1. Starting from the definition of the Chebyshev function

$$C_N(x) = f(x) = \begin{cases} cos(N \cos^{-1} x), & |x| \leq 1\\ cosh(N \cosh^{-1} x), & |x| > 1 \end{cases}$$

Show that $C_N(x)$ obeys the following recursion: $C_{N+1}(x) = 2xC_N(x) - C_{N-1}(x)$. Consider the cases $|x| \le 1$ and |x| > 1 separately.

For
$$|x| \le 1$$
:

$$\begin{split} C_{N}(x) &= \cos(N\cos^{-1}x) \\ C_{N+1}(x) &= \cos\left((N+1)\cos^{-1}x\right) = \cos(N\cos^{-1}x + \cos^{-1}x) \\ &= \cos(N\cos^{-1}x)\cos(\cos^{-1}x) - \sin(N\cos^{-1}x)\sin(\cos^{-1}x) \\ &= \cos(N\cos^{-1}x)x - \sin(N\cos^{-1}x)\sin(\cos^{-1}x) \\ C_{N-1}(x) &= \cos\left((N-1)\cos^{-1}x\right) = \cos(N\cos^{-1}x - \cos^{-1}x) \\ &= \cos(N\cos^{-1}x)\cos(\cos^{-1}x) + \sin(N\cos^{-1}x)\sin(\cos^{-1}x) \\ &= \cos(N\cos^{-1}x)\cos(\cos^{-1}x) + \sin(N\cos^{-1}x)\sin(\cos^{-1}x) \\ &= \cos(N\cos^{-1}x)x + \sin(N\cos^{-1}x)\sin(\cos^{-1}x) \\ C_{N+1}(x) + C_{N-1}(x) &= \cos(N\cos^{-1}x)x - \frac{\sin(N\cos^{-1}x)\sin(\cos^{-1}x)}{\cos^{-1}x} \\ C_{N+1}(x) + C_{N-1}(x) &= 2x * \cos(N\cos^{-1}x) \\ C_{N+1}(x) &= 2x * \cos(N\cos^{-1}x) - C_{N-1}(x) = \frac{2xC_{N}(x) - C_{N-1}(x)}{\cos^{-1}x} \end{split}$$

For |x| > 1:

$$C_N(x) = \cosh(N\cosh^{-1}x)$$
:

$$\begin{aligned} \mathsf{C}_{\mathsf{N+1}}(\mathsf{x}) &= \cosh \left((\mathsf{N}+1) \mathsf{cosh}^{-1} \mathsf{x} \right) = \cosh (\mathsf{N} \mathsf{cosh}^{-1} \mathsf{x} + \mathsf{cosh}^{-1} \mathsf{x}) \\ &= \cosh (\mathsf{N} \mathsf{cosh}^{-1} \mathsf{x}) \mathsf{cosh} (\mathsf{cosh}^{-1} \mathsf{x}) + \sinh (\mathsf{N} \mathsf{cosh}^{-1} \mathsf{x}) \mathsf{sinh} (\mathsf{cosh}^{-1} \mathsf{x}) \\ &= \cosh (\mathsf{N} \mathsf{cosh}^{-1} \mathsf{x}) \mathsf{x} + \sinh (\mathsf{N} \mathsf{cosh}^{-1} \mathsf{x}) \mathsf{sinh} (\mathsf{cosh}^{-1} \mathsf{x}) \end{aligned}$$

$$\begin{aligned} \mathsf{C}_{\mathsf{N}-1}(\mathsf{x}) &= \cosh \left((\mathsf{N}-1) \cosh^{-1} \mathsf{x} \right) = \cosh (\mathsf{N} \cosh^{-1} \mathsf{x} \, - \, \cosh^{-1} \mathsf{x}) \\ &= \cosh (\mathsf{N} \cosh^{-1} \mathsf{x}) \cosh (\cosh^{-1} \mathsf{x}) - \sinh (\mathsf{N} \cosh^{-1} \mathsf{x}) \sinh (\cosh^{-1} \mathsf{x}) \\ &= \cosh (\mathsf{N} \cosh^{-1} \mathsf{x}) \mathsf{x} - \sinh (\mathsf{N} \cosh^{-1} \mathsf{x}) \sinh (\cosh^{-1} \mathsf{x}) \end{aligned}$$

$$C_{N+1}(x) + C_{N-1}(x)$$

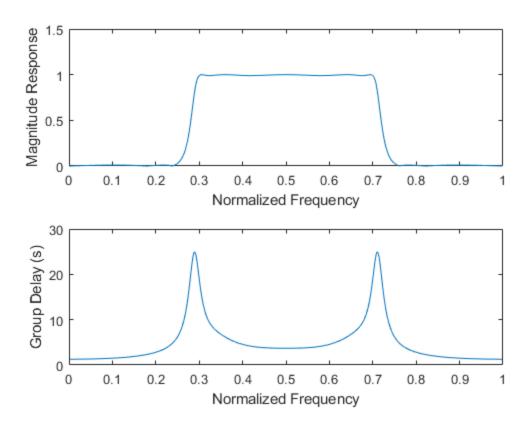
$$= cosh(Ncosh^{-1}x)x + \frac{sinh(Ncosh^{-1}x)sinh(cosh^{-1}x)}{+ cosh(Ncosh^{-1}x)x - \frac{sinh(Ncosh^{-1}x)sinh(cosh^{-1}x)}{+ cosh(Ncosh^{-1}x)sinh(cosh^{-1}x)}$$

