Lecture 9
Wavelet and Other Image Transforms

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Outline

❑ **2D Unitary transform**

❑ **Frequency Domain Extension**

➢ Discrete Cosine Transform（余弦变换）

➢ Hadamard Transform （哈德马变换）

➢ Discrete Wavelet Transform （小波变换）

❑ **Discrete Wavelet Transform (DWT)**

➢ An example for 1D-DWT

➢ Generalization of 1D-DWT

➢ 2D-DWT

Unitary Transform

❑ **Forward Transform:**

$$t = Af$$

$$t[k] = \sum_{n=0}^{N-1} A[k,n]f[n]$$

❑ **Inverse Transform:**

$$f = A^H t \quad \text{if } A^H = (A^T)^* \text{ and } AA^H = I$$

Example for 1D Unitary Transform
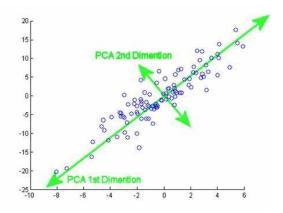
❑ **Image rotation:**

$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \begin{bmatrix} cos\theta & sin\theta \\ -sin\theta & cos\theta \end{bmatrix} * \begin{bmatrix} x \\ y \end{bmatrix}$$

❑ **Principle Component Analysis (PCA):**

$Y = PX$ that satisfies $C = XX^T$ $D = PCP^T$

and $PP^T = I$

Discrete Fourier Transform

➢ **Forward Transform:**
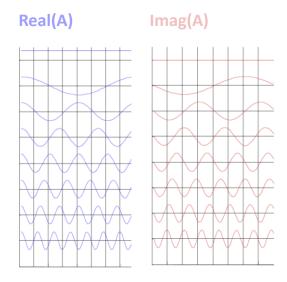
$$t = Af; \quad t[k] = \sum_{n=0}^{N-1} A[k,n]f[n]$$

➢ **Inverse Transform：**

$$f = A^H t; \quad f[n] = \sum_{k=0}^{N-1} A^H[k,n]t[k]$$

➢ **1-D DFT**

$$X[k] = \sum_{n=0}^{N-1} x[n]\, e^{-j\frac{2\pi kn}{N}}, (k = 1,2,\cdots,N)$$

$$A[k,n] = e^{-j\frac{2\pi kn}{N}} = cos(2\pi\frac{kn}{N})\text{-}jsin(2\pi\frac{kn}{N})$$

Real(A)   **Imag(A)**

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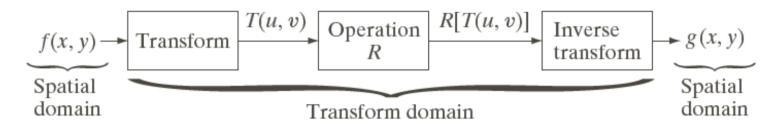ShanghaiTech University

2D Unitary Transform

❑ Forward Transform:

$$F(u,v) = \sum_{x=0}^{M-1}\sum_{y=0}^{N-1} f(x,y)e^{-j2\pi(\frac{ux}{M}+\frac{vy}{N})}$$
$$= A_M f A_N$$

❑ Inverse Transform

$$f = A_M^H F A_N^H \qquad\qquad AA^H = I$$

Image Transform

❑ The general approach for operating in linear transform domain

❑ The unitary transform satisfies

$$\sum_{x=0}^{M} \sum_{y=0}^{N} (f[x, y])^2 = \sum_{u=0}^{M} \sum_{v=0}^{N} (F[u, v])^2$$

➤ i.e., the energy is preserved.

