

# Linear Quadratic Tracking Control with Sparsity-Promoting Regularization

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# Linear Quadratic (LQ) Tracking

Master System:  $\dot{z}_m(t) = Az_m(t)$ ,  $t \geq 0$ ,  $z_m(0) = \xi_m \in \mathbb{R}^n$

Slave System:  $\dot{z}_s(t) = Az_s(t) + Bu(t)$ ,  $t \geq 0$ ,  $z_s(0) = \xi_s \in \mathbb{R}^n$

Tracking Goal:

$$\lim_{t \rightarrow \infty} \|z_s(t) - z_m(t)\| \doteq \lim_{t \rightarrow \infty} \|x(t)\| = 0$$

Tracking Error System:

$$\dot{x}(t) = Ax(t) + Bu(t), \quad x(0) = \xi_s - \xi_m = \xi \in \mathbb{R}^n, \quad t \geq 0 \quad (1)$$

Performance Index (LQ Cost)

$$J_{LQ} = \frac{1}{2} \int_0^T \left\{ x(t)^\top Q x(t) + r u(t)^2 \right\} dt, \quad Q = Q^\top \geq 0, \quad r > 0$$

$\mathcal{Q}$ : Is it possible to find a *feasible control*  $\{u(t): 0 \leq t \leq T\}$  that

*achieves tracking as well as minimizes the control effort ?*

# Recall Control Signals

▷ Minimum energy  $\Leftrightarrow \mathcal{L}^2$  norm of control signal  $\int_0^T u(t)^2 dt$

▷ Minimum fuel  $\Leftrightarrow \mathcal{L}^1$  norm of control signal  $\int_0^T |u(t)| dt$

▷ ?  $\Leftrightarrow \mathcal{L}^0$  norm of control signal  $\int_0^T |u(t)|^0 dt$

where  $|u|^0 = 1$  if  $u \neq 0$  and 0 otherwise

$\Rightarrow \mathcal{L}^0$  norm of control signal is related to the *sparsity*.

It can induce *more zero elements* compared to its dimension

$\mathcal{A}$ : *Sparsity-promoting method* is a powerful technique !

- Compressed Sensing
- Maximum Hands-off Control

# Sparse Optimization: LQ Hands-off Control

A novel LQ hands-off control problem via sparse optimization

$$\begin{aligned} \min \quad & \underbrace{\frac{1}{2} \int_0^T \left\{ x(t)^\top Q x(t) + r u(t)^2 \right\} dt}_{J_{LQ}: \text{LQ cost}} + \underbrace{\lambda \int_0^T |u(t)|^0 dt}_{\text{weighted } \mathcal{L}^0 \text{ norm}} \\ \text{s.t.} \quad & \dot{x}(t) = Ax(t) + Bu(t) \\ & x(0) = \xi, \quad x(T) = 0 \\ & |u(t)| \leq 1, \quad \forall t \in [0, T] \end{aligned} \quad (P_0)$$

*Pros:* Minimize the control inputs and achieve tracking

*Cons:*  $(P_0)$  is non-convex, non-smooth and discontinuous

# Necessary Conditions: Non-smooth Maximum Principle

*Lemma:* Let  $(x^*, u^*)$  be a local minimizer for  $(P_0)$ . Then there exist the  $\{p(t) \in \mathbb{R}^n : t \in [0, T]\}$  and Hamiltonian function

$$H^\eta(x, p, u) \triangleq p^\top (Ax + Bu) - \eta \left( \frac{1}{2} Ru^2 - \lambda |u|^0 \right) \quad (2)$$

with a scalar  $\eta \in \{0, 1\}$  satisfying the following properties:

1. the non-triviality condition for a.e.  $t \in [0, T]$ :

$$(\eta, p(t)) \neq 0$$

2. the adjoint equation for a.e.  $t \in [0, T]$  :

$$\dot{p}(t) = -A^\top p(t) + \eta Qx^*(t)$$

3. the maximum condition for a.e.  $t \in [0, T]$ :

$$u^*(t) = \arg \max_{u \in [-1, 1]} p(t)^\top Bu - \eta \left( \frac{1}{2} ru^2 - \lambda |u|^0 \right)$$

