

# Estimation of Detected ML in Tone Reservation and Evaluate of Iterative Flicking PTS, SLM Based Riemann Matrix for Reducing PAPR in OFDM

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**Abstract**—Separation process of receiving signal is not efficient and destroyed in a fast fading channel of the orthogonally of consecutive Space Time Block Code STBC Orthogonal frequency division multiplexing OFDM. So we take the tone reservation technique to generate a avoidance of signal in time domain reserving from frequency domain and also to remove high peaks to average power ratio. To improve the tone reservation method by introducing space time block code. We find the value of working ability of orthogonal frequency division multiplexing system under the unite of the influence of tone reservation and Space time block code in maximum likelihood ML to act of valuing by using BPSK modulations are presented for variation of BER of the different sub blocks. The Complementary Cumulative Distribution Function CCDF of Riemann matrix based on Selective Mapping Technique SMT and Iterative Flicking of Partial Transmit Sequence PTS for PAPR rapidly reducing from exciting system in this technique and algorithm used to stimulate the result by using MATLAB to reducing the Peak to Average Power Ratio by increasing SNR by BPSK modulators on OFDM system.

**Key words**—STBC-Space Time Block Codes; OFDM-Orthogonal Frequency Division Multiplexing; Tone Reservation; ML-Maximum-Likelihood; SNR-Signal to Noise Ratio; SMT-Selective Mapping Technique; PTS-Partial Transmit Sequence.

## I. INTRODUCTION

At present Orthogonal Frequency Division Multiplexing which is OFDM has been applied widely in wireless communication systems due to its high data rate otherwise higher bandwidth transmission ability with high bandwidth efficiency and its toughness to multi-path fading and delay. It convert a signal which are from higher data rate signal to lower data rate signal are transducing over parallel and serial. In mobile communication system launched fourth generation (4G) based on this OFDM core technique. The high spectral helpfulness in the system is attained by overlapping. However, the main drawbacks of OFDM are its sensitivity against carrier frequency offset between transducer, co-channel interference which destroys the orthogonally between sub carriers and generates Inter Carrier Interference (ICI). The signal amplitude is reduction and rough guide of the ICI are two destructive effects make happen by carrier frequency offset in OFDM systems. Hence wireless applications demands with high data rates or higher bandwidth. Naturally we dealing with ever-unpredictable wireless channel at higher

data rate otherwise higher bandwidth communications is not task. The idea of multipath transmission has surfaced recently to be used for struggling video conferencing, the hostility of wireless channel and providing high data rate communication. OFDM is to divide the obtainable spectrum into many narrow band signals, low data rate carriers otherwise sub carriers. To attain high spectral productivity the frequency response of the sub carriers of the signals are overlapping and orthogonal, hence the label of OFDM each narrow band sub carrier can be modulated signal using various modulation formats where use in BPSK signal, QPSK and QAM signal are commonly used to single carrier systems, OFDM is adaptable modulation system for multiple right of entry systems in that it fundamentally simplifies both TDMA which means Time Division Multiple Access and FDMA which expands of Frequency Division Multiple Access. In accumulation, significant responsiveness has been given to the arrangement of the OFDM transmission technique and Code-Division Multiple Access which is also termed as CDMA in Multi Carrier-CDMA systems and also MC-CDMA.

## II. PAPR SYSTEMS WITH STBC CODES ANALYSIS

Voguish Orthogonal Frequency Division Multiplexing (OFDM) with separate blocking codes are available which are Space Time Block Coding (STBC-OFDM) otherwise the Space Frequency Block Coding which is SFBC-OFDM. It has been scrutinized in the up-to-date wireless communication systems such as IEEE 802.16 due to its high bandwidth efficiency and IEEE 802.11 a/g, such as Digital Video Broadcasting for Handheld (DVB-H) and the ability of overwhelming frequency signal which discriminating fading channel. A conservative OFDM system transforming of the frequency discriminating fading channel into several flat fading channels so that the transmit of orthogonal diversity techniques i.e., SFBC otherwise STBC can be applied. However, In a fast fading channel, the orthogonally of the two consecutive STBC-OFDM symbols will be destroyed. As a result, the performance of Bit Error Rate which is BER the OFDM system will be degraded

dramatically and the Co-Channel Interference which is CCI has been arises. Correspondingly, the SFBC-OFDM suffers the badly behaved similar to STBC when the frequency responses of the head-to-head subcarriers are not unique. In this case, the Adjacent Channel Interference which is ACI come about and the received signals are not perfect which means it has been not in order of separate the channel signal.

