  $\Phi(t_1, t)B(t)$  are linearly independent over  $t \in [t_0, t_1]$  iff  $W(t_0, t_1)$  is nonsingular.

#### Definition: (Reachability matrix)

For a time-invariant system, the reachability matrix is defined as  $\Gamma := [B, AB, \dots, A^{n-1}B]$ . (A is  $n \times n$ )

#### Theorem:

For a time-invariant system, im  $W(t_0, t_1) = \operatorname{im} \Gamma$ .

### Definition: (Reachable subspace)

For a time-invariant system, we denote the reachable subspace as  $\mathcal{R} := \operatorname{im} \Gamma$ .

#### Theorem:

The reachable subspace  $\mathcal{R}$  is A-invariant, that is  $\forall x \in \mathcal{R}, Ax \in \mathcal{R}$ .

 $\hookrightarrow$  Corollary: Also  $A^kx \in \mathbb{R}$  for any  $k \in \mathbb{N}$ , and  $e^{At}x \in \mathbb{R}$  for any t.

#### Theorem:

A time-invariant can be transferred from  $x(t_0) \in \mathbb{R}$  to  $x(t_1) \in \mathbb{R}$  in time  $\varepsilon$  for any  $\varepsilon > 0$ .

#### Theorem:

For a time-invariant system, if  $x(t_0) \in \mathcal{R}$  then it is impossible for  $x(t) \notin \mathcal{R}$  for any t.

If  $x(t_0) \notin \mathcal{R}$  then it is impossible for  $x(t) \in \mathcal{R}$  for any t.

### Definition: (Observable)

A state-space model is observable if given u(t), y(t) we can reconstruct x(t).

#### Theorem:

 $\overline{\text{A state-space model}}$  is obserable iff  $C(t)\Phi(t,t_0)x_0=v(t)$  has a unique solution  $x_0$  for all v(t).

#### Definition: (Observability Gramian)

The observability Gramian to a system is

$$M(t_0, t_1) := \int_{t_0}^{t_1} \Phi^{\mathrm{T}}(t, t_0) C^{\mathrm{T}}(t) C(t) \Phi(t, t_0) \, \mathrm{d}t.$$

#### Theorem:

A system is observable iff its observability Gramian  $M(t_0, t_1)$  is nonsingular.

#### Theorem:

If  $M(t_0, t_1)$  is singular, then two initial states  $x(t_0) = a$ ,  $x(t_0) = b$  will produce the same y(t) on  $t \in [t_0, t_1]$  iff  $a - b \in \ker M(t_0, t_1)$ .

### Definition: (Observability matrix)

For a time-invariant system, the observability matrix is defined as

$$\Omega := \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix}.$$

#### Theorem:

For a time-invariant system,  $\ker M(t_0, t_1) = \ker \Omega$ .

 $\hookrightarrow$  Corollary: A time-invariant system is observable iff  $\Omega$  has full column rank. In the case where y is one-dimensional, this means that  $\Omega$  is nonsingular.

Definition: (Unobservable subspace)

For a time-invariant system, we call the unobservable subspace  $\ker \Omega$ .

Theorem:

The unobservable subspace is A-invariant.

 $\hookrightarrow$  Corollary: If  $x_0 \in \ker \Omega$ , then  $y = Ce^{At}x_0 = 0$ .

# Stability

Definition: (Equilibrium)

For a general time-invariant system  $\dot{x} = f(x)$ , we say that  $x^0$  is an equilibrium if  $f(x^0) = 0$ .

Definition: (Asymptotically stable)

The system  $\dot{x} = Ax$  is asymptotically stable if  $x(t) \to 0$  as  $t \to \infty$ , for all  $x(t_0) \in \mathbb{R}^n$ .

Definition: (Stable, Unstable)

The system  $\dot{x} = Ax$  is stable if  $||x(t)|| < \infty$ , for all  $x(t_0) \in \mathbb{R}^n$ . Otherwise it is unstable.

Theorem:

For  $\dot{x} = Ax$ :

- Asymptotically stable  $\iff e^{At} \to 0$  as  $t \to \infty$
- Stable  $\iff$   $||e^{At}|| < \infty, \forall t \ge 0$

Definition: (Stable matrix)

A matrix is stable if the real parts of all its eigenvalues are strictly negative.

Theorem:

The system  $\dot{x} = Ax$  is asymptotically stable iff A is a stable matrix.

