

Channel Code

Random code

- Only used for proof, not in practice.
- Only good for long block length.
- Difficult to decode, requiring exhaustive search in an exponentially large codebook

In practice

- Finite block length (for low latency system, eg. Auto drive)
- Simple decoder (relaxing as computing power increase)

Hamming Code.

$$H = \begin{bmatrix} 0 & 0 & 0 & 1 & 1 & 1 & 1 \\ 0 & 1 & 1 & 0 & 0 & 1 & 1 \\ 1 & 0 & 1 & 0 & 1 & 0 & 1 \end{bmatrix}_{3 \times 7} \rightarrow \text{Parity check matrix}$$

- block length - $n = 7$.
- H contains all set of non-zero binary vectors of length 3
- Null space of H : $\{v : Hv = 0\}, v \in \mathbb{R}^{7 \times 1}$
has dimension $4 = n - \text{rank}(H)$

These > 4 codewords are

0000000	0100101	1000011	1100110
0001111	0101010	1001100	1101001
0010110	0110011	1010101	1110000
0011001	0111100	1011010	1111111

$Hc = 0$
column vector

properties of 2^4 codewords.

Codeword {

0000000	0100101	1000011	1100110
0001111	0101010	1001100	1101001
0010110	0110011	1010101	1110000
0011001	0111100	1011010	1111111

- Linear: any two codewords is also a codeword
- All codewords forms a linear subspace of dimension 4 in the vector space of length 7
- minimum weight:
 - ① minimum # of 1's in any codeword, except all-0 codeword, e.g. 0011001 \rightarrow 3
 - ② no two columns of H can add to 000
- minimum distance: minimum number of places in which two codewords differ.

Minimum Distance.

Codeword {

0000000	0100101	1000011	1100110
0001111	0101010	1001100	1101001
0010110	0110011	1010101	1110000
0011001	0111100	1011010	1111111

- Measure how far apart the codewords are.
- Determine how distinguishable the codewords will be at the output of channel.
- Aim to develop codes having large minimum distance
- For linear code, $\text{minimum weight} = \underline{\text{minimum distance}}$ of
 $000\ 000$ + any codeword with minimum weight

Decoder:

• Let

$$Q_i = (0 \ 0 \ \dots \ 1 \ 0 \ \dots) \quad \begin{matrix} \downarrow \\ i\text{-th position} \end{matrix}$$

• Received vector if i -th position has error :

$$r = c + e_i$$

$$r = c + e_i + e_j$$

• Check with H by

$$Hr = H(c + e_i) = Hc + H \cdot e_i$$

e.g. $H \cdot e_2 =$

$$H = \begin{bmatrix} 0 & 0 & 0 & 1 & 1 & 1 & 1 \\ 0 & 1 & 1 & 0 & 0 & 1 & 1 \\ 1 & 0 & 1 & 0 & 1 & 0 & 1 \end{bmatrix} \cdot \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$$

Nul space

The $\begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$ position has error.

info. bits \rightarrow Parity check bits

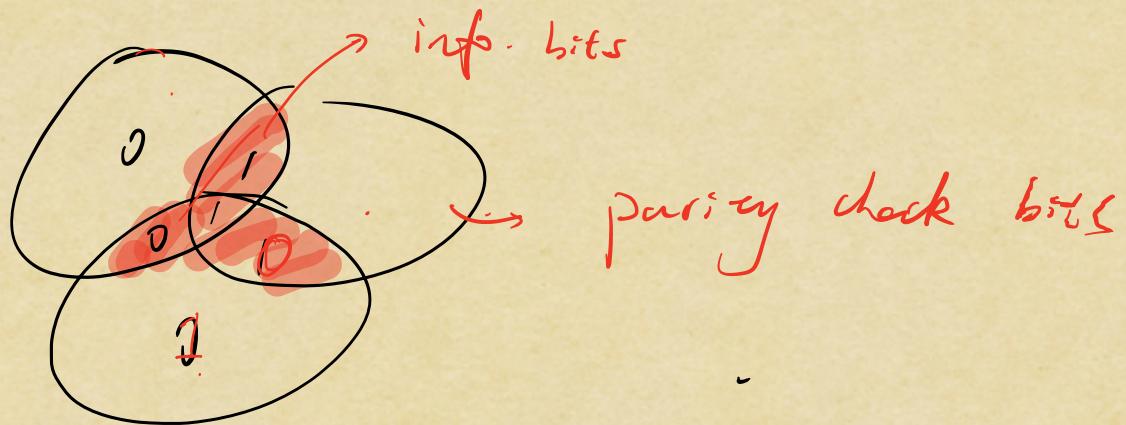
0000000	0100101	1000011	1100110
0001111	0101010	1001100	1101001
0010110	0110011	1010101	1110000
0011001	0111100	1011010	1111111