### III. EXCITING SYSTEM

#### A. Selection Mapping Technique

In SLM technique has phase sequence can be multiplied from input sequence. Bauml was introduced the technique. IDFT operation is made on each this alternative input data sequence. The set of sufficiently different signals which are the represented same information the signal having lowest PAPR. Original data can be recover from receiver side by multiplexing sequence.

#### B. Partial Transmit Sequences (PTS)

Intended for PAPR reduction using partial transmit sequence a distinctive OFDM system with input data block in  $X$  has been separated into  $Y$  disjoint sub-blocks of clusters. In general, for PTS scheme, the known sub block splitting methods can be classified into three categories: adjacent divider, interleaved divider and pseudo-random divider. These partial sequences are self-sufficiently rotated by phase factors. The objective has been to optimally combine the  $M$  sub blocks to attain the time domain of OFDM signals with the lowest PAPR. Assuming that there are  $G$  phase angles to be permitted, thus has the possibility of  $W$  poles apart values. And so, at hand  $GY$  alternative depictions for an OFDM symbol. The PTS technique pointedly reduces the PAPR, but inappropriately, finding the optimal phase factors has been a highly difficult problem. In order to reduce the complexity, the variety of the phase factors has been restricted to a set of finite number of fundamentals. Introducing the Exhaustive Search Algorithm which is ESA. It has been hired to find the finest phase factor. However, the ESA requires an exhaustive examine over all groupings of the permitted phase factors and the number of sub blocks. It has exponential examine the complexity. To reduce the computational complexity, some shortened research techniques have been proposed such as the Iterative Flicking Algorithm which is IFA. Although the IFA significantly expressively reduces the examine complexity, some gap between PAPR reduction performance. A Cross-Entropy which shows CE established system has been proposed by Jung-CheihChen for attaining the ideal phase factors for the technique of PTS mainly to reduce the PAPR. He has projected a Quantum-Inspired Evolutionary Algorithm which is QEA founded method to find the optimal phase factor for the PTS technique. A.Ghas semietal has planned an ACF which means Auto-Correlation Function to expand a different PTS sub blocking technique using Error Correcting Codes which is ECCs. This technique reduces the number of repeated subcarrier with a sub block and improved PAPR

reduction than pseudo-random otherwise Y-sequence sub blocking.

### IV. PROPOSED SYSTEM

#### A. Maximum Likelihood

To improve the tone reservation method by introducing space time block code. We find the value of employed ability of orthogonal frequency division multiplexing system under influence of tone reservation technique and Space time block code in Maximum Likelihood (ML) to action of valuing by using BPSK modulator are presented for difference of BER variation of the different sub blocks.

