

MODULE 06 HOMEWORK

3/12/18

EN.525.718.81.SP18 MULTIRATE SIGNAL PROCESSING

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Problem 1

Using the Parks-McClellan algorithm, design a lowpass decimation filter for $M=5$ with the following specifications: $\omega_p=\pi/M-\Delta\omega$, $\omega_s=\pi/M+\Delta\omega$, $R_p=1\text{dB}$, $R_s=30\text{dB}$ with $\Delta\omega=\pi/100$. Plot the impulse response and the frequency response magnitude. Show clearly that your design meets all specifications by plotting the specification template on the frequency response graph.

```
% Decimation Parameter
M = 5;

% Transition band
dw = pi/100;

% Passband/Stopband parameters
wp = pi/M - dw;
ws = pi/M + dw;

% Passband/Stopband ripple parameters
Rp = 1;    %dB
Rs = 30;   %dB

Rp_linear = min([(1-10^(-Rp/20)) (10^(Rp/20)-1)]);

% Compute estimated filter order and other design params using firpmord()
[N,fo,mo,w] = firpmord([wp/pi ws/pi], [1 0], [Rp_linear 10^(-Rs/20)]);

% Increase filter order by 3 since original order was under estimated.
% Compute FIR filter impulse response
N = N + 3;
b = firpm(N,fo,mo,w);

% Compute Frequency Response
[H,w] = freqz(b);

% Plot Impulse Response
figure(1);
stem(0:N,b,'MarkerFaceColor','b');
ylim([min(b)-0.1 max(b)+0.1]);
title('Impulse Response');
xlabel('Samples (n)'); ylabel('Amplitude');

% Plot Magnitude Frequency Response
