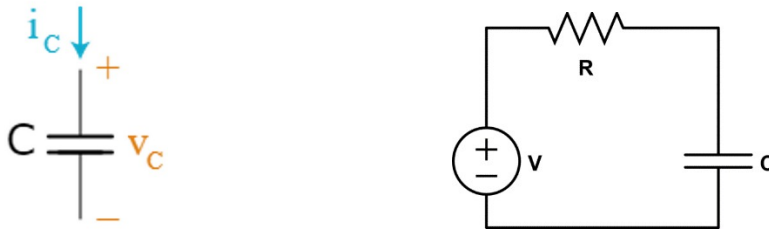


Homework 7

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Problem 7.1 – Capacitors in circuits (1h)

In class, we mainly learned about capacitors in terms of the charge-voltage relationship, $Q = CV$. This becomes a new circuit component for us. In the circuits context, as usual, we focus on its current-voltage relationship, $i = C dV/dt$.



While we're not yet prepared to tackle every circuit with any number of capacitors, with basic calculus we can do some classes of problems. Consider the so-called RC circuit in the Figure above, where we would like to solve for the voltage across the capacitor.

- Write the KVL, KCL, and component law equations that define how this circuit works (as usual)
- Combine the equations into a single equation in which v_c is the only unknown (as usual)
- Your equation will be a differential equation, which is different from the algebraic equations we've seen so far. Let $V = 0$ and solve the differential equation assuming that the capacitor's voltage starts at V_{ic} . Write a phrase/sentence stating how the capacitor's **initial condition** enters into the calculation.
- Your solution will rely on the product $R \times C$. What are the units of RC ? What is the significance of this number, known as the **time constant** of the system?
- Create a table with the percent of the overall transition that the capacitor voltage has completed after N time constants, where $N = 1, 2, 3, 4, 5$.
- Use Thevenin's Theorem to demonstrate that a system with a single capacitor (plus any number of resistors and sources) will result in the same solution as above and therefore will be characterized by a single time constant. Do the sources affect the time constant?
- Use LTSpice to simulate the RC circuit with $V = 0V$, $R = 10\Omega$, $C = 3\mu F$. Label the node below the capacitor as ground and the node above the capacitor as "vc." Place a "spice directive" anywhere on the schematic that says ".ic V(vc)=5" to set the initial condition on the capacitor. Use a transient simulation and make sure to simulate for a long enough time to see the whole transition but not so long of a time that you can't see the transition. Include a screenshot of the schematic and the capacitor voltage as your solution to this part.

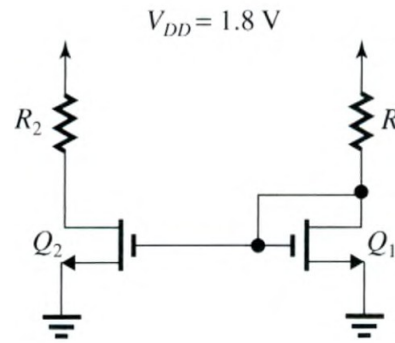
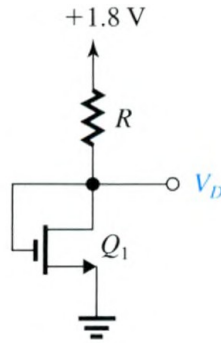
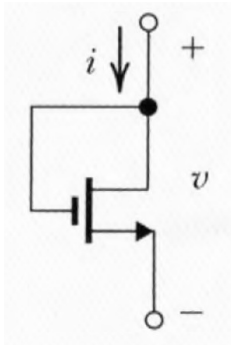
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Problem 7.2 – Diode-Connected Device (1h)

Sometimes NMOS transistors have their gates shorted to their drains as shown in the Figure. This is known as a diode connected device.

