

# Fourier Transform for Continuous Functions (or Signals)

- How to define the frequency or spectrum for general continuous-time signals?
- Central goal: representing a signal by a set of orthogonal bases corresponding to frequencies or spectrum.
- Fourier series allows to find the spectrum of only periodic functions.

# From Periodic to Non-Periodic

- Fourier series transforms a periodic continuous signal into the frequency domain.
- What will happen when the continuous signal is not periodic?
- Consider the period of a signal with the fundamental frequency  $\omega_0$ .
- $T_0$  specifies the fundamental period,

$$T_0 = \frac{2\pi}{\omega_0} = \frac{1}{f}$$

# Review of Fourier Series

- Fourier series representation of a periodic signal  $x_{T_0}(t)$  can be given by the pair of equations

Forward  
Transform

$$a_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt$$

Integrals over a period  $[0, T_0]$  and  $[-T_0/2, T_0/2]$  are the same

Inverse  
Transform

$$x_{T_0}(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

where  $\omega_0 = \frac{2\pi}{T_0}$

# Imaging $T_0 \rightarrow \infty$

- A non-periodic signal can be conceptually thought of as a periodic signal whose fundamental period  $T_0$  is infinitely long,

$$T_0 \rightarrow \infty.$$

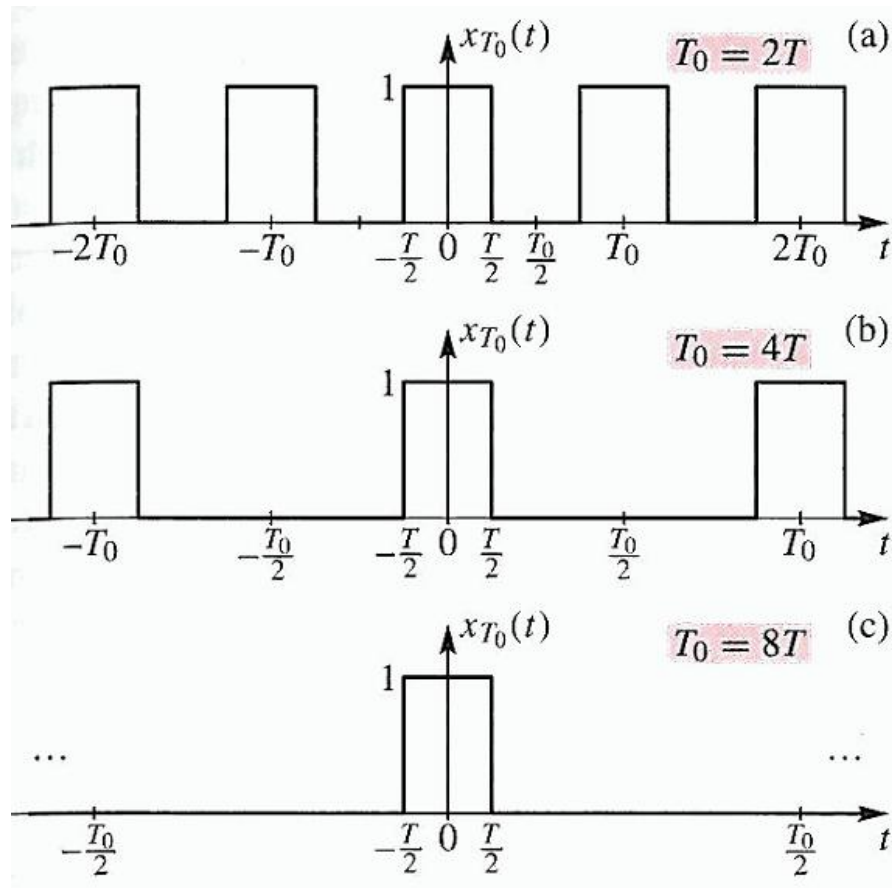
- In this case, the fundamental frequency