figure(2);
plot(w/pi,20*log10(abs(H)));
axis([0 1 -60 10]);
title('Magnitude Response');
xlabel('Normalized Frequency (x pi rad/sample)'); ylabel('Magnitude (dB)');
line([wp/pi wp/pi],[-60 Rp],'color','black','LineStyle','--');
line([ws/pi ws/pi],[-60 Rp],'color','black','LineStyle','--');
line([0 wp/pi],[-Rp -Rp],'color','red','LineStyle','--');
line([0 wp/pi],[Rp Rp],'color','red','LineStyle','--');
line([ws/pi 1],[-Rs -Rs],'color','red','LineStyle','--');

% Zoomed view of passband
figure(3);
plot(w/pi,20*log10(abs(H)));
axis([0 0.2 -1.01 1.01]);
title('Zoomed View of Passband');
```

```

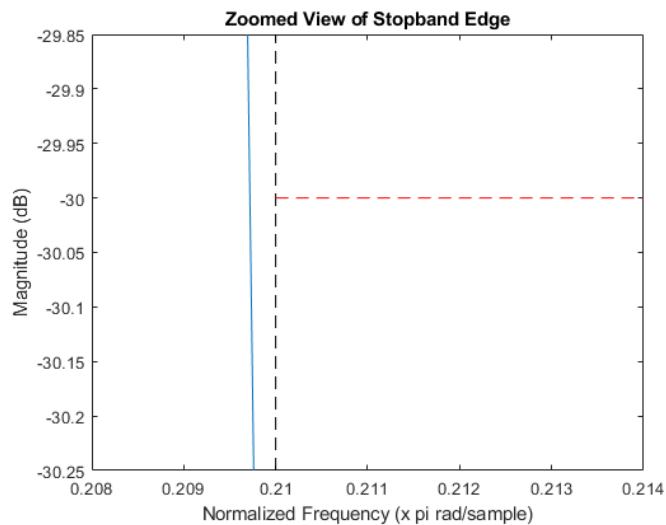
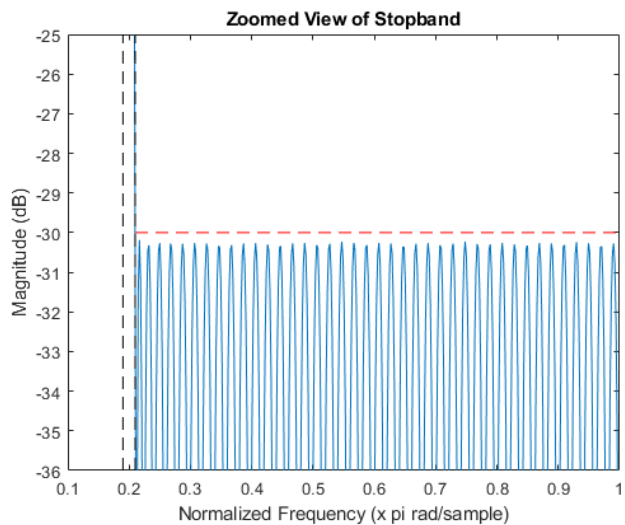
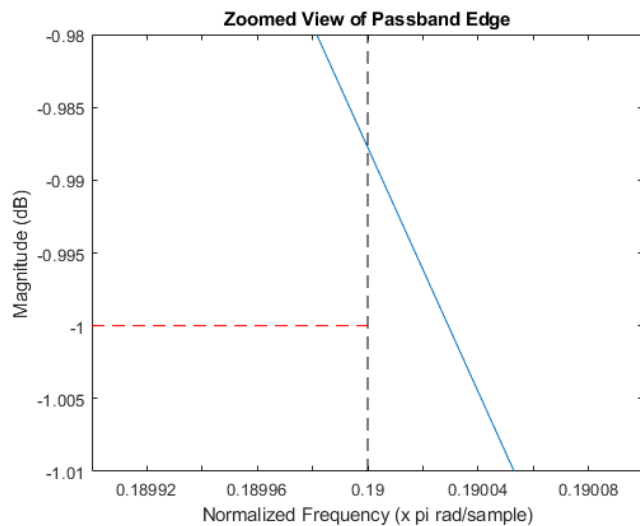
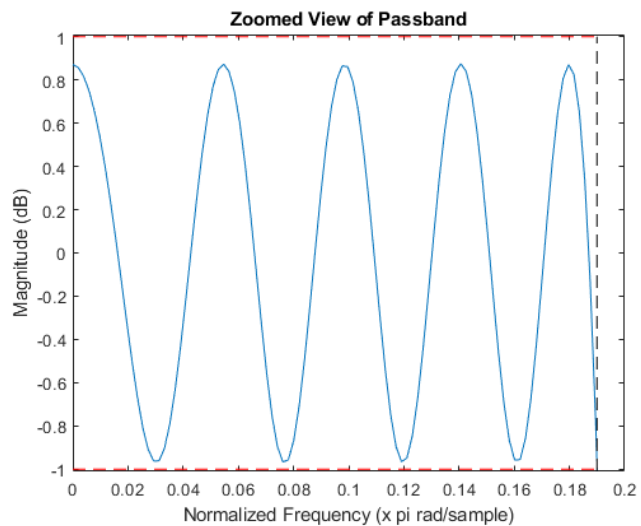
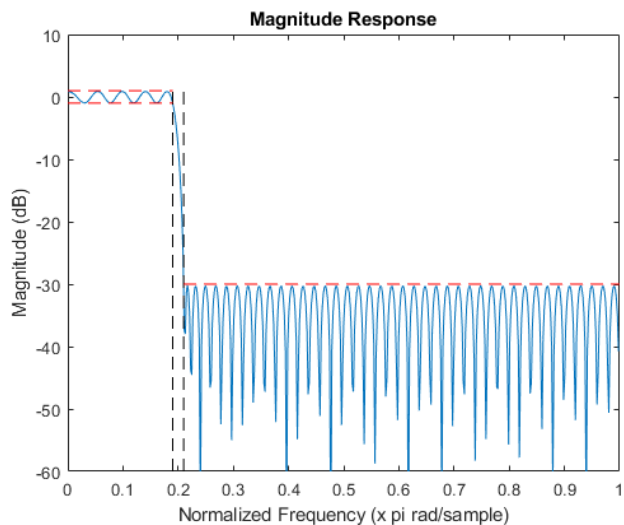
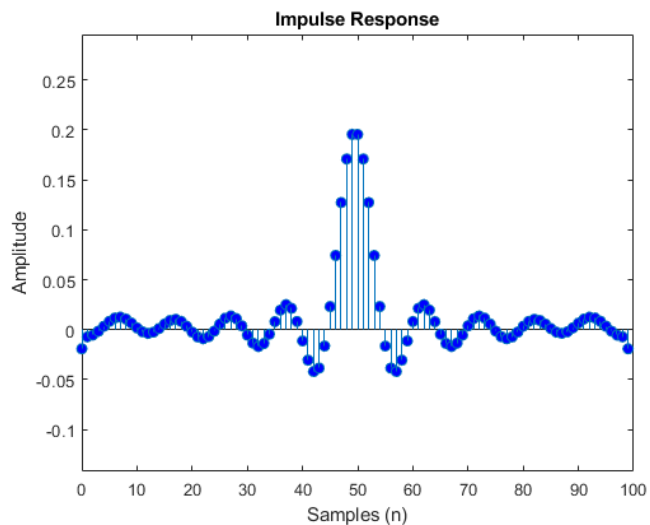
xlabel('Normalized Frequency (x pi rad/sample)'); ylabel('Magnitude (dB)');
line([wp/pi wp/pi],[-60 Rp],'color','black','LineStyle','--');
line([ws/pi ws/pi],[-60 Rp],'color','black','LineStyle','--');
line([0 wp/pi],[-Rp -Rp],'color','red','LineStyle','--');
line([0 wp/pi],[Rp Rp],'color','red','LineStyle','--');
line([ws/pi 1],[-Rs -Rs],'color','red','LineStyle','--');

% Zoomed view of passband edge
figure(4);
plot(w/pi,20*log10(abs(H)));
axis([0.1899 0.1901 -1.01 -0.980]);
title('Zoomed View of Passband Edge');