Good and Bad things about DFT

❑ **Positive：**

➢ Energy is usually packed into low-frequency coefficients

➢ Convolution property

➢ Fast implementation

❑ **Negative：**

➢ Transform is complex, even if image is real

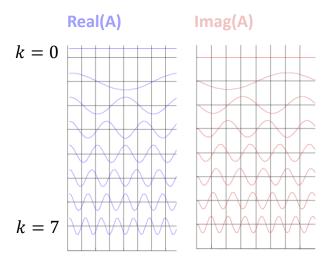➢ The basis function span image height/width

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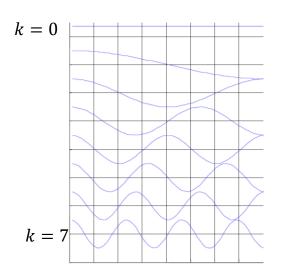
DFT vs. DCT (Discrete Cosine Transform)

➢ **1D-DFT**

$$A[k,n] = e^{-j\frac{2\pi kn}{N}}$$

$$= cos\left(2\pi\frac{kn}{N}\right) + jsin\left(2\pi\frac{kn}{N}\right)$$

➢ **1D-DCT**

$$A[k,n] = \sqrt{\frac{2}{N}}cos\frac{\pi(2n+1)k}{2N}$$

Real(A)   **Imag(A)**

$k = 0$

$k = 7$

$k = 0$

$k = 7$

What's the difference???

2D DCT

❑ Forward Transform

$$F(0,0) = \frac{1}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y)$$

$$F(u,0) = \frac{\sqrt{2}}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos\frac{(2x+1)u\pi}{2N}$$

$$F(0,v) = \frac{\sqrt{2}}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos\frac{(2y+1)v\pi}{2N}$$

$$F(u,v) = \frac{2}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos\frac{(2x+1)u\pi}{2N} \cos\frac{(2y+1)v\pi}{2N}$$

2D IDCT

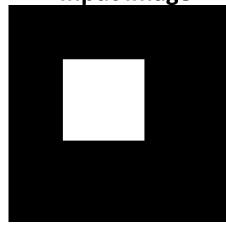## ❑ Inverse Transform

$$f(x, y) = \frac{1}{N} F(0,0)$$

$$+ \frac{\sqrt{2}}{N} \sum_{u=1}^{N-1} F(u, 0) \cos \frac{(2x+1)u\pi}{2N}$$
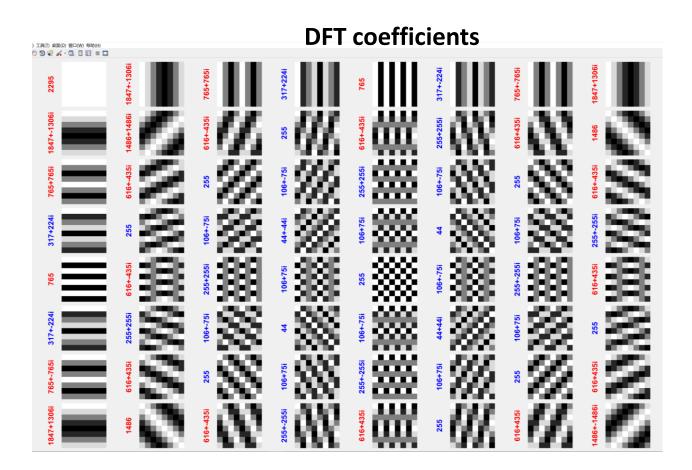
$$+ \frac{\sqrt{2}}{N} \sum_{v=1}^{N-1} F(0, v) \cos \frac{(2y+1)v\pi}{2N}$$

$$+ \frac{2}{N} \sum_{x=1}^{N-1} \sum_{y=1}^{N-1} F(u, v) \cos \frac{(2x+1)u\pi}{2N} \cos \frac{(2y+1)v\pi}{2N}$$

DFT example

Input image

DFT coefficients
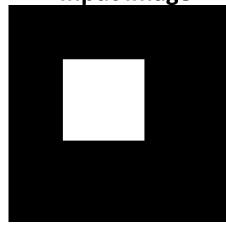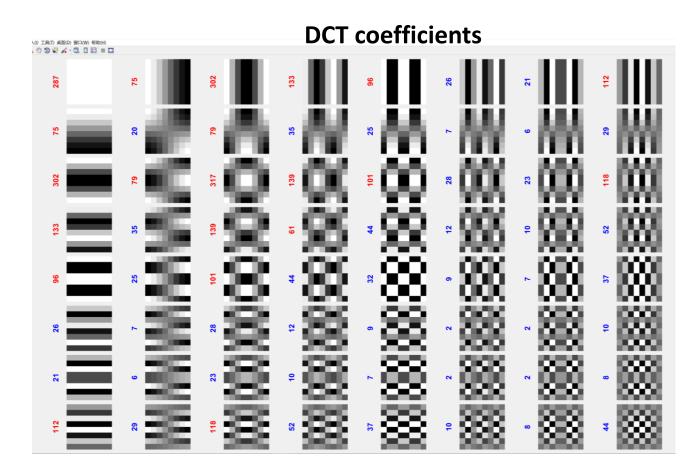
DCT example

