Rate Distortion Theory

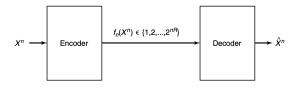
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- Quantization
- Definitions
- Calculation of the rate-distortion function
- Example: Gaussian Source with MSE distortion

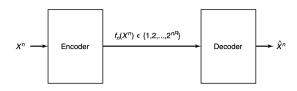
Rate Distortion Theory



Rate-distortion theory describes the trade-off between lossy compression rate and the resulting distortion.

- Lossless Source coding: Recover source data X without error
- Lossy source coding: Recover source with some error and distortion

Quantization

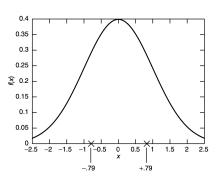


- \bullet Given a continuous RV X, if it is lossless source coding, we need infinite bits to perfectly recover X
- Question: What is the best possible representation of X for a given data rate?
- X: random variable to be represented
- $\hat{X}(X)$: representation of X
- R bits for the representation $\rightarrow |\hat{X}| = 2^{nR}$
- ullet Want to find the optimal set of values for \hat{X} and associated regions

Quantization Example: 1-bit Gaussian

Given $X \sim \mathcal{N}(0, \sigma^2)$ and $MSE = \|X - \hat{X}\|^2$, find \hat{X} such that 1) \hat{X} takes on two values; 2) minimizes MSE

$$X = \begin{cases} \sqrt{\frac{2}{\pi}\sigma}, & \text{if } X \ge 0\\ -\sqrt{\frac{2}{\pi}\sigma}, & \text{if } X < 0 \end{cases}$$
 (1)



Quantization

Objective: Map the incoming sequence U_1,U_2,\ldots into a sequence of discrete RVs V_1,V_2,\ldots , where V_m should represent U_m with as little distortion as possible.

- Scalar Quantization: Each analog RV in the sequence is quantized independently of the other RVs.
- Vector Quantization: The analog sequence is first segmented into blocks of n RVs each; then each n-tuple is quantized as a unit.

Scalar Quantization

- Partition the region $\mathbb R$ into M regions $\mathcal R_1,\ldots,\mathcal R_M$.
- Each region R_j is mapped to a symbol a_j called the representation point for \mathcal{R}_j .

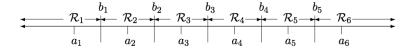
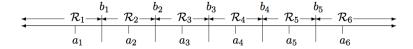


Figure: Quantization Region [Gallagar'Book]

- Each source value $u \in \mathcal{R}_j$ is mapped into the representation point a_j .
- After discrete coding and channel transmission, the receiver sees a_j and the distortion is $u-a_j$.

Scalar Quantization

- ullet View the source value u as a sample value of a RV U.
- The representation a_j is a sample value of the RV V, where V is the quantization of U. If $U \in \mathcal{R}_j$, then $V = a_j$
- The source sequence is U_1, U_2, \ldots The representation is V_1, V_2, \ldots , where if $U_k \in \mathcal{R}_j$, then $V_k = a_j$
- Aussme that U_1, U_2, \ldots is DMS
- ullet For a scalar quantizer, we can look at just a single U and a single V



Mean Square Distortion of a Scalar Quantizer

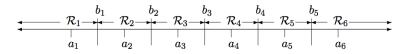
$$\mathsf{MSE} = E[(U - V)^2]$$

Aim: Given pdf $f_U(u)$ and alphabet size M, choose $\{\mathcal{R}_j, 1 \leq j \leq M\}$ and $\{a_j, 1 \leq j \leq M\}$ to miminize MSE.

Explore it into two ways:

- ullet Given a set of $\{a_j\}$, how should the intervals $\{\mathcal{R}_j\}$ be chosen?
- Given a set of intervals $\{\mathcal{R}_j\}$, how to choose $\{a_j\}$?

