

eMCP Specification

32GB eMMC (x8)
+ 32Gb LPDDR3 (x32, 2CS)

Document Title

eMCP
32GB eMMC(x8) Flash / 32Gb (x32, 2CS) LPDDR3

Revision History

Revision No.	History	Draft Date	Remark
0.1	- Initial Draft	Oct. 2015	Preliminary

FEATURES

[eMCP]

- Operation Temperature
 - (-25)°C ~ 85°C
- Package
 - 221-ball FBGA
 - 11.5x13.0mm², 1.2t, 0.5mm pitch
 - Lead & Halogen Free

[eMMC]

• eMMC5.1 compatible

(Backward compatible to eMMC4.5&eMMC5.0)

• Bus mode

- Data bus width : 1bit(default), 4bits, 8bits
- Data transfer rate: up to 400MB/s (HS400)
- MMC I/F Clock frequency : 0~200MHz
- MMC I/F Boot frequency : 0~52MHz

• Operating voltage range

- V_{cc} (NAND) : 2.7V - 3.6V
- V_{ccq} (Controller) : 1.7V - 1.95V / 2.7V ~ 3.3V

• Temperature

- Operation (-25°C ~ +85°C)
- Storage without operation (-40°C ~ +85°C)

• Others

- This product is compliance with the RoHS directive

• Supported features

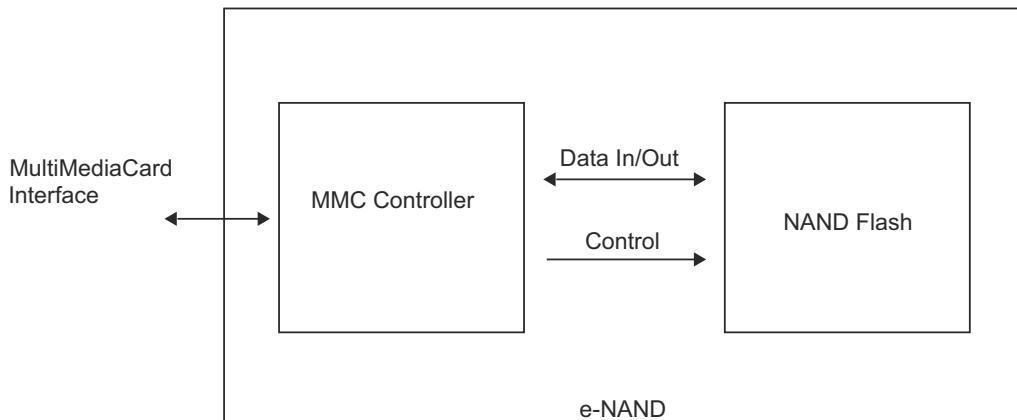
- HS400, HS200
- HPI, BKOPS, BKOP operation control
- Packed CMD, CMD queuing
- Cache, Cache barrier, Cache flushing report
- Partitioning, RPMB, RPMB throughput improve
- Discard, Trim, Erase, Sanitize
- Write protect, Secure write protection
- Lock/Unlock
- PON, Sleep/Awake
- Reliable Write
- Boot feature, Boot partition
- HW/SW Reset
- Field Firmware Update
- Configurable driver strength
- Health(Smart) report
- Production state awareness
- Secure removal type
- Data Strobe pin, Enhanced data strobe

[LPDDR3]

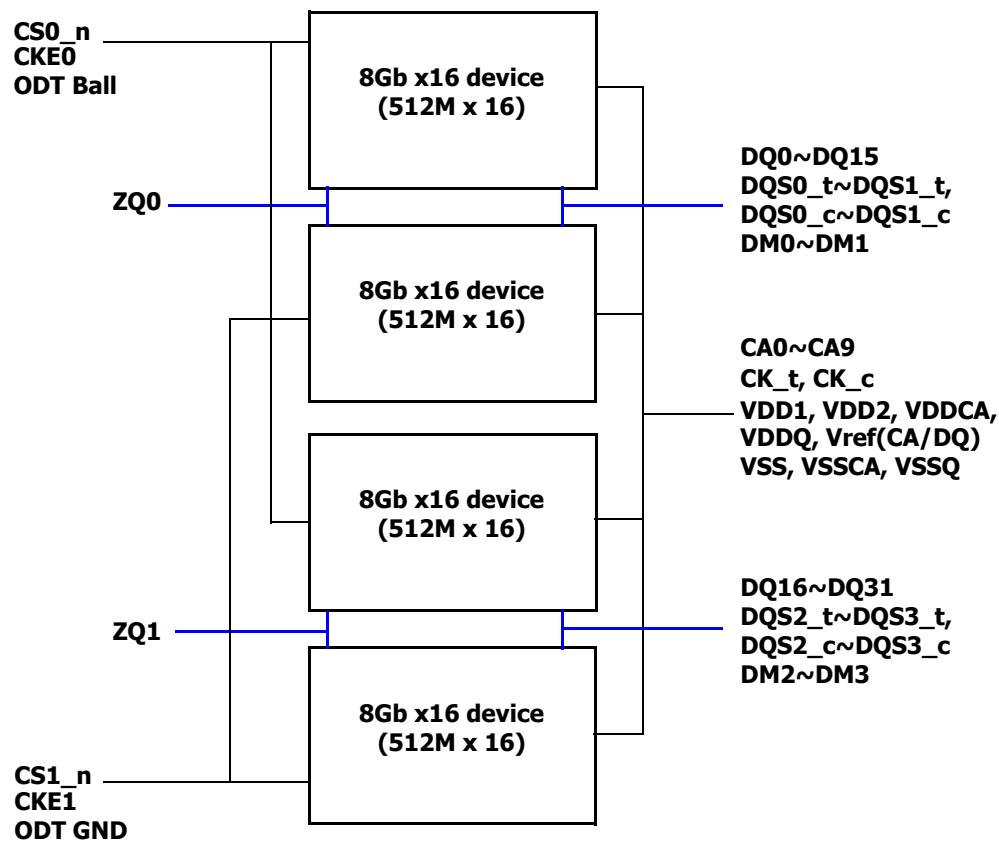
- VDD1 = 1.8V (1.7V to 1.95V)
- VDD2, VDDCA and VDDQ = 1.2V (1.14V to 1.30)
- HSUL_12 interface (High Speed Untermminated Logic 1.2V)
- Double data rate architecture for command, address and data Bus;
- all control and address except CS_n, CKE latched at both rising and falling edge of the clock
- CS_n, CKE latched at rising edge of the clock
- two data accesses per clock cycle
- Differential clock inputs (CK_t, CK_c)
- Bi-directional differential data strobe (DQS_t, DQS_c)
- Source synchronous data transaction aligned to bi-directional differential data strobe (DQS_t, DQS_c)
- Data outputs aligned to the edge of the data strobe (DQS_t, DQS_c) when READ operation
- Data inputs aligned to the center of the data strobe (DQS_t, DQS_c) when WRITE operation
- DM masks write data at the both rising and falling edge of the data strobe
- Programmable RL (Read Latency) and WL (Write Latency)
- Programmable burst length: 8
- Auto refresh and self refresh supported
- All bank auto refresh and per bank auto refresh supported
- Auto TCSR (Temperature Compensated Self Refresh)
- PASR (Partial Array Self Refresh) by Bank Mask and Segment Mask
- DS (Drive Strength)
- ZQ (Calibration)
- ODT (On Die Termination)

Functional Block Diagram

eMMC Block Diagram



DRAM Block Diagram

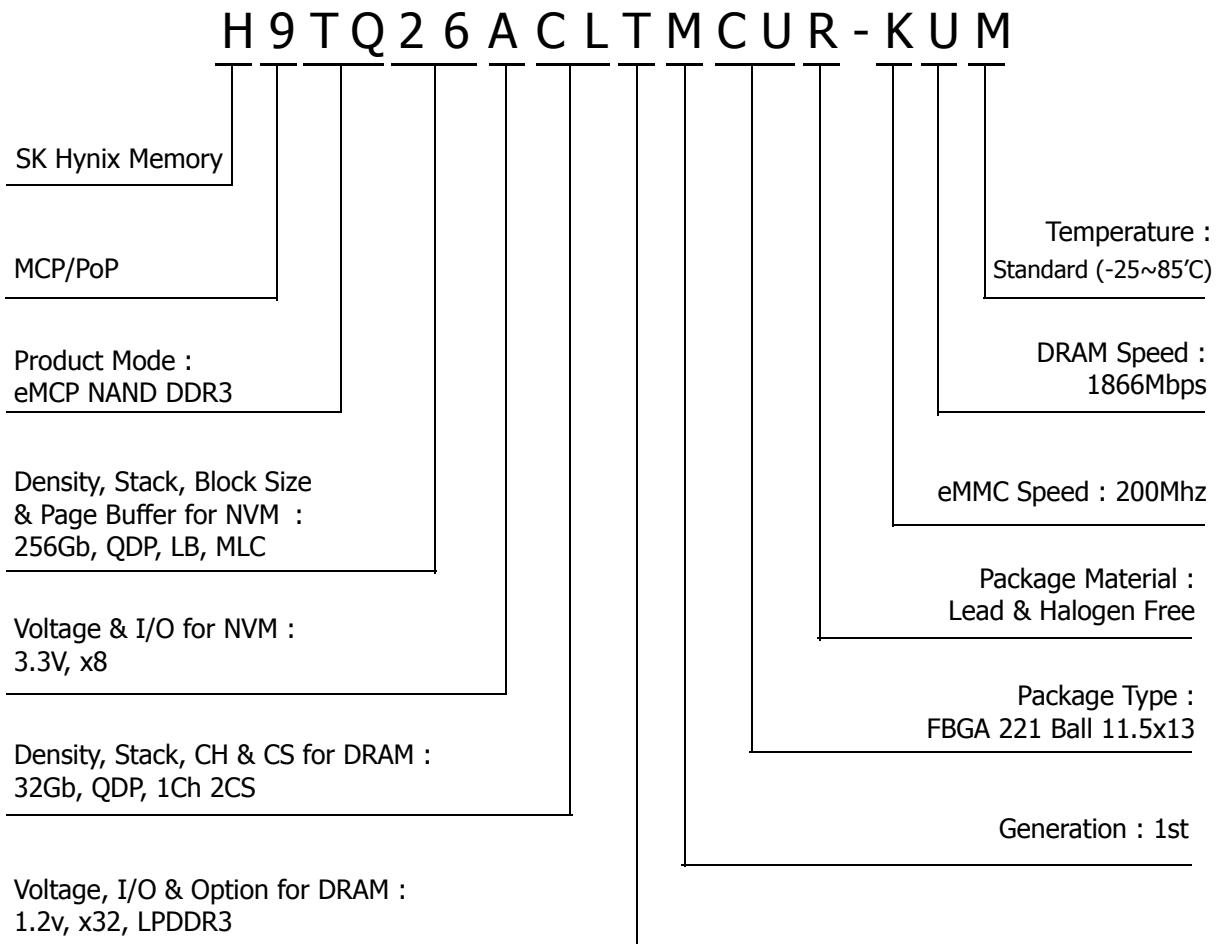


Note

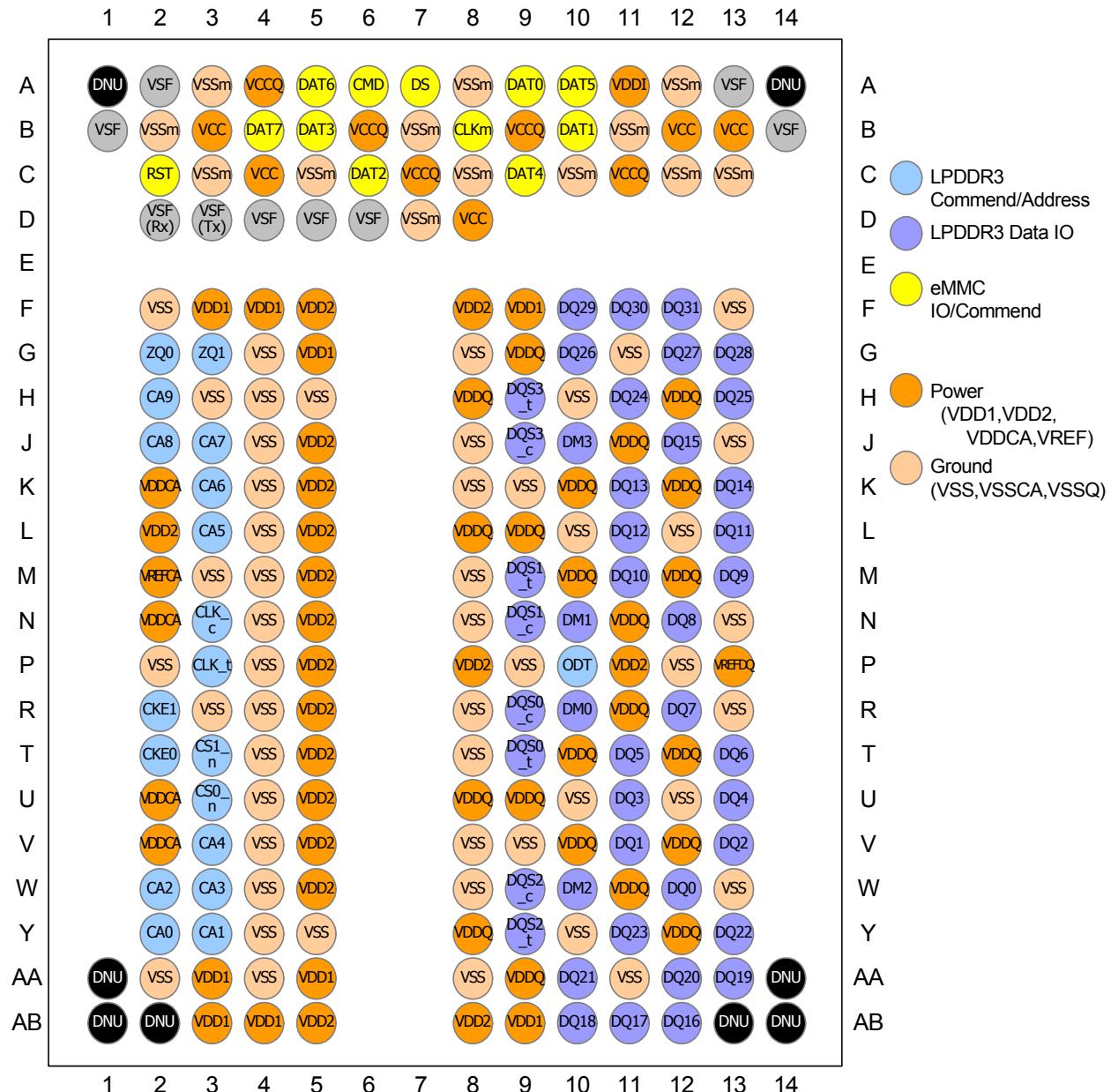
1. Total current consumption is dependent to user operating conditions.
2. VDD, VSS, VREF, CLK, CA are not shown for simplicity. They are connected to all dies. DQS are differential signals.

ORDERING INFORMATION

Part Number	Memory Combination	Operation Voltage	Density	Speed	Package
H9TQ26ACLTMCUR-KUM	eMMC LPDDR3	3.3V 1.8V/1.2/1.2/1.2	32GB (x8) 32Gb (x32)	400MHz DDR3 1866	221Ball FBGA (Lead & Halogen Free)



Ball ASSIGNMENT



Top View

221ball MCP
eMMC + x32 LPDDR3 (1CH)

Note: 1. Vendor specific function (VSF) - this terminal should not have any external electrical connections, but it may have an internal connection. The terminal may be routed to provide accessibility and may be used for general purpose vendor specific operations.

Pin Description

SYMBOL	DESCRIPTION	Type
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< 32GB (x8, MMC) eMMC >

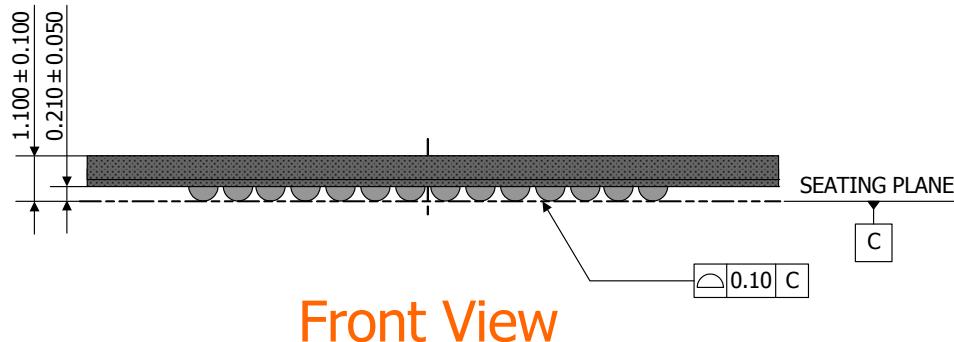
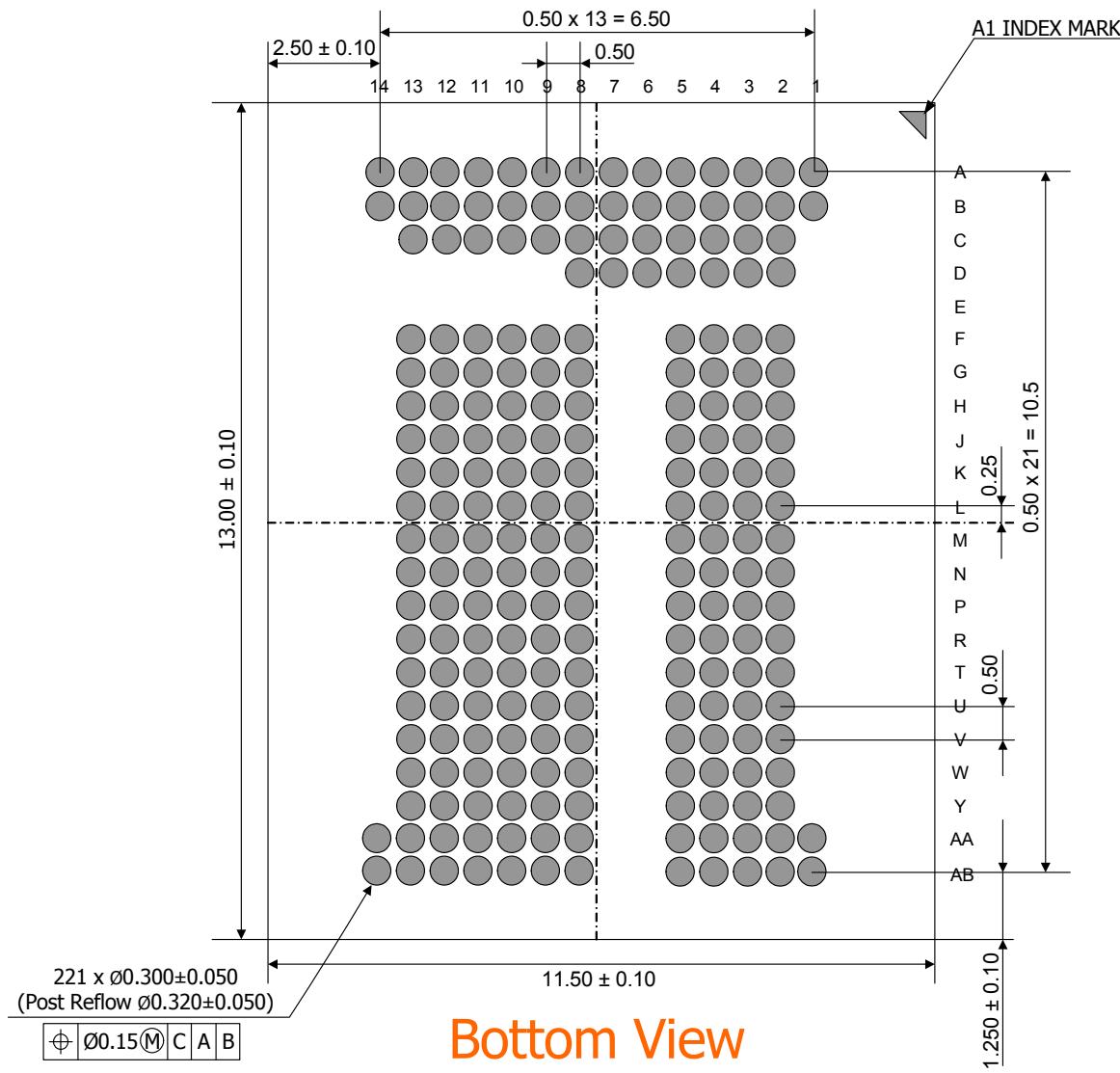
CLK	Clock	Input
CMD	Command	Input/Output
DAT0~DAT7	Data Input/Output	Input/Output
DS	Data Strobe	Output
VCC	Core Power Supply	Power
VCCQ	I/O Power Supply	Power
VSS	Ground	Ground
VDDI	By pass	Power
VSF	Vendor Specific Function	Supply
RST	Reset signal	Input

< 32Gb (x32, 2CS) LPDDR3>

CS_n	Chip Select	Input
CK_c, CK_t	Differential Clocks	Input
CKE	Clock Enable	Input
CA0 ~ CA9	Command / Address	Input
DQ0 ~ DQ31	Data I/O	Input/Output
DM0 ~ DM3	Input Data Mask	Input
DQS0_t ~ DQS3_t	Differential Data Strobe (rising edge)	Input/Output
DQS0_c ~ DQS3_c	Differential Data Strobe (falling edge)	Input/Output
ZQ	Drive Strength Calibration	Input/Output
VDD1	Core Power Supply	Power
VDD2	Core Power Supply	Power
VSS	Ground	Ground
VDDQ	I/O Power Supply	Power
VDDCA	CA Power Supply	Power
VSSCA	CA Ground	Ground
VSSQ	I/O Ground	Ground
VREF	Reference Voltage	Power
ODT	On Die Termination Enable	Input

PACKAGE INFORMATION

221 Ball 0.5mm pitch 11.5mm x 13.0mm FBGA [$t = 1.2\text{mm max}$]



32GB(x8) eMMC Flash

1. Introduction

1.1 General Description

SK hynix eMMC consists of NAND flash and MMC controller.
eMMC has the built-in intelligent controller which manages interface protocols, wear leveling, bad block management, garbage collection, and ECC. eMMC protects the data contents from the host sudden power off failure.
eMMC is compatible with JEDEC standard eMMC5.1 specification.

2. eMMC Characteristics

2.1 Performance

- **Packed CMD ON (Packed depth=8)**

Density	Sequential Write (MB/s)	Sequential Read (MB/s)	Random Write (IOPS)	Random Read (IOPS)
32GB (QDP)	140	280	15K	6.5K

- **Packed CMD OFF**

Density	Sequential Write (MB/s)	Sequential Read (MB/s)	Random Write (IOPS)	Random Read (IOPS)
32GB (QDP)	140	280	8K	5K

- Tool : Device level
- eMMC I/F speed : HS400
- Area : 1GB
- Seq. chunk size : 512KB
- Ran. chunk size : 4KB
- Adapted feature : Cache on
- Status : Sustain
- Not 100% tested

2.2 Power

2.1.1 Active Power Consumption During Operation

Density		I_{cc}	I_{ccq}
32GB (QDP)	Avg (mA)	180	130 (HS200) 140 (HS400)
	Peak (mA)	300	200 (HS200) 230 (HS400)

- Temperature : 25°C
- Average current consumption : over a period of 100ms
- Peak current consumption : over a period of 20us
- Vcc : 3.3V
- Vccq : 1.8V
- Not 100% tested

2.1.2 Low Power Mode (Idle)

Density	Room Temperature (25 °C)		Hot Temperature (85 °C)	
	I _{cc}	I _{ccq}	I _{cc}	I _{ccq}
32GB (QDP)	Typ. 200uA	Typ. 150uA	Max. 400uA	Max. 650uA

- In Standby Power mode, CTRL V_{ccq} & NAND V_{cc} power supply is switched on
- No data transaction period before entering sleep status
- Not 100% tested.

2.1.3 Low Power Mode (CMD5 Sleep)

Density	Room Temperature (25 °C)		Hot Temperature (85 °C)	
	I _{cc}	I _{ccq}	I _{cc}	I _{ccq}
32GB (QDP)	0	Typ. 150uA	0	Max. 650uA

- In Sleep state, triggered by CMD5, NAND V_{cc} power supply is switched off (CTRL V_{ccq} on)
- Temperature : 25 °C
- Not 100% tested.

3. eMMC New features (eMMC5.0 and eMMC5.1)

3.1 eMMC5.0 New features

3.1.1 HS400 mode

eMMC supports HS400 signaling to achieve a bus speed of 400MB/s via a 200MHz DDR clock frequency. HS400 mode supports only 8-bit bus width and the 1.8V V_{ccq}. Due to the speed, the host may need to have an adjustable sampling point to reliably receive the incoming data (Read Data and CRC Response) with DS pin. eMMC supports up to 5 Driver Strength.

Driver type values	Support	Nominal Impedance	Approximated driving capability compared to Type_0	Remark
0	Mandatory	50Ω	x 1	Default Driver Type. Supports up to 200MHz operation.
1	Optional	33Ω	x 1.5	Supports up to 200MHz operation.
2		66Ω	x 0.75	The weakest driver that supports up to 200MHz operation.
3		100Ω	x 0.5	For low noise and low EMI systems. Maximal operating frequency is decided by host design.
4		40Ω	x 1.2	

[Table 1] I/O Driver strength types

Selecting **HS_Timing** depends on Host I/F speed, default is 0, but all of value can be selected by host.

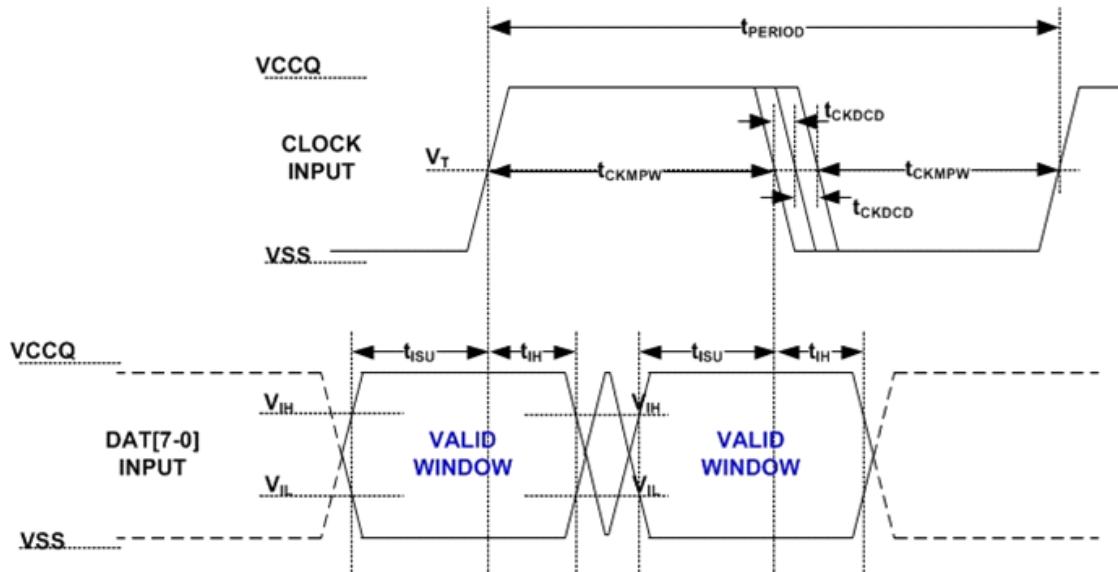
Value	Timing	Supportability for eMMC
0x00	Selecting backward compatibility interface timing	Support
0x01	High speed	Support
0x02	HS200	Support
0x03	HS400	Support

[Table 2] HS_Timing values

3.1.1.1 Bus timing specification in HS400 mode

■ HS400 Device input timing

The CMD input timing for HS400 mode is the same as CMD input timing for HS200 mode.



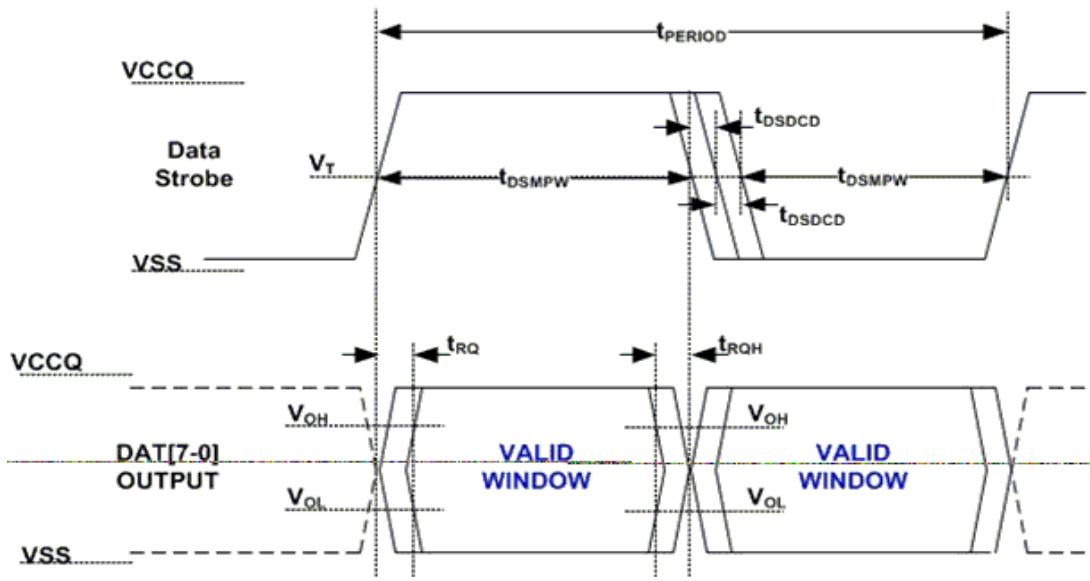
[Figure 1] HS400 Device input timing

Parameter	Symbol	Min	Max	Unit	Remark
Input CLK					
Cycle time data transfer mode	t_{PERIOD}	5			200MHz(Max), between rising edges with respect to V_T
Slew rate	SR	1.125		V/ns	With respect to $V_{\text{IH}}/V_{\text{IL}}$
Duty cycle distortion	t_{CKDCD}	0.0	0.3	ns	Allowable deviation from an ideal 50% duty cycle. With respect to V_T . Includes jitter, phase noise
Minimum pulse width	t_{CKMPW}	2.2		ns	With respect to V_T
Input DAT (referenced to CLK)					
Input set-up time	$t_{\text{ISU}}_{\text{ddr}}$	0.4		ns	$C_{\text{DEVICE}} \leq 6\text{pF}$ With respect to $V_{\text{IH}}/V_{\text{IL}}$
Input hold time	$t_{\text{IH}}_{\text{ddr}}$	0.4		ns	$C_{\text{DEVICE}} \leq 6\text{pF}$ With respect to $V_{\text{IH}}/V_{\text{IL}}$
Slew rate	SR	1.125		V/ns	With respect to $V_{\text{IH}}/V_{\text{IL}}$

[Table 3] HS400 Device input timing

■ HS400 Device output timing

Data strobe is for reading data in HS400 mode. Data strobe is toggled only during data read or CRC status response.



[Figure 2] HS400 Device output timing

Parameter	Symbol	Min	Max	Unit	Remark
Input CLK					
Cycle time data transfer mode	t_{PERIOD}	5			200MHz(Max), between rising edges with respect to V_T
Slew rate	SR	1.125		V/ns	With respect to V_{OH}/V_{OL} and HS400 reference load
Duty cycle distortion	t_{DSDCD}	0.0	0.2	ns	Allowable deviation from the input CLK duty cycle distortion(t_{CKDCD}) With respect to V_T Includes jitter, phase noise
Minimum pulse width	t_{DSMPW}	2.0		ns	With respect to V_T
Read pre-amble	t_{RPRE}	0.4	5 (One Clock Cycle)	t_{PERIOD}	Max value is specified by manufacturer. value up to infinite is valid.
Read post-amble	t_{RPST}	0.4	2.5 (Half Clock Cycle)	t_{PERIOD}	Max value is specified by manufacturer. value up to infinite is valid.
Output DAT (referenced to Data strobe)					
Output skew	t_{RQ}		0.4	ns	With respect to V_{OH}/V_{OL} and HS400 reference load
Output hold skew	t_{RQH}		0.4	ns	With respect to V_{OH}/V_{OL} and HS400 reference load
Slew rate	SR	1.125		V/ns	With respect to V_{OH}/V_{OL} and HS400 reference load

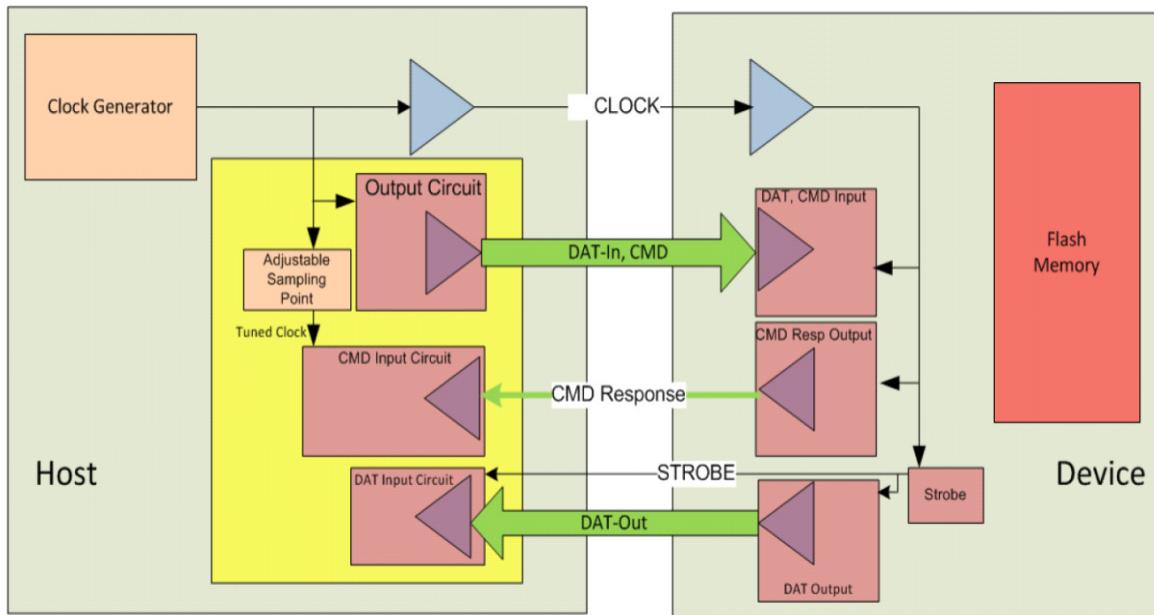
[Table 4] HS400 Device output timing

Parameter	Symbol	Min	Type	Max	Unit	Remark
Pull-up resistance for CMD	R_{CMD}	4.7		100	Kohm	
Pull-up resistance for DAT0-7	R_{DAT}	10		100	Kohm	
Pull-down resistance for Data strobe	R_{DS}	10		100	Kohm	
Internal pull up resistance DAT1-DAT7	R_{int}	10		150	Kohm	
Bus signal line capacitance	C_L			13	pF	
Single Device capacitance	C_{Device}			6	pF	

[Table 5] HS400 Device input timing

■ Data Strobe for HS400

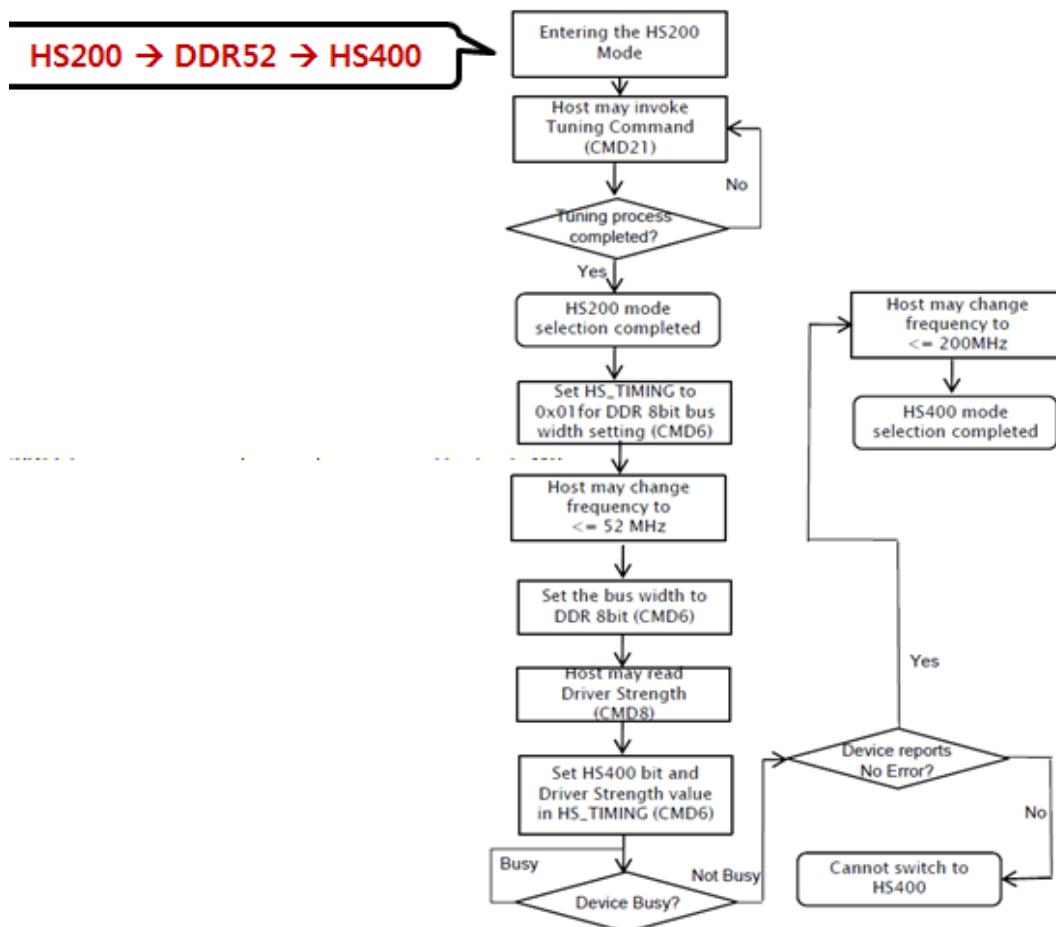
Data strobe is Return Clock signal used in HS400 mode. This signal is generated by the device and used for data output and CRC status response output in HS400 mode. The frequency of this signal follows the frequency of CLK. For data output each cycle of this signal directs two bits transfer(2x) on the data - one bit for positive edge and the other bit for negative edge. For CRC status response output, the CRC status is latched on the positive edge only, and don't care on the negative edge. Data strobe signal is toggled only for Data out and CRC response (Align CMD response as well as CRC response to the DS in eMMC5.1)



[Figure 3] HS400 Host and Device block diagram

3.1.1.2 HS400 Mode selection

Following JEDEC standard for eMMC5.0, changing bus mode directly from HS200 to HS400 is not allowed. It has a rule for changing bus width from SDR mode to DDR mode that HS_TIMING must be set to "0x01"(HS mode : 52MHz) before setting BUS_WIDTH for DDR operation. We recommend the HS400 bus mode selection sequence as following.
(eMMC5.1 has basically same flow, but 'enhanced strobe feature' is added. Please refer to 4.2.6 Enhanced strobe)



[Figure 4] HS400 Bus mode selection sequence

■ EXT_CSD register for Data strobe

- Enhanced Strobe field in BUS_WIDTH [183]

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Enhanced Strobe		Reserved				Bus Mode Selection	

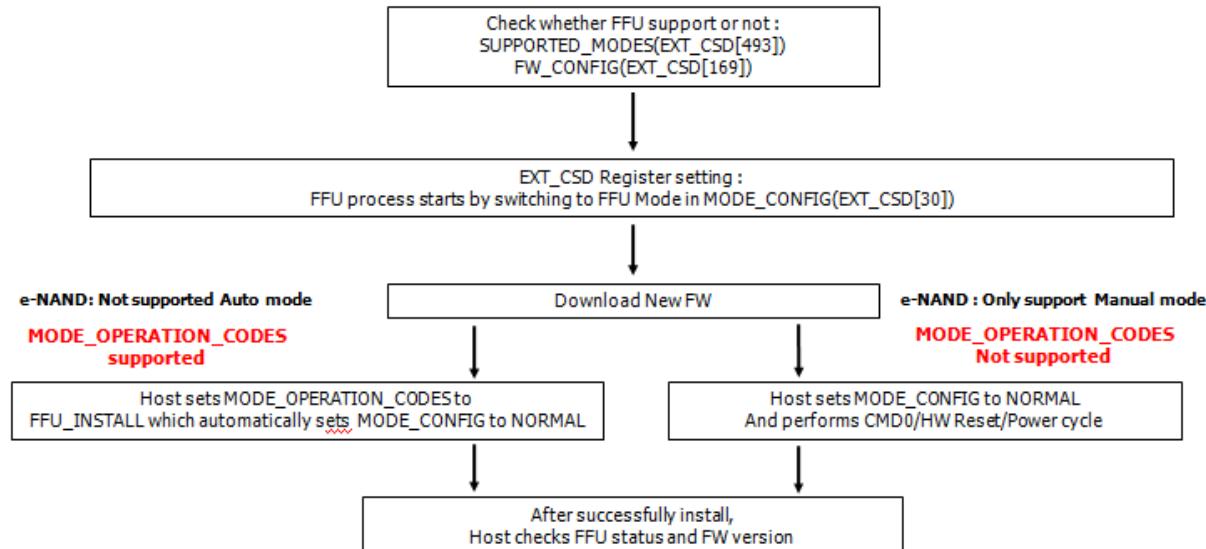
BIT[7] : 0: Strobe is provided only during data out and CTC response [Default]

1: Strobe is provided during data out, CRC response and CMD response

The support of STROBE_ENHANCED mode is optional for devices. STROBE_SUPPORT[184] register of EXT_CSD indicates whether a device supports that mode.