xlabel('Normalized Frequency (x pi rad/sample)'); ylabel('Magnitude (dB)');
line([wp/pi wp/pi],[-60 Rp],'color','black','LineStyle','--');
line([ws/pi ws/pi],[-60 Rp],'color','black','LineStyle','--');
line([0 wp/pi],[-Rp -Rp],'color','red','LineStyle','--');
line([0 wp/pi],[Rp Rp],'color','red','LineStyle','--');
line([ws/pi 1],[-Rs -Rs],'color','red','LineStyle','--');

% Zoomed view of stopband
figure(5);
plot(w/pi,20*log10(abs(H)));
axis([0.1 1 -36 -25]);
title('Zoomed View of Stopband');
xlabel('Normalized Frequency (x pi rad/sample)'); ylabel('Magnitude (dB)');
line([wp/pi wp/pi],[-60 Rp],'color','black','LineStyle','--');
line([ws/pi ws/pi],[-60 Rp],'color','black','LineStyle','--');
line([0 wp/pi],[-Rp -Rp],'color','red','LineStyle','--');
line([0 wp/pi],[Rp Rp],'color','red','LineStyle','--');
line([ws/pi 1],[-Rs -Rs],'color','red','LineStyle','--');

% Zoomed view of stopband edge
figure(6);
plot(w/pi,20*log10(abs(H)));
axis([0.208 0.214 -30.25 -29.85]);
title('Zoomed View of Stopband Edge');
xlabel('Normalized Frequency (x pi rad/sample)'); ylabel('Magnitude (dB)');
line([wp/pi wp/pi],[-60 Rp],'color','black','LineStyle','--');
line([ws/pi ws/pi],[-60 Rp],'color','black','LineStyle','--');
line([0 wp/pi],[-Rp -Rp],'color','red','LineStyle','--');
line([0 wp/pi],[Rp Rp],'color','red','LineStyle','--');
line([ws/pi 1],[-Rs -Rs],'color','red','LineStyle','--');