4. the constancy of the Hamiltonian for a.e.  $t \in [0, T]$  :

$$H^\eta(x^*(t), p(t), u^*(t)) = h \in \mathbb{R}$$

## Optimal Solution

*Theorem:* The optimal control  $u^*(t)$  (if it exists) satisfies

1. If  $\eta = 1$ , then

$$u^*(t) = \text{sat}(\mathbf{H}_\theta(r^{-1}B^\top p(t))), \quad (3)$$

where  $\theta = \sqrt{2\lambda/r}$ ,  $\text{sat}(\cdot)$  is the saturation function defined by

$$\text{sat}(v) \triangleq \begin{cases} -1, & \text{if } v < -1 \\ v, & \text{if } -1 \leq v \leq 1 \\ 1, & \text{if } v > 1. \end{cases}$$

and  $\mathbf{H}_\theta(\cdot)$  is the *hard-thresholding function* defined by

$$\mathbf{H}_\theta(w) \triangleq \begin{cases} 0, & \text{if } -\theta < w < \theta \\ w, & \text{if } w < -\theta \text{ or } \theta < w, \end{cases}$$

and  $\mathbf{H}_\theta(w) \in \{0, w\}$ , if  $w = \pm\theta$ .

# Numerical Computation

Convex relaxation:

$$\min J_{LQ} + \lambda \|u\|_1 \quad \text{s.t. } (P_0) \text{ constraints} \quad (P_1)$$

Time-discretization:

$(P_1) \rightarrow$  Finite-dimensional Optimization Problem  $(P_2)$

$$\begin{aligned} \min \quad & \frac{1}{2} \sum_{k=0}^{m-1} \begin{bmatrix} x_d^k \\ u_d^k \end{bmatrix}^\top \begin{bmatrix} Q_d & S_d \\ S_d^\top & R_d \end{bmatrix} \begin{bmatrix} x_d^k \\ u_d^k \end{bmatrix} + \lambda \frac{T}{m} \sum_{k=0}^{m-1} |u_d^k| \\ \text{s.t.} \quad & x_d^{k+1} = A_d x_d^k + B_d u_d^k \\ & x_d^0 = \xi, \quad x_d^m = 0 \\ & |u_d^k| \leq 1, \quad k = 0, 1, \dots, m-1 \end{aligned} \quad (P_2)$$

- Method:  $L^1$  relaxation, Time-discretization, CVX

# Robustness Analysis: Uncertainties in the initial states

We assume that the initial states are perturbed as

$$z_m(0) = z_m + \delta_m, \quad z_s(0) = z_s + \delta_s, \quad (4)$$

where  $\delta_m$  and  $\delta_s$  are uncertain vectors in  $\mathbb{R}^n$ . Define

$$\delta \triangleq \delta_s - \delta_m. \quad (5)$$

Then, the initial state  $x(0)$  in (1) is described as

$$x(0) = \xi + \delta. \quad (6)$$

*Lemma:* Let  $u^*(t)$  be the LQ hands-off control that solves  $(P_1)$  with initial state  $x(0) = \xi$ . Let  $x(t; \delta)$  denote the state variable of (1) with the optimal control  $u^*$  from the perturbed initial state in (6). Then we have

$$\|x(T; \delta)\|_{\ell^2} \leq \|e^{AT}\| \|\delta\|_{\ell^2}, \quad (7)$$

where  $\|e^{AT}\|$  is the maximum singular value of  $e^{AT}$ , and  $\|\delta\|_{\ell^2}$  is the  $\ell^2$  norm of vector  $\delta$  defined by  $\|\delta\|_{\ell^2} \triangleq \sqrt{\delta^\top \delta}$ .