Theorem:

For  $\dot{x} = Ax$  where A has eigenvalues  $\lambda_{\nu} = \sigma_{\nu} + i\omega_{\nu}$ :

- Asymptotically stable  $\iff \sigma_{\nu} < 0 \ \forall \nu$
- Stable  $\iff \sigma_{\nu} \leq 0 \ \forall \nu$ ; and  $J_{\nu} = [\lambda_{\nu}]$  for all  $\nu$  s.t.  $\sigma_{\nu} = 0$  (If the real part of an eigenvalue is zero, its Jordan block must be  $1 \times 1$ )
- Unstable  $\iff \exists \sigma_{\nu} > 0$

<u>Definition</u>: (Input-output stability)

The system  $\dot{x} = Ax + Bu$ , y = Cx (assumed: x(0) = 0) is input-output stable if for all bounded u(t) (assumed: bounded by 1), y(t) will be bounded.

Theorem:

For  $\dot{x} = Ax + Bu$ , y = Cx, assume that (A, B) is reachable and (C, A) is observable. Then:

Input-output stability  $\iff$  A is a stable matrix

 $\,\, \hookrightarrow \,\,$  Comment: Without the assumptions, only "  $\Longleftarrow$  " holds.

Theorem:

 $\overline{\text{For } \dot{x} = Ax + Bu}, \ y = Cx, \text{ if } A \text{ is a stable matrix then } y(t) \in L_p[0,\infty) \ \forall u(t) \in L_p[0,\infty).$ 

L Comment:  $u(t) \in L_p[0,\infty)$  means that  $\int_0^\infty ||u(t)||^p dt < \infty$ .

Definition: (Positive definite)

 $\overline{A \text{ matrix }} P$  is positive definite if  $P^{T} = P$  and all eigenvalues to P are positive.

Theorem:

For a positive definite matrix P with eigenvalues  $\lambda_1 \leq \lambda_2 \leq \ldots \leq \lambda_n$ :

$$\lambda_1 \|x\|^2 \le x^{\mathrm{T}} P x \le \lambda_n \|x\|^2$$

Definition: (Lyapunov equation)

For given matrices A and Q, the Lyapunov equation is

$$A^{\mathrm{T}}P + PA = -Q.$$

Theorem:

For any  $Q \in \mathbb{R}^{n \times n}$ , if A is a stable matrix, then

$$P = \int_0^\infty e^{A^{\mathrm{T}}t} Q e^{At} \, \mathrm{d}t$$

is the unique solution to the Lyapunov equation.

Theorem:

 $\overline{\text{Let }(C,A)}$  be observable. Then the following are equivalent:

- $A^{\mathrm{T}}P + PA = -C^{\mathrm{T}}C$  has a solution P > 0.
- $\bullet$  A is a stable matrix.

# Realizations

Definition: (Realization, dimension, transfer matrix)

Matrices (A, B, C) that satisfy  $Ce^{At}B = G(t)$  is a realization of G(t).

Alternatively,  $C(sI - A)^{-1}B = R(s)$  where  $\mathcal{L}\{G(t)\} = R(s)$  is called the transfer matrix of the system.

The dimension of A is called the dimension of the realization.

 $\hookrightarrow$  Comment: In the case where D is present, we say that  $R(s) = C(sI - A)^{-1}B + D$ .

Definition: (Minimal realization)

A realization is minimal if no other realization has lower dimension.

Theorem:

A transfer function R(s) is realizable iff it is a strictly proper rational matrix.

 $\vdash$  Comment: If R(s) is just proper, we can let  $\bar{R}(s) = R(s) - R(\infty)$  so that  $\bar{R}(s)$  is strictly proper and  $D = R(\infty)$ .

Method: (Standard reachable realization)

Let R(s) be a  $m \times k$  strictly proper rational matrix.

- Take  $\chi(s)$  be the least common denominator of all the elements of R(s). Write  $\chi(s) = s^r + a_1 s^{r-1} + \cdots + a_r$ .
- Write  $\chi(s)R(s) = N_0 + N_1s + \dots + N_{r-1}s^{r-1}$ .