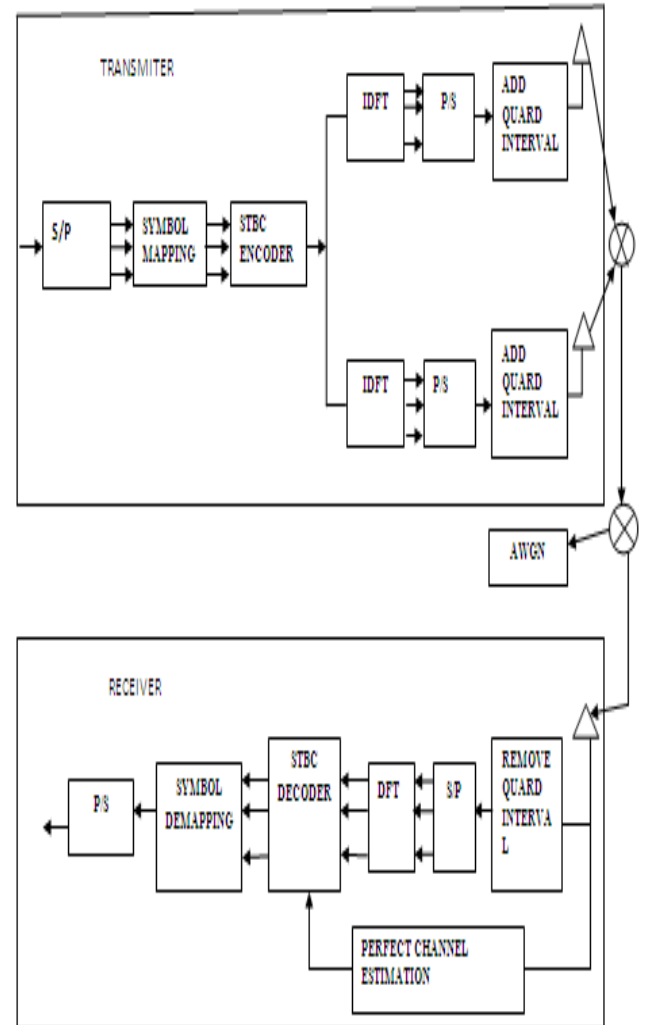


Figure 1. BLOCK DIAGRAM OF STBC-OFDM

In encoder two symbols  $x_1$  and  $x_2$  are encoded with STBC matrix two transmit antennas over two symbol periods. During  $t=T$  is first period, simultaneously transmitted the two symbols of  $x_1$  and  $x_2$  are from the double transmit antennas. Through period at  $t=2T$ , then transmit the signal again. Where  $x_1^*$  transmitted from second antenna and  $-x_2^*$  is transmitted from first antenna. Note that we have assumed the perfect knowledge of the channel responses the receiver. However, now the actual fast fading system of environments, the channel response of two sequential OFDM symbol are not the same. Essentially, the two channel responses for the

time fluctuating are STBC-OFDM system and frequency selective. In such a fast fading situation, the channel matrix  $H$  will be no longer orthogonal that is,

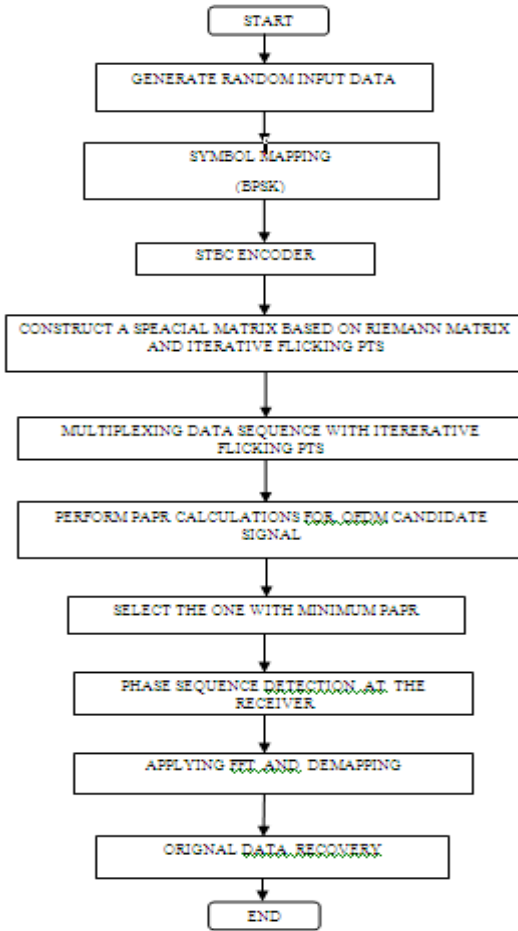


Figure 2. PROPOSED SYSTEM OF SIGNAL FLOWING

#### TIME-VARYING FADING CHANNEL IN A STBC-OFDM

In this paper, superscripts,  $(\cdot)^*$ ,  $W(\cdot)$ ,  $(\cdot)^T$  and  $(\cdot)^H$  denoted the transpose, complex conjugate, hard choice and complex conjugate transpose correspondingly. The system typical of a STBC-OFDM system can be described. To easily the analysis, we consider the simple STBC  $g_2$  encoder with dual transmits antennas and single receive antenna. The input vector of the STBC encoder is represented as

$$x = [X(0), X(1), \dots, X(2N-1)]^T,$$

where  $N$  is the number of the subcarriers.

Let  $x_1 = [X(0), X(1), \dots, X(N-1)]^T$  and  $x_2 = [X(N), X(N+1), \dots, X(2N-1)]^T$ , after the STBC encoder, the generated  $g_2^{STBC}$  coded data symbols are

$$g_2^{STBC} = \begin{bmatrix} x_1 & x_2 \\ -x_2^* & x_1^* \end{bmatrix} \quad (1)$$

Where,  $x_1$  and  $x_2$  are encoded with the matrix of STBC. The channel output matrix is given by

$$Z = HX + Y \quad (2)$$

Where,  $K=1,2,3,\dots,N-1$ ,  $H$  are channel matrix and  $Y$  are Gaussian Noises self-governing of  $x$ .

