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Analysis of Systems

Transfer functions

State-space representations

- Block diagram manipulation
- Relation of transfer function to pulse response
- BIBO stability

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Transfer Functions

• The Z-transform has the same role in discrete-time systems that the Laplace transform has in the analysis of continuous-time systems.

If u(k) and e(k) are related by a difference equation, the transfer function, H(z), is defined as the ratio of the transform of the "output" u(k) of the system to the transform of the "input" e(k) of the system.

For a general difference equation:

$$u_k = -a_1u_{k-1} - a_2u_{k-2} - \cdots - a_nu_{k-n} + b_0e_k + b_1e_{k-1} + \cdots + b_me_{k-m}$$

• Take Z-transform (and use the time-shift property):

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_m z^{-m}}{1 + a_1 z^{-1} + a_2 z^{-2} + \dots + a_n z^{-n}}$$

$$H(z) = \frac{b_0 z^n + b_1 z^{n-1} + \dots + b_m z^{n-m}}{z^n + a_1 z^{n-1} + a_2 z^{n-2} + \dots + a_n} = \frac{b(z)}{a(z)}$$

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$$H(z) = \frac{z^{n-m}(b_0z^m + b_1z^{m-1} + \dots + b_m)}{z^n + a_1z^{n-1} + a_2z^{n-2} + \dots + a_n} = \frac{b(z)}{a(z)}$$

- H(z) is a rational function of a complex variable.
 - Solutions of b(z)=0 are zeros of the transfer function.
 - Solutions of a(z)=0 are poles of the transfer function.
- Special example

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State-Space Representations

- You should know how to convert an nth-order ordinary differential equation (O.D.E.) into a set of n 1st-order O.D.E.'s.
- Similarly, an nth-order difference equation can be converted into a set of n 1st-order difference equations.
- Often done because set of 1st-order equations is easier to solve.
- Consider 3rd-order system as an example:

There are <u>many</u> state-space representations of this system.

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Control Canonical Form

$$U(z) = H(z)E(z) = \frac{b(z)}{a(z)}E(z)$$

Let
$$\xi(z) = \frac{E(z)}{a(z)}$$

then
$$U(z) = b(z)\xi(z)$$
 and $a(z)\xi(z) = E(z)$

$$\xi(k+3) = e(k) - a_1 \xi(k+2) - a_2 \xi(k+1) - a_3 \xi(k)$$

$$u(k) = b_0 \xi(k+3) + b_1 \xi(k+2) + b_2 \xi(k+1) + b_3 \xi(k)$$

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Block Diagrams of Systems

• Can be drawn directly from state-space representations using only simple delay z^{-1} blocks, gains, and summers.

$$x_1(k+1) = -a_1x_1(k) - a_2x_2(k) - a_3x_3(k) + e(k)$$

 $x_2(k+1) = x_1(k)$
 $x_3(k+1) = x_2(k)$

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Control Canonical Form (and Observer Canonical Form, which
we will discuss much later . . .) is a <u>Direct Canonical</u> realization
because all the gains in the block diagram are just the
coefficients in the transfer function polynomials.

Will discuss Control and Observer forms more later . . .

General Control Canonical Form for nth-order system

$$H(z) = \frac{b_0 z^n + b_1 z^{n-1} + \dots + b_n}{z^n + a_1 z^{n-1} + a_2 z^{n-2} + \dots + a_n}$$

$$\mathbf{A}_{c} = \begin{bmatrix} \mathbf{I} & \mathbf{a}_{1} & -\mathbf{a}_{2} & \cdots & -\mathbf{a}_{n} \\ \mathbf{I} & & & & \\ \mathbf{I} & \mathbf{I} & & & \\ \mathbf{I} & & \ddots & & \\ \mathbf{I} & & \mathbf{I} & & \\ \mathbf{I} & & & \\ \mathbf{I} & & & & \\ \mathbf{I} &$$

$$\mathbf{C}_c = \begin{bmatrix} b_1 - a_1 b_0 & b_2 - a_2 b_0 & \cdots & b_n - a_n b_0 \end{bmatrix} \quad , \quad \mathbf{D}_c = b_0$$

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Non-Uniqueness of State-Space Representations

 Any nonsingular transformation of state yields another state-space representation of the same transfer function.

