

# Circuit Quantum Electrodynamics

## Superconducting platform

Covering: basic concepts, measurement techniques, implementations, qubit approaches, current trends

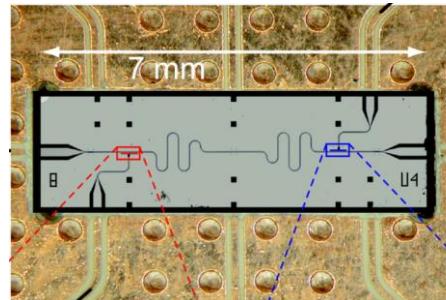
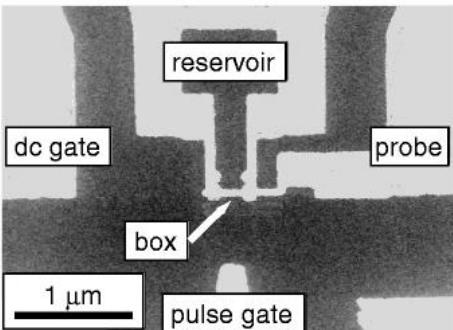
With figures and slides borrowed from

A. Wallraff (ETH-Zurich), A. Fedorov (University of Queensland), C. Eichler (FAU Erlangen).

# Goals of this part of the course

- Superconducting
  - Understand basic superconducting theory
  - What is a Josephson junction?
- Quantum
  - How do we make quantum circuits with superconductor?
- Bits
  - Superconductor as information carriers
  - How to design a qubit
  - How to operate a qubit

# History of superconducting qubits



1980: Leggett  
– Macroscopic quantum tunnelling

1985: Devoret, Martinis:  
Measurements of MQT in Josephson junction

1999: Nakamura:  
First superconducting qubit

2003: Hans Mooij:  
First flux qubit

2004: Wallraff, Blais et al:  
Circuit QED

2007: Koch et al:  
Transmon qubit

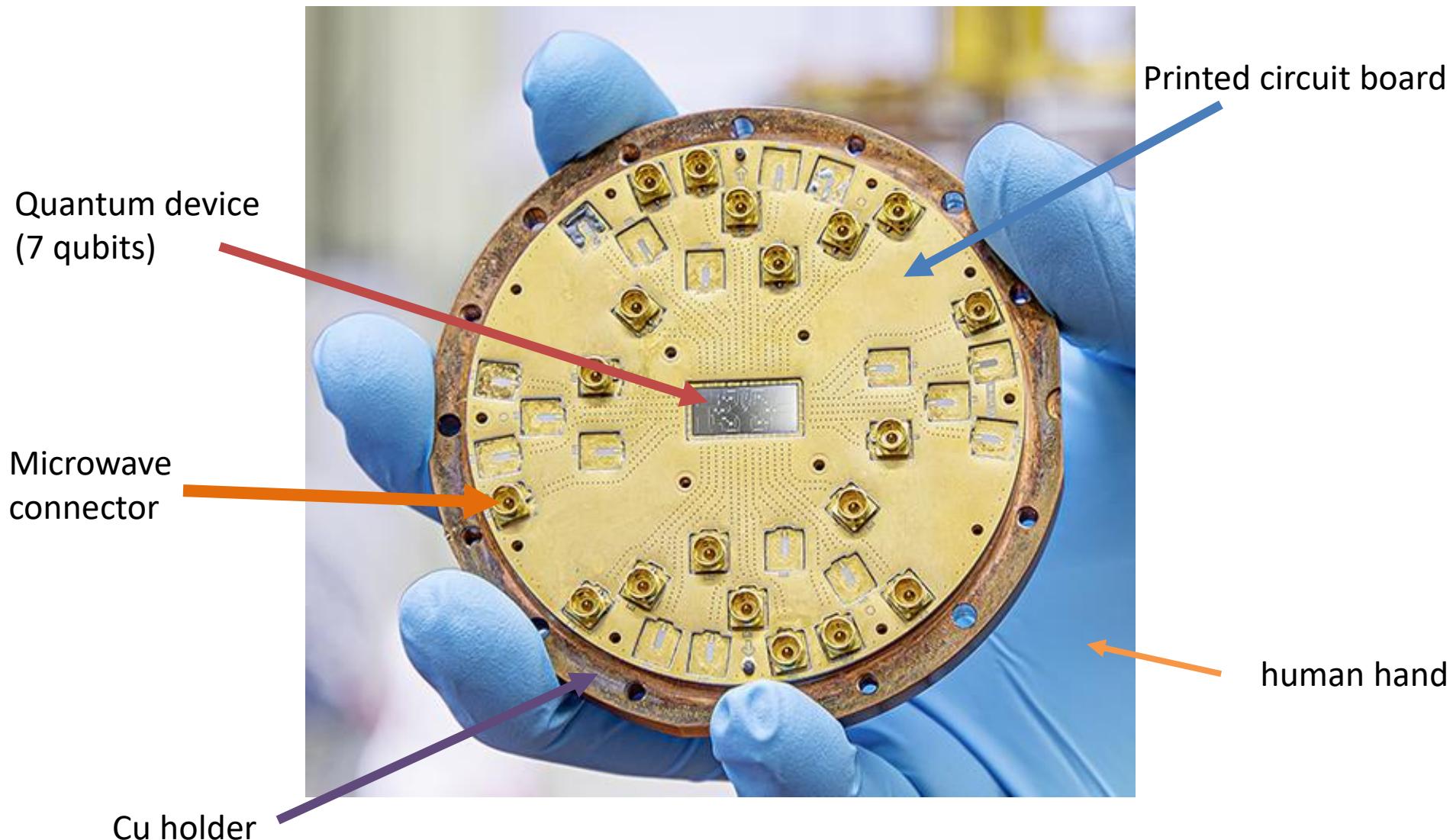
2009: Dicarlo et al:  
Demonstration of two-qubit algorithms

2011: Paik et al:  
High coherence transmon with 3D

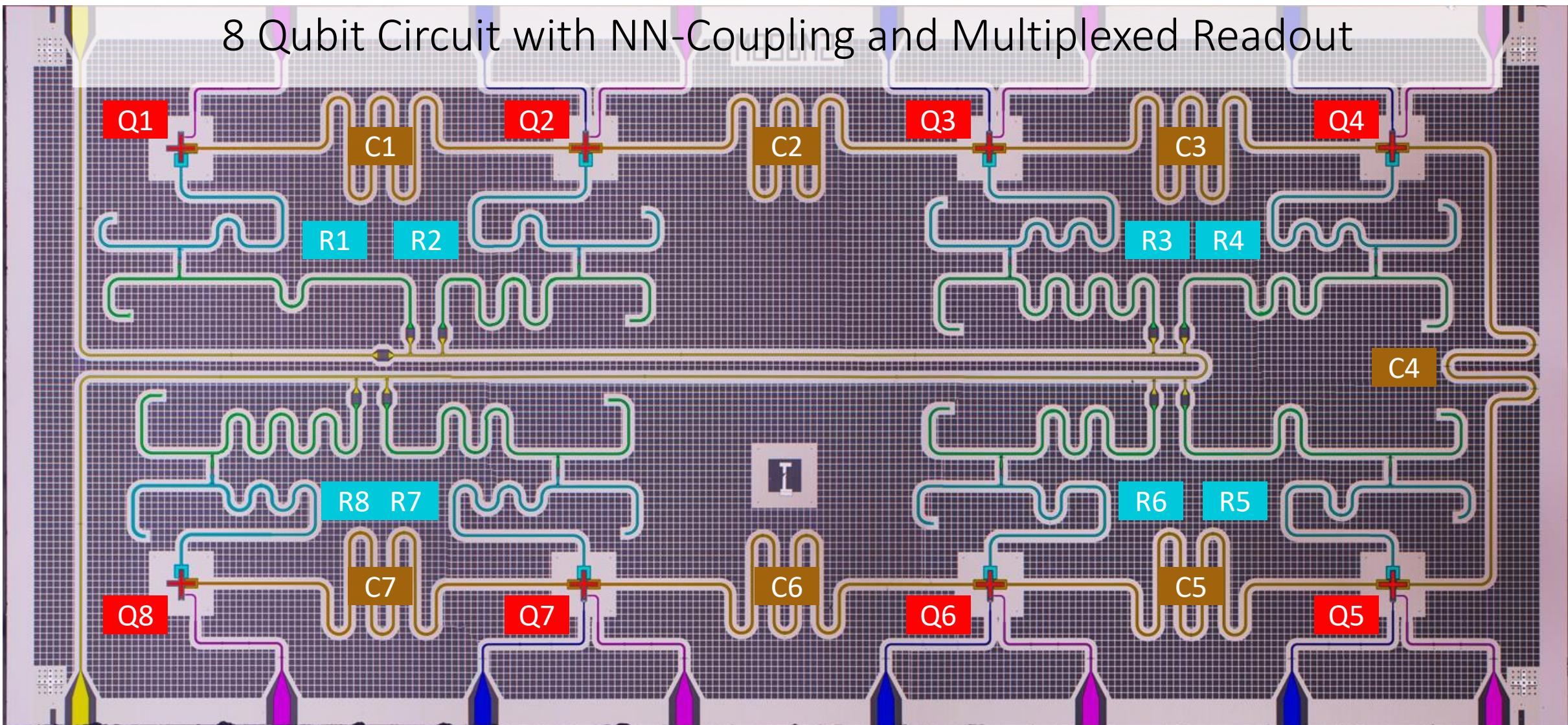
2016: IBM:  
5 qubit device on the cloud

2019: Google:  
Quantum computation beyond classical with 53 qubits

# A superconducting quantum processor



## 8 Qubit Circuit with NN-Coupling and Multiplexed Readout



Qubits

Readout resonators

Purcell filters

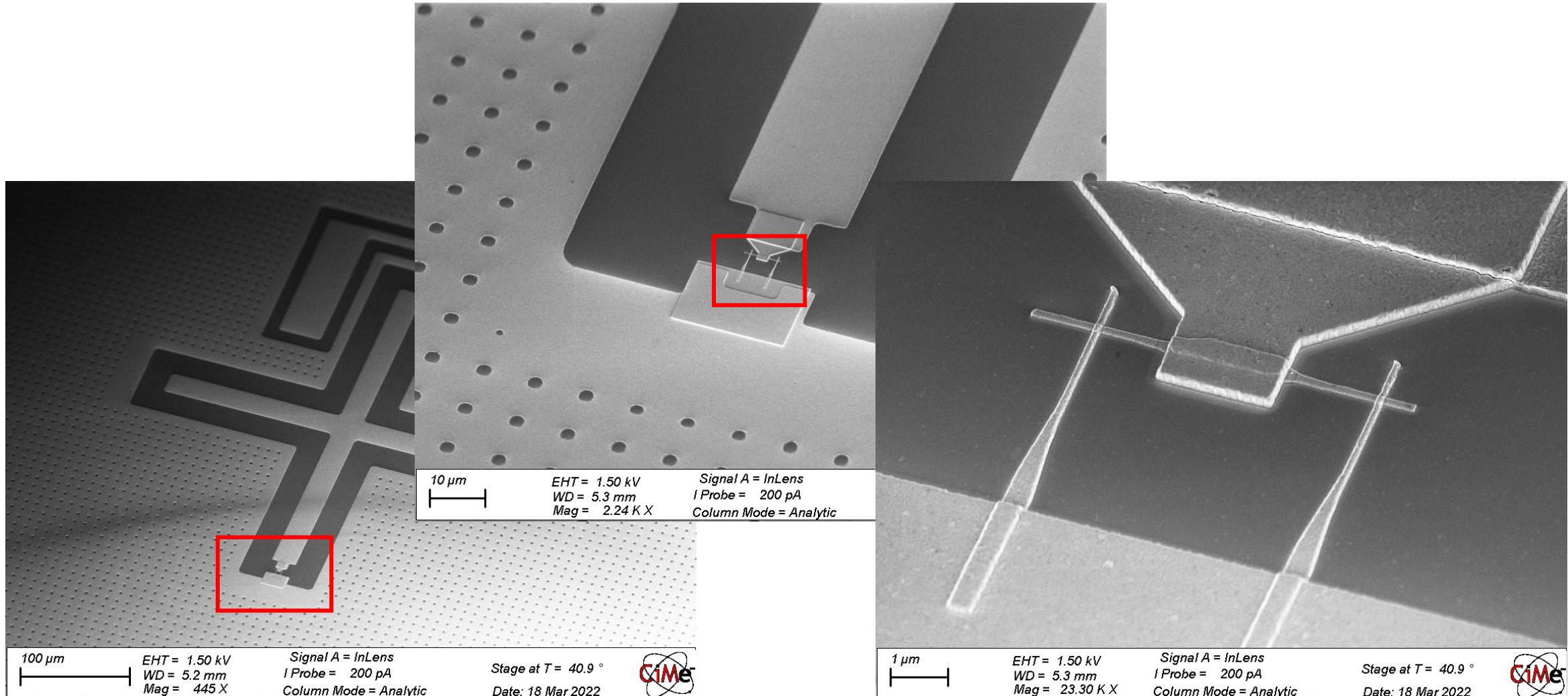
Coupling Bus resonators

Charge lines

Flux lines

Feed line

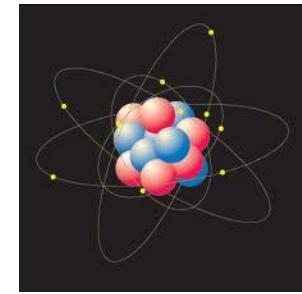
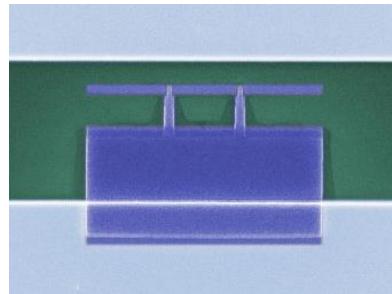
# Transmons in EPFL



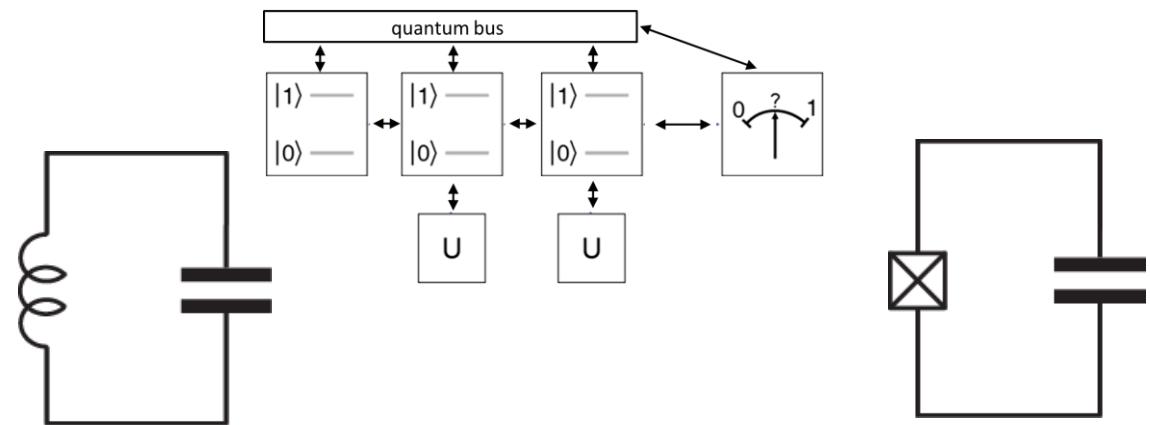
In collaboration with AQUA and LPQM labs in EPFL (pictures credit: Simone Frasca)

# Outline

- Brief introduction and motivation. Artificial vs natural quantum systems



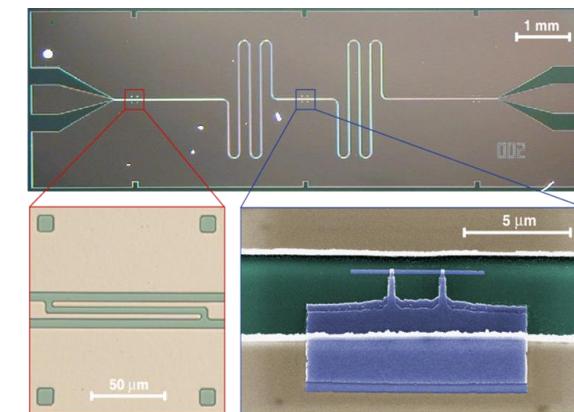
- Superconductivity
- Quantization of an LC oscillator. Classical and quantum state of the oscillator



- Dissipation less nonlinearity: Josephson junction

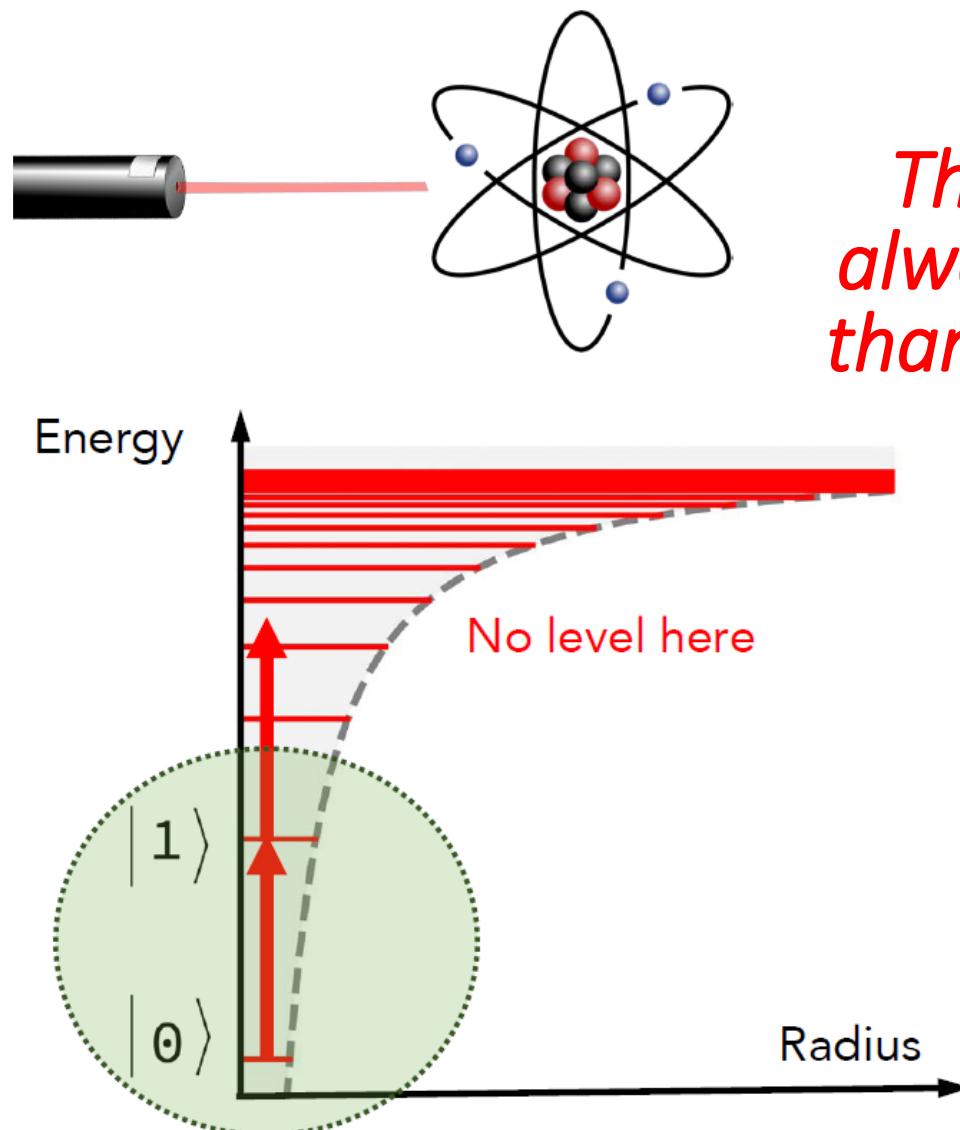


- Cooper Pair Box and Transmon qubits



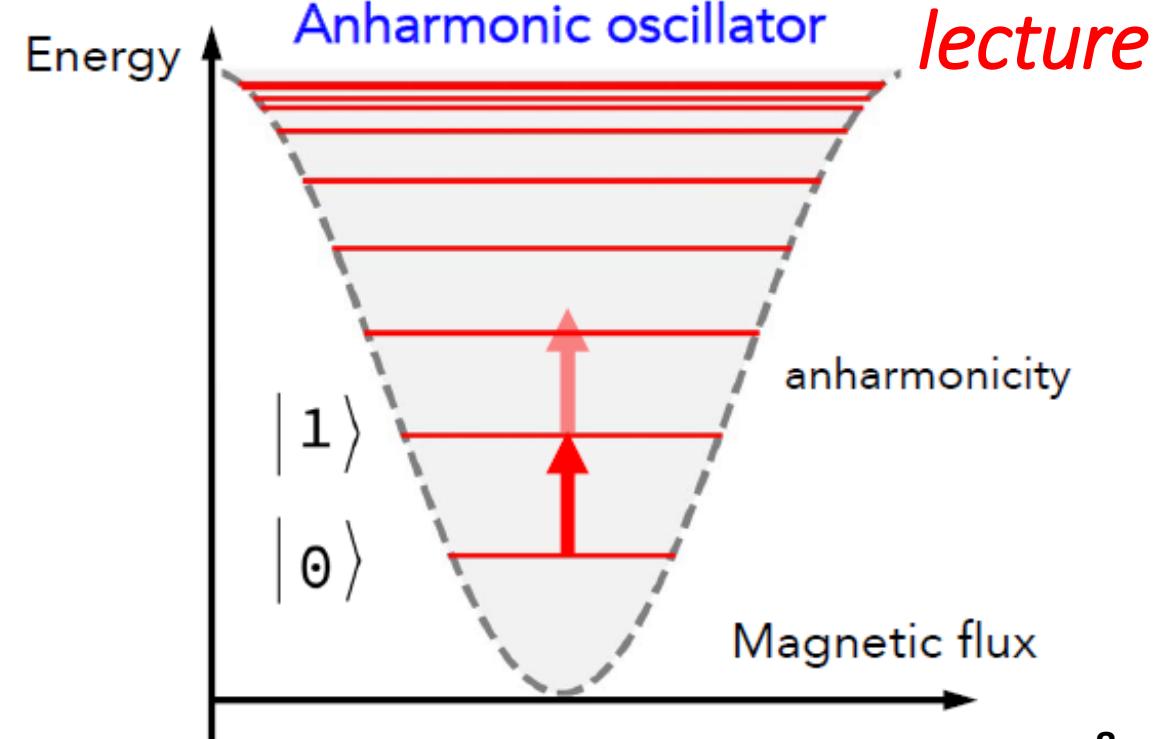
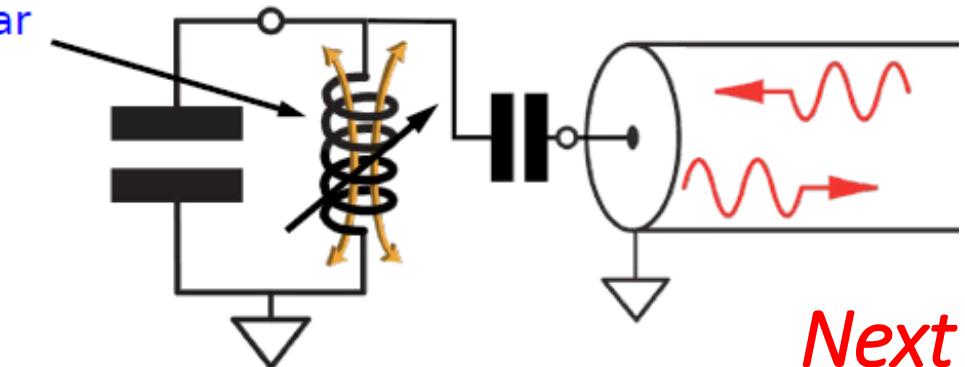
- Circuit Quantum Electrodynamics architecture

# Real Atoms



*There are  
always more  
than 2 levels!*

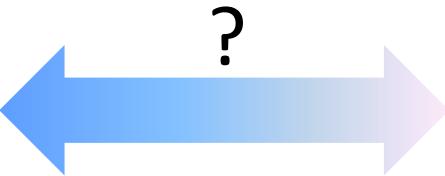
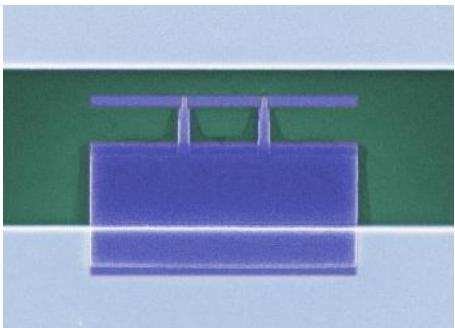
# Artificial Systems



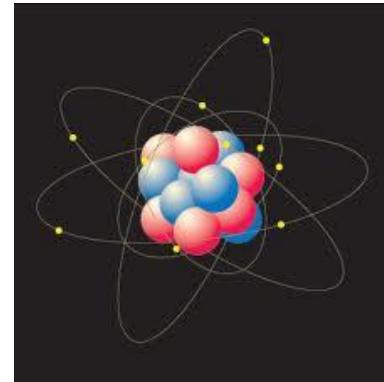
*Next  
lecture*

# Artificial and “natural”

Engineered quantum systems



“Natural” quantum particles



## Pros

- Parameters can be designed for specific purpose
- Many parameters can be also controlled *in situ*

