LOW POWER N-BIT ADDERS AND MULTIPLIER USING LOWEST-NUMBER-OF-TRANSISTOR 1-BIT ADDERS

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Abstract

4-bit Ripple Carry Adders (RCA), 12-bit Carry Select Adders (CSA), and a 4×4 Braun Multiplier, based on lowest-number-of-Transistor full adders, were designed and simulated. The designed full adders consist of 10 Transistors and were used for n-bit adders with output voltage levels having a maximum of one threshold voltage (V_T) degradation. The 10 Transistors adder achieved a 43.68% reduction in the power dissipation compared to the standard CMOS-28T Adder. Power consumption can be further reduced by using an extra stack transistor. A 12-Transistor Adder was also designed for low area Array Multipliers.

Keywords: Full Adders; Voltage Threshold Loss; Pass Transistors; n-bit Adder; Array Multiplier.

I. Introduction

Design of digital Integrated Circuits for many applications relies on three major criteria: Low power consumption, reduced chip area, and high speed. Using lower number of transistors to implement a logic function is beneficial in reducing the device and interconnect parasitic and reducing the chip area, resulting in lower time delay and potentially lower power consumption. However; the problem of threshold-voltage (V_T) loss of the output voltage levels, of those low-number-transistor circuits [2], may lead to faulty operation and higher leakage currents, especially for $0.18\mu m$ and subsequent CMOS technologies.

Reducing the number of transistors of 1-bit adder design, for lower power consumption and higher speed in $0.35\mu m$ and older CMOS technologies was investigated [1-4]. However, many low-number-transistor adders do not operate correctly at low supply voltage in $0.18\mu m$ and subsequent CMOS technologies due to V_T loss problem. The goal of this work is to design the lowest–number-of-transistors that can be used successfully in an n-bit adder or n-bit array multiplier. The targeted technology is $0.18\mu m$ CMOS Technology.

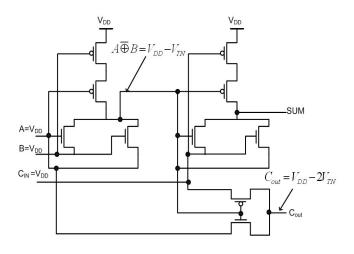
This paper presents two 10-Transistor (10-T) and one 12-T Adder designs for n-bit adders and multiplier respectively. The two 10-T designs were not analyzed or studied before even though they were two of the forty-one 10-T adders mentioned in [2]. Adding one series-connected (stack) transistor to the

circuit does reduce the power dissipation. The design of 12-T adders is reported here for the first time targeting their use in n-bit Multipliers. The two 10-T adders suffer from one V_T loss instead of two V_T as in many other designs [1], [3], [4]. The 12-T adder design has only one output (Sum) suffering from one V_T loss while C_{OUT} has full voltage swing. A 4-bit RCA, a 12-bit CSA adder, and a 4-bit Braun multiplier based on 10-T or 12-T designs, are presented.

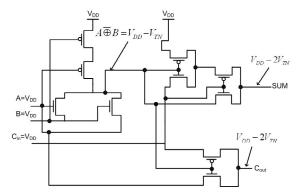
II- Review of previous 10-T 1-bit Adders

Previous designs of 10-T adder are used in this paper for comparison purposes (Figure 1). The SERF adder [3], the Yuke Wang's three 10-T adders (named: 9A, 9B, and 13A) [1], and Junming's 10-T 1-bit adder [4]. They all suffer from $2V_T$ loss of the output voltage levels for certain input signals sets. This was not critical for $0.35\mu m$ and older CMOS technologies. However, for the $0.18~\mu m$ and subsequent CMOS technologies, the proper operation of these adder designs is not possible including their use in n-bit adders.

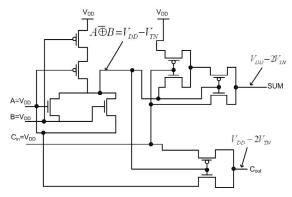
The voltage level degradation (V_T loss) occurs when either the PMOS transistor is passing logic "0" (V_{TP} loss) or the NMOS transistor is passing logic "1" (V_{TN} loss). Fig. 1 shows some of existing 10T adders with the input signals set that causes $2V_T$ loss of its output signal.



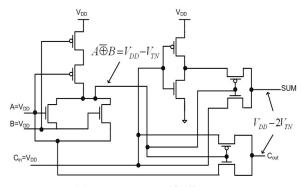
(a) SERF Adder



(b) Yuke-Wang-9A adder



(c) Yuke-Wang-9B adder



(d) Yuke-Wang-13A" adder

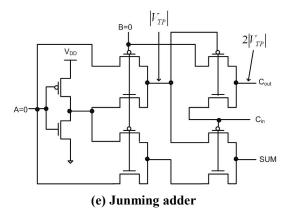


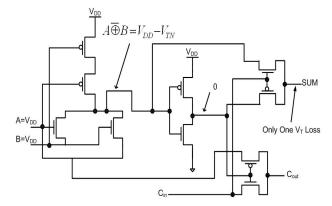
Fig. 1: Five previous 10-Transistors (10-T) full-adders showing the wrest output voltages.