Systematic code

minimum distance

(n, k, d) or (n, k) code

codeword \downarrow info. bits
length length



In general. Hamming code

$$(n=2^l-1, k=2^l-l-1, d=3)$$

$$l=3, n=7, k=4, d=3.$$

$$R = \frac{4}{7} = \frac{60}{105}$$

$$l=4, n=15, k=15-4=11, d=3$$

$$R = \frac{11}{15} = \frac{77}{105}$$

Linear Block code — Digital Communications Jhon G. Proakis

- 1, codeword $C_m = (C_{m1}, C_{m2}, \dots, C_{mn})$ n codeword length., $C_{mj} \in \mathbb{F}_{2^b}$
- 2, (n, k) code. k : length of information bits.

Generator Matrix: G $k \times n$ dimension

0000000	0100101	1000011	1100110
0001111	0101010	1001100	1101001
0010110	0110011	1010101	1110000
0011001	0111100	1011010	1111111

$$C_m = U_m \cdot G \quad 1 \leq m \leq 2^k$$

U_m : length $1 \times k$ binary vector.

$$G = \begin{pmatrix} g_1 \\ g_2 \\ \vdots \\ g_k \end{pmatrix} \quad \{g_i\} \text{ are the codewords for message } (1, 0 \dots 0), (0100 \dots 0) \\ (0, 0, \dots, 1)$$

Thus, $C_m = \sum_{i=1}^k U_{mi} \cdot g_i \quad (GF(2)) \mod 2$

Property : ① $G = [I_k | P]$

I_k is $k \times k$ identity matrix, P is a $k \times (n-k)$ matrix.

This code is systematic code.

② In systematic code



→ information bits → parity check bits

③ $C : (n, k)$ code $\xrightarrow{\text{Nullspace}} (n : n-k)$ code C^\perp (dual code of C)

The generation matrix of C^\perp is call parity check matrix: H

$$C_m H^t = 0 \quad \forall C_m \in C$$

④ If $G = [I_k | P]$ then $H = [P^t | I_{n-k}]$

Hamming code

$$G = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 1 \end{bmatrix} = [I_3 | P] \rightarrow H = \begin{bmatrix} 0 & 0 & 1 & 1 & 1 & 1 & 0 & 0 \\ 1 & 0 & 0 & 1 & 1 & 0 & 1 & 0 \\ 1 & 1 & 0 & 0 & 1 & 1 & 0 & 1 \end{bmatrix}$$

$$\sum_{i=0}^{n-1} \binom{n}{i} = ? = 2^n$$

$$\left\{ \begin{array}{l} n = 2^e - 1 \quad \hookrightarrow H = \begin{bmatrix} 1 & \dots & 1 \\ 1 & \dots & 1 \\ \vdots & \ddots & \vdots \end{bmatrix}_{(e-1) \times n} \\ k = 2^e - e - 1 \quad \hookrightarrow k = n - \text{rank}(H) = n - e \end{array} \right.$$

1. Reed-Muller Code: (with flexible parameter)

Reed-Muller code: $n = 2^{m-1}$, $k = 2^{m-m-1}$, $d_m = 3$

RM code: $n = 2^m$, $k = \sum_{r=0}^m \binom{m}{r}$, $d_{\min} = 2^{m-r}$

1) G matrix:

$$G = \begin{pmatrix} G_0 \\ \vdots \\ G_k \end{pmatrix}, \quad \underline{m \gg r}$$

where

$$G_0 = (1 \ 1 \ 1 \ \dots \ 1)_{1 \times n}$$

$$G_1 = \begin{pmatrix} 0 & 0 & \dots & 1 \\ 0 & 1 & \dots & 1 \\ 0 & 0 & \dots & 1 \\ 0 & 1 & \dots & 1 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 1 & 2 & \dots & 2^{m-1} \end{pmatrix}_{m \times 2^m}$$

EXAMPLE 7.3-2. The first-order Reed-Muller code with block length 8 is an (8, 4) code with generator matrix

$$G = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 0 & 0 & 1 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \end{bmatrix} \quad \begin{array}{l} G_0 \\ G_1 \\ G_2 \end{array} \quad 2^m = n \Rightarrow m = 3$$
(7.3-9)