## Cons

- Typically, solid-state fabricated structure hence relatively “dirty”

## Pros

- “Clean” quantum systems: long quantum coherence

## Cons

- Many parameters are fixed and cannot be tuned
- Hard to couple with each other

# Conventional electronic circuits

Conventional circuit elements:

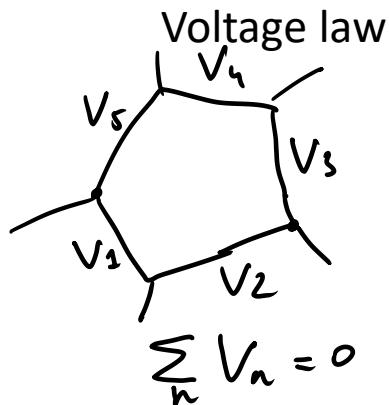
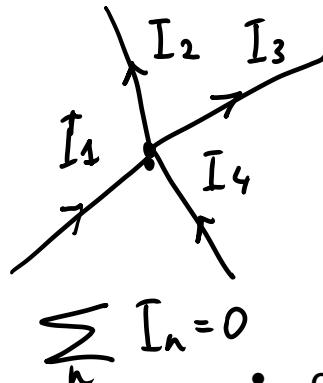


Described by classical variables (real numbers):

- $V$  (voltages) and  $q$  (charges)
- $\Phi$  (fluxes) and  $I$  (currents)

## Kirghoff's circuit laws:

Current law



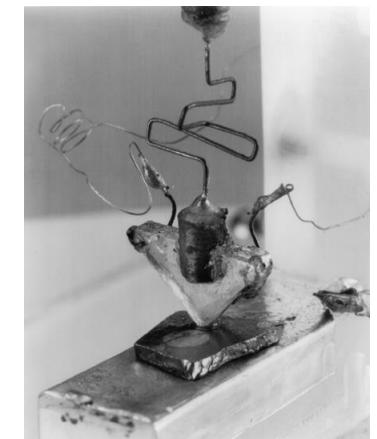
- classical physics
- no quantum mechanics
- no superposition principle
- no quantization of fields

Very complex electronic circuits as the base for modern electronic technology:



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first transistor at Bell Labs (1947)



# Quantum Electronic Circuits

basic circuit elements:

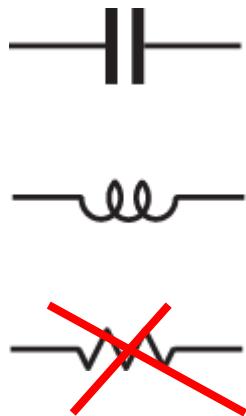


properties :

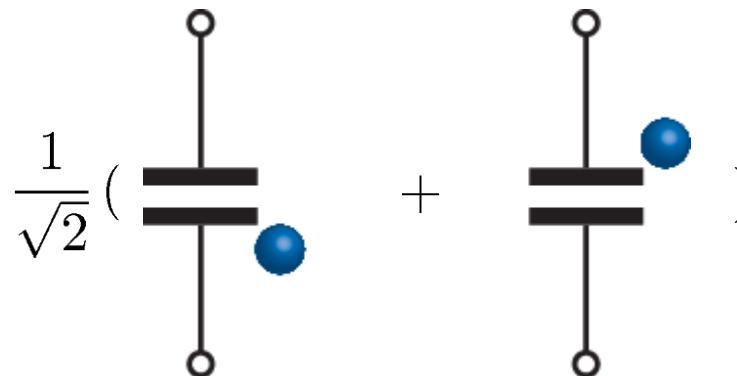
- classical physics
- no quantum mechanics
- no superposition principle
- no quantization of fields

# Quantum Electronic Circuits

basic circuit elements:



charge on a capacitor:

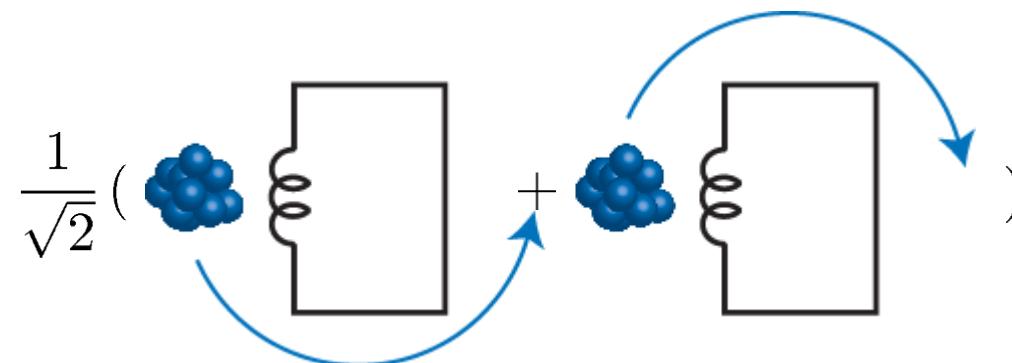


quantum superposition states of:

- charge  $Q$
- flux  $\phi$

$Q, \phi$  are conjugate variables

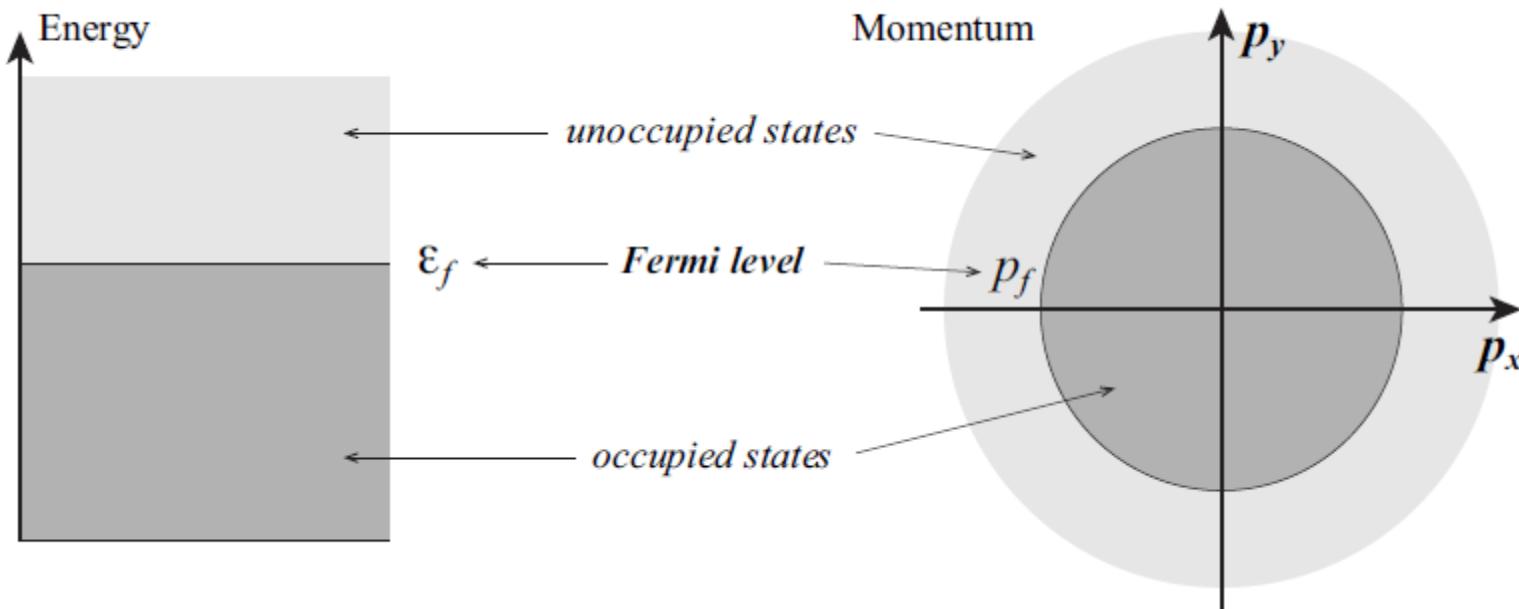
current or magnetic flux in an inductor:



quantum uncertainty relation

$$\Delta\phi\Delta Q > h$$

# Review: Normal Metal and Superconductors

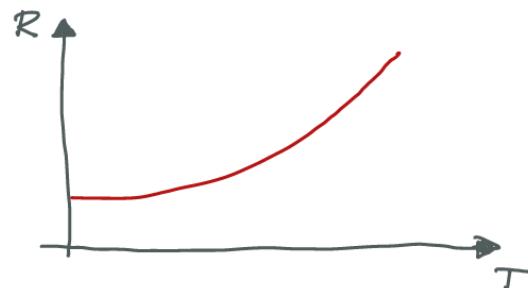


# Superconductivity

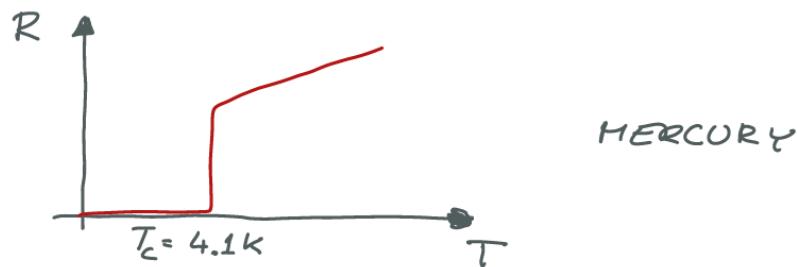
## FIRST MEASUREMENT:

- 1911 KAMERLINGH ONNES (LEIDEN)

MEASURING RESISTANCE OF METALS IN LIQUID  ${}^4\text{He}$



GOLD, PLATINUM  
(PHONONS FREEZE OUT  
RESIDUAL RESISTANCE)  
LINEAR DEPENDENCE

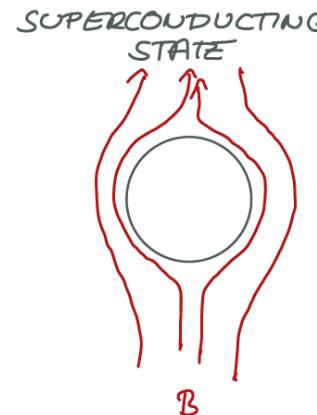
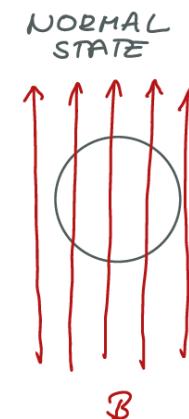


FOR EXAMPLE,  
Al  $\sim 1.2\text{K}$   
Nb  $\sim 9\text{K}$   
Ta  $\sim 4\text{K}$

① ZERO ELECTRICAL RESISTANCE BELOW  $T_c$

## "SECOND" MEASUREMENT

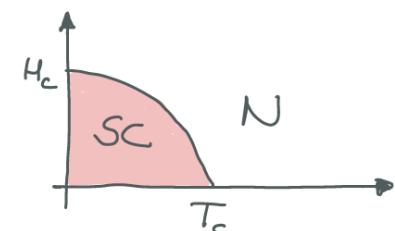
- 1933 MEISSNER & OCHSENFELD



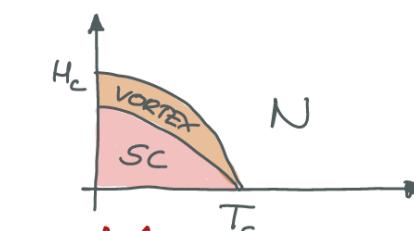
② THE MAGNETIC FIELD IS EXPELLED FROM SC.

→ IT BREAKS DOWN ABOVE  $H_c(T)$

PHASE DIAGRAM OF  
TYPE I SC

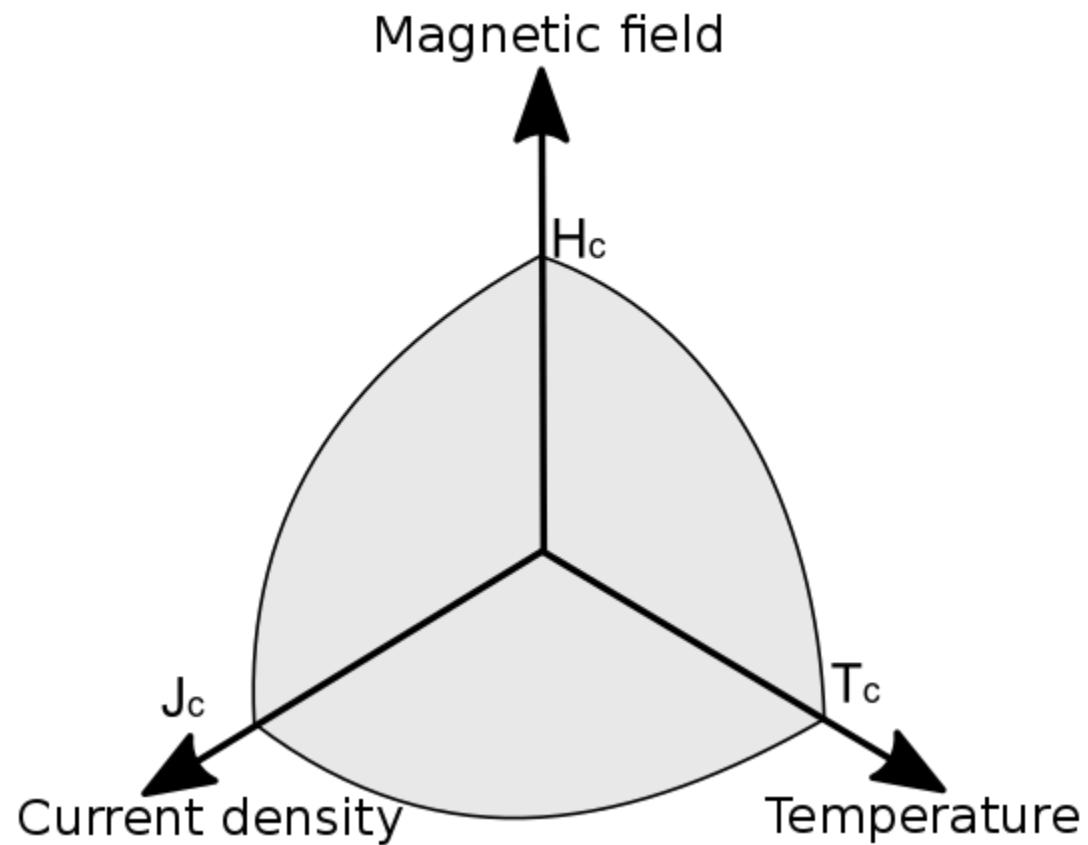


PHASE DIAGRAM OF  
TYPE II SC



EACH VORTEX  
CORE CARRIES  
 $\phi_0 = h/2e$  FLUX  
14

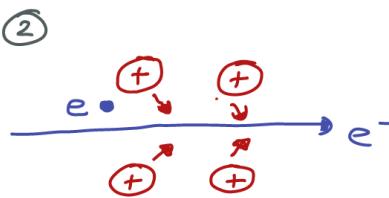
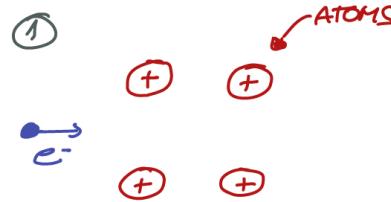
# How to destroy superconductivity?



## BCS THEORY (1957) BARDEEN, COOPER, SCHRIEFFER

### MICROSCOPIC THEORY OF SUPERCONDUCTIVITY

- HOW CAN ELECTRONS ATTRACT EACH OTHER ??

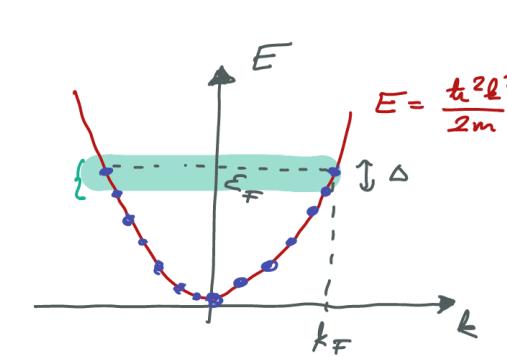
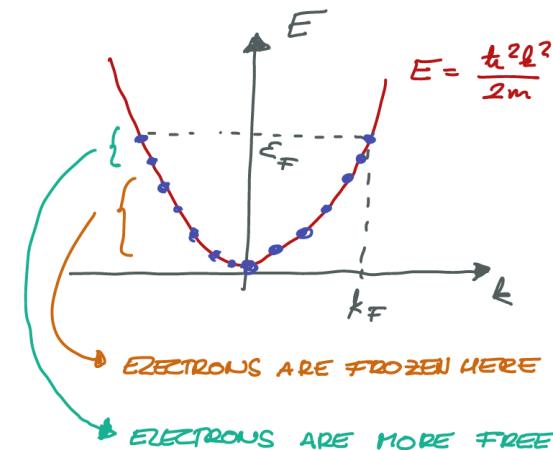


③

LATTICE - MEDIATED  
ELECTRON-ELECTRON  
INTERACTION

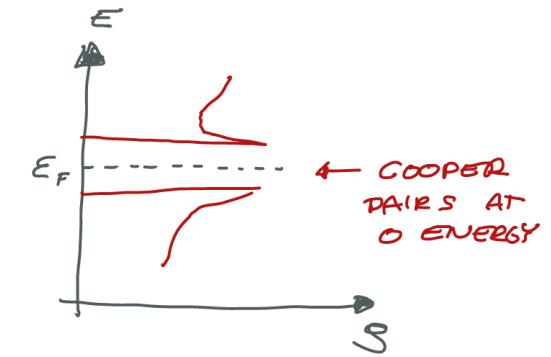
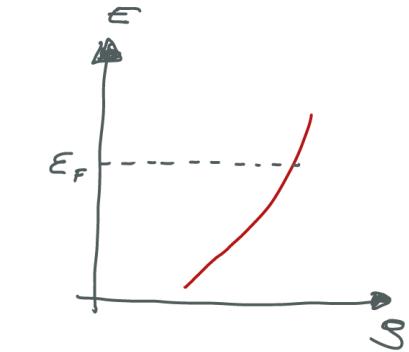


WHICH ELECTRONS FORM COOPER PAIRS IN A METAL ?



COOPER PAIRS ARE FORMED AROUND THE FERMI SURFACE ( $\Delta$ ) ,  $\Delta = 1.76 L_B T_C$

GAP IS FORMED

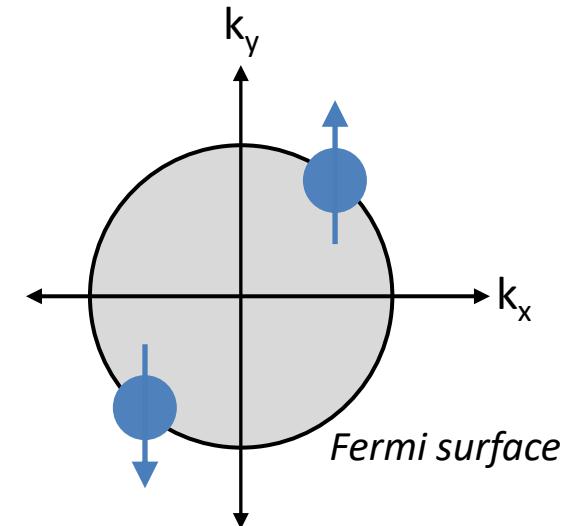


- CAN NOT GIVE ENERGY DURING SCATTERING.
- MAKE A QUBIT
- HOW CAN WE PUT ENERGY LEVELS INTO IT? QUANTUM CIRCUITS.

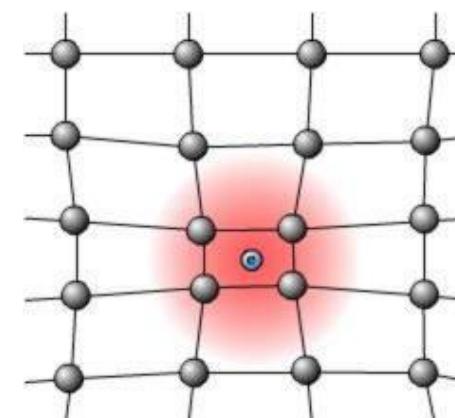
# BCS theory

- Bardeen, Cooper, Schrieffer (1957)
- Microscopic theory of superconductivity

[1] Electrons with opposite momentum and opposite spin form bound states: **Cooper pairs**

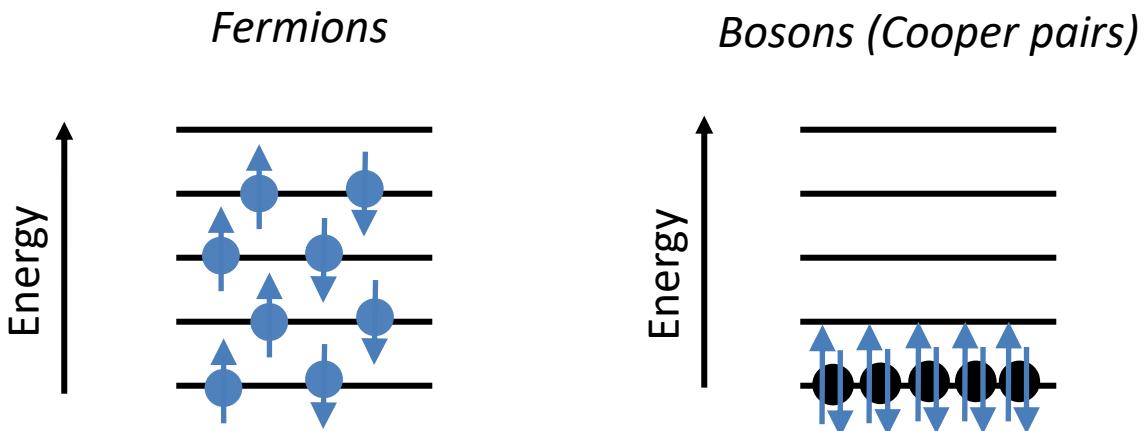


[2] The attraction between the electrons is provided by **electron-phonon interaction** (i.e., interaction with lattice vibrations)

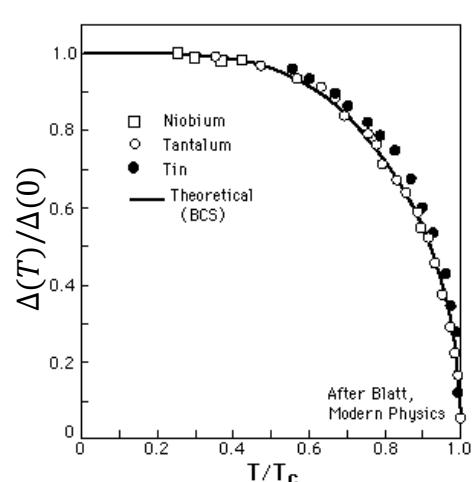
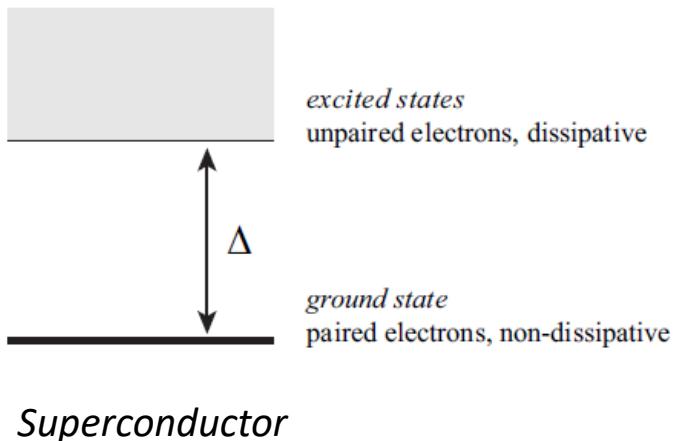


# BCS theory

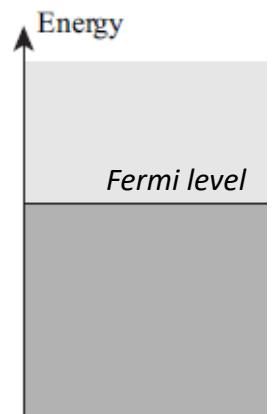
[3] Cooper pairs are (composite) **bosons**, they condense to a single ground state



[4] Pairing results in an **energy gap** in the



Dimensionless energy gap  $\Delta(T)/\Delta(0)$   
in niobium, tantalum, and tin.



# BCS theory

[5] The superconducting state is described by a ***single macroscopic wave function***

$$\psi = |\psi| e^{i\phi}$$

The superconducting state can be described by an order parameter  $\psi$

$$|\psi|^2 = n_s \text{ (density of Cooper pairs)}, \phi: \text{phase of the condensate}$$

**Wave function is the same everywhere in the superconductor**

(\*provided there are no magnetic fields)

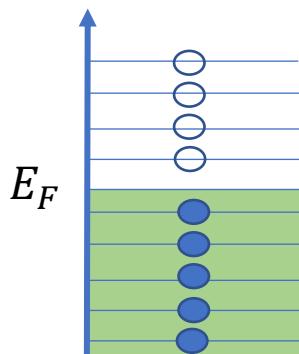
In the superconducting state, the wavefunction acquires a certain rigidity in that there is a global phase  $\phi$ .

This has important consequences (including the Meissner effect and flux quantization).