$$C_{N+1}(x) + C_{N-1}(x) = 2x * cosh(Ncosh^{-1}x)$$

$$C_{N+1}(x) = 2x * cosh(Ncosh^{-1}x) - C_{N-1}(x) = \frac{2xC_N(x) - C_{N-1}(x)}{2xC_N(x) - C_{N-1}(x)}$$

2. Design a bandpass discrete-time Elliptic filter with the following specifications: $\omega_{s1}=0.2\pi$, $\omega_{p1}=0.3\pi$, $\omega_{p2}=0.7\pi$, $\omega_{s2}=0.8\pi$, $G_p=0.99$, $G_s=0.01$. Plot the Magnitude response and group delay for the resulting filter design. Realize the transfer function as a cascade of first-order and second-order sections with real-valued coefficients. List the coefficients for each section.

```
ws1 = 0.2*pi;
wp1 = 0.3*pi;
wp2 = 0.7*pi;
ws2 = 0.8*pi;
Gp = 0.99;
Gs = 0.01;
Rp = -20*log10(Gp);
Rs = -20*log10(Gs);
% Calculating digital Elliptic filter order and new w using normalized pass
% band and stop band frequencies.
[N,wpnew] = ellipord([wp1/pi wp2/pi],[ws1/pi ws2/pi],Rp,Rs);
% Calculate Zeros, Poles, and Gain for digital Elliptic filter
[z,p,k] = ellip(N,Rp,Rs,wpnew);
% Calculate Z domain numerator polynomial
Bz = poly(z);
% Calculate Z domain denominator polynomial
Az = poly(p);
% Adjust for gain using k computed in line 24
Bz = k*Bz;
% Calculating Magnitude Response
[H,wh] = freqz(Bz,Az);
% Calculating Group Delay
[Gpd,wg] = grpdelay(Bz,Az);
% Plotting Magnitude Response and Group Delay
figure(1)
subplot(211)
plot(wh/pi,abs(H),'-')
xlabel('Normalized Frequency'); ylabel('Magnitude Response');
subplot(212)
plot(wg/pi,Gpd,'-')
xlabel('Normalized Frequency'); ylabel('Group Delay (s)');
```



Realize transfer function as cascaded 2nd order terms

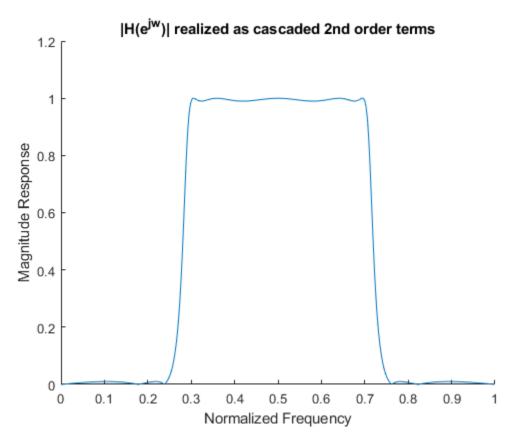
The 2nd Order terms were acquired by using MATLAB poly function on complex-conjugate pairs for variables z (zeros) and p (poles).

```
% 2nd Order Numerator coefficients indexed as follows:
% Z^2*p(1) + Z*p(2) + p(3)
z0 = [1 \ 0 \ -1];
z1 = [1 -1.6954 1];
z2 = [1 \ 1.6954 \ 1];
z3 = [1 \ 1.4663 \ 1];
z4 = [1 -1.4663 1];
% 2nd Order Denominator coefficients indexed as follows:
% Z^2*p(1) + Z*p(2) + p(3)
p0 = [1 \ 0 \ 0.3335];
p1 = [1 -0.8406 \ 0.6311];
p2 = [1 \ 0.8406 \ 0.6311];
p3 = [1 \ 1.1701 \ 0.9050];
p4 = [1 -1.1701 0.9050];
% Frequency response of each term
h0 = freqz(z0,p0);
h1 = freqz(z1,p1);
h2 = freqz(z2,p2);
h3 = freqz(z3,p3);
```

```
[h4,ww] = freqz(z4,p4);

% Cascading 2nd order Transfer functions
% Compensating for Gain
H_cascaded = k*(h0.*h1.*h2.*h3.*h4);

figure(2)
hold on
title('|H(e^{jw})| realized as cascaded 2nd order terms');
xlabel('Normalized Frequency'); ylabel('Magnitude Response');
plot(ww/pi, abs(H_cascaded),'-');
```



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