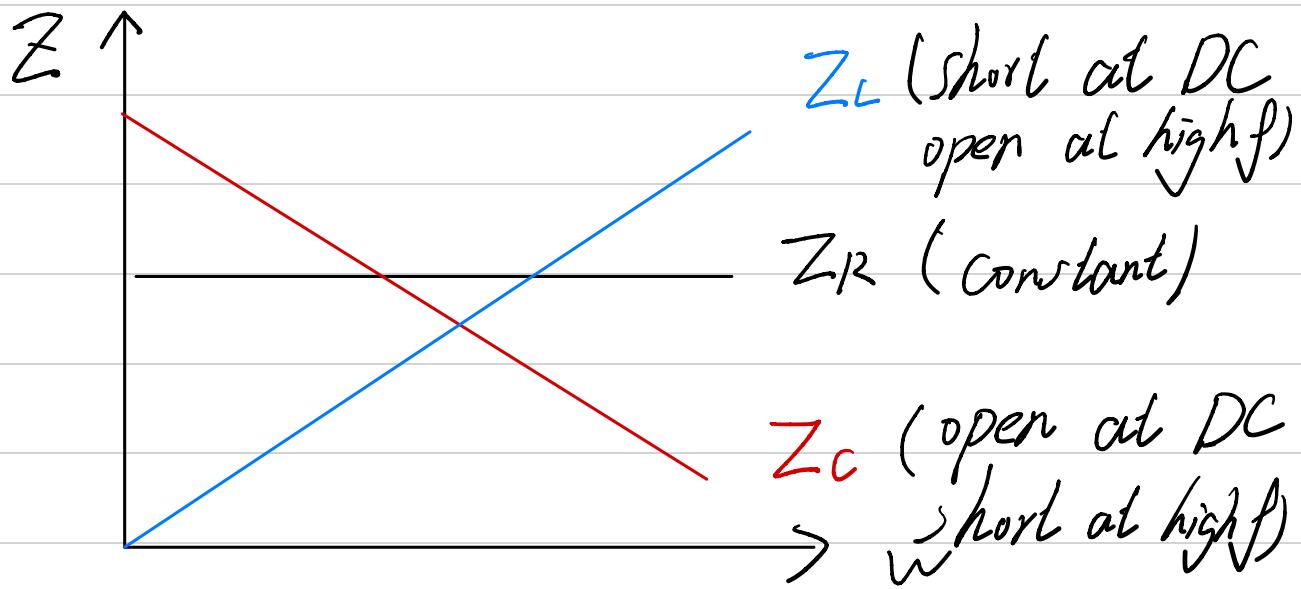


$$Z_R = R$$

$$Z_C = \frac{1}{j\omega C}$$

$$Z_L = j\omega L$$



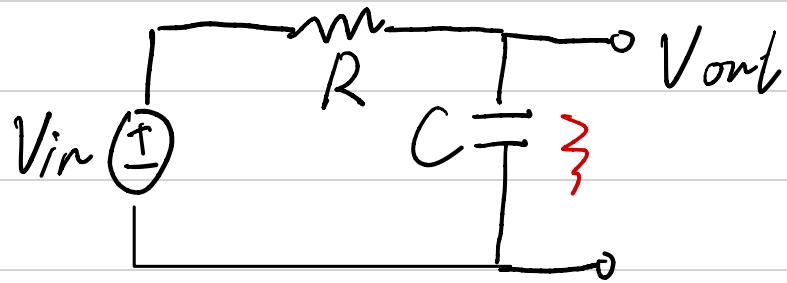
Why frequency domain ?