III- Two 10-T Full-bit adder with minimum degradation of its output voltage level

Two designs of 10-T full adder, with only one V_T loss of the output voltage levels, are described and simulated in this paper. The critical part of the design is having a full swing XOR or XNOR gate in the first module to design the rest of the 1-bit adder with multiplexers to generate the C_{OUT} and SUM outputs with improved voltage levels (Figure 2).

To generate a full swing XOR gate signal in n-10T adder (fig. 2a), an inverter is used. The XOR-gate output and $C_{\rm IN}$ input signals, both with full voltage swing, will be used as control signals in the following multiplexer stages to generate $C_{\rm OUT}$ and SUM outputs with a maximum of one $V_{\rm T}$ loss. Similarly, an inverter is used in the second design (fig. 2b).

The Power dissipation of the two 10-T adders may further be reduced if the leakage current of the inverter is suppressed or decreased. The output signal of XOR gate, for the P-10T adder, is $|V_{TP}|$ for a specific input signals set (A=B=0). This causes the NMOS transistor of the inverter not to be completely turned off, giving rise to an unwanted leakage current and higher power dissipation. This also applies to the N-10T transistor when the input signals A and B are high. This problem is illustrated in Figure 3.



(a) The N-10T 1-bit Adder

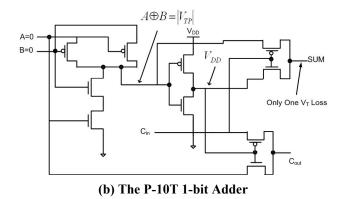


Fig. 2: Two 10-T Adders with one V_T loss.

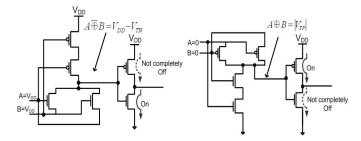


Fig. 3: The inverter leakage current for specific input signals

Sizing the transistors to have a skewed inverter, such that the switching voltage (V_S) of the inverter in the P-10T adder is higher than 0.5 V_{DD}, while it is lower in the N-10T adder. Furthermore, adding a narrow series-connected transistor to the inverter decreases the leakage current. (This technique is targeting the 0.13µm and subsequent CMOS technologies to reduce static power dissipation). This transistor is an NMOS for P-10T adder and a PMOS for N-10T adder. Figure 4 shows adding the series-connected transistor to the inverter. The inverter, used in the P-10T adder, may have an input of $|V_{TP}|$ in This makes the n-MOS transistor not the worst case. completely off. With sizing ($W_N = 300 \text{nm}$, $W_P = 900 \text{nm}$) of the inverter transistors, the switching voltage is changed to V_S=0.86V. Adding an n-MOS stack transistor, the switching voltage further increases to Vs=0.97V. For the N-10T adder, the input of its inverter degrades to V_{DD} - V_{TN} in the worse case. This makes the p-MOS transistor not completely off. With sizing $(W_N=600nm, W_P = 300nm)$, the switching voltage decreases to Vs= 0.67V. Adding a p-MOS stack transistor, Vs further decreases to 0.63V.

The power dissipation and the time delay of the two sized 10-T adders, the two adders with an extra transistor, CMOS (28T), and Transmission Gate Adder (TGA 20T) are simulated. The maximum frequency of input signals is 100MHz. The simulation results, in Table 1, show that both the power consumption and the Time delay of both P-10T and P-11T are higher than those of N-10T and N-11T. Therefore, we use N-10T adder in the n-bit adders and the modified N-12T adder in the array multiplier as described in the next sections.

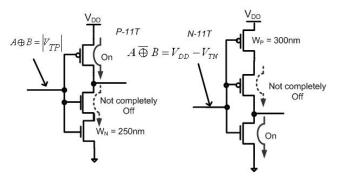


Fig. 4: Addition of a narrow Transistor to decrease the leakage current in an inverter (Stack Transistor).