3.1.2 Field firmware update (FFU)

To download a new firmware, the eMMC requires instruction sequence following JEDEC standard. SK hynix eMMC only supports Manual mode (MODE_OPERATION_CODES is not supported). For more details, see as the following chart and register table given below.



[Figure 5] FFU flow chart

3.1.2.1 Field F/W update flow - CMD sequence

Operation	CMD	Remark
Set bus width (1bit or 4bit)		Bus width should be 1bit or 4bit
Set block length 512B	CMD16, arg : 0x00000200	
Enter FFU mode	CMD6, arg : 0x031E0100	
Send FW to device(Download)	CMD25, arg : 0xFFFFFFF0	Sending CMD25 is followed by sending FW data
CMD12 : Stop	CMD12, arg : 0x00000000	
CMD6 : Exit FFU mode	CMD6, arg : 0x031E0000	
CMD0/HW Reset/Power cycle		
Re-Init to trans state	CMD0, CMD1 ...	
Check if FFU is succeeded	CMD8, arg : 0x00000000	Check EXT_CSD[26] : FFU_SUCCESS If FFU_SUCCESS is 0, FFU is succeeded, otherwise FFU is failed.

3.1.2.2 EXT_CSD Register for FFU

■ SUPPORTED_MODE[493] (Read Only)

BIT[0] : '0' FFU is not supported by the device. '1' FFU is supported by the device.

BIT[1] : '0' Vendor specific mode (VSM) is not supported by the device. '1' Vendor specific mode is supported by the device.

Bit	Field	Supportability
Bit[7:2]	Reserved	-
Bit[1]	VSM	Not support
Bit[0]	FFU	Supported

■ FFU_FEATURE[492] (Read Only)

BIT[0] : '0' Device does not support MODE_OPERATION_CODES field (Manual mode)

'1' Device supports MODE_OPERATION_CODES field (Auto mode)

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	SUPPORTED_MODE_OPERATION_CODES	Not supported

■ FF_ARG[490-487] (Read Only)

Using this field the device reports to the host which value the host should set as an argument for read and write commands is FFU mode.

■ FF_CONFIG[169] (R/W)

BIT[0] : Update disable

0x0 : FW updates enabled. / 0x01 : FW update disabled permanently

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	Update disable	FW updates enabled (0x0)

■ FFU_STATUS[26] (R/W/E_P)

Using this field the device reports to the host the state of HHU process.

Value	Description
0x13 ~ 0xFF	Reserved
0x12	Error in downloading Firmware
0x11	Firmware install error
0x10	General error
0x01 ~ 0x0F	Reserved
0x00	Success

■ OPERATION_CODES_TIMEOUT[491](Read Only)

Maximum timeout for the SWITCH command when setting a value to the MODE_OPERATION_CODES field

The register is set to '0', because the eMMC doesn't support MODE_OPERATION_CODES.

Value	Description	Timeout value
0x01 ~ 0x17	MODE_OPERATION_CODES_TIMEOUT = 100us × 2 _{OPERATION_CODES_TIMEOUT}	0 (Not defined)
0x18 ~ 0xFF	Reserved	-

■ MODE_OPERATION_CODES[29] (W/E_P)

The host sets the operation to be performed at the selected modes, in case MODE_CONFIGS is set to FFU_MODE, MODE_OPERATION_CODES could have the following values :

Bit	Description
0x01	FFU_INSTALL
0x02	FFU_ABOUT
0x00, others	Reserved

3.1.3 Health(Smart) report

Using this feature is for monitoring device status and preventing the error and failure in advance.

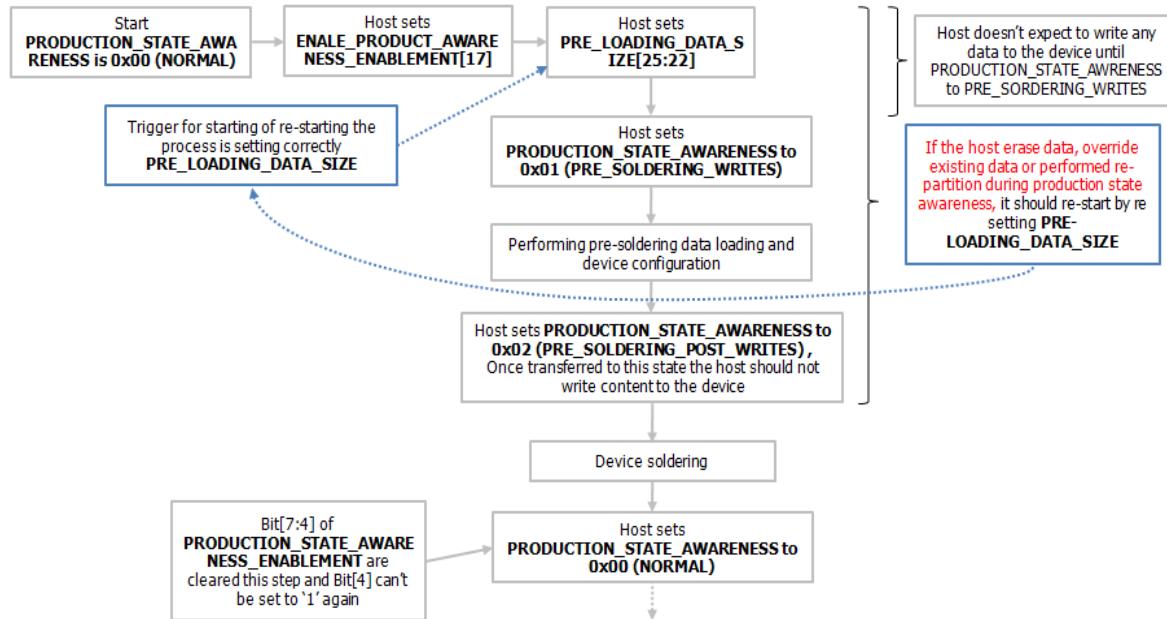
Host can check device information with EXT_CSD as the register table given below.

Field	CSD slice	Description
VENDOR_PROPRIETARY_HEALTH_REPORT	[301:270]	Reserved for vendor proprietary health report. (NONE)
DEVICE_LIFE_TIME_EST_TYPE_A/B	[268:269]	Current average P/E cycle of memory of Type A(SLC) / Type B(MLC) relative to its maximum estimated capability
PRE_EOL_INFO	[267]	Consumed reserved blocks to notify before reaching the EOL (End of life) status
OPTIMAL_TRIM/WRITE_READ_SIZE	[264:266]	Minimum optimal (for the device) Erase / Write / Read unit size for the different partitions
DEVICE_VERSION	[263:262]	Device version
FIRMWARE_VERSION	[261:254]	Device FW version

[Table 6] Using EXT_CSD for health report (Read only)

3.1.4 Production state awareness

This new feature is added for eMMC5.0 JEDEC Spec. to prevent the data break during device soldering. For this feature implementation, eMMC supports only manual mode and PRODUCT_STATE_AWARENESS_TIMEOUT is 0x17(maximum). For more detail, see as the flow chart and register table given below.



[Figure 6] Production State Awareness manual mode flowchart

■ PRODUCTION_STATE_AWARENESS_TIMEOUT[218] (Read Only)

This field indicates maximum timeout for the SWITCH command when setting a value to the PRODUCTION_STATE_AWARENESS[133]field

Value	Description	Timeout value
0x01 ~ 0x17	Production State Timeout = 100us x 2 _{PRODUCTION_STATE_AWARENESS_TIMEOUT}	0x17 (838.86s)
0x18 ~ 0xFF	Reserved	-

■ PRODUCTION_STATE_AWARENESS[133](R/W/E)

eMMC doesn't support 0x03 state.

Value	Device State	Description
0x00	NORMAL (Field)	Regular operation
0x01	PRE_SOLDERING_WRITES	-
0x02	PRE_SOLDERING_POST_WRITES	Once transferred to this state the host should not write content to the device
0x03	AUTO_PRE_SOLDERING	Not supported
0x04 ~ 0x0F	Reserved	-
0x10 ~ 0x1F	Reserved for Vendor Proprietary Usage	-

■ PRODUCTION_STATE_ENABLEMENT[17]

eMMC only supports manual mode for PRODUCTION_STATE_AWARENESS

Enablement(R/W/F)				Capabilities(R)			
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
Reserved	Mode	Production State Awareness enable		Reserved		Auto mode Supported	Manual mode Supported
Cleared when PRODUCTION_STATE_AWARENESS is charged to Normal (either automatically or by setting PRODUCTION_STATE_AWARENESS to Normal)				This bit could be set to '1' only once			

3.1.5 Sleep notification

Host may use a power off notification when it intends to turn-off V_{CC}. After moving the device to sleep state. Some features are added to clarify the spec for entering sleep mode when power off notification is enabled.

■ Add the SLEEP_NOTIFICATION on the interruptible Command List

CMD	Description	Is interruptible?
CMD6	SWITCH, Byte POWER_OFF_NOTIFICATION, Value POWER_OFF_LONG or SLEEP_NOTIFICATION	Yes

■ SLEEP_NOTIFICATION_TIME[216](Read Only)

Maximum timeout for the SWITCH command when notifying the device that it is about to move to sleep state by writing SLEEP_NOTIFICATION to POWER_OFF_NOTIFICATION[34]byte. (unit : 10us)

Value	Description	Timeout value
0x01 ~ 0x17	Sleep Notification Timeout = 10us x 2 _{SLEEP_NOTIFICATION_TIME}	0xC (40.96ms)
0x18 ~ 0xFF	Reserved	-

■ POWER_OFF_NOTIFICATION[34]

Add 0x04h for the SLEEP_NOTIFICATION as a valid value

Value	Field	Description
:	:	:
0x03	POWER_OFF_LONG	Host is going to power off the device. The device shall respond within POWER_OFF_LONG_TIME
0x04	SLEEP_NOTIFICATION	Host is going to put device in sleep mode. The device shall respond within SLEEP_NOTIFICATION_TIME

3.1.6 Secure removal type

This feature is used for how information is removed from the physical memory during a purge operation.

■ Secure Removal Type[16]

Among four options for secure removal type, eMMC supports 0x3, 0x1 and 0x0 (0x2 option is not supported)
eMMC recommends using a vendor defined removal type(type 3). If host want to erase the device physically using removal type0.

Secure erase & Secure trim time is longer than using removal type0

BIT	Description of Secure Removal Type	Description		Supportability
BIT[5:4]	Configure Secure Removal Type (R/W)	0x3	Information removed using a vendor defined	Support
		0x2	Information removed by an overwriting the addressed locations with a character, its complement, then a random character	-
		0x1	Information removed by an overwriting the addressed locations with a character followed by an erase	-
		0x0	Information removed by an erase of the physical memory	-
BIT[3:0]	Supported Secure Removal Type (R)	BIT[3]	Information removed using a vendor defined	Support
		BIT[2]	Information removed by an overwriting the addressed locations with a character, its complement, then a random character	Not support
		BIT[1]	Information removed by an overwriting the addressed locations with a character followed by an erase	Support
		BIT[0]	Information removed by an erase of the physical memory	Support

3.2 eMMC5.1 New features

3.2.1 Command queuing

eMMC manages an internal task queue to which the host can queue data transfer tasks to efficient operate.

3.2.1.1 CMD list for Command queuing

Index	Abbreviation	Argument	Remark
CMD44	Queued Task parameter	[31] Reliable Write Request [30] Data Direction (Read=1, Write=0) [29] Tag Request [28:25] context ID [24]Forced Programming [23] Priority (simple =0, high=1) [20:16] Task ID(31~0) [15:0] # of BLK	Encodes parameter which are necessary for queuing the task and executing the transfer
CMD45	Queued Task Address	[31:0] Block address for the transaction	The host instructs the device to queue a data transfer task
CMD46	Execute Read Task	[20:16] Task ID of the requested task which is must marked as "ready for execution" in the Queue Status Register	In order to execute a data read task (CMD46) or write task(CMD47) which is already queued
CMD47	Execute Write Task		
CMD48	CMDQ_TASK_MGMT	[31:21] reserved [20:16]: TaskID [15:4]: reserved [3:0] TM op-code	Device shall discard a specific task or entire queue (all tasks in the queue) [20:16] when TM op-code = 2h these bits represent TaskID. When TM op-code=1h these bits are reserved.

Index	Abbreviation	Argument	Remark
CMD13	SEND_STATUS	[31:16] RCA [15] SQS [14] Stuff bits [0] HPI	In case SQS bit = 1: indicate that this is CMD13 of CMD Queue. In response device shall send the QSR. In this case HPI bit must be set to '0'.

3.2.1.2 EXT_CSD Register for Command queuing

■ CMDQ_SUPPORT[308](Read only)

This field indicates whether command queuing is supported by the device.

BIT[0]: 0: Command queuing is not supported

1: Command queuing is supported

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	CMDQ Support	Supported (0x1)

■ CMDQ_DEPTH[307](Read only)

This field is used to calculate the depth of the queue supported by the device. The maximum depth allowed by the standard is 32.

The range of allowed Task IDs is 0 through N.

BIT[4:0]: N, a parameter used to calculate the Queue Depth of task queue in the device.

Queue Depth = N+1.

Bit	Field	CMD queue depth value
Bit[7:5]	Reserved	-
Bit[4:0]	N	N=31 (0x1F)

■ CMDQ_MODE_EN[15](R/W)

This field is used by the host enable command queuing mechanism if supported by the device.

BIT[0]: 0: Command queuing is disabled. Host should clear the queue empty using CMD48 prior disabling the queue.

1: Command queuing is enabled

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	CMDQ enable	-

To maintain backward compatibility with hosts, which do not support command queuing, when the command queuing is disabled other functionality of the device is as if the device does not support command queuing.

3.2.1.3 New register for Command queuing

■ QSR (Queue Status Register)

The 32bit QSR carries the state of tasks in the queue at a specific point in time.

The host has read access to this register through device response to SEND_STATUS command(CMD13 with bit[15] = "1"), R1's argument will be the QSR.

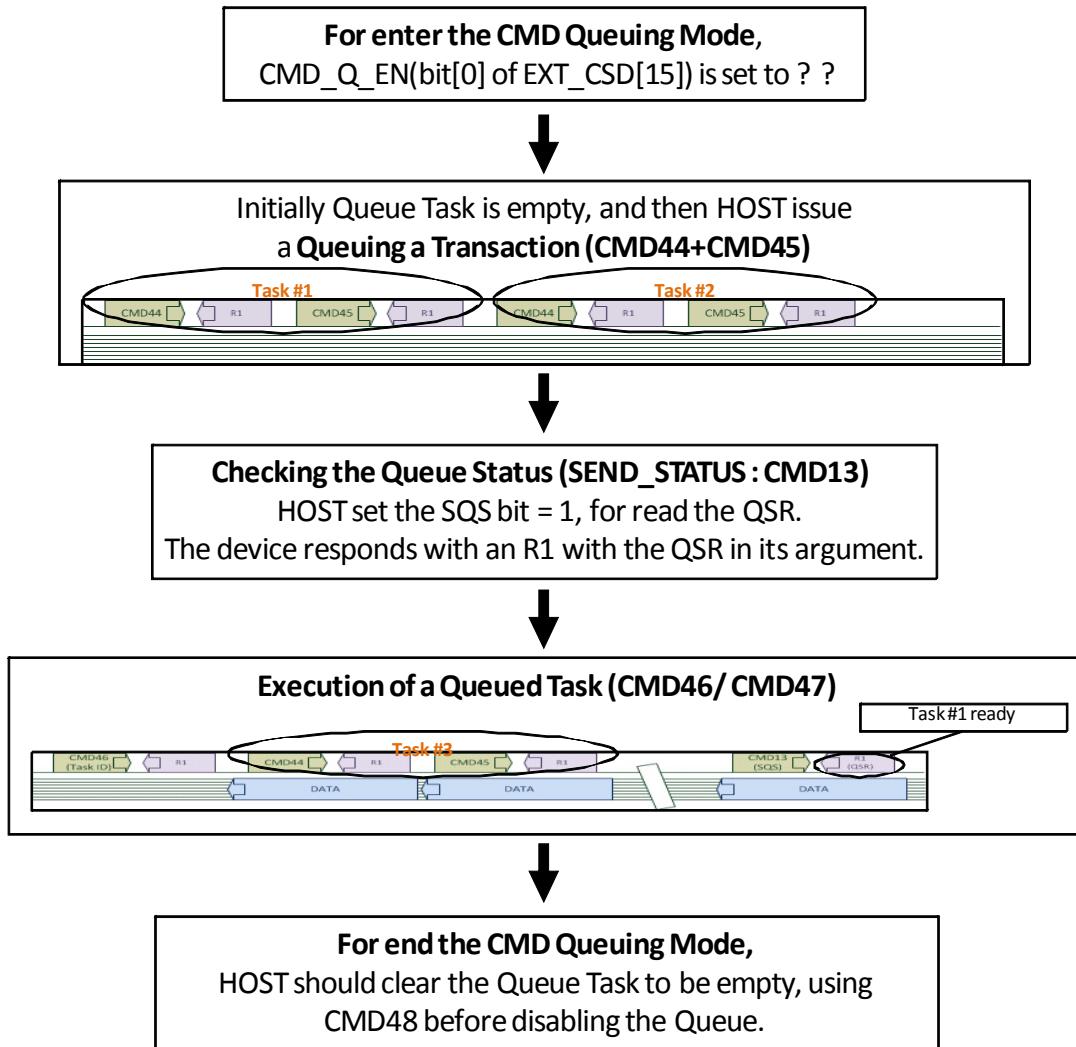
Every bit in the QSR represents the task who's ID corresponds to the bit index.

If bit QSR[i] = '0', the queued task with a Task ID i is not ready for execution.

The task may be queued and pending, or the Task ID is unused.

If bit QSR[i] = '1', the queued task with Task ID i is ready for execution.

3.2.1.4 Command queuing Flows



3.2.2 Cache Barrier

There are cases where the host is not interested in flushing the data right away, however it would like to keep an order between different cached data.

The flushing can be delayed by the device to some later idle time. Barrier commands avoid the long delay by flush commands.

- Arrival sequence



- Flush sequence



3.2.2.1 EXT_CSD Register for Cache Barrier

■ BARRIER_SUPPORT[488](Read only)

This field indicates whether the device supports the barrier command.

BIT[7:0]: 0: Barrier command is not supported

1: Barrier command is supported

Bit	Field	Supportability
Bit[7:0]	BARRIER_SUPPORT	Supported (0x1)

■ FLUSH_CACHE[32](W/E_P)

A barrier command is issued by setting BARRIER bit. All data cached before the barrier shall be flushed to the non-volatile memory before any request after the barrier command.

Data in the cache shall be flushed to the non-volatile storage by setting the FLUSH bit.

BIT[1]: 0: Reset value

1: Set barrier

BIT[0]: 0: Reset value

1: Triggers the flush

Bit	Field	Supportability
Bit[7:2]	Reserved	-
Bit[1]	BARRIER	-
Bit[0]	FLUSH	-

■ BARRIER_CTRL[31](R/W)

This field is used by the host enable barrier command mechanism if supported by the device.

BIT[0]: 0: Barrier feature is OFF

1: Barrier feature is ON

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	BARRIER_EN	-

3.2.2.2 Cache barrier Flows**1. Support Barrier command**

- The device exposes its barrier support capability via the BARRIER_SUPPORT (EXT_CSD byte [486])

2. Enable Barrier command

- The host shall set bit 0 of BARRIER_EN (EXT_CSD byte [31])

3. Cache on**4. Send data****5. Set Barrier**

- The host shall set both BARRIER bit and FLUSH bit of the FLUSH_CACHE (EXT_CSD byte [32])

3.2.3 Cache Flushing Report

For devices which flush cached data in an in-order manner, cache barrier commands are redundant and impose a needless overhead to the device and host.

3.2.3.1 EXT_CSD Register for Cache Flushing Report

■ CACHE_FLUSH_POLICY[240](Read only)

BIT[0]: 0: Device flushing policy is not provided by the device.

1: Device is using a FIFO policy for cache flushing

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	FIFO	Supported (0x01)

3.2.4 BKOP Control

This feature allows the host to indicate to the device if it is expected to periodically manually start background operations by writing to the BKOPS_START field.

3.2.4.1 EXT_CSD Register for BKOP Control

■ BKOP_EN[163](R/W/E, R/W)

BIT[1](R/W/E): 0: Device shall not perform background operations while not servicing the host.

1: Device may perform background operations while not servicing the host.

BIT[0] (R/W): 0: Host does not support background operations handling and is not expected to write to BKOPS_START field.

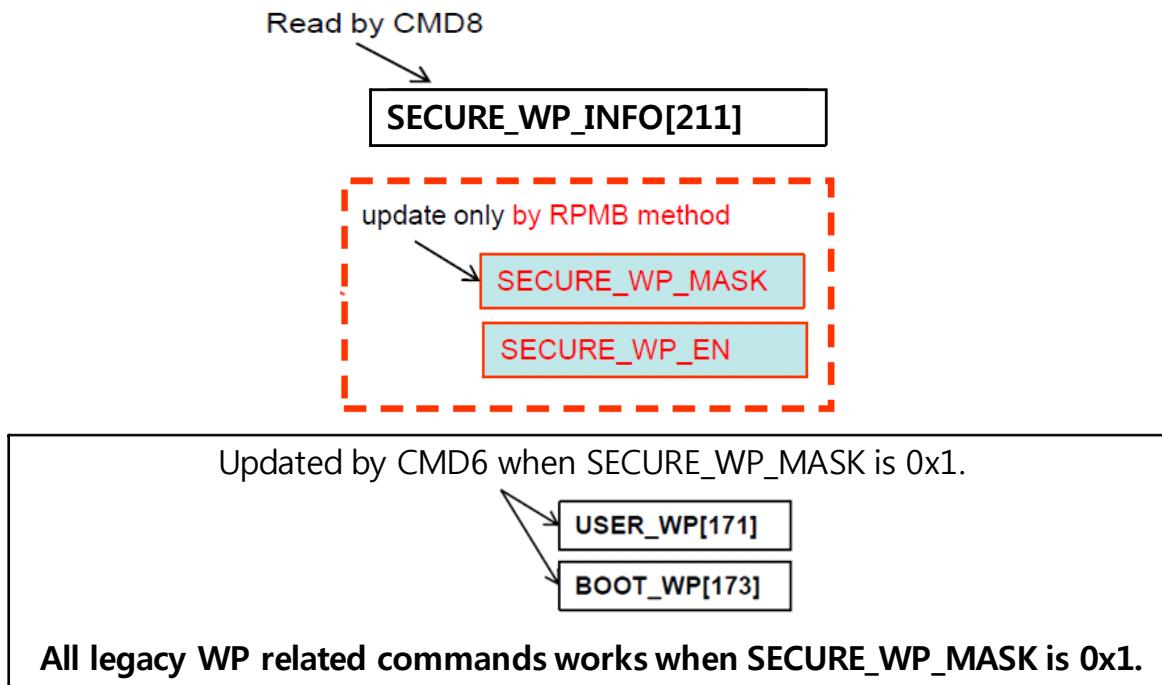
1: Host is indicating that it shall periodically write to BKOPS_START field to manually start background operations.

Bit	Field	Supportability
Bit[7:2]	Reserved	-
Bit[1]	AUTO_EN	-
Bit[0]	MANUAL_EN	-

3.2.5 Secure Write Protection

Any application running on the host may issue write protection by updating fields of write protection related EXT_CSD, like USER_WP[171], BOOT_WP[173], by issuing CMD6, CMD8, CMD28 and CMD29 (Legacy mode). However there are weak points in the legacy mode. To prevent un-authorized changes, host should enter the secure write protect mode

- In Secure WP Mode, WP related EXT_CSDs (EXT_CSD[171],[173]) can be updated only if SECURE_WP_MASK fields is 0x1.
- Secureness is provided by allowing only RPMB method to update the register for SECURE_WP_MASK.
- Automatic Write protection mode is added to prevent security hole by power-control security attack.



3.2.5.1 EXT_CSD Register for Secure Write Protection

■ SECURE_WP_INFO[211](Read Only)

The SECURE_WP_SUPPORT field indicates whether the device is supporting secure write protection mode. The SECURE_WP_EN_STATUS is showing the value of SECURE_WP_EN defined in Authenticated Device Configuration Area.

BIT[1] 0: Legacy Write Protection mode
1: Secure Write Protection mode

BIT[0] 0: Secure Write Protection is NOT supported by this device
1: Secure Write Protection is supported by this device

Bit	Field	Supportability
Bit[7:2]	Reserved	-
Bit[1]	SECURE_WP_EN_STATUS	-
Bit[0]	SECURE_WP_SUPPROT	Supported (0x01)

3.2.5.2 New register for Secure Write Protection

■ Authenticated Device Configuration Area

- Hidden register instead of EXT_CSD register for SECURE_WP_MODE_CONFIG, ENABLE.
- Those two SECURE_WP_MODE_CONFIG and SECURE_WP_MODE_ENABLE registers are defined in Device Configuration area, and those register should be updated only by Authenticated Device Configuration write request. (RPMB)

Name	Field	Size (Bytes)	Cell Type	Address
Reserved		253	-	[255:3]
Secure Write Protect Configuration	SECURE_WP_MODE_CONFIG	1	R/W/E_P	[2]
Secure Write Protect Enable	SECURE_WP_MODE_ENABLE	1	R/W/E	[1]
Reserved		1		[0]

■ Authenticated Device Configuration Area (1) : SECURE_WP_MODE_ENABLE (R/W/E)

The byte is to enter/exit the secure write protection mode.

If host want a device to enter the secure Write Protection mode, host set the SECURE_WP_EN bit as '0x1' in this register using Authenticated Device Configuration Write request. This register can be read using Authenticated Device Configuration Read request. If there are already write protected groups or write protected boot partitions, those will be preserved when entering or exiting secure Write protected mode.

Bit [0] 0: Legacy Write Protection mode.

(TMP_WRITE_PROTECT[12] , PERM_WRITE_PROTECT[13] is updated by CMD27. USER_WP[171],
BOOT_WP[173] and BOOT_WP_STATUS[174] are updated by CMD6.)

1: Secure Write Protection mode.

(The access to the write protection related EXT_CSD and CSD fields depends on the value of
SECURE_WP_MASK bit in SECURE_WP_MODE_CONFIG field.)

The default value of this field is 0x0.

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	SECURE_WP_MODE_ENABLE	-

■ Authenticated Device Configuration Area (2) : SECURE_WP_MODE_CONFIG (R/W/E_P)

In secure write protected mode, the updatability of USER_WP[171], BOOT_WP[173], TMP_WRITE_PROTECT[12] and PERM_WRITE_PROTECT[13] are controlled by this mask value.

Bit [0] 0: Disabling updating WP related EXT_CSD and CSD fields. CMD27 (Program CSD) will generate generic error for setting TMP_WRITE_PROTECT[12] , PERM_WRITE_PROTECT[13]. CMD6 for updating USER_WP[171], BOOT_WP[173] and BOOT_WP_STATUS[174] generates SWITCH_ERROR. If a force erase command is issued, the command will fail (Device stays locked) and the LOCK_UNLOCK_FAILED error bit will be set in the status register. If CMD28 or CMD29 is issued, then generic error will be occurred. Power-on WriteProtected boot partitions will keep protected mode after power failure, H/W reset assertion and any CMD0 reset. The device keeps the current value of BOOT_WP in the EXT_CSD register to be same after power cycle, H/W reset assertion, and any CMD0 reset.

1: Enabling updating WP related EXT_CSD and CSD fields.

(TMP_WRITE_PROTECT[12] , PERM_WRITE_PROTECT[13] , USER_WP[171], BOOT_WP[173] and BOOT_WP_STATUS[174] are accessed using CMD6, CMD8 and CMD27. If a force erase command is issued and accepted, then ALL THE DEVICE CONTENT WILL BE ERASED including the PWD and PWD_LEN register content and the locked Device will get unlocked. If a force erase command is issued and power-on protected or a permanently-write-protected write protect groups exist on the device, the command will fail (Device stays locked) and the LOCK_UNLOCK_FAILED error bit will be set in the status register. An attempt to force erase on an unlocked Device will fail and LOCK_UNLOCK_FAILED error bit will be set in the status register. Write Protection is applied to the WPG indicated by CMD28 with the WP type indicated by the bit[2] and bit[0] of USER_WP[171]. All temporary WP Groups and power-on Write Protected boot partitions become writable/erasable temporarily which means write protect type is not changed. All power-on and permanent WP Groups in user area will not become writable/erasable temporarily. Those temporarily writable/erasable area will become write protected when this bit is cleared to 0x0 by the host or when there is power failure, H/W reset assertion and any CMD0 reset. The device keeps the current value of BOOT_WP CSD register to be same after power cycle, H/W reset assertion, and any CMD0 reset.)

The default value of this field is 0x0.

Bit	Field	Supportability
Bit[7:1]	Reserved	-
Bit[0]	SECURE_WP_MODE_CONFIG	-

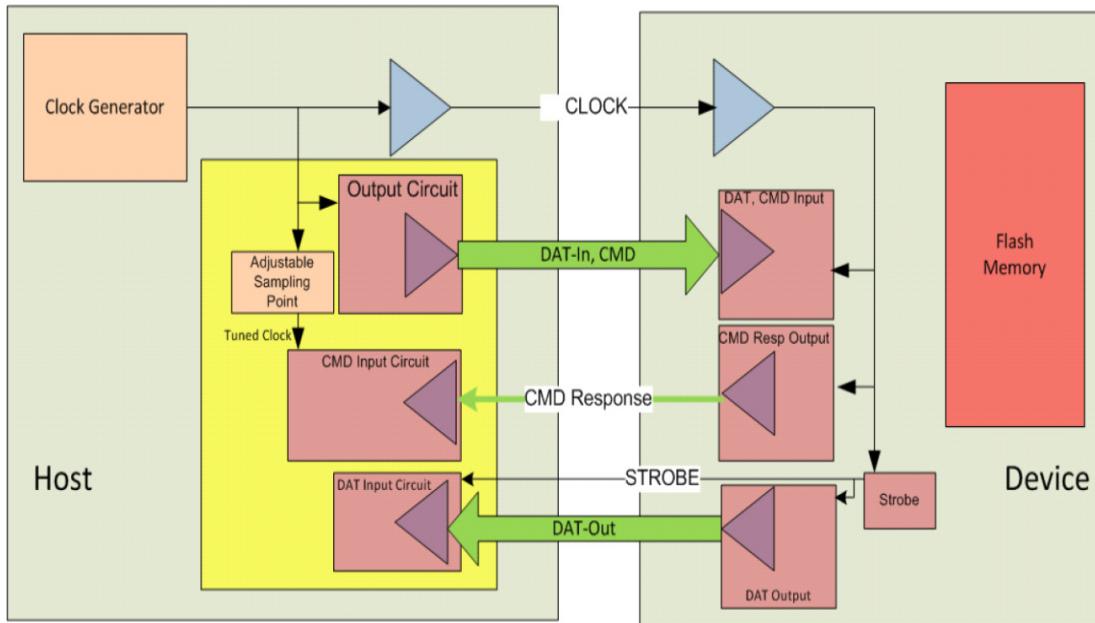
3.2.5.3 RPMB Types for accessing Authenticated Device Configuration Area

Secure WP Enable & Configuration registers are defined in Authenticated Device Configuration Area which only can be accessible by new RPMB operations.

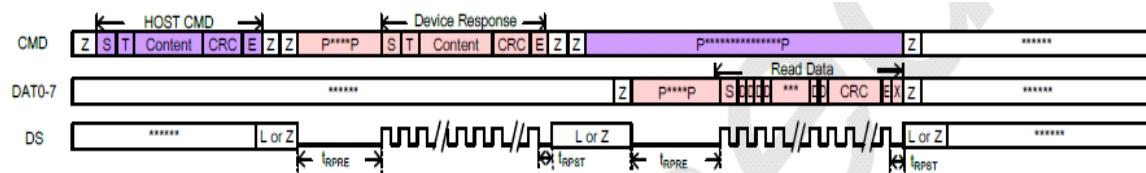
Request Message Types	
0x0001	Authentication key programming request
0x0002	Reading of the Write Counter value –request
0x0003	Authenticated data write request
0x0004	Authenticated data read request
0x0005	Result read request
0x0006	Authenticated Device Configuration Write request
0x0007	Authenticated Device Configuration Read request
Response Message Types	
0x0100	Authentication key programming response
0x0200	Reading of the Write Counter value –response
0x0300	Authenticated data write response
0x0400	Authenticated data read response
0x0600	Authenticated Device Configuration Write response
0x0700	Authenticated Device Configuration Read response

4.2.6 Enhanced strobe

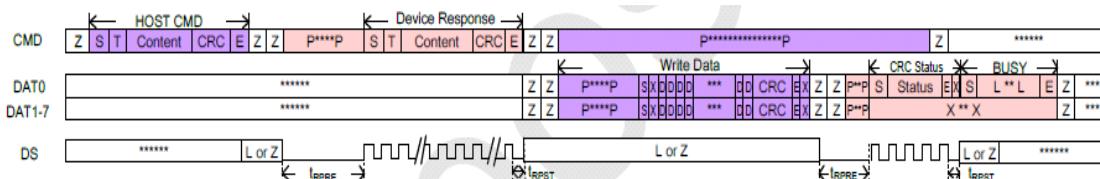
In Enhanced Strobe mode DATA OUT, CRC Response and CMD Response are all synched to STROBE clocks. The timing relation between CMD Response output signals and STROBE clocks is the same as defined for DATA Out to STROBE clocks.



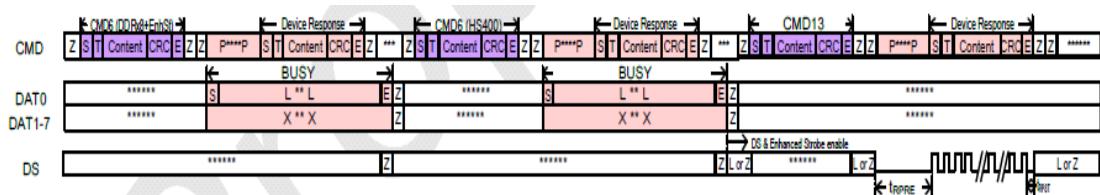
[Figure 7] HS400 Host and Device block diagram (when Enhanced Strobe is enabled)



[Figure 8] Enhanced Strobe signals for CMD Response and Data Out (Read operation)



[Figure 9] Enhanced Strobe signals for CMD Response and Data Out (Write operation)



[Figure 10] HS400 mode change with Enhanced Strobe

3.2.6.1 EXT_CSD Register for Enhanced Strobe

■ STROBE_SUPPORT[184](Read only)

This register indicates whether a device supports Enhanced Strobe mode for operation modes that STROBE is used for HS400.

BIT[7:0]: 0: Indicates No support of Enhanced Strobe mode

1: Indicates the device supports Enhanced Strobe mode

Bit	Field	Supportability
Bit[7:0]	STROBE_SUPPORT	Supported (0x01)

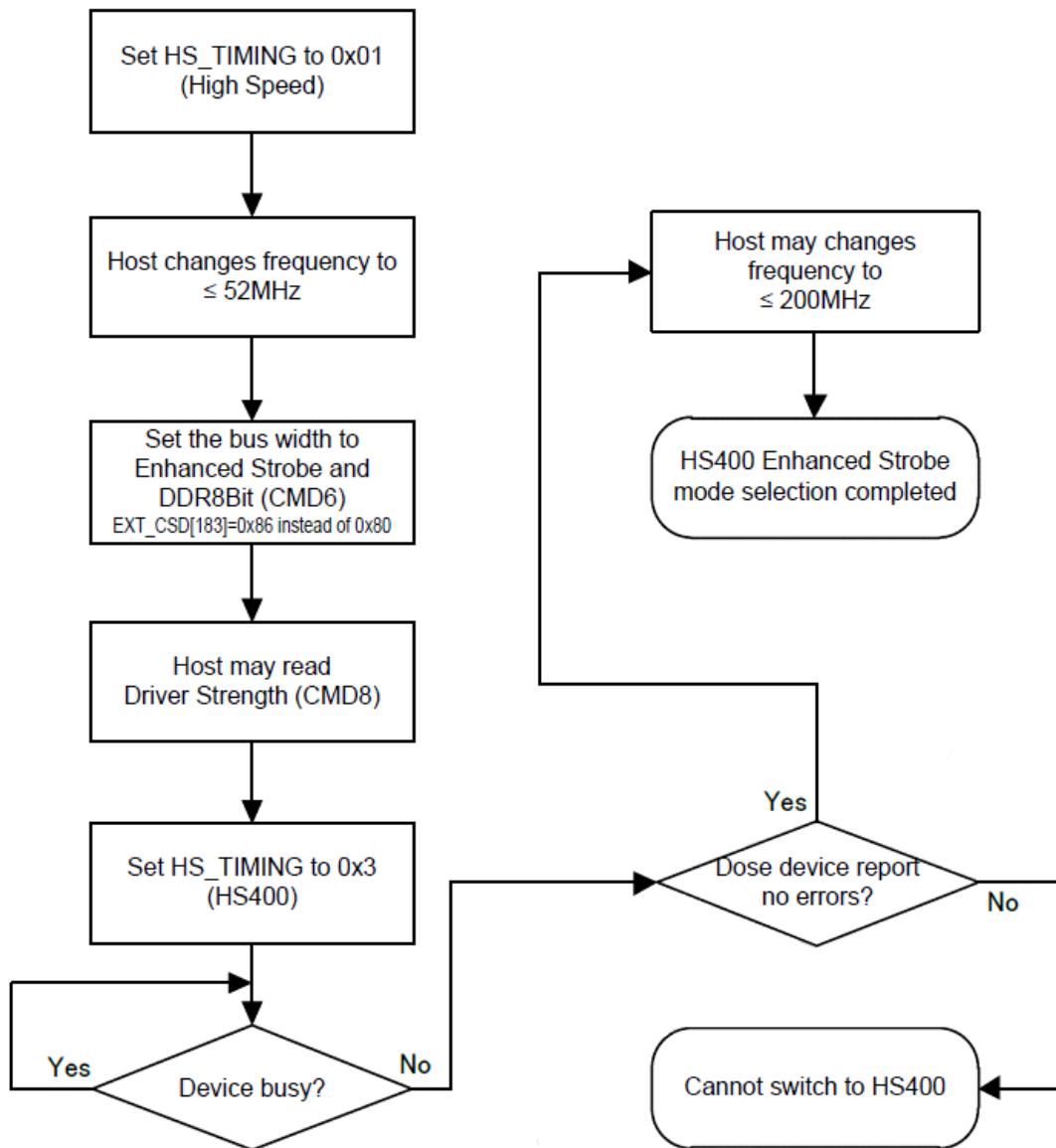
3.2.6.2 HS400 mode (Enhanced Strobe) selection

This selection flow describes how to initialize the eMMC device in HS400 mode while enabling Enhanced Strobe without the need for tuning procedure.

After the host initializes the device, host check whether the device supports the HS400 mode and Enhanced Strobe by reading the DEVICE_TYPE and STROBE_SUPPORT fields in the Extended CSD register.

After power-on or software reset (CMD0), the interface timing of the device is set as the default 'Backward Compatible Timing'.

In order to switch to HS400 mode with Enhanced Strobe, host should perform the following steps.



[Figure 11] HS400 Bus mode (Enhanced Strobe) selection flow diagram

3.2.7 RPMB throughput improvement

This feature is proposed for RPMB write data size to improve the RPMB throughput in eMMC5.1 spec. In the eMMC5.0 spec, REL_WR_SEC_C[222] register shall be set to 1 (hence the granularity is always 512B.)

- For reliable write to RPMB partition, there is limitation that block count can not exceed the size of REL_WR_SEC_C x 512B. In eMMC5.1, the supported RPMB write access size is 256B, 512B, and 8KB.