Choice of $\{\mathcal{R}_j\}$ for given $\{a_j\}$



Given a_j , choose b_j such that $E[(U-V)^2]$ is minimized. **Solution:**

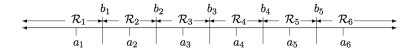
Choice of $\{\mathcal{R}_j\}$ for given $\{a_j\}$

Given a_j , choose b_j such that $E[(U-V)^2]$ is minimized.

Solution:

$$\begin{split} E[(U-V)^2] &= \sum_{j=1}^M \int_{\mathbf{R_j}} f_U(u) (u-a_j)^2 du = \sum_{j=1}^M \int_{b_{j-1}}^{b_j} f_U(u) (u-a_j)^2 du \\ &= \dots + \int_{b_{j-1}}^{b_j} f_U(u) (u-a_j)^2 du + \int_{b_j}^{b_{j+1}} f_U(u) (u-a_{j+1})^2 du + \dots \\ \text{Let } \frac{\partial E[(U-V)^2]}{\partial b_j} &= 0, \text{ we have } \big(\frac{\partial \int_{q(x)}^{g(x)} f(u) du}{\partial x} = f(g(x)) \frac{\partial g(x)}{\partial x} - f(q(x)) \frac{\partial q(x)}{\partial x} \big) \\ f_U(b_j) (b_j - a_j)^2 - f_U(b_j) (b_j - a_{j+1})^2 &= 0 \\ 2b_j (a_{j+1} - a_j) &= a_{j+1}^2 - a_j^2 \\ b_j &= \frac{a_j + a_{j+1}}{2} \end{split}$$

Choice of $\{\mathcal{R}_j\}$ for given $\{a_j\}$



- For source output u, squared error to a_j is $|u a_j|^2$
- Minimize by choosing closest a_j .
- Thus \mathcal{R}_j is a region closer to a_j than any a_i , for all $i \neq j$. (The boundary of b_j is between \mathcal{R}_j and \mathcal{R}_{j+1} must lie hafway between a_j and a_{j+1})
- ullet \mathcal{R}_j is bounded by

$$b_j = \frac{a_j + a_{j+1}}{2}$$

Choice of $\{a_j\}$ for given \mathcal{R}_j

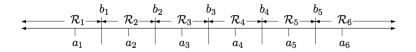
$$MSE = E[(U - V)^{2}] = \int_{-\infty}^{\infty} f(u)(u - v)^{2} du = \sum_{j=1}^{M} \int_{\mathcal{R}_{j}} f(u)(u - a_{j})^{2} du$$
$$= \sum_{j=1}^{M} \int_{\mathcal{R}_{j}} f(u)(u^{2} - 2a_{j}u + a_{j}^{2}) du$$

Let
$$\frac{\partial E[(U-V)^2]}{\partial a_j} = 0$$
, we have

$$-2\int_{\mathcal{R}_j} f(u)udu + 2\int_{\mathcal{R}_j} f(u)a_jdu = 0$$

$$\implies a_j = \frac{\int_{R_j} uf(u)du}{\int_{R_i} f(u)du}$$

Lloyd-Max Algorithm



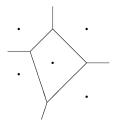
An optimal scalar quantizer must satisfy both $b_j=(a_j+a_{j+1})/2$ and $a_j=E[U_{(j)}].$

- 1. Choose $a_1 < a_2 < \cdots < a_M$
- 2. set $b_j = (a_j + a_{j+1})/2$ for $1 \le j \le M 1$
- 3. Set $a_j = \frac{\int_{\mathcal{R}_j} uf(u)du}{\int_{\mathcal{R}_j} f(u)du}$ where $\mathcal{R}_j = (b_{j-1}, b_j]$ for $1 \leq j \leq M-1$
- 4. Iterate on 2 and 3 until improvement is negligible.

It find local min, not necessarily global min.