1. physical insight

2. Math simple

TF: $H(j\omega) = \frac{V_{out}(j\omega)}{V_{in}(j\omega)}$

1. RC low-pass filter



$$V_{out} = V_{in} \times \frac{Z_C}{Z_R + Z_C}$$

\downarrow

$$R \quad \frac{1}{j\omega C}$$

$$H(j\omega) = \frac{V_{out}}{V_{in}} = \frac{\frac{1}{j\omega C}}{R + \frac{1}{j\omega C}} = \frac{1}{1 + j\omega RC}$$

$$|H(j\omega)| = \frac{1}{\sqrt{1 + (\omega RC)^2}}$$

$$\angle H(j\omega) = \tan^{-1}\left(\frac{0}{1}\right) - \tan^{-1}\left(\frac{\omega RC}{1}\right)$$

$$= -\tan^{-1}(\omega RC)$$

①

N Phase Behavior

$\omega \ll \frac{1}{RC}$

≈ 1 (0dB)

0°

passband

$\omega = \frac{1}{RC}$

$\frac{1}{\sqrt{2}}$ (-3dB)

-45°

cutoff

$\omega \gg \frac{1}{RC}$

$\frac{1}{\omega RC} (-20\text{dB/dec})$

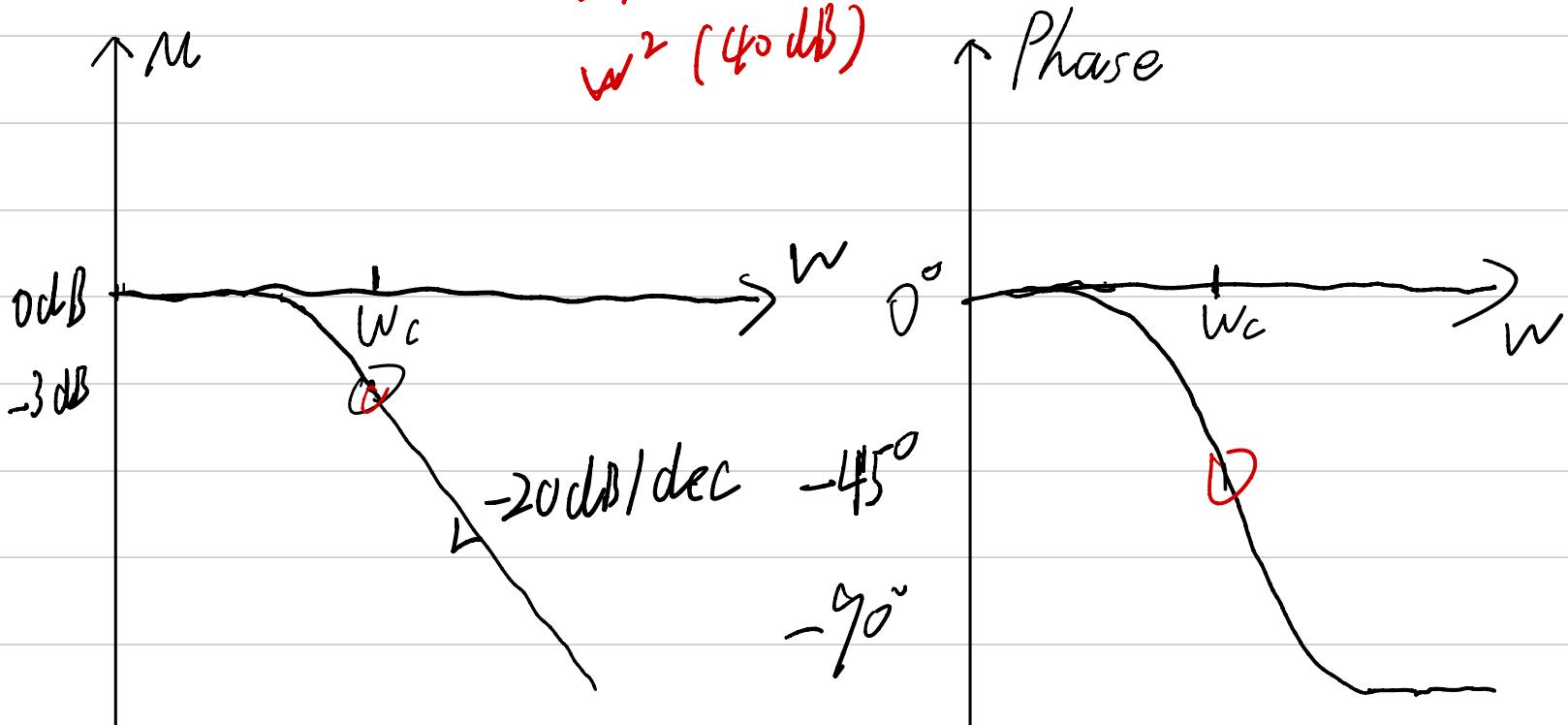
-90°

stopband

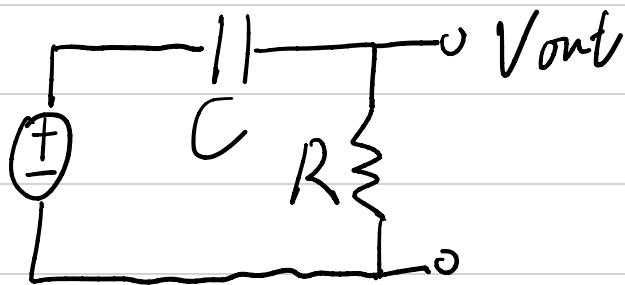
$$20 \log_{10} \left(\frac{1}{\omega} \right) = -20$$

$$\omega RC (20\text{dB})$$

$$\omega^2 (40\text{dB})$$



RC high-pass filter



$$H(j\omega) = - \times \frac{Z_R}{Z_R + Z_C}$$

$$= \frac{R}{R + \frac{1}{j\omega C}}$$

$$\frac{j\omega C}{1} = j\omega C$$

$$= \frac{j\omega RC}{1 + j\omega RC}$$

$$|H(j\omega)| = \boxed{\frac{j\omega RC}{\sqrt{1 + (\omega RC)^2}}}$$

$$\angle H(j\omega) = 90^\circ - \tan^{-1}(\frac{\omega RC}{1})$$

$$\tan^{-1}(\frac{\omega RC}{1})$$

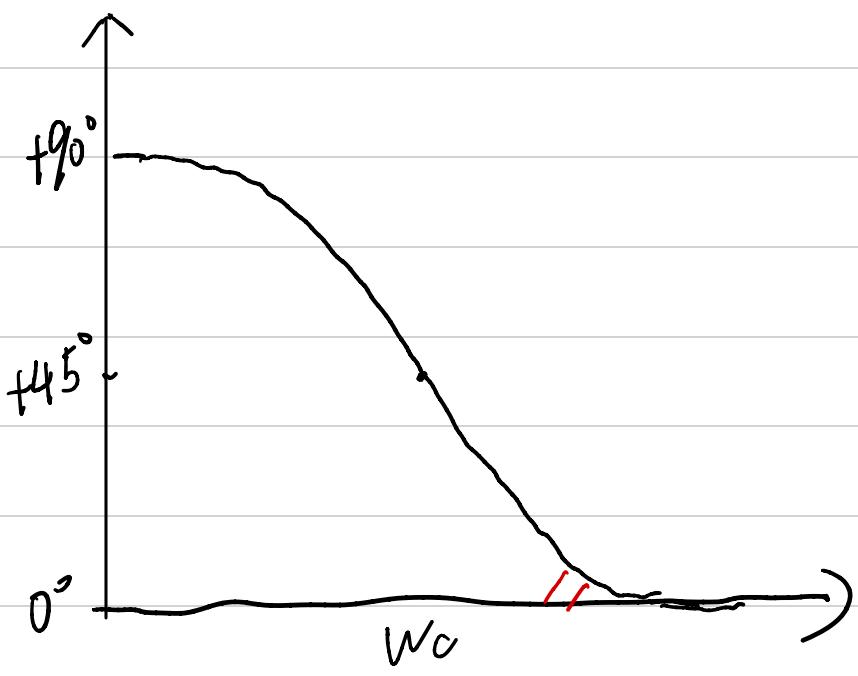
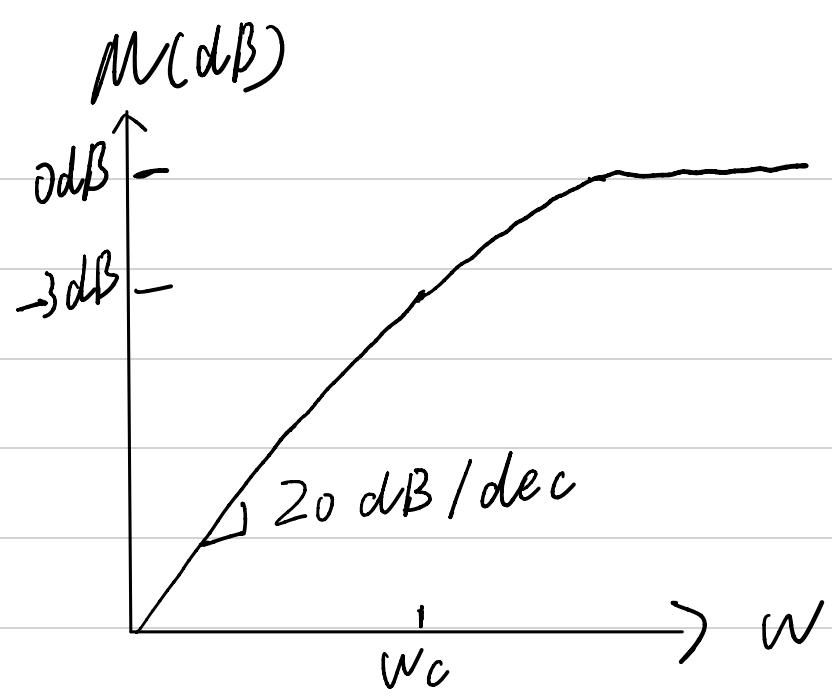
\downarrow M Phase Behavior

