#### **DFT**

• *N*-point signals

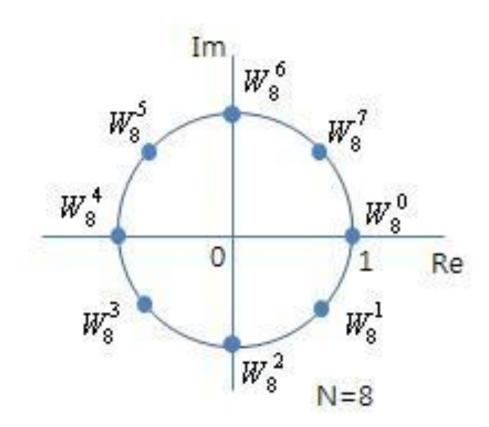
DFT 
$$X[k] = \sum_{n=0}^{N-1} x[n]W_N^{kn}, \qquad k = 0,1...,N-1$$

IDFT (inverse DFT)  $x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k]W_N^{-kn}, \qquad n = 0,1...,N-1$ 

where  $W_N = e^{-j(2\pi/N)}$ , and  $W_N^n$  are the roots of the polynomial  $W^N = 1$ .

## What is $W_N^n$ ? n-th Root of **1**

- $W_N^n$  is the n-th root of the equation  $W^N = 1$ .
- Eg., when *N*=8,



#### **Fast Fourier Transform (FFT)**

- An important characteristic of DFT is that it can be computed very fast, referred to as FFT.
- So, the spectrogram are often computed using FFT.

DFT pairs: 
$$X[k] = \sum_{n=0}^{N-1} x[n]W_N^{kn}$$
,  $k = 0,1...,N-1$  
$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k]W_N^{-kn}$$
,  $n = 0,1...,N-1$ 

 $W_N = e^{-j2\pi/N}$  is a root of the equation  $W^{N}=1$ .

Originally, it requires  $N^2$  complex multiplications and (N-1)N complex additions for direct computation.

### Fast Fourier Transform (FFT)

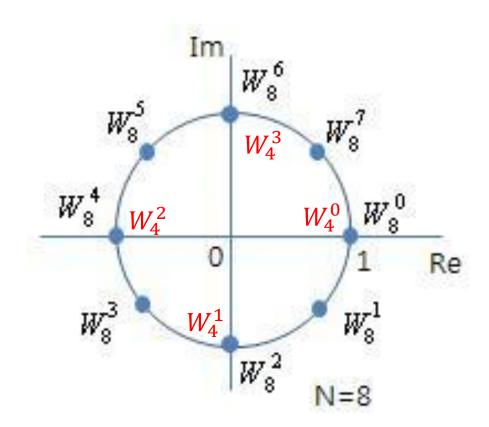
- One of the important algorithms for FFT is the Cooley-Tukey algorithm.
- The key principle used in the Cooley-Tukey algorithm is

$$- W_N^2 = W_{N/2}$$
(eg.,  $W_8^2 = W_4$ )

Divide and concur (iteratively)

### What is $W_N^n$ ? n-th Root of **1**

- $W_N^n$  is the n-th root of the equation  $W^N = 1$ .
- Eg., when n=8,



### **Decimation-in-time FFT algorithm**

Most conveniently illustrated by considering the special case of N an integer power of 2, i.e,  $N=2^v$ .