Input image

DCT coefficients
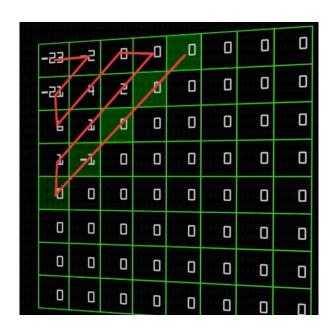
Good things about DCT

❑ **Positive：**

➤ Transform is real, $C^{-1} = C^T$ (unitary transform).

➤ Excellent energy compaction for nature images.

➤ Fast transform.

➤ JPEG algorithms.
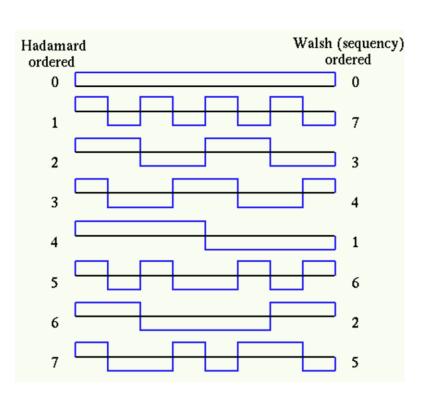
Walsh Transform

➢ Consist of ± 1 arranged in a checkerboard pattern.

➢ Transform:
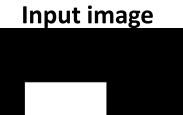
$$W(i) = \frac{1}{N} \sum_{t=0}^{N-1} f(t)\, Wal(i, t)$$

$$f(t) = \sum_{t=0}^{N-1} W(i)\, Wal(i, t)$$
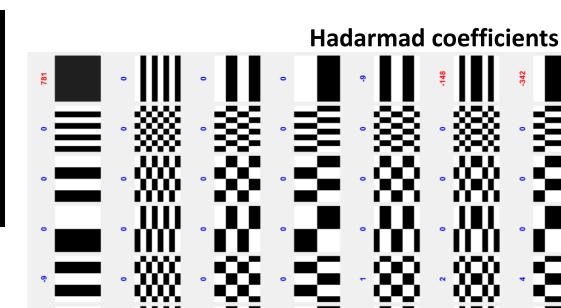
➢ Types of Wal(i, t).

➢ Walsh Ordering (沃尔什定序)

➢ Paley Ordering (佩利定序)

➢ Hadamard Matrix Ordering (哈达玛矩阵定序)

Hadamard Matrix Ordering

$$V_8 = \begin{pmatrix} W_4 & W_4 \\ W_4 & -W_4 \end{pmatrix} = \begin{pmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & -1 & 1 & -1 & 1 & -1 & 1 & -1 \\ 1 & 1 & -1 & -1 & 1 & 1 & -1 & -1 \\ 1 & -1 & -1 & 1 & 1 & -1 & -1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ 1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 \\ 1 & 1 & -1 & -1 & -1 & -1 & 1 & 1 \\ 1 & -1 & -1 & 1 & -1 & 1 & 1 & -1 \end{pmatrix}$$

Hadarmad Transform

Input image

Hadarmad coefficients

DFT, DCT, & Hadamard
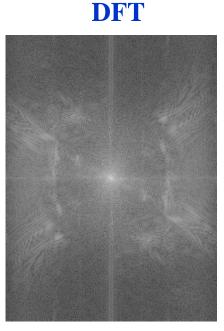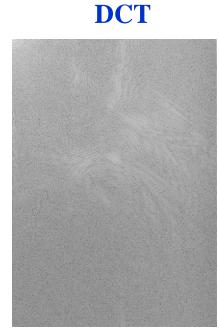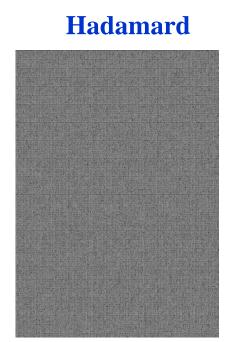
DFT **DCT** **Hadamard**

Any connection
between DFT and DCT?