Through the signal operations such as the Inverse Discrete Fourier Transform (IDFT), further addition and removal of the cyclic prefix and the Discrete Fourier Transform, the received symbol vector  $Z$  as be,

$$Z = \begin{bmatrix} Z_t(k) \\ Z_{2,t+1}^*(k) \end{bmatrix}$$

$$Z = \begin{bmatrix} H_{1,t}(k) & H_{2,t}(k) \\ H_{2,t+1}^*(k) & -H_{1,t+1}^*(k) \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} Y_1(k) \\ Y_{t+1}^*(k) \end{bmatrix} \quad (3)$$

where  $t$  identify as the time constant,  $k$  is the index subcarrier,  $H_{1,t}(k)$ ,  $H_{2,t}(k)$ ,  $H_{2,t+1}^*(k)$  and  $Y_{t+1}^*(k)$  are the DFT System of the channel impulse responses for the first antenna and the second transmit antenna and the channel noises respectively,  $H^H H$

$$H^H H = \begin{bmatrix} H_{1,t}(k) & H_{2,t}(k) \\ H_{2,t+1}^*(k) & -H_{1,t+1}^*(k) \end{bmatrix}^H \begin{bmatrix} H_{1,t}(k) & H_{2,t}(k) \\ H_{2,t+1}^*(k) & -H_{1,t+1}^*(k) \end{bmatrix} = \begin{bmatrix} \phi_K & 0 \\ 0 & \phi_K \end{bmatrix}$$

$$H^H H = \phi_K I_2 \quad (4)$$

$$\text{Where, } \phi_1(k) = |H_{1,t}(k)|^2 + |H_{2,t+1}(k)|^2$$

$$= |H_{1,t+1}(k)|^2 + |H_{2,t+1}(k)|^2$$

$$H^H H = \begin{bmatrix} \phi_1(k) & \varepsilon(k) \\ \varepsilon^*(k) & \phi_2(k) \end{bmatrix} \neq \phi_K I_2 \quad (5)$$

$$\phi_1(k) = |H_{1,t}(k)|^2 + |H_{2,t+1}(k)|^2$$

$$= |H_{1,t+1}(k)|^2 + |H_{2,t+1}(k)|^2$$

The detected output vector  $X$  can be rewritten as

$$\tilde{X} = \begin{bmatrix} \tilde{X}_1(k) \\ \tilde{X}_2(k) \end{bmatrix} = \begin{bmatrix} \phi_1(k)X_1(k) + \varepsilon(k)x_2(k) + Y_t(k) \\ \varepsilon^2(k)X_1(k) + \phi_2(k)X_2(k) + Y_{t+1}(k) \end{bmatrix}$$

$$Y' = \begin{bmatrix} y'_t(k) \\ y'_{t+1}(k) \end{bmatrix} = H^H Y$$

Where,

$\varepsilon(k)x_2(k)$  and  $\varepsilon^2(k)X_1(k)$  are the Co-Channel Interferences (CCI).

Voguish the STBC-OFDM system is performance of BER is used for various detection methods when the speed of the mobile communication system is at 120 km/hr. The method of Alamouti which suffers severe performance degradation due to the Co-Channel Interference and Inter Carrier Interference.

#### B. Riemann Matrix Based New Sequence SLM Technique

SLM was used to PAPR reduction large amount of complexity which is costly and this procedure used to reducing the timing overwhelming in transmission system. Riemann matrix used to problem arises from OFDM which has been PAPR. It can be reducing by this new technique so the transmission becomes faster than other communication system. This method is SLM based new Phase Sequence Transmit technique (PTS). In this approach phase rotation vector is used to rows of normalized Riemann matrix. By deleting of first column and first row of matrix  $D$  where

$$D(i, j) = \begin{cases} i - 1 & \text{if } i \text{ divides } j \\ 1 & \text{otherwise} \end{cases} \quad (6)$$

Using Equation (1), Riemann Matrix (R) of order R=[4\*4] matrix

$$\begin{aligned} R_{11}=1, R_{12}=-1, R_{13}=-1, R_{14}=-1 \\ R_{21}=-1, R_{22}=2, R_{23}=-1, R_{24}=-1 \\ R_{31}=-1, R_{32}=-1, R_{33}=3, R_{34}=-1 \\ R_{41}=-1, R_{42}=-1, R_{43}=-1, R_{44}=4 \end{aligned}$$

Therefore,

$$R = \begin{bmatrix} 1 & -1 & -1 & -1 \\ -1 & 2 & -1 & -1 \\ -1 & -1 & 3 & -1 \\ -1 & -1 & -1 & 4 \end{bmatrix} \quad (7)$$

Normalized Riemann matrix can be expressed as F,

$$\begin{aligned} \text{FMR} &= 1 \\ \text{F} &= 1/\text{MR} \end{aligned} \quad (8)$$

In this approach to make desired of the amplifier and the selection of data block has been reduced the complexity. Threshold value of PAPR is based on RFPA which is termed as Radio Frequency Power Amplifier. With certain constant dipping threshold. The Riemann matrix (R) is of size N\*N the normalized Riemann matrix B.