```



Problem 2

Determine the Type I polyphase representation for the filter designed in Part 1 and plot the magnitude and phase response for each polyphase section.

```
h = b;

% Create each polyphase impulse response by resampling h(n) from problem 1.
e0 = M*h(1:M:end);
e1 = M*h(2:M:end);
e2 = M*h(3:M:end);
e3 = M*h(4:M:end);
e4 = M*h(5:M:end);

% Interleaving polyphase components so that I could plot the impulse
% response and make sure it matched the direct form impulse response.
eL = zeros(1, length(e0)+length(e1)+length(e2)+length(e3)+length(e4));
eL(1:M:end) = e0;
eL(2:M:end) = e1;
eL(3:M:end) = e2;
eL(4:M:end) = e3;
eL(5:M:end) = e4;

% Compute magnitude response for each polyphase component
[E0,w0] = freqz(e0);
[E1,w1] = freqz(e1);
[E2,w2] = freqz(e2);
[E3,w3] = freqz(e3);
[E4,w4] = freqz(e4);

% Compute phase response for each polyphase component
[P0,wp0] = phasez(e0);
[P1,wp1] = phasez(e1);
[P2,wp2] = phasez(e2);
[P3,wp3] = phasez(e3);
[P4,wp4] = phasez(e4);

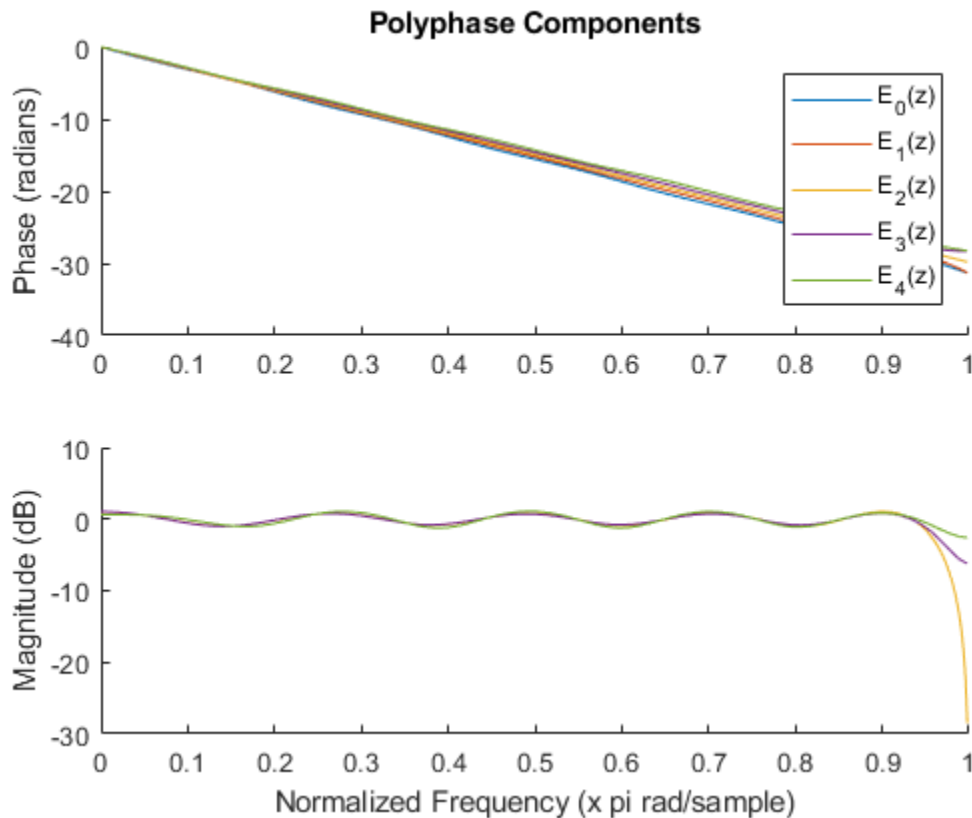
% Plot phase response for e0-e4
figure(7);
subplot(211)
hold on
plot(wp0/pi, P0);
plot(wp1/pi, P1);
plot(wp2/pi, P2);
plot(wp3/pi, P3);
plot(wp4/pi, P4);
title('Polyphase Components');
ylabel('Phase (radians)');
legend('E_0(z)', 'E_1(z)', 'E_2(z)', 'E_3(z)', 'E_4(z)');
hold off

% Plot magnitude response for e0-e4
subplot(212)
hold on
plot(w0/pi, 20*log10(abs(E0)));
plot(w1/pi, 20*log10(abs(E1)));
plot(w2/pi, 20*log10(abs(E2)));
plot(w3/pi, 20*log10(abs(E3)));
plot(w4/pi, 20*log10(abs(E4)));
```

```

xlabel('Normalized Frequency (x pi rad/sample)');
ylabel('Magnitude (dB)');
hold off