# Robustness Analysis: Gap between the $A$ matrices

We consider the case when there is a gap between the  $A$  matrices in master-slave system. We denote the gap matrix by  $\Delta$ , that is, we consider the following state-space equation instead of (1):

$$\dot{x}(t; \Delta) = (A + \Delta)x(t; \Delta) + Bu(t). \quad (8)$$

*Lemma:* Let  $u^*(t)$  be the LQ hands-off control that solves  $(P_1)$  for the ideal plant (1). Then we have

$$\begin{aligned} \|x(T; \Delta)\|_{\ell^2} &\leq \alpha(\Delta)(\|\xi\|_{\ell^2} + \|B\|\|u^*\|_1) \\ &\leq \alpha(\Delta)(\|\xi\|_{\ell^2} + \|B\|\|u^*\|_0) \end{aligned} \quad (9)$$

where

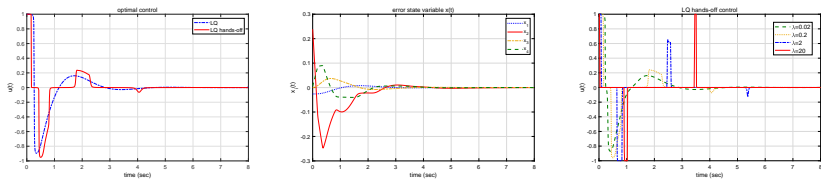
$$\alpha(\Delta) \triangleq e^{\min\{\|A\|, \|A+\Delta\|\}T} (e^{\|\Delta\|T} - 1). \quad (10)$$

# Simulations

Consider a inverted pendulum-cart system without perturbations, which can be written as a linearized model

$$\begin{cases} \ddot{\epsilon} = \frac{(M_c + m_p)g}{M_c l_p} \epsilon - \frac{1}{M_c l_p} u, \\ \ddot{q} = -\frac{m_p g}{M_c} \epsilon + \frac{1}{M_c} u, \end{cases} \quad (11)$$

where parameters are:  $M_c = 2.4$ ,  $m_p = 0.23$ ,  $l_p = 0.36$ ,  $g = 9.81$ ,  $Q = 5I$ ,  $R = 1$ , and  $x(0) = [-\pi/120, \pi/12, 0, 0]^T$ .



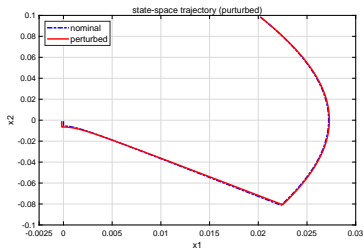
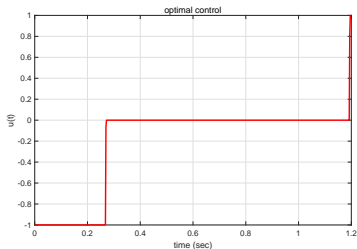
**Fig.** LQ and LQ hands-off control (left), tracking error states (middle), and different weights for LQ hands-off control (right)

# Simulations

Consider a second-order linearized inverted pendulum system with a small disturbance

$$A = \begin{bmatrix} 0 & 1 \\ \frac{m_p l_p g}{L_p} & \frac{-b}{L_p} \end{bmatrix}, \quad \Delta = \begin{bmatrix} 0 & 0 \\ 0 & 0.1 \end{bmatrix} \quad \text{and} \quad B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}.$$

The parameters are  $m_p=3$ ,  $l_p=1.5$ ,  $g=9.81$ ,  $L_p = m_p l_p^2/2$ ,  $b = 0.06$ ,  $\lambda = 1$ ,  $Q=3I$ ,  $R=1$  and  $x(0) = [0.02, 0.1]^\top$ .



**Fig.** LQ hands-off control (Left) and Phase portraits (Right) of the perturbed system with gap  $\Delta$ .

# Conclusion

- Necessary conditions for LQ hands-off control
- LQ hands-off control may not be continuous
- Future work
  - $\mathcal{L}^p$  norm approximation, Majorization-minimization algorithms
  - Improve Robustness: Probabilistic Methods, Scenario approach

THANK YOU FOR YOUR ATTENTION !