This code can be obtained from a (7, 3) maximum-length code by adding one extra parity bit to make the overall weight of each codeword even. This code has a minimum distance of 4. The second-order Reed-Muller code with block length 8 has the generator matrix

$$G = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 0 & 1 & 1 & 1 & 1 \\ 0 & 0 & 1 & 1 & 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 \end{bmatrix} \quad \begin{array}{l} G_0 \\ G_1 \\ G_2 \\ G_3 \end{array} \quad \begin{array}{l} 1 \times 8 \\ 4 \times 8 \\ 4 \times 8 \\ 1 \times 8 \end{array}$$
(7.3-10)

and has a minimum distance of 2.

$$R_{\text{RMM}} = \frac{2^{m-m-1}}{2^{m-1}}$$

$$R_{\text{RM}} = \frac{\sum_{r=0}^m \binom{m}{r}}{2^m}$$

G_2 is a $\binom{m}{2} \times n$ matrix whose rows are obtained by bitwise multiplication of two rows of G_1 .

$G_i : 2 \leq i \leq r$, is $\binom{m}{r} \times n$ matrix, bitwise
bitwise multiplication of r rows of G_1

↳ Hadamard Codes.

$$M_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix},$$

$$M_{2^n} = \begin{bmatrix} M_n & M_n \\ M_n & \bar{M}_n \end{bmatrix}_{2n \times 2n}$$

$$n = 2^m$$

\bar{M}_n : 0's is replaced by 1 and vice versa.

Each row represent as a code, not systematic code, $d_{\min} = \frac{n}{2}$

$$n = 2^m, \quad k = \log_2 2^n = m+1, \quad d_{\min} = \frac{n}{2} = 2^{m-1}$$

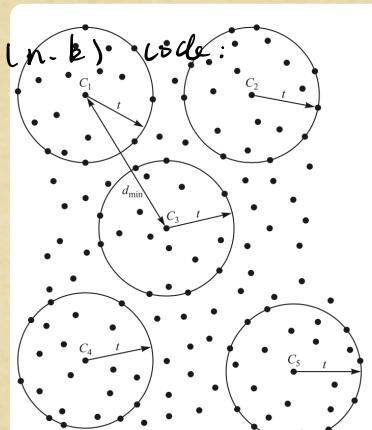


FIGURE 7.5-1
A representation of codewords as center of spheres with radius $t = \lfloor \frac{1}{2}(d_{\min} - 1) \rfloor$.

Detect
 $d_{\min} - 1$ errors
Correct
 $\frac{d_{\min} - 1}{2} + 1$ error

2^k codewords is a point of n -dimensional space; let t : the maximum # of det. error

$$\text{# of } \begin{cases} \text{err} \\ \text{corr} \end{cases} \quad d \geqslant -1$$

$$2^k \geq 2^{d-1} \quad d \geq 2t+1$$

LDPC (Low-density Parity-check)

- Robert Gallager, PhD Thesis, 1963
- A linear block code: $Hc=0$
- It can approach channel capacity
- Widely used in satellite communications, 5G, optic communications, etc
- Allow parallel decoding, and thus easy to implement.

Example:

Information bits: $b=(c_1, c_2, c_3)=(1 \ 1 \ 0)$

LDPC code: $C(s)=bG = (1 \ 1 \ 0 \ 0 \ 1 \ 0)$

$$C(s) = (c_1 \ c_2 \ c_3) \begin{pmatrix} 1 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 \end{pmatrix}$$

G

- (c_4, c_5, c_6) are parity check bits

$$c_4 = c_1 \oplus c_2$$

$$c_5 = c_2 \oplus c_3$$

$$c_6 = c_1 \oplus c_2 \oplus c_3$$

$$c_1 \oplus c_2 \oplus c_4 = 0$$

$$c_2 \oplus c_3 \oplus c_5 = 0$$

$$c_1 \oplus c_2 \oplus c_3 \oplus c_6 = 0$$

$$\begin{pmatrix} 1 & 1 & 0 & 1 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 \end{pmatrix} \begin{pmatrix} c_1 \\ c_2 \\ c_3 \\ c_4 \\ c_5 \\ c_6 \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \end{pmatrix}$$

H

- Note: $GH^T=0$

$$c4 = c1 \oplus c2$$

$$c5 = c2 \oplus c3$$

$$c6 = c1 \oplus c2 \oplus c3$$

$$\begin{pmatrix} 1 & 1 & 0 & 1 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 \end{pmatrix} \begin{pmatrix} c1 \\ c2 \\ c3 \\ c4 \\ c5 \\ c6 \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \end{pmatrix}$$

$$H = \begin{pmatrix} 1 & 1 & 0 & 1 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 \end{pmatrix}$$