$\omega < \omega_c$ $\omega RC (+20dB/\text{dec})$ $+90^\circ$ stopband

$\omega = \omega_c$ $\frac{1}{\sqrt{2}} (-3 \text{ dB})$ $+45^\circ$

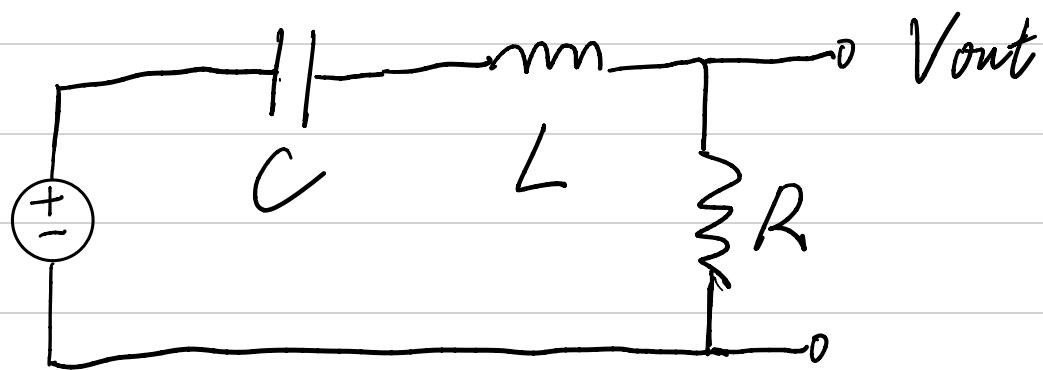
cutoff
passband

$\omega > \omega_c$ $1 (0 \text{ dB})$ 0°



RL (high / low pass)