Then the realization on block form is:

$$A = \begin{bmatrix} 0 & I_k & & 0 \\ & \ddots & \ddots & \\ 0 & & 0 & I_k \\ -a_r I_k & \cdots & -a_2 I_k & -a_1 I_k \end{bmatrix} (rk \times rk) \qquad B = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ I_k \end{bmatrix} (rk \times k) \qquad C = \begin{bmatrix} N_0 & \cdots & N_{r-1} \end{bmatrix} (m \times rk)$$

Definition: (Markov parameters)

For a (strictly proper) transfer matrix R(s), its Markov parameters  $R_i$  are given by the Laurent expansion  $R(s) = R_1 s^{-1} + R_2 s^{-2} + \dots$ 

Theorem:

Some useful Laurent expansions are:

$$\frac{1}{s-a} = \sum_{k=1}^{\infty} \underbrace{a^{k-1}}_{R_k} s^{-k}, \qquad \frac{1}{(s-a)^2} = \sum_{k=1}^{\infty} \underbrace{a^{k-2}(k-1)}_{R_k} s^{-k}, \qquad \frac{1}{(s-a)(s-b)} = \sum_{k=1}^{\infty} \underbrace{\frac{1}{a-b} \left(a^{k-1} - b^{k-1}\right)}_{R_k} s^{-k}$$

Method: (Standard observable realization)

Let R(s) be a  $m \times k$  strictly proper rational matrix.

- Take  $\chi(s)$  be the least common denominator of all the elements of R(s). Write  $\chi(s) = s^r + a_1 s^{r-1} + \cdots + a_r$ .
- Find the Markov parameters  $R_1, \ldots, R_r$  of R(s).

Then the realization on block form is:

$$A = \begin{bmatrix} 0 & I_m & 0 \\ & \ddots & \ddots & \\ 0 & & 0 & I_m \\ -a_r I_m & \cdots & -a_2 I_m & -a_1 I_m \end{bmatrix} (rm \times rm) \qquad B = \begin{bmatrix} R_1 \\ \vdots \\ R_r \end{bmatrix} (rm \times k) \qquad C = \begin{bmatrix} I_m & 0 & \cdots & 0 \end{bmatrix} (m \times rm)$$

#### Theorem:

For a transfer matrix R(s) with Markov parameters  $R_i$ , the matrices (A, B, C) are a realization of R(s) iff  $CA^{i-1}B = R_i$  for i = 1, 2, ...

#### Method: (Kalman decomposition)

Given a realization (A, B, C) of R(s) where A is  $n \times n$ , we want to find a new realization with lower dimension if possible.

- Let  $V_{\bar{o}r} = \operatorname{im} \Gamma \cap \ker \Omega$  (Not observable; Reachable)
- Define  $V_{or}$  by  $\operatorname{im} \Gamma = V_{\bar{o}r} \oplus V_{or}$  (Observable; Reachable)
- Define  $V_{\bar{o}\bar{r}}$  by  $\ker \Omega = V_{\bar{o}r} \oplus V_{\bar{o}\bar{r}}$  (Not observable; Not reachable)
- Define  $V_{o\bar{r}}$  by  $\mathbb{R}^n = V_{\bar{o}r} \oplus V_{or} \oplus V_{\bar{o}\bar{r}} \oplus V_{o\bar{r}}$  (Observable; Not reachable)
- Let  $T \in \mathbb{R}^{n \times n}$  be a matrix with the basis vectors in  $V_{\bar{o}r}$ ,  $V_{or}$ ,  $V_{\bar{o}\bar{r}}$ ,  $V_{o\bar{r}}$  as columns.
- Compute

$$\tilde{A} = T^{-1}AT = \begin{bmatrix} A_{11} & A_{12} & A_{13} & A_{14} \\ 0 & A_{22} & 0 & A_{24} \\ 0 & 0 & A_{33} & A_{34} \\ 0 & 0 & 0 & A_{44} \end{bmatrix}, \qquad \tilde{B} = T^{-1}B = \begin{bmatrix} B_1 \\ B_2 \\ 0 \\ 0 \end{bmatrix}, \qquad \tilde{C} = CT = \begin{bmatrix} 0 & C_2 & 0 & C_4 \end{bmatrix}$$

- The new realization is  $(A_{22}, B_2, C_2)$ .
- Intuition: Note that we build  $V_{or}$ ,  $V_{\bar{o}\bar{r}}$  and  $V_{o\bar{r}}$  as complements. There is no "unreachable (vector) space" or "observable (vector) space".
- → Comment: This new realization will be minimal!

### Definition: (Hankel matrix)

For a transfer matrix R(s) with Markov parameters  $R_i$ , the corresponding Hankel matrix is

$$H_i = \begin{bmatrix} R_1 & \cdots & R_i \\ \vdots & \ddots & \vdots \\ R_i & \cdots & R_{2i-1} \end{bmatrix}.$$

#### Theorem:

For a transfer matrix R(s), rank  $H_i = \operatorname{rank} H_r \ \forall i \geq r$  where  $r = \deg \chi$  is the degree of the least common denominator of all the elements of R(s).