# Superconductors vs normal conductors

## Normal metal

Charge carrier: electrons, fermions ( $S=1/2$ )

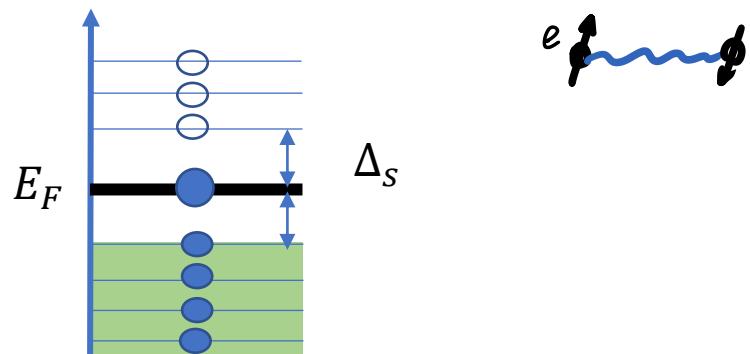


$$T = 0$$

Scattering with impurities can take some energy -> create **resistance**

## Superconductor

Charge carrier: Cooper pairs, pairs of electrons, bosons ( $S=0$ )



Excitations are **gapped** ->

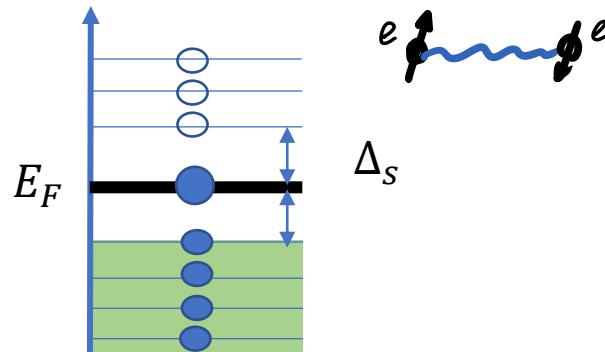
No electrical resistance for  $T < T_c = \frac{\Delta_s}{k_B}$

Typical superconducting materials:

- Niobium (Nb):  $T_c = 9.2$  K,  $\frac{2\Delta_s}{h} = 725$  GHz
- Aluminium (Al):  $T_c = 1.2$  K,  $\frac{2\Delta_s}{h} = 100$  GHz

# Superconducting phase

Charge carrier: Cooper pairs, pairs of electrons, bosons ( $S=0$ )



Chunk of a superconductor



No magnetic field:

All Cooper pairs condensed in one macroscopic wave function

$$\psi = \sqrt{n} \exp(i\delta)$$

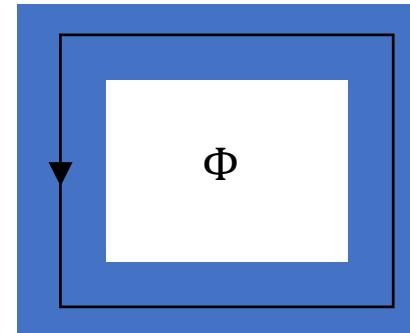
$n$  is Cooper pair density and  $\delta$  is global phase.

With magnetic field:

$\Delta\delta = \frac{2\pi\Phi}{\Phi_0}$ , where  $\Phi_0 = \frac{\hbar}{2e} = 2.067 \times 10^{-15}$  Wb is called the flux quantum

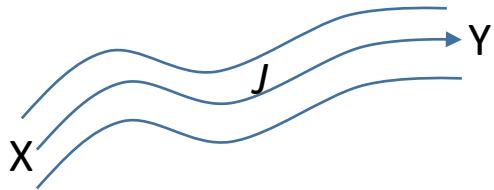
As  $\Delta\delta = 2\pi n \rightarrow$  flux quantization:

$$\Phi = n \Phi_0$$



# Flux quantization

Consider the change in phase when current flows through a superconductor from X → Y

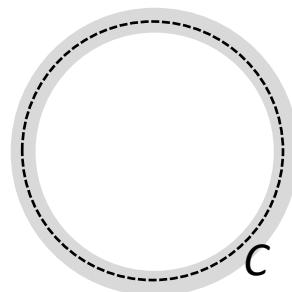


$$\Delta\phi_{XY} = \frac{1}{\hbar} \int_X^Y p(\mathbf{r}) \cdot d\mathbf{r}$$

In the presence of a magnetic field,  $\mathbf{B}$ :  $\longrightarrow$   $\Delta\phi_{XY} = \frac{m}{en_s \hbar} \int_X^Y \mathbf{J}(\mathbf{r}) \cdot d\mathbf{r} + \frac{2e}{\hbar} \int_X^Y \mathbf{A}(\mathbf{r}) \cdot d\mathbf{r}$

$$p = 2mv + 2e\mathbf{A} \quad \nabla \times \mathbf{A} = \mathbf{B}$$

Consider supercurrent flowing around a ring: **flux quantization**



$$\Delta\phi_C = n2\pi$$

phase of the  
superconducting  
condensate must be  
single-valued

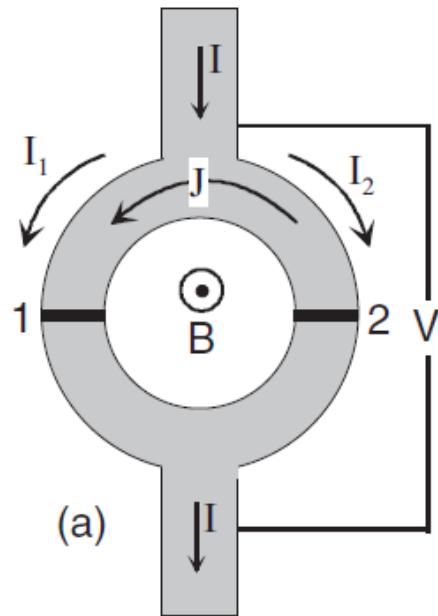
$$\Phi' \equiv \frac{m}{2e^2 n} \oint_C \mathbf{J}(\mathbf{r}) \cdot d\mathbf{r} + \Phi_A = n\Phi_0$$

Total flux in SC ring quantized in units of SC  
flux quantum,  $\Phi_0 = h/2e$

In general, phase difference in a SC circuit can  
be expressed as:  $\phi = 2\pi\Phi_A/\Phi_0$

# SQUID loop (following class)

- Superconducting quantum *interference* device
- Results in a flux-tunable critical current
- Used very often in SC qubits to tune  $E_J$



$$H = -E_J \cos \phi_1 - E_J \cos \phi_2$$
$$\phi_1 - \phi_2 - \frac{2\pi\Phi}{\Phi_0} = 2\pi n$$
$$\frac{\phi_1 + \phi_2}{2} = \phi$$

Small exercise: show that the Hamiltonian can be written as

$$H = -E_J(\Phi) \cos \phi$$

# Brief recap of Electromagnetism

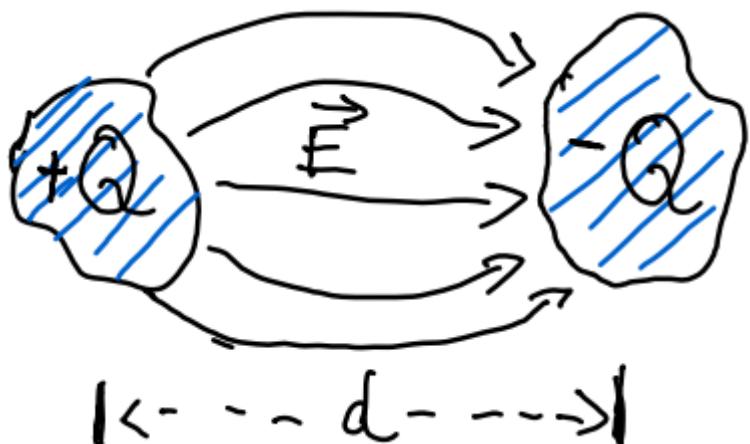
# Quantization of Electrical Circuits

**GOAL:** given an electrical circuit composed of inductors and capacitors, find the system Hamiltonian.

N.B. (for the moment) there are no Josephson junctions and no resistors.

**Lumped Element representation:**

Consider two metallic islands charged with charge + and - Q



$\vec{E}$  electric field

$d$  characteristic length  
of the circuit

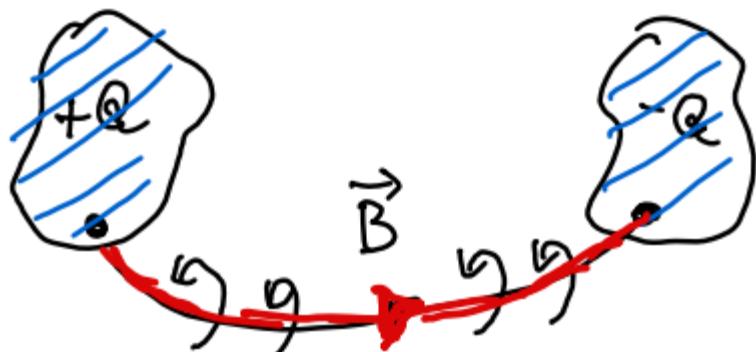
- In a static configuration, charges rearrange on the surface of the metals and the electric field vanishes inside the metal.
- The time to reach a static charge configuration is  $\tau \sim d/c$ , with  $d$  the typical conductor dimension and  $c$  the velocity of light.
- If  $Q(t)$  changes slowly compared to  $\tau$ , the electric field follows quasi-instantaneously.

The energy required to move one unit of charge from one to the other island is path-independent and given by the voltage  $V = Q/C$ , with  $C$  the capacitance of the conductor configuration (which depends on geometry and dielectric medium).

Total energy stored  
in the electric field

$$\rightarrow E_{el} = \int_0^Q dQ' V(Q') = \int_0^Q dQ' \frac{Q'}{C} = \frac{Q^2}{2C}$$

Now consider an additional metallic wire connecting the two islands:



$$I = -\dot{Q}$$

magnetic flux

$$\dot{\Phi} = L I \rightarrow \text{current}$$

Inductance  
(geometry-dependent)

Total energy stored  
in a magnetic field:

$$\rightarrow E_{mag} = \int_0^I dI' \phi(I') = \frac{1}{2} L I^2 = \dot{\Phi}^2 / 2L$$

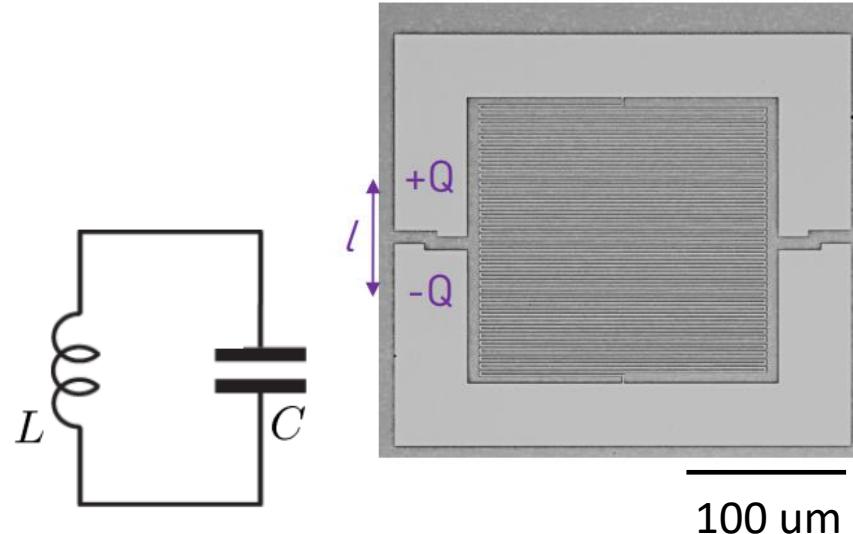
# There are two types of on-chip electronic components:

Lumped elements and Distributed elements.

We can distinguish the two types by their typical dimension 'd' compared to the wavelength ' $\lambda$ ' of the microwave radiation at the relevant frequency.

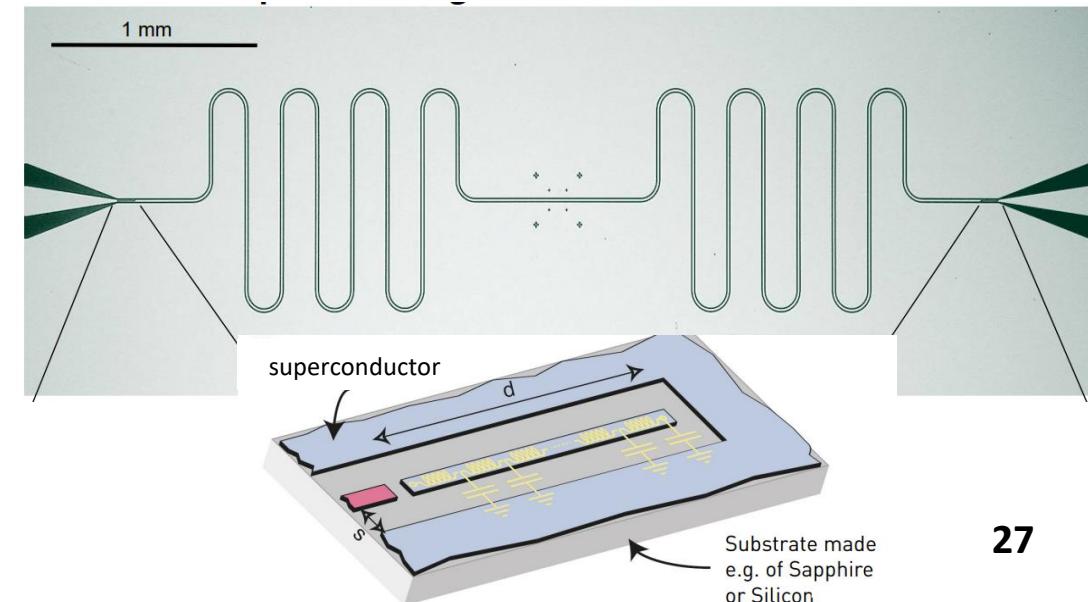
- For lumped elements, the size 'd' of the components is much smaller than the wavelength ' $\lambda$ ':  $d \ll \lambda$
- For distributed elements, the size of the components 'd' is roughly of the same size as ' $\lambda$ ':  $d \sim \lambda$

## Lumped elements resonators



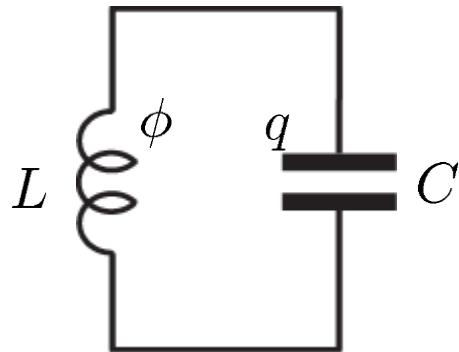
Resonance frequencies typically between  
4 – 8 GHz

## Distributed resonators:



# Classical dynamics of harmonic oscillator

harmonic LC oscillator:



From Kirchhoff's rules and classical electrodynamics, the variables are related to each other:

$$q = CV_C$$

$$\Phi = LI$$

$$V_L = -LI = -\dot{\Phi} = -V_C$$

Take  $\Phi$  as a new coordinate of a fictitious particle and identify :

$$\omega = \frac{1}{\sqrt{LC}}$$

Inductive energy  $U(\phi) = \frac{\phi^2}{2L}$  -> as potential energy

Electrostatic energy  $K(\dot{\phi}) = \frac{CV_C^2}{2} = \frac{C\dot{\phi}^2}{2}$  -> as kinetic energy

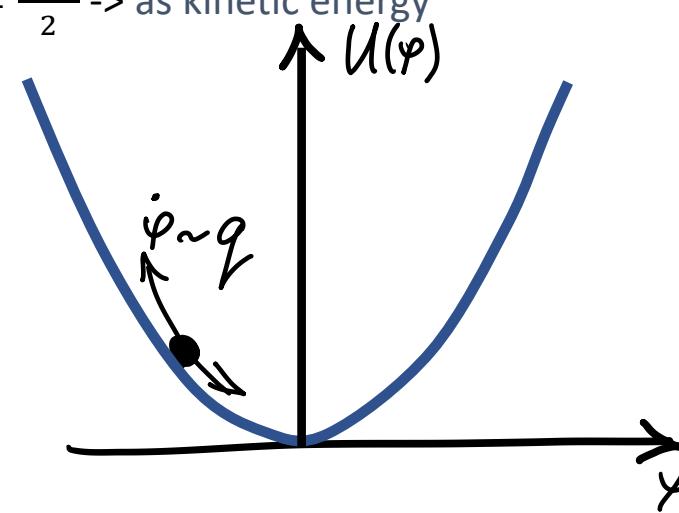
General procedure to find quantum Hamiltonian

1) Set up the Lagrangian function:

$$\mathcal{L}(\dot{\phi}, \phi) = K - U = \frac{C\dot{\phi}^2}{2} - \frac{\phi^2}{2L}$$

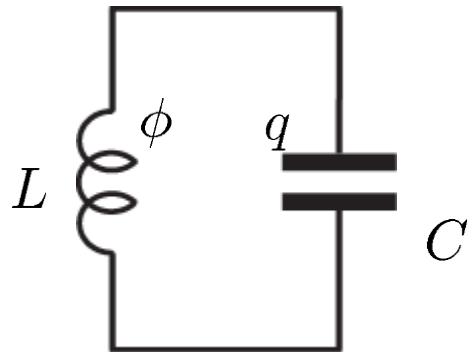
$$\rightarrow \text{generalized momentum } p = \frac{\partial \mathcal{L}}{\partial \dot{\phi}} = C\dot{\phi} = q$$

Particle with mass  $C$  momentum  $q$  and coordinate  $\Phi$  in a harmonic trap



# Quantization of harmonic oscillator

harmonic LC oscillator:



$$\omega = \frac{1}{\sqrt{LC}}$$

2) Legendre Transformation:

$$q = \frac{\partial \mathcal{L}}{\partial \dot{\Phi}} = C\dot{\Phi}$$

3) The Hamiltonian function:

$$H(q, \Phi) = \dot{\Phi}q - \mathcal{L} = \frac{q^2}{C} - \frac{q^2}{2C} - \left(-\frac{\Phi^2}{2L}\right) = \frac{q^2}{2C} + \frac{\Phi^2}{2L}$$

where we have chosen to write the Hamiltonian in terms of the charge and magnetic flux.

Now we are stuck since we do not know the quantum operators for  $Q$  and  $\Phi$ .

For inspiration, let us turn to the [mechanical harmonic oscillator](#). Recall the Hamiltonian for the classic mechanical harmonic oscillator, for example a mass on a spring, is:

$$H_{\text{cl,mech}} = \frac{p^2}{2m} + \frac{kx^2}{2}$$

$$H_{\text{cl,LC}} = \frac{Q^2}{2C} + \frac{\Phi^2}{2L}$$

where  $p$  is the momentum of the mass  $m$ ,  $k$  the spring constant and  $x$  is the displacement of the mass from equilibrium.

Comparing these two Hamiltonians gives us the idea to think of  $Q$  as  $p$  and  $x$  as  $\Phi$ .

As further motivation we recall that  $p$  and  $x$  satisfy the canonical equations of motion

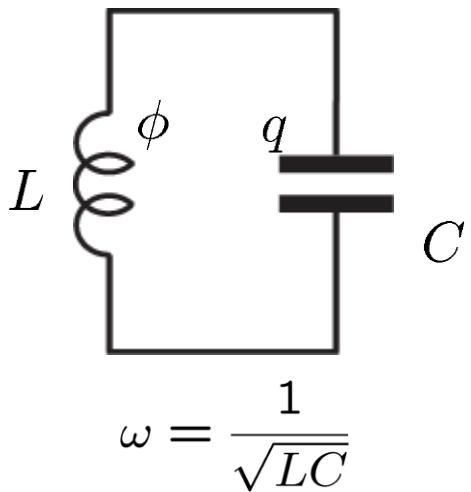
$$\frac{\delta H}{\delta p} = \dot{x} \quad \frac{\delta H}{\delta x} = -\dot{p}$$

	Classic Mechanical	Classic Electronic	Quantum Mechanical	Quantum Electronic
Displacement	$x$	$\Phi$	$\hat{x}$	$\hat{\Phi}$
Flow	$p$	$Q$	$\hat{p} = -i\hbar \frac{d}{dx}$	$\hat{Q} = -i\hbar \frac{d}{d\Phi}$
Force	$m$	$C$	$m$	$C$
Proportionality				
Restoring Proportionality	$k$	$\frac{1}{L}$	$k$	$\frac{1}{L}$
Resonant Frequency	$\omega = \sqrt{\frac{k}{m}}$	$\omega = \frac{1}{\sqrt{LC}}$	$\omega = \sqrt{\frac{k}{m}}$	$\omega = \frac{1}{\sqrt{LC}}$
Commutation Relations	-	-	$[\hat{x}, \hat{p}] = i\hbar$	$[\hat{\Phi}, \hat{Q}] = i\hbar$

Relationship of variables between the mechanical and electronic harmonic oscillator as well as classically and quantum mechanically.

# Quantization of harmonic oscillator

harmonic LC oscillator:



2) Legendre Transformation:

$$q = \frac{\partial \mathcal{L}}{\partial \dot{\Phi}} = C\dot{\Phi}$$

3) The Hamiltonian function:

$$H(q, \Phi) = \dot{\Phi}q - \mathcal{L} = \frac{q^2}{C} - \frac{q^2}{2C} - \left(-\frac{\Phi^2}{2L}\right) = \frac{q^2}{2C} + \frac{\Phi^2}{2L}$$

4) Quantization:

Flux and charge operator (similar to coordinate and momentum) with commutation relation  $[\hat{\Phi}, \hat{q}] = i\hbar$

$$\hat{\Phi} = \Phi$$

$$\hat{q} = -i\hbar \frac{\partial}{\partial \Phi}$$

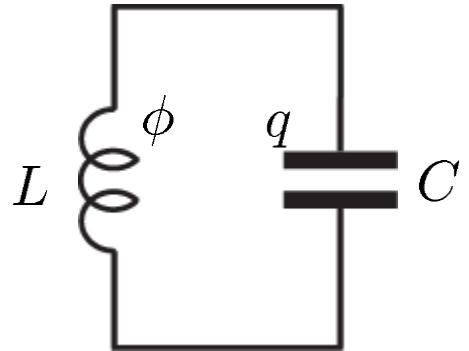
The quantum Hamiltonian

$$\hat{H}(q, \Phi) = \frac{\hat{q}^2}{2C} + \frac{\hat{\Phi}^2}{2L}$$

Diagonalizing the Hamiltonian -> find the allowed energies of the system

# Creation and annihilation operators

harmonic LC oscillator:



$$\omega = \frac{1}{\sqrt{LC}}$$

5) Express H in terms of annihilation and creation operators

$$\hat{H}(q, \Phi) = \frac{\hat{q}^2}{2C} + \frac{\hat{\Phi}^2}{2L}$$

Introduce creation and annihilation operators as

$$\hat{a}^\dagger = \sqrt{\frac{\omega C}{2\hbar}} (\hat{\Phi} - i \frac{\hat{q}}{C\omega})$$

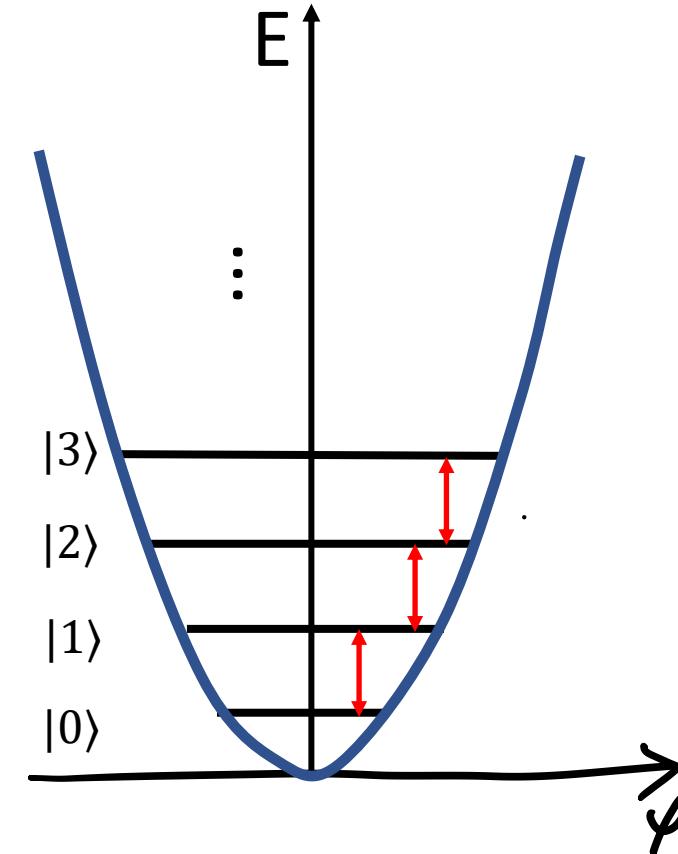
$$\hat{a} = \sqrt{\frac{\omega C}{2\hbar}} (\hat{\Phi} + i \frac{\hat{q}}{C\omega})$$

Solution of the Hamiltonian are number states  $|n\rangle$ :

$$\hat{a}|n\rangle = \sqrt{n}|n-1\rangle$$

$$\hat{a}^\dagger|n\rangle = \sqrt{n+1}|n+1\rangle$$

$$\hat{a}^\dagger\hat{a}|n\rangle = n|n\rangle$$



Can we define a qubit with an harmonic system?