■ WRITE_REL_PARAM (EXT_CSD[166]): Read only

When EXT_CSD[166][4] (R) = 0

- Device does not support large RPMB write transfer
- The behavior is same as eMMC v5.0 or earlier

When EXT_CSD[166][4] (R) = 1

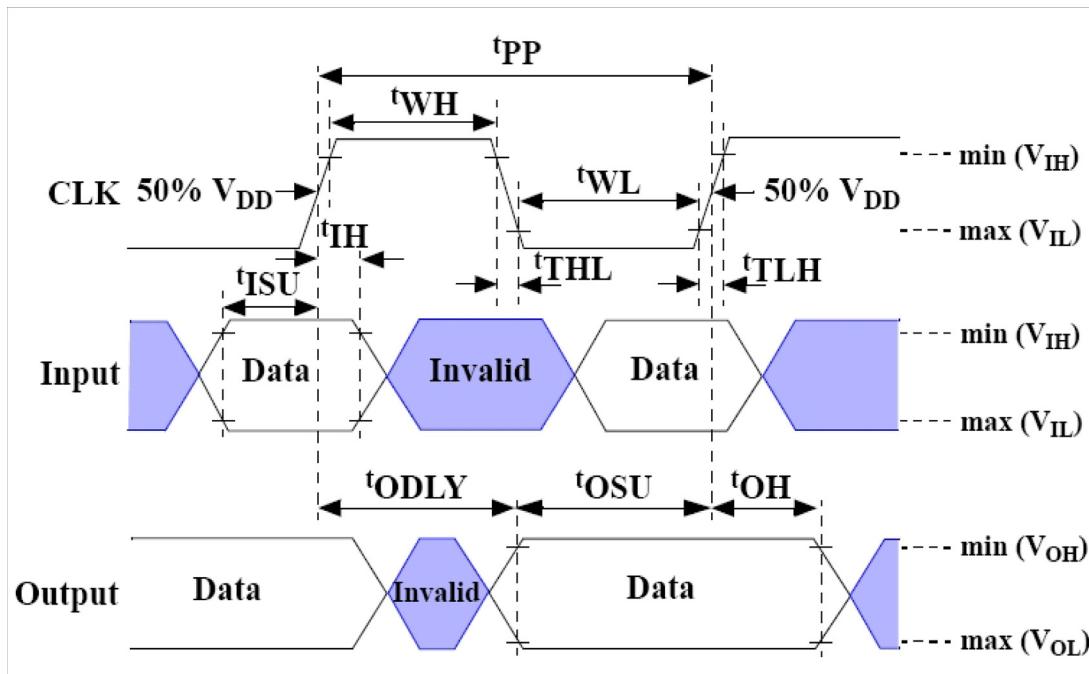
- Device supports large RPMB write transfer
- Host transfers small RPMB write with block count = 1 or 2 (256/512B)
- Host transfers large RPMB write with block count = 32 (8KB)
 - * the start address should be 8KB aligned, and the transferred data (8KB) is all-new or all-old

Bit	Field	Supportability
Bit[4]	EN_RPMB_REL_WR	Supported (0x01)

4. eMMC general parameters

4.1 Timing

4.1.1 Bus timing



Data must always be sampled on the rising edge of the clock.

[Figure 12] Timing diagram: data input/output

Parameter	Symbol	Min	Max	Unit	Remark
Clock CLK					
Clock frequency data transfer mode (PP)	f_{PP}	0	52	MHz	$C_L \leq 30 \text{ pF}$ Tolerance: +100KHz
Clock frequency identification mode (OD)	f_{OD}	0	400	MHz	Tolerance: +20KHz
Clock high time	t_{WH}	6.5		ns	$C_L \leq 30 \text{ pF}$
Clock low time	t_{WL}	6.5		ns	$C_L \leq 30 \text{ pF}$
Clock rise time	t_{TLH}		3	ns	$C_L \leq 30 \text{ pF}$
Clock fall time	t_{THL}		3	ns	$C_L \leq 30 \text{ pF}$
Inputs CMD, DAT (referenced to CLK)					
Input set-up time	t_{ISU}	3		ns	$C_L \leq 30 \text{ pF}$
Input hold time	t_{IH}	3		ns	$C_L \leq 30 \text{ pF}$
Outputs CMD, DAT (referenced to CLK)					
Output delay time during data transfer	t_{ODLY}		13.7	ns	$C_L \leq 30 \text{ pF}$
Output hold time	t_{OH}	2.5		ns	$C_L \leq 30 \text{ pF}$
Signal rise time	t_{RISE}		3	ns	$C_L \leq 30 \text{ pF}$
Signal fall time	t_{FALL}		3	ns	$C_L \leq 30 \text{ pF}$

[Table 7] High-speed eMMC interface timing

- CLK timing is measured at 50% of VDD.
- eMMC shall support the full frequency range from 0-26Mhz, or 0-52MHz
- CLK rising and falling times are measured by min (V_{IH}) and max (V_{IL}).
- Input CMD, DAT rising and falling times are measured by min (V_{IH}) and max (V_{IL}), and output CMD, DAT rising and falling times are measured by min (V_{OH}) and max (V_{OL}).

Parameter	Symbol	Min	Max	Unit	Remark
Clock CLK					
Clock frequency data transfer mode (PP)	f_{PP}	0	26	400	$C_L \leq 30 \text{ pF}$
Clock frequency identification mode (OD)	f_{OD}	0	400	MHz	
Clock high time	t_{WH}	10		ns	$C_L \leq 30 \text{ pF}$
Clock low time	t_{WL}	10		ns	$CL \leq 30 \text{ pF}$
Clock rise time	t_{TLH}		10	ns	$CL \leq 30 \text{ pF}$
Clock fall time	t_{THL}		10	ns	$CL \leq 30 \text{ pF}$
Inputs CMD, DAT (referenced to CLK)					
Input set-up time	t_{ISU}	3		ns	$CL \leq 30 \text{ pF}$
Input hold time	t_{IH}	3		ns	$CL \leq 30 \text{ pF}$
Outputs CMD, DAT (referenced to CLK)					
Output set-up time	t_{OSU}	11.7		ns	$CL \leq 30 \text{ pF}$
Output hold time	t_{OH}	8.3		ns	$CL \leq 30 \text{ pF}$

[Table 8] Backward-compatible eMMC interface timing

- eMMC must always start with the backward-compatible interface timing. The timing mode can be switched to high-speed timing by the host sending the switch command (CMD6) with the argument for high speed interface select.
- CLK timing is measured at 50% of VDD.
- CLK rising and falling times are measured by min (V_{IH}) and max (V_{IL}).
- t_{OSU} and t_{OH} are defined as values from clock rising edge. However, there may be cards or devices which utilize clock falling edge to output data in backward compatibility mode.

Therefore, it is recommended for hosts either to set t_{WL} value as long as possible within the range which should not go over

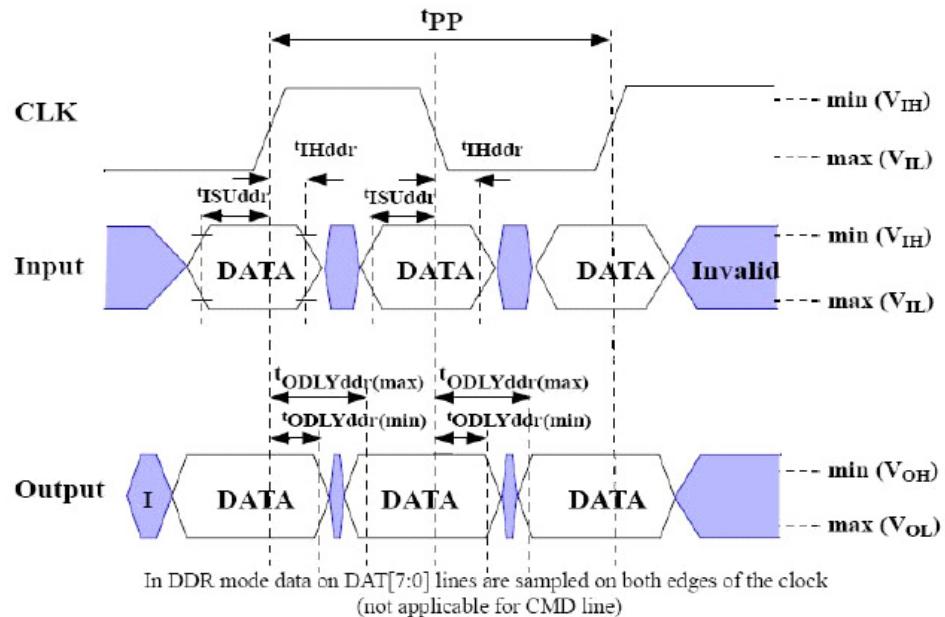
$t_{CK} - t_{OH}(\text{min})$ in the system or to use slow clock frequency, so that host could have data set up margin for those devices.

In this case, each device which utilizes clock falling edge might show the correlation either between t_{WL} and t_{OSU} or between

t_{CK} and t_{OSU} for the device.

4.1.2 Bus timing for DAT Signals During 2x Data Rate Operation

These timings apply to the DAT[7:0] signals only when the device is configured for dual data mode operation. In dual data mode, the DAT signals operate synchronously of both the rising and the falling edges of CLK.



[Figure 13] Timing diagram: data input/output in dual data rate mode

Parameter	Symbol	Min.	Max.	Unit	Remark
Input CLK⁽¹⁾					
Clock duty cycle		45	55	%	Includes jitter, phase noise
Clock rise time	t_{TLH}		3	ns	$CL \leq 30 \text{ pF}$
Clock fall time	t_{THL}		3	ns	$CL \leq 30 \text{ pF}$
Input CMD (referenced to CLK-SDR mode)					
Input set-up time	t_{ISUddr}	3		ns	$CL \leq 20 \text{ pF}$
Input hold time	t_{IHDDR}	3		ns	$CL \leq 20 \text{ pF}$
Output CMD (referenced to CLK-SDR mode)					
Output delay time during data transfer	t_{ODLY}		13.7	ns	$CL \leq 20 \text{ pF}$
Output hold time	t_{OH}	2.5		ns	$CL \leq 20 \text{ pF}$
Signal rise time	t_{RISE}		3	ns	$CL \leq 20 \text{ pF}$
Signal fall time	t_{FALL}		3	ns	$CL \leq 20 \text{ pF}$
Input DAT (referenced to CLK-DDR mode)					
Input set-up time	t_{ISUddr}	2.5		ns	$CL \leq 20 \text{ pF}$
Input hold time	t_{IHddr}	2.5		ns	$CL \leq 20 \text{ pF}$
Outputs DAT (referenced to CLK-DDR mode)					
Output delay time during data transfer	$t_{ODLYddr}$	1.5	7	ns	$CL \leq 20 \text{ pF}$
Signal rise time(DAT0-7) ⁽²⁾	t_{RISE}		2	ns	$CL \leq 20 \text{ pF}$
Signal fall time (DAT0-7)	t_{FALL}		2	ns	$CL \leq 20 \text{ pF}$

[Table 9] Dual data rate interface timings

- **NOTE 1.** CLK timing is measured at 50% of VDD.
- **NOTE 2.** Inputs DAT rising and falling times are measured by min (V_{IH}) and max (V_{IL}), and outputs CMD, DAT rising and falling times are measured by min (V_{OH}) and max (V_{OL})

4.2 Power mode

4.2.1 eMMC power-up guidelines

eMMC power-up must adhere to the following guidelines:

- When power-up is initiated, either V_{CC} or V_{CCQ} can be ramped up first, or both can be ramped up simultaneously.
- After power up, eMMC enters the pre-idle state. The power up time of each supply voltage should be less than the specified tPRU (tPRUH, tPRUL or tPRUV) for the appropriate voltage range.

• If eMMC does not support boot mode or its BOOT_PARTITION_ENABLE bit is cleared, eMMC moves immediately to the idle state. While in the idle state, eMMC ignores all bus transactions until receiving CMD1.

eMMC begins boot operation with the argument of 0xFFFFFFFFFA. If boot acknowledge is finished, eMMC shall send acknowledge pattern "010" to the host within the specified time. After boot operation is terminated, eMMC enters the idle state and shall be ready for CMD1 operation. If eMMC receives CMD1 in the pre-boot state, it begins to respond to the command and moves to the card identification mode.

• When eMMC is initiated by alternative boot command(CMD0 with arg=0xFFFFFFFFFA), all the data will be read from the boot partition and then eMMC automatically goes to idle state, but hosts are still required to issue CMD0 with arg=0x0000000000 in order to complete a boot mode properly and move to the idle state. While in the idle state, eMMC ignores all bus transactions until it receives CMD1.

• CMD1 is a special synchronization command which is used to negotiate the operating voltage range and poll the device until it is out of its power-up sequence. In addition to the operating voltage profile of the device, the response to CMD1 contains a busy flag indicating that the device is still working on its power-up procedure and is not ready for identification. This bit informs the host that the device is not ready, and the host must wait until this bit is cleared. The device must complete its initialization within 1 second of the first CMD1 issued with a valid OCR range.

• If the eMMC device was successfully partitioned during the previous power up session (bit 0 of EXT_CSD byte [155] PARTITION_SETTING_COMPLETE successfully set) then the initialization delay is (instead of 1s) calculated from INI_TIMEOUT_PA (EXT_CSD byte [241]). This timeout applies only for the very first initialization after successful partitioning. For all the consecutive initialization 1sec time out will be applied.

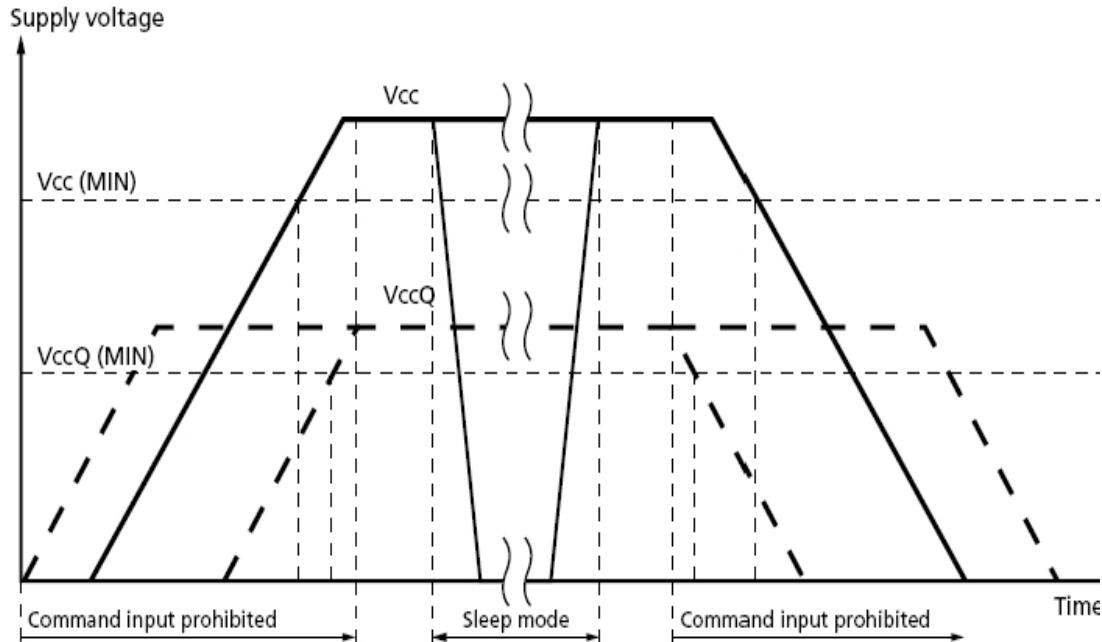
• The bus master moves the device out of the idle state. Because the power-up time and the supply ramp-up time depend on the application parameters such as the bus length and the power supply unit, the host must ensure that power is built up to the operating level (the same level that will be specified in CMD1) before CMD1 is transmitted.

• After power-up, the host starts the clock and sends the initializing sequence on the CMD line. The sequence length is the longest of: 1ms, 74 clocks, the supply ramp-up time, or the boot operation period. An additional 10 clocks (beyond the 64 clocks of the power-up sequence) are provided to eliminate power-up synchronization problems.

- Every bus master must implement CMD1.

4.2.2 eMMC Power Cycling

The master can execute any sequence of V_{CC} and V_{CCQ} power-up/power-down. However, the master must not issue any commands until V_{CC} and V_{CCQ} are stable within each operating voltage range. After the slave enters sleep mode, the master can power-down V_{CC} to reduce power consumption. It is necessary for the slave to be ramped up to V_{CC} before the host issues CMD5 (SLEEP_AWAKE) to wake the slave unit.



[Figure 14] eMMC power cycle

If V_{CC} or V_{CCQ} is below 0.5 V for longer than 1 ms, the slave shall always return to the pre-idle state, and perform the appropriate boot behavior. The slave will behave as in a standard power up condition once the voltages have returned to their functional ranges.

An exception to this behavior is if the device is in sleep state, in which the voltage on V_{CC} is not monitored.

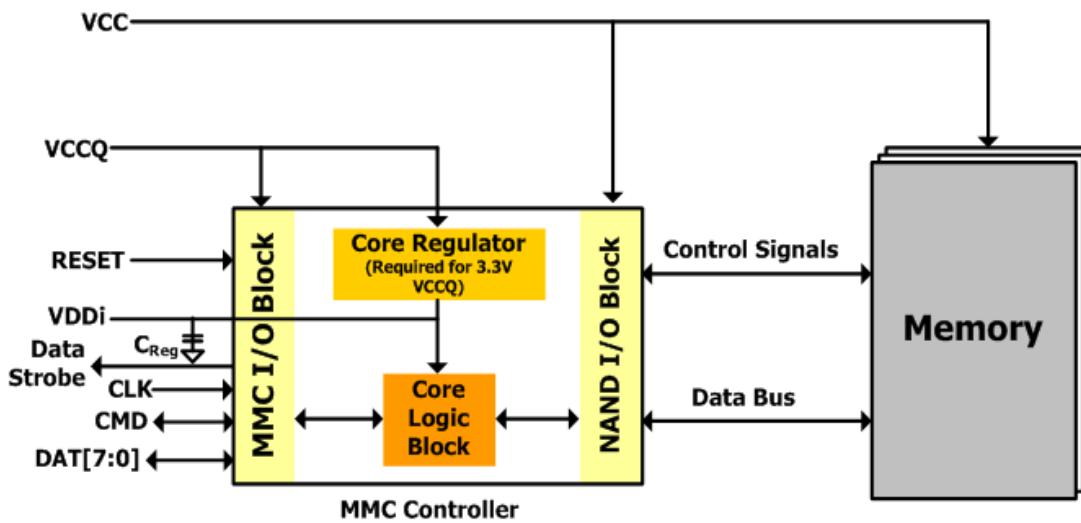
4.2.3 Leakage

Parameter	Symbol	Min	Max.	Unit	Remark
	BGA	-0.5	$V_{ccq} + 0.5$	V	
All inputs					
Input leakage current (before initialization sequence and/or the internal pull up resistors connected)		-100	100	μA	
All outputs					
Output leakage current (before initialization sequence)		-100	100	μA	
Output leakage current (after initialization sequence)		-2	2	μA	

[Table 10] General operation conditions

4.2.4 Power Supply

In eMMC, V_{cc} is used for the NAND core voltage and NAND interface; V_{ccq} is for the controller core and eMMC interface voltage shown in Figure 15. The core regulator is optional and only required when internal core logic voltage is regulated from V_{ccq} . A C_{reg} capacitor must be connected to the $VDDi$ terminal to stabilize regulator output on the system.



[Figure 15] eMMC internal power diagram

eMMC supports one or more combinations of V_{CC} and V_{CCQ} as shown in Table 11.

The available voltage configuration is shown in Table 12.

Parameter	Symbol	Min	Max.	Unit	Remark
Supply voltage (NAND)	V_{CC}	2.7	3.6	V	
		1.7	1.95	V	Not supported
Supply voltage (I/O)	V_{CCQ}	2.7	3.6	V	
		1.7	1.95	V	
Supply power-up for 3.3V	t_{PRUH}		35	ms	
Supply power-up for 1.8V	t_{PRUL}		35	ms	

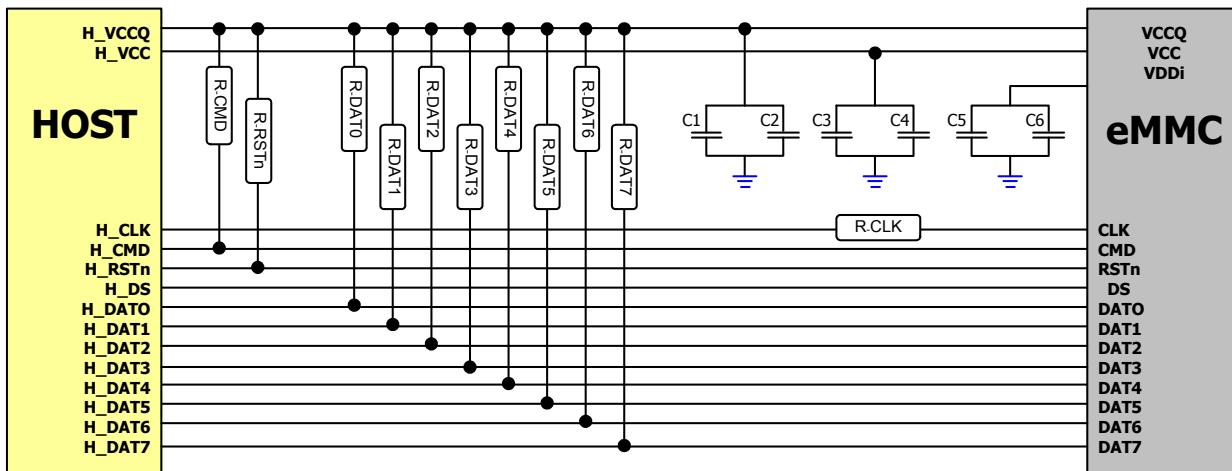
[Table 11] eMMC power supply voltage

		V_{CCQ}	
		1.7V ~ 1.95V	2.7V ~ 3.6V
V_{CC}	2.7V–3.6V	Valid	Valid (1)
	1.7V–1.95V	Not Valid	Not Valid

[Table 12] eMMC voltage combinations

- NOTE 1. $V_{CCQ}(I/O)$ 3.3 volt range is not supported in either HS200 or HS400 devices.

4.3 Connection Guide



[Figure 16] Connection guide drawing

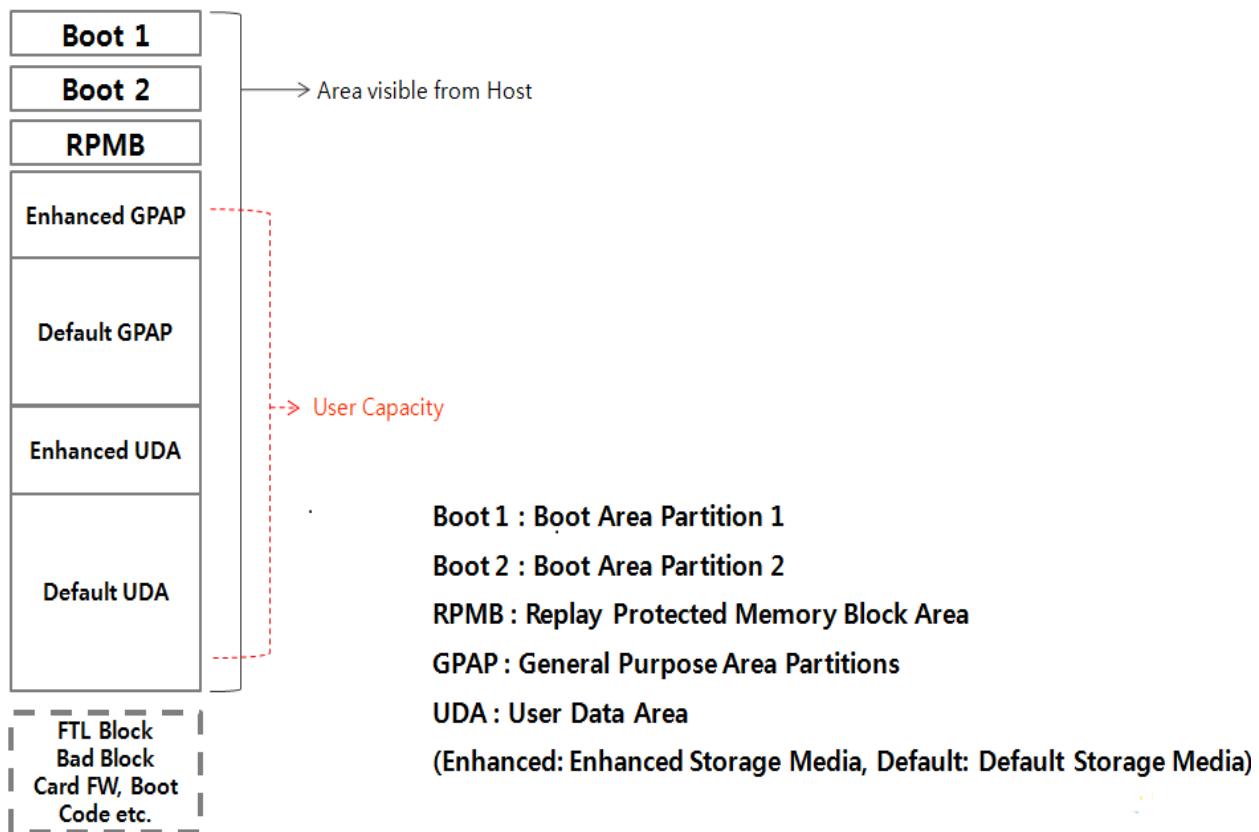
Parameter	Symbol	Min	Max	Recommend	Unit	Remark
Pull-up resistance for CMD	R_CMD	4.7	100	10	kohm	Pull-up resistance should be put on CMD line to prevent bus floating.
Pull-up resistance for DAT0~7	R_DAT	10	100	50	kohm	Pull-up resistance should be put on DAT line to prevent bus floating.
Data strobe(DS)	R_DS	NC	NC	NC	-	It is not necessary to put pull-up/pull-down resistance on DS line since DS is internally pulled down. Direct connection to host is required and please float this pin if it is not used
Pull-up resistance for RSTn	R_RSTn	10	100	50	kohm	It is not necessary to put pull-up resistance on RSTn line if host does not use H/W reset. (Extended CSD register [162] = 0b)
Serial resistance on CLK	R_CLK	0	30	27	ohm	To reduce overshooting/undershooting Note: If the host uses HS200, we recommend to remove this resister for better CLK signal
V _{ccq} capacitor value	C1 & C2	2±0.22	4.7	2±0.22	uF	Coupling cap should be connected with V _{ccq} closely.
V _{cc} capacitor value(≤8GB)	C3 & C4	4.72±10%	10	4.72±10%	uF	Coupling cap should be connected with V _{cc} closely. V _{cc} /V _{ccq} cap. value would be up to Host requirement and the application system characteristics.
VDDi capacitor value	C5 & C6	0	2.2	0.1	uF	Coupling cap should be connected with VDDi and Vssq as closely possible. (Internal Cap : 1uF)

[Table 13] Connection guide specification

5. eMMC basic operations

5.1 Partitioning

5.1.1 User density



[Figure 17] Partition diagram

■ Boot and RPMD partition size

Density	Boot 1,2 and RPMB partition size
32GB	4096KB (4MB)

■ User density size

Capacity	SEC_COUNT	Capacity	Percentile
32GB	61,071,360 (0x3A3E0000)	31,268,536,320 Bytes (29.12GB)	91.00%

- 1sector=512 bytes.
- The total usable capacity of the eMMC may be less than total physical capacity because a small portion of the capacity is used for NAND flash management and maintenance purpose.

■ Maximum enhanced partition size

Enhanced user data area can be configured to store read-centric data such as sensitive data or for other host usage models.

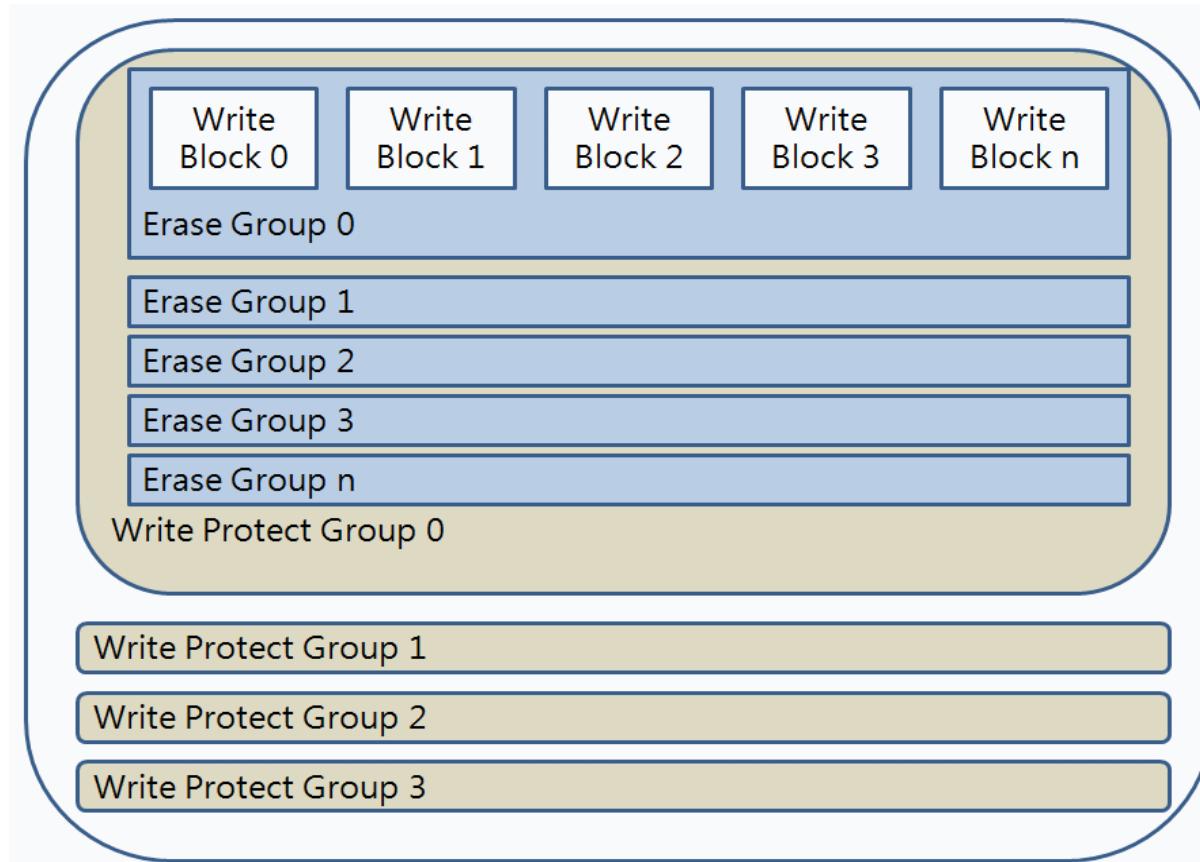
SK hynix eMMC supports Enhanced User Data Area as SLC Mode. When customer adopts some portion as enhanced user data area in User Data Area, that area occupies double the size of the original set-up size.

Capacity	Max ENH_SIZE_MULTI	HC_ERASE_GRP_SIZE	HC_WP_GRP_SIZE
32GB	0x000E8F	1h	08h

- Max Enhanced Partition Size is defined as **MAX_ENH_SIZE_MULTI x HC_WP_GRP_SIZE x HC_ERASE_GRP_SIZE x 512KByte**.

Capacity	Capacity (KB)
32GB	15,265,792

5.1.2 Erase / Write protect group size

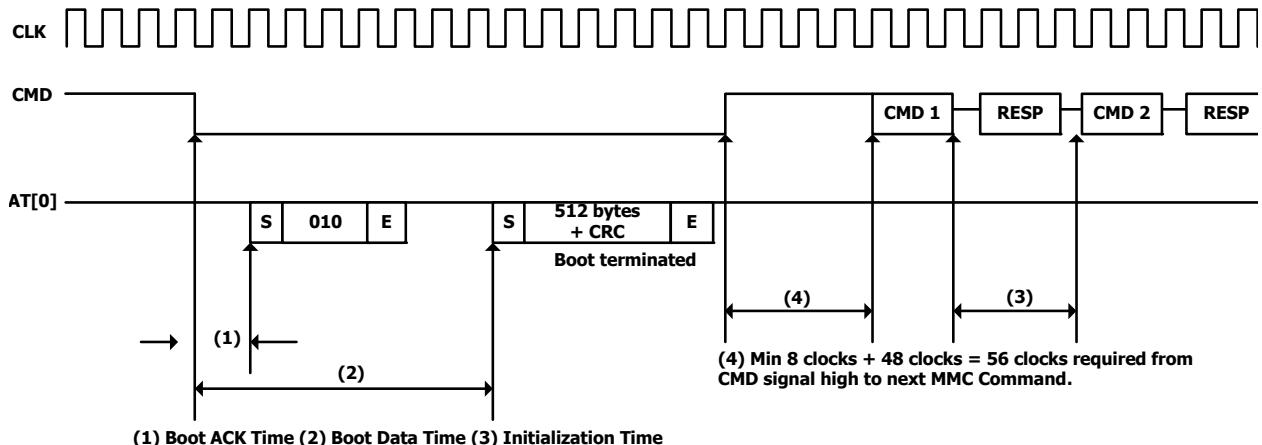


Density	Erase group size		Write protect group size
	ERASE_GROUP_DEF=0	ERASE_GROUP_DEF=1	
32GB	512KB	512KB	4MB

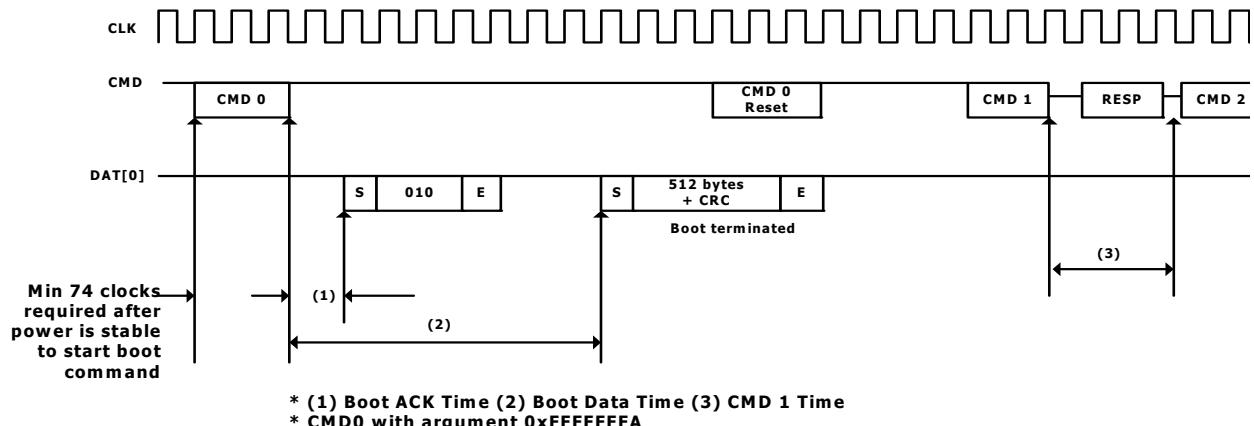
[Table 14] Erase/Write protect Group size

5.2 Boot operation

eMMC supports boot mode and alternative boot mode. eMMC also, supports high speed timing and dual data rate during boot.



[Figure 18] eMMC state diagram (Boot mode)



[Figure 19] eMMC state diagram (Alternative boot mode)

Timing Factor	Value
(1) Boot ACK Time	< 50 ms
(2) Boot Data Time	< 1 sec
(3) Initialization Time	< 1 sec

- Initialization time includes partition setting, Please refer to INI_TIMEOUT_AP in 7.4 Extended CSD Register. Initialization time is completed within 1sec from issuing CMD1 until receiving response.
- The device has to send the acknowledge pattern "010" to the master within 50ms after the CMD0 with the argument of 0xFFFFFFF is received.

6. Time out

Latency Item	Value	Remark
Write Time out (CMD To 512KB Write Done)	Max 1s	No read/program/erase failure case
Write Time out (Data To Data)	Max 500ms	
Read Time Out (CMD To the first data out)	Max 100ms	No read/program/erase failure case
Initialization Time	Max 1s	
Initialization Time (After Partitioning)	Max 3s	
Initialization after PON (Short/long)	Max 180ms	
Pre-Boot to ACK	Max 50ms	
Pre-Boot to Boot Data	Max 200ms	
Partition Switch	Max 10ms	
CMD6 Switch	Max 50ms	
CMD8 time out	Max 5ms	
Erase (Erase Group)	Max 600ms	
Trim (512B~512KB)	Max 600ms	
Discard (512B~512KB)	Max 600ms	

[Table 15] Time out value

Latency Item	Value		Remark	
Secure Trim1 Type 0, 1 (512B~512KB)	Max 5.1s		Type 0 : information removed by an erase of the physical memory Type 1 : information removed by an overwriting the addressed locations with a character followed by an erase Type 2 : Not support Type 3 : information unmapped	
Secure Trim1 Type 3 (512B~512KB)	Max 600ms			
Secure Erase Type 0,1 (Write Block 512KB)	Max 8.1s			
Secure Erase Type 3 (Write Block 512KB)	Max 600ms			
Sanitize (UDA Area)	32GB	Max 30min		
Force Erase (UDA Area)	Max 3min			
HPI	Max 50ms			
Sliding Window	Max 256ms		For 256kB randomly write case	
PON Busy Time (Short/Long)	Max 50ms / 1000ms		PON long includes GC PON short not include GC	
CMD6 Switch	Max 50ms			

[Table 16] Time out value (Continue)

- Be advised timeout values specified in table above are for testing purpose under SK hynix test pattern only and actual timeout situations may vary

7. Device registers

There are six different registers within the device interface:

- Operation conditions register (OCR)
- Card identification register (CID)
- Card specific data register (CSD)
- Relative card address register (RCA)
- DSR (Driver Stage Register)
- Extended card specific data register (EXT_CSD).

These registers are used for the serial data communication and can be accessed only using the corresponding commands. eMMC has a status register to provide information about the current device state and completion codes for the last host command.

7.1 Operation conditions register (OCR)

The 32-bit operation conditions register (OCR) stores the VDD voltage profile of eMMC and the access mode indication. In addition, this register includes a status information bit. This status bit is set if eMMC power up procedure has been finished.

OCR bit	Description	SK hynix eMMC
[6:0]	Reserved	000 0000b
[7]	1.70 - 1.95V	1b
[14:8]	2.0 - 2.6	000 0000b
[23:15]	2.7 - 3.6 (High V _{ccq} range)	1111 1111 1b
[28:24]	Reserved	000 000b
[30:29]	Access mode	10b (sector mode)
[31]	(card power up status bit (busy)) ⁽¹⁾	

[Table 17] OCR register definition

- NOTE 1. This bit is set to LOW if the card has not finished the power up routine

7.2 Card identification (CID) register

The card identification (CID) register is 128 bits wide. It contains eMMC identification information used during eMMC identification phase (eMMC protocol). Every individual eMMC has a unique identification number. The structure of the CID register is defined in the following sections.

Name	Field	Width	CID slice	CID value	Remark
Manufacturer ID	MID	8	[127:120]	0X90	
Reserved		6	[119:114]		
Card/BGA	CBX	2	[113:112]	0X1	BGA
OEM/application ID	OID	8	[111:104]	0X4A	
Product name	PNM	48	[103:56]	32GB: 0X484247346132	
Product revision	PRV	8	[55:48]	-	Not Fixed ¹⁾
Product serial number	PSN	32	[47:16]	-	Not Fixed
Manufacturing date	MDT	8	[15:8]	-	Not Fixed
CRC7 checksum	CRC	7	[7:1]	-	Not Fixed
Not used, always '1'		1	[0:0]	0X1	

1) PRV composed of the revision count of controller and the revision count of F/W patch

[Table 18] Card identification (CID) fields

7.3 Card specific data register (CSD)

The card specific data (CSD) register provides information on how to access eMMC contents. The CSD defines the data format, error correction type, maximum data access time, data transfer speed and so on. The programmable part of the register (entries marked by W or E, see below) can be changed by CMD27. The type of the CSD Registry entries in the Table 19 below is coded as follows:

- R: Read only.
- W: One time programmable and not readable.
- R/W: One time programmable and readable.
- W/E: Multiple writable with value kept after power failure, H/W reset assertion and any CMD0 reset and not readable.
- R/W/E: Multiple writable with value kept after power failure, H/W reset assertion and any CMD0 reset and readable.
- R/W/C_P: Writable after value cleared by power failure and HW/reset assertion (the value not cleared by CMD0 reset) and readable.
- R/W/E_P: Multiple writable with value reset after power failure, H/W reset assertion and any CMD0 reset and readable.
- W/E_P: Multiple writable with value reset after power failure, H/W reset assertion and any CMD0 reset and not readable.