Since N is an even integer, we can compute X[k] by separating x[n] into two (N/2)-point sequences consisting of the even numbered point in x[n] and the odd-numbered points in x[n], respectively

$$X[k] = \sum_{n \text{ even}} x[n]W_N^{nk} + \sum_{n \text{ odd}} x[n]W_N^{nk}$$

### **Decimation-in-time FFT algorithm**

With the substitution of variable n=2r for n even and n=2r+1 for n odd:

$$X[k] = \sum_{r=0}^{(N/2)-1} x[2r]W_N^{2rk} + \sum_{r=0}^{(N/2)-1} x[2r+1]W_N^{(2r+1)k}$$

By the key property:  $W_N^2=e^{-2j(2\pi/N)}=e^{-j2\pi/(N/2)}=W_{N/2}$  We have

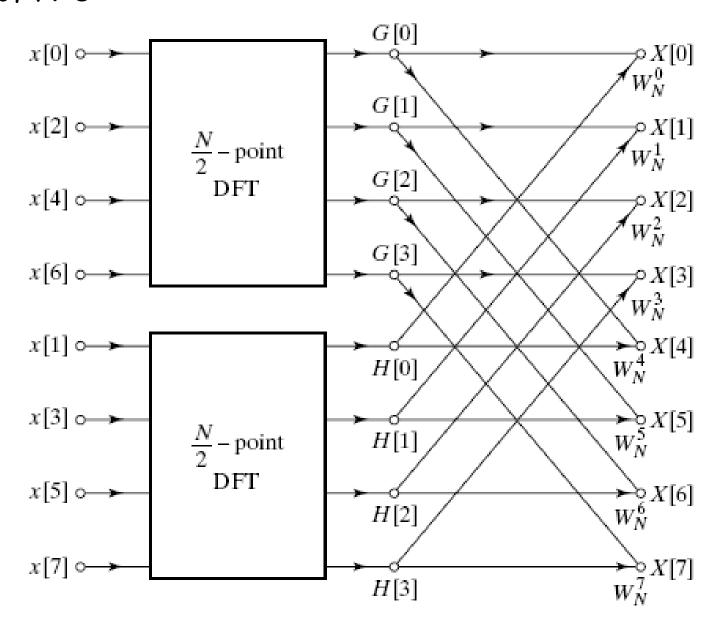
$$X[k] = \sum_{r=0}^{(N/2)-1} x[2r](W_N^2)^{rk} + W_N^k \sum_{r=0}^{(N/2)-1} x[2r+1](W_N^2)^{rk}$$
$$= \sum_{r=0}^{(N/2)-1} x[2r](W_{N/2})^{rk} + W_N^k \sum_{r=0}^{(N/2)-1} x[2r+1](W_{N/2})^{rk}$$

In sum,

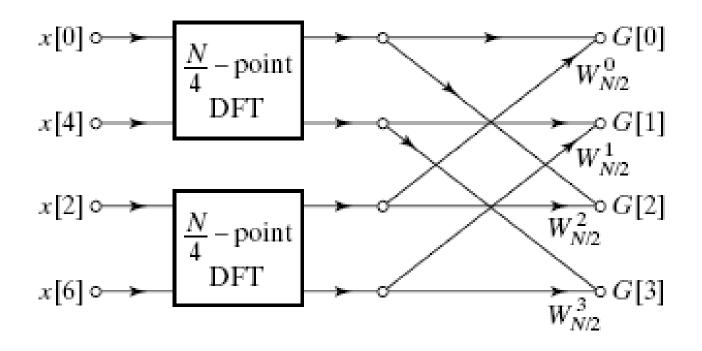
$$X[k] = \sum_{r=0}^{(N/2)-1} x[2r](W_{N/2})^{rk} + W_N^k \sum_{r=0}^{(N/2)-1} x[2r+1](W_{N/2})^{rk}$$
$$= G[k] + W_N^k H[k], \quad k = 0,1,...,N-1$$

- Both G[k] and H[k] can be computed by (N/2)-point DFT
- G[k]: the (N/2)-point DFT of the even numbered points of the original sequence
- H(k): the (N/2)-point DFT of the odd-numbered points of the original sequence.
- Although the index ranges over N values, k = 0, 1, ..., N-1, they must be computed only for k between 0 and (N/2)-1, since G[k] and H[k] are each periodic in k with period N/2.