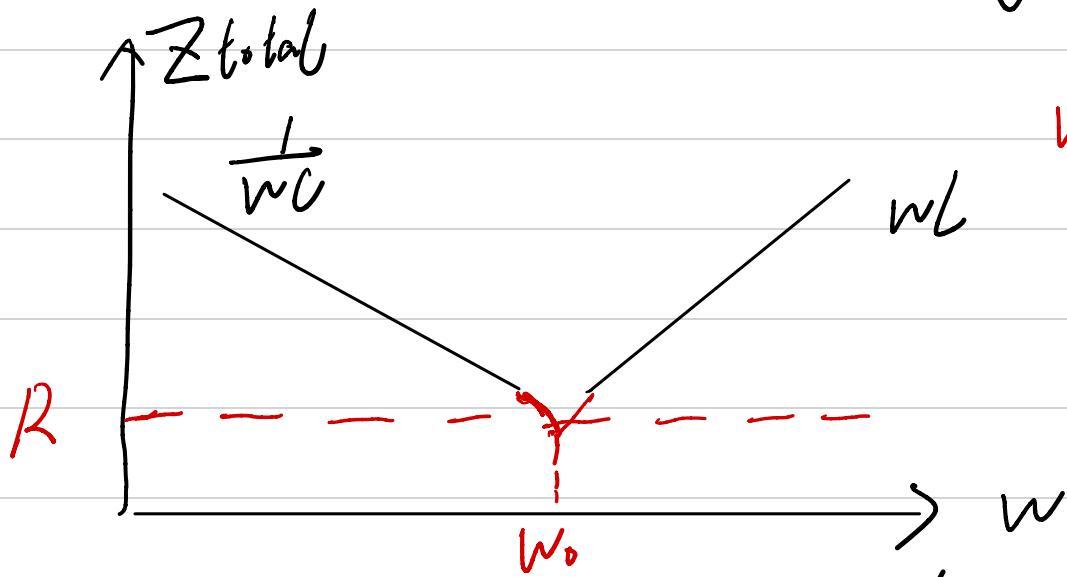
RLC circuit (series)



$$Z = Z_C + Z_L + Z_R = R + j\omega L + \frac{1}{j\omega C}$$

$$= R + j(\omega L - \frac{1}{\omega C})$$

$\omega L = \frac{1}{\omega C}$



$$I = \frac{V_{in}}{Z_{total}}$$

ω_0 resonant frequency $\nu = \frac{1}{\sqrt{LC}}$

$$H(j\omega) = \frac{Z_R}{Z_{total}} = \frac{R}{R + j\omega L + \frac{1}{j\omega C}} = \frac{R}{R + j(\omega L - \frac{1}{\omega C})}$$

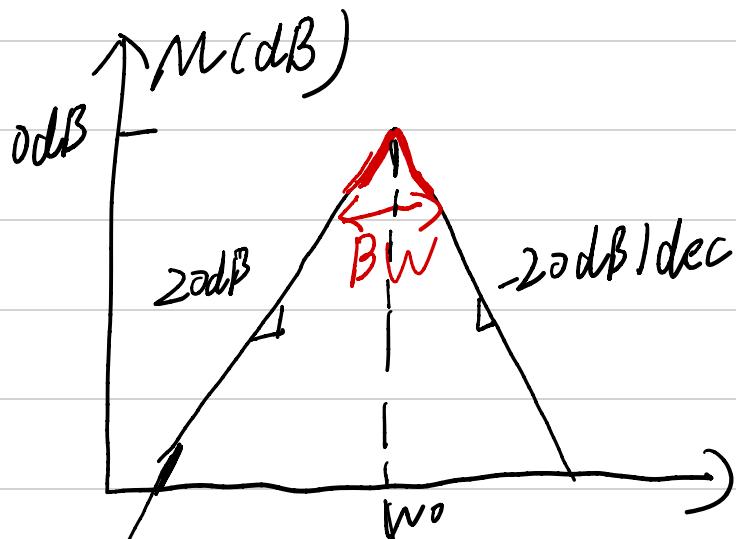
$$|H(j\omega)| = \frac{R}{\sqrt{R^2 + (\omega L - \frac{1}{\omega C})^2}}$$

$$\angle H(j\omega) = \tan^{-1}(\frac{\omega}{1}) \approx$$

$$- \tan^{-1}\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)$$

$$= - \tan^{-1}\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)$$

\int	M	Phase	Behavior
$\omega < \omega_0$	$\frac{R}{\omega C} = \omega CR$ $(+20dB/dec)$	$+90^\circ$	Capacitive
$\omega = \omega_0$	$1 (0dB)$	0°	resistive
$\omega > \omega_0$	$\frac{R}{\omega L} (-20dB/dec)$	-90°	Inductive