$$\omega_0 \rightarrow 0.$$

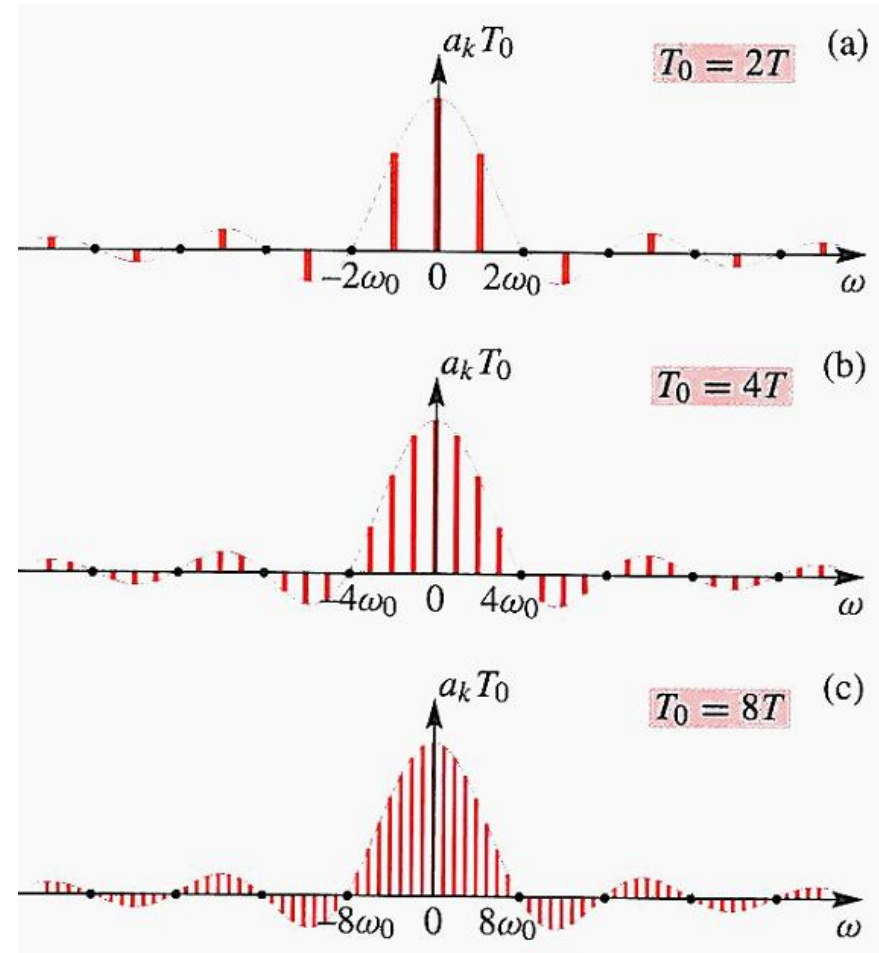
# Interval between adjacent frequency

- Remember that the spectrum (in the frequency domain) of a periodic continuous signal is discrete, specified by  $k\omega_0$  ( $k$  is an integer).
- Therefore, the **interval** between adjacent frequencies,  $k\omega_0$  and  $(k + 1)\omega_0$ , **is just  $\omega_0$** .

# Example: recall the case of squared wave



Time domain



Fourier series domain 6

# Notation change of the forward transform of Fourier series

- Because  $\frac{1}{T_0} = \frac{\omega_0}{2\pi}$ , let us re-denote  $f(t) = x_{T_0}(t)$  and **rewrite the forward transform of Fourier series as**

$$\begin{aligned} a_k &= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt \\ &= \frac{\omega_0}{2\pi} \int_{-\pi/\omega_0}^{\pi/\omega_0} f(t) e^{-jk\omega_0 t} dt \end{aligned}$$

# Extreme Case of Fourier Series: $T_0 \rightarrow \infty$ (i.e., $\omega_0 \rightarrow 0$ )

- **Further changing the notation:**
- Let  $b_{k\omega_0} = a_k$ , the forward transform becomes

$$b_{k\omega_0} = \frac{\omega_0}{2\pi} \int_{-\pi/\omega_0}^{\pi/\omega_0} f(t) e^{-jk\omega_0 t} dt$$

- The inverse transform becomes

$$f(t) = \sum_{k=-\infty}^{\infty} b_{k\omega_0} e^{jk\omega_0 t}$$



## Extreme Case of Fourier Series: $\omega_0 \rightarrow 0$

- When  $\omega_0 \rightarrow 0$ , we can *image* that the frequency becomes continuous:
- The interval  $\omega_0$  becomes  $d\omega$
- Let  $\omega = k\omega_0$ . Then, when  $\omega_0 \rightarrow 0$ , the forward transform approaches

$$b(\omega) = \frac{d\omega}{2\pi} \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

# Extreme Case of Fourier Series

- Combining with the inverse transform,

$$f(t) = \sum_{k=-\infty}^{\infty} b(w)e^{j\omega t}$$

- we have

$$f(t) = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} \left[ \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt \right] e^{j\omega t}$$

Define this as a transform

$b(\omega)$

when  $\omega_0 \rightarrow 0$ , where the **summation** in the inverse transform **becomes integral**.