Example:

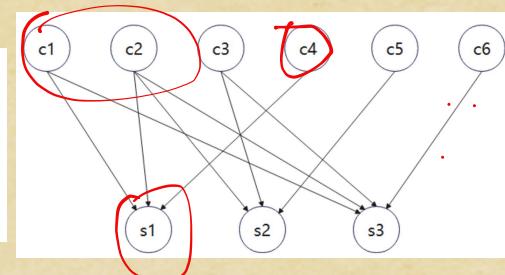
- Each information bit will be checked by at least two parity check bits.

$$c4 = c1 \oplus c2$$

$$\text{C1} \times \quad c1 \oplus c2 \oplus c4 = 1 \neq 0$$
$$c2 \oplus c3 \oplus c5 = 0$$

$$c6 = c1 \oplus c2 \oplus c3$$

$$c1 \oplus c2 \oplus c3 \oplus c6 = 1 \neq 0$$



- Decoding: Hard decoding and soft decoding
 - Hard Decoding: $s = Hr$, where r is the received code
 - Step 1: if $s=0$, correct decoding, otherwise
 - Step 2: $f_n = \sum_{m=1}^3 (s_m \cdot h_{mn}), n = 1, 2, \dots, 6$

if the f_n of r_n is larger than a threshold, then revert the bit

Cyclic Codes: C若为码字，则其所有循环位均分码字。

one an important class of linear block code.

codeword: $C = (c_{n-1}, c_{n-2}, \dots, c_1, c_0) = (c \in \mathbb{F}_{2^m})$.

$$C(x) = c_{n-1}x^{n-1} + c_{n-2}x^{n-2} + \dots + c_1x + c_0$$

$$x \cdot C(x) = c_{n-1}x^n + c_{n-2}x^{n-1} + \dots + c_0x$$

$$\frac{x \cdot C(x)}{x^{n+1}} = c_{n-1} + \frac{C'(x)}{x^{n+1}}, \quad \text{where } C'(x) = c_{n-2}x^{n-1} + \dots + c_1x + c_0$$

- ① Here $C'(x)$ represents codeword: $C' = (c_{n-2}, c_{n-3}, \dots, c_1, c_0)$, which is one unit shift of codeword $C = (c_{n-1}, c_{n-2}, \dots, c_0)$

codeword: $C = (c_{n-1}, c_{n-2}, \dots, c_1, c_0) = C \in \mathbb{F}_{10,14}$.

$$C(x) = c_{n-1}x^{n-1} + c_{n-2}x^{n-2} + \dots + c_1x + c_0$$

$$X \cdot C(x) = c_{n-1}x^n + c_{n-2}x^{n-1} + \dots + c_0x$$

$$\frac{X \cdot C(x)}{x^{n+1}} = c_{n-1} + \frac{C'(x)}{x^{n+1}}, \quad \text{where } C'(x) = c_{n-2}x^{n-1} + \dots + c_0x + c_{n-1}$$

$$\textcircled{2} \quad C''(x) = X \cdot C(x) \mod (x^{n+1}) \Leftrightarrow X \cdot C(x) = c_{n-1}(x^{n+1}) + C''(x)$$

$$C^{(i)}(x) = X^i C(x) \mod (x^{n+1}) \Leftrightarrow X^i \cdot C(x) = Q(x) \cdot (x^{n+1}) + C^{(i)}(x)$$

codeword 循环右移 i 位

\downarrow Quotient

$C^{(i)}(x)$ 为余式. remainder polynomial

\textcircled{3} generator polynomial (生成多项式) of degree $n-k$. ²⁾ is factor of (x^n+1)

$$g(x) = x^{n-k} + g_{n-k-1}x^{n-k+1} + \dots + g_1x + 1,$$

$$\text{i.e. } (x^{n+1}) = g(x) \cdot b(x)$$

message polynomial with message bits (u_{k-1}, \dots, u_0)

$$u(x) = u_{k-1}x^{k-1} + u_{k-2}x^{k-2} + \dots + u_1x + u_0 \Rightarrow 2^k \mid u_m(x)^4$$

$$m=1 \dots 2^k$$

degree $(u(x) \cdot g(x)) \leq n-1$, $u(x) \cdot g(x)$ represents a codeword

(4)
Genc:

$$c_m(x) = u_m(x) \cdot g(x) \quad m=1 \dots 2^k$$

Given a generator polynomial $g(x)$, and for each message $m \in G[2^k]$

$c_m(x) = u_m(x) \cdot g(x) \rightarrow$ its coefficients construct a codeword.

