

# Notes in ECEN 5448

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## Missed the first part

Talking about state space rep of a simple system.  
State space matrices are  $F=A$ ,  $G=B$ ,  $H=C$ , and  $J=D$ .  
 $J$  will be zero for any strictly proper transfer function.

you lazy slut, read the fucking slides cause you aren't paying attention.

## Real notess

a zero of a discrete-time system is a value of  $z$  such that the system output  $y=0$  even if the initial state  $x(t_0)$  and the forcing input  $u$  are nonzero.

$$\begin{aligned}x(k+1) &= \Phi x(k) + \Gamma u(k) \\y(k) &= Hx(k) + Ju(k) \rightarrow \\(zI - \Phi)X(z) - \Gamma U(z) &= 0 \\HX(z) + JU(z) &= 0 \\\begin{bmatrix} zI - \Phi & \Gamma \\ H & J \end{bmatrix} \begin{bmatrix} X(z) \\ U(z) \end{bmatrix} &= 0\end{aligned}$$

first matrix is  $(n+1) \times (n+1)$ , second matrix is  $(n+1) \times 1$ .

## Discrete Time Systems and Dynamic Response

talking about signal convergence.

$$\begin{aligned}e_4(k) &= r^k \cos(k\theta) 1(k) \\E_f(z) &= \frac{1}{2} \sum_{k=0}^{\infty} (r^k e^{jk\theta} z^{-k}) + r^k e^{-jk\theta} z^{-k} \\&= \frac{1}{2} \left( \frac{1}{1 - re^{j\theta} z^{-1}} + \frac{1}{1 - re^{-j\theta} z^{-1}} \right)\end{aligned}$$

ROC  $|re^{j\theta} z^{-1}| < 1 \rightarrow |rz^{-1}| < 1$ .

$$E_4(z) = \frac{1}{2} \left( \frac{z}{z - re^{j\theta}} + \frac{z}{z - re^{-j\theta}} \right)$$

after some work, you get:

$$E_4(z) = \frac{z(z - r \cos(\theta))}{z^2 - 2r(\cos \theta)z + r^2}, \quad |z| > |r|$$

zeros at  $0, r \cos \theta$ .

poles at  $\frac{2r \cos(\theta) \pm \sqrt{4r^2 \cos^2 \theta - 4r^2}}{2} = r \cos(\theta) \pm jr \sin(\theta) = re^{\pm j\theta}$

closer poles are to the origin, faster the decay is.

figure 4.24 is not a smith chart.

Correspondence of s plane poles and zeros with z plane poles and zeros.

$$Y(s) = \frac{s + a}{(s + a)^2 + b^2} = \frac{s + a}{s^2 + 2\zeta\omega s + \omega^2}$$

$$y(t) = e^{-at} \cos(bt)1(t)$$

poles:  $s_1, s_2 = -a \pm jb = -\zeta\omega \pm j\omega\sqrt{1 - \zeta^2}$

$$y(k) = y(t) \Big|_{t=kT} = e^{-aKT} \cos(bkT)1(kT)$$

poles:  $z = e^{(-a \pm jb)T} = e^{s_1 T}, e^{s_2 T}$

In general, if a continuous signal  $y(t)$  has Laplace transform  $Y(s)$  with poles  $s_1, s_2, \dots$ , then the sampled (discrete) signal  $y(kT)$  has z-transform  $Y(z)$  with poles  $z_1, z_2, \dots$  where:

$$z_i = e^{s_i T}$$

$$s_1, s_2 = -\zeta\omega T + j\omega\sqrt{1 - \zeta^2}$$

only interested in that till the phase is  $\pm \frac{\pi}{T}$  because of aliasing.

this all gives log spirals, which are the graph in her notes. Again, they only go to phase= $\pi$  because aliasing produces the rest of the spiral.