# It becomes:

## Continuous Fourier Transform

- Consider the above recovering equation of  $f(t)$ . Let us decompose the equation into **forward transform** and **inverse transform**:

$$f(t) = \int_{-\infty}^{\infty} \frac{1}{2\pi} \left[ \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt \right] e^{j\omega t} d\omega$$

Forward Transform of  $f(t)$  to  
the frequency domain  $F(\omega)$

Inverse Transform of  $F(\omega)$  to the time domain  $f(t)$

# Continuous Fourier Transform

a.k.a. Continuous-time Fourier Transform

- Forward Transform

$$F(j\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

- Inverse Transform

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega$$

- Remark: in the continuous domain,  $e^{j\omega t}$  ( $\omega \in \mathbb{R}$ ) still form a set of orthogonal bases (with the amount uncountable infinite), no matter whether  $\omega$  is a multiple of an integer or continuous real value.

# Continuous Fourier Transform

- Both time and frequency domains in continuous Fourier transform are continuous.
- The frequency waveform is also referred to as the 'spectrum'.

# Continuous Fourier Transform

- Continuous Fourier transform is **the most fundamental Fourier transform**. **We will see later that other Fourier transforms (including Fourier series) are its special cases.**
- It is also the most 'standard' Fourier transform. When there is no specification, we usually referred to Fourier Transform as this type.

# Continuous Fourier Transform Pair

- Transform pair

**Forward**

$$F(j\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

**Backward**

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega$$

- Sometimes also written as

$$F(j\omega) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$f(t) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega$$

depending on how we decompose the normalization constant  $1/(2\pi)$ .

## Note that there are Many Variations of the Forms of Continuous F. T. (example from Kuhn's slides 2005)

### Recall: Fourier transform

The Fourier integral transform and its inverse are defined as

$$\mathcal{F}\{g(t)\}(\omega) = G(\omega) = \alpha \int_{-\infty}^{\infty} g(t) \cdot e^{\mp j\omega t} dt$$

$$\mathcal{F}^{-1}\{G(\omega)\}(t) = g(t) = \beta \int_{-\infty}^{\infty} G(\omega) \cdot e^{\pm j\omega t} d\omega$$

where  $\alpha$  and  $\beta$  are constants chosen such that  $\alpha\beta = 1/(2\pi)$ .

Many equivalent forms of the Fourier transform are used in the literature, and there is no strong consensus on whether the forward transform uses  $e^{-j\omega t}$  and the backwards transform  $e^{j\omega t}$ , or vice versa. Some authors set  $\alpha = 1$  and  $\beta = 1/(2\pi)$ , to keep the convolution theorem free of a constant prefactor; others use  $\alpha = \beta = 1/\sqrt{2\pi}$ , in the interest of symmetry.

The substitution  $\omega = 2\pi f$  leads to a form without prefactors:

$$\mathcal{F}\{h(t)\}(f) = H(f) = \int_{-\infty}^{\infty} h(t) \cdot e^{\mp 2\pi jft} dt$$

$$\mathcal{F}^{-1}\{H(f)\}(t) = h(t) = \int_{-\infty}^{\infty} H(f) \cdot e^{\pm 2\pi jft} df$$



# Symmetry between Time and Frequency of Continuous Fourier Transform

- Unlike Fourier series, the continuous Fourier transform has **very similar forward and inverse transforms**.
- Except to the normalization constant, the only difference is that **the forward uses  $-j$**  and the **inverse uses  $j$**  in the complex exponential basis.
- This suggests that **the roles of time and frequency can be exchanged**, and some properties are symmetric to each other.