#### Theorem:

A realization is minimal if and only if it is reachable and observable.

#### Theorem:

If (A, B, C) is a realization of R(s) and T is a transform, then  $(\tilde{A} = TAT^{-1}, \tilde{B} = TB, \tilde{C} = CT^{-1})$  is also a realization of R(s).  $(\tilde{x} = Tx)$ 

#### Theorem:

Given two minimal realizations (A, B, C) and  $(\tilde{A}, \tilde{B}, \tilde{C})$  there exists a transform T such that  $\tilde{A} = TAT^{-1}$ ,  $\tilde{B} = TB$ ,  $\tilde{C} = CT^{-1}$ .  $(\tilde{x} = Tx)$ 

 $\hookrightarrow$  Comment: This transform is given by  $T = (\tilde{\Omega}^T \tilde{\Omega})^{-1} \tilde{\Omega}^T \Omega = \tilde{\Gamma} \tilde{\Gamma}^T (\Gamma \tilde{\Gamma}^T)^{-1}$ .

Definition: (McMillan degree)

The McMillan degree  $\delta(R)$  of a transfer matrix R(s) is the dimension of its minimal realization.

#### Theorem:

 $\delta(R) = \operatorname{rank} H_r$  where  $r = \deg \chi$  is the degree of the least common denominator of all the elements of R(s).

#### Theorem:

Let R(s) be a transfer matrix. The minors of R(s) are the determinants of all square matrices contained in R(s) (from all combinations of row and column indices). Let  $\rho(R)$  be the least common denominator of all minors of R(s). Then,  $\delta(R) = \deg \rho(R)$ .

# Pole Placement & Observers

#### Theorem:

Let 
$$\mathcal{R} = \operatorname{im}[B, AB, \dots, A^{n-1}B]$$
 and  $\mathcal{R}_K = [B, (A+BK)B, \dots, (A+BK)^{n-1}B]$ . Then  $\mathcal{R} = \mathcal{R}_K$ .

 $\rightarrow$  Intuition: Using the controller u = Kx + v does not affect reachability.

#### Theorem:

The pole placement problem is solvable iff (A, B) is reachable. That is, for any polynomial  $\varphi(s)$  of degree n, it is possible to find a K such that  $\det(sI - (A + BK)) = \varphi(s)$ .

Ly Intuition: Solving the pole placement means using a controller u = Kx to move the poles of the system, or the eigenvalues of the matrix, to arbitrary locations.

#### Theorem:

We can assign arbitrary eigenvalues to A - LC by choosing L iff (C, A) is observable.

Ly Intuition: This corresponds to using an observer  $\hat{x}$  defined by  $\dot{\hat{x}} = A\hat{x} + Bu + L(y - C\hat{x})$ . Then, the error  $e(t) := x(t) - \hat{x}(t)$  obeys  $\dot{e}(t) = (A - LC)e(t)$ . Therefore we want to choose the eigenvalues of A - LC to have negative real parts.

# Linear Quadratic Optimal Control

#### Theorem & Notation:

Consider the linear quadratic optimal control problem

$$\min_{u} J(u) = x(t_1)^{\mathrm{T}} S x(t_1) + \int_{t_0}^{t_1} x(t)^{\mathrm{T}} Q x(t) + u(t)^{\mathrm{T}} R u(t) \, \mathrm{d}t$$
 such that 
$$\begin{cases} \dot{x} = A x + B u \\ x(t_0) = x_0 \end{cases}$$

where  $S \geq 0$ ,  $Q \geq 0$ , R > 0. Then the optimal control is given by

$$u^*(t) = -R^{-1}B^{\mathrm{T}}P(t)x(t) =: -K(t)x(t)$$

where  $K(t) = R^{-1}B^{T}P(t)$  is called the Kalman gain, and P(t) satisfies the Dynamical Riccati Equation (DRE)

$$\begin{cases} \dot{P}(t) = -A^{\mathrm{T}}P - PA + PBR^{-1}B^{\mathrm{T}}P - Q\\ P(t_1) = S \end{cases}$$

and the optimal cost is given by

$$V(x_0) = x_0^{\mathrm{T}} P(t_0) x_0$$

So, the resulting optimal x can be expressed by:

$$\dot{x}(t) = (A - BK(t))x(t) =: A_K(t)x(t) \implies x(t) = \Phi_K(t_1, t_0)x_0$$

#### Theorem

The Dynamical Riccati Equation has a unique solution P on the interval  $[t_0, t_1]$  which is positive semidefinite and bounded.