```



Problem 3

Evaluate the response of a system consisting of the decimation filter designed in Part 1 followed by a M-sample compressor for the following inputs: $x(n) = \cos(\omega_0 n)$, $\omega_0 = \pi/50$, $\omega_0 = \pi/50 + 2\pi/5$, $\omega_0 = \pi/50 + 4\pi/5$. Determine the output frequency and approximate peak amplitude (steady state) for each input frequency. Make a rough comparison of the relative amplitudes and compare with predictions based on the passband/stopband specifications for the decimation filter.

This problem clearly exemplifies the effects of aliasing that can occur during decimation. The lowpass filter from problem 1 did not have enough stop band attenuation to prevent the signals of frequency $\omega_0 = \pi/50 + 2\pi/5$ and $\omega_0 = \pi/50 + 4\pi/5$ from having a noticeable presence in the output of the system. As a result of the aliasing these two signals appeared on the output of the system as having a frequency of $\omega_0 = \pi/50$.

```

% Re-Compute Frequency response of decimation filter from problem 1
[H,w] = freqz(b);

% Input signal frequencies
w_0 = pi/50;
w_1 = w_0 + (2*pi)/5;
w_2 = w_0 + (4*pi)/5;

% Create input signals x0, x1, and x2 according to the above frequencies

```

```

n = (0:10000);
x0 = cos(w_0*n);
x1 = cos(w_1*n);
x2 = cos(w_2*n);

%***** Output Signal Predictions *****
% This small section is where I predict what the output will look like for
% each input signal.

% Interpolate H and w of Decimation filter in problem 1 and find the
% frequencies closest to w_0, w_1, and w_2 for more accurate predictions.
H_interp = interp(H,1000);
w_interp = interp(w,1000);
[m,i0] = min(abs(w_interp-w_0));
[m,i1] = min(abs(w_interp-w_1));
[m,i2] = min(abs(w_interp-w_2));

% Predict output amplitude for input signals x0, x1, and x2
A0 = abs(H_interp(i0));
A1 = abs(H_interp(i1));
A2 = abs(H_interp(i2));

% Predict phase offset for input signals x0, x1, and x2
ph0 = angle(H_interp(i0));
ph1 = angle(H_interp(i1));
ph2 = angle(H_interp(i2));

% Predict normalized frequency for input signals x0,x1, and x2
f = min(abs((-M-1):M-1)*(2*pi/M) + [-w_0, w_0]'));
f0_est = min(f)/pi;
f = min(abs((-M-1):M-1)*(2*pi/M) + [-w_1, w_1]'));
f1_est = min(f)/pi;
f = min(abs((-M-1):M-1)*(2*pi/M) + [-w_2, w_2]'));
f2_est = min(f)/pi;

% Apply amplitude, phase, and frequency predictions to signal template.
x0_est = A0*cos(w_interp(i0)*n - ph0);
x1_est = A1*cos(w_interp(i1)*n - ph1);
x2_est = A2*cos(w_interp(i2)*n - ph2);

% Downsample the prediction signals by factor M
x0_est = downsample(x0_est,M);
x1_est = downsample(x1_est,M);
x2_est = downsample(x2_est,M);

%***** End of Output Signal Predictions *****

%***** Compute Actual System Output *****

% Apply convolve filter impulse response with each input signal.
x0_filt = conv(x0,h);
x1_filt = conv(x1,h);
x2_filt = conv(x2,h);

% Keep only the steady state portions of the filtered signals. Steady
%state begins at N+1 and ends at length(filtered signal) - N+1 where
% N = filter order