# Existence and Convergence of Fourier Transform

- We have 'derived' continuous Fourier transform as an extreme extension of Fourier series.
- To ensure the existence of continuous Fourier transform, we should consider the conditions where the integrals exist.
- A sufficient condition is

*Sufficient Condition  
for Existence of  $X(j\omega)$*

$$\int_{-\infty}^{\infty} |x(t)| dt < \infty$$

# Rational

$$\begin{aligned} |X(j\omega)| &= \left| \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \right| \\ &\leq \int_{-\infty}^{\infty} |x(t) e^{-j\omega t}| dt = \int_{-\infty}^{\infty} |x(t)| dt \end{aligned}$$

- Hence, if  $\int_{-\infty}^{\infty} |x(t)| dt < \infty$ , or equivalently, the integral is

bounded, then the continuous Fourier transform is also bounded.

# Existence and Convergence of Fourier Transform

- The above is **a sufficient condition** but **not a necessary condition**.
- Many functions that do not satisfy the above condition, **but** we can still obtain a useful Fourier transform representation particularly when **the impulse signals** are allowed to be used.
- **The impulse signals will be introduced in the next course.**
  - In engineering, we usually do not care much about the exact necessary and sufficient conditions, despite there are mathematically rigorous ways to specify these conditions.

# Examples of Fourier Transform Pairs

- Rectangular function (rectangular pulse signal)

$$x(t) = \begin{cases} 1 & -\frac{1}{2}T \leq t < \frac{1}{2}T \\ 0 & \text{otherwise} \end{cases}$$

- Derivation of its continuous F. T.

$$\begin{aligned} X(j\omega) &= \int_{-T/2}^{T/2} e^{-j\omega t} dt \\ &= \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-T/2}^{T/2} = \frac{e^{-j\omega T/2} - e^{j\omega T/2}}{-j\omega} \\ &= \frac{\sin(\omega T/2)}{\omega/2} \quad (\text{Sinc function}) \end{aligned}$$

Note that in this case, the Fourier transform is a real function. The phase is zero

# Fourier transform of rectangular function

- Rectangular function can also be represented by the unit-pulse function  $u(t)$  as

$$u\left(t + \frac{T}{2}\right) - u\left(t - \frac{T}{2}\right)$$

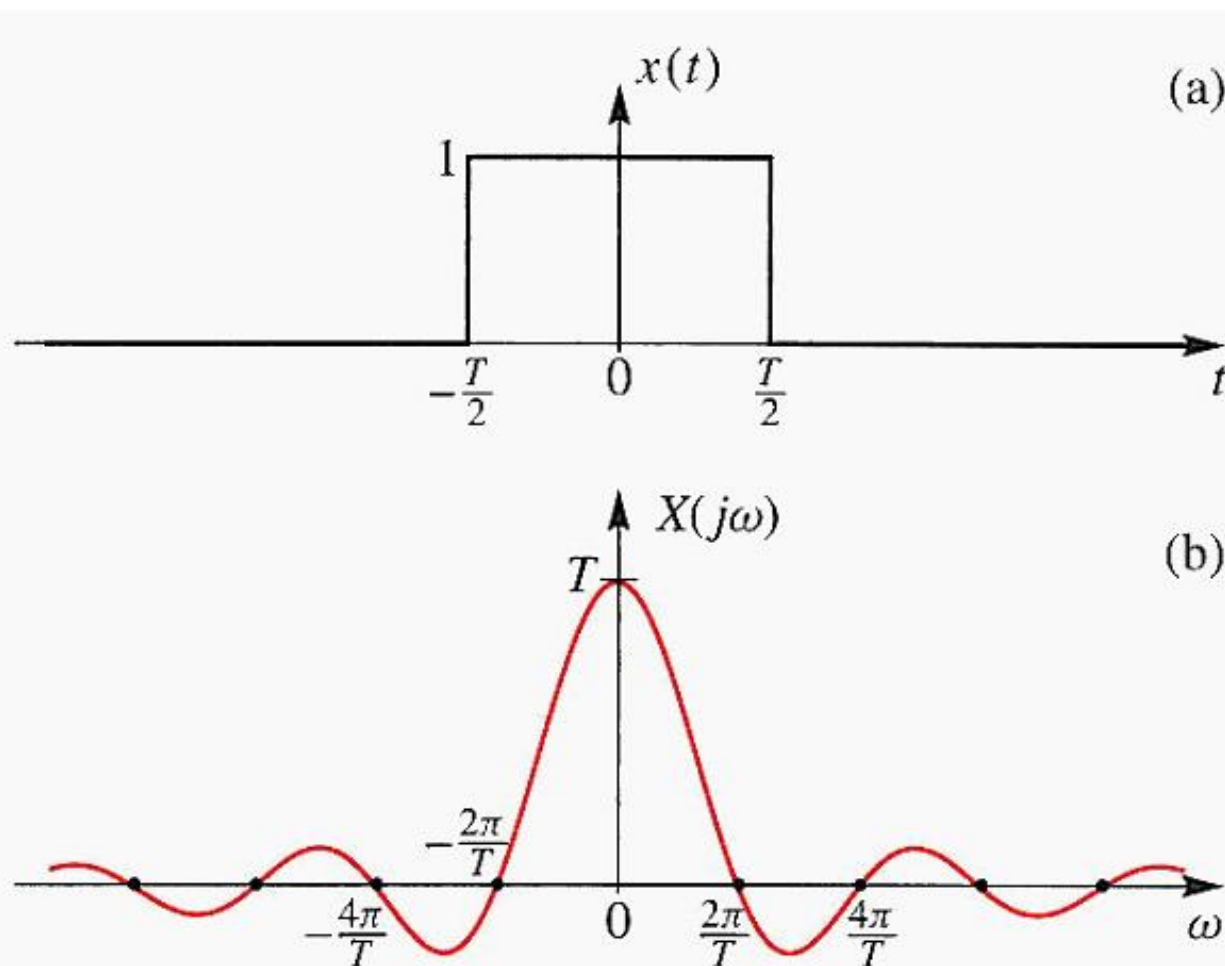
where the **unit-step function** is defined as