**Find the system Hamiltonian  
for a general circuit network**

# Lagrangian formalism

- Formalism of mechanics that deals with energies rather than forces (Newtonian mechanics)
- Lagrangian of a system is expressed with terms involving the generalized coordinates only and generalized velocities only.

$$\mathcal{L} = \mathcal{L}(X, \dot{X})$$

- Often represents the difference between the kinetic energy (velocity dependent) and potential energy (position dependent)

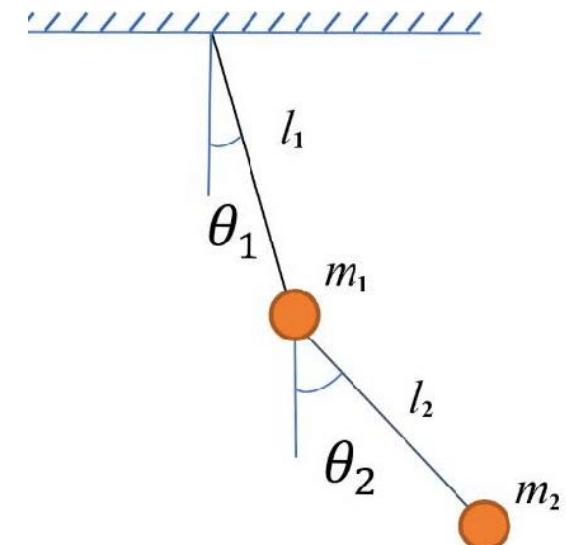
$$\mathcal{L} = E_k - E_p$$

- Hamiltonian of the system is expressed with terms involving generalized coordinates only and generalized momenta only

$$\mathcal{H} = \mathcal{H}(X, P)$$

- Conjugate variable (generalized momentum) can be expressed as:  $P = \frac{\partial \mathcal{L}}{\partial \dot{X}}$
- Connection between Lagrangian and Hamiltonian via the Legendre transformation:

$$\mathcal{H}(X, P) = \dot{X}P - \mathcal{L}(X, \dot{X})$$



# Quantize the LC circuit – for reals!

- Step 1: Write down Lagrangian

$$\mathcal{L} = E_{cap} - E_{ind} = \frac{1}{2}C\dot{\Phi}^2 - \frac{1}{2L}\Phi^2$$

- Step 2: Find conjugate variable

$$Q = \frac{\partial \mathcal{L}}{\partial \dot{\Phi}} = C\dot{\Phi}$$

- Step 3: Calculate classical Hamiltonian

$$\mathcal{H}(\Phi, P) = C\dot{\Phi}^2 - \mathcal{L}(\Phi, \dot{\Phi}) = \frac{1}{2C}Q^2 + \frac{1}{2L}\Phi^2$$

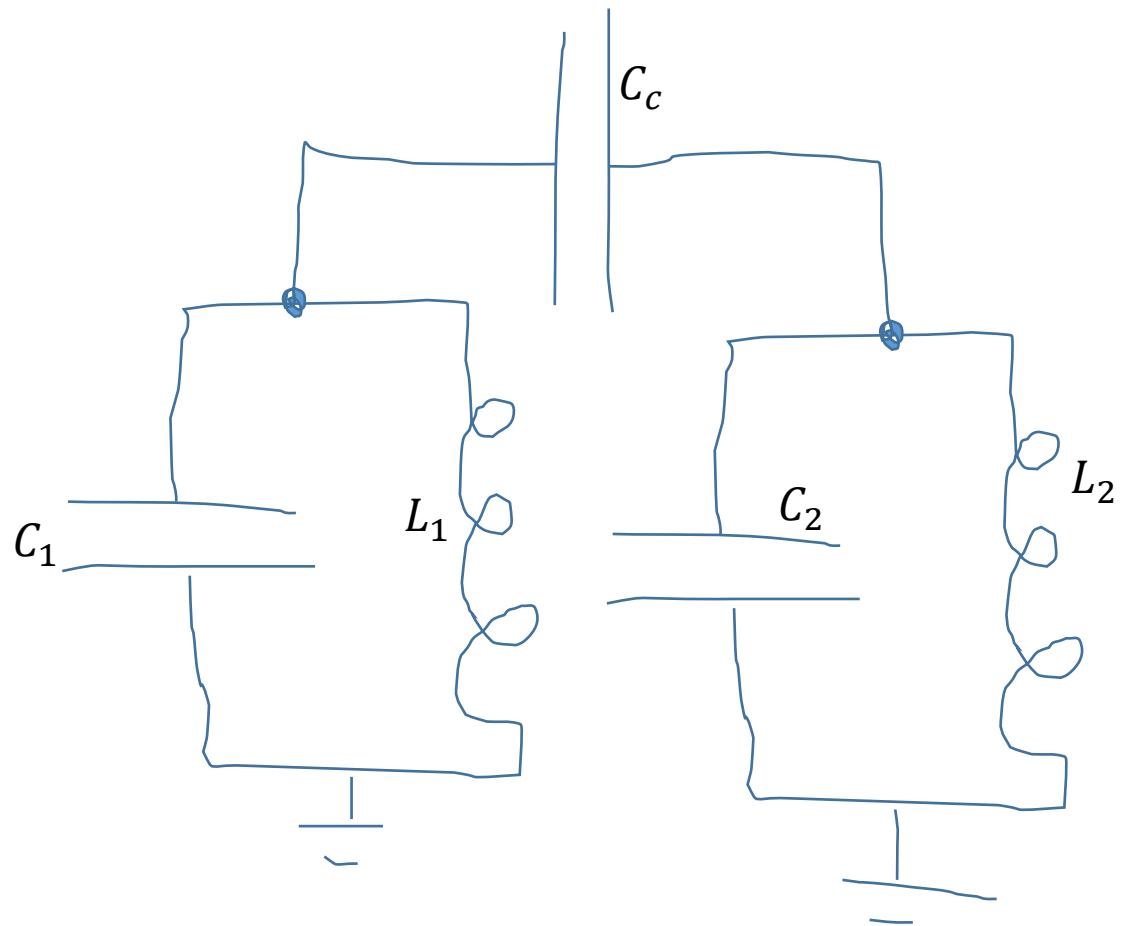
- Step 4: Quantize the Hamiltonian

$$[\hat{\Phi}, \hat{Q}] = i\hbar$$

$$H = \frac{1}{2L}\hat{\Phi}^2 + \frac{1}{2C}\hat{Q}^2$$

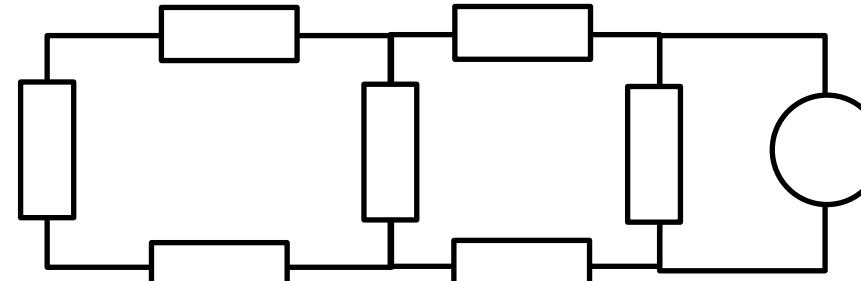
# Exercise 1

- Consider the circuit of two capacitively coupled LC circuit
  - Write the Lagrangian
  - Calculate the quantum Hamiltonian
- Hints:
  - consider each node to have a given flux  $\phi_i$
  - Energy of capacitor connecting nodes  $\phi_i$  and  $\phi_j$ :  $\frac{1}{2}C(\dot{\phi}_i - \dot{\phi}_j)^2$
  - Energy of inductor connecting nodes  $\phi_i$  and  $\phi_j$ :  $\frac{1}{2L}(\phi_i - \phi_j)^2$
  - Node flux of ground is 0



# Hamiltonian of an arbitrary circuit

$$\boxed{\quad} = \left\{ \begin{array}{l} \text{---} \parallel \\ \text{---} \circlearrowleft \\ \times \\ \text{---} \perp \\ \text{---} \otimes \end{array} \right.$$



HAMILTONIAN ???

Correct procedure described in :

M. H. Devoret, p. 351 in *Quantum fluctuations* (Les Houches 1995)

Girvin, S. M. (2011), Circuit QED: superconducting qubits coupled to microwave photons.

G. Wendum and V. Shumeiko, cond-mat/0508729

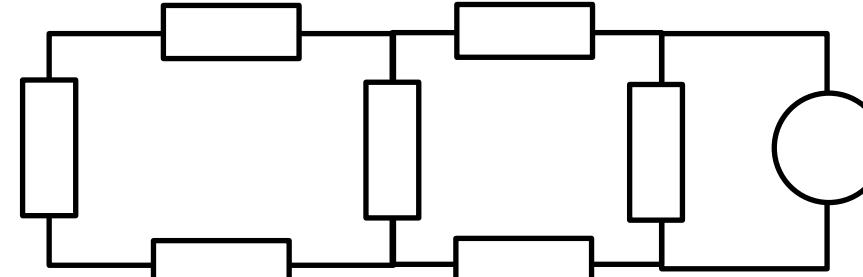
U. Vool, M.H. Devoret, Introduction to quantum electromagnetic circuits,  
<https://doi.org/10.1002/cta.2359>

Credit P. Bertet

SPEC, CEA Saclay (France),  
 Quantronics group

# Hamiltonian of an arbitrary circuit

$$\boxed{\quad} = \begin{cases} \text{---} || \\ -\text{---} \text{---} \\ \times \\ \text{---} | \\ \text{---} \text{---} \end{cases}$$



HAMILTONIAN ???

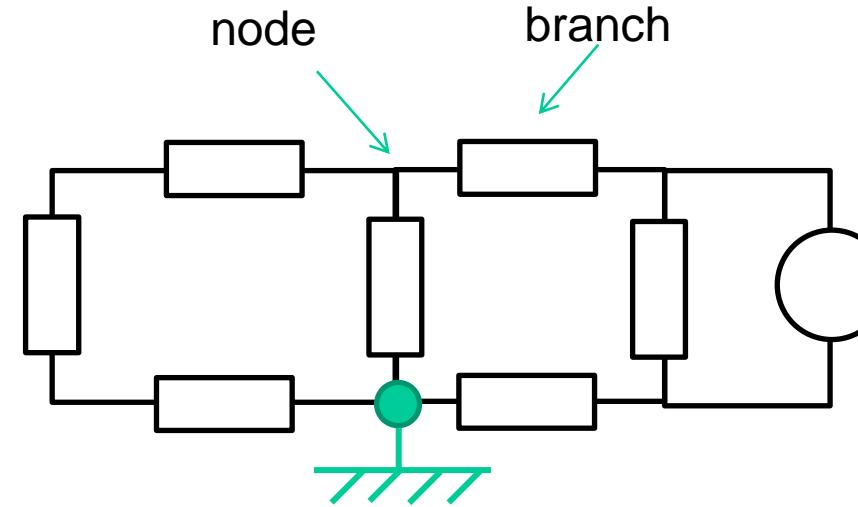
- 1) Identify the relevant independent circuit variables
- 2) Write the circuit Lagrangian
- 3) Determine the canonical conjugate variables and the Hamiltonian

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Quantronics group

# Hamiltonian of an arbitrary circuit

$$\boxed{\quad} = \begin{cases} \text{---} || \text{---} \\ -\text{---} \text{---} \text{---} \\ \times \\ \text{---} | \text{---} \\ \text{---} \text{---} \end{cases}$$



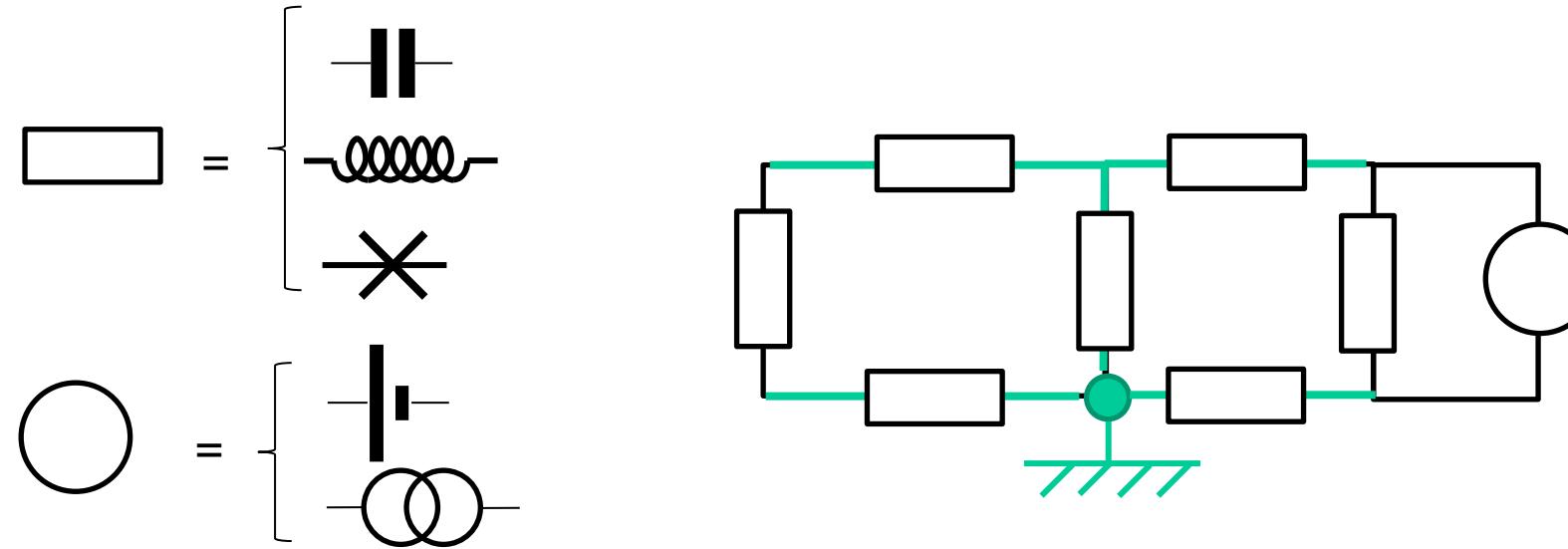
Identifying the **relevant independent circuit variables**

- 1) Choose reference node (ground)

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# Hamiltonian of an arbitrary circuit



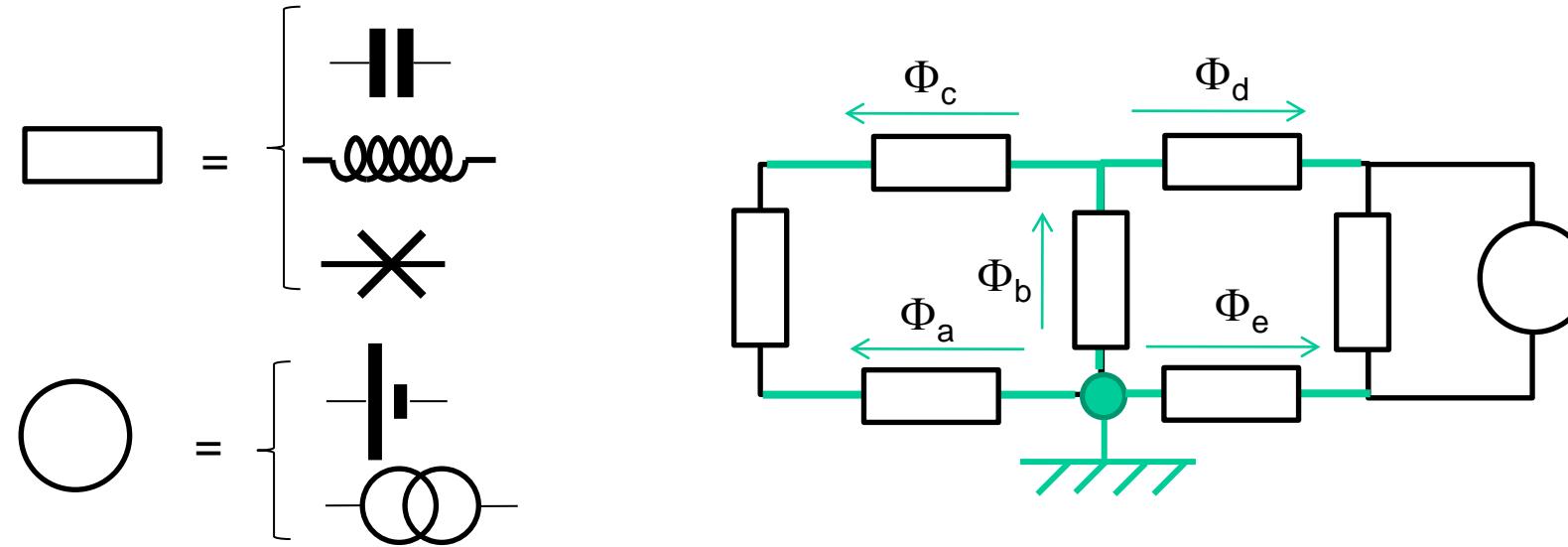
Identifying the **relevant independent circuit variables**

- 1) Choose reference node (ground)
- 2) Choose « spanning tree » (no loop)

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Quantronics group

# Hamiltonian of an arbitrary circuit



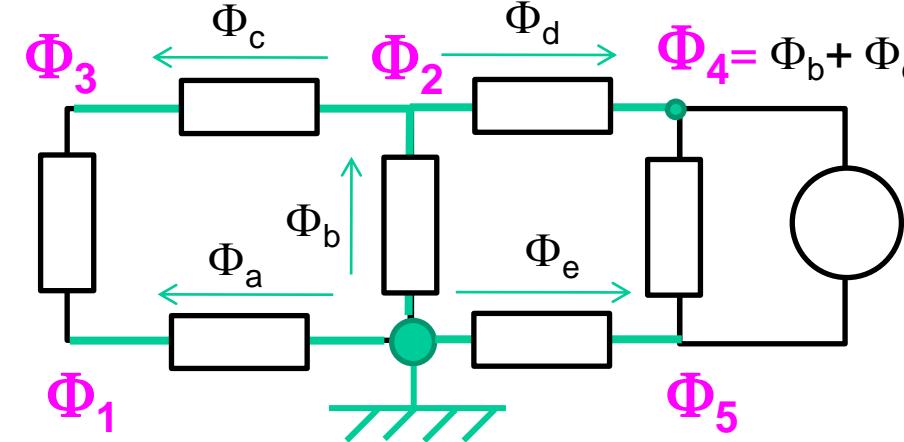
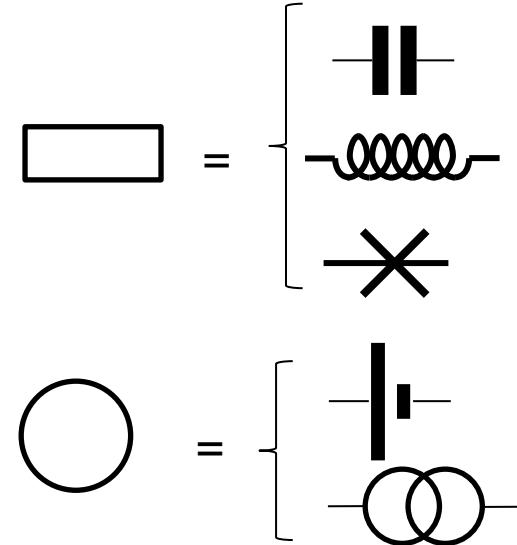
Identifying the **relevant independent circuit variables**

- 1) Choose reference node (ground)
- 2) Choose « spanning tree » (no loop)
- 3) Define « tree branch fluxes »  $\Phi_i(t) = \int_{-\infty}^t V(t')dt'$

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# Hamiltonian of an arbitrary circuit



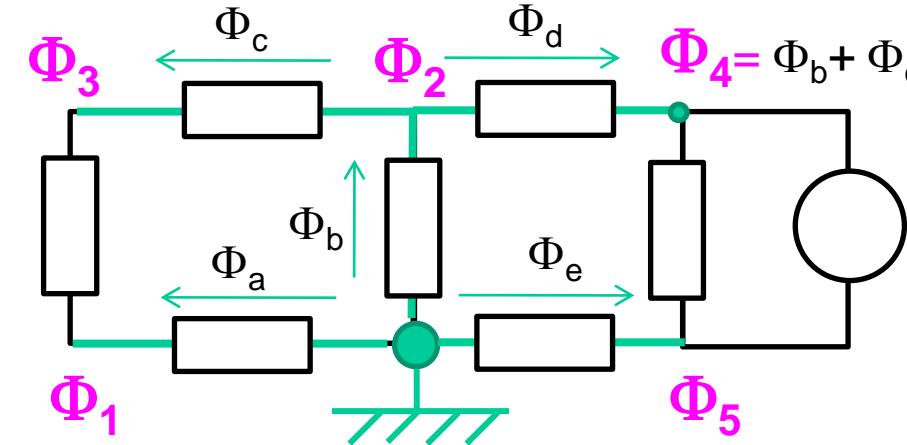
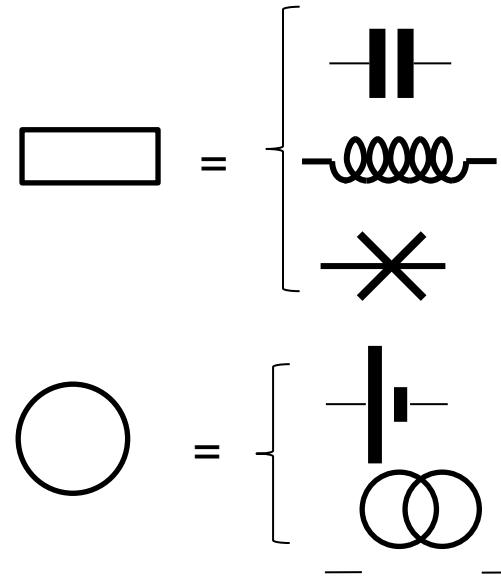
Identifying the **relevant independent circuit variables**

- 1) Choose reference node (ground)
- 2) Choose « spanning tree » (no loop)
- 3) Define « tree branch fluxes »  $\Phi_i(t) = \int_{-\infty}^t V(t')dt'$
- 4) Define **node fluxes** = sum of branch fluxes from ground

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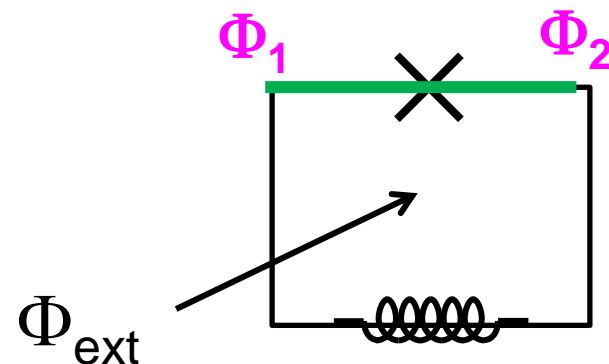
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# Hamiltonian of an arbitrary circuit



Write Lagrangian  $L(\Phi_i, \dot{\Phi}_i) = L_{el}(\dot{\Phi}_i) - L_{pot}(\Phi_i)$

taking into account constraints imposed by external biases (fluxes or charges)

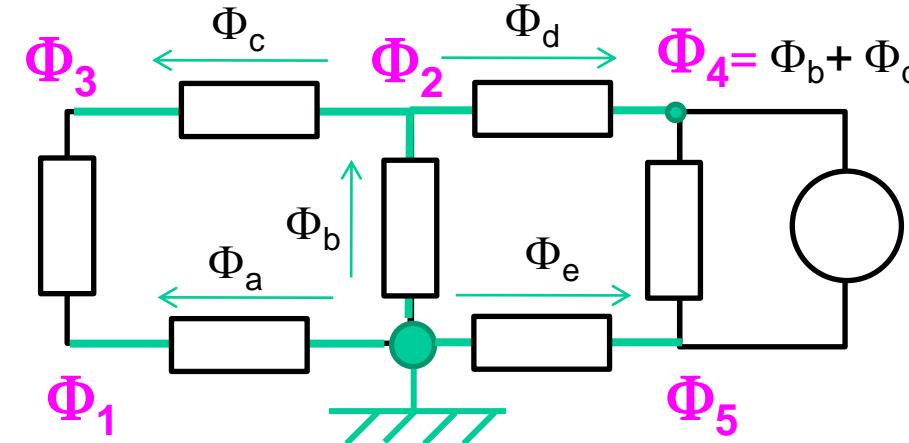
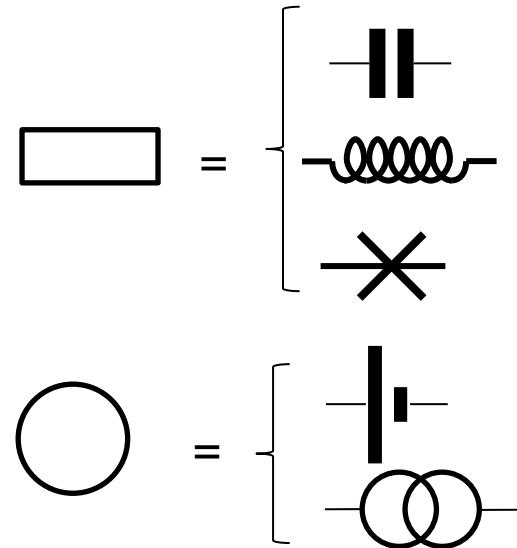


$$L_{pot} = E_J \cos(\Phi_2 - \Phi_1) - \frac{(\Phi_1 - \Phi_2 - \Phi_{ext})^2}{2L}$$

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Quantronics group

# Hamiltonian of an arbitrary circuit



Conjugate variables :  $Q_i = \frac{\partial L}{\partial \dot{\Phi}_i}$

→ Classical Hamiltonian  $H(\Phi_i, Q_i) = \sum Q_i \dot{\Phi}_i - L$

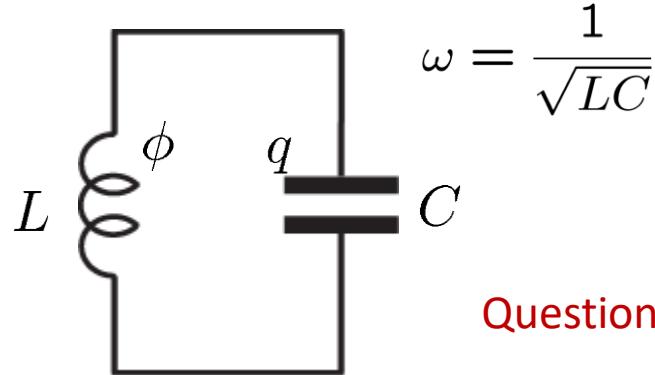
→ Quantum Hamiltonian  $H(\hat{\Phi}_i, \hat{Q}_i)$  With  $[\hat{\Phi}_i, \hat{Q}_i] = i\hbar$

or  $H(\hat{\theta}_i, \hat{n}_i)$  with  $\hat{n}_i = \hat{Q}_i / 2e$   
 $\hat{\theta}_i = \hat{\Phi}_i (2e / \hbar)$

$$[\hat{\theta}_i, \hat{n}_i] = i$$

# Classical and quantum states of the harmonic oscillator

harmonic LC oscillator:

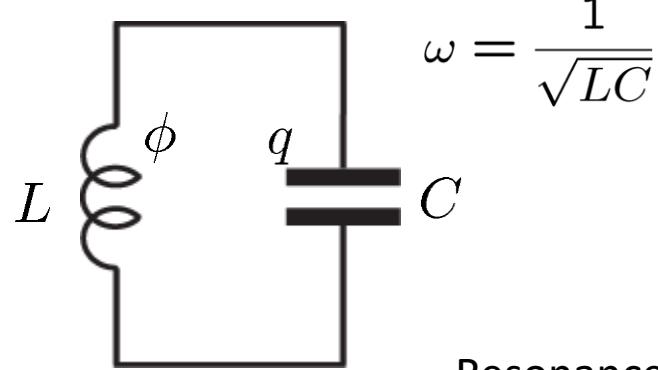


**Question:** take 1 nF capacitor and 1 nH inductor and solder an LC oscillator. Why we don't see quantum effects with it?