Table 1: Simulation results of the adder designs

Frequency	Power	t _d (SUM)	t _d (C _{OUT})
100 MHz	$[\mu W]$	[ps]	[ps]
P-10T	20.9	792	629
N-10T	9.94	566	598
P-11T	15.56	814	544
N-11T	9.17	574	606
CMOS 28T	17.65	307	222
TGA 20T	12.19	215	263

IV- n-bit Adders

One of the advantages of N-10T adder is that it can operate properly even when one of the inputs of XNOR gate has one V_T loss. By applying the signal with one V_T loss to input B, fig. 2, the V_T loss does not propagate to the output of the adder, and consequently does not accumulate when used in n-bit Adders. The wave shown in figure 5 illustrates the full swing wave, called "A" $[0,\ V_{DD}]$, and the wave with one V_T loss, called "B" $[V_{TP},\ V_{DD}$ - $V_{TN}]$, can be added together in the N-10T adder. The output of XOR, after the inverter, is not affected by the V_T loss of one of its two inputs.

The input B, in the simulation, varied from 0.6 V to 1.2V representing the threshold-voltage loss of the signal level. However, this is quite larger than the value of the threshold-voltage with no body effect ($V_{TN0} = 0.487V$) in 0.18 μ m CMOS technology. These values reflect the voltage levels obtained from the circuit simulation. This is due to the high body effect of the transistors. The V_T loss can be reduced by using lower V_T and/or lower-body-effect transistors, if accommodated by the adopted CMOS technology.

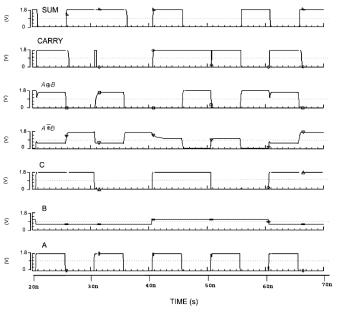


Fig. 5: Full swing of the output voltage of the XOR gate used in the N-10T adder and the correct operation of the 1-bit adder.

Therefore, the described N-10T design can be used in n-bit RCA and CSA Adders [5] if the carry-out ($C_{\rm OUT}$) of previous adder is applied at input B. However, this is not desirable for RCA adders since the path of the propagation of the carry signal is made longer. All the inputs signals are applied through buffers and have been fed into the adder cells, while all the outputs are loaded with buffer circuits. The power consumption and maximum time-delay values of the 4-bit and 12-bit adders are shown in Table 3. The simulation was carried out at a frequency of 50MHz for the 4-bit RCA and the 12-Bit CSA.

V- The 12-T 1-bit Adder and Array Multiplier

Previous 12-T adders [2] do not operate correctly when used in a multiplier due to accumulation of V_T loss, causing erroneous or ambiguous output logic value. Here we present a solution by modifying the design of the 10-T adders. One of the adder two outputs must have full voltage swing while the other can have a maximum of one V_T loss. Adding two transistors to the multiplexer, used to generate C_{OUT} , in the 10-T adders, allows a full voltage swing of C_{OUT} . This increases the number of transistors of the improved adder to 12 (fig. 6).

The 4x4 multiplier employs AND gates and an array of 12-T 1-bit adders to generate the final result [5-6]. The simulation is performed at a frequency of 50 MHz, for 0.18µm CMOS technology. The results are presented in Table 2 and some of the Output waves with one input wave are shown in figure 7.

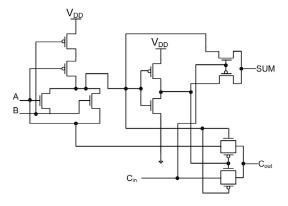


Fig. 6: N-12T Adder: Adding 2 transistors, to the N-10T design, to remove the V_T -loss of the output signal $C_{\rm OUT}$.

Table 2: Simulation results of a 4-bit RCA adder, a 12-bit CSA adder, and an array multiplier

n-bit adder, multiplier	The used 1-bit adder	Power [μW]	Time Delay [ns]
4-bit RCA	N-10T	34.28	3.103
12-bit CSA	N-10T	90.14	5.28
4×4 Array Multiplier	N-12T	145.48	3.97

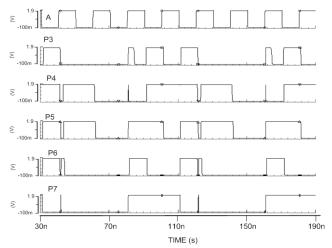


Fig. 7: One set of selected input (A) and output (P3-P7) pulses from the simulation of the 4x4 Array Multiplier.

VI- Conclusion

Minimum-Number-Transistor adders and multiplier which have up to one V_T loss at their output signals are described in this paper. Two n-bit adders and one array-multiplier, with acceptable operation, were designed and simulated using the lowest-number-of-transistors (10-T and 12-T) full adders. Challenges in lowering the power dissipation and further reducing the V_T losses can be solved in 0.13 μ m and subsequent CMOS technologies where multi V_T devices are provided. MOSFET Transistors with lower body effect will also reduce the V_T losses.

Acknowledgments

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