$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

- Hence, we have the Fourier transform pair:

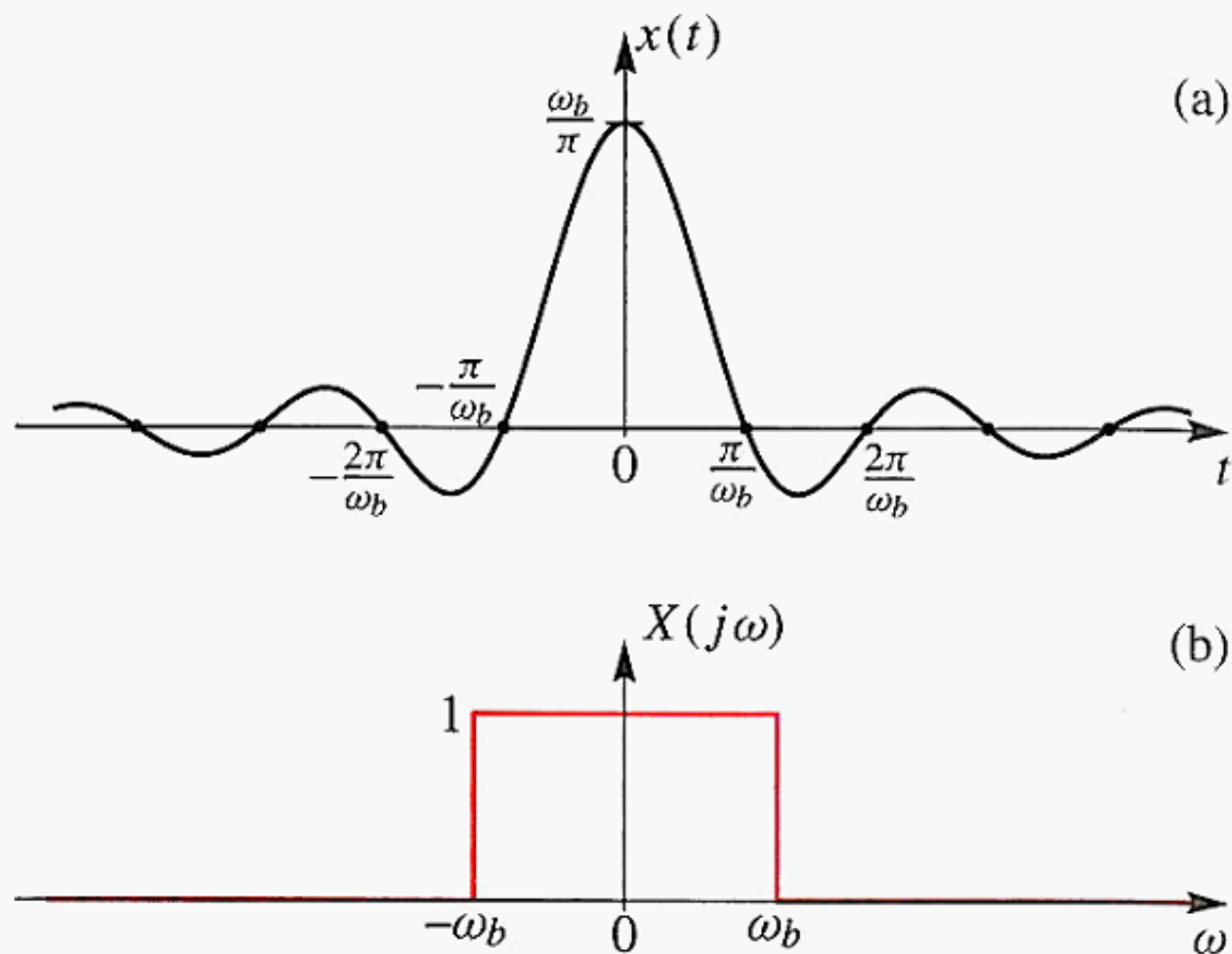
<i>Time-Domain</i>		<i>Frequency-Domain</i>
$\left[ u\left(t + \frac{1}{2}T\right) - u\left(t - \frac{1}{2}T\right) \right]$	$\xleftrightarrow{\mathcal{F}}$	$\frac{\sin(\omega T/2)}{\omega/2}$