### C. Iterative Flicking PTS

To find the optimum set of phase factor we need to estimate all the combination of phase factor. We use PTS technique to reduce the computation complexity of number of sub blocks (N) so we bring together the Iterative Flicking algorithm of PTS. In that, each sub blocks keeps only one phase. In the set phase shows minimum PAPR first sub blocks, toward following all sub block is carry on means to decrease the PAPR, performance the method upto end of sub blocks.

Subsequently allocating the data block into M disjoint sub blocks, one take up that  $a(n) = 1$ ;  $n = 1, 2, \dots, N$  for all of sub-blocks and scheming the PAPR of the OFDM signal. Then the phase factor of one deviations the signal of the first sub-block from 1 to -1 where  $a(1) = -1$  and scheming the PAPR signal again and again continuously. If the PAPR of the until that time planned signal is greater than of the current signal, keep  $a(n) = -1$ . Otherwise, invert the previous phase factor,  $a(n) = 1$ . Suppose one chooses  $a(n) = -1$ . Then the first phase factor is absolute, and thus kept fixed for the left over part of the algorithm. Subsequent, we follow the same procedure used in the second sub-block. Since all of the phase factors can be assumed were 1, in the second sub-block, change the order  $a(2) = 1$  to  $a(2) = -1$ , and scheming the PAPR of the OFDM signal. If the PAPR of considered the previous signal is larger than of current signal  $a(2) = -1$  and to revert previous phase factor of  $a(2) = 1$ . further we analysis of performing this same action of technique with the first sub-block is the same as that with the second sub-block. One by one simultaneously continuing this performance of procedure iteratively upto the end of sub-blocks

TABLE I.

PARAMETERS	Value
ModulationType	BPSK
NumberofSubcarriers(N)	256
NumberofSubblocks	2,4,8,16
OversamplingFactor	4
Phase Factor	[1,-1]

### D. Algorithm

In Proposed technique are Iterative Flicking PTS, Riemann matrix of SLM the rows of the normalized Riemann matrix are used as phase order instruction set for PAPR reduction. Here, in proposed system first SLM technique using by means of rows of return to normal Riemann matrix as a phase sequence set is applied to select the best grouping of phase and input data which contributes the lowest PAPR. After selecting the finest grouping of modulated phase sequence set and input data are spread over this combination to the Iterative Flicking PTS and Riemann Matrix for SLM technique by means of columns of normalized Riemann matrix as a phase sequence set.

The set of rules for this technique is described in following steps:

Step 1: Orders of data bits are mapped to gathering points BPSK to produce sequence symbols  $Y_0, Y_1, Y_2, \dots, Y_{N-1}$ .

Step 2: These representation orders are separated into sub blocks of length is N. where N is the number of subcarriers.

Step 3: Each block  $Y = [Y_0, Y_1, Y_2, \dots, Y_{N-1}]$  is multiplied by U different phase sequence vectors  $C^{(u)} = [C_0^{(u)}, C_1^{(u)}, C_2^{(u)}, \dots, C_{N-1}^{(u)}]^T$  where every row of the normalized Riemann matrix C is taken as  $C^{(u)}$ ,  $u = 1, 2, \dots, U$ .

Step 4: A set of U different OFDM data blocks  $Y^{(u)} = [Y_0^{(u)}, Y_1^{(u)}, Y_2^{(u)}, \dots, Y_{N-1}^{(u)}]^T$  are formed, where  $Y_N^{(u)} = Y_n * C_n^{(u)}$   
Where,  $n = 0, 1, \dots, N-1$ ,  $u = 1, 2, \dots, U$ .

Step 5: Transform of  $\{X^{(u)} i^{(u)}\}$  into time domain to get  $Y^{(u)} = \text{IDFT} \{Y\}$

Step 6: Select the one from  $Y^{(u)}$ ,  $u = 1, 2, \dots, U$ , which has the minimum PAPR.

Step 7: Select conforming combination of phase sequence set and input data signal in advance the IDFT operation that's  $X^{(u)}$ .



Step 8: Phase sequence set used to apply  $Y^{(u)}$  as an input data given to Iterative flicking technique with normalized Riemann matrix.

Step 9: Obtain signal with reduced PAPR after applying the Iterative flicking technique with Riemann matrix.

Step 10: Finally the required output can be estimated.