```

```

ind_start = N+1;
ind_end = length(x0_filt)-(N+1);
x0_filt = x0_filt(ind_start:ind_end);
x1_filt = x1_filt(ind_start:ind_end);
x2_filt = x2_filt(ind_start:ind_end);

% Downsample the filtered signal by factor M by retrieving every Mth
% sample from each filtered signal
x0_ds = downsample(x0_filt,M);
x1_ds = downsample(x1_filt,M);
x2_ds = downsample(x2_filt,M);

% Plot filtered/downsampled signals x0_ds, x1_ds, and x2_ds as well as
% predicted outputs
figure(11);
hold on
plot(x0_ds);
plot(x0_est);
line([0 length(x0_ds)], [A0 A0], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(pi/50*n)'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_0 Decimator Output', 'x_0 Predicted Output'); hold off

figure(12);
hold on
plot(x0_ds);
plot(x0_est);
line([0 length(x0_ds)], [A0 A0], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(pi/50*n) Zoomed'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_0 Decimator Output', 'x_0 Predicted Output');
axis([10.96 11.05 0.9578 0.960]); hold off

figure(13);
hold on
plot(x1_ds);
plot(x1_est);
line([0 length(x1_ds)], [A1 A1], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(((pi/50)+(2*pi/5))*n)'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_1 Decimator Output', 'x_1 Predicted Output'); hold off

figure(14);
hold on
plot(x1_ds);
plot(x1_est);
line([0 length(x1_ds)], [A1 A1], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(((pi/50)+(2*pi/5))*n) Zoomed'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_1 Decimator Output', 'x_1 Predicted Output');
axis([450 456 0.015 0.017]); hold off

figure(15);
hold on
plot(x2_ds);
plot(x2_est);
line([0 length(x2_ds)], [A2 A2], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(((pi/50)+(4*pi/5))*n)'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_2 Decimator Output', 'x_2 Predicted Output'); hold off

figure(16);
hold on
plot(x2_ds);

```



```

plot(x2_est);
line([0 length(x2_ds)], [A2 A2], 'color', 'red', 'LineStyle', '--');
title('Decimated Signal cos(((pi/50)+(4*pi/5))*n) Zoomed'); xlabel('Samples (n)'); ylabel('Amplitude');
legend('x_2 Decimator Output', 'x_2 Predicted Output');
axis([550 560 3e-3 8e-3]); hold off

% Compute FFT of output signal x0_ds, x1_ds, and x2_ds
H0_ds = fft(x0_ds);
H1_ds = fft(x1_ds);
H2_ds = fft(x2_ds);
L = length(H0_ds);
w = linspace(0,1,L/2)/M;

% Find the peak amplitude, frequency, and phase offset of output signals
% by finding the index corresponding to the highest absolute magnitude
% of the FFT.
[m0,i0] = max(abs(H0_ds/L)*2);
[m1,i1] = max(abs(H1_ds/L)*2);
[m2,i2] = max(abs(H2_ds/L)*2);

H0_actual = abs(H0_ds(i0)/L)*2;
H1_actual = abs(H1_ds(i1)/L)*2;
H2_actual = abs(H2_ds(i2)/L)*2;

ph0_actual = angle(H0_ds(i0));
ph1_actual = angle(H1_ds(i1));
ph2_actual = angle(H2_ds(i2));

w0_actual = w(i0);
w1_actual = w(i1);
w2_actual = w(i2);

% Compare to Predictions made previously
fprintf('\nPredicted Output Signal for Input x0\n');
fprintf('Predicted:   Amplitude = %f, Frequency = %f, Phase Offset = %f\n', A0, f0_est, ph0);
fprintf('Actual:       Amplitude = %f, Frequency = %f, Phase Offset = %f\n', H0_actual, w0_actual, ph0_actual);
fprintf('\nPredicted Vs Actual Output Signal for Input x1\n');
fprintf('Predicted:   Amplitude = %f, Frequency = %f, Phase Offset = %f\n', A1, f1_est, ph1);
fprintf('Actual:       Amplitude = %f, Frequency = %f, Phase Offset = %f\n', H1_actual, w1_actual, ph1_actual);
fprintf('\nPredicted Vs Actual Output Signal for Input x2\n');
fprintf('Predicted:   Amplitude = %f, Frequency = %f, Phase Offset = %f\n', A2, f1_est, ph2);
fprintf('Actual:       Amplitude = %f, Frequency = %f, Phase Offset = %f\n', H2_actual, w2_actual, ph2_actual);