Name	Field	Width	Cell type	CSD slice	CSD value	Remark
CSD structure	CSD_STRUCTURE	2	R	[127:126]	0X03	
System specification version	SPEC_VERS	4	R	[125:122]	0X04	
Reserved		2	R	[121:120]	-	
Data read access-time 1	TAAC	8	R	[119:112]	0X27	
Data read access-time 2 in CLK cycles (NSAC*100)	NSAC	8	R	[111:104]	0X01	
Max. bus clock frequency	TRAN_SPEED	8	R	[103:96]	0X32	
Card command classes	CCC	12	R	[95:84]	0X8F5	
Max. read data block length	READ_BL_LEN	4	R	[83:80]	0X09	
Partial blocks for read allowed	READ_BL_PARTIAL	1	R	[79:79]	0X00	
Write block misalignment	WRITE_BLK_MISALIGN	1	R	[78:78]	0X00	
Read block misalignment	READ_BLK_MISALIGN	1	R	[77:77]	0X00	
DSR implemented	DSR_IMP	1	R	[76:76]	0X00	
Reserved		2	R	[75:74]	-	
Device size	C_SIZE	12	R	[73:62]	0FFF	
Max. read current @ VDD min	VDD_R_CURR_MIN	3	R	[61:59]	0X07	
Max. read current @ VDD max	VDD_R_CURR_MAX	3	R	[58:56]	0X07	
Max. write current @ VDD min	VDD_W_CURR_MIN	3	R	[55:53]	0X07	
Max. write current @ VDD max	VDD_W_CURR_MAX	3	R	[52:50]	0X07	
Device size multiplier	C_SIZE_MULT	3	R	[49:47]	0X07	
Erase group size	ERASE_GRP_SIZE	5	R	[46:42]	0X1F	
Erase group size multiplier	ERASE_GRP_MULT	5	R	[41:37]	0X1F	
Write protect group size	WP_GRP_SIZE	5	R	[36:32]	32GB : 0x07	
Write protect group enable	WP_GRP_ENABLE	1	R	[31:31]	0X01	
Manufacturer default ECC	DEFAULT_ECC	2	R	[30:29]	0X00	
Write speed factor	R2W_FACTOR	3	R	[28:26]	0X02	

[Table 19]CSD fields

Name	Field	Width	Cell type	CSD slice	CSD value	Remark
Max. write data block length	WRITE_BL_LEN	4	R	[25:22]	0X09	
Partial blocks for write allowed	WRITE_BL_PARTIAL	1	R	[21:21]	0X00	
Reserved		4	R	[20:17]	-	
Content protection application	CONTENT_PROT_APP	1	R	[16:16]	0X00	
File format group	FILE_FORMAT_GRP	1	R/W	[15:15]	0X00	
Copy flag (OTP)	COPY	1	R/W	[14:14]	0X00	
Permanent write protection	PERM_WRITE_PROTECT	1	R/W	[13:13]	0X00	
Temporary write protection	TMP_WRITE_PROTECT	1	R/W/E	[12:12]	0X00	
File format	FILE_FORMAT	2	R/W	[11:10]	0X00	
ECC code	ECC	2	R/W/E	[9:8]	0X00	
CRC	CRC	7	R/W/E	[7:1]	-	
Not used, always '1'		1		[0:0]	0X01	

[Table 19] CSD fields (continued)

The following sections describe the CSD fields and the relevant data types. If not explicitly defined otherwise, all bit strings are interpreted as binary coded numbers starting with the left bit first.

7.4 Extended CSD register

The Extended CSD register defines eMMC properties and selected modes. It is 512 bytes long. The most significant 320 bytes are the Properties segment, which defines eMMC capabilities and cannot be modified by the host. The lower 192 bytes are the modes segment, which defines the configuration eMMC is working in. These modes can be changed by the host by means of the switch command.

Name	Field	CSD slice	Cell Type	CSD value	Remark
Properties segment					
Reserved		[511:506]			
Extended Security Commands Error	EXT_SECURITY_ERR	[505]	R	0x00	
Supported command sets	S_CMD_SET	[504]	R	0x01	
HPI features	HPI_FEATURES	[503]	R	0x01	
Background operations support	BKOPS_SUPPORT	[502]	R	0x01	
Max packed read commands	MAX_PACKED_READS	[501]	R	0x3F	
Max packed write commands	MAX_PACKED_WRITES	[500]	R	0x3F	
Data Tag Support	DATA_TAG_SUPPORT	[499]	R	0x01	
Tag Unit Size	TAG_UNIT_SIZE	[498]	R	0x00	
Tag Resources Size	TAG_RES_SIZE	[497]	R	0x00	
Context management capabilities	CONTEXT_CAPABILITIES	[496]	R	0x78	
Large Unit size	LARGE_UNIT_SIZE_M1	[495]	R	0x01	
Extended partitions attribute support	EXT_SUPPORT	[494]	R	0x03	
Supported modes	SUPPORTED_MODES	[493]	R	0x01	
FFU features	FFU_FEATURES	[492]	R	0x00	
Operation codes timeout	OPERATION_CODE_TIME_OUT	[491]	R	0x00	
FFU Argument	FFU_ARG	[490:487]	R	0xFFFFAFFF0	
Barrier support	BARRIER_SUPPORT	[486]	R	0x01	
Reserved		[485:309]		-	
CMD Queuing Support	CMDQ_SUPPORT	[308]	R	0x1	
CMD Queuing Depth	CMDQ_DEPTH	[307]	R	0x1F	
Reserved		[306]		-	
Number of FW sectors correctly programmed	NUMBER_OF_FW_SECTORS_CORRECTLY_PROGRAMMED	[305:302]	R	0x00000000	
Vendor proprietary health report	VENDOR_PROPRIETARY_HEALTH_REPORT	[301:270]	R	-	None

[Table 20] Extended CSD

Name	Field	CSD slice	Cell Type	CSD value	Remark
Device life time estimation type B	DEVICE_LIFE_TIME_EST_TYP_B	[269]	R	0x01	
Device life time estimation type A	DEVICE_LIFE_TIME_EST_TYP_A	[268]	R	0x01	
Pre EOL information	PRE_EOL_INFO	[267]	R	0x01	
Optimal read size	OPTIMAL_READ_SIZE	[266]	R	0x40	
Optimal write size	OPTIMAL_WRITE_SIZE	[265]	R	0x40	
Optimal trim unit size	OPTIMAL_TRIM_UNIT_SIZE	[264]	R	0x07	
Device version	DEVICE_VERSION	[263:262]	R	-	Not fixed
Firmware version	FIRMWARE_VERSION	[261:254]	R	-	Same to PRV
Power class for 200MHz, DDR at Vcc=3.6V	PWR_CL_DDR_200_360	[253]	R	0x22	
Cache size	CACHE_SIZE	[252:249]	R	0x0000400	
Generic CMD6 timeout	GENERIC_CMD6_TIME	[248]	R	0x05	
Power off Notification (long) timeout	POWER_OFF_LONG_TIME	[247]	R	0x64	
Background operations status	BKOPS_STATUS	[246]	R	0x00	
Number of correctly programmed sectors	CORRECTLY_PRG_SECTORS_NUM	[245:242]	R	-	Not fixed
1st initialization time after partitioning	INI_TIMEOUT_AP	[241]	R	0x0A	
Cache Flushing Policy	CACHE_FLUSH_POLICY	[240]	R	0x01	
Power class for 52MHz, DDR at Vcc=3.6V	PWR_CL_DDR_52_360	[239]	R	0x11	
Power class for 52MHz, DDR at Vcc=1.95V	PWR_CL_DDR_52_195	[238]	R	0x00	Not support
Power class for 200MHz at Vccq=1.95, Vcc=3.6V	PWR_CL_200_195	[237]	R	0x22	
Power class for 200MHz at Vccq=1.3, Vcc=3.6V	PWR_CL_200_130	[236]	R	0x00	Not support
Minimum write performance for 8bit at 52MHz in DDR mode	MIN_PERF_DDR_W_8_52	[235]	R	0x78	
Minimum read performance for 8bit at 52MHz in DDR mode	MIN_PERF_DDR_R_8_52	[234]	R	0x8C	
Reserved		[233]			
TRIM multiplier	TRIM_MULT	[232]	R	0x01	
Secure feature support	SEC_FEATURE_SUPPORT	[231]	R	0x55	
Secure erase multiplier	SEC_ERASE_MULT	[230]	R	32GB : 0x64	
Secure TRIM multiplier	SEC_TRIM_MULT	[229]	R	0x0A	

[Table 20] Extended CSD(Continued)

Name	Field	CSD slice	Cell Type	CSD value	Remark
Boot information	BOOT_INFO	[228]	R	0x07	
Reserved		[227]		-	
Boot partition size	BOOT_SIZE_MULTI	[226]	R	0x20	
Access size	ACC_SIZE	[225]	R	0x06	
High-capacity erase unit size	HC_ERASE_GRP_SIZE	[224]	R	0x01	
High-capacity erase timeout	ERASE_TIMEOUT_MULT	[223]	R	0x02	
Reliable write sector count	REL_WR_SEC_C	[222]	R	0x01	
High-capacity write protect group size	HC_WP_GRP_SIZE	[221]	R	32GB : 0x08	
Sleep current(Vcc)	S_C_Vcc	[220]	R	0x07	
Sleep current(Vccq)	S_C_Vccq	[219]	R	0x07	
Production state awareness timeout	PRODUCTION_STATE_AWARENESS_TIMEOUT	[218]	R	0x00	
Sleep/awake timeout	S_A_TIMEOUT	[217]	R	0x11	
Sleep Notification Timeout	SLEEP_NOTIFICATION_TIME	[216]	R	0x0C	
Sector count	SEC_COUNT	[215:212]	R	32GB: 0X03A3E000	
Secure Write Protect information	SECURE_WP_INFO	[211]	R	0x01	
Minimum write performance for 8bit at 52MHz	MIN_PERF_W_8_52	[210]	R	0x8C	
Minimum read performance for 8bit at 52MHz	MIN_PERF_R_8_52	[209]	R	0x8C	
Minimum write performance for 8bit at 26MHz, for 4bit at 52MHz	MIN_PERF_W_8_26_4_52	[208]	R	0x46	
Minimum read performance for 8bit at 26MHz, for 4bit at 52MHz	MIN_PERF_R_8_26_4_52	[207]	R	0x46	
Minimum write performance for 4bit at 26MHz	MIN_PERF_W_4_26	[206]	R	0x1E	
Minimum read performance for 4bit at 26MHz	MIN_PERF_R_4_26	[205]	R	0x1E	
Reserved		[204]			
Power class for 26MHz at 3.6V	PWR_CL_26_360	[203]	R	0x00	
Power class for 52MHz at 3.6V	PWR_CL_52_360	[202]	R	0x00	
Power class for 26MHz at 1.95V	PWR_CL_26_195	[201]	R	0x00	Not support
Power class for 52MHz at 1.95V	PWR_CL_52_195	[200]	R	0x00	Not support

[Table 20] Extended CSD(Continued)

Name	Field	CSD slice	Cell Type	CSD value	Remark
Partition switching timing	PARTITION_SWITCH_TIME	[199]	R	0x01	
Out-of-interrupt busy timing	OUT_OF_INTERRUPT_TIME	[198]	R	0x05	
I/O Driver Strength	DRIVER_STRENGTH	[197]	R	0x1F	
Device type	DEVICE_TYPE	[196]	R	0x57	
Reserved		[195]		-	
CSD structure	CSD_STRUCTURE	[194]	R	0x02	
Reserved		[193]		-	
Extended CSD revision	EXT_CSD_REV	[192]	R	0x08	
Modes Segment					
Command set	CMD_SET	[191]	R/W/E_P	0x00	
Reserved		[190]		-	
Command set revision	CMD_SET_REV	[189]	R	0x00	
Reserved		[188]		-	
Power class	POWER_CLASS	[187]	R/W/E_P	0x00	
Reserved		[186]		-	
High-speed interface timing	HS_TIMING	[185]	R/W/E_P	0x00	
Strobe Support	STROBE_SUPPORT	[184]	R	0x01	
Bus width mode	BUS_WIDTH	[183]	W/E_P	0x00	
Reserved		[182]		-	
Erased memory content	ERASED_MEM_CONT	[181]	R	0x00	
Reserved		[180]		-	
Partition configuration	PARTITION_CONFIG	[179]	R/W/E & R/W/E_P	0x00	
Boot config protection	BOOT_CONFIG_PROT	[178]	R/W/E & R/W/C_P	0x00	
Boot bus conditions	BOOT_BUS_CONDITIONS	[177]	R/W/E	0x00	
Reserved		[176]		-	
High-density erase group definition	ERASE_GROUP_DEF	[175]	R/W/E_P	0x00	
Boot write protection status registers	BOOT_WP_STATUS	[174]	R	0x00	
Boot area write protection register	BOOT_WP	[173]	R/W/E & R/W/C_P	0x00	
Reserved		[172]		-	

[Table 20] Extended CSD(Continued)

Name	Field	CSD slice	Cell Type	CSD value	Remark
User area write protection register	USER_WP	[171]	R/W, R/W/ C_P & /W/ E_P	0x00	
Reserved		[170]		-	
FW configuration	FW_CONFIG	[169]	R/W	0x00	
RPMB Size	RPMB_SIZE_MULT	[168]	R	0x20	
Write reliability setting register	WR_REL_SET	[167]	R/W	0x1F	
Write reliability parameter register	WR_REL_PARAM	[166]	R	0x15	
Sanitize start	SANITIZE_START	[165]	W/E_P	0x00	
Manually start background operations	BKOPS_START	[164]	W/E_P	0x00	
Enable background operations Handshake	BKOPS_EN	[163]	R/W	0x00	
H/W reset function	RST_n_FUNCTION	[162]	R/W	0x00	
HPI management	HPI_MGMT	[161]	R/W/E_P	0x00	
Partitioning support	PARTITIONING_SUPPORT	[160]	R	0x07	
Max enhanced area size	MAX_ENH_SIZE_MULT	[159:157]	R	32GB: 0X000E8F	
Partitions attribute	PARTITIONS_ATTRIBUTE	[156]	R/W	0x00	
Partitioning setting	PARTITION_SETTING_COMPLETED	[155]	R/W	0x00	
General purpose partition size	GP_SIZE_MULT	[154:143]	R/W	0x00	
Enhanced user data area size	ENH_SIZE_MULT	[142:140]	R/W	0x00	
Enhanced user data start address	ENH_START_ADDR	[139:136]	R/W	0x00	
Reserved		[135]		-	
Bad Block management mode	SEC_BAD_BLK_MGMNT	[134]	R/W	0x00	
Production state awareness	PRODUCTION_STATE_AWARENESS	[133]	R/W/E	-	
Package Case Temperature is Controlled	TCASE_SUPPORT	[132]	W/E_P	0x00	
Periodic Wake-up	PERIODIC_WAKEUP	[131]	R/W/E	0x00	
Program CID/CSD in DDR mode support	PROGRAM_CID_CSD_DDR_SUPPORT	[130]	R	0x00	
Reserved		[129:128]		-	
Vendor Specific Fields	VENDOR_SPECIFIC_FIELD	[127:64]	Vendor specific	-	

[Table 20] Extended CSD(Continued)

Name	Field	CSD slice	Cell Type	CSD value	Remark
Native sector size	NATIVE_SECTOR_SIZE	[63]	R	0x01	
Sector size emulation	USE_NATIVE_SECTOR	[62]	R/W	0x00	
Sector size	DATA_SECTOR_SIZE	[61]	R	0x00	
1st initialization after disabling sector size emulation	INI_TIMEOUT_EMU	[60]	R	0x0A	
Class 6 commands control	Class6_CTRL	[59]	R/W/E_P	0x00	
Number of addressed group to be Released	DYNCAP_NEEDED	[58]	R	0x00	
Exception events control	EXCEPTION_EVENTS_CTRL	[57:56]	R/W/E_P	0x00	
Exception events status	EXCEPTION_EVENTS_STATUS	[55:54]	R	0x0000	
Extended Partitions Attribute	EXT_PARTITIONS_ATTRIBUTE	[53:52]	R/W	0x0000	
Context configuration	CONTEXT_CONF	[51:37]	R/W/E_P	0x00...00	
Packed command status	PACKED_COMMAND_STATUS	[36]	R	0x00	
Packed command failure index	PACKED_FAILURE_INDEX	[35]	R	0x00	
Power Off Notification	POWER_OFF_NOTIFICATION	[34]	R/W/E_P	0x00	
Control to turn the Cache ON/OFF	CACHE_CTRL	[33]	R/W/E_P	0x00	
Flushing of the cache	FLUSH_CACHE	[32]	W/E_P	0x00	
Control to turn the barrier ON/OFF	BARRIER_CONTROL	[31]	R/W	0x00	
Mode config	MODE_CONFIG	[30]	R/W/E_P	0x00	
Mode operation codes	MODE_OPERATION_CODES	[29]	W/E_P	0x00	
Reserved		[28:27]		-	
FFU Status	FFU_STATUS	[26]	R	0x00	
Pre loading data size	PRE_LOADING_DATA_SIZE	[25:22]	R/W/E_P	0X00000000	
Max pre loading data size	MAX_PRE_LOADING_DATA_SIZE	[21:18]	R	32GB: 0X03A3E000	
Product state awareness enablement	PRODUCT_STATE_AWARENESS_ENABLEMENT	[17]	R/W/E & R	0x01	
Secure Removal Type	SECURE_REMOVAL_TYPE	[16]	R/W & R	0x3B	
Command Queue Mode Enable	CMDQ_MODE_EN	[15]	R/W/E_P	0x00	
Reserved		[14:0]		-	

[Table 20] Extended CSD(Continued)

- Reserved bits should read as "0"
- Obsolete values should be don't care

7.5 RCA (Relative card address)

The writable 16-bit relative card address (RCA) register carries the card address assigned by the host during the card identification. This address is used for the addressed host-card communication after the card identification procedure. The default value of the RCA register is 0x0001. The value 0x0000 is reserved to set all cards into the Stand-by State with CMD7.

7.6 DSR (Driver stage register)

It can be optionally used to improve the bus performance for extended operating conditions (depending on parameters like bus length, transfer rate or number of Devices). The CSD register carries the information about the DSR register usage. The default value of the DSR register is 0x404.

8Gb LPDDR3 SDRAM

Input/Output Functional Description

SYMBOL	TYPE	DESCRIPTION
CK_t, CK_c	Input	Clock: CK_t and CK_c are differential clock inputs. All Double Data Rate (DDR) CA inputs are sampled on both positive and negative edge of CK_t. Single Data Rate (SDR) inputs, CS_n and CKE, are sampled at the positive Clock edge. Clock is defined as the differential pair, CK_t and CK_c. The positive Clock edge is defined by the crosspoint of a rising CK_t and a falling CK_c. The negative Clock edge is defined by the crosspoint of a falling CK_t and a rising CK_c.
CKE	Input	Clock Enable: CKE HIGH activates and CKE LOW deactivates internal clock signals and therefore device input buffers and output drivers. Power savings modes are entered and exited through CKE transitions. CKE is considered part of the command code. CKE is sampled at the positive Clock edge.
CS_n	Input	Chip Select: CS_n is considered part of the command code. CS_n is sampled at the positive Clock edge.
CA0 - CA9	Input	DDR Command/Address Inputs: Uni-directional command/address bus inputs. CA is considered part of the command code.
DQ0 - DQ15 (x16) DQ0 - DQ31 (x32)	I/O	Data Input/Output: Bi-directional data bus
DQS0_t, DQS1_t, DQS0_c, DQS1_c (x16) DQS0_t - DQS3_t, DQS0_c - DQS3_c (x32)	I/O	Data Strobe (Bi-directional, Differential): The data strobe is bi-directional (used for read and write data) and differential (DQS_t and DQS_c). It is output with read data and input with write data. DQS_t is edge-aligned to read data and centered with write data. For x16, DQS0_t and DQS0_c correspond to the data on DQ0 - DQ7; DQS1_t and DQS1_c to the data on DQ8 - DQ15. For x32 DQS0_t and DQS0_c correspond to the data on DQ0 - DQ7, DQS1_t and DQS1_c to the data on DQ8 - DQ15, DQS2_t and DQS2_c to the data on DQ16 - DQ23, DQS3_t and DQS3_c to the data on DQ24 - DQ31.
DM0-DM1 (x16) DM0-DM3 (x32)	Input	Input Data Mask: DM is the input mask signal for write data. Input data is masked when DM is sampled HIGH coincident with that input data during a Write access. DM is sampled on both edges of DQS_t. Although DM is for input only, the DM loading shall match the DQ and DQS_t (or DQS_c). For x16 and x32 devices, DM0 is the input data mask signal for the data on DQ0-7. DM1 is the input data mask signal for the data on DQ8-15. For x32 devices, DM2 is the input data mask signal for the data on DQ16-23 and DM3 is the input data mask signal for the data on DQ24-31.
ODT	Input	On-Die Termination: This signal enables and disables termination on the DRAM DQ bus according to the specified mode register settings.
VDD1	Supply	Core Power Supply 1
VDD2	Supply	Core Power Supply 2
VDDCA	Supply	Input Receiver Power Supply: Power for CA0-9, CKE, CS_n, CK_t and CK_c input buffers.
VDDQ	Supply	I/O Power Supply: Power supply for data input/output buffers.
VREFCA	Supply	Reference Voltage for CA Command and Control Input Receiver: Reference voltage for all CA0-9, CKE, CS_n, CK_t and CK_c input buffers.
VREFDQ	Supply	Reference Voltage for DQ Input Receiver: Reference voltage for all Data input buffers.
VSS	Supply	Ground
VSSCA	Supply	Ground for Input Receivers
VSSQ	Supply	I/O Ground: Ground for data input/output buffers
ZQ	I/O	Reference Pin for Output Drive Strength Calibration

Functional Description

LPDDR3-SDRAM is a high-speed synchronous DRAM device internally configured as an 8-bank memory.

These devices contain the following number of bits:

8 Gb has 8,589,934,592 bits

LPDDR3 devices use a double data rate architecture on the Command/Address (CA) bus to reduce the number of input pins in the system. The 10-bit CA bus contains command, address, and bank information. Each command uses one clock cycle, during which command information is transferred on both the positive and negative edge of the clock.

These devices also use a double data rate architecture on the DQ pins to achieve high speed operation. The double data rate architecture is essentially an 8n prefetch architecture with an interface designed to transfer two data bits per DQ every clock cycle at the I/O pins. A single read or write access for the LPDDR3 SDRAM effectively consists of a single 8n-bit wide, one clock cycle data transfer at the internal DRAM core and eight corresponding n-bit wide, one-half-clock-cycle data transfers at the I/O pins.

Read and write accesses to the LPDDR3 SDRAMs are burst oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an Activate command, which is then followed by a Read or Write command. The address and BA bits registered coincident with the Activate command are used to select the row and the bank to be accessed. The address bits registered coincident with the Read or Write command are used to select the bank and the starting column location for the burst access.

Prior to normal operation, the LPDDR3 SDRAM must be initialized. The following section provides detailed information covering device initialization, register definition, command description and device operation.

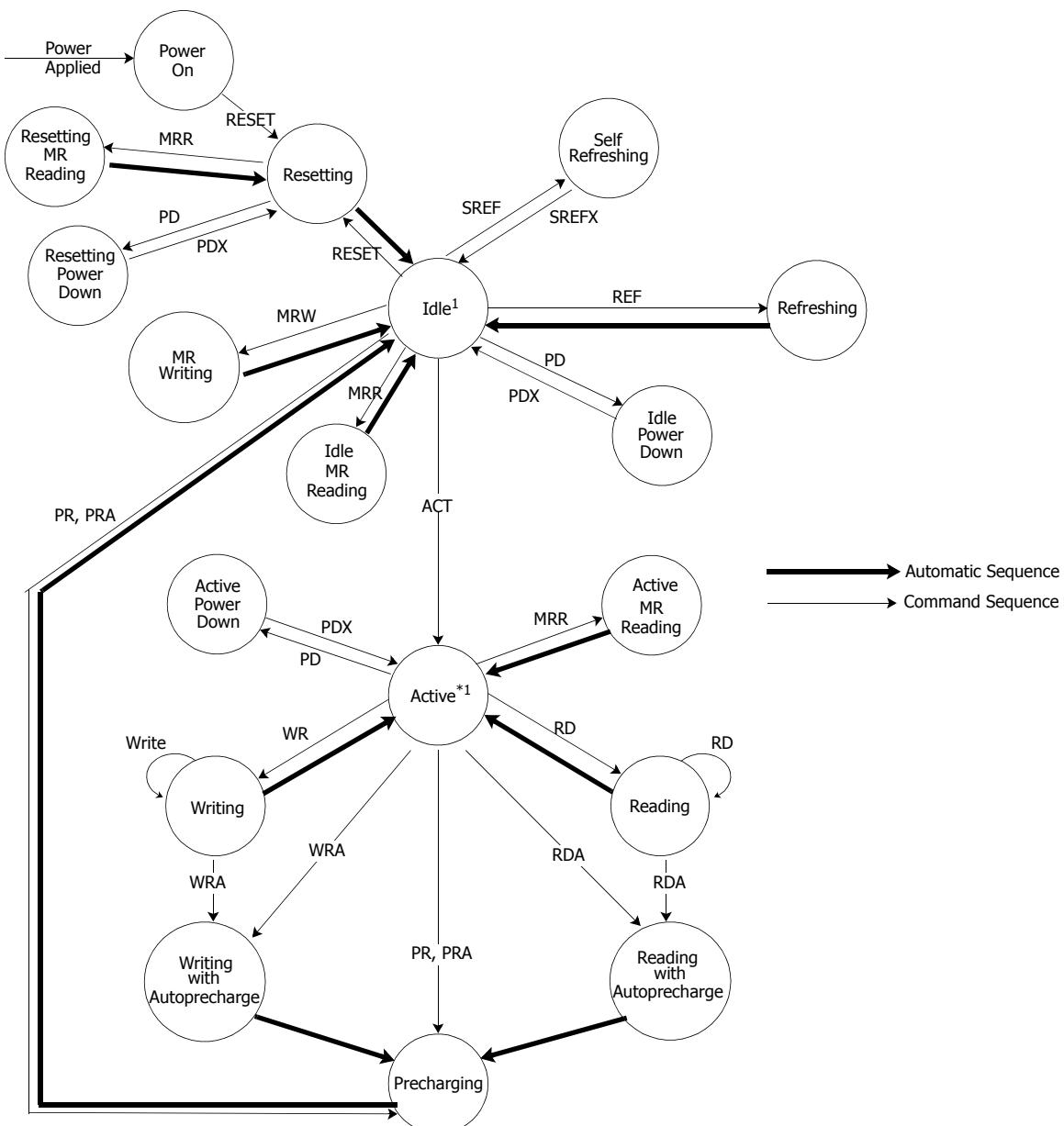
LPDDR3 SDRAM Addressing

Density		8Gb
Number of Banks		8
Bank Addresses		BA0 - BA2
$t_{REFI}(\mu s)^2$		3.9
x16	Row Addresses	R0 - R14
	Column Addresses ¹	C0 - C10
x32	Row Addresses	R0 - R14
	Column Addresses ¹	C0 - C9

Note:

1. The least-significant column address C0 is not transmitted on the CA bus, and is implied to be zero.
2. tREFI values for all bank refresh is $T_c = -30 \sim 85^{\circ}\text{C}$, T_c means Operating Case Temperature.
3. Row and Column Address values on the CA bus which are not used are "don't care".
4. No memory present at addresses with R13=R14=HIGH. ACT command with R13=R14=HIGH is ignored (NOP). Write to R13=R14=HIGH is ignored (NOP).

STATE DIAGRAM



PR(A) = Precharge (All)

PD = Enter Power Down

ACT = Activate

PDX = Exit Power Down

WR(A) = Write (with Autoprecharge)

SREF = Enter Self Refresh

RD(A) = Read (with Autoprecharge)

SREFX = Exit Self Refresh

RESET = Reset is achieved through MRW command

REF = Refresh

MRW = Mode Register Write

MRR = Mode Register Read

Note:

1. In the Idle state, all banks are precharged.
2. In the case of MRW to enter CA Training mode or Write Leveling Mode, the state machine will not automatically return to the Idle state. In these cases an additional MRW command is required to exit either operating mode and return to the Idle state. See sections "CA Training" or "Write Leveling".
3. Terminated bursts are not allowed. For these state transitions, the burst operation must be completed before the transition can occur.
4. Use caution with this diagram. It is intended to provide a floorplan of the possible state transitions and commands to control them, not all details. In particular, situations involving more than one bank are not captured in full detail.

Power-up, Initialization and Power-off

Voltage Ramp and Device Initialization

The following sequence must be used to power up the device. Unless specified otherwise, this procedure is mandatory.

1. Voltage Ramp

While applying power (after Ta), CKE must be held LOW ($\leq 0.2 \times VDDCA$), and all other inputs must be between VILmin and VIHmax. The device outputs remain at High-Z while CKE is held LOW.

Following the completion of the voltage ramp (Tb), CKE must be maintained LOW. DQ, DM, DQS_t and DQS_c voltage levels must be between VSSQ and VDDQ during voltage ramp to avoid latchup. CK_t, CK_c, CS_n, and CA input levels must be between VSSCA and VDDCA during voltage ramp to avoid latch-up. Voltage ramp power supply requirements are provided in the table "Voltage Ramp Conditions".

Table. Voltage Ramp Conditions

After...	Applicable Conditions
Ta is reached	VDD1 must be greater than VDD2-200mV.
	VDD1 and VDD2 must be greater than VDDCA-200mV.
	VDD1 and VDD2 must be greater than VDDQ-200mV.
	VREF must always be less than all other supply voltages.

Note:

1. Ta is the point when any power supply first reaches 300mV.
2. Noted conditions apply between Ta and power-off (controlled or uncontrolled).
3. Tb is the point at which all supply and reference voltages are within their defined operating ranges.
4. Power ramp duration tINIT0 (Tb - Ta) must not exceed 20ms.
5. The voltage difference between any of VSS, VSSQ, and VSSCA pins must not exceed 100mV.

Beginning at Tb, CKE must remain LOW for at least tINIT1, after which CKE can be asserted HIGH. The clock must be stable at least tINIT2 prior to the first CKE LOW-to-HIGH transition (Tc). CKE, CS_n, and CA inputs must observe setup and hold requirements (tIS, tIH) with respect to the first rising clock edge (as well as to subsequent falling and rising edges).

If any MRR commands are issued, the clock period must be within the range defined for tCKb. MRW commands can be issued at normal clock frequencies as long as all AC timings are met. Some AC parameters (for example, tDQSCK) could have relaxed timings (such as tDQSCKb) before the system is appropriately configured. While keeping CKE HIGH, NOP commands must be issued for at least tINIT3 (Td). The ODT input signal may be in undefined state until tIS before CKE is registered HIGH. When CKE is registered HIGH, the ODT input signal shall be statically held at either LOW or HIGH. The ODT input signal remains static until the power up initialization sequence is finished, including the expiration of tZQINIT.

2. Reset Command

After tINIT3 is satisfied, the MRW RESET command must be issued (Td).

An optional PRECHARGE ALL command can be issued prior to the MRW RESET command. Wait at least tINIT4 while keeping CKE asserted and issuing NOP commands. Only NOP commands are allowed during time tINIT4.

3. MRRs and Device Auto Initialization (DAI) Polling

After tINIT4 is satisfied (T_e), only MRR commands and power-down entry/exit commands are supported. After T_e , CKE can go LOW in alignment with power-down entry and exit specifications. MRR commands are only valid at this time if the CA bus does not need to be trained. may only begin after time Tf. **User may issue MRR command to poll the DAI bit which will indicate if device auto initialization is complete; once DAI bit indicates completion, SDRAM is in idle state. Device will also be in idle state after tINIT5(max) has expired (whether or not DAI bit has been read by MRR command).** As the memory output buffers are not properly configured by T_e , some AC parameters must have relaxed timings before the system is appropriately configured.

After the DAI bit (MR0, DAI) is set to zero by the memory device (DAI complete), the device is in the idle state (Tf). DAI status can be determined by issuing the MRR command to MR0. The device sets the DAI bit no later than tINIT5 after the RESET command. The controller must wait at least tINIT5 or until the DAI bit is set before proceeding.

4. ZQ Calibration

If CA Training is not required, the MRW initialization calibration (ZQ_CAL) command can be issued to the memory (MR10) after time Tf. If CA Training is required, the CA Training may begin at time Tf. See the section of "Mode Register Write - CA Training Mode" for the CA Training command. No other CA commands (other than RESET or NOP) may be issued prior to the completion of CA Training. At the completion of CA Training (Tf'), the MRW initialization calibration (ZQ_CAL) command can be issued to the memory (MR10).

This command is used to calibrate output impedance over process, voltage, and temperature. In systems where more than one LPDDR3 device exists on the same bus, the controller must not overlap MRW ZQ_CAL commands. The device is ready for normal operation after tZQINIT.

5. Normal Operation

After tZQINIT (T_g), MRW commands must be used to properly configure the memory (for example the output buffer drive strength, latencies, etc.). Specifically, MR1, MR2, and MR3 must be set to configure the memory for the target frequency and memory configuration.

After the initialization sequence is complete, the device is ready for any valid command. After T_g , the clock frequency can be changed using the procedure described in the LPDDR3 specification.

Table. Timing Parameters for initialization

Symbol	Parameter	Value		Unit
		min	max	
tINIT0	Maximum Voltage Ramp Time	-	20	ms
tINIT1	Minimum CKE low time after completion of voltage ramp	100	-	ns
tINIT2	Minimum stable clock before first CKE high	5	-	tCK
tINIT3	Minimum idle time after first CKE assertion	200	-	us
tINIT4	Minimum idle time after Reset command	1	-	us
tINIT5	Maximum duration of Device Auto-Initialization	-	10	us
tZQINIT	ZQ Initial Calibration for LPDDR3 devices	1	-	us
tCKb	Clock cycle time during boot	18	100	ns

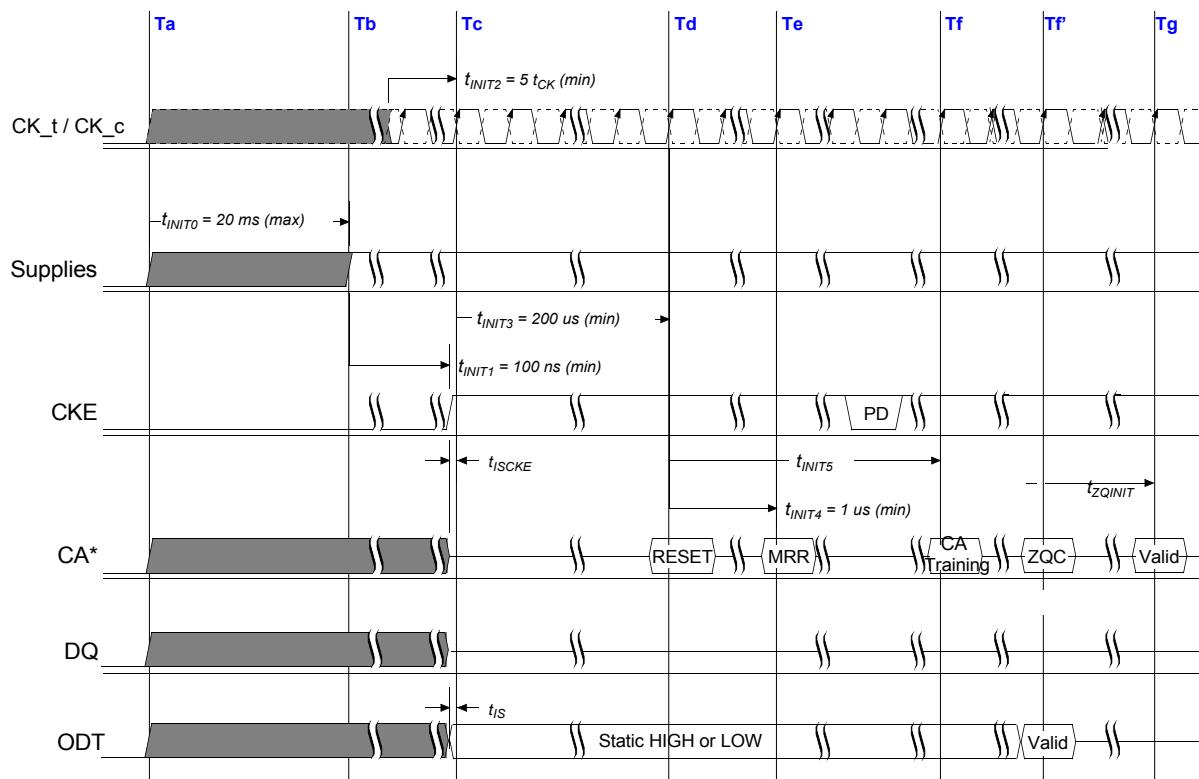


Figure. Power Ramp and Initialization Sequence

Notes

1. High-Z on the CA bus indicates NOP.
2. For tINIT values, see the table "Timing Parameters for Initialization".
3. After RESET command (time Te), RTT is disabled until ODT function is enabled by MRW to MR11 following Tg.
4. CA Training is optional.

Initialization After Reset (without Power ramp)

If the RESET command is issued before or after the power-up initialization sequence, the re-initialization procedure must begin at Td.

Power-off Sequence

The following procedure is required to power off the device.

While powering off, CKE must be held LOW ($\leq 0.2 \times VDDCA$); all other inputs must be between VILmin and VIHmax. The device outputs remain at High-Z while CKE is held LOW.

DQ, DM, DQS_t, and DQS_c voltage levels must be between VSSQ and VDDQ during the power-off sequence to avoid latch-up. CK_t, CK_c, CS_n, and CA input levels must be between VSSCA and VDDCA during the power-off sequence to avoid latch-up.

Tx is the point where any power supply drops below the minimum value specified.

Tz is the point where all power supplies are below 300mV. After Tz, the device is powered off (see the table "Power Supply Conditions").

Table. Power Supply Conditions

Between...	Applicable Conditions
Tx and Tz	VDD1 must be greater than VDD2—200mV
	VDD1 must be greater than VDDCA—200mV
	VDD1 must be greater than VDDQ—200mV
	VREF must always be less than all other supply voltages

The voltage difference between any of VSS, VSSQ, and VSSCA pins must not exceed 100mV.

Uncontrolled Power-Off Sequence

When an uncontrolled power-off occurs, the following conditions must be met:

At Tx, when the power supply drops below the minimum values specified, all power supplies must be turned off and all power-supply current capacity must be at zero, except for any static charge remaining in the system.

After Tz (the point at which all power supplies first reach 300mV), the device must power off. The time between Tx and Tz must not exceed 10ms. During this period, the relative voltage between power supplies is uncontrolled. VDD1 and VDD2 must decrease with a slope lower than 0.5 V/μs between Tx and Tz.

An uncontrolled power-off sequence can occur a maximum of 400 times over the life of the device.

Table. Power-Off Timing

Symbol	Parameter	Value		Unit
		min	max	
tPOFF	Maximum power-off ramp time	-	2	sec

Mode Register Definition

Table below shows the mode registers for LPDDR3 SDRAM. Each register is denoted as "R" if it can be read but not written, "W" if it can be written but not read, and "R/W" if it can be read and written. A Mode Register Read command shall be used to read a mode register. A Mode Register Write command shall be used to write a mode register.

Table. Mode Register Assignment

MR #	MA <7:0>	Function	Access	OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0	Link
0	00H	Device Info.	R	RL3	WL set B	(RFU)	RZQI (Optional)	(RFU)	(RFU)	DAI	go to MR0	
1	01H	Device Feature1	W	nWR (for AP)		(RFU)	BT		BL		go to MR1	
2	02H	Device Feature 2	W	WR Lev	WL Select	(RFU)	nWRE		RL & WL		go to MR2	
3	03H	I/O Config-1	W		(RFU)			DS			go to MR3	
4	04H	Device Temperature	R	TUF		(RFU)		Refresh Rate			go to MR4	
5	05H	Basic Config-1	R				Manufacturer ID				go to MR5	
6	06H	Basic Config-2	R				Revision ID1				go to MR6	
7	07H	Basic Config-3	R				Revision ID2				go to MR7	
8	08H	Basic Config-4	R	I/O width		Density		Type			go to MR8	
9	09H	Test Mode	W				Vendor-Specific Test Mode				go to MR9	
10	0AH	Calibration	W				Calibration Code				go to MR10	
11	0BH	ODT	W			(RFU)		PD CTL	DQ ODT		go to MR11	
16	10H	PASR_Bank	W				PASR Bank Mask				go to MR16	
17	11H	PASR_Segment	W				PASR Segment Mask				go to MR17	
32	20H	DQ Calibration Pattern A	R				See the section "DQ Calibration"				go to MR32	
40	28H	DQ Calibration Pattern B	R				See the section "DQ Calibration"				go to MR40	
41	29H	CA Training Entry for CA0-3, CA5-8	W				See the section "Mode Register Write - CA Training Mode"				go to MR41	
42	2AH	CA Training Exit	W				See the section "Mode Register Write - CA Training Mode"				go to MR42	
48	30H	CA Training Entry for CA4, 9	W				See the section "Mode Register Write - CA Training Mode"				go to MR48	
63	3FH	Reset	W				X				go to MR63	

Note:

1. RFU bits shall be set to `0' during Mode Register writes.
2. RFU bits shall be read as `0' during Mode Register reads.
3. All Mode Registers that are specified as RFU or write-only shall return undefined data when read and DQS_t, DQS_c shall be toggled.
4. All Mode Registers that are specified as RFU shall not be written.
5. Writes to read-only registers shall have no impacts on the functionality of the device.