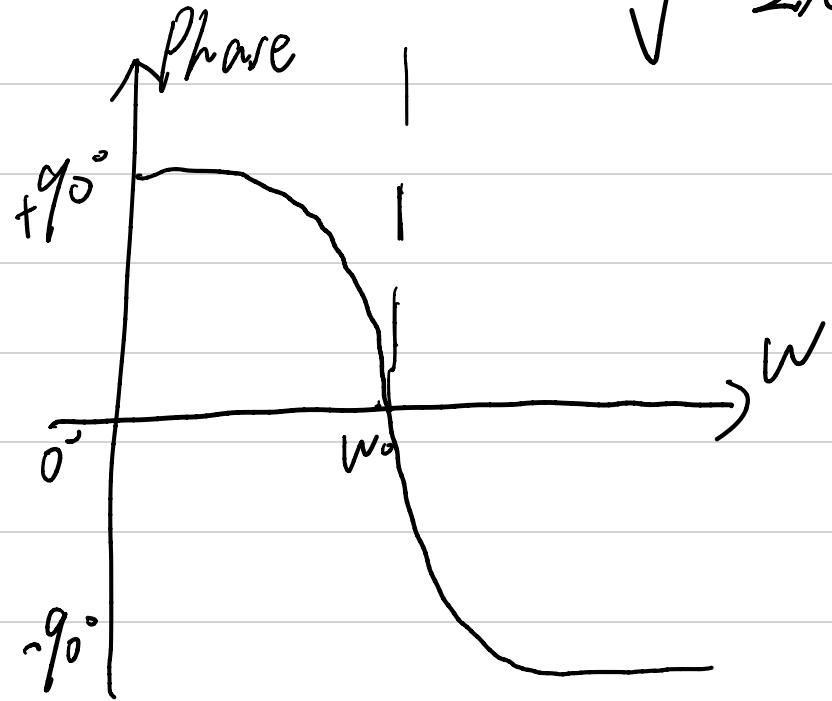


band-pass filter

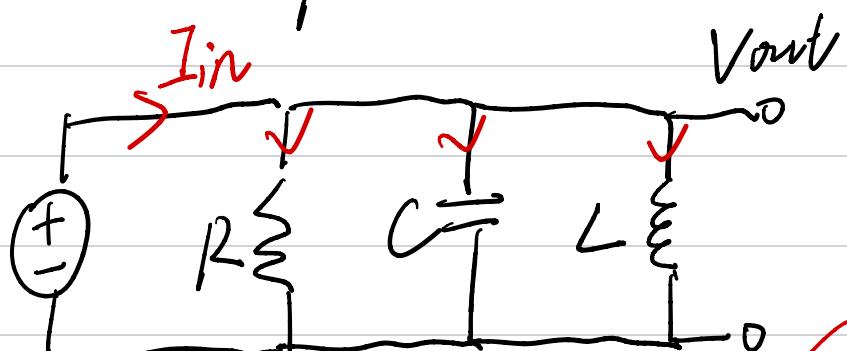
$$\omega_0 L - \frac{1}{\omega_0 C} = 0$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$f = \frac{1}{2\pi\sqrt{LC}}$$



RLC parallel



$$H(j\omega) = \frac{V_{out}}{V_{in}} = 1$$

$$H(j\omega) = \frac{I_{in}}{V_{in}} = \frac{1}{Z_{total}}$$

$$\frac{I_{in}}{V_{in}} = \frac{1}{Z_{total}} \Rightarrow \frac{V_{in}}{I_{in}} = Z_{total}$$

$$\frac{1}{Z_{\text{total}}} = \frac{1}{Z_R} + \frac{1}{Z_C} + \frac{1}{Z_L}$$

↙

$$\text{Admittance} = \frac{1}{R} + \frac{1}{jwC} + \frac{1}{jwL}$$

$$= \frac{1}{R} + j(wC - \frac{1}{wL})$$

$$|H(jw)| = \sqrt{\frac{1}{R^2} + (wC - \frac{1}{wL})^2}$$

$$\angle H(jw) = \tan^{-1} \left(\frac{wC - \frac{1}{wL}}{\frac{1}{R}} \right) = \tan^{-1} (R(wC - \frac{1}{wL}))$$

\int	M	Phase	Behavior
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$w \ll w_0$	$\frac{1}{wL} (-20 \text{ dB/dec})$	-90°	Inductive
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$w = w_0$	$\frac{1}{R}$	0°	resistive
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$w \gg w_0$	$wC (+20 \text{ dB/dec})$	$+90^\circ$	Capacitive
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