```

Predicted Vs Actual Output Signal for Input x0

Predicted: Amplitude = 0.959098, Frequency = 0.020000, Phase Offset = -3.110810

Actual: Amplitude = 0.955399, Frequency = 0.020020, Phase Offset = -3.015992

Predicted Vs Actual Output Signal for Input x1

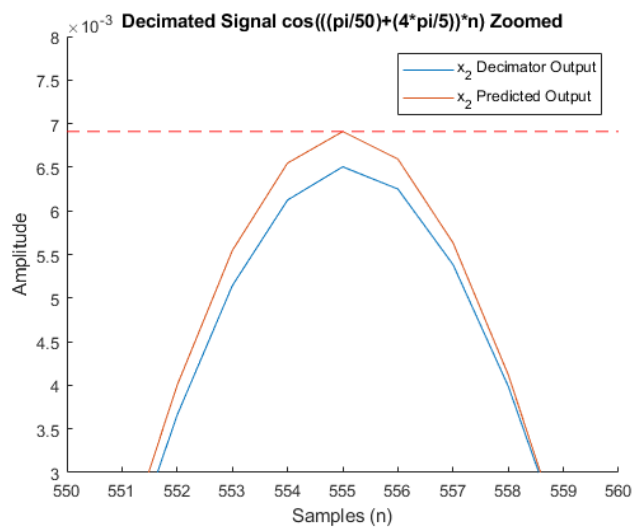
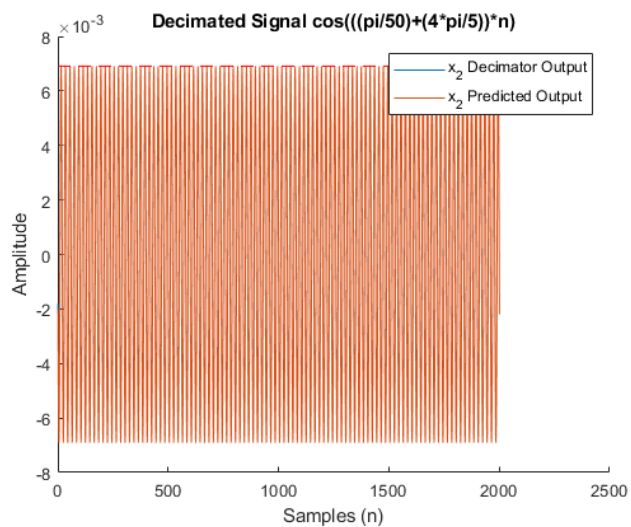
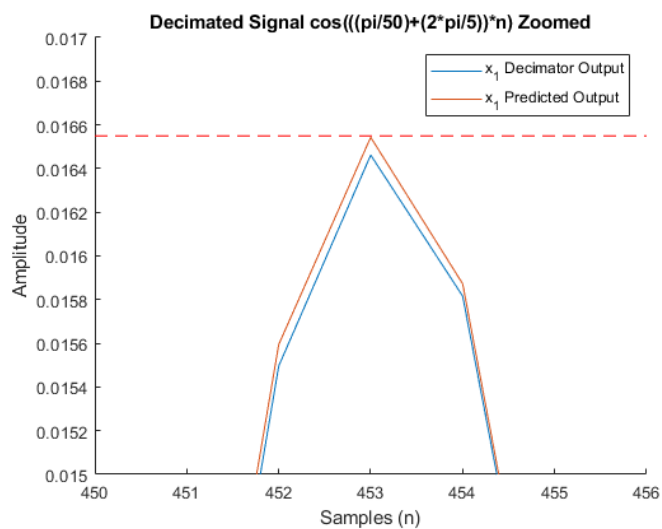
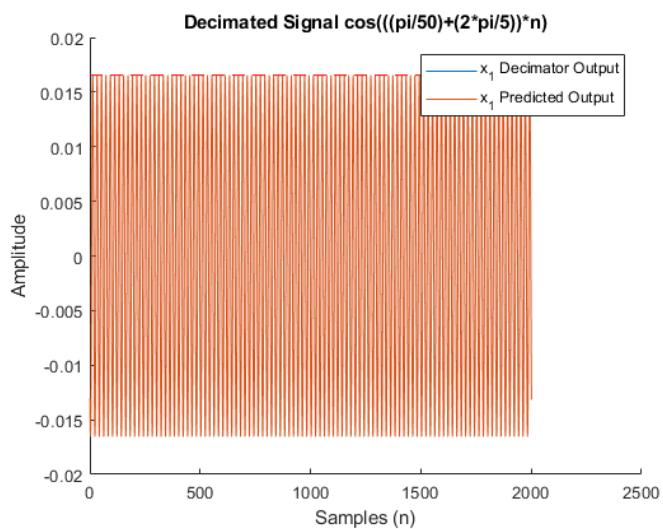
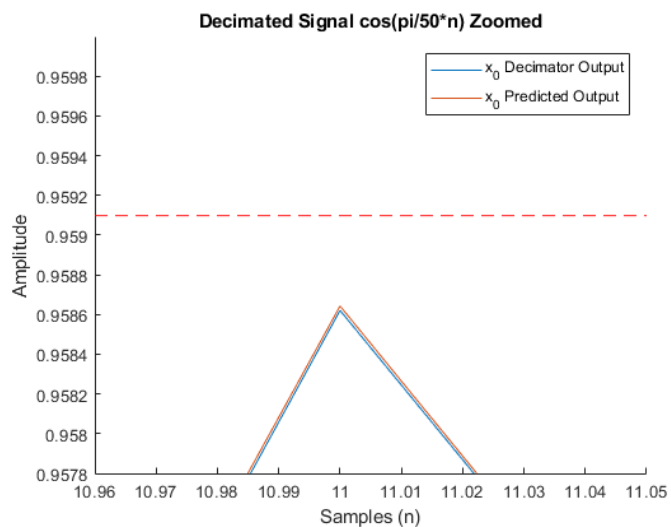
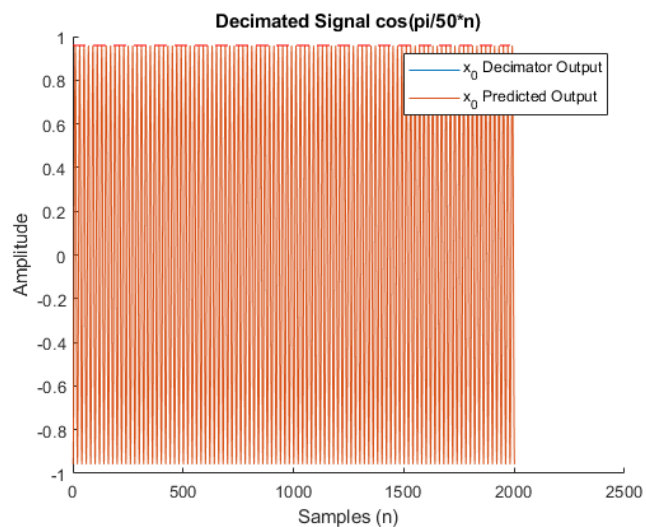
Predicted: Amplitude = 0.016549, Frequency = 0.020000, Phase Offset = -2.484833

Actual: Amplitude = 0.016403, Frequency = 0.020020, Phase Offset = 2.639107

Predicted Vs Actual Output Signal for Input x2

Predicted: Amplitude = 0.006909, Frequency = 0.020000, Phase Offset = -1.866467

Actual: Amplitude = 0.006480, Frequency = 0.020020, Phase Offset = 2.010678



Problem 4

Describe, in general terms, how your observations in Part 3 would be changed by the use of a decimation filter with a stopband decay rate of $1/f$ (as opposed to the minimax stopband characteristic of the Parks-McClellan design).

```
% If the stop band of the filter in part 1 decayed at a rate of 1/f the  
% amplitude of signals x1 and x2 would be less significant at the output  
% of the filter. Since the stopband of the filter in part 1 has a mini-max  
% characteristic all frequencies in the stopband are generally attenuated  
% the same. On the other hand, with a stopband decay rate of 1/f, higher  
% frequency input signals are attenuated more. This would reduce the  
% effects of aliasing in the output of the system.
```

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