Take home message

❑ The key idea for unitary transform is to find a proper basis for data decomposition.

❑ DCT provides better frequency consistency than DFT.

❑ Hadamard transform is able to present a simple image with simple coefficients. But can not keep energy compact for image full of details.

Wavelet transform Outline

❑ Discrete Wavelet Transform (DWT)（小波变换）

  ➢ An example for 1D-DWT

  ➢ Generalization of 1D-DWT

  ➢ 2D-DWT
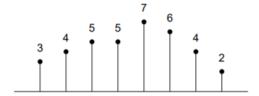
Discrete Wavelet Transform (DWT)
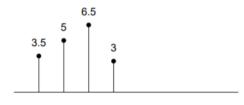
❑ Based on small waves called Wavelets-1) limited; 2) oscillation.

❑ Key idea: Translation & Scaling.

❑ Localized both time/space and frequency.

❑ Efficient for noise reduction and image compression.

❑ Two types of DWT one for image processing (easy invertible) and one for signal processing (invertible but computational expensive).

A simple example
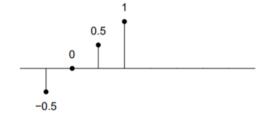
❑ **We can decompose an eight-point signal x(n):**

into two four-point signals:

$$c(n) = 0.5\,x(2n) + 0.5\,x(2n+1) \qquad d(n) = 0.5\,x(2n) - 0.5\,x(2n+1)$$

A simple example

➢ **The above process can be represented by a block diagram:**

$$x(n) \longrightarrow \boxed{\begin{array}{c}\text{AVE/}\\\text{DIFF}\end{array}} \begin{array}{c}\longrightarrow c(n)\\\longrightarrow d(n)\end{array}$$
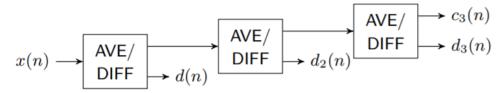
It is clear that this decomposition can be easily reversed:

$$y(2n) = c(n) + d(n)$$
$$y(2n + 1) = c(n) - d(n)$$

Which is also represented by a block diagram:

$$\begin{array}{c}c(n) \longrightarrow\\ d(n) \longrightarrow\end{array} \boxed{\text{INV}} \longrightarrow y(n)$$

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A simple example

❑ **When we repeat the simple AVE/DIFF signal decomposition:**

$$x(n) = [x_1, x_2, x_3, x_4, x_5, x_6, x_7, x_8] = [3, 4, 5, 5, 7, 6, 4, 2]$$

Level 1

$$c_1 = \frac{1}{2}[x_1 + x_2, x_3 + x_4, x_5 + x_6, x_7 + x_8]$$

$$d = d_1 = \frac{1}{2}[x_1 - x_2, x_3 - x_4, x_5 - x_6, x_7 - x_8]$$

Level 2

$$c_2 = \frac{1}{4}[x_1 + x_2 + x_3 + x_4, x_5 + x_6 + x_7 + x_8]$$

$$d_2 = \frac{1}{4}[x_1 - x_3 + x_2 - x_4, x_5 - x_7 + x_6 - x_8]$$
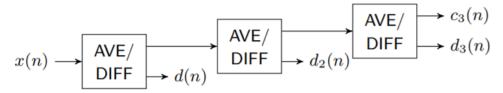
Level 3

