



# Some Statistical Fundamental Limits of Learning for Dynamics and Control

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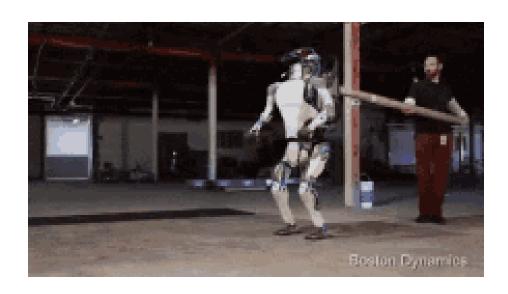
# Some Statistical Fundamental Limits of Learning for Dynamics and Control with Insights for Future Algorithms

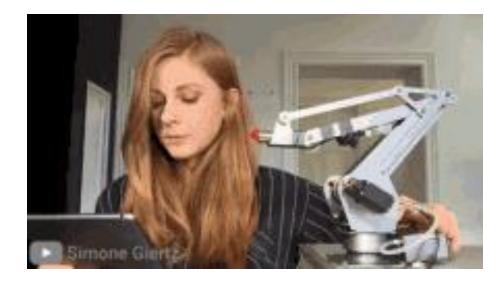
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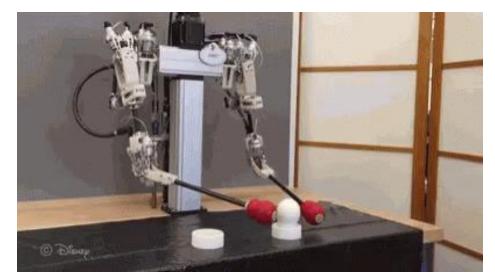
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#### Unlike ChatGPT, the physical agents are sensitive and dangerous







#### Why?

- Unlike cyberspace with discrete states, the physical world is a continuous state space.
- In continuous state space, infinitesimal error can lead to catastrophic failures in stability and safety.
- Learning for infinitesimal error might be arbitrarily hard.

Therefore, we focus on the statistical fundamental limits of agent learning in continuous state space, like system identification and learning-based control, etc.

#### **Summary of Statistical Fundamental Limits**

 Statistical Consistency: Does the algorithm converge to the ground-truth solution with respect to sample size?

 Statistical Optimality: If the algorithm is statistically consistent, does the algorithm achieve the minimax sample complexity lower bound?

 Statistical Hardness: If the algorithm is statistically optimal, does the optimal sample complexity increase moderately with the system complexity?

#### **Summary of Our Contributions**

 Chapter 2 for Statistical Consistency --- Noise Sensitivity of the Semidefinite Programs for Direct Data-Driven LQR (ACC 2025 and the extension is submitted to TAC)

 Chapter 3 for Statistical Optimality --- System Identification Under Bounded Noise: Optimal Rates Beyond Least Squares (Submitted to CDC 2025 and L-CSS)

 Chapter 4 for Statistical Hardness --- On the Hardness of Learning to Stabilize Linear Systems (CDC 2023)

## **Chapter 2**

# Noise Sensitivity of the Semidefinite Programs for Direct Data-Driven LQR

Given a linear time-invariant (LTI) system:

$$x_{t+1} = Ax_t + Bu_t + w_t,$$

where  $x_t, w_t \in \mathbb{R}^n$ ,  $u_t \in \mathbb{R}^m$ ,  $w_t = 0$  or  $w_t \sim \mathcal{N}(0, \sigma_w^2 I_n)$ , and (A, B) is controllable.

Linear quadratic regulator (LQR) problem:

$$\min_{\mathbf{u}_0, \mathbf{u}_1, \dots} \lim_{T \to \infty} \mathbb{E}\left[\frac{1}{T} \sum_{t=0}^{T} (\mathbf{x}_t^T \mathbf{Q} \mathbf{x}_t + \mathbf{u}_t^T \mathbf{R} \mathbf{u}_t)\right]$$

s.t. the previous LTI system,

where Q > 0, R > 0. When (A, B) is known, the above optimal solution is  $u_t = K_{lar}x_t$ , where

$$K_{lqr} = -(R + B^{T}PB)^{-1}B^{T}PA,$$

and P is the PSD solution of discrete-time algebraic Riccati equation.