Method: (Solving DRE)

Solve

$$\frac{\mathrm{d}}{\mathrm{d}t} \begin{bmatrix} X(t) \\ Y(t) \end{bmatrix} = \begin{bmatrix} A & -BR^{-1}B^{\mathrm{T}} \\ -Q & -A^{\mathrm{T}} \end{bmatrix} \begin{bmatrix} X(t) \\ Y(t) \end{bmatrix}$$
$$\begin{bmatrix} X(t_1) \\ Y(t_1) \end{bmatrix} = \begin{bmatrix} I \\ S \end{bmatrix}$$

Then the solution to the Dynamical Riccati Equation is given by  $P = YX^{-1}$ .

Definition: (Infinite time horizon LQ)

A LQ control problem on infinite time horizon is on the form

$$\min_{u} J(u) = \int_{0}^{\infty} x(t)^{\mathrm{T}} Q x(t) + u(t)^{\mathrm{T}} R u(t) \, \mathrm{d}t$$
  
such that 
$$\begin{cases} \dot{x} = Ax + Bu \\ x(t_0) = x_0 \end{cases}$$

where  $Q \ge 0$ , R > 0.

Related to this, we have

$$\min_{u} J(u) = \int_{0}^{t_1} x(t)^{\mathrm{T}} Q x(t) + u(t)^{\mathrm{T}} R u(t) \ \mathrm{d}t$$
 such that 
$$\begin{cases} \dot{x} = Ax + Bu \\ x(t_0) = x_0 \end{cases}$$

where the optimal cost is  $x_0^T P(t_1 - 0) x_0$  (abuse of notation:  $P(t) \leftrightarrow P(t_1 - t)$ ).

Definition: (Feasible)

For an infinite time horizon LQ problem, a control u(t) is feasible if J(u) is finite.

Theorem:

For an finite time horizon LQ problem with S = 0, let  $Q = C^{T}C$ . If (C, A) is observable, then  $P(t_1 - 0) > 0 \ \forall t_1 > 0$ .

Theorem:

For an finite time horizon LQ problem with S=0, assume P>0. Then for any feasible control u, we have  $x(t)\to 0$  as  $t\to \infty$ .

Theorem:

Let (A, B, C) be a minimal realization. Then the Algebraic Riccati Equation (ARE)

$$A^{\mathrm{T}}P + PA - PBR^{-1}B^{\mathrm{T}}P + Q = 0$$

has a unique real positive definite solution P. The optimal control corresponding to the infinite time horizon LQ problem (with  $Q = C^{T}C$ ) is

$$u = -R^{-1}B^{\mathrm{T}}Px$$

and the optimal cost is given by  $x_0^{\mathrm{T}} P x_0$ .

# Kalman Filtering

Theorem: (Least squares estimate)

Let y be a vector of elements  $y_1, \ldots, y_N$  in a Hilbert space  $\mathcal{H}$ . Let  $k^T \in \mathbb{R}^N$ . Let  $x \in \mathcal{H}$ . The problem of least squares estimation, that is choosing k to minimize

$$\|x - ky\|^2$$

is given by

$$k^* = x \cdot y^{\mathrm{T}} (y \cdot y^{\mathrm{T}})^{-1} \qquad \Longrightarrow \hat{x} = k^* y = x \cdot y^{\mathrm{T}} (y \cdot y^{\mathrm{T}})^{-1} y$$

assuming that  $y \cdot y^T$  is invertible, which is true if the components of y are linearly independent.

Definition: ([y])

Let y be a vector of elements  $y_1, \ldots, y_N$  in a Hilbert space  $\mathcal{H}$ . Then we define

$$[y] := \left\{ ky : k^{\mathrm{T}} \in \mathbb{R}^{N} \right\}$$

i.e. the span of the components of y.

Theorem: (Orthogonal projection)

Let  $\mathcal{H}$  be a Hilbert space. Let  $h \in \mathcal{H}$ . Let  $M \subset \mathcal{H}$  be a subspace of  $\mathcal{H}$  consisting of estimations ("[y]"). Then,  $\hat{m} \in M$  is the best estimation of h among all points in M if and only if  $(h - \hat{m}) \cdot m = 0 \ \forall m \in M$ .

Definition:  $(E^M)$ 

Reusing notations as in the theorem above, we denote  $E^M h$  as the orthogonal projection of  $h \in \mathcal{H}$  onto the subspace  $M \subset \mathcal{H}$ , i.e.  $\hat{m} =: E^M h$ .