## V. SIMULATION RESULTS

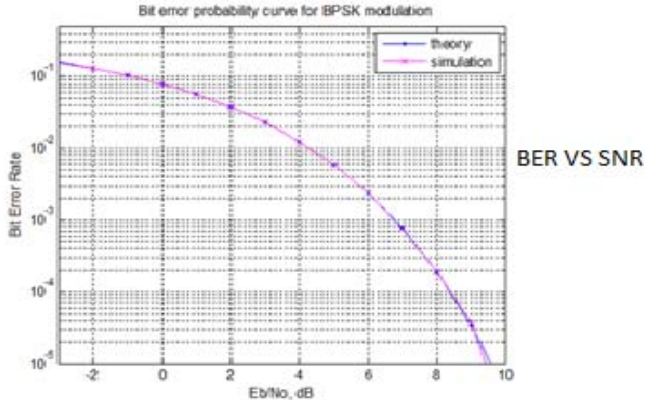


Figure 3. OFDM with Bit Error Rate BER Versus SNR in BPSK Modulation.

Figure 3 shows the presentation of OFDM with Bit Error Rate Versus Signal to Noise Ratio SNR of BPSK modulator used in STBC-OFDM system. In the graph we see the decreasing of BER from  $10^{-12}$  to  $10^{-5}$  by way of increasing SNR from -1db to 9.8db shown on it.

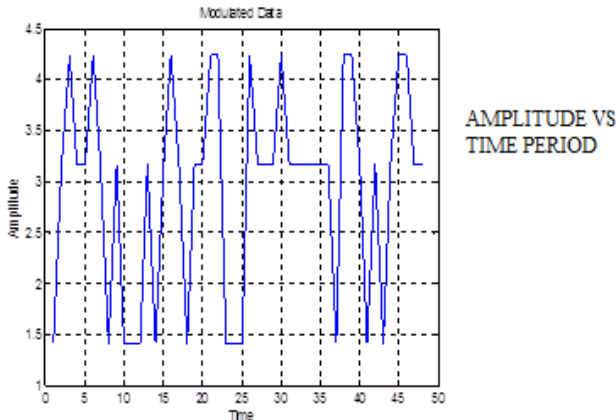


Figure 4. STBC in Modulated Signal for Amplitude Versus Time Period

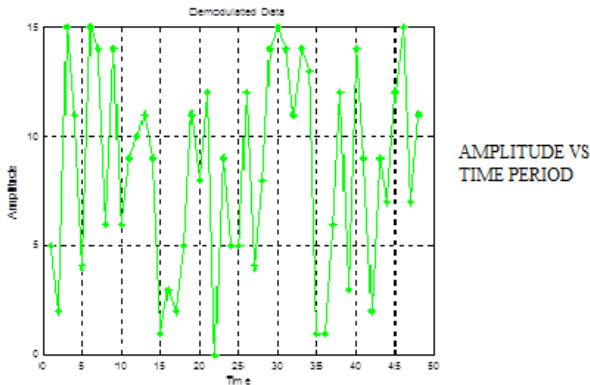


Figure 5. Demodulated Signal for Amplitude with STBC

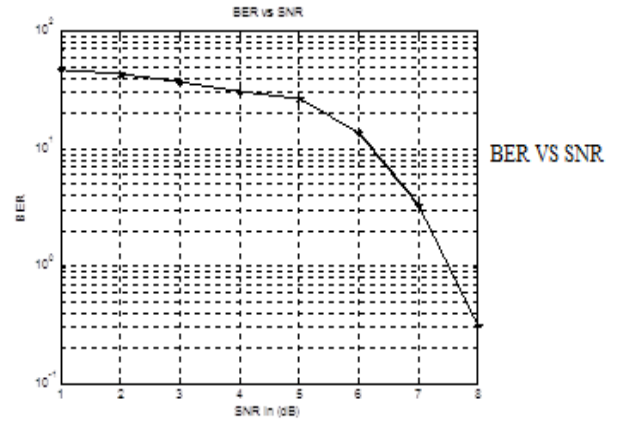


Figure 6. STBC-OFDM Bit Error Rate Vs SNR.

Above figure 6 shows that the presentation of BER Versus SNR using Space Time Block Codes which is STBC-OFDM Orthogonal Frequency Division Multiplexing, In that graph BER decreases from  $10^{-1.3}$  to  $10^{-0.8}$  SNR increases from 1db to 8db is explained.