#### When (A, B) is unknown

#### Offline Data matrices:

$$\begin{aligned} X_0 &= [x_0 \ x_1 \ ... x_{T-1}], \\ U_0 &= [u_0 \ u_1 \ ... u_{T-1}], \\ X_1 &= [x_1 \ x_2 \ ... x_T], \\ \text{where } u_t \sim N(0, \sigma_u^2 I_m). \end{aligned}$$

Consider direct data-driven (DDD) control.

#### **Certainty Equivalence (CE) DDD LQR**

(De Persis & Tesi, TAC 2019):

$$\min_{X,Y} \operatorname{trace}(QX_0Y) + \operatorname{trace}(X)$$

s.t. 
$$\begin{bmatrix} X_0 Y - I_n & X_1 Y \\ Y^T X_1^T & X_0 Y \end{bmatrix} \geqslant 0$$
$$\begin{bmatrix} X & \sqrt{R} U_0 Y \\ (\sqrt{R} U_0 Y)^T & X_0 Y \end{bmatrix} \geqslant 0.$$

Let  $Y_{ce}^*$  be an optimal solution, an estimate of  $K_{lqr}$   $K_{ce}(T) := -U_0 Y_{ce}^* (X_0 Y_{ce}^*)^{-1}$ .

**Theorem 1** (De Persis & Tesi, 2019). Let  $\operatorname{rank}\left(\begin{bmatrix} X_0 \\ U_0 \end{bmatrix}\right) = m + n$ . When  $w_t = 0$  for all t,

$$K_{ce}(T) = K_{lqr}$$
.

(CE DDD LQR is perfect for noiseless case)



**Theorem 2** (our work). Assume  $\sigma_w^2 > 0$ . When  $T \ge (m+n)(n+1) + n$ ,

$$P(K_{ce}(T) = 0_{m \times n}) = 1.$$



(CE DDD LQR is trivial for almost all noise)

**Key observation:** The following equalities hold for all  $Y_{ce}^*$  when  $\sigma_w^2 > 0$ 

$$\begin{cases} U_{0}Y_{ce}^{*} = 0_{m \times n} \\ X_{0}Y_{ce}^{*} = I_{n} \\ X_{1}Y_{ce}^{*} = 0_{n \times n} \end{cases}$$

for any  $T \ge (m + n)(n + 1) + n$ . Recall  $K_{ce} = -U_0 Y_{ce}^* (X_0 Y_{ce}^*)^{-1}$ .

$$\min_{X,Y} \operatorname{trace}(QX_{0}Y) + \operatorname{trace}(X)$$
s.t. 
$$\begin{bmatrix} X_{0}Y - I_{n} & X_{1}Y \\ Y^{T}X_{1}^{T} & X_{0}Y \end{bmatrix} \geqslant 0$$

$$\begin{bmatrix} X & \sqrt{R}U_{0}Y \\ (\sqrt{R}U_{0}Y)^{T} & X_{0}Y \end{bmatrix} \geqslant 0.$$

#### **Explanation for key observations:**

- $X_0Y I_n \ge 0$  and  $X \ge 0 \implies trace(QX_0Y) + trace(X) \ge trace(Q)$ .
- All Y with  $\begin{cases} U_0Y = \mathbf{0_{m \times n}} \\ X_0Y = \mathbf{I_n} \\ X_1Y = \mathbf{0_{n \times n}} \end{cases}$  are feasible solutions, for which trace(QX<sub>0</sub>Y) + trace(X) = trace(Q).

#### Robustness-Promoting (RP) DDD LQR (De Persis & Tesi, Automatica 2021):

$$\min_{X,Y,S} \operatorname{trace}(QX_0Y) + \operatorname{trace}(X) + \operatorname{trace}(S)$$

s.t. 
$$\begin{bmatrix} X_0 Y - I_n & X_1 Y \\ Y^T X_1^T & X_0 Y \end{bmatrix} \geqslant 0$$

$$\begin{bmatrix} X & \sqrt{R} U_0 Y \\ (\sqrt{R} U_0 Y)^T & X_0 Y \end{bmatrix} \geqslant 0$$

$$\begin{bmatrix} S & Y \\ Y^T & X_0 Y \end{bmatrix} \geqslant 0.$$

Let  $Y_{rp}^*$  denote its optimal solution,  $K_{rp}(T) := -U_0 Y_{rp}^* (X_0 Y_{rp}^*)^{-1}$ .