Vector Quantization

Quantize n source variables at a time. (In scalar quantization, n=1)



Given $\{(a_j, a_j')\}$, how to choose $\{\mathcal{R}_j\}$

- The square error is $(u-a_j)^2+(u'-a_j')^2$, the point $\{a_j,a_j'\}$ which is the closest to (u,u') in Euclidean distance should be chosen.
- $\{\mathcal{R}_j\}$ contains points that are closer to (a_j, a'_j) than any other representation points, i.e., Voronoi regions.

Given a set of Voronoi region, how to find the $\{a_j, a_i'\}$?.

• Choose $\{a_j, a_j'\}$ to be the conditional means within those regions.

2D Quantization

The symbol + represents the updated point, and • the original point

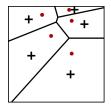


Figure: Iteration 1

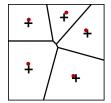


Figure: Iteration 3

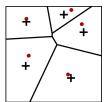


Figure: Iteration 2

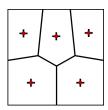
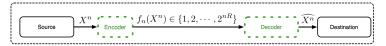


Figure: Iteration 4

More about Vector Quantization

- Popular research topic, related to deep learning algorithm
- ullet Quantizing complexity goes up exponentially with n
- Reduction in MSE with increasing n is quite modest
- Application: Video, Audio



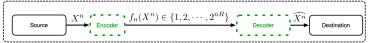
- X_1, X_2, \cdots, X_n i.i.d. $\sim p(x), x \in \mathcal{X}$
- A distortion function or distortion measure is a mapping

$$d: \mathcal{X} \times \hat{\mathcal{X}} \to \mathbb{R}^+$$

from the set of source alphabet-reproduction alphabet pairs into the set of non-negative real numbers. Measures the "cost" of representing symbol x by \hat{x} .

 A distortion measure is said to be bounded if the maximum value of the distortion is finite,

$$d_{max} := \max_{x \in \mathcal{X}, \hat{x} \in \hat{\mathcal{X}}} d(x, \hat{x}) < \infty$$



- Two most common distortion functions:
 - Hamming distortion:

$$d(x,\hat{x}) = \begin{cases} 0 \text{ if } x = \hat{x} \\ 1 \text{ if } x \neq \hat{x} \end{cases}$$

- Squared-error distortion:

$$d(x,\hat{x}) = (x - \hat{x})^2$$

ullet We define the distortion between sequences x^n and \hat{x}^n as

$$d(x^n, \hat{x}^n) = \frac{1}{n} \sum_{i=1}^n d(x_i, \hat{x}_i).$$

Definitions



 \bullet A $(2^{nR},n)\text{-}rate$ distortion code consists of an encoding function

$$f_n: \mathcal{X}^n \to \{1, 2, \cdots, 2^{nR}\},$$

and a decoding (reproduction) function,

$$g_n:\{1,2,\cdots,2^{nR}\}\to\hat{\mathcal{X}}^n.$$

• The distortion associated with the $(2^{nR}, n)$ code is defined as

$$D = E[d(X^n, g_n(f_n(X^n))),$$

where the expectation is with respect to the probability distribution on \mathcal{X} ,

$$D = \sum p(x^n)d(x^n, g_n(f_n(x^n))).$$

• The set of n-tuples $g_n(1), g_n(2), \cdots, g_n(2^{nR})$, denoted by $\hat{X}^n(1), \hat{X}^n(2), \cdots, \hat{X}^n(2^{nR})$ constitutes the codebook and $f_n^{-1}(1), \cdots f_n^{-1}(2^{nR})$ are the associated assignment regions.

- A rate-distortion pair (R,D) is said to be achievable if there exists a sequence of $(2^{nR},n)$ -rate distortion codes (f_n,g_n) with $\lim_{n\to\infty} E[d(X^n,g_n(f_n(X^n)))] \leq D$.
- The rate-distortion region for a source is the closure of the set of achievable rate distortion pairs (R, D).
- The rate-distortion function R(D) is the **infimum** of rates R such that (R, D) is in the rate distortion region of the source for a given distortion D.
- The distortion-rate function D(R) is the **infimum** of all distortions D such that (R, D) is in the rate distortion region of the source for a given rate R.

