



Module 2

Infinite Impulse Response Filter Design, Part I



Overview

- Analog filter types
 - Butterworth
 - Chebyshev I/II
 - Elliptic
 - Bessel
- Bilinear transformation
- Other transformations



Elliptic Response

- Elliptic response is equiripple in both passband and stopband
- Both poles and zeros in finite s -plane
- Filter order as well as poles and zeros are computed from elliptic integrals (see 'Lecture Notes on Elliptic Filter Design' for additional details and a design example)



Bessel Response

- Bessel filter has *maximally flat* group delay response at $\Omega=0$

- Transfer function is
$$H(s) = \frac{b_0}{\sum_{k=0}^N b_k s^k}$$

with
$$b_k = \frac{(2N-k)!}{2^{N-k} k! (N-k)!}$$



MATLAB Function Summary

- Butterworth
 - *butter, buttap, buttord*
- Chebyshev Type I
 - *cheb1ord, cheby1*
- Chebyshev Type II
 - *cheb2ord, cheby2*
- Elliptic
 - *ellip, ellipap, ellipord*
- Bessel
 - *besself, besslap*



Filter Response Comparison (1 of 4)

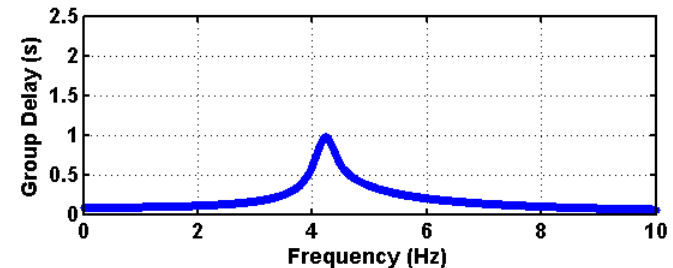
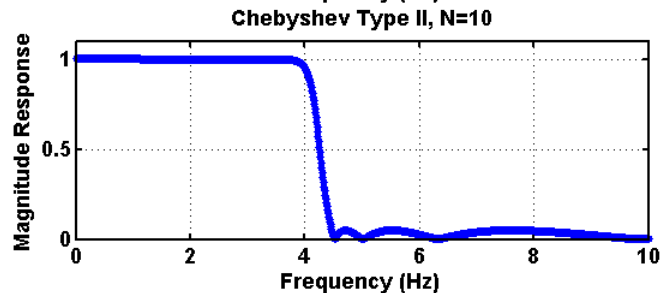
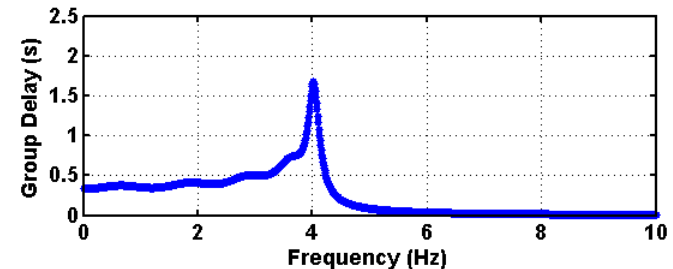
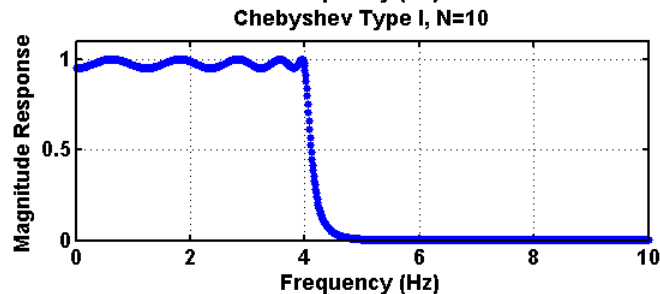
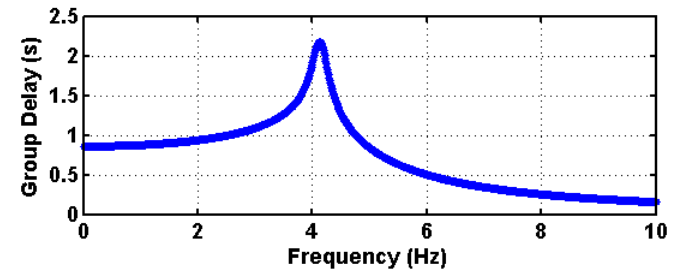
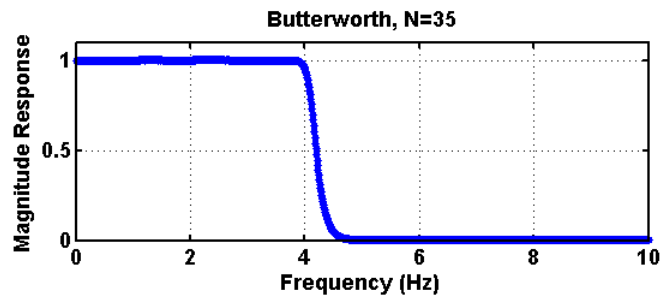
- **Butterworth** filter has maximally flat magnitude response with no ripple in either the passband or stopband
- **Chebyshev Type I** is equiripple in the passband and monotonic in the stopband
- **Chebyshev Type II** is equiripple in the stopband and monotonic in the passband