**Theorem 3** (our work). Assume  $w_t = 0$ ,

$$\lim_{T\to\infty} P(K_{rp}(T) = K_{lqr}) = 1.$$



(RP DDD LQR is statistically consistent for noiseless case)

**Theorem 4** (our work). Assume  $\sigma_w^2 > 0$ ,

$$\lim_{T\to\infty} P(K_{rp}(T) = \mathbf{0}_{m\times n}) = 1.$$

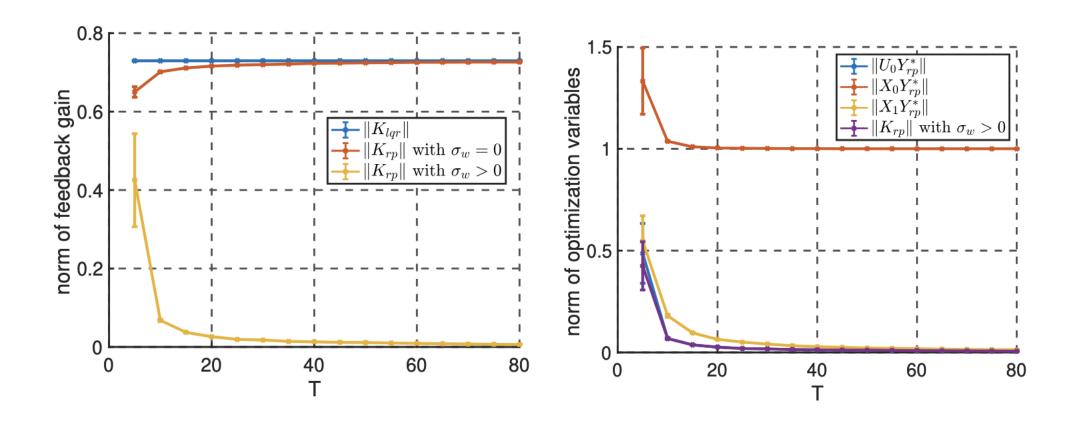


(RP DDD LQR is not statistically consistent for noisy case)

**Key Observation:** The following equalities always hold when  $\sigma_w^2 > 0$ :

$$\begin{cases} \lim_{T \to \infty} P(U_0 Y_{rp}^* = \mathbf{0}_{m \times n}) = 1 \\ \lim_{T \to \infty} P(X_0 Y_{rp}^* = \mathbf{I}_n) = 1 \\ \lim_{T \to \infty} P(X_1 Y_{rp}^* = \mathbf{0}_{n \times n}) = 1 \end{cases}$$

and recall  $K_{rp}(T) = -U_0 Y_{rp}^* (X_0 Y_{rp}^*)^{-1}$ .



Experiments for RP DDD LQR: Consider an order-2 single-input unstable system, for which  $K_{lqr} = [-0.7112 - 0.2046]$ .  $\sigma_w^2 = 1$  and  $\sigma_u^2 = 1$ .

	CE DDD LQR	RP DDD LQR
noiseless	$\mathbb{P}(\mathbf{K}_{ce} = \mathbf{K}_{ ext{lqr}}) = 1$	$\mathbf{K}_{rp}(T) \stackrel{p}{\to} \mathbf{K}_{lqr}$
$(\mathbf{w}_t = 0)$	(De Persis & Tesi, 2019)	(Theorem 3)
noisy	$\mathbb{P}(\mathbf{K}_{ce} = 0_{m \times n}) = 1$	$\mathbf{K}_{rp}(T) \stackrel{p}{\to} 0_{m \times n}$
$(\sigma_w > 0)$	(Theorem 2)	(Theorem 4)

#### Summary and Future Work

- 1. Some SDPs for DDD LQR are sensitive to noise.
- 2. Check the statistical fundamental limits when designing new DDD control algorithms.
- 3. The fundamental limits for DDD Robust Control by matrix S-lemma (Waarde et al. 2020, Waarde et al. 2023) are unclear now.