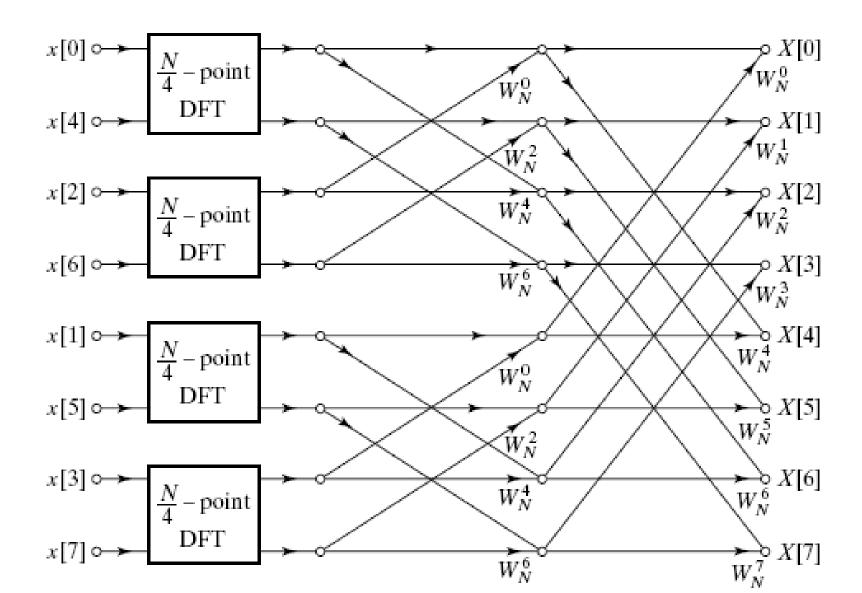
# Decomposing N-point DFT into two (N/2)-point DFTs for the case of N=8



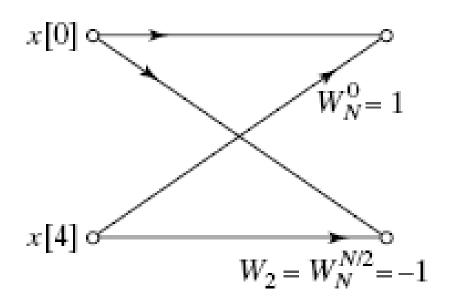
We can further decompose the (N/2)-point DFT into two (N/4)-point DFTs. For example, the upper half of the previous diagram can be decomposed as



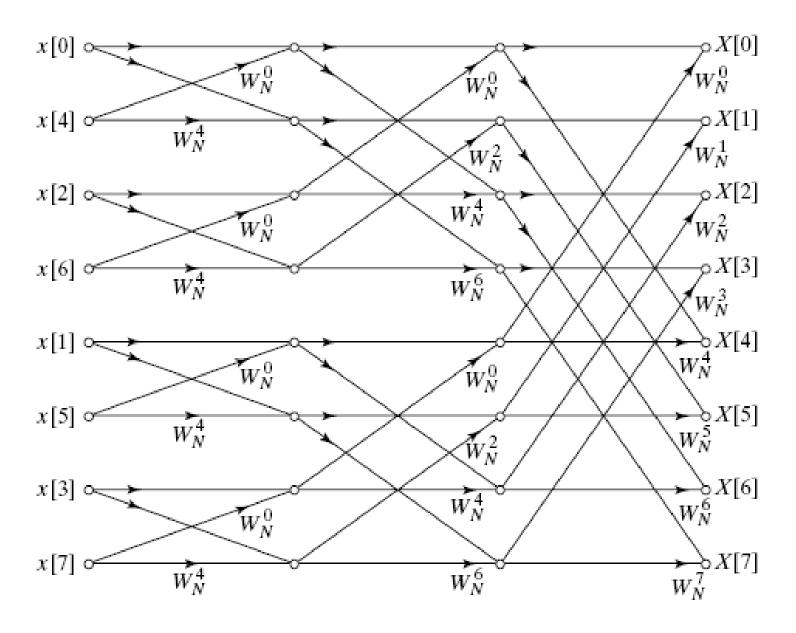
Hence, the 8-point DFT can be obtained by the following diagram with four 2-point DFTs.