Check: Given any $c(c_{n-1}, c_{n-2}, \dots, c_0)$, see if 其循环移位是否为一个码字
 $c(x) = u(x) \cdot g(x)$

$c^{(1)}(x) = x \cdot c(x) + c_{n-1}x^{n-1}$, since $g(x)$ 整除 $x^n + 1$ 和 $c(x)$, we can write

$$c^{(1)}(x) = x(u(x)g(x) + c_{n-1}g(x) \cdot b(x)) = \underbrace{(u(x) + c_{n-1}b(x))}_{u_1(x)} \cdot g(x) \quad c^{(1)}(x) \text{ 可由 } u_m(x)g(x) \text{ 构成}$$

⑤ How to find $g(x)$?

$\deg(g(x)) = n-k$, and $\frac{f(x)}{x^n+1} = h(x)$, $g(x) \Rightarrow x^{n-k}$ 的同式.

e.g. $x^7+1 = (x+1)(x^3+x^2+1)(x^3+x+1)$. ∵ $g(x) = x^3+x^2+1 \Leftrightarrow g(x) = x^3+x+1$

Information Bits				Codewords							
X^3	X^2	X^1	X^0	X^6	X^5	X^4	X^3	X^2	X^1	X^0	
0	0	0	0	0	0	0	0	0	0	0	
0	0	0	1	0	0	0	1	1	0	1	
0	0	1	0	0	0	1	1	0	1	0	
0	0	1	1	0	0	1	0	1	1	1	
0	1	0	0	0	1	1	0	1	0	0	
0	1	0	1	0	1	1	1	0	0	1	
0	1	1	0	0	1	0	1	1	1	0	
0	1	1	1	0	1	0	0	0	1	1	
1	0	0	0	1	1	0	1	0	0	0	
1	0	0	1	1	1	0	0	1	0	1	
1	0	1	0	1	1	1	0	0	1	0	
1	0	1	1	1	1	1	1	1	1	1	
1	1	0	0	1	0	1	1	1	0	0	
1	1	0	1	1	0	1	0	0	0	1	
1	1	1	0	1	0	0	0	1	1	0	
1	1	1	1	1	0	0	1	0	1	1	

⑥ Decoder: $X^n+1 = g(x) \cdot h(x)$

奇偶校验多项式
 $\rightarrow h(x)$ is parity check polynomial
 or reciprocal

BCH code: 高效译码. n 中小值常用

$$n = 2^m - 1, \quad n-k \leq m+t, \quad d_{\min} \geq 2t+1$$

$$\frac{t}{n}$$

$$g(X) = \text{LCM} \{ \phi_\alpha(X), \phi_{\alpha^3}(X), \phi_{\alpha^5}(X), \dots, \phi_{\alpha^{2t-1}}(X) \} \quad (7.10-3)$$

LCM: 最小公倍式. $\alpha, \alpha^3, \alpha^5 \in GF(2^m)$

$$c(\alpha^i) = 0 \quad 1 \leq i \leq 2t \quad \text{monic}$$

$\phi_p(x) = p$ 的最小多项式. 即满足 $\phi_p(p)=0$ 的最低次非零多项式
 minimal polynomial $\Rightarrow \phi_p(x) = \prod_{i=0}^{l-1} (x+p^{2^i})$ l s.t. $p^{2^{l-1}}=1$ 称为
 最小多项式

$$\text{eg: } p = \alpha^5, \Rightarrow l=2 \quad l=2$$

$$\begin{aligned} \phi_p(x) &= (x+p)(x+p^2) = (x+\alpha^5)(x+\alpha^{10}) \\ &= x^2 + (\alpha^5 + \alpha^{10})x + \alpha^{15} = x^2 + x + 1 \end{aligned}$$

LDPC (Low-density Parity-check

Reed-Solomon code:

$$\begin{aligned}g(x) &= (x+a)(x+a^2) \cdots (x+a^{2t}) \\&= x^{2t} + g_{2t-1}x^{2t-1} \cdots + g_1x + g_0\end{aligned}$$

$$g_i \in GF(2^m)$$

$$n = 2^{m-1}, \quad n-k = 2t, \quad d_{min} = 2t+1 = n-k+1$$

数论
特征多项式