## **Chapter 3**

# System Identification Under Bounded Noise: Optimal Rates Beyond Least Squares

Consider an unknown LTI system:

$$\mathbf{x}_{t+1} = \mathbf{A}\mathbf{x}_t + \mathbf{w}_t,$$

where  $x_t, w_t \in \mathbb{R}^n$ . Assume

- $||\mathbf{w}_{\mathsf{t}}||_{\infty} < \overline{w} \ and \ \mathbf{w}_{\mathsf{t}} \ are i.i.d.$  for all  $\mathsf{t}$ ,
- $\rho(A) < 1$ ,
- and for any  $\epsilon \in [0, \overline{w}]$ , there exists C > 0, such that  $\forall j \in [n]$ ,

$$\max\left(P\left(w_t^{(j)}<-\overline{w}+\epsilon\right),P\left(w_t^{(j)}>\overline{w}-\epsilon\right)\right)< C\epsilon.$$

**Theorem 5** (our work). Given a single trajectory  $\{x_t\}_{t\in[T]}$ .  $\mathcal{F}_T$  denotes the  $\sigma$ -algebra generated by  $\{x_t\}_{t\in[T]}$  and  $\hat{A}_T$  denotes a  $\mathcal{F}_T$ -measurable estimator. Then,  $\forall \delta \in (0,1)$  and small  $\epsilon > 0$ ,

$$\sup_{\hat{A}_T} \inf_{A \in \mathbb{R}^{n \times n}} P(||\hat{A}_T - A||_2 < \epsilon) \ge 1 - \delta \quad \text{only if} \quad T > \frac{1}{4\bar{w}C\epsilon} (1 - \frac{2\delta}{n}).$$
(The best estimator can achieve  $\Omega(\frac{1}{\epsilon})$ )

The ordinary least squares (OLS) for scalar case:

$$\hat{a}_T^{OLS} = \operatorname{argmin}_a \sum_{t=1}^{T-1} ||x_{t+1} - ax_t||^2$$

**Theorem 6** (our work). Assume |a| < 1. Then,  $\forall \delta \in (0,1)$  and small  $\epsilon > 0$ ,

$$P(||\hat{a}_T^{OLS} - a||_2 < \epsilon) \ge 1 - \delta$$
 only if  $T > \Omega(\frac{1}{\epsilon^2})$ .  
(OLS only achieves  $\Omega(\frac{1}{\epsilon^2})$ )



The set membership estimator (SME) based on  $\{x_t\}_{t\in[T]}$ 

$$\mathcal{S}_{\mathrm{T}} = \Big\{ A \in \mathbb{R}^{n \times n} \colon \big| \big| x_{t+1} - A x_t \big| \big|_{\infty} \le \overline{w}, \forall t \in [T-1] \Big\}.$$

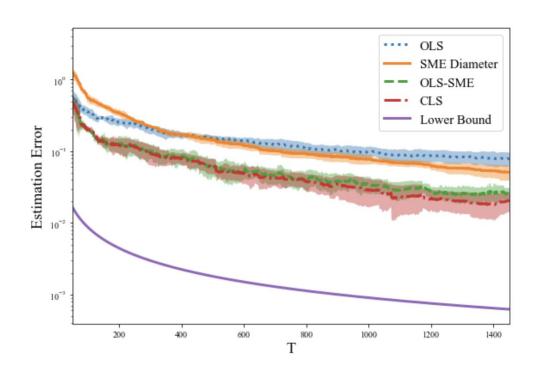
**Theorem 7** (Li & Yu et al., ICML 2024).  $\forall \delta \in (0,1)$  and small  $\epsilon > 0$ , with or without knowing  $\overline{w}$ ,

$$\forall \hat{A}_T \in \mathcal{S}_T, \ P(||\hat{A}_T - A||_2 < \epsilon) \ge 1 - \delta \quad \text{if} \quad T > \Omega(\frac{1}{\epsilon}).$$
(SME achieves the optimal  $\Omega(\frac{1}{\epsilon})$ )