Main Results

Theorem: The rate distortion function for an i.i.d. source X with distributed p(x) and bounded distortion function $d(x, \hat{x})$ is equal to the associated information rate distortion function. Thus,

$$R(D) = R^{(I)}(D) = \min_{p(\hat{x}|x): \sum_{x,\hat{x}} p(x)p(\hat{x}|x)d(x,\hat{x}) \le D} I(X;\hat{X})$$

is the minimum achievable rate at distortion D.

Theorem: The rate distortion function for a Bernoulli(p) source with Hamming distortion is given by

$$R(D) = \left\{ \begin{array}{ll} H(p) - H(D), & 0 \leq D \leq \min\{p, 1-p\} \\ 0, & D \, \min\{p, 1-p\} \end{array} \right.$$

Theorem: The rate distortion function for a $\mathcal{N}(0, \sigma^2)$ source with squared-error distortion is given by

$$R(D) = \left\{ \begin{array}{ll} \frac{1}{2} \log \frac{\sigma^2}{D}, & 0 \le D \le \sigma^2 \\ 0, & D > \sigma^2 \end{array} \right\}$$

The key idea of the Proof:

- ullet Converse: Find a lower bound on $I(X;\hat{X})$
- Achievability: Show that the lower bound is achievable

R(D) of Gaussian and Binary Sources

$$R(D) = \begin{cases} \frac{1}{2} \log \frac{\sigma^2}{D}, & 0 \le D \le 0, \\ 0, & D > \sigma^2 \end{cases}$$

Figure: Gaussian Source $X \sim \mathcal{N}(0, \sigma^2)$ with MSE distortion D

$$R(D) = \begin{cases} \frac{1}{2} \log \frac{\sigma^2}{D}, & 0 \le D \le \sigma^2 \\ 0, & D > \sigma^2 \end{cases} \quad R(D) = \begin{cases} H(p) - H(D), & 0 \le D \le \min\{p, 1 - p\} \\ 0, & D > \min\{p, 1 - p\} \end{cases}$$

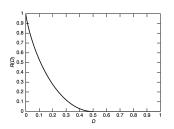


Figure: Binary Source $X \sim \text{Bern}(p)$ with Hamming distortion D

Converse Proof: R(D) of Gaussian

Converse:

$$\begin{split} I(X; \hat{X}) &= h(X) - h(X|\hat{X}) \\ &= \frac{1}{2} \log(2\pi e) \sigma^2 - h(X - \hat{X}|\hat{X}) \\ &\geq \frac{1}{2} \log(2\pi e) \sigma^2 - h(X - \hat{X}) \\ &\geq \frac{1}{2} \log(2\pi e) \sigma^2 - h(\mathcal{N}(0, E(X - \hat{X})^2)) \\ &= \frac{1}{2} \log(2\pi e) \sigma^2 - \frac{1}{2} \log(2\pi e) E(X - \hat{X})^2 \\ &\geq \frac{1}{2} \log(2\pi e) \sigma^2 - \frac{1}{2} \log(2\pi e) D \\ &= \frac{1}{2} \log \frac{\sigma^2}{D}, \end{split}$$

Achievability Proof: R(D) of Gaussian

Achievability: Find a $f(x|\hat{x})$ to achieve the lower bound

- If $D < \sigma^2$
 - Choose $X=\hat{X}+Z$, where $\hat{X}\sim\mathcal{N}(0,\sigma^2-D)$, $Z\sim\mathcal{N}(0,D)$ are independent
 - With the above choice, $I(X;\hat{X})=\frac{1}{2}\log\frac{\sigma^2}{D}$, and $E(X-\hat{X})^2=D$
- If $D > \sigma^2$, we choose $\hat{X} = 0$ with probability 1, then R(D) = 0