$$c_3 = \frac{1}{8}[x_1 + x_2 + x_3 + x_4 + x_5 + x_6 + x_7 + x_8]$$

$$d_3 = \frac{1}{8}[x_1 + x_2 + x_3 + x_4 - x_5 - x_6 - x_7 - x_8]$$

A simple example

❑ **When we repeat the simple AVE/DIFF signal decomposition:**

The Haar wavelet representation of the eight-point signal x[n] is simply the set of four output signals produced by this three-level operation :

$$c_3 = [4.5]$$
$$d_3 = [-0.25]$$
$$d_2 = [-0.75, \quad 1.75]$$
$$d = [-0.5, \quad 0, \quad 0.5, \quad 1]$$

$$\mathbf{H}_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix} \begin{matrix} c_3 \\ d_3 \\ d_2(1) \\ d_2(2) \\ d(1) \\ d(2) \\ d(3) \\ d(4) \end{matrix}$$

Haar Transform matrix

➢ **When N=2 we have:**

$$\mathbf{H}_2 = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

➢ **When N=4 we have:**

$$\mathbf{H}_4 = \frac{1}{2} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & 1 & -1 & -1 \\ \sqrt{2} & -\sqrt{2} & 0 & 0 \\ 0 & 0 & \sqrt{2} & -\sqrt{2} \end{bmatrix}$$

➢ **When N=8 we have:**

$$\mathbf{H}_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix}$$

Haar Transform matrix

➢ **The family of N Haar functions $h_u(x), (u = 0, ..., N-1)$ are defined on the interval $0 \leq x \leq 1$. The shape of the specific function $h_u(x)$ of a given index u depends on two parameters p and q:**
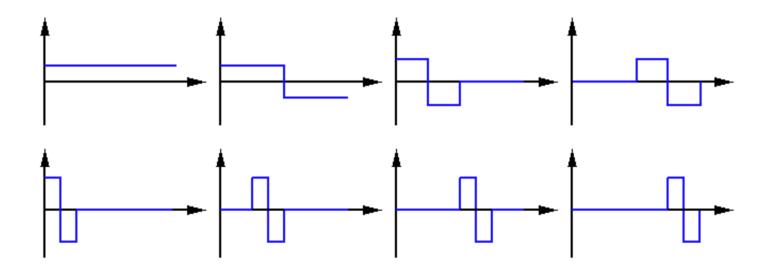
$$u = 2^p + q$$

u	p	q
1	0	0
2	1	0
3	1	1

➢ **The Haar basis functions are defined by:**

$$h_u(x) = \begin{cases} 1 & u = 0 \text{ and } 0 \leq x < 1 \\ 2^{p/2} & u > 0 \text{ and } q/2^p \leq x < (q+0.5)/2^p \\ -2^{p/2} & u > 0 \text{ and } (q+0.5)/2^p \leq x < (q+1)/2^p \\ 0 & \text{otherwise} \end{cases}$$

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Haar Transform matrix

$$\mathbf{H}_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix}$$

Generalization of 1D-DWT

➤ **Discrete Wavelet Transform (DWT):**

$$W_\varphi(j_0, k) = \frac{1}{\sqrt{M}} \sum_n f(n)\, \varphi_{j_0,k}(n)$$

$$W_\psi(j, k) = \frac{1}{\sqrt{M}} \sum_n f(n)\, \psi_{j,k}(n) \quad j \geq j_0$$

➤ **Inverse Discrete Wavelet Transform (IDWT):**

$$f(n) = \frac{1}{\sqrt{M}} \sum_k W_\varphi(j_0, k)\, \varphi_{j_0,k}(n) + \frac{1}{\sqrt{M}} \sum_{j=j_0}^{\infty} \sum_k W_\psi(j, k)\, \psi_{j,k}(n)$$

Where

$\varphi_{j_0,k}(n)$: scaling function (尺度函数)　　　　　$\psi_{j,k}(n)$: Wavelet (小波)

$W_\varphi(j_0, k)$: Approximation coefficients (近似系数)　　$W_\psi(j, k)$: detail coefficients (细节系数)