MRO Device Information (MA<7:0> = 00H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0		
RL3	WL (Set B) Support	(RFU)	RZQI (Optional)			(RFU)			
DAI (Device Auto-Initialization Status)			Read-only	OP0	0B: DAI complete 1B: DAI still in progress				
RZQI (Built in Self Test for RZQ Information)			Read-only	OP4:OP3	00B: RZQ self test not supported 01B: ZQ-pin may connect to VDDCA or float 10B: ZQ-pin may short to GND 11B: ZQ-pin self test completed, no error condition detected (ZQ-pin may not connect to VDD or float nor short to GND)				
WL (Set B) Support			Read-only	OP<6>	0B: DRAM does not support WL (Set B) 1B: DRAM supports WL (Set B)				
RL3 Option Support			Read-only	OP<7>	0B : DRAM does not support RL=3, nWR=3, WL=1 1B : DRAM supports RL=3, nWR=3, WL=1 for frequencies <=166				

Note:

1. RZQI, if supported, will be set upon completion of the MRW ZQ Initialization Calibration command.
2. If ZQ is connected to VDDCA to set default calibration, OP[4:3] shall be set to 01. If ZQ is not connected to VDDCA, either OP[4:3]=01 or OP[4:3]=10 might indicate a ZQ-pin assembly error. It is recommended that the assembly error is corrected.
3. In the case of possible assembly error (either OP[4:3]=01 or OP[4:3]=10 per Note 4), the LPDDR3 device will default to factory trim settings for RON, and will ignore ZQ calibration commands. In either case, the system may not function as intended.
4. In the case of the ZQ self-test returning a value of 11b, this result indicates that the device has detected a resistor connection to the ZQ pin. However, this result cannot be used to validate the ZQ resistor value or that the ZQ resistor tolerance meets the specified limits (i.e. 240-ohm +/-1%).

MR1 Device Feature 1 (MA<7:0> = 01H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
nWR (for AP)		(RFU)		BT	BL		
BL		Write-only		OP<2:0>	011B: BL8 (default)		
					100B: Reserved		
					All others: reserved		
BT		Write-only		OP<3>	0B: Don't care		
nWR		Write-only		OP<7:5>	If nWRE (in MR2 OP4) = 0		
					001B : nWR=3 (optional)		
					100B : nWR=6		
					110B : nWR=8		
					111B : nWR=9		
					If nWRE (in MR2 OP4) = 1		
					000B : nWR=10 (default)		
					001B : nWR=11		
					010B : nWR=12		
					100B : nWR=14		
					110B : nWR=16		
					All others: reserved		

Note:

- Programmed value in nWR register is the number of clock cycles which determines when to start internal precharge operation for a write burst with AP enabled. It is determined by RU(tWR/tCK).

Table. Burst Sequence by BL and BT

C2	C1	C0	BT	BL	Burst Cycle Number and Burst Address Sequence							
					1	2	3	4	5	6	7	8
0B	0B	0B	seq	8	0	1	2	3	4	5	6	7
0B	1B	0B			2	3	4	5	6	7	0	1
1B	0B	0B			4	5	6	7	0	1	2	3
1B	1B	0B			6	7	0	1	2	3	4	5

Note:

- C0 inputs are not present on CA bus. Those are implied zero.
- For BL=8, the burst address represents C2 - C0.

MR2 Device Feature 2 (MA<7:0> = 02H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
WR Lev	WL Select	(RFU)	nWRE			RL & WL	
RL & WL	Write-only		OP<3:0>	If OP<6> =0 (WL Set A, default) 0001B: RL = 3 / WL = 1 (\leq 166 MHz, optional ¹) 0100B: RL = 6 / WL = 3 (\leq 400 MHz) 0110B: RL = 8 / WL = 4 (\leq 533 MHz) 0111B: RL = 9 / WL = 5 (\leq 600 MHz) 1000B: RL = 10 / WL = 6 (\leq 667 MHz, default) 1001B: RL = 11 / WL = 6 (\leq 733 MHz) 1010B: RL = 12 / WL = 6 (\leq 800 MHz) 1100B: RL = 14 / WL = 8 (\leq 933 MHz) 1110B: RL = 16 / WL = 8 (\leq 1066 MHz) All others: reserved			
				If OP<6> =1 (WL Set B, optional ²) 0001B: RL = 3 / WL = 1 (\leq 166 MHz, optional ¹) 0100B: RL = 6 / WL = 3 (\leq 400 MHz) 0110B: RL = 8 / WL = 4 (\leq 533 MHz) 0111B: RL = 9 / WL = 5 (\leq 600 MHz) 1000B: RL = 10 / WL = 8 (\leq 667 MHz, default) 1001B: RL = 11 / WL = 9 (\leq 733 MHz) 1010B: RL = 12 / WL = 9 (\leq 800 MHz) 1100B: RL = 14 / WL = 11 (\leq 933MHz) 1110B: RL = 16 / WL = 13 (\leq 1066MHz) All others: reserved			
nWRE	Write-only		OP<4>	0B : Enable nWR programing \leq 9 1B : Enable nWR programing $>$ 9 (default)			
WL Select	Write-only		OP<6>	0B : Select WL Set A (default) 1B : Select WL Set B (optional ²)			
Write Leveling	Write-only		OP<7>	0B : Disabled (default) 1B : Enabled			

Note:

1. See MR0, OP<7>.
2. See MR0, OP<6>

MR3 I/O Configuration 1 (MA<7:0> = 03H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
(RFU)				DS			
DS	Write-only	OP<3:0>		0000B: reserved			
				0001B: 34.3Ω typical pull-down/pull-up			
				0010B: 40Ω typical pull-down/pull-up (default)			
				0011B: 48Ω typical pull-down/pull-up			
				0100B: reserved for 60Ω typical pull-down/pull-up			
				0110B: reserved for 80Ω typical pull-down/pull-up			
				1001B: 34.3Ω typical pull-down, 40Ω Typical Pull-up (optional ¹)			
				1010B: 40Ω typical pull-down, 48Ω Typical Pull-up (optional ¹)			
				1011B: 34.3Ω typical pull-down, 48Ω Typical Pull-up (optional ¹)			
				All others: reserved			

Note:

1. Please contact us, for the supportability of the optional feature.

MR4 Device Temperature (MA<7:0> = 04H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
TUF	(RFU)				Refresh Rate		
Refresh Rate	Read-only	OP<2:0>	000B: Low temperature operating limit exceeded				
			001B: 4 x tREFI, 4 x tREFIpb, 4 x tREFW				
			010B: 2 x tREFI, 2 x tREFIpb, 2 x tREFW				
			011B: 1 x tREFI, 1 x tREFIpb, 1 x tREFW ($\leq 85^{\circ}\text{C}$)				
			100B: 1/2 x tREFI, 1/2 x tREFIpb, 1/2 x tREFW, do not de-rate AC timing				
			101B: 1/4 x tREFI, 1/4 x tREFIpb, 1/4 x tREFW, do not de-rate AC timing				
			110B: 1/4 x tREFI, 1/4 x tREFIpb, 1/4 x tREFW, de-rate AC timing				
			111B: High temperature operating limit exceeded				
Temperature Update Flag (TUF)	Read-only	OP<7>	0B: OP<2:0> value has not changed since last read of MR4 1B: OP<2:0> value has changed since last read of MR4				

Note:

1. A Mode Register Read from MR4 will reset OP7 to '0'.
2. OP7 is reset to '0' at power-up.
3. If OP2 equals '1', the device temperature is greater than 85°C .
4. OP7 is set to '1' if OP2:OP0 has changed at any time since the last read of MR4.
5. LPDDR3 might not operate properly when OP[2:0] = 000B or 111B.
6. For specified operating temperature range and maximum operating temperature refer to the section of Operating Temperature Range.
7. LPDDR3 devices shall be de-rated by adding derating values to the following core timing parameters: tRCD, tRC, tRAS, tRP and tRRD. tDQSCK shall be de-rated according to the tDQSCK de-rating in "AC timing table". Prevailing clock frequency spec and related setup and hold timings shall remain unchanged.
8. See the section of Temperature Sensor for information on the recommended frequency of reading MR4.

MR5 Basic Configuration1 (MA<7:0> = 05H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Manufacturer ID							
Company ID	Read-only	OP<7:0>	0000 0110B: Hynix Semiconductor				

MR6 Basic Configuration2 (MA<7:0> = 06H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Revision ID 1							
Revision ID1	Read-only	OP<7:0>	00000001B				

MR7 Basic Configuration3 (MA<7:0> = 07H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Revision ID 2							
Revision ID2	Read-only	OP<7:0>	00000000B: A-version				

MR8 Basic Configuration4 (MA<7:0> = 08H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0		
I/O width			Density			Type			
Type	Read-only	OP<1:0>	11B: S8		All Others : Reserved				
Density	Read-only	OP<5:2>	0110B : 4Gb		1110B : 6Gb				
			0111B : 8Gb		1101B : 12Gb				
			1000B : 16Gb		1001B : 32Gb				
			All Others : Reserved						
I/O width	Read-only	OP<7:6>	00B: x32		01B: x16				
			All Others : Reserved						

MR9 Test Mode (MA<7:0> = 09H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Vendor-specific Test Mode							

MR10 Calibration (MA<7:0> = 0AH)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Calibration Code							
Calibration Code	Write Only	OP<7:0>		1111 1111B: Calibration command after initialization			
				1010 1011B: Long Calibration			
				0101 0110B: Short Calibration			
				1100 0011B: ZQ Reset			
				others: reserved			

Note:

1. Host processor shall not write MR10 with "Reserved" values
2. LPDDR3 devices shall ignore calibration command when a "Reserved" value is written into MR10.
3. See AC timing table for the calibration latency.
4. If ZQ is connected to VSSCA through RZQ, either the ZQ calibration function (see "Mode Register Write ZQ Calibration Command") or default calibration (through the ZQRESET command) is supported. If ZQ is connected to VDDCA, the device operates with default calibration, and ZQ calibration commands are ignored. In both cases, the ZQ connection shall not change after power is applied to the device.
5. LPDDR3 devices that do not support calibration shall ignore the ZQ Calibration command.
6. Optionally, the MRW ZQ Initialization Calibration command will update MR0 to indicate RZQ pin connection.

MR11 ODT (MA<7:0> = 0BH)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
(RFU)					PD Control	DQ ODT	
DQ ODT	Write Only	OP<1:0>		00B : Disable (Default)			
				01B : RZQ/4 (See the Note 1.)			
				10B : RZQ/2			
				11B : RZQ/1			
Power Down Control	Write Only	OP<2>		0B : ODT disabled by DRAM during power down			
				1B : ODT enabled by DRAM during power down			

Note:

1. RZQ/4 shall be supported for LPDDR3-1866 devices. RZQ/4 support is optional for LPDDR3-1333 and LPDDR3-1600 devices. Consult manufacturer specifications for RZQ/4 support for LPDDR3-1333 and LPDDR3-1600.

MR12:15 (Reserved) (MA<7:0> = 0CH - 0FH)

MR16 PASR Bank Mask (MA<7:0> = 10H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Bank Mask							

Bank <7:0> Mask	Write-only	OP<7:0>	0B : refresh enable to the bank (=unmasked, default) 1B : refresh blocked (=masked)
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OP	Bank Mask	LPDDR3 SDRAM
0	XXXXXXX1	Bank 0
1	XXXXXX1X	Bank 1
2	XXXXX1XX	Bank 2
3	XXXX1XXX	Bank 3
4	XXX1XXXX	Bank 4
5	XX1XXXXX	Bank 5
6	X1XXXXXX	Bank 6
7	1XXXXXXX	Bank 7

MR17 PASR Segment Mask (MA<7:0> = 11H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
Segment Mask							

Segment <7:0> Mask	Write-only	OP<7:0>	0B : refresh enable to the segment (=unmasked, default) 1B : refresh blocked (=masked)
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Segment	OP	Segment Mask	4Gb R13:11	6Gb ² R14:12	8Gb R14:12	12Gb ² R14:12	16Gb R14:12
0	0	XXXXXXX1			000B		
1	1	XXXXXX1X			001B		
2	2	XXXXX1XX			010B		
3	3	XXXX1XXX			011B		
4	4	XXX1XXXX			100B		
5	5	XX1XXXXX			101B		
6	6	X1XXXXXX			110B		
7	7	1XXXXXXX			111B		

Note:

1. This table indicates the range of row addresses in each masked segment. X is do not care for a particular segment.
2. No memory present at addresses with R13=R14=HIGH. Segment masks 6 and 7 are ignored.

MR18:31 (Reserved) (MA<7:0> = 12H - 1FH)
MR32 DQ Calibration Pattern A (MA<7:0> = 20H): MRR only

Reads to MR32 return DQ Calibration Pattern A. See the section of DQ Calibration.

MR33:39 (Reserved) (MA<7:0> = 21H - 27H)
MR40 DQ Calibration Pattern B (MA<7:0> = 28H): MRR only

Reads to MR40 return DQ Calibration Pattern B. See the section of DQ Calibration.

MR41 CA Calibration Mode Entry for CA0-3, CA5-8 (MA<7:0> = 29H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
A4							

See the section of CA Calibration.

MR42 CA Calibration Mode Exit (MA<7:0> = 2AH)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
A8							

See the section of CA Calibration.

MR43:47 (Reserved) (MA<7:0> = 2BH - 2FH)
MR48 CA Calibration Mode Entry for CA4, 9 (MA<7:0> = 30H)

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
C0							

See the section of CA Calibration.

MR49:62 (Reserved) (MA<7:0> = 31H - 3EH)
MR63 Reset (MA<7:0> = 3FH): MRW only

OP7	OP6	OP5	OP4	OP3	OP2	OP1	OP0
X or 0xFC							

Note: For additional information on MRW RESET, see Mode Register Write Command section.

TRUTH TABLES

Operation or timing that is not specified is illegal and after such an event, in order to guarantee proper operation, the LPDDR3 device must be powered down and then restarted through the specified initialization sequence before normal operation can continue.

COMMAND TRUTH TABLE

Command	SDR Command Pins (2)				DDR CA Pins (10)										CK_t edge
	CKE		CS_n	CA0	CA1	CA2	CA3	CA4	CA5	CA6	CA7	CA8	CA9		
	CK_t(n-1)	CK_t(n)													
MRW	H	H	L	L	L	L	L	MA0	MA1	MA2	MA3	MA4	MA5	rising	
			X	MA6	MA7	OP0	OP1	OP2	OP3	OP4	OP5	OP6	OP7	falling	
MRR	H	H	L	L	L	L	H	MA0	MA1	MA2	MA3	MA4	MA5	rising	
			X	MA6	MA7									falling	
Refresh (per bank)	H	H	L	L	L	H	L							rising	
			X											falling	
Refresh (all bank)	H	H	L	L	L	H	H							rising	
			X											falling	
Enter Self Refresh	H	L	L	L	L	H								rising	
			X											falling	
Active (bank)	H	H	L	L	H	R8	R9	R10	R11	R12	BA0	BA1	BA2	rising	
			X	R0	R1	R2	R3	R4	R5	R6	R7	R13	R14	falling	
Write (bank)	H	H	L	H	L	L	RFU	RFU	C1	C2	BA0	BA1	BA2	rising	
			X	Ap ³	C3	C4	C5	C6	C7	C8	C9	C10	C11	falling	
Read (bank)	H	H	L	H	L	H	RFU	RFU	C1	C2	BA0	BA1	BA2	rising	
			X	Ap ³	C3	C4	C5	C6	C7	C8	C9	C10	C11	falling	
Precharge (per bank, all bank) ¹¹	H	H	L	H	H	L	H	AB	X	X	BA0	BA1	BA2	rising	
			X	X	X	X	X	X	X	X	X	X	X	falling	
NOP	H	H	L	H	H	H								rising	
			X											falling	
Maintain SREF, PD (NOP) ⁴	L	L	X	H	H	H								rising	
			X											falling	
NOP	H	H	H											rising	
			X											falling	
Maintain PD, SREF (NOP) ⁴	L	L	X											rising	
			X											falling	
Enter Power Down	H	L	H											rising	
			X											falling	
Exit PD, SREF	L	H	H											rising	
			X											falling	

Note:

1. All LPDDR3 commands are defined by states of CS_n, CA0, CA1, CA2, CA3, and CKE at the rising edge of the clock.
2. Bank addresses BA0, BA1, BA2 (BA) determine which bank is to be operated upon.
3. AP "high" during a READ or WRITE command indicates that an auto-precharge will occur to the bank associated with the READ or WRITE command.
4. "X" means "H or L (but a defined logic level)", except when the LPDDR3 SDRAM is in PD, or SREF, in which case CS_n, CK_t/CK_c, and CA can be floated after the required tCPDED time is satisfied, and until the required exit procedure is initiated as described in the respective entry/exit procedure.
5. Self refresh exit is asynchronous.
6. VREF must be between 0 and VDDQ during Self Refresh operation.
7. CAxr refers to command/address bit "x" on the rising edge of clock.
8. CAxf refers to command/address bit "x" on the falling edge of clock.

9. CS_n and CKE are sampled at the rising edge of clock.
10. The least-significant column address C0 is not transmitted on the CA bus, and is implied to be zero.
11. AB "high" during Precharge command indicates that all bank Precharge will occur. In this case, Bank Address is do-not-care.
12. When CS_n is HIGH, LPDDR3 CA bus can be floated.

CKE TRUTH TABLE

Current State³	CKE_{n-1}⁴	CKE_n⁴	CS_n⁵	Command n⁶	Operation n⁶	Next State	Notes
Active Power Down	L	L	X	X	Maintain Active Power Down	Active Power Down	
	L	H	H	NOP	Exit Active Power Down	Active	7
Idle Power Down	L	L	X	X	Maintain Idle Power Down	Idle Power Down	
	L	H	H	NOP	Exit Idle Power Down	Idle	7
Resetting Power Down	L	L	X	X	Maintain Resetting Power Down	Resetting Power Down	
	L	H	H	NOP	Exit Resetting Power Down	Idle or Resetting	7,9
Self Refresh	L	L	X	X	Maintain Self Refresh	Self Refresh	
	L	H	H	NOP	Exit Self Refresh	Idle	8
Bank(s) Active	H	L	H	NOP	Enter Active Power Down	Active Power-Down	
All Banks Idle	H	L	H	NOP	Enter Idle Power Down	Idle Power Down	
	H	L	L	Enter Self Refresh	Enter Self Refresh	Self Refresh	10
Resetting	H	L	H	NOP	Enter Resetting Power Down	Resetting Power Down	
	H	H			Refer to the Command Truth Table		

Note:

1. All states and sequences not shown are illegal or reserved unless explicitly described elsewhere in this document.
2. 'X' means 'Don't care'.
3. "Current state" is the state of the LPDDR3 device immediately prior to clock edge n.
4. "CKEn" is the logic state of CKE at clock rising edge n; "CKEn-1" was the state of CKE at the previous clock edge.
5. "CS_n" is the logic state of CS_n at the clock rising edge n.
6. "Command n" is the command registered at clock edge N, and "Operation n" is a result of "Command n".
7. Power Down exit time (tXP) should elapse before a command other than NOP is issued. The clock must toggle at least twice during the tXP period.
8. Self-Refresh exit time (tXSR) should elapse before a command other than NOP is issued. The clock must toggle at least twice during the tXSR time.
9. Upon exiting Resetting Power Down, the device will return to the Idle state if tINIT5 has expired.
10. In the case of ODT disabled, all DQ output shall be Hi-Z. In the case of ODT enabled, all DQ shall be terminated to VDDQ.

Current State Bank n - Command to Bank n

Current State	Command	Operation	Next State	Note
Any	NOP	Continue previous operation	Current State	
Idle	Activate	Select and activate row	Active	
	Refresh (Per Bank)	Begin to refresh	Refreshing (Per Bank)	6
	Refresh (All Bank)	Begin to refresh	Refreshing (All Bank)	7
	MRW	Write value to Mode Register	MR Writing	7
	MRR	Read value from Mode Register	Idle MR Reading	
	Reset	Begin Device Auto-Initialization	Resetting	8
	Precharge	Deactive row in bank or banks	Precharging	9, 12
Row Active	Read	Select Column, and start read burst	Reading	
	Write	Select Column, and start write burst	Writing	
	MRR	Read value from Mode Register	Active MR Reading	
	Precharge	Deactivate row in bank or banks	Precharging	9
Reading	Read	Select column, and start new read burst	Reading	10,11
	Write	Select column, and start write burst	Writing	10,11,13
Writing	Write	Select Column, and start new write burst	Writing	10,11
	Read	Select column, and start read burst	Reading	10,11,14
Power On	Reset	Begin Device Auto-Initialization	Resetting	7, 9
Resetting	MRR	Read value from Mode Register	Resetting MR Reading	

Note:

1. The table applies when both CKE_{n-1} and CKE_n are HIGH, and after tXSR or tXP has been met if the previous state was Power Down.
2. All states and sequences not shown are illegal or reserved.

3. Current State Definitions:

Idle: The bank or banks have been precharged, and tRP has been met.

Row Active: A row in the bank has been activated, and tRCD has been met. No data bursts / accesses and no register accesses are in progress.

Reading: A READ burst has been initiated, with Auto Precharge disabled.

Writing: A WRITE burst has been initiated, with Auto Precharge disabled.

4. The following states must not be interrupted by a command issued to the same bank. NOP commands or allowable commands to the other bank should be issued on any clock edge occurring during these states. Allowable commands to the other banks are determined by its current state and Table "Current State Bank n - Command to Bank n", and according to Table "Current State Bank n - Command to Bank m".

Precharging: starts with the registration of a PRECHARGE command and ends when tRP is met. Once tRP is met, the bank will be in the idle state.

Row Activating: starts with registration of an ACTIVE command and ends when tRCD is met. Once tRCD is met, the bank will be in the 'Active' state.

Read with AP Enabled: starts with the registration of the READ command with Auto Precharge enabled and ends when tRP has been met. Once tRP has been met, the bank will be in the idle state.

Write with AP Enabled: starts with registration of a WRITE command with Auto Precharge enabled and ends when tRP has been met. Once tRP is met, the bank will be in the idle state.

5. The following states must not be interrupted by any executable command; NOP commands must be applied to each positive clock edge during these states.

Refreshing (Per Bank): starts with registration of a REFRESH (Per Bank) command and ends when tRFCpb is met. Once tRFCpb is met, the bank will be in an 'idle' state.

Refreshing (All Bank): starts with registration of a REFRESH(All Bank) command and ends when tRFCab is met. Once tRFCab is met, the device will be in an 'all banks idle' state.

Idle MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Idle state.

Resetting MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Resetting state.

Active MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Row Active state.

MR Writing: starts with the registration of a MRW command and ends when tMRW has been met. Once tMRW has been met, the bank will be in the Idle state.

Precharging All: starts with the registration of a PRECHARGE ALL command and ends when tRP is met. Once tRP is met, the bank will be in the idle state.

6. Bank-specific; requires that the bank is idle and no bursts are in progress.
7. Not bank-specific; requires that all banks are idle and no bursts are in progress.
8. Not bank-specific reset command is achieved through MODE REGISTER WRITE command.
9. This command may or may not be bank specific. If all banks are being precharged, they must be in a valid state for precharging.
10. A command other than NOP should not be issued to the same bank while a READ or WRITE burst with Auto Precharge is enabled.
11. The new Read or Write command could be Auto Precharge enabled or Auto Precharge disabled.
12. If a Precharge command is issued to a bank in the Idle state, tRP shall still apply.
13. A Write command may be applied after the completion of the Read burst, burst terminates are not permitted.
14. A Read command may be applied after the completion of the Write burst, burst terminates are not permitted.

Current State Bank n - Command to Bank m

Current State of Bank n	Command for Bank m	Operation	Next State for Bank m	Note
Any	NOP	Continue previous operation	Current State of Bank m	
Idle	Any	Any command allowed to Bank m	-	
Row Activating, Active, or Precharging	Activate	Select and activate row in Bank m	Active	6
	Read	Select column, and start read burst from Bank m	Reading	7
	Write	Select column, and start write burst to Bank m	Writing	7
	Precharge	Deactivate row in bank or banks	Precharging	8
	MRR	Read value from Mode Register	Idle MR Reading or Active MR Reading	9,10,12
Reading (Autoprecharge disabled)	Read	Select column, and start read burst from Bank m	Reading	7
	Write	Select column, and start write burst to Bank m	Writing	7,15
	Activate	Select and activate row in Bank m	Active	
	Precharge	Deactivate row in bank or banks	Precharging	8
Writing (Autoprecharge disabled)	Read	Select column, and start read burst from Bank m	Reading	7,16
	Write	Select column, and start write burst to Bank m	Writing	7
	Activate	Select and activate row in Bank m	Active	
	Precharge	Deactivate row in bank or banks	Precharging	8
Reading with Autoprecharge	Read	Select column, and start read burst from Bank m	Reading	7,13
	Write	Select column, and start write burst to Bank m	Writing	7,15,13
	Activate	Select and activate row in Bank m	Active	
	Precharge	Deactivate row in bank or banks	Precharging	8
Writing with Autoprecharge	Read	Select column, and start read burst from Bank m	Reading	7,13,16
	Write	Select column, and start write burst to Bank m	Writing	7,13
	Activate	Select and activate row in Bank m	Active	
	Precharge	Deactivate row in bank or banks	Precharging	8
Power On	Reset	Begin Device Auto-Initialization	Resetting	11, 14
Resetting	MRR	Read value from Mode Register	Resetting MR Reading	

Note:

1. The table applies when both CKE_{n-1} and CKE_n are HIGH, and after tXSR or tXP has been met if the previous state was Self Refresh or Power Down.

2. All states and sequences not shown are illegal or reserved.

3. Current State Definitions:

Idle: the bank has been precharged, and tRP has been met.

Active: a row in the bank has been activated, and tRCD has been met. No data bursts/accesses and no register accesses are in progress.

Reading: a READ burst has been initiated, with Auto Precharge disabled.

Writing: a WRITE burst has been initiated, with Auto Precharge disabled.

4. REFRESH, SELF REFRESH, and MODE REGISTER WRITE commands may only be issued when all bank are idle.

5. The following states must not be interrupted by any executable command; NOP commands must be applied during each clock cycle while in these states:

Idle MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Idle state.

Resetting MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Resetting state.

Active MR Reading: starts with the registration of a MRR command and ends when tMRR has been met. Once tMRR has been met, the bank will be in the Row Active state.

MR Writing: starts with the registration of a MRW command and ends when tMRW has been met. Once tMRW has been met, the bank will be in the Idle state.

6. tRRD must be met between Activate command to Bank n and a subsequent Activate command to Bank m.

7. READs or WRITEs listed in the Command column include READs and WRITEs with Auto Precharge enabled and READs and WRITEs with Auto Precharge disabled.
8. This command may or may not be bank specific. If all banks are being precharged, they must be in a valid state for precharging.
9. MRR is allowed during the Row Activating state and MRW is prohibited during the Row Activating state. (Row Activating starts with registration of an Activate command and ends when tRCD is met.)
10. MRR is allowed during the Precharging state. (Precharging starts with registration of a Precharge command and ends when tRP is met.)
11. Not bank-specific; requires that all banks are idle and no bursts are in progress.
12. The next state for Bank m depends on the current state of Bank m (Idle, Row Activating, Precharging, or Active). The reader shall note that the state may be in transition when a MRR is issued. Therefore, if Bank m is in the Row Activating state and Precharging, the next state may be Active and Precharge dependent upon tRCD and tRP respectively.
13. Read with auto precharge enabled or a Write with auto precharge enabled may be followed by any valid command to other banks provided that the timing restrictions in the section of Precharge and Auto Precharge clarification are followed.
14. Reset command is achieved through MODE REGISTER WRITE command.
15. A Write command may be applied after the completion of the Read burst, burst terminates are not permitted.
16. A Read command may be applied after the completion of the Write burst, burst terminates are not permitted.

DATA MASK TRUTH TABLE

Function	DM	DQ	Note
Write Enable	L	Valid	1
Write Inhibit	H	X	1

Note:

1. Used to mask write data, provided coincident with the corresponding data.

Absolute Maximum DC Ratings

Stresses greater than those listed may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Parameter	Symbol	Min	Max	Unit	Notes
VDD1 supply voltage relative to VSS	VDD1	-0.4	2.3	V	1
VDD2 supply voltage relative to VSS	VDD2	-0.4	1.6	V	1
VDDCA supply voltage relative to VSSCA	VDDCA	-0.4	1.6	V	1, 2
VDDQ supply voltage relative to VSSQ	VDDQ	-0.4	1.6	V	1, 3
Voltage on Any Pin relative to VSS	VIN, VOUT	-0.4	1.6	V	
Storage Temperature	TSTG	-55	125	°C	4

Note:

1. See the section "Power-up, Initialization, and Power-off" for relationships between power supplies.
2. $VREFCA \leq 0.6 \times VDDCA$; however, $VREFCA$ may be $\geq VDDCA$ provided that $VREFCA \leq 300mV$.
3. $VREFDQ \leq 0.7 \times VDDQ$; however, $VREFDQ$ may be $\geq VDDQ$ provided that $VREFDQ \leq 300mV$.
4. Storage Temperature is the case surface temperature on the center/top side of the device. For the measurement conditions, please refer to JESD51-2 standard.

AC and DC Operating Conditions

Operation or timing that is not specified is illegal, and after such an event, in order to guarantee proper operation, the LPDDR3 Device must be powered down and then restarted through the specialized initialization sequence before normal operation can continue.

Recommended DC Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit
Core Power 1	VDD1	1.70	1.80	1.95	V
Core Power 2	VDD2	1.14	1.20	1.30	V
Input Buffer Power	VDDCA	1.14	1.20	1.30	V
I/O Buffer Power	VDDQ	1.14	1.20	1.30	V

Note :

1. VDD1 uses significantly less current than VDD2.
2. The voltage range is for DC voltage only. DC is defined as the voltage supplied at the DRAM and is inclusive of all noise up to 1MHz at the DRAM package ball.

Input Leakage Current

Parameter	Symbol	Min	Max	Unit	Note
Input Leakage current	IL	-2	2	uA	2
VREF supply leakage current	IVREF	-1	1	uA	1

Note:

1. For CA, CKE, CS_n, CK_t, CK_c. Any input $0V \leq VIN \leq VDDCA$ (All other pins not under test = 0V)
2. Although DM is for input only, the DM leakage shall match the DQ and DQS_t/DQS_c output leakage specification.
3. The minimum limit requirement is for testing purposes. The leakage current on VREFCA and VREFDQ pins should be minimal.
4. $VREFDQ = VDDQ/2$ or $VREFCA = VDDCA/2$. (All other pins not under test = 0V)

Operating Temperature

Parameter	Symbol	Min	Max	Unit	Note
Operating Temperature	Standard	T_{OPER}	-30	85	°C
	Extended		-30	105	

Note:

1. Operating Temperature is the case surface temperature on the center-top side of the LPDDR3 device. For the measurement conditions, please refer to JESD51-2 standard.
2. Some applications require operation of LPDDR3 in the maximum temperature conditions in the Extended Temperature Range between -30°C and 105°C case temperature. For LPDDR3 devices, derating may be necessary to operate in this range. See MR4 on the section "Mode Register".
3. Either the device case temperature rating or the temperature sensor (See the section of "Temperature Sensor") may be used to set an appropriate refresh rate, determine the need for AC timing de-rating and/or monitor the operating temperature. When using the temperature sensor, the actual device case temperature may be higher than the TOPER rating that applies for the Standard or Elevated Temperature Ranges. For example, TCASE may be above 85°C when the temperature sensor indicates a temperature of less than 85°C.

AC and DC Input Measurement Levels

AC and DC Logic Input Levels for Single-Ended CA and CS_n Signals

Parameter	Symbol	LPDDR3 1866		LPDDR3 1600/1333		Unit	Note
		Min	Max	Min	Max		
AC Input Logic High	VIHCA	VREF + 0.135	Note 2	VREF + 0.150	Note 2	V	1,2
AC Input Logic Low	VILCA	Note 2	VREF - 0.135	Note 2	VREF - 0.150	V	1,2
DC Input Logic High	VIHCA	VREF + 0.100	VDDCA	VREF + 0.100	VDDCA	V	1
DC Input Logic Low	VILCA	VSSCA	VREF - 0.100	VSSCA	VREF - 0.100	V	1
Reference Voltage for CA and CS_n Inputs	VREFCA(DC)	0.49 * VDDCA	0.51 * VDDCA	0.49 * VDDCA	0.51 * VDDCA	V	3,4

Note:

1. For CA and CS_n input only pins. VREF = VREFCA(DC).
2. See the section "Overshoot and Undershoot Specifications".
3. The ac peak noise on VREFCA may not allow VREFCA to deviate from VREFCA(DC) by more than +/-1% VDDCA (for reference: approx. +/- 12 mV).
4. For reference: approx. VDDCA/2 +/- 12 mV.

AC and DC Logic Input Levels for CKE

Parameter	Symbol	Min	Max	Unit	Note
CKE Input High Level	VIHCKE	0.65 * VDDCA	Note 1	V	1
CKE Input Low Level	VILCKE	Note 1	0.35 * VDDCA	V	1

Note: 1. See the section "Overshoot and Undershoot Specifications".

AC and DC Logic Input Levels for Single-Ended Data (DQ and DM) Signals

Parameter	Symbol	LPDDR3 1866		LPDDR3 1600/1333		Unit	Note
		Min	Max	Min	Max		
AC Input High Voltage	VIHDQ	VREF + 0.135	Note 2	VREF + 0.150	Note 2	V	1,2
AC Input Low Voltage	VILDQ	Note 2	VREF - 0.135	Note 2	VREF - 0.150	V	1,2
DC Input High Voltage	VIHDQ	VREF + 0.100	VDDCA	VREF + 0.100	VDDQ	V	1
DC Input Low Voltage	VILDQ	VSSCA	VREF - 0.100	VSSCA	VREF - 0.100	V	1
Reference Voltage for DQ and DM Inputs	VREFDQ(DC) (DQ ODT disabled)	0.49 * VDDQ	0.51 * VDDQ	0.49 * VDDQ	0.51 * VDDQ	V	3,4
Reference Voltage for DQ and DM Inputs	VREFDQ(DC) (DQ ODT enabled)	VODTR/2 - 0.01 * VDDQ	VODTR/2 + 0.01 * VDDQ	0.5 * Vodtr - 0.01 * VDDQ	0.5 * Vodtr + 0.01 * VDDQ	V	3,5,6

Note:

1. For DQ input only pins. VREF = VREFDQ(DC).
2. See the section of Overshoot and Undershoot Specifications.
3. The ac peak noise on VREFDQ may not allow VREFDQ to deviate from VREFDQ(DC) by more than +/-1% VDDQ (for reference: approx. +/- 12 mV).
4. For reference: approx. VDDQ/2 +/- 12 mV.
5. For reference: approx. VODTR/2 +/- 12 mV.
6. The nominal mode register programmed value for RODT and the nominal controller output impedance RON are used for the calculation of VODTR. For testing purposes a controller RON value of 50 Ω is used.

$$Vodtr = (2 * RON + RTT) / (RON + RTT) * VDDQ$$

VREF Tolerances

The dc-tolerance limits and ac-noise limits for the reference voltages VREFCA and VREFDQ are illustrated in Figure below. It shows a valid reference voltage $V_{REF}(t)$ as a function of time. (VREF stands for VREFCA and VREFDQ likewise). VDD stands for VDDCS for VREFCA and VDDQ for VREFDQ. $V_{REF}(DC)$ is the linear average of $V_{REF}(t)$ over a very long period of time (e.g. 1 sec) and is specified as a fraction of the linear average of VDDCA or VDDQ also over a very long period of time (e.g. 1 sec). This average has to meet the min/max requirements in Table "Electrical Characteristics and Operating Conditions". Furthermore $V_{REF}(t)$ may temporarily deviate from $V_{REF}(DC)$ by no more than $\pm 1\%$ VDD. $V_{REF}(t)$ cannot track noise on VDDQ or VDDCA if this would send VREF outside these specifications.

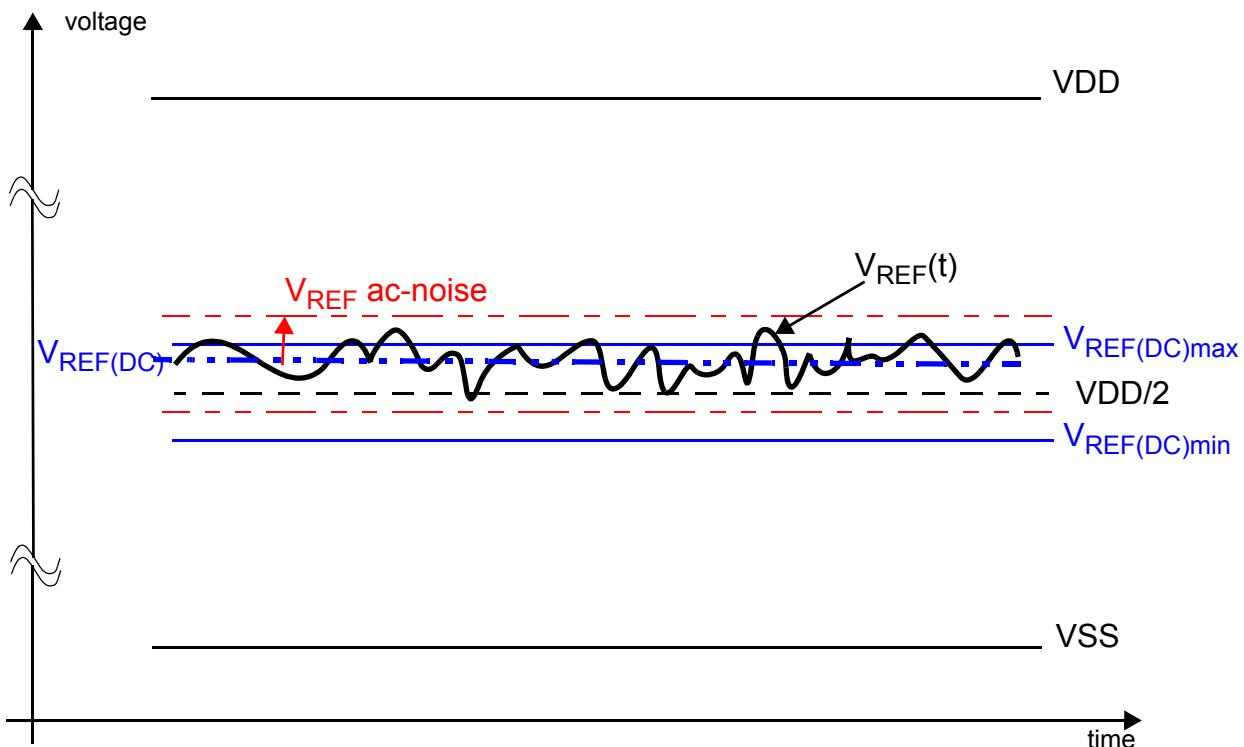


Figure. Illustration of $V_{REF}(DC)$ tolerance and V_{REF} ac-noise limits

The voltage levels for setup and hold time measurements VIH(AC), VIH(DC), VIL(AC) and VIL(DC) are dependent on VREF. "VREF" shall be understood as $V_{REF}(DC)$, as defined in Figure above.

This clarifies that dc-variations of VRef affect the absolute voltage a signal has to reach to achieve a valid high or low level and therefore the time to which setup and hold is measured. System timing and voltage budgets need to account for $V_{REF}(DC)$ deviations from the optimum position within the data-eye of the input signals.

This also clarifies that the LPDDR3 setup/hold specification and derating values need to include time and voltage associated with VREF ac-noise. Timing and voltage effects due to ac-noise on VREF up to the specified limit ($\pm 1\%$ of VDD) are included in LPDDR3 timings and their associated deratings.

Input Signal

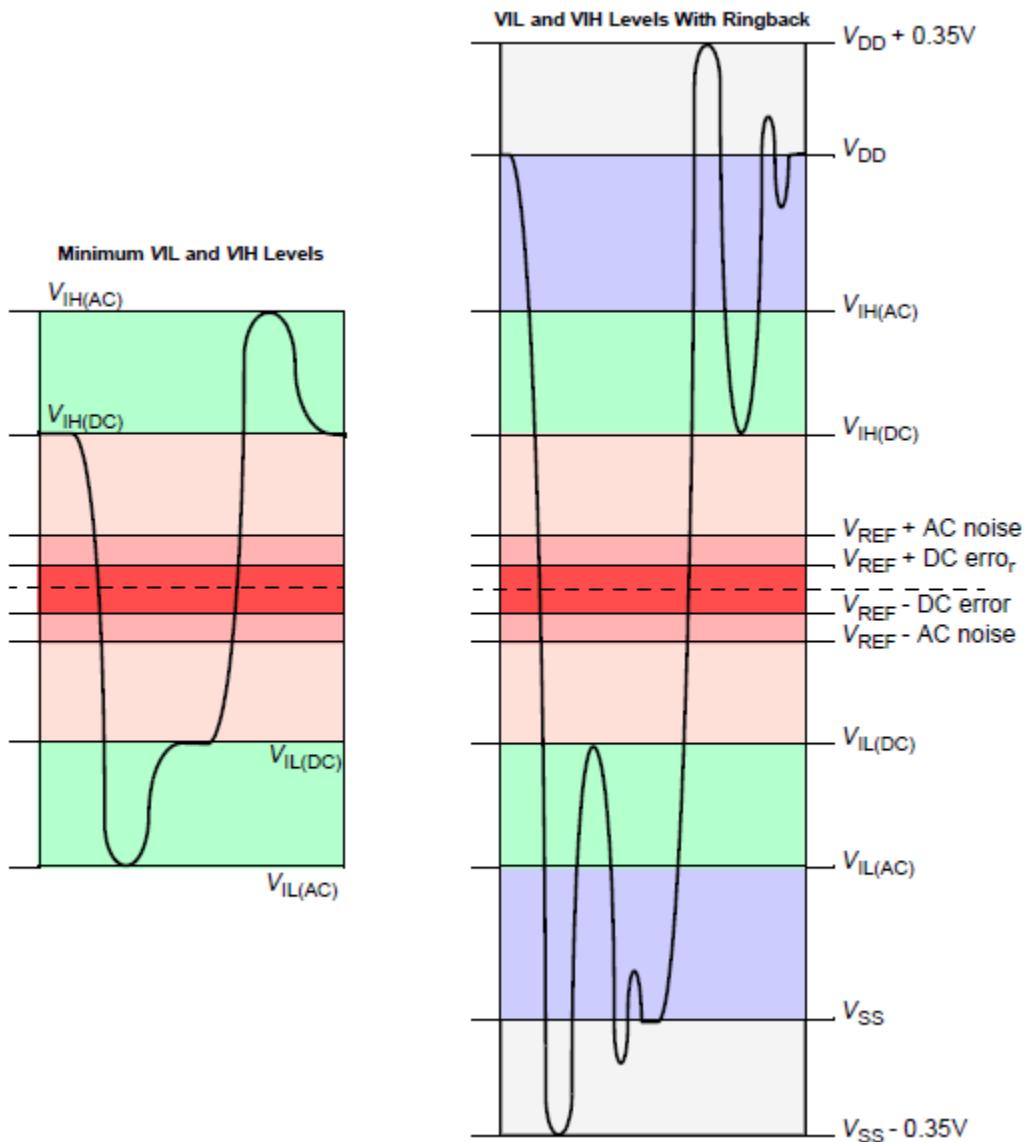


Figure. LPDDR3 Input signal

Note:

1. Numbers reflect nominal values.
2. For CA0-9, CK_t, CK_c and CS_n, VDD stands for VDDCA. For DQ, DM/DNV, DQS_t and DQS_c, VDD stands for VDDQ.
3. For CA0-9, CK_t, CK_c and CS_n, VSS stands for VSSCA. For DQ, DM/DNV, DQS_t and DQS_c, VSS stands for VSSQ.