Filter Response Comparison (2 of 4)

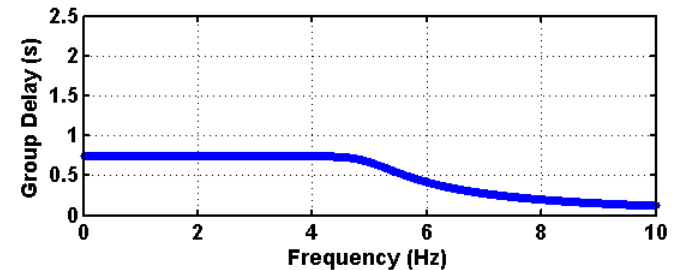
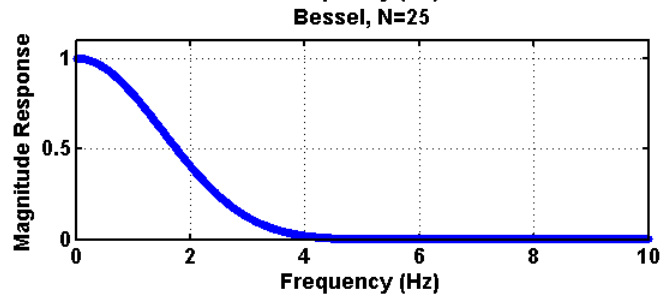
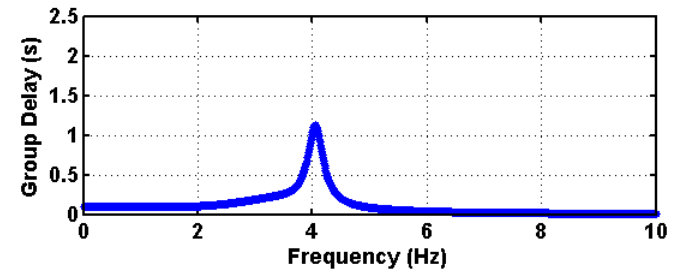
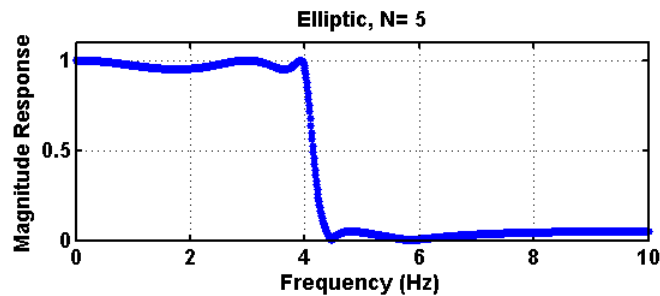
- **Elliptic** filter is most economical in the sense that a given specification can be met with a lower filter order than the other response types
 - Phase response of the elliptic filter is highly non-linear, however
- The **Bessel** filter offers nearly linear phase response (constant group delay) but has poor magnitude response

Filter Response Comparison (3 of 4)





Filter Response Comparison (4 of 4)





Bilinear Transformation

- A lowpass analog prototype filter is transformed to a lowpass discrete-time filter using the *bilinear transformation*

$$H(z) = H(s) \Big|_{s = \frac{1-z^{-1}}{1+z^{-1}}}$$

- The bilinear transformation maps the poles/zeros of $H(s)$ to $H(z)$ according to

$$z = \frac{1+s}{1-s} \quad \begin{array}{l} s = \infty \text{ maps to } z=-1, \quad s = 0 \text{ maps to } z=1 \\ s=j\Omega \text{ maps to unit circle } (|z|=1) \end{array}$$



Bilinear Transformation

- The bilinear transformation introduces *frequency warping* described by

$$\Omega = \tan\left(\frac{\omega}{2}\right)$$

- Passband/stopband frequencies must be *pre-warped* using this expression when designing the analog prototype filter



Other Transformations

- Lowpass→Lowpass, Lowpass→Highpass, Lowpass→Bandpass and Lowpass→Bandstop frequency transformations are given by

$$\begin{aligned} \text{(LP)} \quad s &= \frac{1 - Z^{-1}}{1 + Z^{-1}}, & \Omega &= \tan\left(\frac{\omega}{2}\right) \\ \text{(HP)} \quad s &= \frac{1 + Z^{-1}}{1 - Z^{-1}}, & \Omega &= -\cot\left(\frac{\omega}{2}\right) \\ \text{(BP)} \quad s &= \frac{1 - 2c_0Z^{-1} + Z^{-2}}{1 - Z^{-2}}, & \Omega &= \frac{c_0 - \cos \omega}{\sin \omega} \\ \text{(BS)} \quad s &= \frac{1 - Z^{-2}}{1 - 2c_0Z^{-1} + Z^{-2}}, & \Omega &= -\frac{\sin \omega}{c_0 - \cos \omega} \end{aligned} \tag{86}$$

where $c_0 = \cos \omega_0$, with ω_0 corresponding to the center of the bandpass or bandstop filter.

Reprinted from: 'Lecture Notes on Elliptic Filter Design' by Sophocles J. Orfanidis