Finally, each 2-point DFT can be implemented by the following signal-flow graph, where no multiplications are needed.



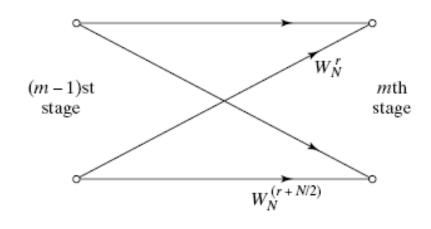
Flow graph of a 2-point DFT



Flow graph of complete decimation-in-time decomposition of an 8-point DFT.

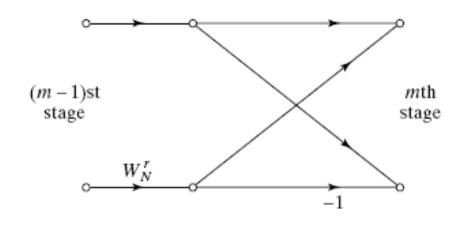
In each stage of the decimation-in-time FFT algorithm, there are a basic structure called the butterfly computation:

$$X_{m}[p] = X_{m-1}[p] + W_{N}^{r} X_{m-1}[q]$$
$$X_{m}[q] = X_{m-1}[p] - W_{N}^{r} X_{m-1}[q]$$

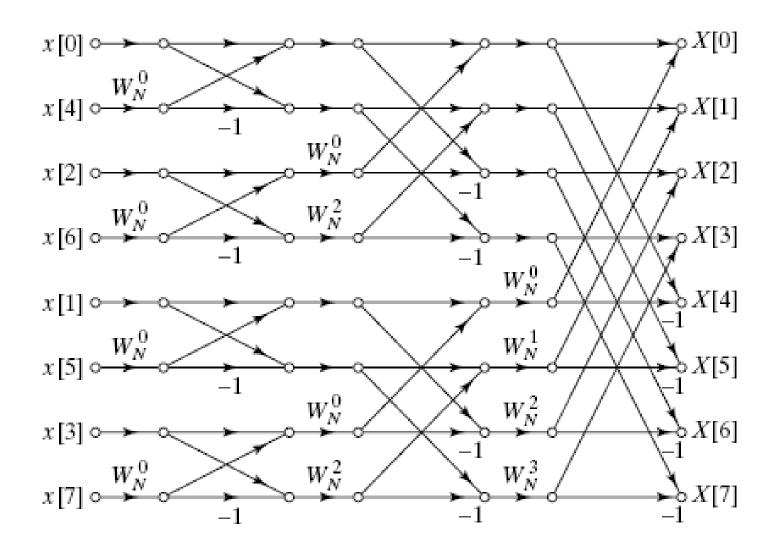


Flow graph of a basic butterfly computation in FFT.

The butterfly computation can be simplified as follows:



Simplified butterfly computation.



Flow graph of 8-point FFT using the simplified butterfly computation

- In the above, we have introduced the decimation-in-time algorithm of FFT.
- Here, we assume that N is the power of 2. For N=2<sup>v</sup>, it requires v=log<sub>2</sub>N stages of computation.
- The number of complex multiplications and additions required was N+N+...N = Nv = N log<sub>2</sub>N.

 In practice, if N is not the power of 2, we can use zero padding to complement zeros to its rear to result in an length-M sequence, where M is the smallet poaer-of-2 integer that is larger than N. Then, we compute the Mpoint FFT instead.

- When N is not the power of 2, we can also apply the same principle that were applied in the power-of-2 case when N is a composite integer. For example, if N=RQ, it is possible to express an N-point DFT as either the sum of R Q-point DFTs or as the sum of Q R-point DFTs.
- The FFT algorithm of power-of-two is also called the Cooley-Tukey algorithm since it was first proposed by them.