# Classical and quantum states of the harmonic oscillator

harmonic LC oscillator:



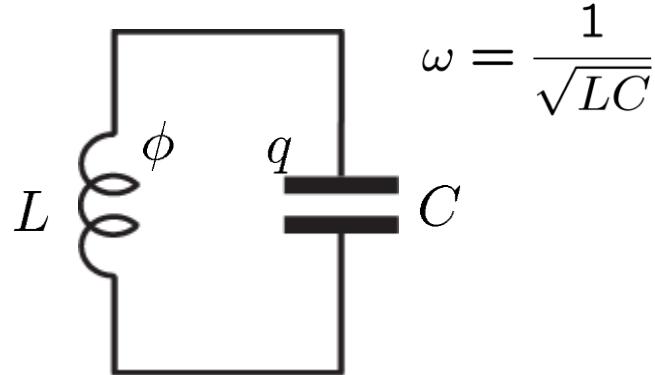
Resonance frequency of this resonator is  $f \simeq 150$  MHz

If compare to thermal energy:

$$\frac{k_B 300K}{\hbar \omega} \simeq 40\,000$$

# Classical and quantum states of the harmonic oscillator

harmonic LC oscillator:



$$\omega = \frac{1}{\sqrt{LC}}$$

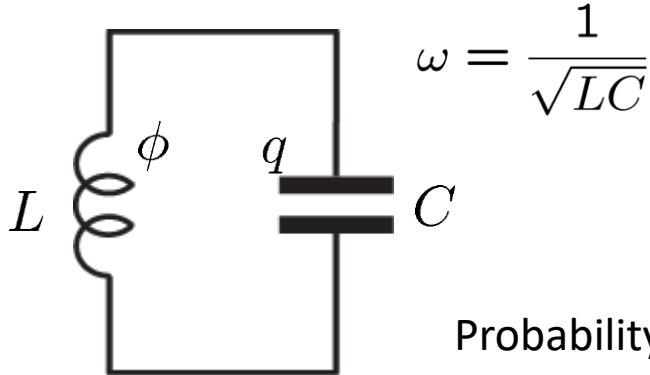
**Question:** what state will the oscillator attain if put in contact with environment with temperature  $T$ ?

**Note that due to the Boltzmann distribution:** The probability to find the state of the system with energy  $E_i$  is given by

$$p_i = \frac{e^{-\frac{E_i}{k_B T}}}{\sum_i e^{-\frac{E_i}{k_B T}}}$$

# Classical and quantum states of the harmonic oscillator

harmonic LC oscillator:



$$\omega = \frac{1}{\sqrt{LC}}$$

Probability to find  $n$  photons is  $p_n = \frac{e^{-\frac{\hbar\omega(n+\frac{1}{2})}{k_B T}}}{\sum_n e^{-\frac{\hbar\omega(n+\frac{1}{2})}{k_B T}}}$

Let's calculate the sum using the geometric progression formula:

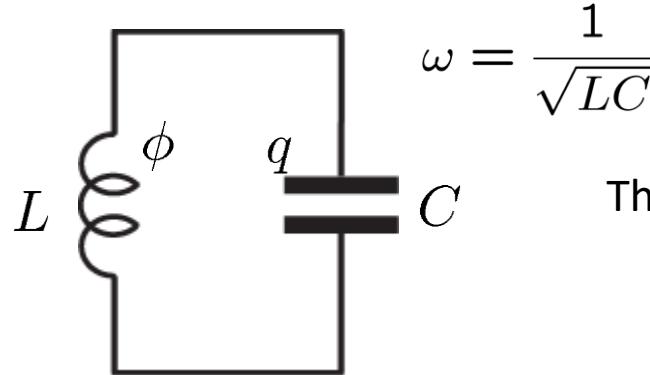
$$\sum_n e^{-\frac{\hbar\omega n}{k_B T}} = \frac{1}{1 - e^{-\frac{\hbar\omega}{k_B T}}}$$

The result is:

$$p_n = e^{-\frac{\hbar\omega n}{k_B T}} (1 - e^{-\frac{\hbar\omega}{k_B T}})$$

# Classical and quantum states of the harmonic oscillator

harmonic LC oscillator:

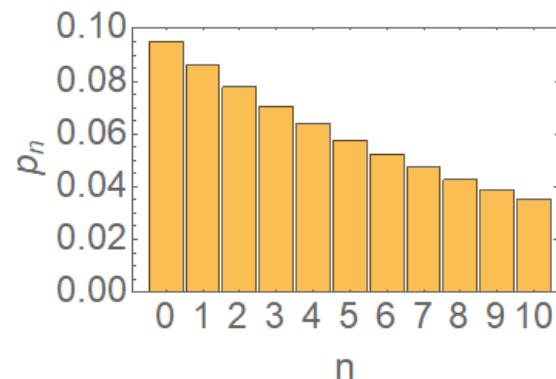


$$\omega = \frac{1}{\sqrt{LC}}$$

The HO oscillator will be described by the density operator:

$$\rho_{th} = \begin{pmatrix} p_0 & 0 & \dots \\ 0 & p_1 & \dots \\ \vdots & \vdots & \ddots \end{pmatrix} \text{ with } p_n = e^{-\frac{\hbar\omega n}{k_B T}} (1 - e^{-\frac{\hbar\omega}{k_B T}})$$

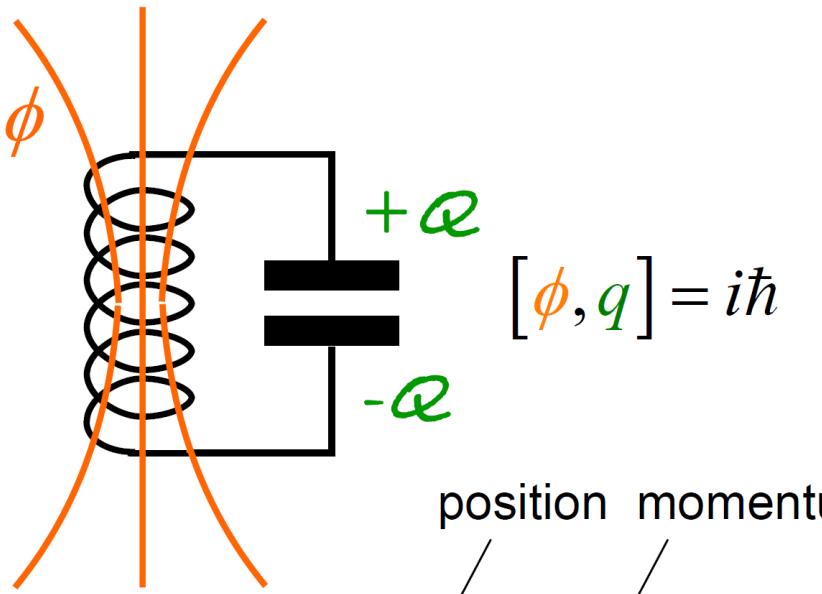
For  $\frac{k_B T}{\hbar\omega} = 10$  (average number of photons):



Statistical mixture not a pure state!  
-> no quantum effects

There is only one thermal state which is a pure state:  
Ground state at T = 0 K

# Classical and quantum states of the harmonic oscillator

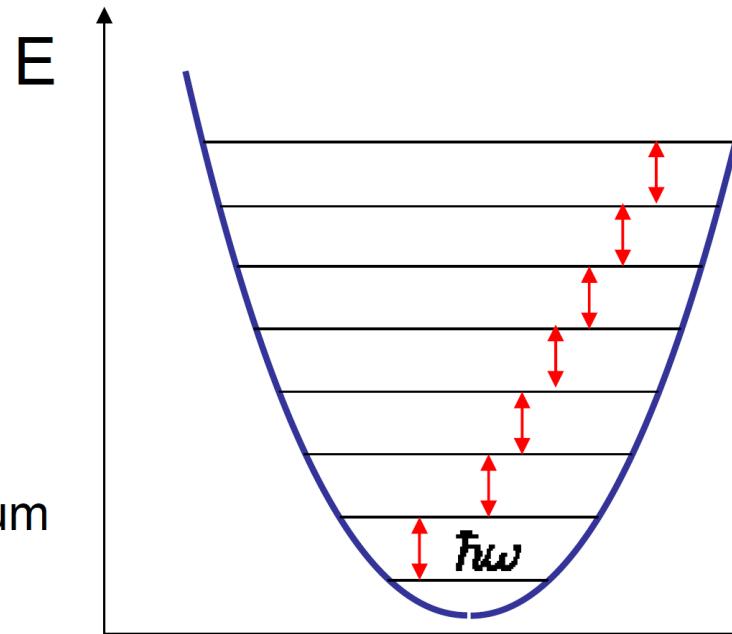


Hamiltonian

$$H = \frac{\phi^2}{2L} + \frac{q^2}{2C}$$

$$\omega = 1/\sqrt{LC}$$

$$H = \hbar\omega \left( a^\dagger a + \frac{1}{2} \right)$$



low temperature required:

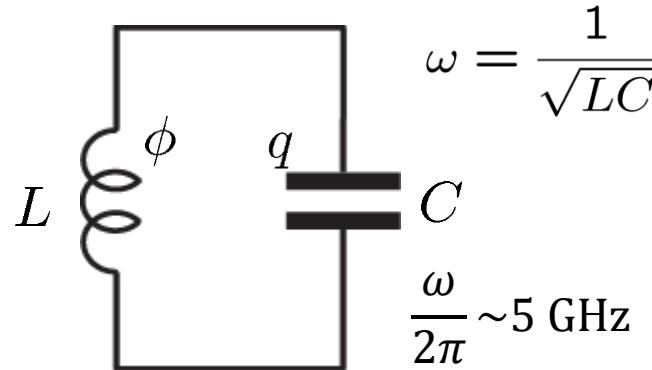
$$\hbar\omega \gg k_B T$$

→                  ←  
10 GHz ~ 500 mK      20 mK

$$\langle n_{\text{th}} \rangle = \frac{1}{\exp(h\nu/k_B T) - 1} \sim 10^{-11}$$

# Different Flavors of Superconducting Resonators

harmonic LC oscillator:

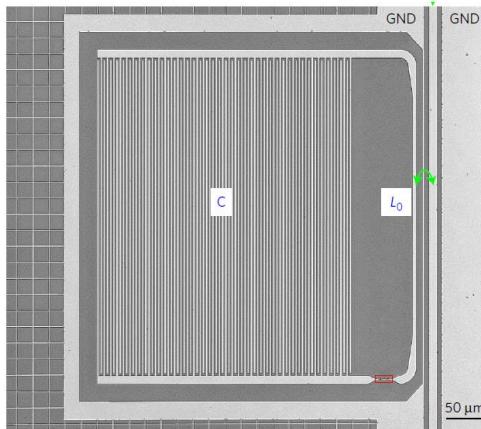


To observe quantum effects we need to initialise an oscillator in a **pure** state.

Use dilution cryostat can be cooled down to its ground state at

$$T = 20 \text{ mK } (\frac{k_B T}{\hbar \omega} = 0.1)$$

**lumped element resonator:**



Use nanofabrication to shrink the dimensions and increase the frequency

F. Yoshihara *et al.*, *Nature Physics* (2017)

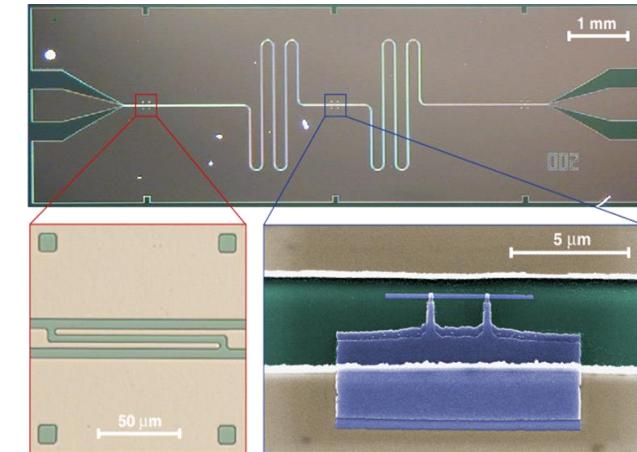
**3D cavity:**



Use 3D cavities as oscillators. Their dimension defines the frequency

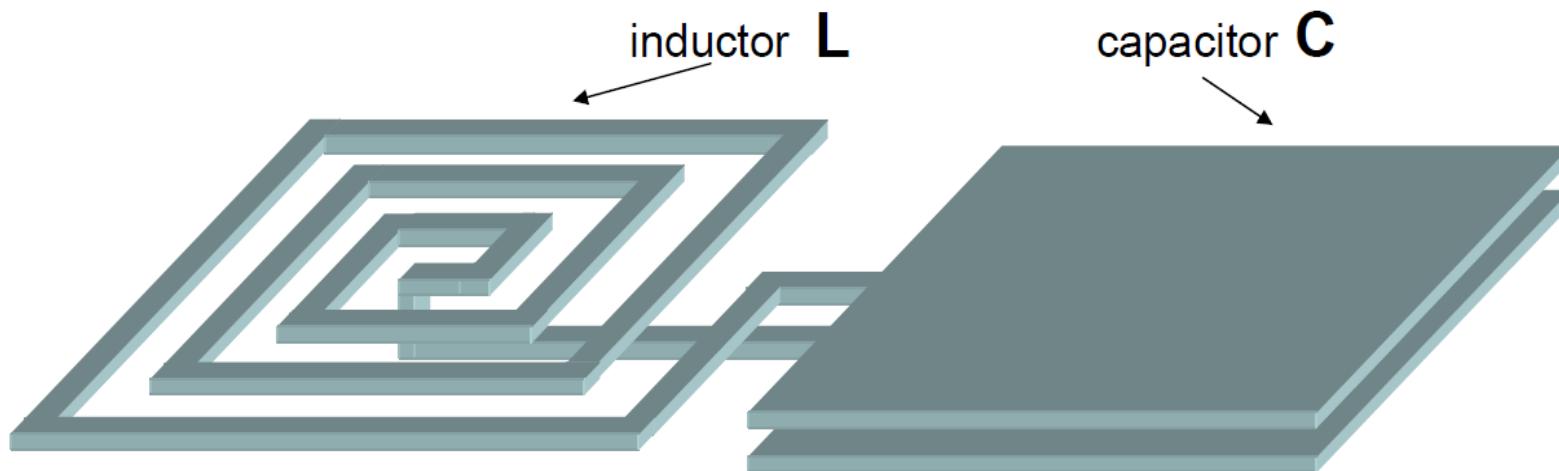
M. Jerger *et al.*, *Nature Comm.* 7, 12930 (2016)

**planar transmission line:**

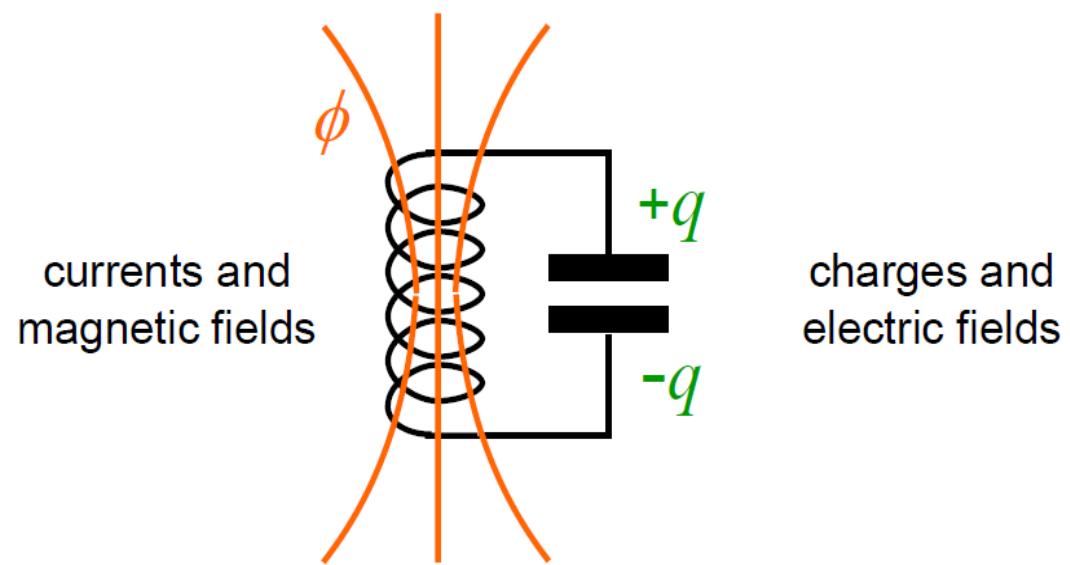


Wallraff *et al.*, *Nature* 431, 162 (2004)

# Lumped element resonator



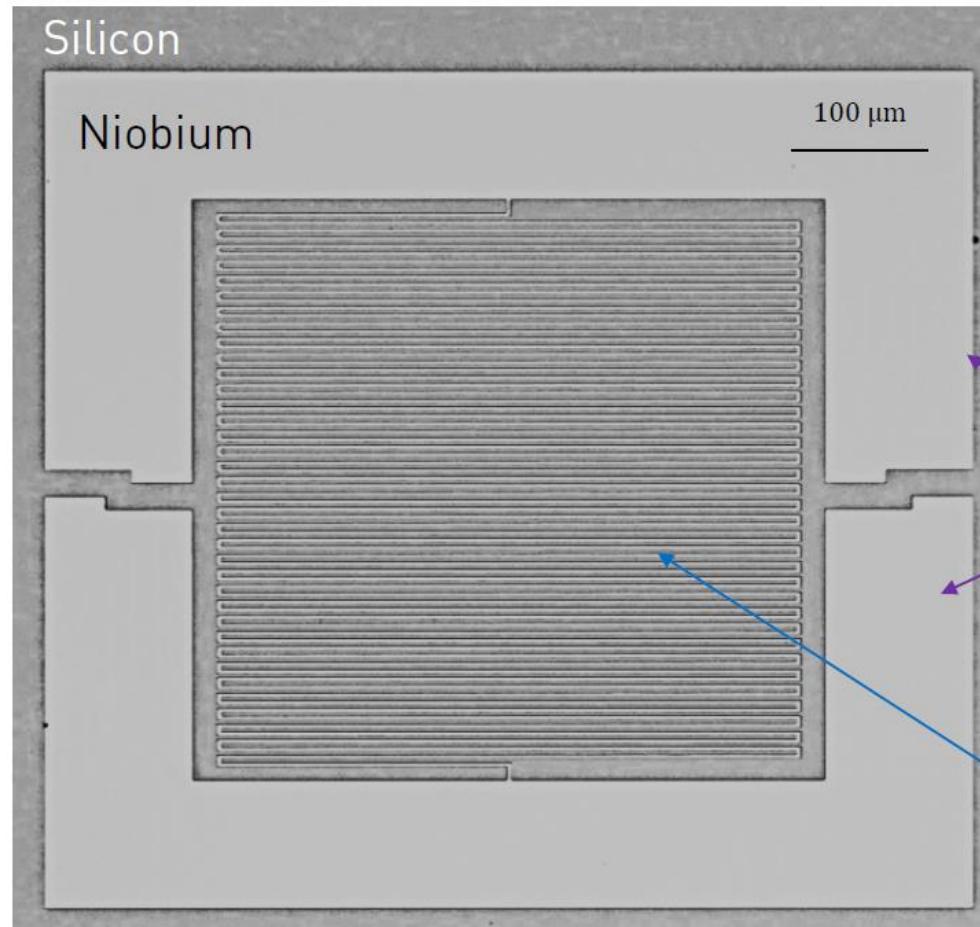
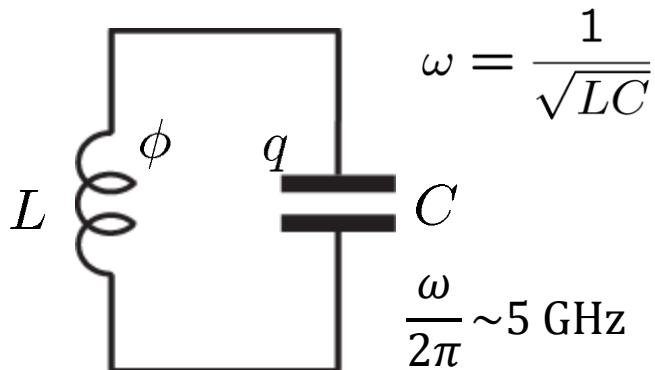
a harmonic oscillator



# Lumped element superconducting resonators

- Resonators fabricated from a 150 nm thin Niobium film on a Silicon substrate.
- Inductance and capacitance controlled by designed geometry.
- Characteristic size much smaller than characteristic wavelength.
- Accurate prediction of resonance frequency needs to take stray capacitance of inductor into account.
- Measured in a metallic cavity with a sender (IN) and receiver (OUT) antenna.

harmonic LC oscillator:



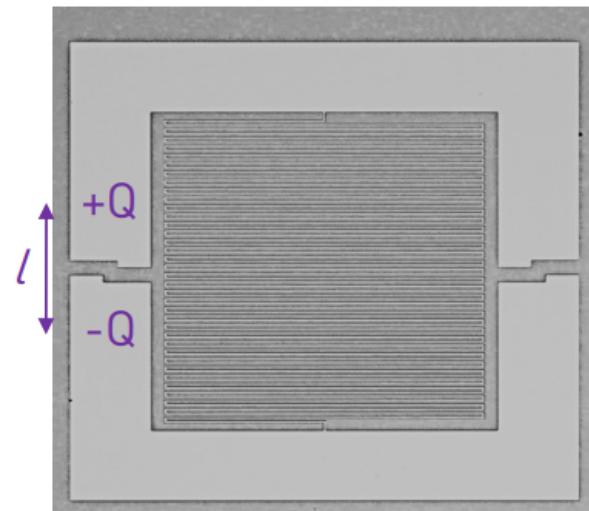
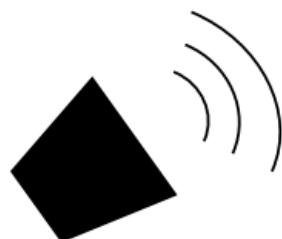
Resonance frequencies typically between 4 – 8 GHz

Two plates forming a capacitor.  
Typical values:  
 $C \sim 50 - 500 \text{ fF}$

Meandering wire forming an inductor.  
Typical values  
 $L \sim 2 - 20 \text{ nH}$