A real-valued  
function in  
frequency domain  
(sinc function)<sup>22</sup>



**Figure 11-4:** Fourier transform of a rectangular pulse. (a) Time function  $x(t) = u(t + \frac{1}{2}T) - u(t - \frac{1}{2}T)$ , and (b) Corresponding Fourier transform  $X(j\omega)$  is a sinc function.

# Time and Frequency domains are dual



**Figure 11-5:** Fourier transform of sinc function: (a) Time function  $x(t) = \sin(\omega_b t)/(\pi t)$ , and (b) corresponding Fourier transform  $X(j\omega) = u(\omega + \omega_b) - u(\omega - \omega_b)$ .



# Fourier transform of right-sided real-exponential signal

$a > 0$

<i>Time-Domain</i>		<i>Frequency-Domain</i>
$e^{-at} \underbrace{u(t)}_{\text{Unit-step function}}$	$\xleftrightarrow{\mathcal{F}}$	$\frac{1}{a + j\omega}$

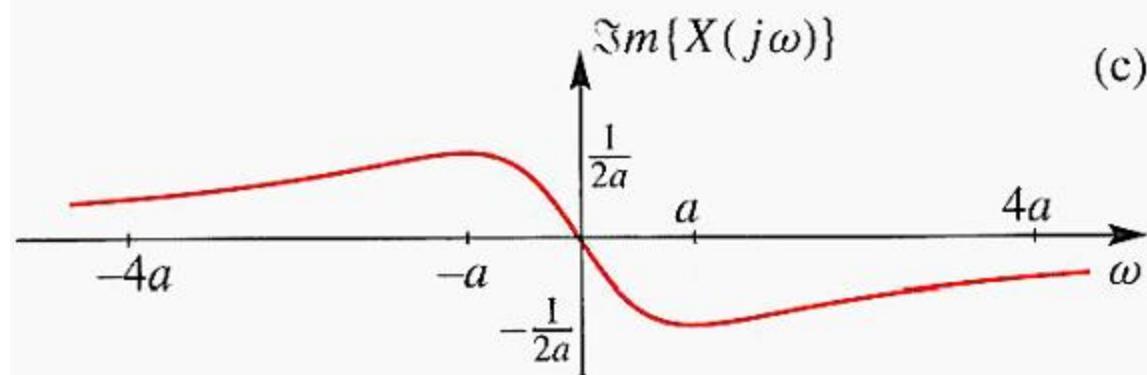