AC and DC Logic Input Levels for Differential Signals

Differential Signal Definition

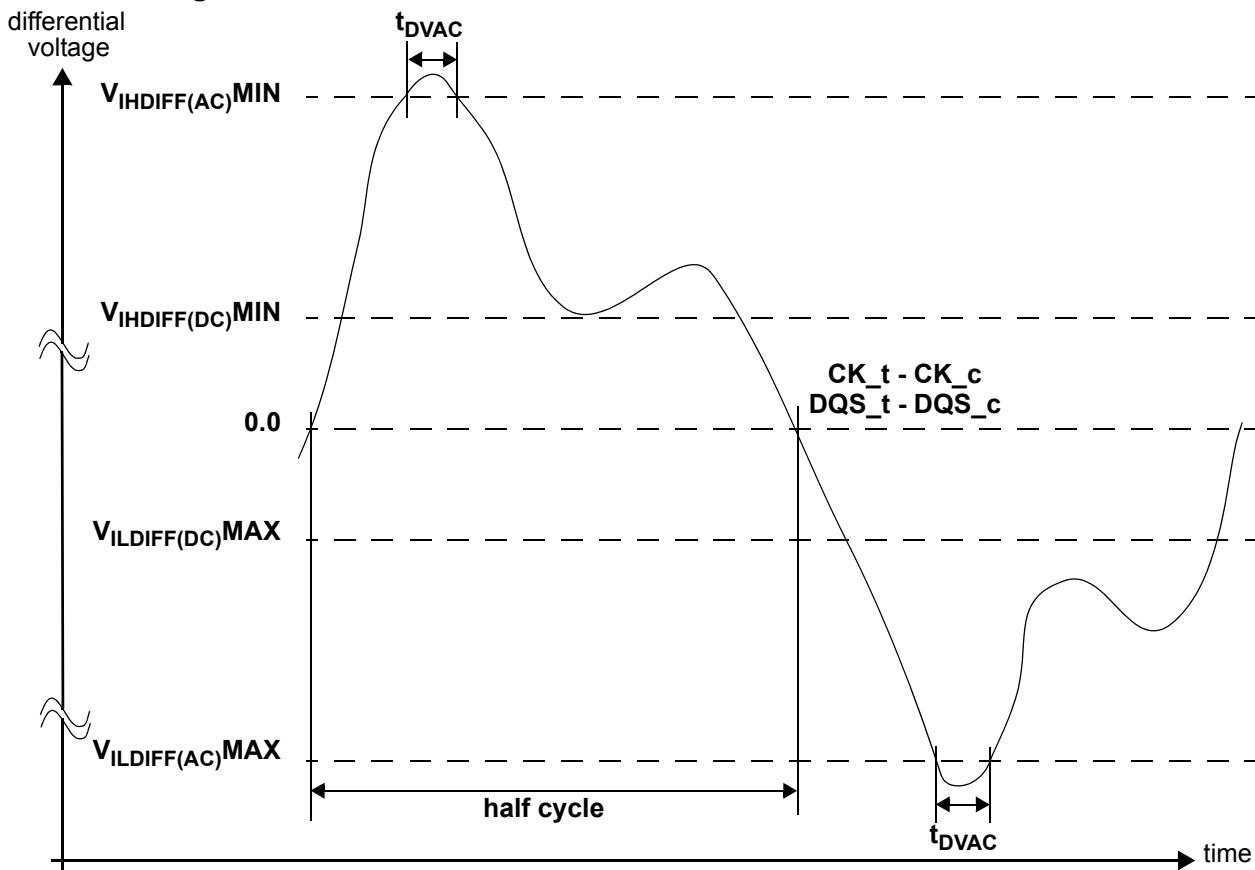


Figure. Definition of differential ac-swing and Time above ac-level t_{DVAC}

Differential swing requirements for clock and strobe

Parameter	Symbol	Min	Max	Unit	Note
DC Differential Input High	VIHDIFF(DC)	2 x (VIH(DC) - VREF)	Note 3	V	1
DC Differential Input Low	VILDIFF(DC)	Note 3	2 x (VIL(DC) - VREF)	V	1
AC Differential Input High	VIHDIFF(AC)	2 x (VIH(AC) - VREF)	Note 3	V	2
AC Differential Input Low	VILDIFF(AC)	Note 3	2 x (VIL(AC) - VREF)	V	2

Note:

- Used to define a differential signal slew-rate. For CK_t - CK_c use VIH/VIL(dc) of CA and VREFCA; for DQS_t - DQS_c, use VIH/VIL(dc) of DQs and VREFDQ; if a reduced dc-high or dc-low level is used for a signal group, then the reduced level applies also here.
- For CK_t - CK_c use VIH/VIL(ac) of CA and VREFCA; for DQS_t - DQS_c, use VIH/VIL(ac) of DQs and VREFDQ; if a reduced ac-high or ac-low level is used for a signal group, then the reduced level applies also here.
- These values are not defined, however the single-ended signals CK_t, CK_c, DQS_t, and DQS_c need to be within the respective limits (VIH(dc) max, VIL(dc)min) for single-ended signals as well as the limitations for overshoot and undershoot. Refer to the section of "Overshoot and Undershoot Specifications".
- For CK_t and CK_c, Vref = VrefCA(DC). For DQS_t and DQS_c, Vref = VrefDQ(DC).

Table. Allowed time before ringback (tDVAC) for DQS_t - DQS_c

Slew Rate [V/ns]	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 270mV	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 300mV	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 300mV
	1866Mbps	1600Mbps	1333Mbps
	MIN	MIN	MIN
> 8.0	40	48	58
8.0	40	48	58
7.0	39	46	56
6.0	36	43	53
5.0	33	40	50
4.0	29	35	45
3.0	21	27	37
<3.0	21	27	37

Table. Allowed time before ringback (tDVAC) for CK_t - CK_c

Slew Rate [V/ns]	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 270mV	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 300mV	t _{DVAC} [ps] @ VIH/Ldiff(ac) = 300mV
	1866Mbps	1600Mbps	1333Mbps
	MIN	MIN	MIN
> 8.0	40	48	58
8.0	40	48	58
7.0	39	46	56
6.0	36	43	53
5.0	33	40	50
4.0	29	35	45
3.0	21	27	37
<3.0	21	27	37

Single-ended Requirements for Differential Signals

Each individual component of a differential signal (CK_t, DQS_t, CK_c, or DQS_c) has also to comply with certain requirements for single-ended signals.

CK_t and CK_c shall meet VSEH(AC)min / VSEL(AC)max in every half-cycle.

DQS_t, DQS_c shall meet VSEH(AC)min / VSEL(AC)max in every half-cycle proceeding and following a valid transition.

Note that the applicable ac-levels for CA and DQ's are different per speed-bin.

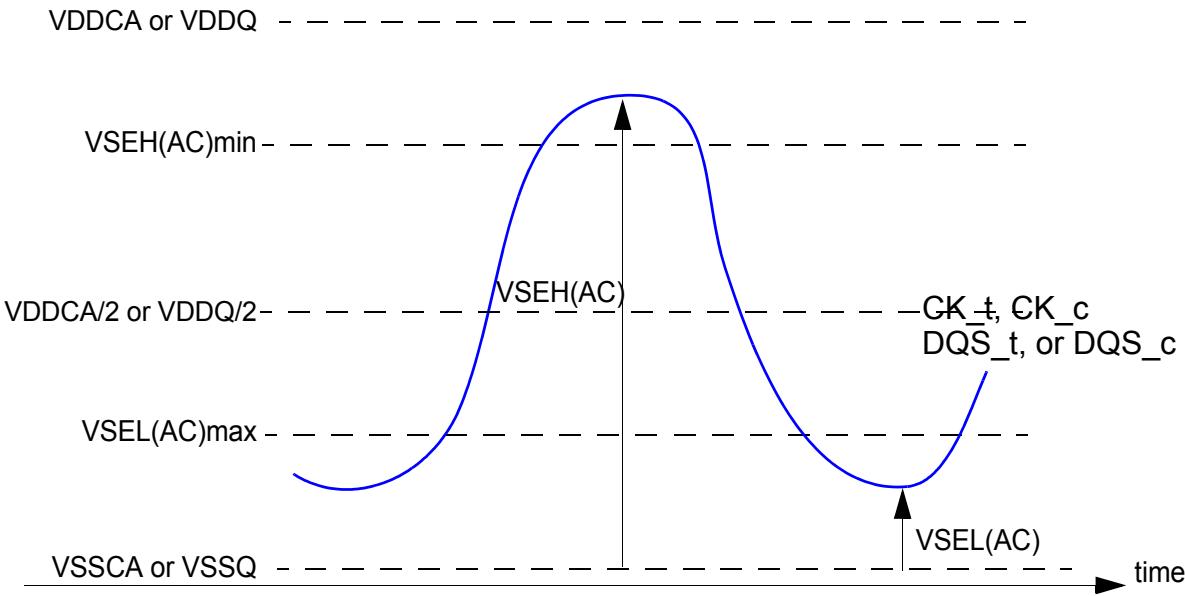


Figure. Single-ended requirement for differential signals

Note that while CA and DQ signal requirements are with respect to VREF, the single-ended components of differential signals have a requirement with respect to VDDQ/2 for DQS_t, DQS_c and VDDCA/2 for CK_t, CK_c; this is nominally the same. The transition of single-ended signals through the ac-levels is used to measure setup time. For single-ended components of differential signals the requirement to reach VSEL(AC)max, VSEH(AC)min has no bearing on timing, but adds a restriction on the common mode characteristics of these signals.

Table. Single-ended Levels for Clock and Strobe

Parameter	Symbol	Min	Max	Unit	Note
Single-ended High Level for strobes	VSEH (AC150)	(VDDQ/2) + 0.150	Note 3	V	1, 2
Single-ended High Level for CK_t and CK_c		(VDDCA/2) + 0.150	Note 3	V	1, 2
Single-ended Low Level for strobes	VSEL (AC150)	Note 3	(VDDQ / 2) - 0.150	V	1, 2
Single-ended Low Level for CK_t and CK_c		Note 3	(VDDCA / 2) - 0.150	V	1, 2
Single-ended High Level for strobes	VSEH (AC135)	(VDDQ/2) + 0.135	Note 3	V	1, 2
Single-ended High Level for CK_t and CK_c		(VDDCA/2) + 0.135	Note 3	V	1, 2
Single-ended Low Level for strobes	VSEL (AC135)	Note 3	(VDDQ / 2) - 0.135	V	1, 2
Single-ended Low Level for CK_t and CK_c		Note 3	(VDDCA / 2) - 0.135	V	1, 2

Note:

1. For CK_t, CK_c use VSEH/VSEL(AC) of CA; for strobes (DQS0_t, DQS0_c, DQS1_t, DQS1_c, DQS2_t, DQS2_c, DQS3_t, DQS3_c) use VIH/VIL(AC) of DQs.
2. VIH(AC)/VIL(AC) for DQs is based on VREFDQ; VSEH(AC)/VSEL(AC) for CA is based on VREFCA; if a reduced ac-high or ac-low level is used for a signal group, then the reduced level applies also here.
3. These values are not defined, however the single-ended signals CK_t, CK_c, DQS0_t, DQS0_c, DQS1_t, DQS1_c, DQS2_t, DQS2_c, DQS3_t, DQS3_c need to be within the respective limits (VIH(DC) max, VIL(DC)min) for single-ended signals as well as the limitations for overshoot and undershoot. Refer to the section of Overshoot and Undershoot Specifications.

Differential Input Cross Point Voltage

To guarantee tight setup and hold times as well as output skew parameters with respect to clock and strobe, each cross point voltage of differential input signals (CK_t, CK_c and DQS_t, DQS_c) must meet the requirements in "Single-ended Levels for Clock and Strobe". The differential input cross point voltage VIX is measured from the actual cross point of true and complement signals to the midlevel between of VDD and VSS.

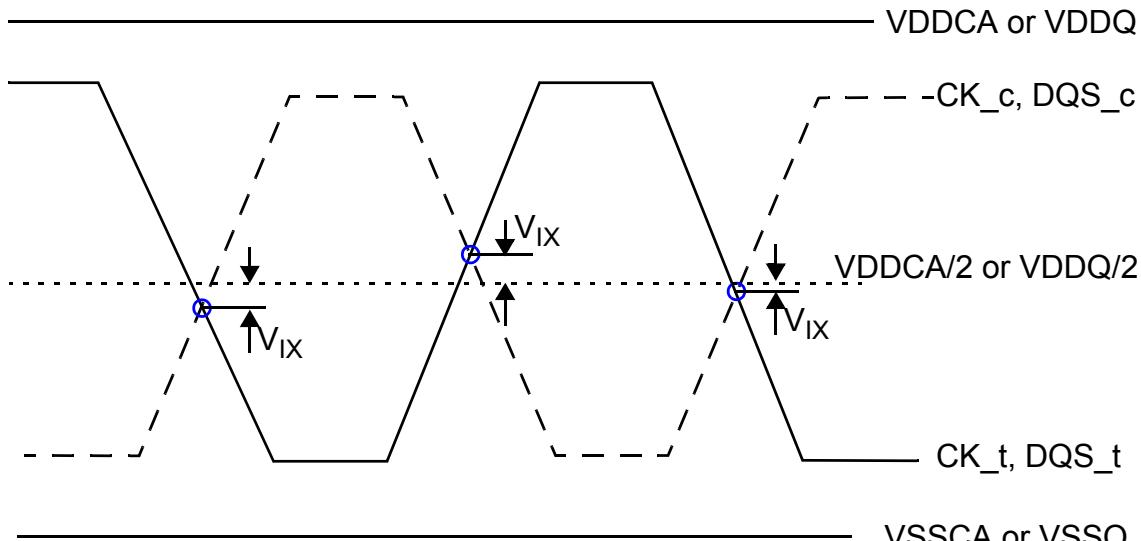


Figure. VIX definition

Table. Cross Point Voltage for Differential Input Signals (Clock and Strobe)

Parameter	Symbol	Min	Max	Unit	Note
Differential Input Cross Point Voltage relative to VDDCA/2 for CK_t and CK_c	VIXCA	-120	120	mV	1, 2
Differential Input Cross Point Voltage relative to VDDQ/2 for DQS_t and DQS_c	VIXDQ	-120	120	mV	1, 2

Note:

1. The typical value of VIX(AC) is expected to be about $0.5 \times VDD$ of the transmitting device, and VIX(AC) is expected to track variations in VDD. VIX(AC) indicates the voltage at which differential input signals must cross.
2. For CK_t and CK_c, VREF = VREFCA(DC). For DQS_t and DQS_c, VREF = VREFDQ(DC).

Slew Rate Definitions for Single-ended Input Signals

See "CA and CS_n Setup, Hold and Derating" for single-ended slew rate definitions for address and command signals.
 See "Data Setup, Hold and Slew Rate Derating" for single-ended slew rate definitions for data signals.

Slew Rate Definitions for Differential Input Signals

Input slew rate for differential signals (CK_t, CK_c and DQS_t, DQS_c) are defined and measured as shown in the table and figure below.

Table. Differential Input Slew Rate Definition

Parameter	Measured		Defined by
	From	To	
Differential Input Slew Rate for Rising Edge (CK_t - CK_c and DQS_t - DQS_c)	$V_{ILdiffmax}$	$V_{IHdiffmin}$	$[V_{IHdiffmin} - V_{ILdiffmax}] / \Delta TRdiff$
Differential Input Slew Rate for Falling Edge (CK_t - CK_c and DQS_t - DQS_c)	$V_{IHdiffmin}$	$V_{ILdiffmax}$	$[V_{IHdiffmin} - V_{ILdiffmax}] / \Delta TFdiff$

Note: 1. The differential signal (i.e. CK_t - CK_c and DQS_t - DQS_c) must be linear between these thresholds.

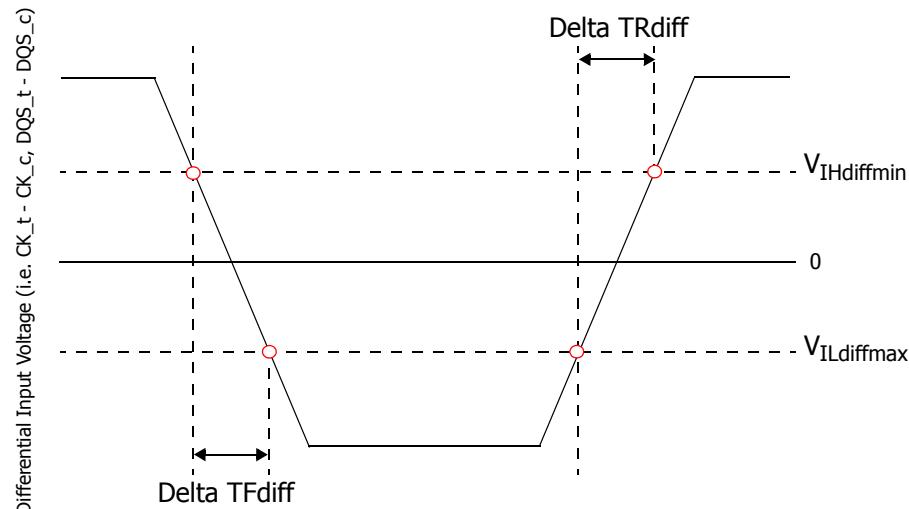


Figure. Differential Input Slew Rate Definition for CK_t, CK_c and DQS_t, DQS_c

AC and DC Output Measurement Levels

Single Ended AC and DC Output Levels

Parameter	Symbol	Levels	Unit	Note
DC Output Logic High Measurement Level (for IV curve linearity)	VOH(DC)	0.9 x VDDQ	V	1
	VOL(DC) ODT disabled	0.1 x VDDQ	V	2
DC Output Logic Low Measurement Level (for IV curve linearity)	VOL(DC) ODT enabled	VDDQ * [0.1 + 0.9 * (RON/ (RTT+RON))]	V	3
AC Output Logic High Measurement Level (for output slew rate)	VOH(AC)	VREFDQ + 0.12	V	
AC Output Logic Low Measurement Level (for output slew rate)	VOL(AC)	VREFDQ - 0.12	V	
Output Leakage current (DQ, DM, DQS_t and DQS_c) (DQ, DQS_t and DQS_c are disabled; 0V ≤ VOUT ≤ VDDQ)	Min Max	I _{OZ}	-5 5	uA
Delta RON between pull-up and pull-down for DQ and DM	Min Max	M _{M_{PUPD}}	-15 15	%

Note:

1. IOH = -0.1mA,

2. IOL = 0.1mA

3. The min value is derived when using RTT, min and RON,max (+/- 30% uncalibrated, +/-15% calibrated).

Differential AC and DC Output Levels (DQS_t, DQS_c)

Parameter	Symbol	Levels	Unit	Note
AC Differential Output High measurement Level (for Output SR)	VOHdiff(AC)	+ 0.20 x VDDQ	V	
AC Differential Output Low measurement Level (for Output SR)	VOLdiff(AC)	- 0.20 x VDDQ	V	

Note:

1. IOH = -0.1mA,

2. IOL = 0.1mA

Single Ended Output Slew Rate

With the reference load for timing measurements, output slew rate for falling and rising edges is defined and measured between VOL(AC) and VOH(AC) for single ended signals as shown in below Table and Figure.

Parameter	Measured		Defined by
	From	To	
Single Ended Output Slew Rate for Rising Edge	VOL(AC)	VOH(AC)	$[VOH(AC) - VOL(AC)] / \Delta TRse$
Single Ended Output Slew Rate for Falling Edge	VOH(AC)	VOL(AC)	$[VOH(AC) - VOL(AC)] / \Delta TFse$

Note: Output slew rate is verified by design and characterization and may not be subject to production test.

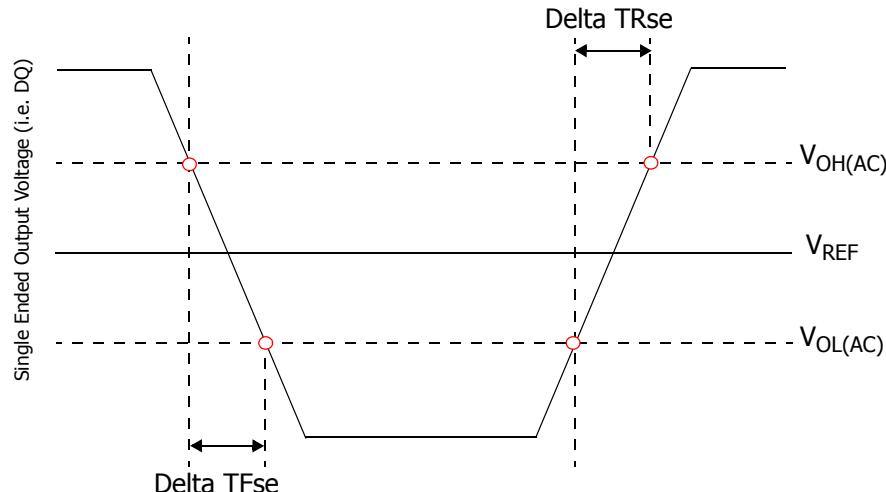


Figure. Single Ended Output Slew Rate Definition

Table. Output Slew Rate (Single Ended)

Parameter	Symbol	Min	Max	Unit	Note
Single-ended Output Slew Rate ($RON = 40\Omega \pm 30\%$)	SRQse	1.5	4.0	V/ns	
Output slew-rate matching Ratio (Pull-up to Pull-down)		0.7	1.4		

Description:

SR: Slew Rate

Q: Query Output (like in DQ, which stands for Data-in, Query-Output)

se: Single-ended Signals

Note:

1. Measured with output reference load.
2. The ratio of pull-up to pull-down slew rate is specified for the same temperature and voltage, over the entire temperature and voltage range. For a given output, it represents the maximum difference between pull-up and pull-down drivers due to process variation.
3. The output slew rate for falling and rising edges is defined and measured between VOL(AC) and VOH(AC).
4. Slew rates are measured under average SSO conditions, with 50% of DQ signals per data byte switching.

Differential Output Slew Rate

With the reference load for timing measurements, output slew rate for falling and rising edges is defined and measured between VOLdiff(AC) and VOHdiff(AC) for differential signals as shown in below Table and Figure.

Parameter	Measured		Defined by
	From	To	
Differential Output Slew Rate for Rising Edge	$V_{OL\text{diff}}(\text{AC})$	$V_{OH\text{diff}}(\text{AC})$	$[V_{OH\text{diff}}(\text{AC}) - V_{OL\text{diff}}(\text{AC})] / \Delta T_{R\text{diff}}$
Differential Output Slew Rate for Falling Edge	$V_{OH\text{diff}}(\text{AC})$	$V_{OL\text{diff}}(\text{AC})$	$[V_{OH\text{diff}}(\text{AC}) - V_{OL\text{diff}}(\text{AC})] / \Delta T_{F\text{diff}}$

Note: 1. Output slew rate is verified by design and characterization, and may not be subject to production test.

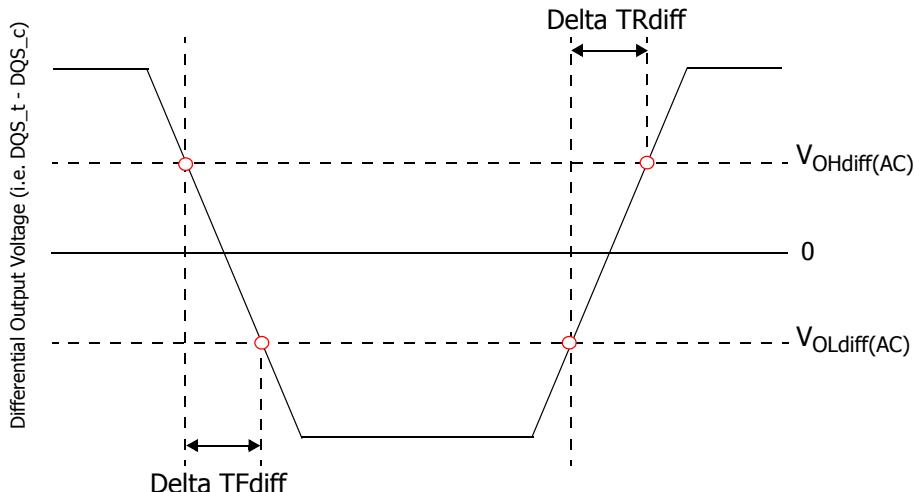


Figure. Differential Output Slew Rate Definition

Table. Output Slew Rate (Differential)

Parameter	Symbol	Min	Max	Unit	Note
Differential Output Slew Rate ($R_{ON} = 40\Omega \pm 30\%$)	$SR_{Q\text{diff}}$	3.0	8.0	V/ns	

Description:

SR: Slew Rate

Q: Query Output (like in DQ, which stands for Data-in, Query-Output)

diff: Differential Signals

Note:

1. Measured with output reference load.
2. The output slew rate for falling and rising edges is defined and measured between VOL(AC) and VOH(AC).
3. Slew rates are measured under average SSO conditions, with 50% of DQ signals per data byte switching.

Overshoot and Undershoot Specifications

Parameter	1866	1600	1333	Unit
Maximum peak amplitude allowed for overshoot area		0.35		V
Maximum peak amplitude allowed for undershoot area		0.35		V
Maximum overshoot area above VDD	0.10	0.10	0.12	V-ns
Maximum undershoot area below VSS	0.10	0.10	0.12	V-ns

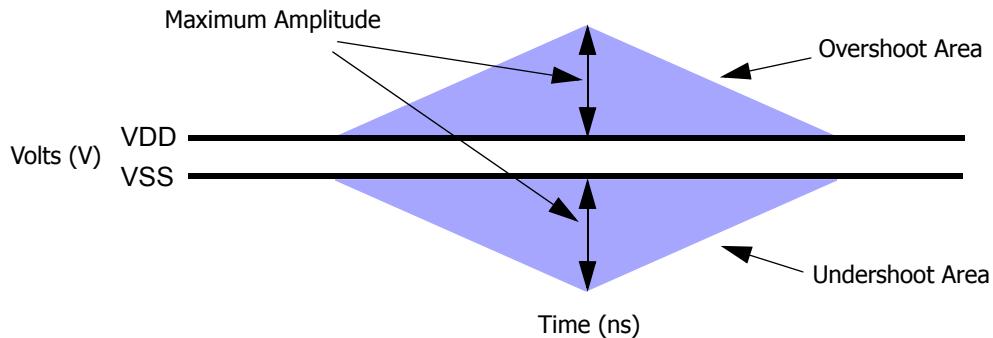


Figure. Overshoot and Undershoot Definition

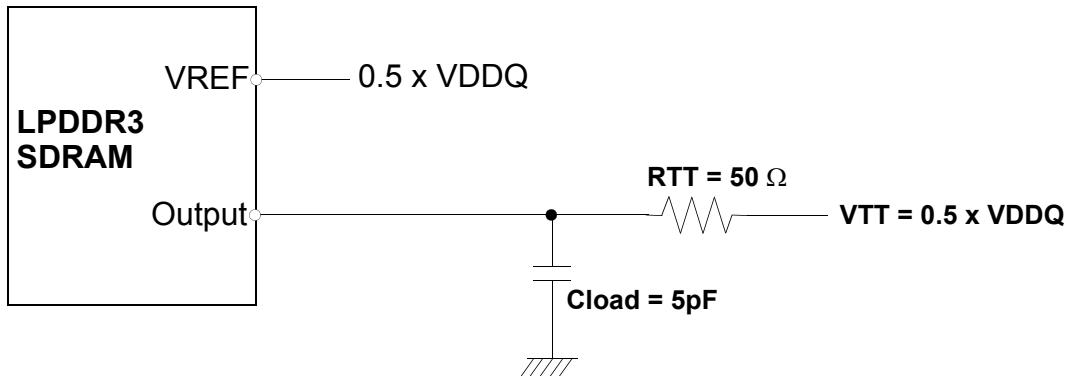
Note:

1. VDD stands for VDDCA for CA0-9, CK_t, CK_c, CS_n, and CKE. VDD stands for VDDQ for DQ, DM, ODT, DQS_t, and DQS_c.
2. VSS stands for VSSCA for CA0-9, CK_t, CK_c, CS_n, and CKE. VSS stands for VSSQ for DQ, DM, ODT, DQS_t, and DQS_c.
3. Absolute maximum requirements apply.
4. Maximum peak amplitude values are referenced from actual VDD and VSS values.
5. Maximum area values are referenced from maximum operating VDD and VSS values.

Output Buffer Characteristics

HSUL_12 Driver Output Timing Reference Load

These 'Timing Reference Loads' are not intended as a precise representation of any particular system environment or a depiction of the actual load presented by a production tester. System designers should use IBIS or other simulation tools to correlate the timing reference load to a system environment. Manufacturers correlate to their production test conditions, generally one or more coaxial transmission lines terminated at the tester electronics.



Note: 1. All output timing parameter values (like t_{DQSCK} , t_{DQSO} , t_{QHS} , t_{HZ} , t_{RPRE} etc.) are reported with respect to this reference load. This reference load is also used to report slew rate.

Figure. HSUL_12 Driver Output Reference Load for Timing and Slew Rate

RON_{PU} and RON_{PD} resistor Definition

$$RON_{PU} = \frac{(VDDQ - V_{out})}{ABS(I_{out})}$$

Note 1: This is under the condition that RON_{PD} is turned off

$$RON_{PD} = \frac{V_{out}}{ABS(I_{out})}$$

Note 1: This is under the condition that RON_{PU} is turned off

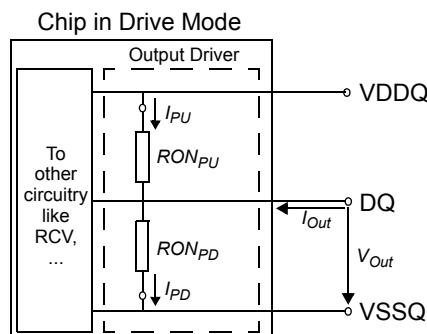


Figure. Output Driver: Definition of Voltages and Currents

RON_PU and RON_PD Characteristics with ZQ Calibration

Output driver impedance RON is defined by the value of the external reference resistor RZQ. Nominal RZQ is 240Ω.

Table - Output Driver DC Electrical Characteristics with ZQ Calibration

RON _{NOM}	Resistor	Vout	Min	Typ	Max	Unit	Notes
34.3Ω	RON34PD	0.5 x VDDQ	0.85	1.00	1.15	RZQ/7	1,2,3,4
	RON34PU	0.5 x VDDQ	0.85	1.00	1.15	RZQ/7	1,2,3,4
40.0Ω	RON40PD	0.5 x VDDQ	0.85	1.00	1.15	RZQ/6	1,2,3,4
	RON40PU	0.5 x VDDQ	0.85	1.00	1.15	RZQ/6	1,2,3,4
48.0Ω	RON48PD	0.5 x VDDQ	0.85	1.00	1.15	RZQ/5	1,2,3,4
	RON48PU	0.5 x VDDQ	0.85	1.00	1.15	RZQ/5	1,2,3,4
Mismatch between pull-up and pull-down	MM _{PUPD}		-15.00		+15.00	%	1,2,3,4,5

Note:

1. Across entire operating temperature range, after calibration.
2. RZQ = 240Ω
3. The tolerance limits are specified after calibration with fixed voltage and temperature. For behavior of the tolerance limits if temperature or voltage changes after calibration, see following section on voltage and temperature sensitivity.
4. Pull-down and pull-up output driver impedances are recommended to be calibrated at 0.5 x VDDQ.
5. Measurement definition for mismatch between pull-up and pull-down,
MMPUPD: Measure RON_PU and RON_PD, both at 0.5 x VDDQ:

$$MMPUPD = \frac{RONPU - RONPD}{RONNOM} \times 100$$

For example, with MMPUPD(max) = 15% and RONPD = 0.85, RONPU must be less than 1.0.
6. Output driver strength measured without ODT.

Output Driver Temperature and Voltage Sensitivity

If temperature and/or voltage change after calibration, the tolerance limits widen according to the Tables shown below.

Table. Output Driver Sensitivity Definition

Resistor	Vout	Min	Max	Unit	Notes
RONPD	0.5 x VDDQ	$85 - (dRONdT \times \Delta T) - (dRONdV \times \Delta V)$	$115 + (dRONdT \times \Delta T) + (dRONdV \times \Delta V)$	%	1,2
RONPU					
RTT	0.5 x VDDQ	$85 - (dRTTdT \times \Delta T) - (dRTTdV \times \Delta V)$	$115 + (dRTTdT \times \Delta T) + (dRTTdV \times \Delta V)$	%	1,2

Note

- $\Delta T = T - T(@ \text{calibration})$, $\Delta V = V - V(@ \text{calibration})$
- dRONdT and dRONdV are not subject to production test but are verified by design and characterization.

Table. Output Driver Temperature and Voltage Sensitivity

Symbol	Parameter	Min	Max	Unit
dRONdT	RON Temperature Sensitivity	0.00	0.75	% / C
dRONdV	RON Voltage Sensitivity	0.00	0.20	% / mV
dRTTdT	RTT Temperature Sensitivity	0.00	0.75	% / C
dRTTdV	RTT Voltage Sensitivity	0.00	0.20	% / mV

RON_{PU} and RON_{PD} Characteristics without ZQ Calibration

Output driver impedance RON is defined by design and characterization as default setting.

Table. Output Driver DC Electrical Characteristics without ZQ Calibration

RON _{NOM}	Resistor	Vout	Min	Nom	Max	Unit	Notes
34.3Ω	RON40PD	0.5 x VDDQ	24	34.3	44.6	Ω	1
	RON40PU	0.5 x VDDQ	24	34.3	44.6	Ω	1
40.0Ω	RON40PD	0.5 x VDDQ	28	40	52	Ω	1
	RON40PU	0.5 x VDDQ	28	40	52	Ω	1
48.0Ω	RON48PD	0.5 x VDDQ	33.6	48	62.4	Ω	1
	RON48PU	0.5 x VDDQ	33.6	48	62.4	Ω	1
60.0Ω (optional)	RON60PD	0.5 x VDDQ	42	60	78	Ω	1
	RON60PU	0.5 x VDDQ	42	60	78	Ω	1
80.0Ω (optional)	RON80PD	0.5 x VDDQ	56	80	104	Ω	1
	RON80PU	0.5 x VDDQ	56	80	104	Ω	1

Note:

- Across entire operating temperature range, without calibration.

RZQ I-V Curve
Table. RZQ I-V Curve

Voltage(V)	RON = 240Ω (RZQ)							
	Pull-Down				Pull-Up			
	Current [mA] / RON [Ω]				Current [mA] / RON [Ω]			
	default value after ZQReset		with Calibration		default value after ZQReset		with Calibration	
	Min	Max	Min	Max	Min	Max	Min	Max
	[mA]	[mA]	[mA]	[mA]	[mA]	[mA]	[mA]	[mA]
0.00	0.00	0.00	n/a	n/a	0.00	0.00	n/a	n/a
0.05	0.17	0.35	n/a	n/a	-0.17	-0.35	n/a	n/a
0.10	0.34	0.70	n/a	n/a	-0.34	-0.70	n/a	n/a
0.15	0.50	1.03	n/a	n/a	-0.50	-1.03	n/a	n/a
0.20	0.67	1.39	n/a	n/a	-0.67	-1.39	n/a	n/a
0.25	0.83	1.73	n/a	n/a	-0.83	-1.73	n/a	n/a
0.30	0.97	2.05	n/a	n/a	-0.97	-2.05	n/a	n/a
0.35	1.13	2.39	n/a	n/a	-1.13	-2.39	n/a	n/a
0.40	1.26	2.71	n/a	n/a	-1.26	-2.71	n/a	n/a
0.45	1.39	3.01	n/a	n/a	-1.39	-3.01	n/a	n/a
0.50	1.51	3.32	n/a	n/a	-1.51	-3.32	n/a	n/a
0.55	1.63	3.63	n/a	n/a	-1.63	-3.63	n/a	n/a
0.60	1.73	3.93	2.17	2.94	-1.73	-3.93	-2.17	-2.94
0.65	1.82	4.21	n/a	n/a	-1.82	-4.21	n/a	n/a
0.70	1.90	4.49	n/a	n/a	-1.90	-4.49	n/a	n/a
0.75	1.97	4.74	n/a	n/a	-1.97	-4.74	n/a	n/a
0.80	2.03	4.99	n/a	n/a	-2.03	-4.99	n/a	n/a
0.85	2.07	5.21	n/a	n/a	-2.07	-5.21	n/a	n/a
0.90	2.11	5.41	n/a	n/a	-2.11	-5.41	n/a	n/a
0.95	2.13	5.59	n/a	n/a	-2.13	-5.59	n/a	n/a
1.00	2.17	5.72	n/a	n/a	-2.17	-5.72	n/a	n/a
1.05	2.19	5.84	n/a	n/a	-2.19	-5.84	n/a	n/a
1.10	2.21	5.95	n/a	n/a	-2.21	-5.95	n/a	n/a
1.15	2.23	6.03	n/a	n/a	-2.23	-6.03	n/a	n/a
1.20	2.25	6.11	n/a	n/a	-2.25	-6.11	n/a	n/a

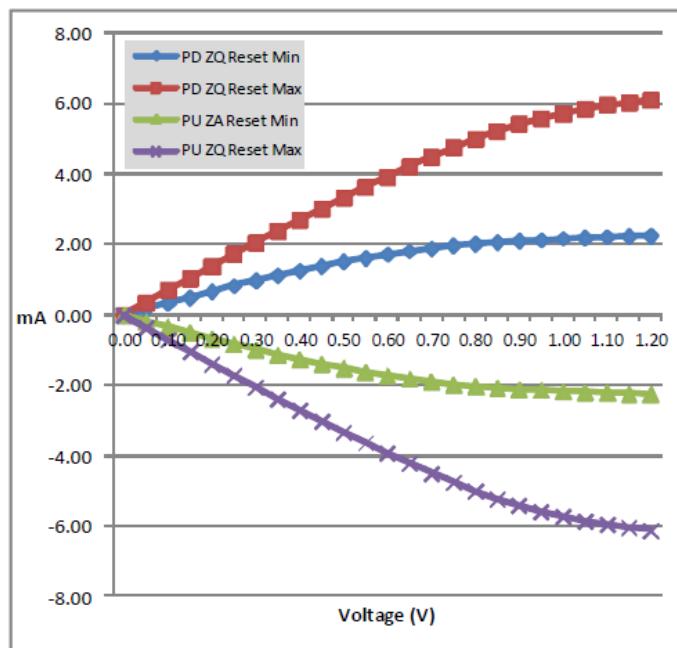


Figure. I-V Curve After ZQ Reset

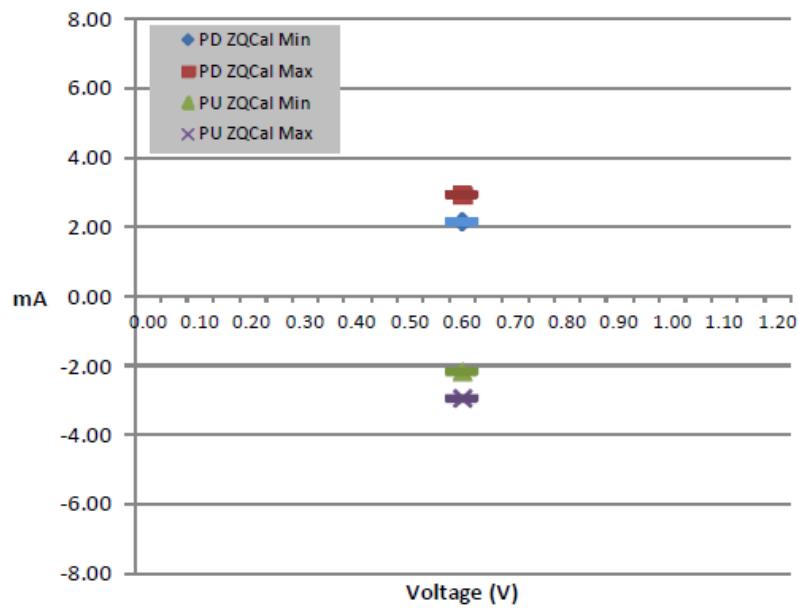


Figure. I-V Curve After Calibration

ODT Levels and I-V Characteristics

On-Die Termination effective resistance, RTT, is defined by mode register MR11[1:0]. ODT is applied to the DQ, DM, and DQS_t/DQS_c pins. A functional representation of the on-die termination is shown in the figure below. RTT is defined by the following formula:

$$RTTPU = (VDDQ - VOut) / | IOut |$$

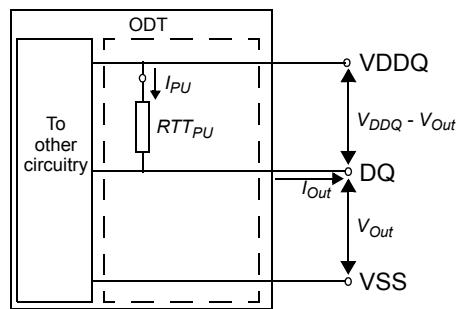


Table. ODT DC Electrical Characteristics, assuming RZQ = 240 ohm after proper ZQ calibration

RTT (ohm)	VOUT (V)	IOUT	
		Min (mA)	Max (mA)
RZQ/1	0.6	-2.17	-2.94
RZQ/2	0.6	-4.34	-5.88
RZQ/4	0.6	-8.68	-11.76

Input/Output Capacitance

Parameter	Symbol	Min	Max	Unit	Note
Input capacitance, CK_t and CK_c	CCK	0.5	1.2	pF	1,2
Input capacitance delta, CK_t and CK_c	CDCK	0	0.15	pF	1,2,3
Input capacitance, all other input-only pins	CI	0.5	1.1	pF	1,2,4
Input capacitance delta, all other input-only pins	CDI	-0.2	0.2	pF	1,2,5
Input/output capacitance, DQ, DM, DQS_t, DQS_c	CIO	1.0	1.8	pF	1,2,6,7
Input/output capacitance delta, DQS_t and DQS_c	CDDQS	0	0.2	pF	1,2,7,8
Input/output capacitance delta, DQ and DM	CDIO	-0.25	0.25	pF	1,2,7,9
Input/Output Capacitance ZQ	CZQ	0	2.0	pF	1,2

(TOPER; VDDQ = 1.14-1.3V; VDDCA = 1.14-1.3V; VDD1 = 1.7-1.95V, VDD2 = 1.14-1.3V)

Note:

1. This parameter applies to die device only (does not include package capacitance).
2. This parameter is not subject to production test. It is verified by design and characterization. The capacitance is measured according to JEP147 (Procedure for measuring input capacitance using a vector network analyzer (VNA) with VDD1, VDD2, VDDQ, VSS, VSSCA, VSSQ applied and all other pins floating).
3. Absolute value of CCK_t - CCK_c.
4. CI applies to CS_n, CKE, CA0-CA9.
5. CDI = CI - 0.5 * (CCK_t + CCK_c)
6. DM loading matches DQ and DQS.
7. MR3 I/O configuration DS OP3-OP0 = 0001B (34.3 Ohm typical)
8. Absolute value of CDQS_t and CDQS_c.
9. CDIO = CIO - 0.5 * (CDQS_t + CDQS_c) in byte-lane.