# Coupling Planar Lumped Element Resonators to an E-field

Apply  
Electromagnetic  
field at  
resonator  
frequency  $\omega_0$



Electric field  $\underline{E}$



Note: Applying  $H_d$  for short duration and with  $\underline{E}(t) = \underline{E}_0 \cos(\omega_0 t)$  generates a coherent state:

$$|0\rangle \rightarrow |\alpha\rangle$$

- Coupling mediated by dipole-field interaction

$$H_d = \underline{d} \cdot \underline{E}$$

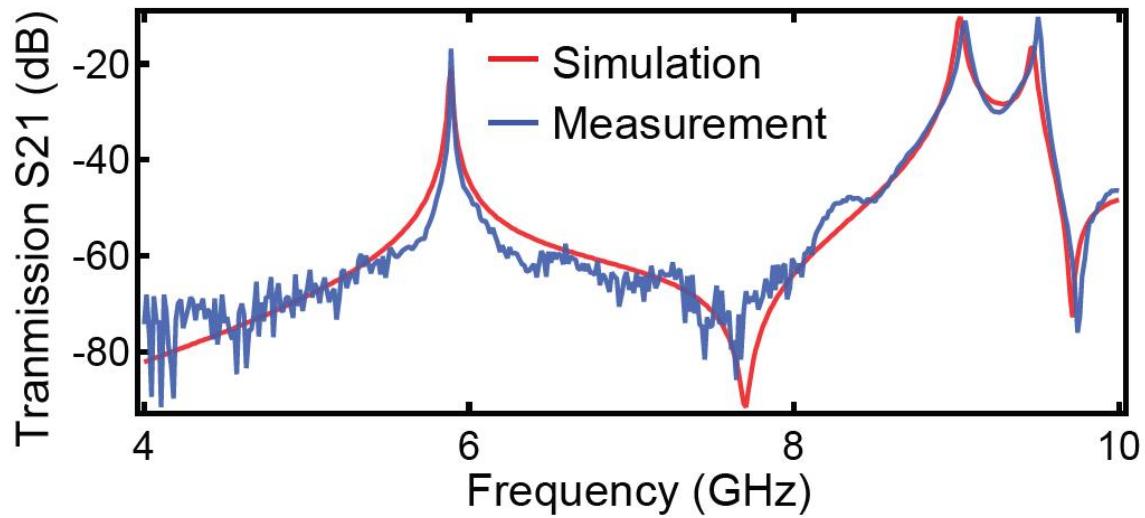
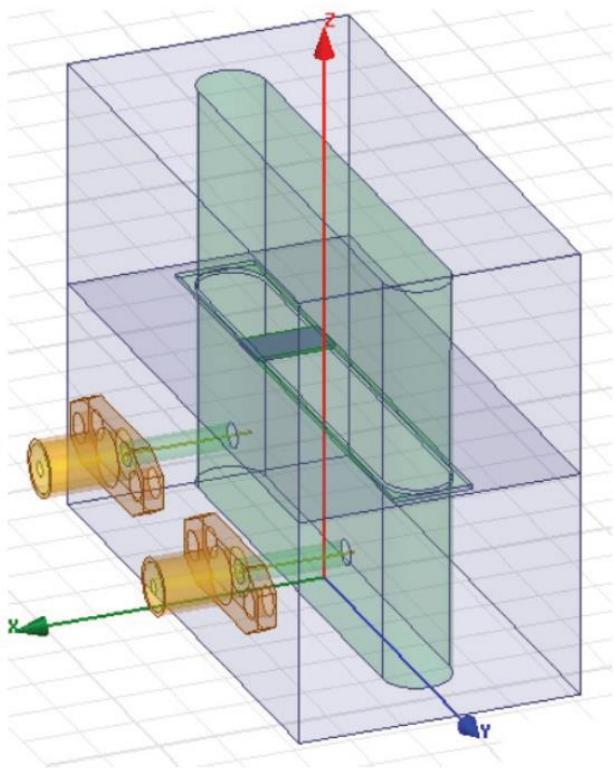
- Dipole moment

$$\hat{\underline{d}} = \int d^3r \underline{r} \rho(\underline{r}) \equiv \hat{Q} \underline{l}$$

- With average charge separation  $\underline{l}$ .
- For an LC resonator we thus obtain

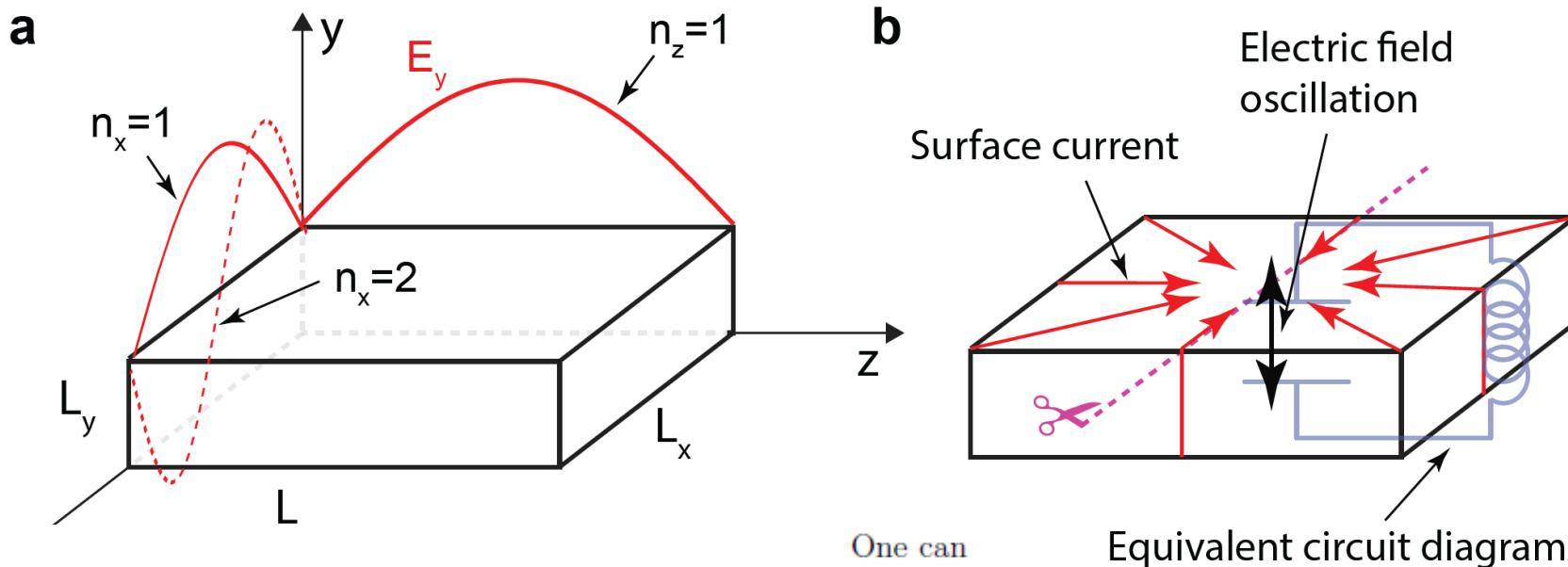
$$H_d = \underline{l} \cdot \underline{E} Q_{\text{zpf}}(a + a^\dagger)$$

# Aluminium 3D cavity



**HFSS simulation for cavity transmission:** The cavity transmission is simulated by HFSS (red curve) which is in agreement with actual measurement (blue curve).

# Electromagnetic Modes of a 3D cavity



show that for a 3D cavity described by  $L_x, L_y, L_z$  dimensions and depicted in Figure 3.1, the Equation 2.1.2 simply generalized to,

$$E(\vec{r}, t) = \mathcal{E} q(t) \sin(\vec{k} \cdot \vec{r})$$

$$B(\vec{r}, t) = \mathcal{E} \frac{\mu_0 \epsilon_0}{k} \dot{q}(t) \cos(\vec{k} \cdot \vec{r}),$$

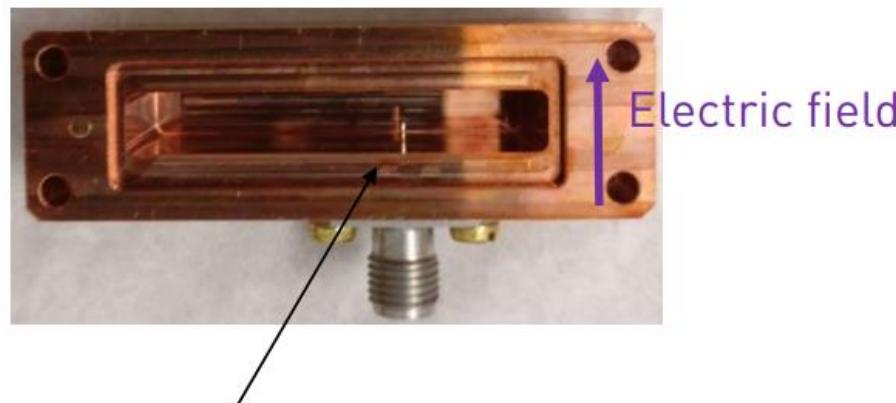
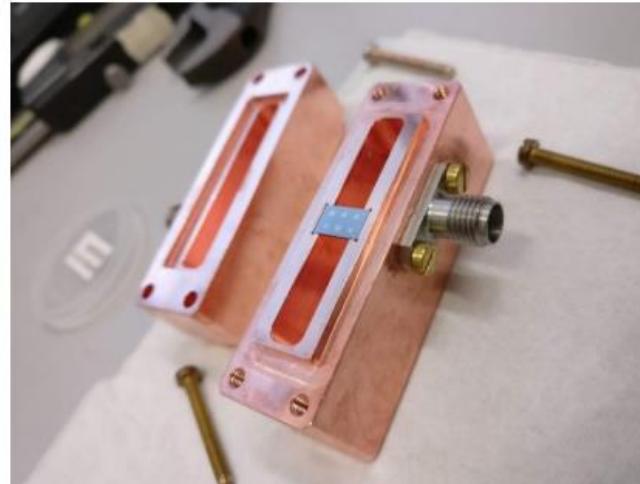
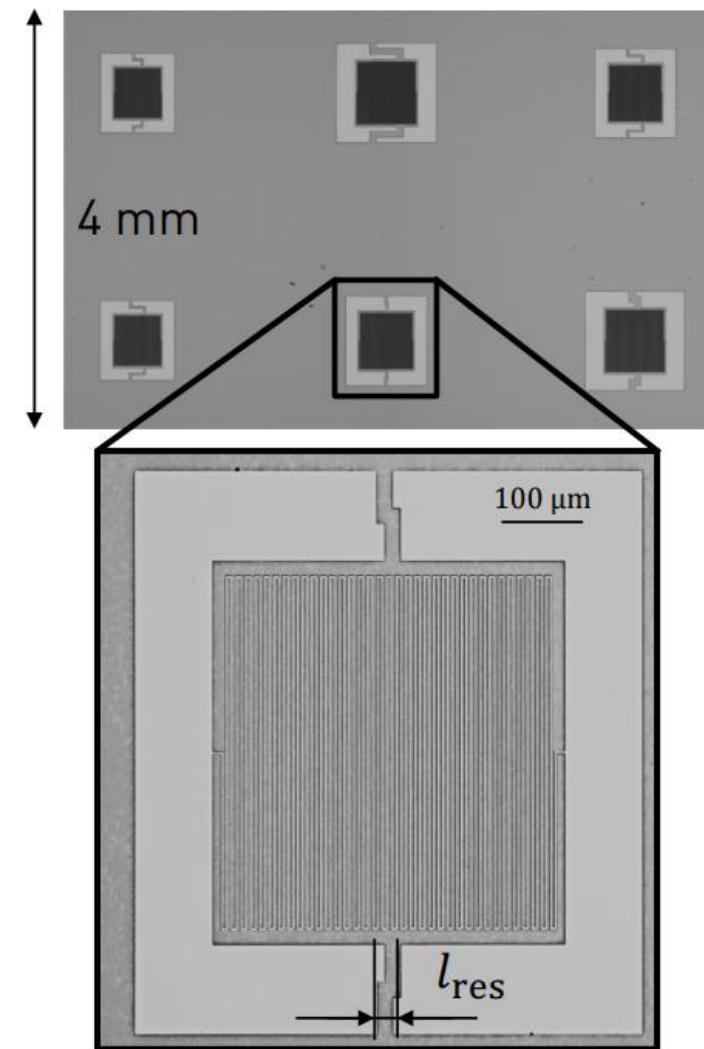
where  $\vec{k} = (n_x \pi / L_x, n_y \pi / L_y, n_z \pi / L_z)$  and  $\vec{r} = (x, y, z)$  and the corresponding resonance frequency of modes are,

$$f = \omega_c / 2\pi = \frac{c}{2} \sqrt{\left(\frac{n_x}{L_x}\right)^2 + \left(\frac{n_y}{L_y}\right)^2 + \left(\frac{n_z}{L_z}\right)^2}, \quad (3.1.2)$$

where  $c$  is the speed of light inside the cavity. Each mode is described by a set of integers  $n_x, n_y, n_z$ ,

typical cavity geometry is shown. Red lines (3.1.1a) TE<sub>101</sub> mode. The electric field oscillations (3.1.1b) oscillations (red arrows) have been depicted at the center (opposite charges at the top and bottom). The cut-line surface current is tangential.

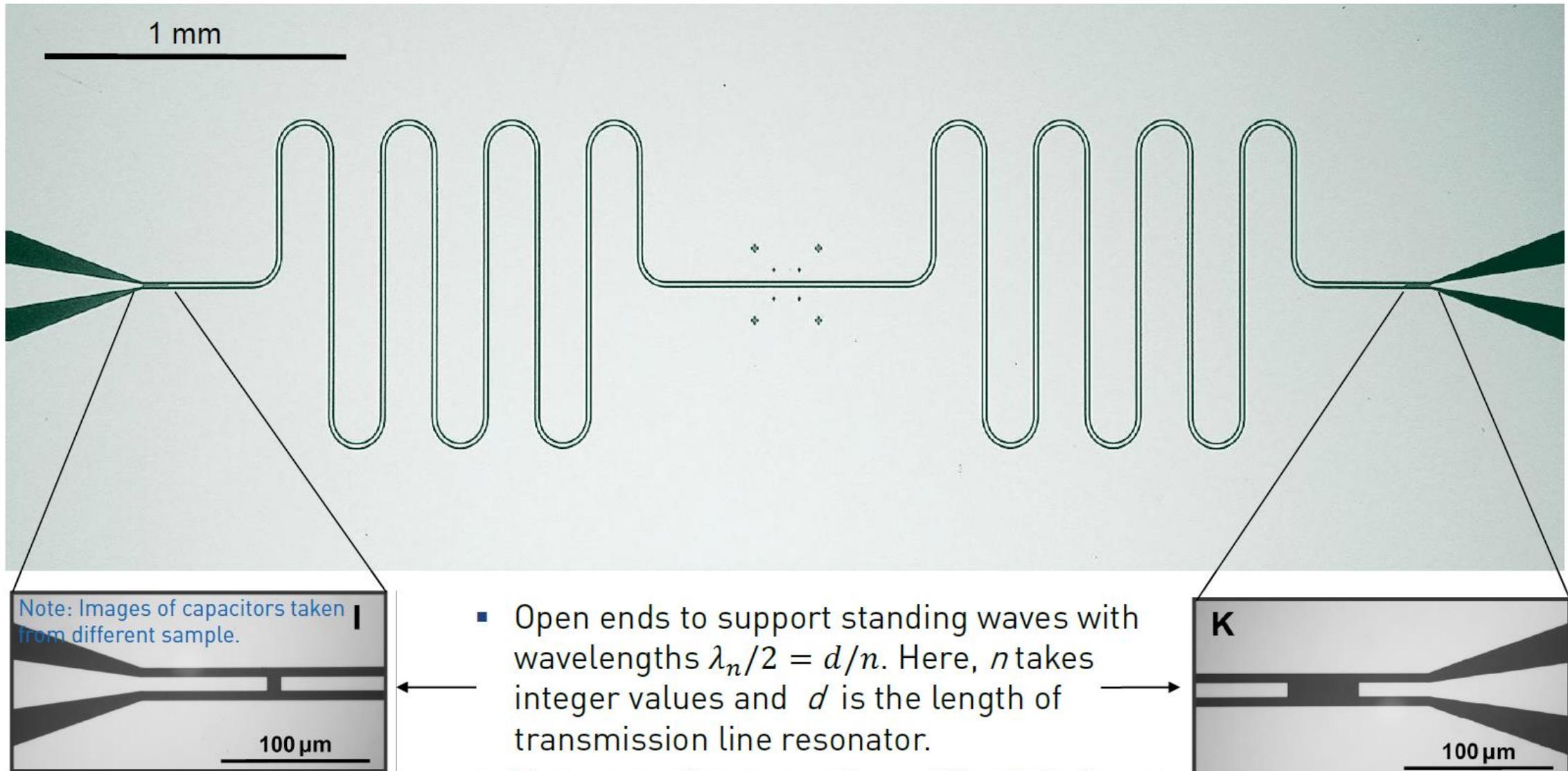
# Lumped Element Resonators in 3D Cavities



Properties of the system:

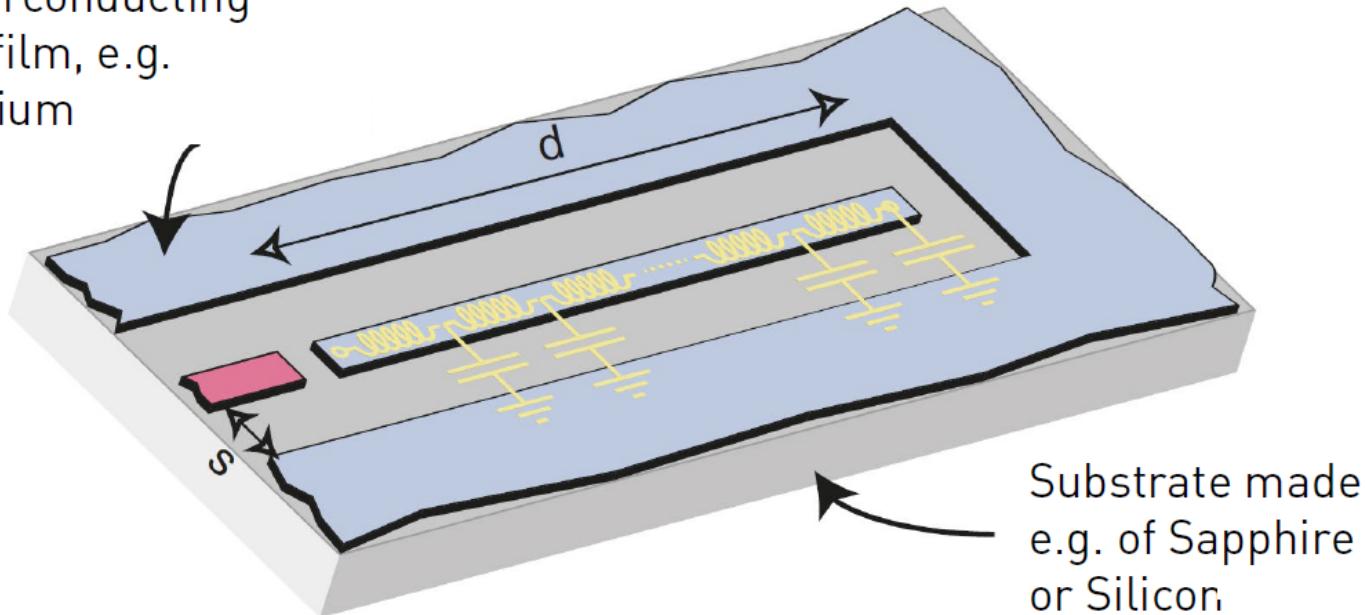
- multiple lumped element with different frequencies on a chip
- 3-Dimensional cavity, typically made of copper or aluminium
- TE101 mode of the cavity couples to the dipole moment of the resonators
- Inner conductor of the connector acts as an antenna to send and receive electromagnetic fields to/from the resonators
- Typical use case: Establish well-controlled environment to test material-related loss mechanisms.

# Distributed coplanar waveguide resonators



# Distributed coplanar waveguide resonators

Superconducting  
thin film, e.g.  
Niobium



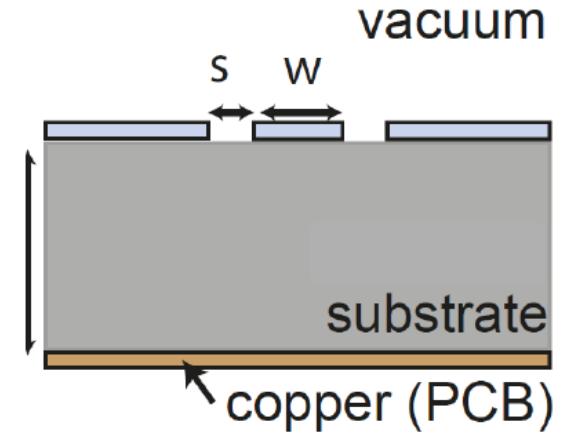
Substrate made  
e.g. of Sapphire  
or Silicon

Radiation field  
stored inside:

$$H = \hbar\omega a^\dagger a$$

Need nonlinear circuit  
element to build qubit.

harmonic  
oscillators

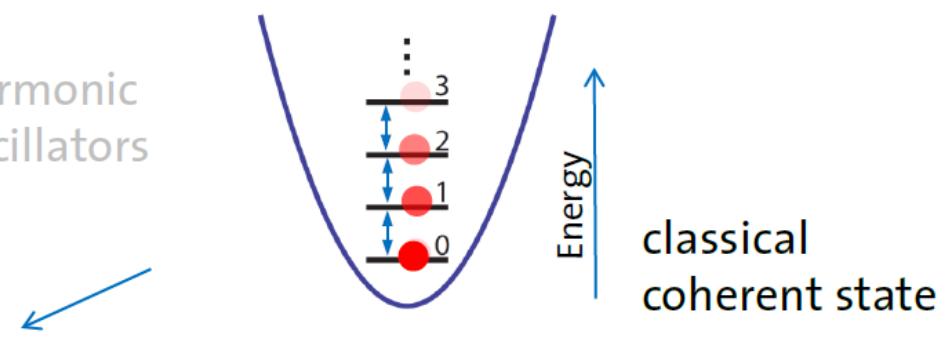


Typical dimensions:

$$s = 4.5 \mu m$$

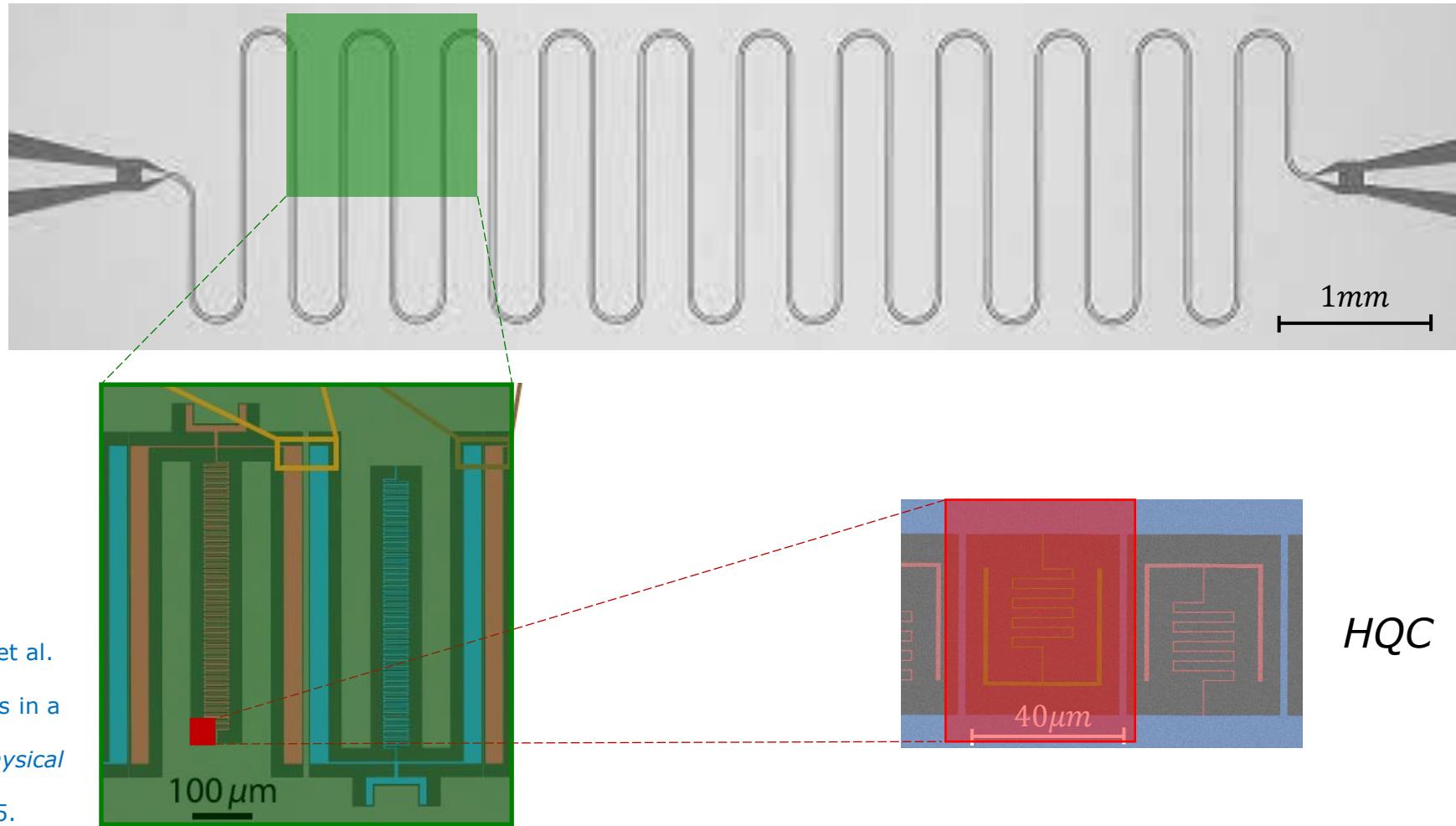
$$w = 10 \mu m$$

$$d \sim 5 mm$$

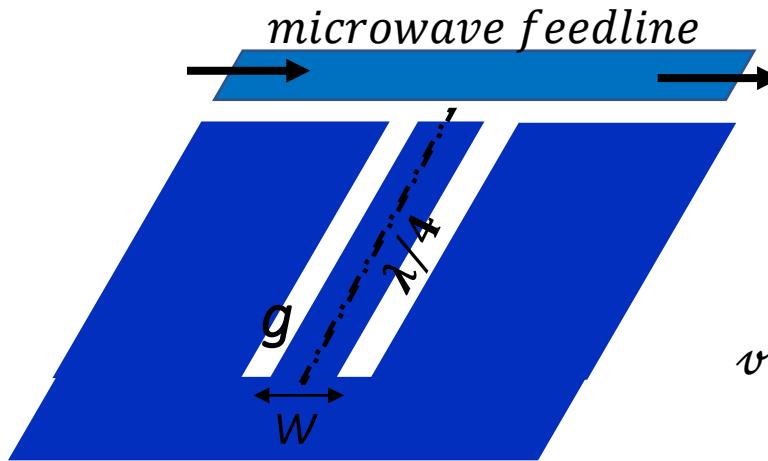


# Comparison to other kind of microwave resonators

Göppl, Martin, et al. "Coplanar waveguide resonators for circuit quantum electrodynamics." *Journal of Applied Physics* 104.11 (2008): 113904.