#### **Summary:**

		Minimax Lower Bound	Lower Bound for OLS
Regression	Gaussian Bounded	$\Omega(1/\sqrt{T})$ (Wainwright, 2019) $\Omega(1/T)$ (Yi & Neykov, 2024)	$\begin{array}{ c c c c c }\hline \Omega(1/\sqrt{T}) \text{ (Mourtada, 2022)}\\ \Omega(1/\sqrt{T}) \text{ (Rudelson & Vershynin, 2008)}\\ \end{array}$
LTI Sys Id	Gaussian Bounded	$\Omega(1/\sqrt{T})$ (Jedra & Proutiere, 2019) $\Omega(1/T)$ (Theorem 5)	$\Omega(1/\sqrt{T})$ (Tu et al., 2024) $\Omega(1/\sqrt{T})$ (Theorem 6)



OLS: 
$$\min_{A} \sum_{t=1}^{T-1} ||x_{t+1} - Ax_t||^2$$

$$\mathsf{SME} : \mathcal{S}_{\mathrm{T}} = \left\{ A \in \mathbb{R}^{n \times n} : \left| \left| x_{t+1} - A x_{t} \right| \right|_{\infty} \leq \overline{w}, \forall t \in [T] \right\}$$

OLS-SME: 
$$\min_{A \in \mathcal{S}_T} \left| \left| A - \hat{A}_T^{OLS} \right| \right|$$

CLS: 
$$\min_{A \in S_T} \sum_{t=1}^{T-1} ||x_{t+1} - Ax_t||^2$$

OLS achieves  $0(\frac{1}{\sqrt{T}})$ 

SME, OLS-SME, and CLS achieves  $0(\frac{1}{T})$ ©

## **Chapter 4**

On the Hardness of Learning to Stabilize Linear Systems

#### Given an unknown LTI system

$$x_{t+1} = Ax_t + Bu_t + w_t,$$

where  $x_t, w_t \in \mathbb{R}^n$ ,  $u_t \in \mathbb{R}^m$ , and  $w_t \sim \mathcal{N}(0, \sigma_w^2 I_n)$ . Assume (A, B) is controllable the energy bounded input  $\mathbb{E}[\|u_t\|^2] \leq \sigma_u^2$ .

#### Consider a learning to stabilize algorithm $\pi$ that

- interacts with the above system for T units of time, and
- outputs a **linear static state feedback** controller  $\widehat{K}_T$  at time T.

We want  $\rho(A + B\widehat{K}_T) < 1$ .

**Theorem 7** (our work). Then  $\forall \delta \in [0,0.5]$ , we have

$$\sup_{\pi} \inf_{(A,B)} P((A + B \widehat{K}_T) \text{ is stable}) \ge 1 - \delta,$$

only if

$$T \ge \Omega(exp(n)).$$

(exp(n)) hardness of learning to stabilize with easy SysId)

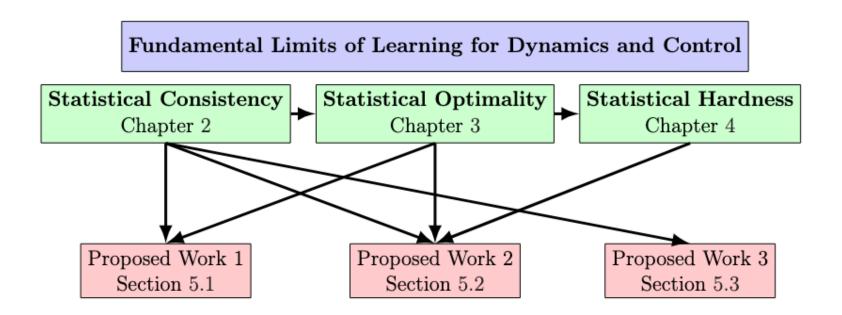
#### **Comparison** with the previous work

Tsiamis et al., COLT 2022	Hard to identify, then hard to learn to stabilize
Our Work	Hard to distinguish and hard to co-stabilize, then hard to learn to stabilize

#### **Summary**

- 1. System identification is easy, but learning to stabilize by linear static state feedback is still exponentially hard with the state dimension
- We can try other state feedback, like online switching (Proposed Work 1) or historical state feedback (Proposed Work 2)

# **Summary and Timeline**



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