IDD Specification Parameters and Test Conditions

IDD Measurement Conditions

The following definitions are used within the IDD measurement tables:

LOW: $V_{IN} \leq V_{IL}(DC)$ MAX

HIGH: $V_{IN} \geq V_{IH}(DC)$ MIN

STABLE: Inputs are stable at a HIGH or LOW level.

SWITCHING: See tables below.

Table. Switching for CA Input Signals

Switching for CA								
	CK_t (Rising) / CK_c (Falling)	CK_t (Falling) / CK_c (Rising)						
Cycle	N		N+1		N+2		N+3	
CS_n	HIGH		HIGH		HIGH		HIGH	
CA0	HIGH	LOW	LOW	LOW	LOW	HIGH	HIGH	HIGH
CA1	HIGH	HIGH	HIGH	LOW	LOW	LOW	LOW	HIGH
CA2	HIGH	LOW	LOW	LOW	LOW	HIGH	HIGH	HIGH
CA3	HIGH	HIGH	HIGH	LOW	LOW	LOW	LOW	HIGH
CA4	HIGH	LOW	LOW	LOW	LOW	HIGH	HIGH	HIGH
CA5	HIGH	HIGH	HIGH	LOW	LOW	LOW	LOW	HIGH
CA6	HIGH	LOW	LOW	LOW	LOW	HIGH	HIGH	HIGH
CA7	HIGH	HIGH	HIGH	LOW	LOW	LOW	LOW	HIGH
CA8	HIGH	LOW	LOW	LOW	LOW	HIGH	HIGH	HIGH
CA9	HIGH	HIGH	HIGH	LOW	LOW	LOW	LOW	HIGH

Note:

1. CS_n must always be driven HIGH.
2. For each clock cycle, 50% of the CA bus is changing between HIGH and LOW once per clock for the CA bus.
3. The above pattern (N, N+1, N+2, N+3...) is used continuously during IDD measurement for IDD values that require switching on the CA bus.

Table. Switching for IDD4R

Clock	CKE	CS_n	Clock Cycle Number	Command	CA[0:2]	CA[3:9]	All DQs
Rising	HIGH	LOW	N	Read_Rising	HLH	LHLHLHL	L
Falling	HIGH	LOW	N	Read_Falling	LLL	LLLLLLL	L
Rising	HIGH	HIGH	N+1	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+1	NOP	LLL	LLLLLLL	L
Rising	HIGH	HIGH	N+2	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+2	NOP	LLL	LLLLLLL	H
Rising	HIGH	HIGH	N+3	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+3	NOP	HLH	HLHLLHL	L
Rising	HIGH	LOW	N+4	Read_Rising	HLH	HLHLLHL	H
Falling	HIGH	LOW	N+4	Read_Falling	LHH	HHHHHHH	H
Rising	HIGH	HIGH	N+5	NOP	HHH	HHHHHHH	H
Falling	HIGH	HIGH	N+5	NOP	HHH	HHHHHHH	L
Rising	HIGH	HIGH	N+6	NOP	HHH	HHHHHHH	L
Falling	HIGH	HIGH	N+6	NOP	HHH	HHHHHHH	L
Rising	HIGH	HIGH	N+7	NOP	HHH	HHHHHHH	H
Falling	HIGH	HIGH	N+7	NOP	HLH	LHLHLHL	L

Note:

1. Data strobe (DQS) is changing between HIGH and LOW every clock cycle.
2. The above pattern (N, N+1, ...) is used continuously during IDD measurement for IDD4R.

Table. Switching for IDD4W

Clock	CKE	CS_n	Clock Cycle Number	Command	CA[0:2]	CA[3:9]	All DQs
Rising	HIGH	LOW	N	Write_Rising	HLL	LHLHLHL	L
Falling	HIGH	LOW	N	Write_Falling	LLL	LLLLLLL	L
Rising	HIGH	HIGH	N+1	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+1	NOP	LLL	LLLLLLL	L
Rising	HIGH	HIGH	N+2	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+2	NOP	LLL	LLLLLLL	H
Rising	HIGH	HIGH	N+3	NOP	LLL	LLLLLLL	H
Falling	HIGH	HIGH	N+3	NOP	HLL	HLHLLHL	L
Rising	HIGH	LOW	N+4	Write_Rising	HLL	HLHLLHL	H
Falling	HIGH	LOW	N+4	Write_Falling	LHH	HHHHHHH	H
Rising	HIGH	HIGH	N+5	NOP	HHH	HHHHHHH	H
Falling	HIGH	HIGH	N+5	NOP	HHH	HHHHHHH	L
Rising	HIGH	HIGH	N+6	NOP	HHH	HHHHHHH	L
Falling	HIGH	HIGH	N+6	NOP	HHH	HHHHHHH	L
Rising	HIGH	HIGH	N+7	NOP	HHH	HHHHHHH	H
Falling	HIGH	HIGH	N+7	NOP	HLL	LHLHLHL	L

Note:

1. Data strobe (DQS) is changing between HIGH and LOW every clock cycle.
2. Data masking (DM) must always be driven LOW.
3. The above pattern (N, N+1...) is used continuously during IDD measurement for IDD4W.

IDD specifications (1/2)

Parameter / Condition	Symbol	Power Supply	1866	1600	Unit	Note
Operating one bank active-precharge current: tCK = tCKmin; tRC = tRCmin; CKE is HIGH; CS_n is HIGH between valid commands; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD0 ₁	VDD1	12		mA	
	IDD0 ₂	VDD2	48		mA	
	IDD0 _{IN}	VDDCA VDDQ	15		mA	4
Idle power-down standby current: tCK = tCKmin; CKE is LOW; CS_n is HIGH; All banks are idle; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD2P ₁	VDD1	1.8		mA	
	IDD2P ₂	VDD2	6		mA	
	IDD2P _{IN}	VDDCA VDDQ	0.4		mA	4,8
Idle power-down standby current with clock stop: CK_t = LOW, CK_c = HIGH; CKE is LOW; CS_n is HIGH; All banks are idle; CA bus inputs are stable; Data bus inputs are stable ODT disabled	IDD2PS ₁	VDD1	1.8		mA	
	IDD2PS ₂	VDD2	6		mA	
	IDD2PS _{IN}	VDDCA VDDQ	0.4		mA	4,8
Idle non-power-down standby current: tCK = tCKmin; CKE is HIGH; CS_n is HIGH; All banks are idle; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD2N ₁	VDD1	4		mA	
	IDD2N ₂	VDD2	20		mA	
	IDD2N _{IN}	VDDCA VDDQ	20		mA	4
Idle non-power-down standby current with clock stopped: CK_t = LOW; CK_c = HIGH; CKE is HIGH; CS_n is HIGH; All banks are idle; CA bus inputs are stable; Data bus inputs are stable ODT disabled	IDD2NS ₁	VDD1	2		mA	
	IDD2NS ₂	VDD2	12		mA	
	IDD2NS _{IN}	VDDCA VDDQ	20		mA	4
Active power-down standby current: tCK = tCKmin; CKE is LOW; CS_n is HIGH; One bank is active; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD3P ₁	VDD1	3		mA	
	IDD3P ₂	VDD2	10.5		mA	
	IDD3P _{IN}	VDDCA VDDQ	0.4		mA	4,8
Active power-down standby current with clock stop: CK = LOW, CK# = HIGH; CKE is LOW; CS_n is HIGH; One bank is active; CA bus inputs are stable; Data bus inputs are stable ODT disabled	IDD3PS ₁	VDD1	3		mA	
	IDD3PS ₂	VDD2	10.5		mA	
	IDD3PS _{IN}	VDDCA VDDQ	0.4		mA	4,8
Active non-power-down standby current: tCK = tCKmin; CKE is HIGH; CS_n is HIGH; One bank is active; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD3N ₁	VDD1	3		mA	
	IDD3N ₂	VDD2	15		mA	
	IDD3N _{IN}	VDDCA VDDQ	20		mA	4

IDD specifications (2/2)

Parameter / Test Condition	Symbol	Power Supply	1866	1600	Unit	Note
Active non-power-down standby current with clock stopped: CK = LOW, CK# = HIGH CKE is HIGH; CS_n is HIGH; One bank is active; CA bus inputs are stable; Data bus inputs are stable ODT disabled	IDD3NS ₁	VDD1	3		mA	
	IDD3NS ₂	VDD2	15		mA	
	IDD3NS _{IN}	VDDCA VDDQ	20		mA	4
Operating burst READ current: tCK = tCKmin; CS_n is HIGH between valid commands; One bank is active; BL = 8; RL = RL (MIN); CA bus inputs are switching; 50% data change each burst transfer ODT disabled	IDD4R ₁	VDD1	22	20	mA	
	IDD4R ₂	VDD2	360	320	mA	
	IDD4R _{IN}	VDDCA	20	20	mA	
	IDD4R _Q	VDDQ	260	200	mA	5
Operating burst WRITE current: tCK = tCKmin; CS_n is HIGH between valid commands; One bank is active; BL = 8; WL = WLmin; CA bus inputs are switching; 50% data change each burst transfer ODT disabled	IDD4W ₁	VDD1	22	20	mA	
	IDD4W ₂	VDD2	380	340	mA	
	IDD4W _{IN}	VDDCA VDDQ	50	50	mA	4
	IDD5 ₁	VDD1	68		mA	
All-bank REFRESH burst current: tCK = tCKmin; CKE is HIGH between valid commands; tRC = tRFCabmin; Burst refresh; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD5 ₂	VDD2	255		mA	
	IDD5 _{IN}	VDDCA VDDQ	17		mA	4
	IDD5ab ₁	VDD1	5.5		mA	
All-bank REFRESH average current: tCK = tCKmin; CKE is HIGH between valid commands; tRC = tREFI; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD5ab ₂	VDD2	24		mA	
	IDD5ab _{IN}	VDDCA VDDQ	17		mA	4
	IDD5pb ₁	VDD1	6		mA	
Per-bank REFRESH average current: tCK = tCKmin; CKE is HIGH between valid commands; tRC = tREFI/8; CA bus inputs are switching; Data bus inputs are stable ODT disabled	IDD5pb ₂	VDD2	25.5		mA	
	IDD5pb _{IN}	VDDCA VDDQ	17		mA	4
	IDD6 ₁	VDD1	6		mA	6
Self refresh current (-30°C to +85°C): CK_t = LOW, CK_c = HIGH; CKE is LOW; CA bus inputs are stable; Data bus inputs are stable Maximum 1x self refresh rate ODT disabled	IDD6 ₂	VDD2	16		mA	6
	IDD6 _{IN}	VDDCA VDDQ	0.4		mA	4,6,8

Note:

1. Published IDD values are the maximum of the distribution of the arithmetic mean at 85°C.
2. IDD current specifications are tested after the device is properly initialized.
3. The 1x self refresh rate is the rate at which the device is refreshed internally during self refresh, before going into the elevated temperature range.
4. Measured currents are the summation of VDDQ and VDDCA.
5. Guaranteed by design with output load = 5 pF and RON = 40 ohm.
6. This is the general definition that applies to full-array SELF REFRESH.
7. For all IDD measurements, VIHCKE = 0.65 x VDDCA, VILCKE = 0.35 x VDDCA.
8. Currents values are measured in one CS enabled condition.
9. DPD(Deep Power down) function is an optional feature. Please contact SK hynix for more information to use this feature.

IDD6 Partial Array Self Refresh Current

Temp. (°C)	Memory Array				Unit
	8 Banks	4 Banks	2 Banks	1 Bank	
25	0.70 / 1.10 / 0.10	0.70 / 0.90 / 0.10	0.60 / 0.64 / 0.10	0.60 / 0.50 / 0.10	mA
85	6.00 / 16.00 / 0.40	3.00 / 8.00 / 0.40	3.00 / 5.00 / 0.40	2.00 / 4.00 / 0.40	mA

Note:

1. IDD6 85°C is the maximum, and IDD6 25°C is typical value.
2. IDD6 currents are measured using bank-masking only.
3. IDD values published are the maximum of the distribution of the arithmetic mean.
4. All values are based on dual dies.

AC TIMING PARAMETERS (1/5)

Parameter	Symbol	min max	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Unit	Note
Maximum clock Frequency		-	933	800	667	MHz	
Clock Timing							
Average Clock Period	tCK(avg)	min	1.071	1.25	1.5	ns	
		max		100			
Average high pulse width	tCH(avg)	min	0.45			tCK(avg)	
		max	0.55				
Average low pulse width	tCL(avg)	min	0.45			tCK(avg)	
		max	0.55				
Absolute Clock Period	tCK(abs)	min	tCK(avg)min + tJIT(per)min			ns	
Absolute clock HIGH pulse width (with allowed jitter)	tCH(abs), allowed	min	0.43			tCK(avg)	
		max	0.57				
Absolute clock LOW pulse width (with allowed jitter)	tCL(abs), allowed	min	0.43			tCK(avg)	
		max	0.57				
Clock Period Jitter (with allowed jitter)	tJIT(per), allowed	min	-60	-70	-80	ps	
		max	60	70	80		
Maximum Clock Jitter between two consecutive clock cycles (with allowed jitter)	tJIT(cc), allowed	max	120	140	160	ps	
Duty cycle Jitter (with allowed jitter)	tJIT(duty), allowed	min	min((tCH(abs)min - tCH(avg)min), (tCL(abs)min - tCL(avg)min)) * tCK(avg)			ps	
		max	max((tCH(abs)max - tCH(avg)max), (tCH(abs)max - tCL(avg)max)) * tCK(avg)				
Cumulative error across 2 cycles	tERR(2per), allowed	min	-88	-103	-118	ps	
		max	88	103	118		
Cumulative error across 3 cycles	tERR(3per), allowed	min	-105	-122	-140	ps	
		max	105	122	140		
Cumulative error across 4 cycles	tERR(4per), allowed	min	-117	-136	-155	ps	
		max	117	136	155		
Cumulative error across 5 cycles	tERR(5per), allowed	min	-126	-147	-168	ps	
		max	126	147	168		
Cumulative error across 6 cycles	tERR(6per), allowed	min	-133	-155	-177	ps	
		max	133	155	177		
Cumulative error across 7 cycles	tERR(7per), allowed	min	-139	-163	-186	ps	
		max	139	163	186		
Cumulative error across 8 cycles	tERR(8per), allowed	min	-145	-169	-193	ps	
		max	145	169	193		
Cumulative error across 9 cycles	tERR(9per), allowed	min	-150	-175	-200	ps	
		max	150	175	200		
Cumulative error across 10 cycles	tERR(10per), allowed	min	-154	-180	-205	ps	
		max	154	180	205		

AC TIMING PARAMETERS (2/5)

Parameter	Symbol	min max	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Unit	Note
Clock Timing (continued)							
Cumulative error across 11 cycles	tERR(11per), allowed	min	-158	-184	-210	ps	
		max	158	184	210		
Cumulative error across 12 cycles	tERR(12per), allowed	min	-161	-188	-215	ps	
		max	161	188	215		
Cumulative error across n cycles (n = 13, 14, ..., 20)	tERR(nper), allowed	min	tERR(nper),allowed min = (1 + 0.68ln(n)) * tJIT(per),allowed min			ps	
		max	tERR(nper),allowed max = (1 + 0.68ln(n)) * tJIT(per),allowed max				
ZQ Calibration Parameters							
Initialization Calibration Time	tZQINIT	min	1			us	
Long Calibration Time	tZQCL	min	360			ns	
Short Calibration Time	tZQCS	min	90			ns	
Calibration Reset Time	tZQRESET	min	max(50ns, 3nCK)			ns	
Read Parameters							3
DQS output access time from CK/CK#	tDQSCK	min	2.5			ns	
		max	5.5				
DQSCK Delta short	tDQSCKDS	max	190	220	265	ps	4
DQSCK Delta Medium	tDQSCKDM	max	435	511	593	ps	5
DQSCK Delta Long	tDQSCKDL	max	525	614	733	ps	6
DQS-DQ skew	tDQSQ	max	115	135	165	ps	
DQS Output High Pulse Width	tQSH	min	tCH(abs) - 0.05			tCK(avg)	
DQS Output Low Pulse Width	tQL	min	tCL(abs) - 0.05			tCK(avg)	
DQ/DQS output hold time from DQS	tQH	min	MIN (tQSH, tQL)			ps	
Read preamble	tRPRE	min	0.9			tCK(avg)	7,10
Read postamble	tRPST	min	0.3			tCK(avg)	7,11
DQS low-Z from clock	tLZ(DQS)	min	tDQSCK(min) - 300			ps	7
DQ low-Z from clock	tLZ(DQ)	min	tDQSCK(min) - 300			ps	7
DQS high-Z from clock	tHZ(DQS)	max	tDQSCK(max) - 100			ps	7
DQ high-Z from clock	tHZ(DQ)	max	tDQSCK(max) + (1.4 x tDQSQ-max)			ps	7

AC TIMING PARAMETERS (3/5)

Parameter	Symbol	min max	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Unit	Note
Write Parameters							3
DQ and DM input setup time (Vref based)	tDS	min	130	150	175	ps	
DQ and DM input hold time (Vref based)	tDH	min	130	150	175	ps	
DQ and DM input pulse width	tDIPW	min	0.35			tCK(avg)	
Write command to 1st DQS latching transition	tDQSS	min	0.75			tCK(avg)	
		max	1.25				
DQS input high-level width	tDQSH	min	0.4			tCK(avg)	
DQS input low-level width	tDQSL	min	0.4			tCK(avg)	
DQS falling edge to CK setup time	tDSS	min	0.2			tCK(avg)	
DQS falling edge hold time from CK	tDSH	min	0.2			tCK(avg)	
Write postamble	tWPST	min	0.4			tCK(avg)	
Write preamble	tWPRE	min	0.8			tCK(avg)	
CKE Input Parameters							
CKE min. pulse width (high/low pulse width)	tCKE	min	max(7.5ns, 3nCK)			ns	
CKE input setup time	tISCKE	min	0.25			tCK(avg)	12
CKE input hold time	tIHCKE	min	0.25			tCK(avg)	13
Command path disable delay	tCPDED	min	2			tCK(avg)	
Command Address Input Parameters							3
Address and control input setup time	tISCA	min	130	150	175	ps	14
Address and control input hold time	tIHCA	min	130	150	175	ps	14
CS_n input setup time	tISCS	min	230	270	290	ps	14
CS_n input hold time	tIHCS	min	230	270	290	ps	14
Address and control input pulse width	tIPWCA	min	0.35			tCK(avg)	
CS_n input pulse width	tIPWCS	min	0.7			tCK(avg)	
Boot Parameters (10MHz-55MHz)							15,16, 17
Clock Cycle Time	tCKb	min	18			ns	
		max	100				
CKE Input Setup Time	tISCKEb	min	2.5			ns	
CKE Input Hold Time	tIHCKEb	min	2.5			ns	
Address & Control Input Setup Time	tISb	min	1150			ps	
Address & Control Input Hold Time	tIHb	min	1150			ps	
DQS Output Data Access Time from CK/CK#	tDQSCkb	min	2.0			ns	
		max	10.0				
Data Strobe Edge to Output Data Edge tDQSqb	tDQSqb	max	1.2			ns	
Mode Register Parameters							
MODE REGISTER Write command period	tMRW	min	10			tCK(avg)	
MODE REGISTER Read command period	tMRR	min	4			tCK(avg)	
Additional time after tXP has expired until MRR command may be issued	tMRRI	min	tRCD(MIN)			ns	

AC TIMING PARAMETERS (4/5)

Parameter	Symbol	min max	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Unit	Note
Core Parameters							18
Read Latency	RL	min	14	12	10	tCK(avg)	
Write Latency (Set A)	WL	min	8	6	6	tCK(avg)	
Write Latency (Set B)	WL	min	11	9	8	tCK(avg)	
ACTIVE to ACTIVE command period	tRC	min	tRAS+tRPab (with all-bank Precharge) tRAS+tRPpb (with per-bank Precharge)		ns		
CKE min. pulse width during Self-Refresh (low pulse width during Self-Refresh)	tCKESR	min	max(15ns, 3nCK)		ns		
Self refresh exit to next valid command delay	tXSR	min	max(tRFCab +10ns, 2nCK)		ns		
Exit power down to next valid command delay	tXP	min	max(7.5ns, 3nCK)		ns		
CAS to CAS delay	tCCD	min	4		tCK(avg)		
Internal Read to Precharge command delay	tRTP	min	max(7.5ns, 4nCK)		ns		
RAS to CAS Delay	tRCD	min	max(18ns, 3nCK)		ns		
Row Precharge Time (single bank)	tRPpb	min	max(18ns ,3nCK)		ns		
Row Precharge Time (all banks) - 8-bank	tRPab	min	max(21ns, 3nCK)		ns		
Row Active Time	tRAS	min	max(42ns, 3nCK)		ns		
		max	70,000				
Write Recovery Time	tWR	min	max(15ns, 4nCK)		ns		
Internal Write to Read Command Delay	tWTR	min	max(7.5ns, 4nCK)		ns		
Active bank A to Active bank B	tRRD	min	max(10ns, 2nCK)		ns		
Four Bank Activate Window	tFAW	min	max(50ns, 8nCK)		ns		
ODT Parameters							
Asynchronous RTT turn-on delay from ODT input	tODTon	min	1.75		ns		
		max			3.5		
Asynchronous RTT turn-off delay from ODT input	tODToff	min	1.75		ns		
		max			3.5		
Automatic RTT turn-on delay after READ data	tAODTon	max	tDQSCKmax + 1.4 * tDQSQ-max + tCK(avg,min)		ps		
Automatic RTT turn-off delay after READ data	tAODToff	min	tDQSCKmin - 300		ps		
RTT disable delay from power down and self-refresh	tODTd	min	12		ns		
RTT enable delay from power down and self refresh exit	tODTe	max	12		ns		

AC TIMING PARAMETERS (5/5)

Parameter	Symbol	min max	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Unit	Note
CA Training Parameters							
First CA calibration command after CA calibration mode is programmed	tCAMRD	min		20		tCK(avg)	
First CA calibration command after CKE is LOW	tCAENT	min		10		tCK(avg)	
CA calibration exit command after CKE is HIGH	tCAEXT	min		10		tCK(avg)	
CKE LOW after CA calibration mode is programmed	tCACKEL	min		10		tCK(avg)	
CKE HIGH after the last CA calibration results are driven	tCACKEH	min		10		tCK(avg)	
Data out delay after CA training calibration command is programmed	tADR	max		20		ns	
MRW CA exit command to DQ tri-state	tMRZ	min		3		ns	
CA calibration command to CA calibration command delay	tCACD	min	RU(tADR+2*tCK)			tCK(avg)	
Write Leveling Parameters							
DQS_t/DQS_c delay after write leveling mode is programmed	tWLQSEN	min	25			ns	
		max	-				
First DQS_t/DQS_c edge after write leveling mode is programmed	tWLMD	min	40			ns	
		max	-				
Write leveling output delay	tWLO	min	0			ns	
		max	20				
Write leveling hold time	tWLH	min	150	175	205	ps	
Write leveling setup time	tWLS	min	150	175	205	ps	
Mode register set command delay	tMRD	min	Max(14ns, 10nCK)			ns	
		max	-				
Temperature De-Rating							17
tDQSCK De-Rating	tDQSCK (Derated)	max	5620			ps	
Core Timings Temperature De-Rating	tRCD (Derated)	min	tRCD + 1.875			ns	
	tRC (Derated)	min	tRC + 1.875			ns	
	tRAS (Derated)	min	tRAS + 1.875			ns	
	tRP (Derated)	min	tRP + 1.875			ns	
	tRRD (Derated)	min	tRRD + 1.875			ns	

Note:

1. Frequency values are for reference only. Clock cycle time (tCK) is used to determine device capabilities.
2. All AC timings assume an input slew rate of 2 V/ns.
3. Measured with 4V/ns differential CK_t/CK_c slew rate and nominal VIX.
4. All timing and voltage measurements are defined 'at the ball'.
5. READ, WRITE, and input setup and hold values are referenced to VREF.
6. tDQSCKDS is the absolute value of the difference between any two tDQSCK measurements (in a byte lane) within a contiguous sequence of bursts in a 160ns rolling window. tDQSCKDS is not tested and is guaranteed by design. Temperature drift in the system is < 10 °C/s. Values do not include clock jitter.
7. tDQSCKDM is the absolute value of the difference between any two tDQSCK measurements (in a byte lane) within a 1.6µs rolling window. tDQSCKDM is not tested and is guaranteed by design. Temperature drift in the system is < 10 °C/s. Values do not include clock jitter.
8. tDQSCKDL is the absolute value of the difference between any two tDQSCK measurements (in a byte lane) within a 32ms rolling window. tDQSCKDL is not tested and is guaranteed by design. Temperature drift in the system is < 10 °C/s. Values do not include clock jitter.
9. For LOW-to-HIGH and HIGH-to-LOW transitions, the timing reference is at the point when the signal crosses the transition threshold (VTT). tHZ and tLZ transitions occur in the same access time (with respect to clock) as valid data transitions. These parameters are not referenced to a specific voltage level but to the time when the device output is no longer driving (for tRPST, tHZ(DQS) and tHZ(DQ)), or begins driving (for tRPRE, tLZ(DQS), tLZ(DQ)). Figure shows a method to calculate the point when device is no longer driving tHZ(DQS) and tHZ(DQ), or begins driving tLZ(DQS), tLZ(DQ) by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent.
10. Output Transition Timing

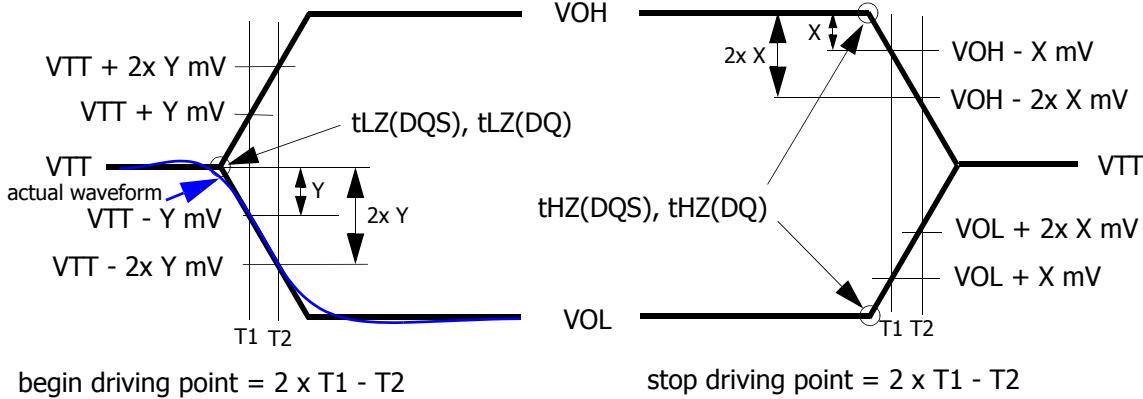


Figure. HSUL_12 Driver Output Reference Load for Timing and Slew Rate

11. The parameters tLZ(DQS), tLZ(DQ), tHZ(DQS), and tHZ(DQ) are defined as single-ended. The timing parameters tRPST and tRPRE are determined from the differential signal DQS/DQS#.
12. Measured from the point when DQS_t/DQS_c begins driving the signal to the point when DQS_t/DQS_c begins driving the first rising strobe edge.
13. Measured from the last falling strobe edge of DQS_t/DQS_c to the point when DQS_t/DQS_c finishes driving the signal.
14. CKE input setup time is measured from CKE reaching a HIGH/LOW voltage level to CK_t/CK_c crossing.
15. CKE input hold time is measured from CK_t/CK_c crossing to CKE reaching a HIGH/LOW voltage level.
16. Input set-up/hold time for signal (CA[9:0], CS_n).
17. To ensure device operation before the device is configured, a number of AC boot-timing parameters are defined in this table. Boot parameter symbols have the letter b appended (for example, tCK during boot is tCKb).
18. The LPDDR3 device will set some mode register default values upon receiving a RESET (MRW) command as specified in "Mode Register Definition".
19. The output skew parameters are measured with default output impedance settings using the reference load.
20. The minimum tCK column applies only when tCK is greater than 6ns.

CA and CS_n Setup, Hold and Derating

For all input signals (CA and CS_n) the total tIS (setup time) and tIH (hold time) required is calculated by adding the data sheet tIS(base) and tIH(base) value to the ΔtIS and ΔtIH derating value respectively. Example: tIS (total setup time) = tIS(base) + ΔtIS

Setup (tIS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VREF(DC) and the first crossing of VIH(AC)min. Setup (tIS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VREF(DC) and the first crossing of VIL(AC)max. If the actual signal is always earlier than the nominal slew rate line between shaded 'VREF(DC) to ac region', use nominal slew rate for derating value. If the actual signal is later than the nominal slew rate line anywhere between shaded 'VREF(DC) to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value.

Hold (tIH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VIL(DC)max and the first crossing of VREF(DC). Hold (tIH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VIH(DC)min and the first crossing of VREF(DC). If the actual signal is always later than the nominal slew rate line between shaded 'DC to VREF(DC) region', use nominal slew rate for derating value. If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'DC to VREF(DC) region', the slew rate of a tangent line to the actual signal from the dc level to VREF(DC) level is used for derating value.

For a valid transition the input signal has to remain above/below VIH/L(AC) for some time tVAC.

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached VIH/L(AC) at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach VIH/L(AC).

For slew rates in between the values listed in Table, the derating values may obtained by linear interpolation. These values are typically not subject to production test. They are verified by design and characterization.

Table. CA Setup and Hold Base-Values

unit [ps]	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Reference
tIS(base)	-	75	100	VIH/L(AC)=VREF(DC)+/-150mV
tIS(base)	62.5	-	-	VIH/L(AC)=VREF(DC)+/-135mV
tIH(base)	80	100	125	VIH/L(DC)=VREF(DC)+/-100mV

Note 1: AC/DC referenced for 2V/ns CA slew rate and 4V/ns differential CK_t/CK_c slew rate.

Table. CS_n Setup and Hold Base-Values

unit [ps]	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Reference
tIS(base)	-	195	215	VIH/L(AC)=VREF(DC)+/-150mV
tIS(base)	162.5	-	-	VIH/L(AC)=VREF(DC)+/-135mV
tIH(base)	180	220	240	VIH/L(DC)=VREF(DC)+/-100mV

Note 1: AC/DC referenced for 2V/ns CS_n slew rate and 4V/ns differential CK_t/CK_c slew rate.

Table. Derating values tIS/tIH - ac/dc based AC150

		ΔtISCA, ΔtIHCA, ΔtISCS, ΔtIHCS derating in [ps] AC/DC based AC150 Threshold -> VIH(ac)=VREF(dc)+150mV, VIL(ac)=VREF(dc)-150mV DC100 Threshold -> VIH(dc)=VREF(dc)+100mV, VIL(dc)=VREF(dc)-100mV											
		8.0 V/ns		7.0 V/ns		6.0 V/ns		5.0 V/ns		4.0 V/ns		3.0 V/ns	
		ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH
CA, CS_n Slew rate V/ns	4.0	38	25	38	25	38	25	38	25	38	25		
	3.0			25	17	25	17	25	17	25	17	38	29
	2.0					0	0	0	0	0	0	13	13
	1.5							-25	-17	-25	-17	-12	-4

Note 1: Cell contents shaded in red are defined as 'not supported'

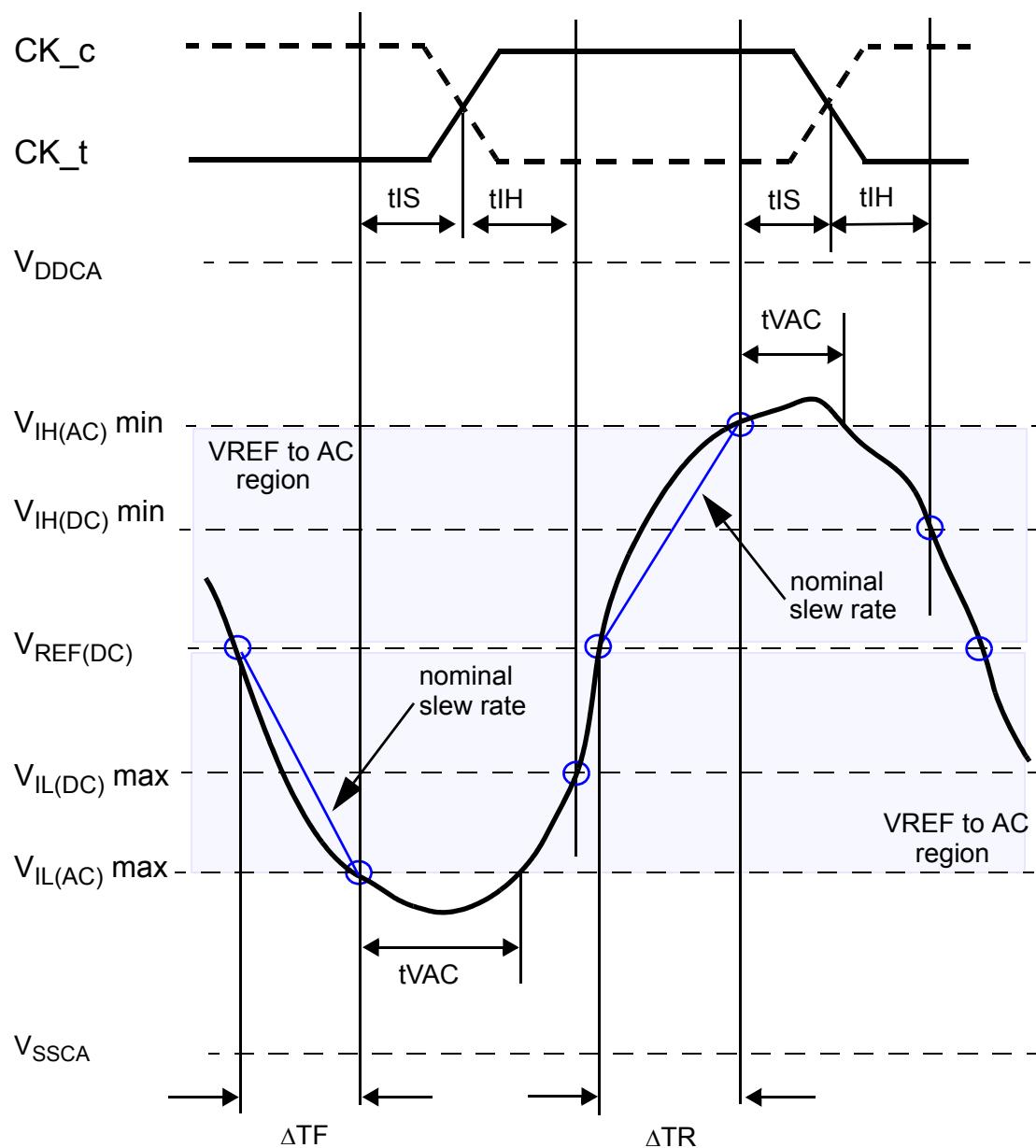
Table. Derating values tIS/tIH - ac/dc based AC135

		ΔtISCA, ΔtIHCA, ΔtISCS, ΔtIHCS derating in [ps] AC/DC based AC135 Threshold -> VIH(ac)=VREF(dc)+135mV, VIL(ac)=VREF(dc)-135mV DC100 Threshold -> VIH(dc)=VREF(dc)+100mV, VIL(dc)=VREF(dc)-100mV											
		8.0 V/ns		7.0 V/ns		6.0 V/ns		5.0 V/ns		4.0 V/ns		3.0 V/ns	
		ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH	ΔtIS	ΔtIH
CA, CS_n Slew rate V/ns	4.0	34	25	34	25	34	25	34	25	34	25		
	3.0			23	17	23	17	23	17	23	17	34	29
	2.0					0	0	0	0	0	0	11	13
	1.5							-23	-17	-23	-17	-12	-4

Note 1: Cell contents shaded in red are defined as 'not supported'