# Compact High Impedance resonators (distributed)



Let's consider a  $\lambda/4$  resonator:

$v_{ph} = 1/\sqrt{\mathcal{LC}}$  is the phase velocity

$$\ell_r \propto v_{ph}/4f_r$$

For a typical CPW  $50\ \Omega$  res. with  $w=10\ \mu m$  and gap  $g=6\ \mu m$  we have:

$\mathcal{L}_g \sim 420\ nH/m$  and  $\mathcal{C}_g \sim 0.166\ fF/m$  so  $v_{pf} \sim 120 * 10^6\ m/s$ .

A  $f_r \sim 6\ GHz$   $\lambda/4$  resonator will be  $\ell_r \sim 5\ mm$  long

$\ell_r$  = resonator length

$\mathcal{L}$  = inductance per unit length

$\mathcal{C}$  = capacitance per unit length

$$Z_r = \sqrt{\mathcal{L}/\mathcal{C}}$$

$$\mathcal{C} = \mathcal{C}_g$$

$$\mathcal{L} = \mathcal{L}_g + \mathcal{L}_k$$

kinetic  
inductance

geometric  
inductance

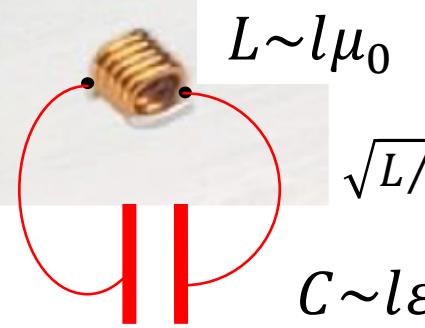
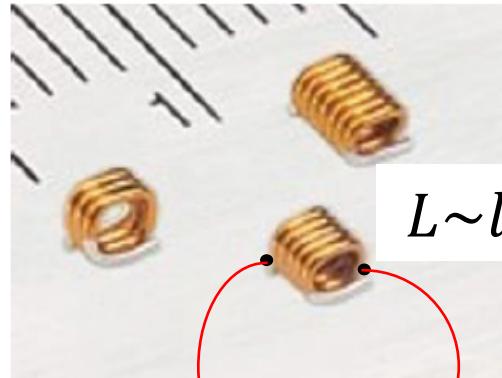
The kinetic inductance of disordered superconducting thin film nanowire resonator (with  $w \sim 200\ nm$ ) can be  $\mathcal{L}_k \sim 1000\ \mathcal{L}_g$  so the length od the resonator can be  $\sim 33$  times shorter

# Need for High Impedance:

ordinary inductors fail before their impedance exceeds  $R_Q$

$$g \propto V_0^{\text{rms}} \propto \sqrt{Z_0}$$

**Coilcraft**



$$\omega_0 = 1/\sqrt{LC} \quad \text{Self resonance frequency}$$

$$Z_0 \equiv \omega_0 L = \sqrt{L/C} \quad \text{Impedance}$$

$$R_Q = h/(2e)^2 \cong 6.5 \text{ k}\Omega$$

$$\alpha \sim 1/137$$

*fine structure constant*

*Vacuum  
Impedance*

$$\sqrt{L/C} \approx \sqrt{\mu_0/\epsilon_0} = 8 \alpha R_Q \equiv Z_{vac} = 377 \Omega$$

$$C \sim l \epsilon_0$$

$$Z_0 \approx Z_{vac}$$

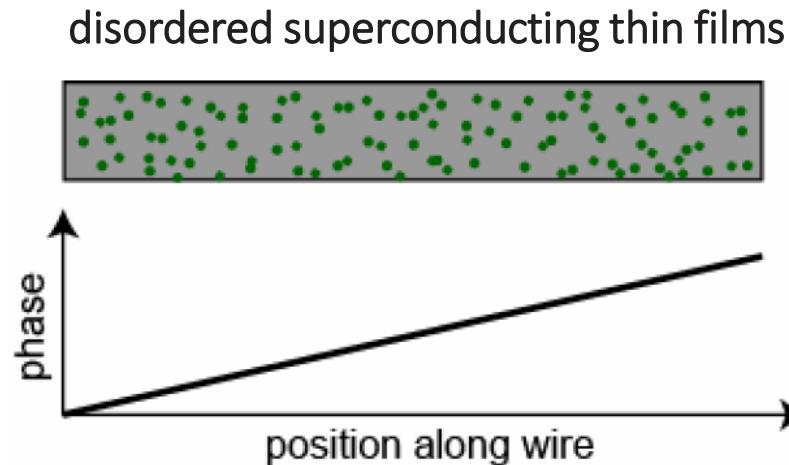
*How can we engineer an environment  
with  $\alpha >> 1/137$ ?*

N. A. Masluk et al., *PRL* **109**, 137002 (2012)

V. Manucharyan's PhD thesis

# Different ways to achieve a high kinetic inductance

Credits  
VI. E. Manucharyan  
 PhD thesis



$$I = (2e)nAv$$

$$E_K = \frac{1}{2}(2m_e)nAdv^2 \equiv \frac{1}{2}L_K I^2$$

$$L_K = \frac{m_e}{2en} \frac{l}{A} = \mu_0 \lambda^2 \frac{l}{A}$$

narrow wire etched in thin film

disordered superconductor

disordered superconducting thin films can be made compatible with in-plane magnetic field

## TiN

- M. R. Vissers, et al., [Appl. Phys. Lett. 97, 232509 \(2010\)](#).  
 H. G. Leduc, et al., [Appl. Phys. Lett. 97, 102509 \(2010\)](#).  
 L. J. Swenson, et al., [J. Appl. Phys. 113, 104501 \(2013\)](#).  
 A. Shearow et al., [Appl. Phys. Lett. 113, 212601 \(2018\)](#)

## NbTiN

- R. Barends, et al., [Appl. Phys. Lett. 97, 033507 \(2010\)](#).  
 N. Samkharadze, et al., [Phys. Rev. Appl. 5, 044004 \(2016\)](#).  
 T. M. Hazard, et al., [Phys. Rev. Lett. 122, 010504 \(2019\)](#).

## Granular aluminum

- L. Grünhaupt, et al., [Phys. Rev. Lett. 121, 117001](#)  
 N. Maleeva et al., [Nat. Commun. 9, 3889 \(2018\)](#)

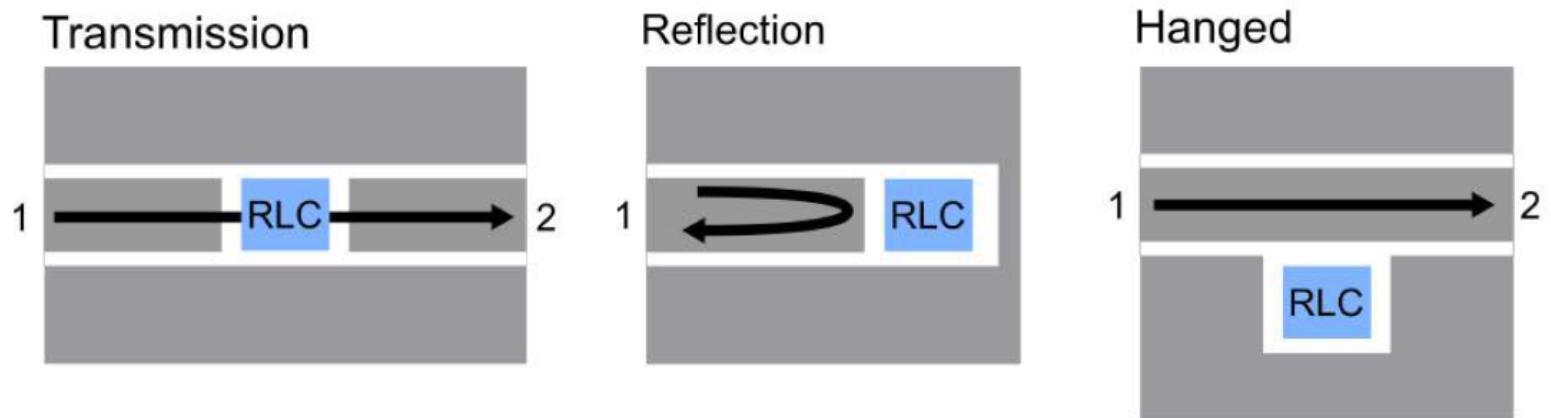
## NbN

- G. J. Grabovskij, et al., [Appl. Phys. Lett. 93, 134102 \(2008\)](#).  
 J. Luomahaara, et al., [Nat. Commun. 5, 4872 \(2014\)](#).  
 D. Niepce, et al., [Phys. Rev. Applied 11, 044014 \(2019\)](#).

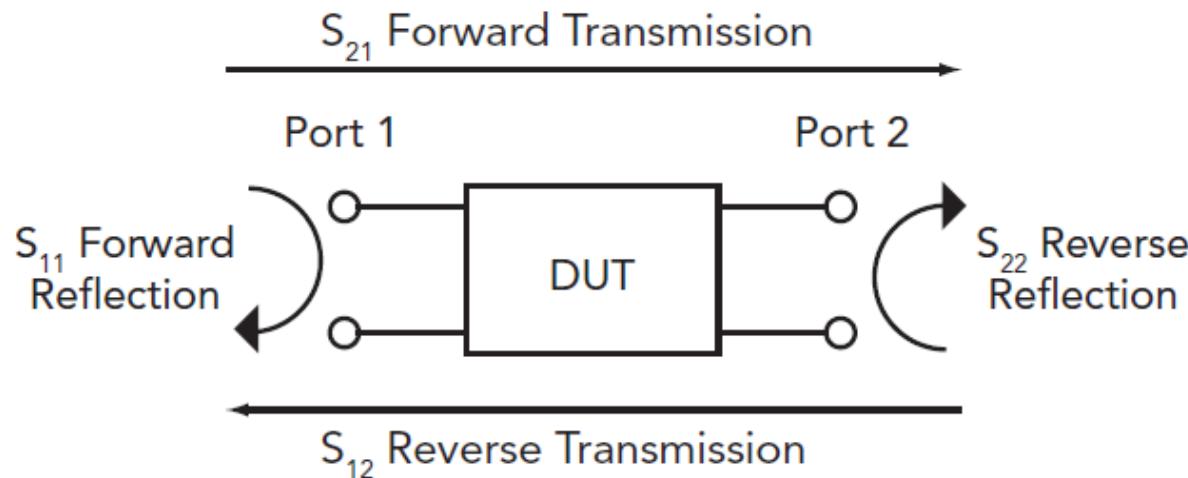
## InOx

- O. Dupre', et al., [Supercond. Sci. Technol. 30, 045007 \(2017\)](#).  
 S. E. de Graaf, et al., [Nat. Phys. 14, 590–594 \(2018\)](#).

# How to measure the scattering parameters of microwave resonators



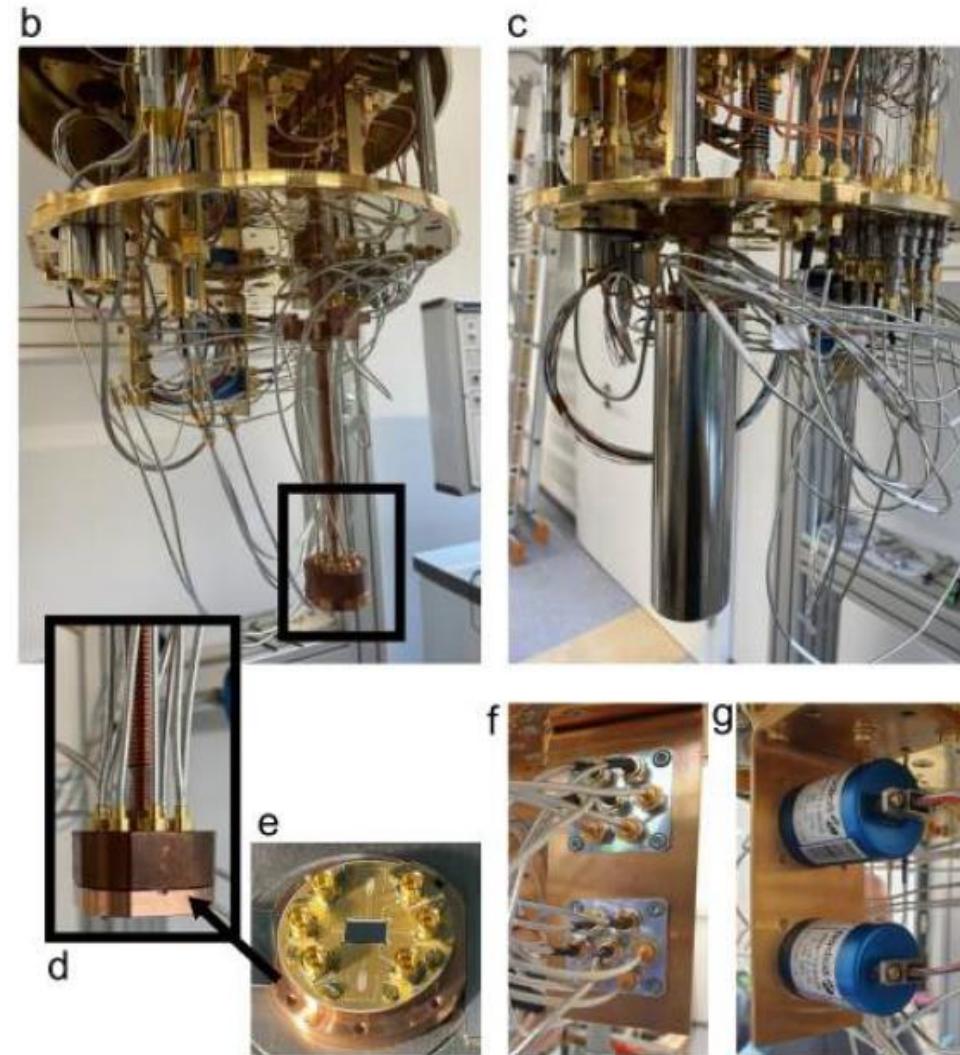
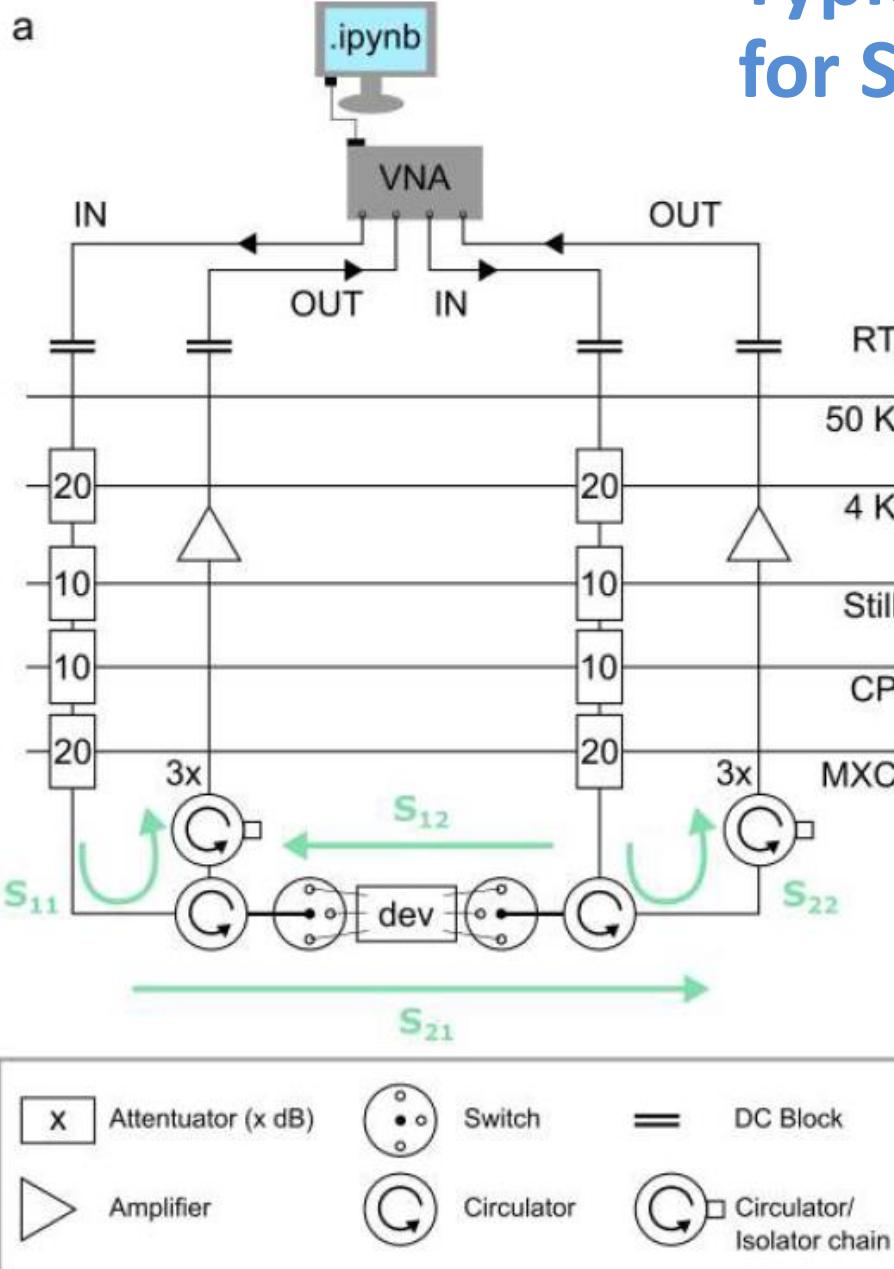
**Three Types of Resonator Measurement.** Transmission: The resonator is coupled to an input and an output feedline on each side. Reflection: The resonator is probed by coupling it to one end of the feedline. Hanged: The resonator couples to a feedline whose transmission is measured.



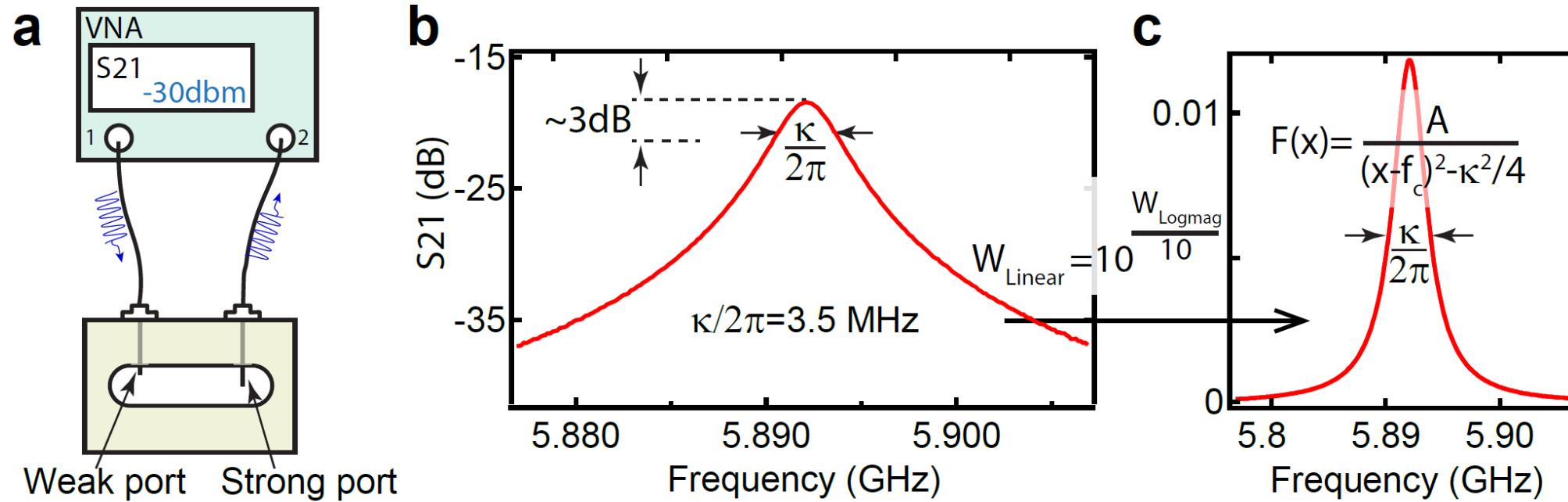
*The four scattering parameters are shown in this figure.*



# Typical Cryogenic measurement setup for Superconducting Resonators

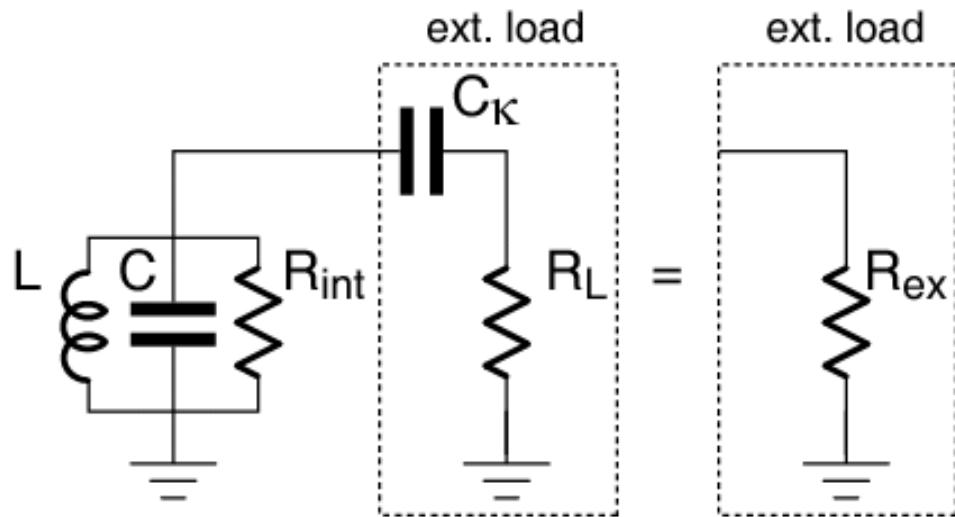


# How to measure the scattering parameters of microwave resonators



**The cavity linewidth characterization:** The cavity linewidth  $\kappa$  can be quantified by measuring the scattering parameters of the cavity. **a,b** In the transmission measurement  $S_{21}$ , the cavity linewidth can be estimated by the bandwidth that the transmission signal drops by  $3 \text{ dB}$ . **c**, More carefully, one can scale the transmission in the linear form and fit to the Lorentzian function. The FWHM of the Lorentzian function would be the cavity linewidth.