2D-DWT

➢ **Define 2D wavelet function: Directionally sensitive wavelet**

$$\psi^H(x,y) = \psi(x)\varphi(y) \qquad \psi^V(x,y) = \varphi(x)\psi(y) \qquad \psi^D(x,y) = \psi(x)\psi(y)$$

➢ **2D-DWT**

$$W_\varphi(j_0, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x,y)\, \varphi_{j_0,m,n}(x,y)$$

$$W_\psi(j, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x,y)\, \psi^i_{j,m,n}(x,y) \qquad i = \{H, V, D\}$$

➢ **2D-IDWT**

$$f(x,y) = \frac{1}{\sqrt{MN}} \sum_m \sum_n W_\varphi(j_0, m, n)\, \varphi_{j_0,m,n}(x,y)$$

$$+ \frac{1}{\sqrt{MN}} \sum_{i=\{H,V,D\}} \sum_{j=j_0}^{\infty} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} W_\psi(j, m, n)\, \psi^i_{j,m,n}(x,y)$$

Haar Transform matrix

Input: image size 8X8 I_{in}

Generate a Haar matrix of 8X8 as shown right
Then clip it into 4 part:
    $H_{L1}(dim = 4 * 8); H_{L2}(dim = 2 * 8);$
    $H_{L3}(dim = 1 * 8); L_{L3}(dim = 1 * 8);$.

$$H_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix}$$

H_{L3}
H_{L2}
H_{L1}

For computing *level 1* components:

LL_1 is down-sample of I_{in} on both X and Y direction	$HL_1 = H_{L1} * I_{in}$ + downsample on Y direction
$LH_1 = I_{in} * H_{L1}$ + downsample on X direction	$HH_1 = H_{L1} * I_{in} * H_{L1}$

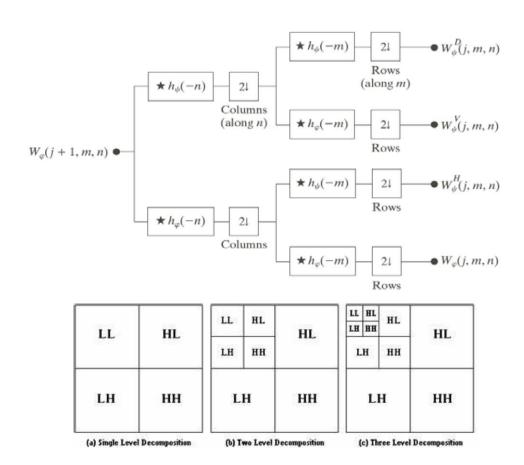For computing *level 2* components:

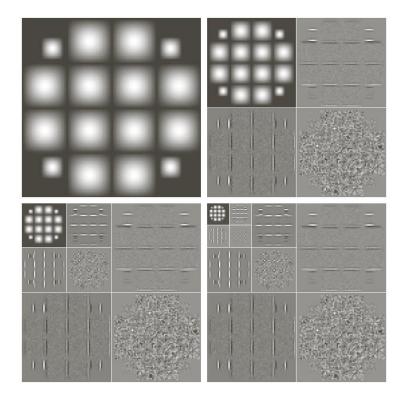|	
LL_2 is down-sample of LL_1 on both X and Y direction	$HL_2 = H_{L2} * I_{in}$ + downsample **twice** on Y direction
$LH_2 = I_{in} * H_{L2}$ + downsample **twice** on X direction	$HH_2 = H_{L2} * I_{in} * H_{L2}$

For computing *level 3* components:

LL_3 is down-sample of LL_2 on both X and Y direction	$HL_3 = H_{L3} * I_{in}$ + downsample **3 times** on Y direction
$LH_3 = I_{in} * H_{L3}$ + downsample **3 times** on X direction	$HH_3 = H_{L3} * I_{in} * H_{L3}$