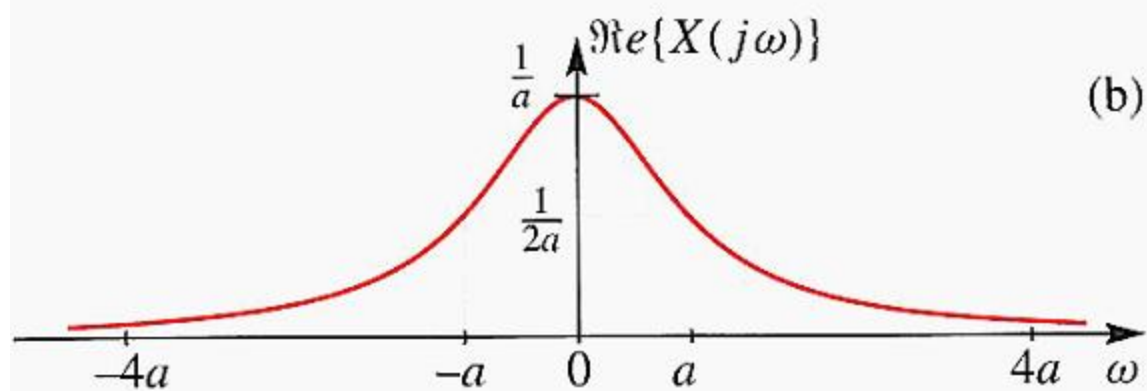
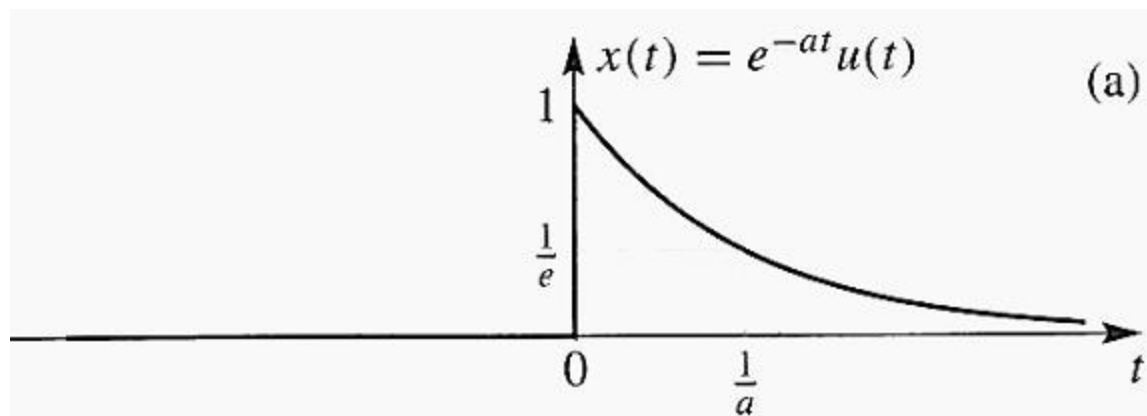
A complex function in frequency domain

- Since

$$\begin{aligned} X(j\omega) &= \frac{1}{a + j\omega} = \frac{1}{a + j\omega} \left( \frac{a - j\omega}{a - j\omega} \right) \\ &= \frac{a}{a^2 + \omega^2} + \frac{-j\omega}{a^2 + \omega^2} \end{aligned}$$

- The real and imaginary parts are

$$\begin{aligned} \Re\{X(j\omega)\} &= \frac{a}{a^2 + \omega^2} \\ \Im\{X(j\omega)\} &= -\frac{\omega}{a^2 + \omega^2} \end{aligned}$$



# Continuous Fourier transform of a Gaussian Function

- Gaussian function:  $e^{-t^2/(2\sigma^2)}$
- The CFT of a Gaussian function is also a Gaussian function (i.e., if time domain is Gaussian, the frequency domain is also Gaussian:

$$e^{-t^2/(2\sigma^2)} \xleftrightarrow{\text{CFT Pair}} \sigma\sqrt{2\pi}e^{-\sigma^2\omega^2/2}$$