# Internal and External Dissipation in an LC Oscillator



internal losses: conductor, dielectric       $R_{\text{int}}$

external losses: radiation, coupling       $R_{\text{ext}}$

total losses       $\frac{1}{R} = \frac{1}{R_{\text{int}}} + \frac{1}{R_{\text{ext}}}$

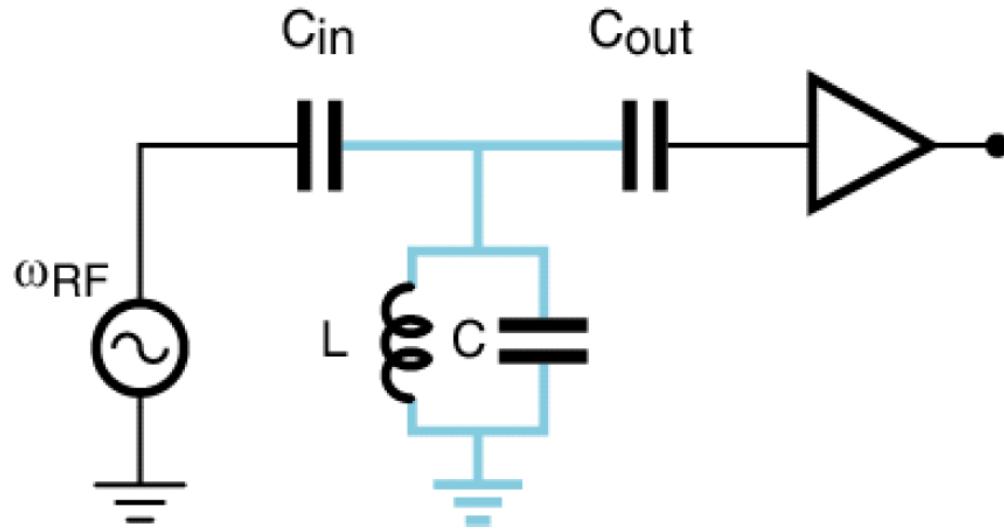
quality factor

$$Q = \frac{R}{Z} = \omega_0 R C \quad \text{with impedance} \quad Z = \sqrt{\frac{L}{C}}$$

excited state decay rate

$$\Gamma_1 = \frac{\omega_0}{Q} = \frac{1}{RC} \quad \begin{aligned} &\text{for } Q = 10^6 \text{ and } \omega_0 \sim 2\pi 1.5 \text{ GHz} \\ &\text{decay rate } \Gamma_1 = 10 \text{ kHz corresponding to lifetime } T_1 = 100 \mu\text{s} \end{aligned}$$

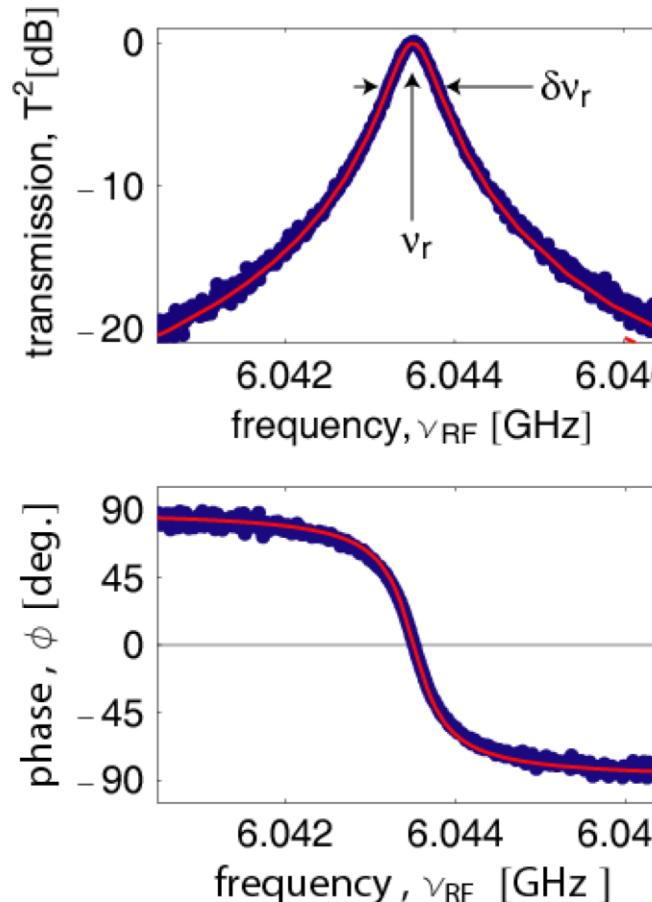
# Measure Transmission through a resonator



$$\frac{1}{Q} = \frac{1}{Q_L} = \frac{1}{Q_{int}} + \frac{1}{Q_{ext}}$$

$Q_{ext} \propto 1/\kappa_{ext}$  represents the effective controlled coupling of the resonator to the MW feedline

$Q_{int} \propto 1/\kappa_{int}$  represents the effective uncontrolled coupling of the resonator to the impurities or defects which can absorb the resonator energy.



resonance frequency:

$$\nu_r = 6.04 \text{ GHz}$$

quality factor:

$$Q = \frac{\nu_r}{\delta\nu_r} \approx 10^4$$

photon decay rate:

$$\frac{\kappa}{2\pi} = \frac{\nu_r}{Q} \approx 0.8 \text{ MHz}$$

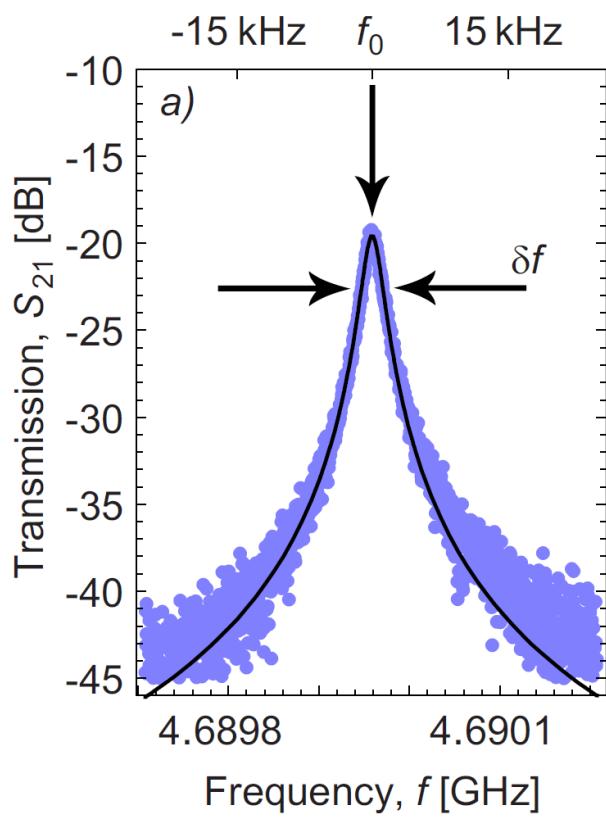
photon lifetime:

$$T_\kappa = 1/\kappa \approx 200 \text{ ns}$$

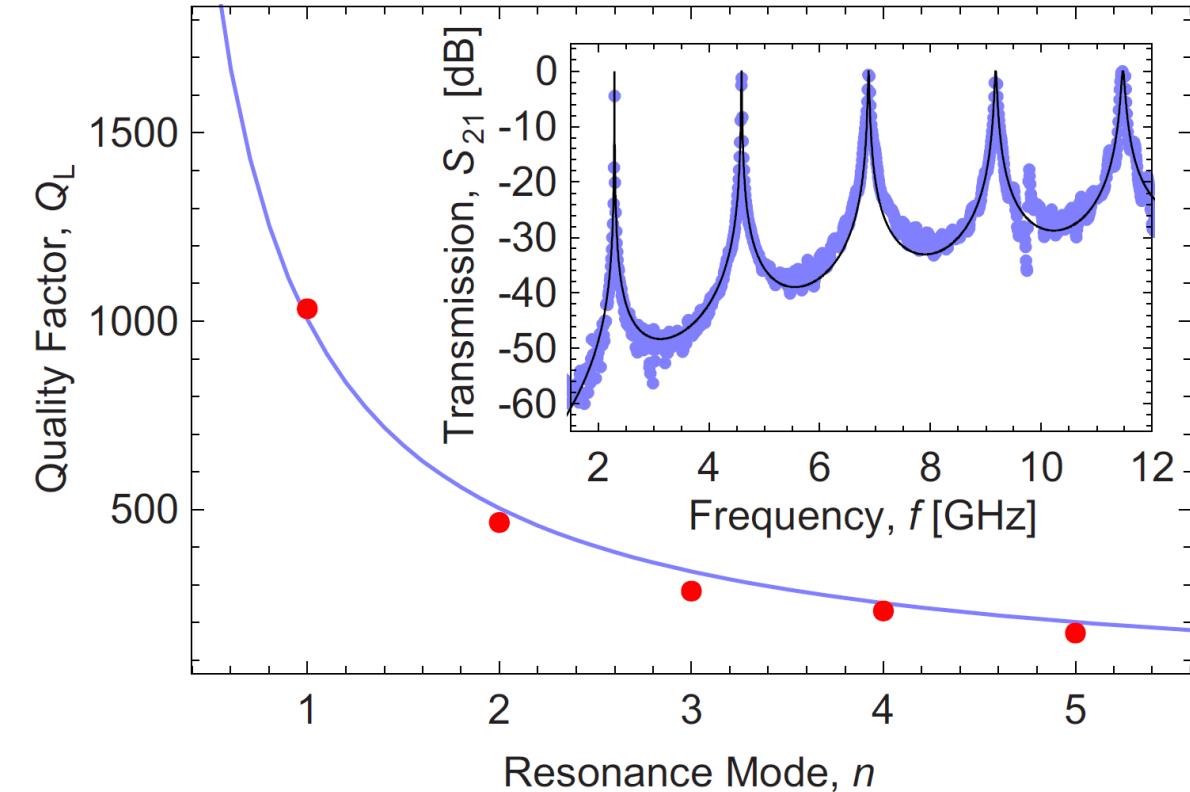
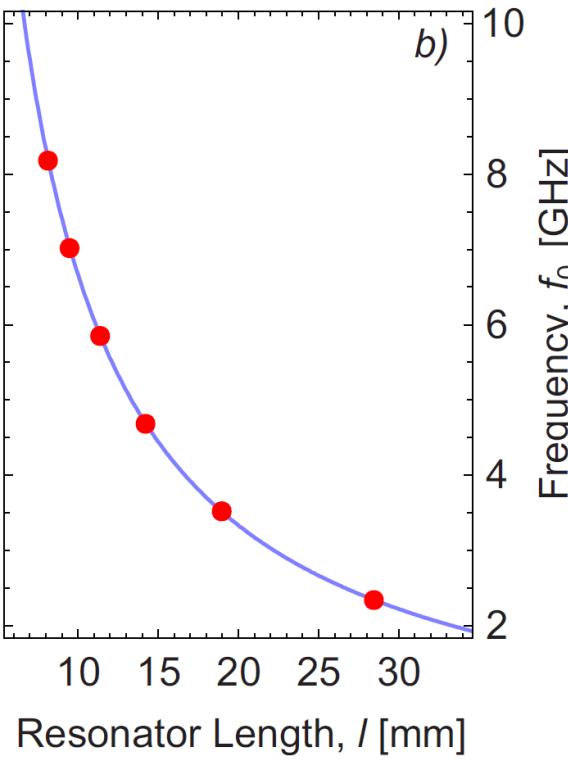
for  $Q = 10^6$  and  $\omega_0 \sim 2\pi 1.5 \text{ GHz}$   
decay rate  $\kappa = 10 \text{ kHz}$  corresponding to lifetime  $T_1 = 100 \mu\text{s}$

# Measure Transmission through a resonator

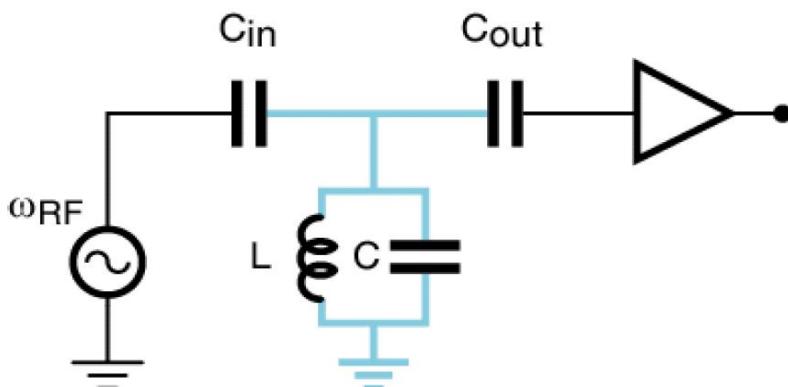
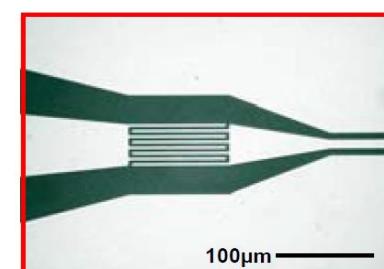
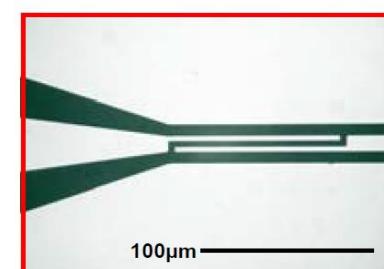
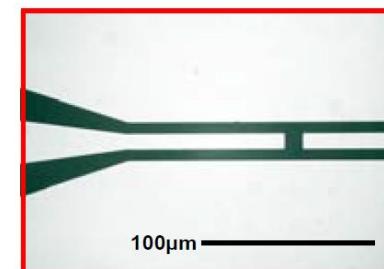
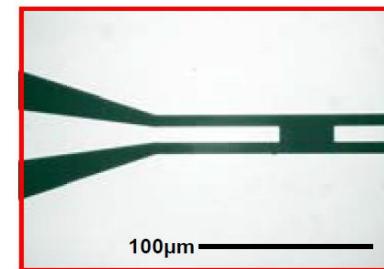
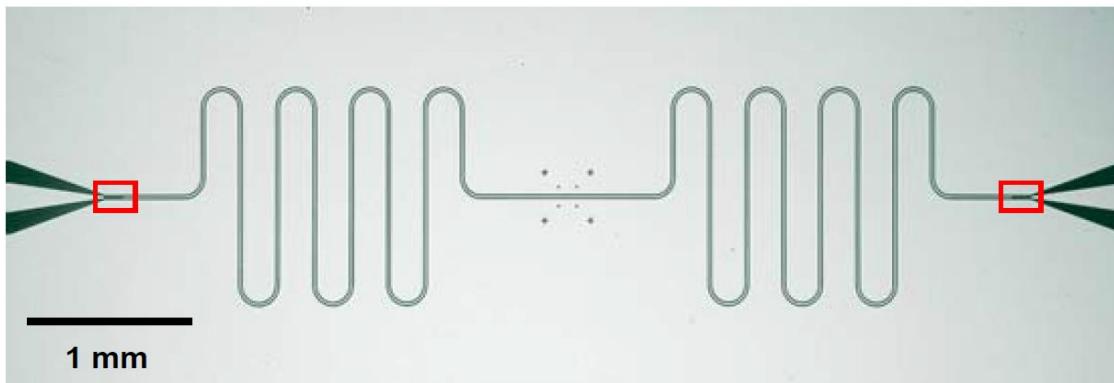
$$f_0 = \frac{c}{\sqrt{\epsilon_{\text{eff}}}} \frac{1}{2l}.$$



$$c / \sqrt{\epsilon_{\text{eff}}} = v_{\text{ph}}$$
$$v_{\text{ph}} = 1 / \sqrt{L_\ell C_\ell}$$



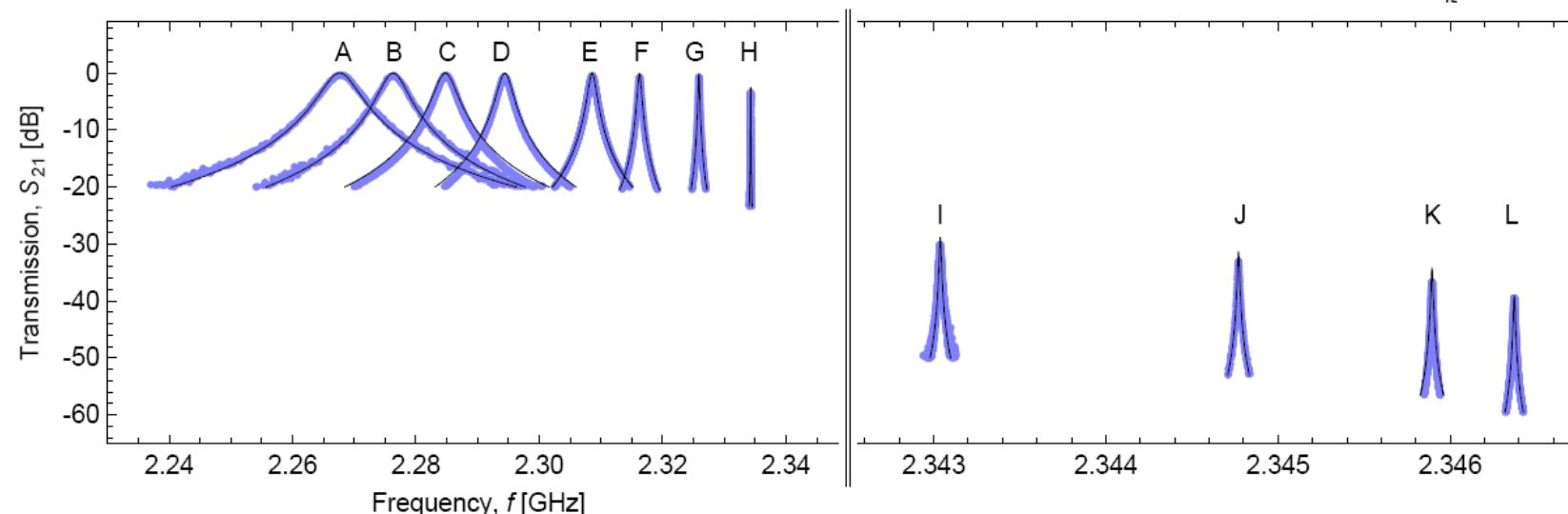
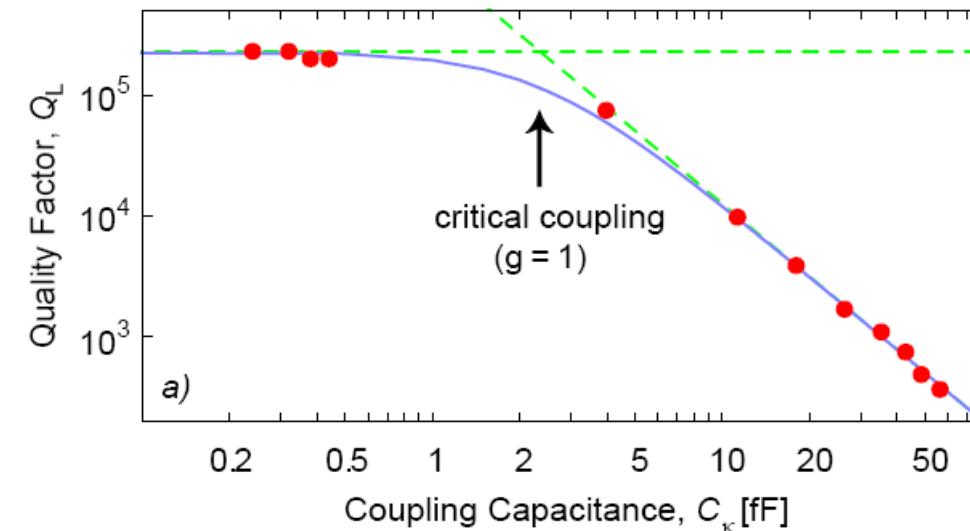
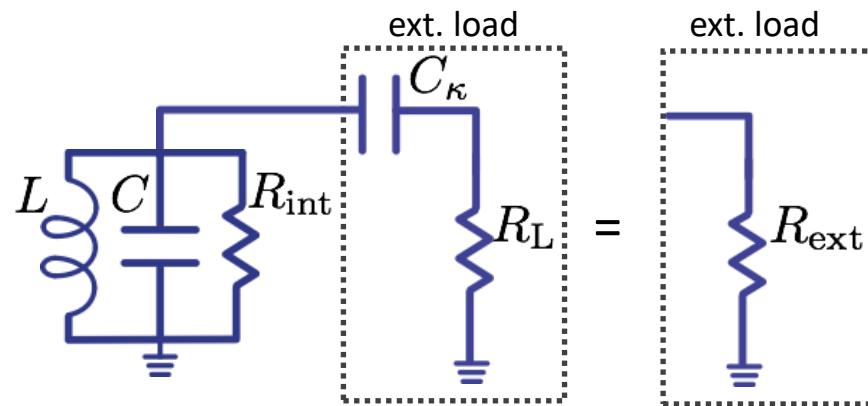
# Controlling the Photon Lifetime $\propto 1/Q_{ext}$



photon lifetime (quality factor)  
controlled by coupling capacitor  $C_{in/out}$

# Internal and External Dissipation in an LC Oscillator

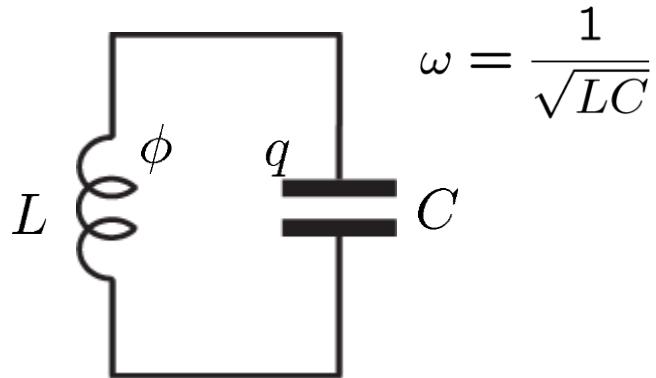
M. Goepp et al., *J. Appl. Phys.* **104**, 113904 (2008)



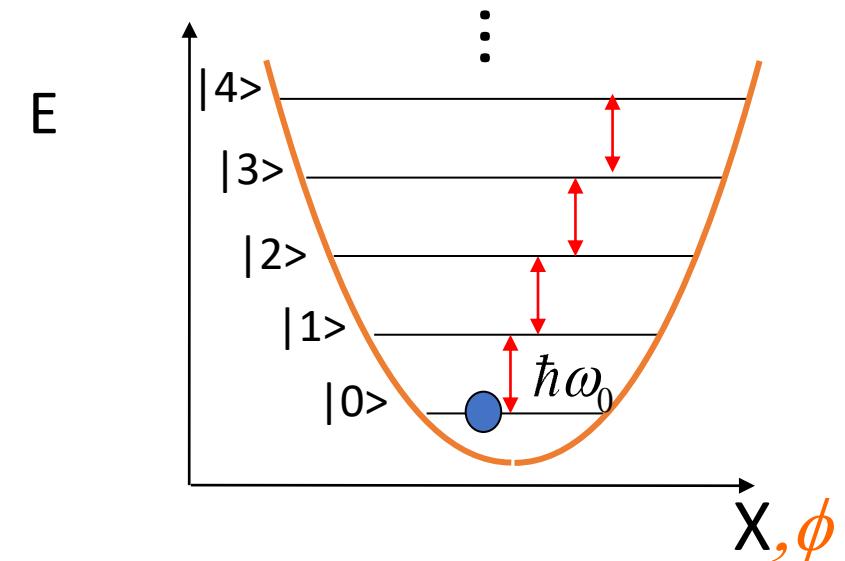
for  $Q = 10^6$  and  $\omega_0 \sim 2\pi 1.5$  GHz  
decay rate  $\Gamma_1 = 10$  kHz corresponding to lifetime  $T_1 = 100$   $\mu$ s

# Harmonic oscillator as a qubit?

harmonic LC oscillator:



Even if we prepare the oscillator in its ground state we do not have an easy way neither to control its quantum states nor to measure.

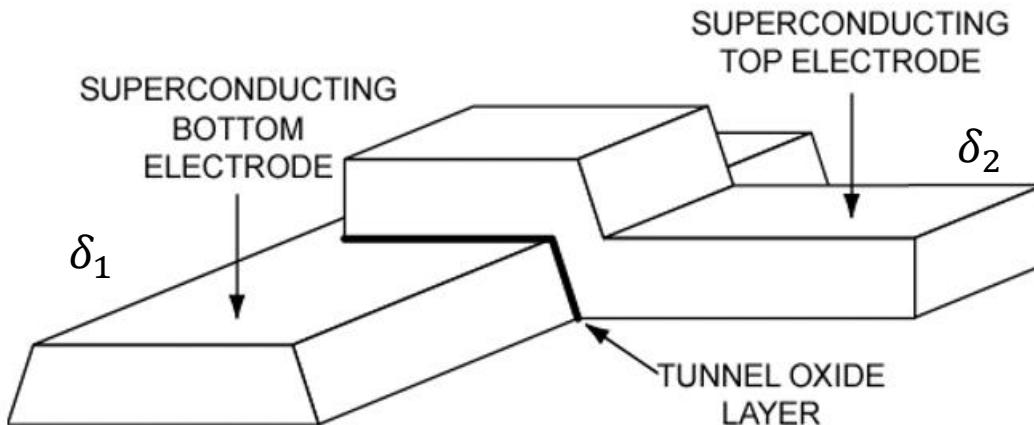


The system is harmonic -> cannot define a qubit

Need to some nonlinear and non-dissipative quantum circuits for that!

# Next lecture

# The Josephson Junction



$\phi_0 = \frac{h}{2e}$  is the flux quantum

$$\phi = \delta_2 - \delta_1$$



## Nobel Prize in Physics 1973

"for his theoretical predictions of the properties of a supercurrent through a tunnel barrier, in particular those phenomena which are generally known as the Josephson effects".

Non-linear current-phase relation  
The potential energy is not harmonic  
No dissipation

Josephson relations:

$$I = I_c \sin \phi$$

$$V = \frac{\Phi_0}{2\pi} \dot{\phi}$$

Josephson inductance

Specific Josephson inductance

$$V = \frac{\Phi_0}{2\pi I_0} \frac{1}{\cos \phi} \dot{I} = L_J \dot{I}$$

Josephson energy

Specific Josephson energy

$$E_J = \int VI dt = \frac{I_0 \Phi_0}{2\pi} (1 - \cos \phi)$$

$I_0 = 100 \text{ nA}$  corresponds to  $L_{J0} \sim 3 \text{ nH}$

$I_0 = 100 \text{ nA}$  corresponds to  $E_{J0}/h \sim 50 \text{ GHz}$

Inductive energy  $U(\phi) = \frac{\phi^2}{2L}$  for an harmonic oscillator