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2D-DWT

(a) Single Level Decomposition

(b) Two Level Decomposition

(c) Three Level Decomposition

2D Haar Transform

$$A = \begin{pmatrix} 88 & 88 & 89 & 90 & 92 & 94 & 96 & 97 \\ 90 & 90 & 91 & 92 & 93 & 95 & 97 & 97 \\ 92 & 92 & 93 & 94 & 95 & 96 & 97 & 97 \\ 93 & 93 & 94 & 95 & 96 & 96 & 96 & 96 \\ 92 & 93 & 95 & 96 & 96 & 96 & 96 & 95 \\ 92 & 94 & 96 & 98 & 99 & 99 & 98 & 97 \\ 94 & 96 & 99 & 101 & 103 & 103 & 102 & 101 \\ 95 & 97 & 101 & 104 & 106 & 106 & 105 & 105 \end{pmatrix}$$

3-level Haar Transform for the first row

$r_1 = (88 \quad 88 \quad 89 \quad 90 \quad 92 \quad 94 \quad 96 \quad 97)$

Group r_1 in pair $[88, 88], [89, 90], [92, 94], [96, 97]$

$r_1 h_1 = (\underline{88 \quad 89.5 \quad 93 \quad 96.5} \quad \underline{0 \quad -0.5 \quad -1 \quad -0.5})$

Approximation coefficients   **Detail coefficients**

Group the first 4 columns in pair $[88, 89.5], [93, 96.5]$

$r_1 h_1 h_2 = (88.75 \quad 94.75 \quad -0.75 \quad -1.75 \quad 0 \quad -0.5 \quad -1 \quad -0.5)$

Group the first 2 columns in pair $[88, 94.75]$

$r_1 h_1 h_2 h_3 = (91.75 \quad -3 \quad -0.75 \quad -1.75 \quad 0 \quad -0.5 \quad -1 \quad -0.5)$

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2D Haar Transform

Repeat the same processing for all the columns and for the rows of the resulting matrix, we get
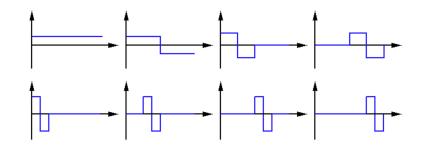
$$
\begin{pmatrix}
96 & -2.0312 & -1.5312 & -0.2188 & -0.4375 & -0.75 & -0.3125 & 0.125 \\
-2.4375 & -0.0312 & 0.7812 & -0.7812 & 0.4375 & 0.25 & -0.3125 & -0.25 \\
-1.125 & -0.625 & 0 & -0.625 & 0 & 0 & -0.375 & -0.125 \\
-2.6875 & 0.75 & 0.5625 & -0.0625 & 0.125 & 0.25 & 0 & 0.125 \\
-0.6875 & -0.3125 & 0 & -0.125 & 0 & 0 & 0 & -0.25 \\
-0.1875 & -0.3125 & 0 & -0.375 & 0 & 0 & -0.25 & 0 \\
-0.875 & 0.375 & 0.25 & -0.25 & 0.25 & 0.25 & 0 & 0 \\
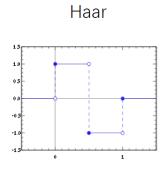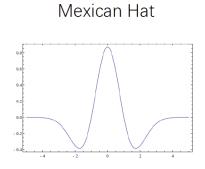-1.25 & 0.375 & 0.375 & 0.125 & 0 & 0.25 & 0 & 0.25
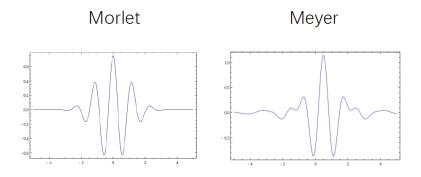\end{pmatrix}
$$

Mother Wavelet (母小波)

➢ **Mother Wavelet should satisfy:**

- $\int_{-\infty}^{\infty} |\psi(t)|^2 dt = 1$

- $\int_{-\infty}^{\infty} |\psi(t)| dt < \infty$

- $\int_{-\infty}^{\infty} \psi(t) dt = 0$

Haar · Mexican Hat · Morlet · Meyer

Take home message

❑ Based on small waves called Wavelets-1) limited; 2) oscillation.

❑ Key idea: Translation & Scaling.

❑ Localized both time/space and frequency.

❑ Efficient for noise reduction and image compression.

❑ JPEG2000, FBI finger printing databased.