$$x(k+1) = Ax(k) + Be(k)$$
$$u(k) = Cx(k) + De(k)$$

Let
$$\bar{\mathbf{x}}(k) = \mathbf{T}\mathbf{x}(k)$$

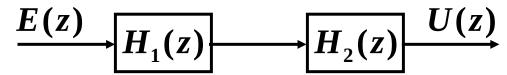
$$\overline{\mathbf{x}}(k+1) = \mathbf{TAT}^{-1}\overline{\mathbf{x}}(k) + \mathbf{TBe}(k)$$
$$u(k) = \mathbf{CT}^{-1}\overline{\mathbf{x}}(k) + \mathbf{De}(k)$$

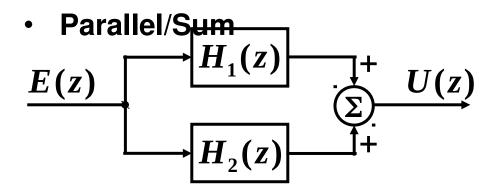
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Block Diagram Manipulation

Four rules of block diagram manipulation, analysis, and simplification:

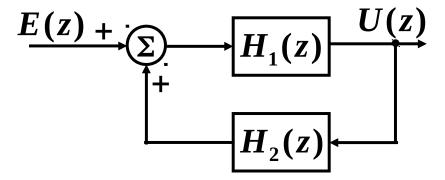
Series/Product





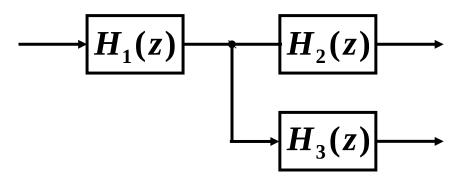
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Single loops



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Moving nodes across blocks (or blocks across nodes)



Multi-path, multi-loop block diagrams can be simplified using

- Above techniques
- Mason's rule

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Relation of Transfer Function to Pulse Response

In continuous-time systems:

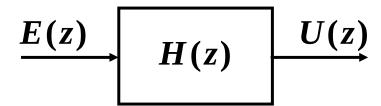
$$E(s)$$
 $H(s)$ $U(s)$

- The transfer function of a system is equal to the Laplace Transform of the impulse response.
- We find a similar result in discrete-time systems: $E(z) \longrightarrow H(z) \xrightarrow{U(z)}$

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BIBO Stability

- Internal vs. External stability
 - Internal stability is concerned with all the internal variables (states) of the system.



- External stability is concerned only with the output.
 - BIBO stability is the most common type of external stability used:
 - If for every <u>Bounded Input</u>, we have a <u>Bounded Output</u>, we say the system is BIBO stable.

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A Test for BIBO Stability

- can be given in terms of the unit-pulse response.
 - First, consider a sufficient condition.
 - Suppose the input e_k is bounded.
 - The output can be written as U(z) = H(z)E(z)

Can we find a bound on the magnitude of the output?

This condition is also necessary . . .

Consider the input

Consider the input
$$\begin{bmatrix}
-1, & h(-j) < 0 \\
e_j = \operatorname{sgn}(h_{-j}) = \operatorname{sgn}(h(-j)) = \begin{bmatrix}
0, & h(-j) = 0 \\
1, & h(-j) > 0
\end{bmatrix}$$

a bounded input . . .

Above provides requirement on unit-pulse response $oldsymbol{h}_k$.

- What are the requirements on H(z) for BIBO stability?
 - That is, what are the requirements in the frequency domain?

$$H(z) = \frac{b_0 z^n + b_1 z^{n-1} + \dots + b_m z^{n-m}}{z^n + a_1 z^{n-1} + a_2 z^{n-2} + \dots + a_n} = \frac{b(z)}{a(z)}$$

$$= c_0 + \frac{c_1 z}{z - p_1} + \frac{c_2 z}{z - p_2} + \dots + \frac{c_n z}{z - p_n}$$
 for $m \le n$

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Example

Suppose in our general difference equation

$$u_k = -a_1u_{k-1} - a_2u_{k-2} - \cdots - a_nu_{k-n} + b_0e_k + b_1e_{k-1} + \cdots + b_me_{k-m}$$

that all the coefficients are zero except a_1 and b_0 .

- What is the unit-pulse response?
- Is the system BIBO stable? Or, under what specific conditions will the system be BIBO stable?

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Jury Test

 Provides a quick way to determine by hand whether the roots of a polynomial are all inside the unit circle.

Analogous to Routh Test for continuous-time systems

 Useful for determining stability of a class of systems where certain parameters may not have fixed quantities.