- Will the transistor operate in the cutoff, linear, or saturation region?
- Assume the MOSFET parameters and geometry (μ_n , C_{ox} , V_t , W , L) are known and ignore channel length modulation ($\lambda=0$). Calculate the I-V characteristic of this two-terminal device when v is positive and when v is negative. In what way is this characteristic similar to the exponential behavior of a diode and in what way is it not similar?
- The NMOS transistor in the second Figure have $V_t=0.5\text{ V}$, $\mu_n C_{ox}=400\text{ }\mu\text{A/V}^2$, $\lambda=0$, $W=0.72\text{ }\mu\text{m}$, and $L=0.18\text{ }\mu\text{m}$. Find the value of R that results in $V_D=0.7\text{ V}$.
- The third Figure is obtained by augmenting the second Figure with a transistor Q_2 identical to Q_1 but with $W=2.16\text{ }\mu\text{m}$. As long as Q_2 is in the saturation region, what will I_{D2} be? Does it matter what value R_2 is?
- What value of R_2 results in Q_2 being at the edge of saturation? Should R_2 be bigger or smaller than this value to keep Q_2 in saturation?



Problem 7.3 – MIS capacitors (0.5h)

An MIS capacitor is composed of an SiO_2 (relative permittivity = 3.9) dielectric of thickness 40 nm on top of p-type silicon with doping level $N_A=10^{17}/\text{cm}^3$. Assume the threshold voltage is 0.5 V and a voltage of 1.8 V is applied to the capacitor.

- Calculate the induced surface charge density (charge per unit area) in the metal electrode. Are these charges positive or negative, mobile or stationary?
- What must the induced charge density be at the semiconductor surface?
- How much of the induced charge density at the semiconductor surface is due to stationary charges vs mobile charges?

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Problem 7.4 – An MOS FET (0.5h)

A MOSFET is made with a dielectric composed of SiO_2 (relative permittivity 3.9) and thickness 40 nm. The channel length is 180 nm and the channel width is 3 μm . The applied gate voltage is 3V, the source voltage is 0 V, and the drain voltage is 0.5 V. The threshold voltage is 0.5 V. Assume the mobility of electrons to be $1000 \text{ cm}^2/\text{Vs}$.

- Calculate the drain current
- Calculate the equivalent resistance of the MOSFET assuming that its drain current rises linearly with its voltage up to the operating point in the problem.
- All else equal, if the drain voltage is gradually increased, at what value of drain voltage will the drain current saturate to a constant value? What will that constant value be?

Problem 7.5 – Small Signal Modeling (1h)

Consider the so-called “common-gate” amplifier below. For all parts, assume the following:

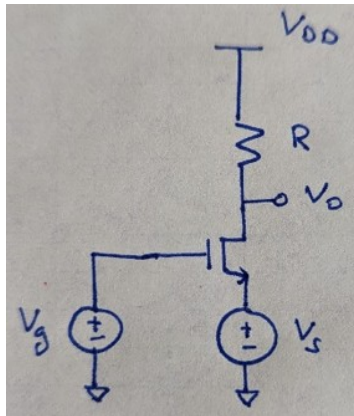
$$\mu_n C_{ox} = 200 \mu\text{A}/\text{V}^2$$

$$\mu_p C_{ox} = 100 \mu\text{A}/\text{V}^2$$

$$V_T = 0.4 \text{ V for NMOS devices}$$

$$V_T = -0.4 \text{ V for PMOS devices}$$

$$W/L = 10$$



- An amplifier is made with an NMOS transistor as shown in the figure. $R = 2 \text{ k}$ and $V_{DD} = 2 \text{ V}$. The DC value of the gate voltage (V_G) is 0 V. Calculate the value of the input, V_S , that will result in the FET being at the edge of saturation.
- Calculate the small signal transconductance g_m for the MOSFET in (a).
- $v_s(t)$ has a dc value V_S calculated in (a) as well as small perturbation such that $v_s(t) = V_S + \tilde{v}_s$. Draw the small signal model of the amplifier and calculate the small signal voltage gain \tilde{v}_o/\tilde{v}_s . Let your answer be symbolic, i.e., a function of g_m and R rather than a number.
- At the output port, calculate the Thevenin equivalent of the small-signal model of the circuit as a function of g_m and R .

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- e) At the input port (i.e., across v_s), calculate the Thevenin input impedance of the small signal model of the circuit as a function of g_m and R , assuming that the output is open circuited.