Theorem: (Properties of  $E^M$ )

- $E^M(\alpha_1 x_1 + \alpha_2 x_2) = \alpha_1 E^M x_1 + \alpha_2 E^M x_2$
- $\bullet$   $E^M Ax = AE^M x$
- $E^{M \oplus N} x = E^M x + E^N x$  if  $M \perp N$ .

Definition: (Kalman Filter setup)

Consider the time discrete system

$$x(t+1) = A(t)x(t) + B(t)v(t)$$
$$y(t) = C(t)x(t) + D(t)w(t)$$

where  $x \in \mathbb{R}^n$ ,  $v \in \mathbb{R}^p$ ,  $y \in \mathbb{R}^m$ ,  $w \in \mathbb{R}^q$ . Here v(t) and w(t) are white noises such that

$$E[v(t)v(s)^{\mathrm{T}}] = Q\delta_{ts}, \quad E[w(t)w(s)^{\mathrm{T}}] = R\delta_{ts}$$

where  $Q \ge 0$  and R > 0 are covariance matrices (no correlation when  $t \ne s$ ). Also, E[v(t)] = 0 and E[w(t)] = 0.

Here, x(t) are vectors of random variables in a Hilbert space  $\mathcal{H}$  with inner product  $x_i(t) \cdot x_i(t) := E[x_i(t)x_i(t)]$ .

Definition: (Things used in Kalman Filtering)

We are given previous measurements  $y_1(0), \ldots, y_m(0), \ldots, y_1(t-1), \ldots, y_m(t-1)$ . We want to find the best estimation  $\hat{x}(t)$  of x(t) based on these measurements. We minimize  $||x_i(t) - \hat{x}_i(t)||^2$  (component-wise).

- $[y_{t-1}] = H_{t-1}y := \operatorname{span} \{y_1(0), \dots, y_m(0), \dots, y_1(t-1), \dots, y_m(t-1)\}$
- $\hat{x}(t) := E^{H_{t-1}}x(t)$  or  $\hat{x}_i(t) := E^{H_{t-1}}x_i(t)$  for  $i = 1, \dots, n$
- $\tilde{y}(t) := y(t) E^{H_{t-1}}y(t)$  called the innovation of y(t)
- $[\tilde{y}] := \operatorname{span} \{\tilde{y}_1, \dots, \tilde{y}_m\}$
- $H_t(y) = H_{t-1}(y) \oplus [\tilde{y}] \text{ (note } H_{t-1}(y) \perp [\tilde{y}])$
- $\bullet \ \hat{x}_t(t) := E^{H_t(y)} x(t)$
- $\hat{x}(t+1) := E^{H_t(y)}x(t+1)$
- $P(t) := E[(x(t) \hat{x}(t))(x(t) \hat{x}(t))^{\mathrm{T}}]$
- $\bullet \ \tilde{x}(t) := x(t) \hat{x}(t)$

Definition: (Kalman Filter)

A Kalman Filter is

$$\begin{cases} \hat{x}(t+1) &= (A-AK(t)C)\hat{x}(t) + AK(t)y(t) \\ K(t) &= P(t)C^{\mathrm{T}} \left(CP(t)C^{\mathrm{T}} + DRD^{\mathrm{T}}\right)^{-1} \quad \text{(Kalman gain)} \\ P(t+1) &= AP(t)A^{\mathrm{T}} - AP(t)C^{\mathrm{T}} \left(CP(t)C^{\mathrm{T}} + DRD^{\mathrm{T}}\right)^{-1} CP(t)A^{\mathrm{T}} + BQB^{\mathrm{T}} \end{cases}$$

Given P(0) and  $\hat{x}(0)$  (usually  $\hat{x}(0) = 0$ , because it is the projection onto the empty set) we can recursively estimate the state.

# Mathematical results

## Theorem:

 $\overline{\text{Consider}} \ x^2 + bx + c.$ 

- The roots have negative real parts if and only if b > 0 and c > 0.
- The roots have nonpositive real parts if and only if  $b \ge 0$  and  $c \ge 0$ .
- If b = 0 and c > 0, then the real parts of the solutions are exactly 0.
- If b > 0 and c = 0, then one solution is 0 while the other is strictly real negative.

### Theorem:

Consider  $x^n + a_1 x^{n-1} + \cdots + a_n$ . If all roots have negative real parts, then  $a_1, \ldots, a_n > 0$ .