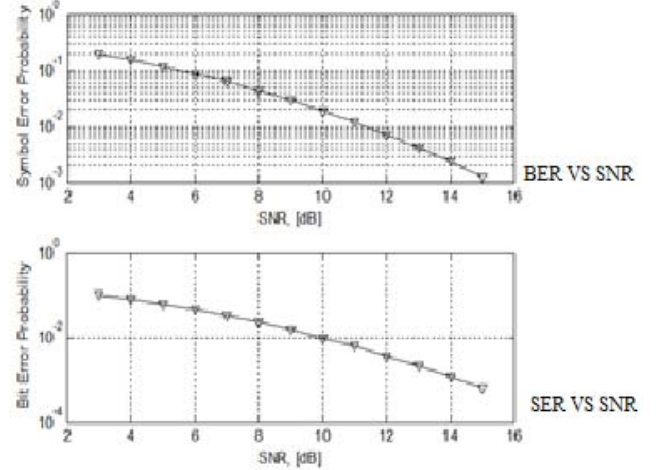


Figure 7. Comparison of STBC-OFDM in BER vs SNR and SER vs SNR.

In that above figure 7 graph shows in MATLAB (Version 7.6.0.424), SER increases from  $10^{-1}$  to  $10^{-3}$  as SNR slightly increases from 3db to 15db is experimental and BER Versus SNR using STBC-OFDM system, Here the BER increases from  $10^{-1}$  to  $10^{-3}$  as SNR increases from 3 db to 15db is experimental.

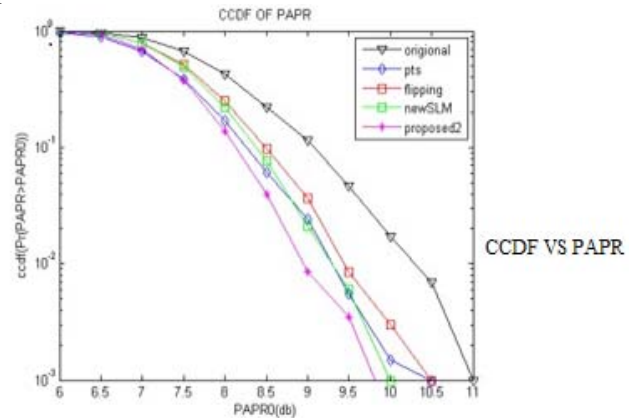


Figure 8. CCDF for comparison of PAPR reduction techniques for M=2 sub-block

In that above figure 8 shows in MATLAB (Version 7.6.0.424), CCDF of PAPR increasing from  $10^0$  to  $10^{-3}$ , these is common for original signal, Partial Transmit Sequence, Flipping, selection mapping technique, proposed system of new SLM Riemann matrix and Iterative Flipping PTS. In  $M=2$  sub block system PAPR is decreasing slightly in original signal of 11db. In the existing systems such as Flipping and pts are used to PAPR reduced from 11db to 10.5db. In this proposed system used to decreasing PAPR from 11db to 9.8db.

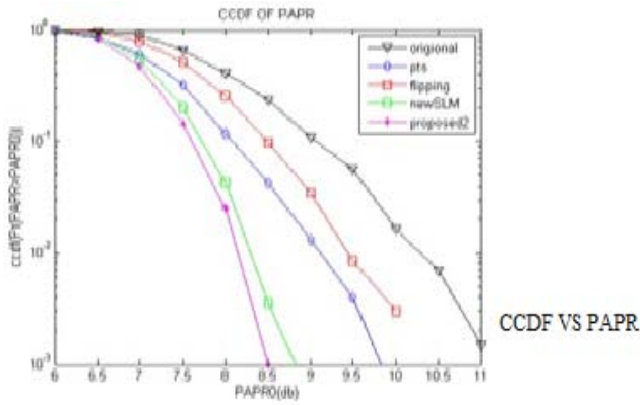


Figure 9. Comparison CCDF of PAPR reduction techniques for  $M=4$  sub-blocks

In that above graph of figure 9 shows in MATLAB (Version 7.6.0.424), the comparison of  $M=2$  system, the CCDF is increasing from  $10^0$  to  $10^{-3}$  and also increasing sub blocks then the  $M=4$ . In Flipping technique is decreasing of -0.2db from  $M=2$  sub block system. In pts technique is decreasing of -0.7db from  $M=2$  sub block system. In the proposed system of new SLM technique which is decreasing from 10db to 8.8db comparing of  $M=2$  sub block system. In of Iterative flicking system comparing of  $M=2$  sub block system, decreasing PAPR from 9.8db to 8.5 db in  $M=4$  sub block system.

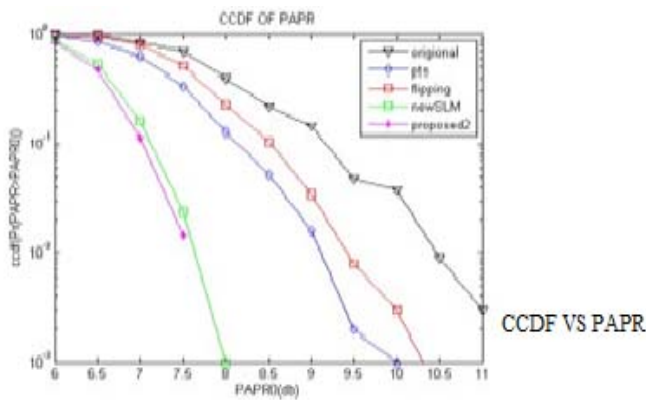


Figure 10. CCDF for comparison of PAPR reduction techniques for  $M=8$  sub-block

In above graph of figure 10 shows in MATLAB version of 7.6.0.424, the CCDF is increasing from the range of  $10^0$  to  $10^{-3}$  differences. PAPR reducing by varies techniques are used in

above graph of  $M=8$  sub block system. Comparing of figure 9, In Flipping technique is remains constant of  $M=4$  sub blocks such as 10.3db. In pts technique is slightly increasing from 0.2db of  $M=4$  sub block system. Then the CCDF of PAPR is slightly increasing from  $10^0$  to  $10^{-0.1}$ . Now use new SLM technique based on Riemann matrix and Iterative Flipping PTS technique shows decreasing of PAPR from increasing of sub block of  $M=8$  so the PAPR has been automatically decreasing from 8.8db to 8db in the above graph.

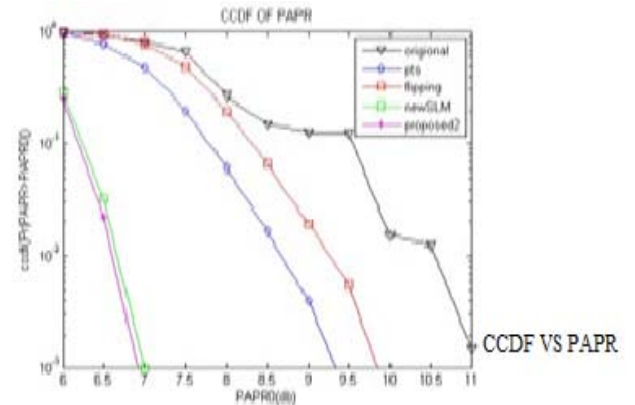


Figure 11. Comparison CCDF of PAPR reduction techniques for  $M=16$  sub-blocks.

In above graph figure 11 explains in MATLAB (Version 7.6.0.424), the deviation of PAPR shows the CCDF of PAPR in unique for PTS, New SLM based on Riemann matrix, Iterative Flipping PTS. Comparison of figure 10 such as  $M=8$  sub block, In  $M=16$  sub block system, the Flipping technique used to reduce PAPR of 10.5db to 9.8db. In pts technique used to reducing of PAPR from  $M=8$  sub block system is -0.7db. Then the graph shows which is CCDF of PAPR is slightly increasing from  $10^{-0.1}$  to  $10^{-0.4}$  in new SLM technique based on Riemann matrix decreasing of PAPR from 8db to 7db and CCDF of PAPR is slightly increasing of  $10^{-0.1}$  to  $10^{-0.5}$  in Iterative Flipping PTS technique decreasing of PAPR from increasing of sub block of  $M=16$  so the PAPR has been decreasing from 8db to 6.9db.

## VI. CONCLUSION

OFDM system with BPSK modulation using Maximum Likelihood of Tone Reservation technique which performance over a number of likely techniques. The analytical equations obtained closely matches the Numerical results. In this paper, we have analysed the bit error rate act of a concatenated low density parity checking of encoded signal which Alamouti-based space-time block code of Orthogonal frequency division multiplexing system under various separate modulations employing the dual transmit antennas and single receive antenna. It has been presented that the proposed system succeeds good error rate performance under BPSK modulation technique. On the basis of results obtained in the current simulation based on

study, In this paper can be established that the OFDM-PAPR Reduction techniques in STBC-OFDM system of Bit Error Rate curve with Maximum Likelihood Estimation based on wireless communication system beneath BPSK modulation is very much effective in suitable identification and recovery of transmitted text message otherwise any other information is noisy and fading situations. This technique gives better performance of OFDM and good BER versus SNR results of the OFDM Systems. The performance of Riemann Matrix based on SLM and Iterative Flicking PTS techniques. However, It is not affected much with the increase the quantity of sub-blocks. From the simulation results shows and used in MATLAB Version 7.6.0.42), It is perfect proposed techniques can complete more PAPR reduction. Moreover, the performance of Proposed Technique such as Iterative Flicking PTS and Riemann matrix based on new SLM technique becomes perfectly we have increasing or using as the number of sub-blocks.

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