Table. Required time t_{VAC} above VIH(ac) {below VIL(ac)} for valid transition

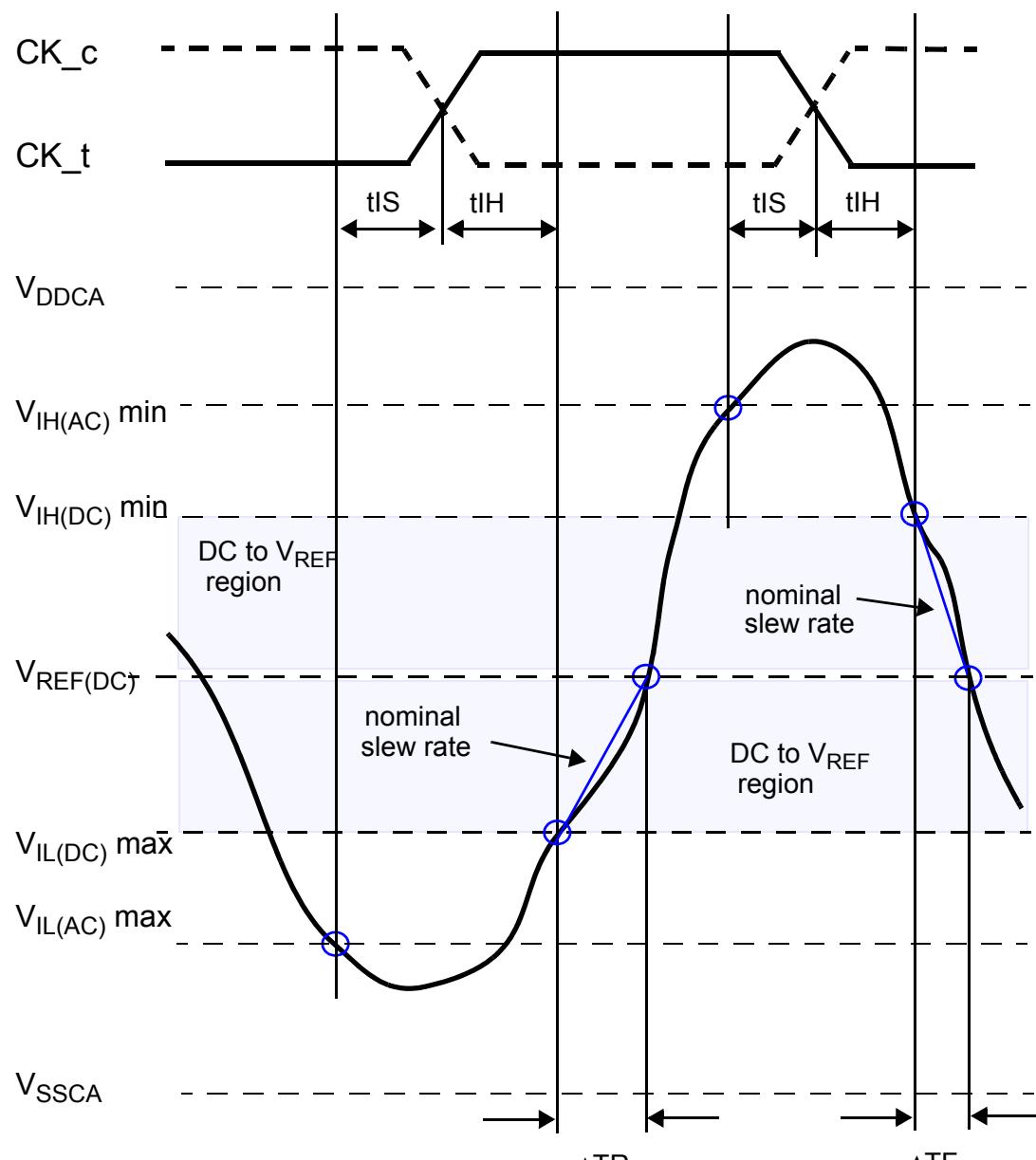
Slew Rate [V/ns]	t _{VAC} [ps] @135mV		t _{VAC} [ps] @150mV		t _{VAC} [ps] @150mV	
	1866Mbps		1600Mbps		1333Mbps	
	MIN	MAX	MIN	MAX	MIN	MAX
> 4.0	40	-	48	-	58	-
4.0	40	-	48	-	58	-
3.5	39	-	46	-	56	-
3.0	36	-	43	-	53	-
2.5	33	-	40	-	50	-
2.0	29	-	35	-	45	-
1.5	21	-	27	-	37	-
<1.5	21	-	27	-	37	-



$$\text{Setup Slew Rate}_{\text{Falling Signal}} = \frac{V_{\text{REF(DC)}} - V_{\text{IL(AC)}\max}}{\Delta TF}$$

$$\text{Setup Slew Rate}_{\text{Rising Signal}} = \frac{V_{\text{IH(AC)}\min} - V_{\text{REF(DC)}}}{\Delta TR}$$

Figure. Illustration of nominal slew rate and t_{VAC} for setup time t_{IS} for CA and CS_n with respect to clock



$$\text{Hold Slew Rate}_{\text{Rising Signal}} = \frac{V_{\text{REF(DC)}} - V_{\text{IL(DC) max}}}{\Delta \text{TR}}$$

$$\text{Hold Slew Rate}_{\text{Falling Signal}} = \frac{V_{\text{IH(DC) min}} - V_{\text{REF(DC)}}}{\Delta \text{TF}}$$

Figure. Illustration of nominal slew rate for hold time tIH for CA and CS_n with respect to clock

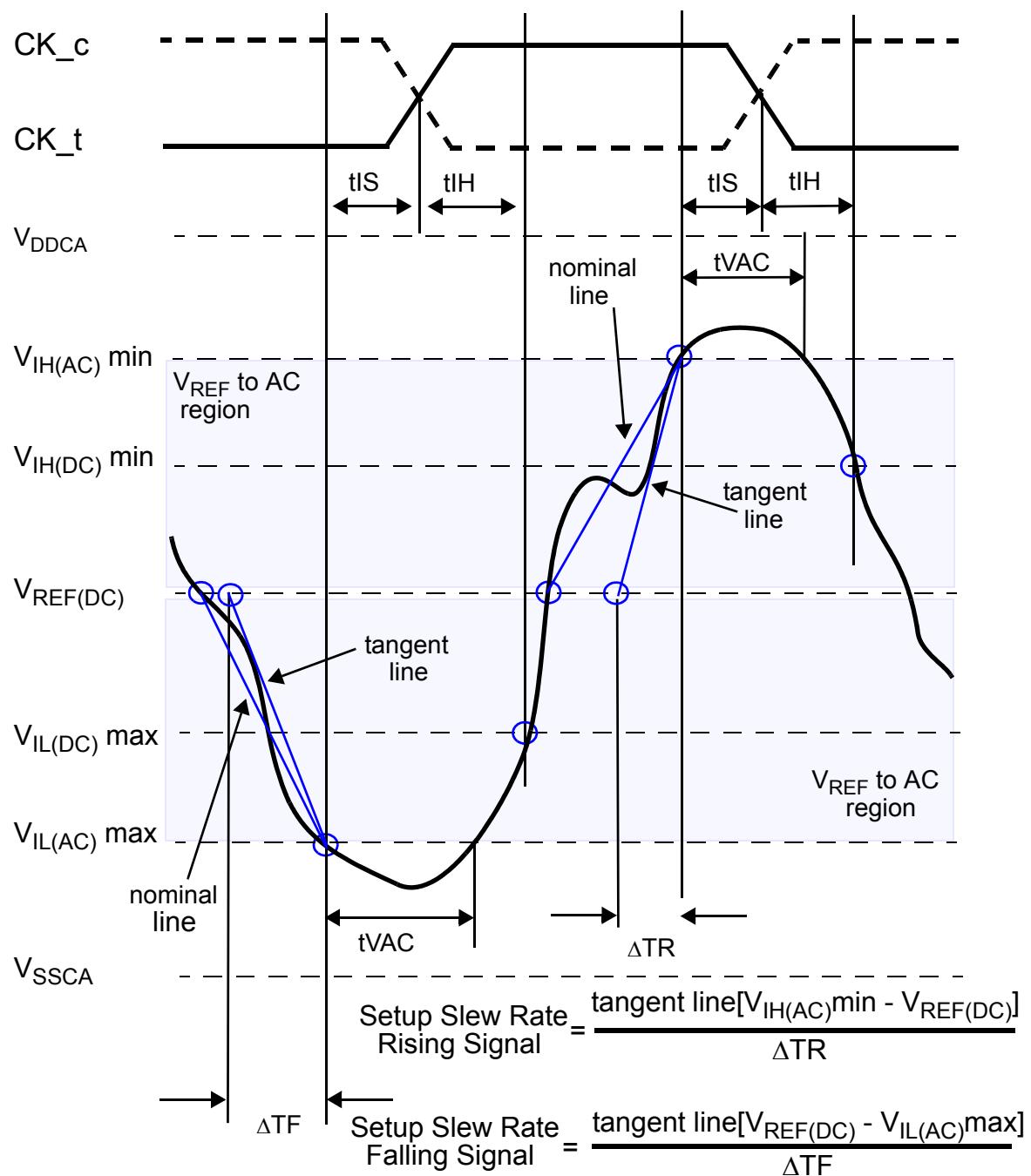
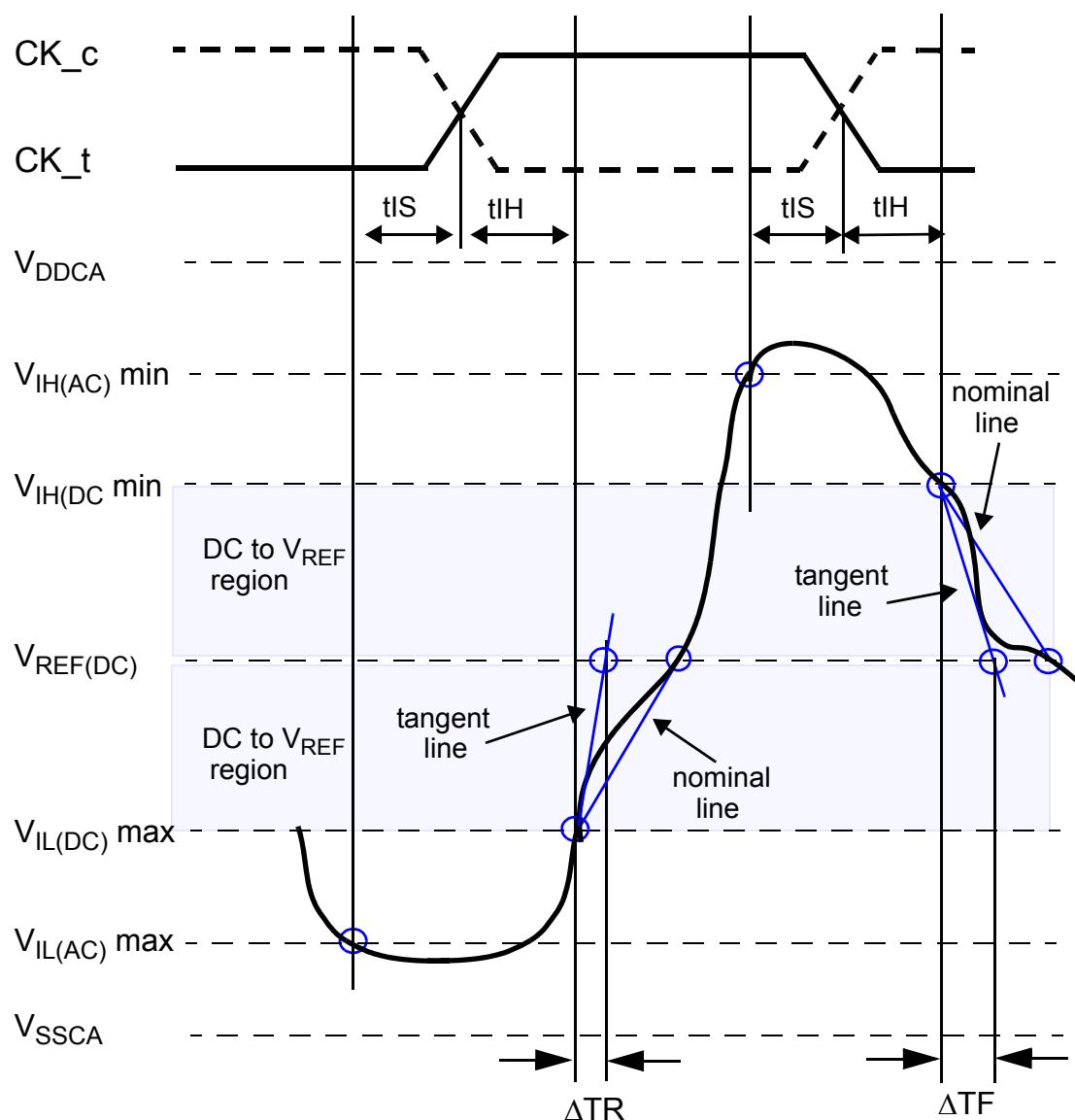


Figure. Illustration of tangent line for setup time tIS for CA and CS_n with respect to clock



$$\text{Hold Slew Rate}_{\text{Rising Signal}} = \frac{\text{tangent line } [V_{\text{REF}(\text{DC})} - V_{\text{IL}(\text{DC})\text{max}}]}{\Delta \text{TR}}$$

$$\text{Hold Slew Rate}_{\text{Falling Signal}} = \frac{\text{tangent line } [V_{\text{IH}(\text{DC})\text{min}} - V_{\text{REF}(\text{DC})}]}{\Delta \text{TF}}$$

Figure. Illustration of tangent line for hold time t_{IH} for CA and CS_n with respect to clock

Data Setup, Hold and Slew Rate Derating

For all input signals (DQ, DM) the total tDS (setup time) and tDH (hold time) required is calculated by adding the data sheet tDS(base) and tDH(base) value to the Δ tDS and Δ tDH derating value respectively. Example: tDS (total setup time) = tDS(base) + Δ tDS.

Setup (tDS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VREF(DC) and the first crossing of VIH(AC)min. Setup (tDS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VREF(DC) and the first crossing of VIL(AC)max. If the actual signal is always earlier than the nominal slew rate line between shaded 'VREF(DC) to ac region', use nominal slew rate for derating value. If the actual signal is later than the nominal slew rate line anywhere between shaded 'VREF(DC) to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value.

Hold (tDH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VIL(AC)max and the first crossing of VREF(DC). Hold (tDH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VIH(DC)min and the first crossing of VREF(DC). If the actual signal is always later than the nominal slew rate line between shaded 'dc level to VREF(DC) region', use nominal slew rate for derating value. If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'dc to VREF(DC) region', the slew rate of a tangent line to the actual signal from the dc level to VREF(DC) level is used for derating value.

For a valid transition the input signal has to remain above/below VIH/IL(AC) for some time tVAC.

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached VIH/IL(AC) at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach VIH/IL(AC).

For slew rates in between the values listed in the tables the derating values may be obtained by linear interpolation. These values are typically not subject to production test. They are verified by design and characterization.

Table. Data Setup and Hold Base-Values

unit [ps]	LPDDR3 1866	LPDDR3 1600	LPDDR3 1333	Reference
tDS(base)	-	75	100	VIH/L(AC)=VREF(DC)+/-150mV
tDS(base)	62.5	-	-	VIH/L(AC)=VREF(DC)+/-135mV
tDH(base)	80	100	125	VIH/L(DC)=VREF(DC)+/-100mV

Note 1: AC/DC referenced for 2V/ns DQ, DM slew rate and 4V/ns differential DQS_t-DQS_c slew rate.

Table. Derating values LPDDR3 tDS/tDH - AC/DC based AC150

		ΔtDS, ΔtDH derating in [ps] AC/DC based AC150 Threshold -> VIH(ac)=VREF(dc)+150mV, VIL(ac)=VREF(dc)-150mV DC100 Threshold -> VIH(dc)=VREF(dc)+100mV, VIL(dc)=VREF(dc)-100mV											
		8.0 V/ns		7.0 V/ns		6.0 V/ns		5.0 V/ns		4.0 V/ns		3.0 V/ns	
		ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH
DQ,DM Slew rate V/ns	4.0	38	25	38	25	38	25	38	25	38	25		
	3.0			25	17	25	17	25	17	25	17	38	29
	2.0					0	0	0	0	0	0	13	13
	1.5							-25	-17	-25	-17	-12	-4

Note 1: Cell contents shaded in red are defined as 'not supported'

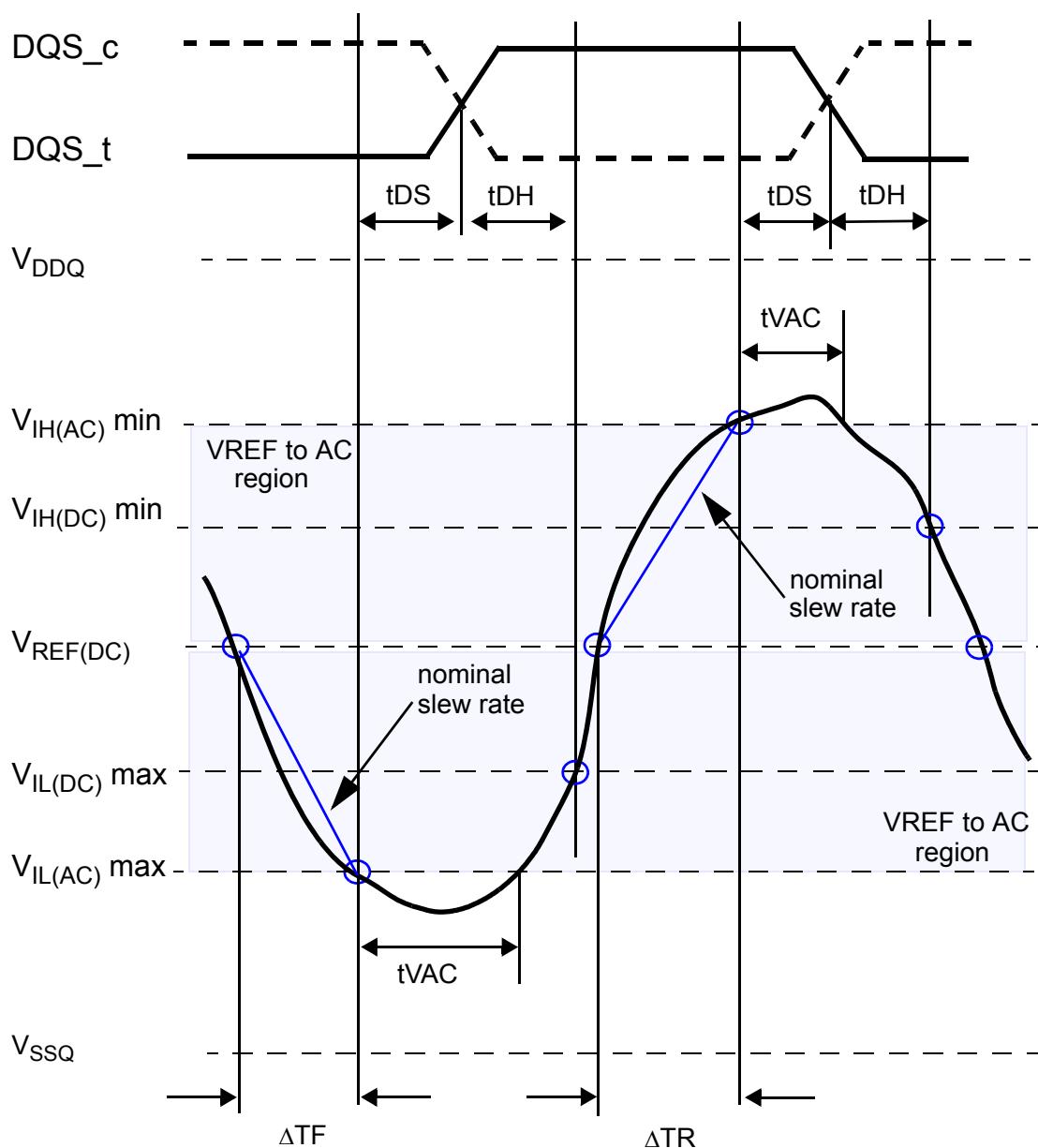
Table. Derating values LPDDR3 tDS/tDH - AC/DC based AC135

		ΔtDS, ΔtDH derating in [ps] AC/DC based AC135 Threshold -> VIH(ac)=VREF(dc)+135mV, VIL(ac)=VREF(dc)-135mV DC100 Threshold -> VIH(dc)=VREF(dc)+100mV, VIL(dc)=VREF(dc)-100mV											
		8.0 V/ns		7.0 V/ns		6.0 V/ns		5.0 V/ns		4.0 V/ns		3.0 V/ns	
		ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH	ΔtDS	ΔtDH
DQ,DM Slew rate V/ns	4.0	34	25	34	25	34	25	34	25	34	25		
	3.0			23	17	23	17	23	17	23	17	34	29
	2.0					0	0	0	0	0	0	11	13
	1.5							-23	-17	-23	-17	-12	-4

Note 1: Cell contents shaded in red are defined as 'not supported'

Table. Required time t_{VAC} above VIH(ac) {below VIL(ac)} for valid transition

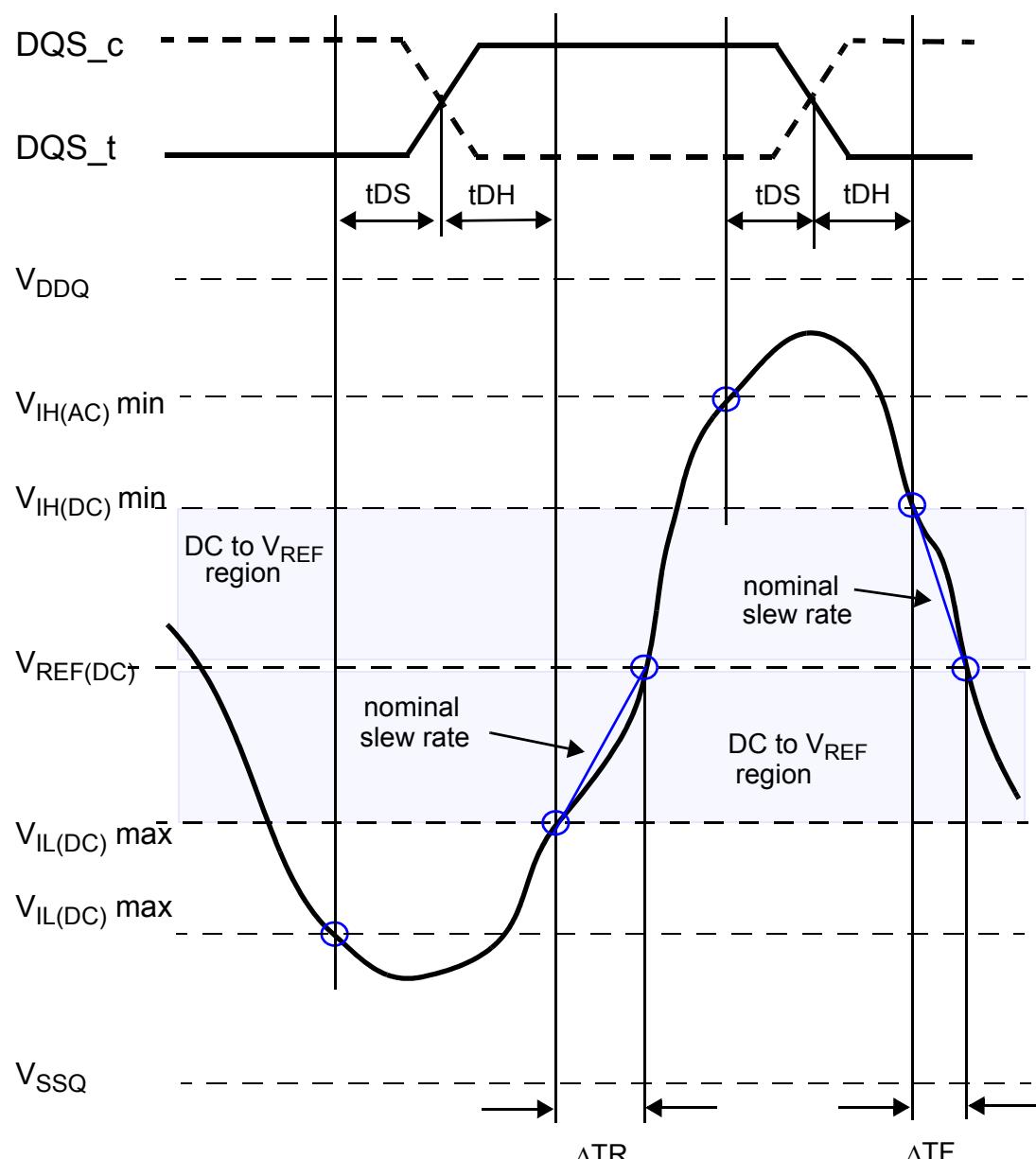
Slew Rate [V/ns]	t_{VAC} [ps] @135mV		t_{VAC} [ps] @150mV		t_{VAC} [ps] @150mV	
	1866Mbps		1600Mbps		1333Mbps	
	MIN	MAX	MIN	MAX	MIN	MAX
> 4.0	40	-	48	-	58	-
4.0	40	-	48	-	58	-
3.5	39	-	46	-	56	-
3.0	36	-	43	-	53	-
2.5	33	-	40	-	50	-
2.0	29	-	35	-	45	-
1.5	21	-	27	-	37	-
<1.5	21	-	27	-	37	-



$$\text{Setup Slew Rate}_{\text{Falling Signal}} = \frac{V_{REF(DC)} - V_{IL(AC)}^{\text{max}}}{\Delta TF}$$

$$\text{Setup Slew Rate}_{\text{Rising Signal}} = \frac{V_{IH(AC)}^{\text{min}} - V_{REF(DC)}}{\Delta TR}$$

Figure. Illustration of nominal slew rate and t_{VAC} for setup time t_{DS} for DQ with respect to strobe



$$\text{Hold Slew Rate}_{\text{Rising Signal}} = \frac{V_{\text{REF}(\text{DC})} - V_{\text{IL}(\text{DC})\text{max}}}{\Delta \text{TR}}$$

$$\text{Hold Slew Rate}_{\text{Falling Signal}} = \frac{V_{\text{IH}(\text{DC})\text{min}} - V_{\text{REF}(\text{DC})}}{\Delta \text{TF}}$$

Figure. Illustration of nominal slew rate for hold time t_{DH} for DQ with respect to strobe

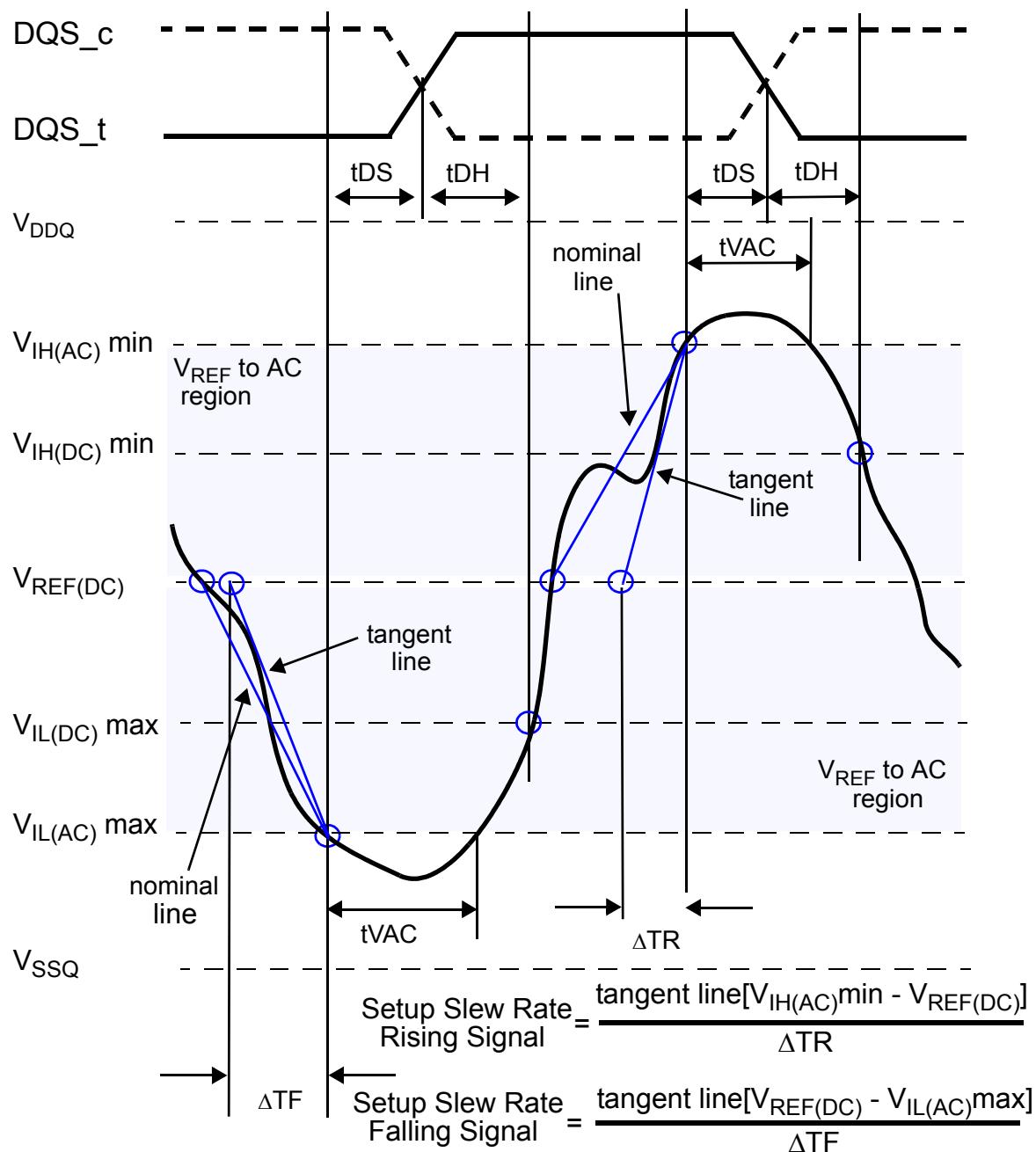


Figure. Illustration of tangent line for setup time t_{DS} for DQ with respect to strobe

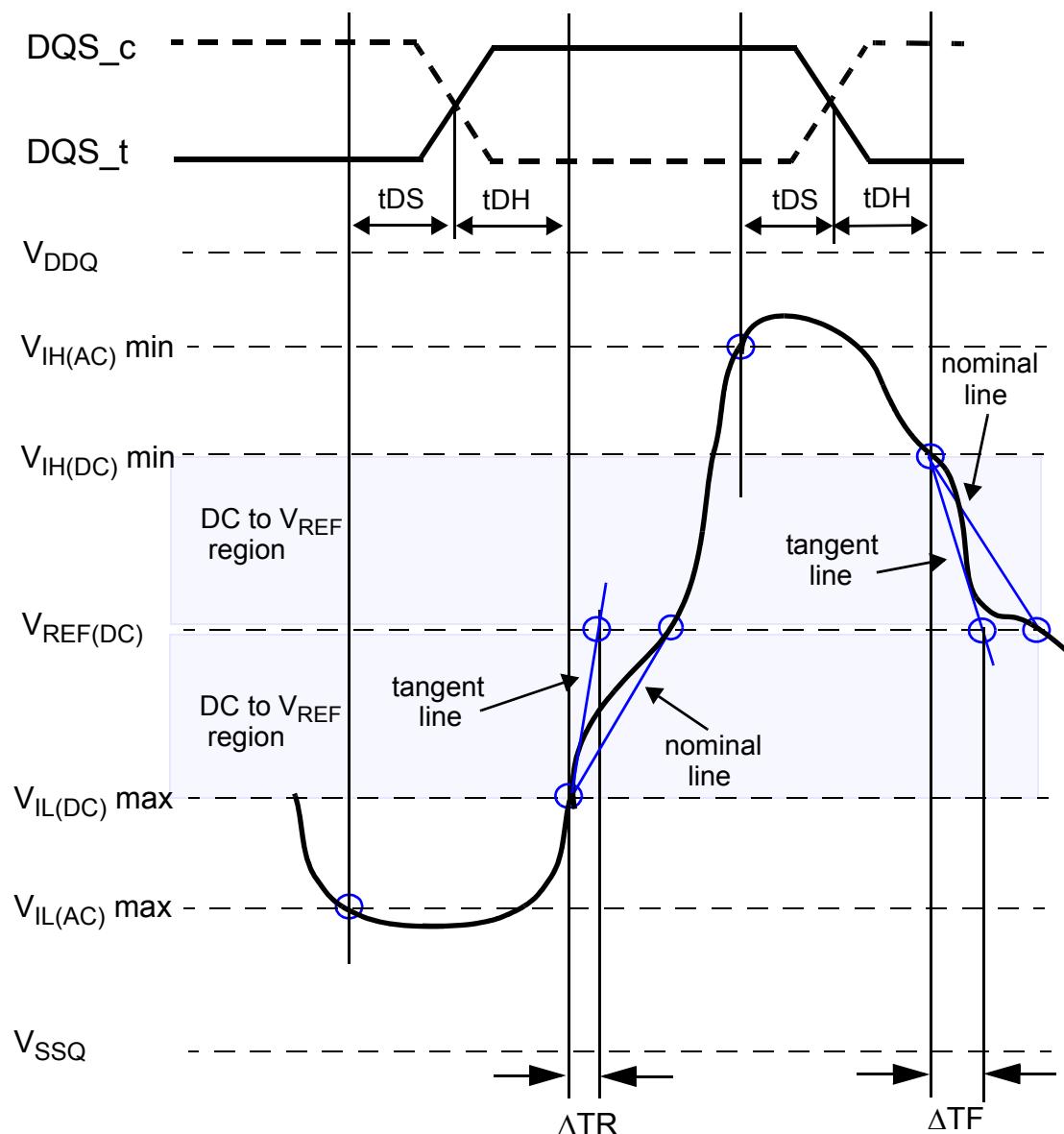


Figure. Illustration of tangent line for hold time tDH for DQ with respect to strobe

Clock Specification

The jitter specified is a random jitter meeting a Gaussian distribution. Input clocks violating the min/max values may result in malfunction of the LPDDR3 device.

Definition for tCK(avg) and nCK

tCK(avg) is calculated as the average clock period across any consecutive 200 cycle window, where each clock period is calculated from rising edge to rising edge.

$$tCK(\text{avg}) = \left(\sum_{j=1}^N tCK_j \right) / N$$

where $N = 200$

Unit 'tCK(avg)' represents the actual clock average tCK(avg) of the input clock under operation. Unit 'nCK' represents one clock cycle of the input clock, counting the actual clock edges.

tCK(avg) may change by up to +/-1% within a 100 clock cycle window, provided that all jitter and timing specifications are met.

Definition for tCK(abs)

tCK(abs) is defined as the absolute clock period, as measured from one rising edge to the next consecutive rising edge. tCK(abs) is not subject to production test.

Definition for tCH(avg) and tCL(avg)

tCH(avg) is defined as the average high pulse width, as calculated across any consecutive 200 high pulses.

$$tCH(\text{avg}) = \left(\sum_{j=1}^N tCH_j \right) / (N \times tCK(\text{avg}))$$

where $N = 200$

tCL(avg) is defined as the average low pulse width, as calculated across any consecutive 200 low pulses.

$$tCL(\text{avg}) = \left(\sum_{j=1}^N tCL_j \right) / (N \times tCK(\text{avg}))$$

where $N = 200$

Definition for tJIT(per)

tJIT(per) is the single period jitter defined as the largest deviation of any signal tCK from tCK(avg).

tJIT(per) = Min/max of {tCK_i - tCK(avg) where i = 1 to 200}. ?

tJIT(per),act is the actual clock jitter for a given system.?

tJIT(per),allowed is the specified allowed clock period jitter.?

tJIT(per) is not subject to production test.

Definition for tJIT(cc)

tJIT(cc) is defined as the absolute difference in clock period between two consecutive clock cycles.

tJIT(cc) = Max of |{tCK_{i+1} - tCK_i}|. ?

tJIT(cc) defines the cycle to cycle jitter.?

tJIT(cc) is not subject to production test.

Definition for tERR(nper)

tERR(nper) is defined as the cumulative error across n multiple consecutive cycles from tCK(avg).

tERR(nper),act is the actual clock jitter over n cycles for a given system.?

tERR(nper),allowed is the specified allowed clock period jitter over n cycles.?

tERR(nper) is not subject to production test.

$$tERR(nper) = \left(\sum_{j=i}^{i+n-1} tCK_j \right) - n \times tCK(\text{avg})$$

tERR(nper),min can be calculated by the formula shown below:

$$tERR(nper), min = (1 + 0.68 LN(n)) \times tJIT(per), min$$

tERR(nper),max can be caculated by the formula shown below:

$$tERR(nper), max = (1 + 0.68 LN(n)) \times tJIT(per), max$$

Using these equations, tERR(nper) tables can be generated for each tJIT(per),act value

Definition for duty cycle jitter tJIT(duty)

tJIT(duty) is defined with absolute and average specification of tCH / tCL.

tJIT(duty),min can be caculated by the formula shown below:

$$tJIT(duty), min = MIN((tCH(abs), min - tCH(avg), min), (tCL(abs), min - tCL(avg), min)) \times tCK(avg)$$

tJIT(duty),max can be caculated by the formula shown below:

$$tJIT(duty), max = MAX((tCH(abs), max - tCH(avg), max), (tCL(abs), max - tCL(avg), max)) \times tCK(avg)$$

Definition for tCK(abs), tCH(abs) and tCL(abs)

These parameters are specified per their average values, however it is understood that the following relationship between the average timing and the absolute instantaneous timing holds at all times.

Parameter	Symbol	Min	Unit
Absolute Clock Period	tCK(abs)	tCK(avg),min + tJIT(per),min	ps
Absolute Clock HIGH Pulse Width	tCH(abs)	tCH(avg),min + tJIT(duty),min / tCK(avg)min	tCK(avg)
Absolute Clock LOW Pulse Width	tCL(abs)	tCL(avg),min + tJIT(duty),min / tCK(avg)min	tCK(avg)

Note:

1. tCK(avg),min is expressed in ps for this table
2. tJIT(duty),min is a negative value

Period Clock Jitter

LPDDR3 devices can tolerate some clock period jitter without core timing parameter de-rating. This section describes device timing requirements in the presence of clock period jitter ($t_{JIT}(per)$) in excess of the values found in "AC timing table" and how to determine cycle time de-rating and clock cycle de-rating.

Clock period jitter effects on core timing parameters

(t_{RCD} , t_{RP} , t_{RTP} , t_{WR} , t_{WRA} , t_{WTR} , t_{RC} , t_{RAS} , t_{RRD} , t_{FAW})

Core timing parameters extend across multiple clock cycles. Period clock jitter will impact these parameters when measured in numbers of clock cycles. When the device is operated with clock jitter within the specification limits, the LPDDR3 device is characterized and verified to support $t_{nPARAM} = RU\{t_{PARAM} / t_{CK(avg)}\}$.

When the device is operated with clock jitter outside specification limits, the number of clocks or $t_{CK(avg)}$ may need to be increased based on the values for each core timing parameter.

Cycle time de-rating for core timing parameters

For a given number of clocks (t_{nPARAM}), for each core timing parameter, average clock period ($t_{CK(avg)}$) and actual cumulative period error ($t_{ERR}(t_{nPARAM},act)$ in excess of the allowed cumulative period error

($t_{ERR}(t_{nPARAM},allowed)$), the equation below calculates the amount of cycle time de-rating (in ns) required if the equation results in a positive value for a core timing parameter (t_{CORE}).

$$Cycle\ Time\ Derating = MAX\left\{\left(\frac{t_{PARAM} + t_{ERR}(t_{nPARAM}, act) - t_{ERR}(t_{nPARAM}, allowed)}{t_{CK(avg)}}\right), 0\right\}$$

A cycle time derating analysis should be conducted for each core timing parameter. The amount of cycle time derating required is the maximum of the cycle time de-ratings determined for each individual core timing parameter.

Clock Cycle de-rating for core timing parameters

For a given number of clocks (t_{nPARAM}) for each core timing parameter, clock cycle de-rating should be specified with amount of period jitter ($t_{JIT}(per)$).

For a given number of clocks (t_{nPARAM}), for each core timing parameter, average clock period ($t_{CK(avg)}$) and actual cumulative period error ($t_{ERR}(t_{nPARAM},act)$ in excess of the allowed cumulative period error ($t_{ERR}(t_{nPARAM},allowed)$), the equation below calculates the clock cycle derating (in clocks) required if the equation results in a positive value for a core timing parameter (t_{CORE}).

$$Clock\ Cycle\ Derating = RU\left\{\frac{t_{PARAM} + t_{ERR}(t_{nPARAM}, act) - t_{ERR}(t_{nPARAM}, allowed)}{t_{CK(avg)}}\right\} - t_{nPARAM}$$

A clock cycle de-rating analysis should be conducted for each core timing parameter.

Clock jitter effects on Command/Address timing parameters

(t_{IS} , t_{IH} , t_{ISCKE} , t_{IHCKE} , t_{ISb} , t_{IHb} , t_{ISCKEb} , t_{IHCKEb})

These parameters are measured from a command/address signal (CKE, CS, CA0 - CA9) transition edge to its respective clock signal (CK_t/CK_c) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. $t_{JIT}(per)$), as the setup and hold are relative to the clock signal crossing that latches the command/address. Regardless of clock jitter values, these values shall be met.

Clock jitter effects on Read timing parameters

tRPRE

When the device is operated with input clock jitter, tRPRE needs to be de-rated by the actual period jitter ($tJIT(per),act,max$) of the input clock in excess of the allowed period jitter ($tJIT(per),allowed,max$). Output de-ratings are relative to the input clock.

$$tRPRE(min, derated) = 0.9 - \left(\frac{tJIT(per), act,max - tJIT(per), allowed,max}{tCK(avg)} \right)$$

tLZ(DQ), tHZ(DQ), tDQSCK, tLZ(DQS), tHZ(DQS)

These parameters are measured from a specific clock edge to a data signal (DMn, DQm: n=0,1,2,3, m=0-31) transition and will be met with respect to that clock edge. Therefore, they are not affected by the amount of clock jitter applied (i.e. $tJIT(per)$).

tQSH, tQSL

These parameters are affected by duty cycle jitter which is represented by $tCH(abs)min$ and $tCL(abs)min$. These parameters determine absolute Data-Valid Window (DVW) at the LPDDR3 device pin.

Absolute min DVW @ LPDDR3 device pin =

$$\min\{ (tQSH(abs)min - tDQSQmax), (tQSL(abs)min - tDQSQmax) \}$$

This minimum DVW shall be met at the target frequency regardless of clock jitter.

tRPST

tRPST is affected by duty cycle jitter which is represented by $tCL(abs)$. Therefore $tRPST(abs)min$ can be specified by $tCL(abs)min$.

$$tRPST(abs)min = tCL(abs)min - 0.05 = tQSL(abs)min$$

Clock jitter effects on Write timing parameters

tDS, tDH

These parameters are measured from a data signal (DM_n, DQ_m: n=0,1,2,3. m=0 –31) transition edge to its respective data strobe signal (DQS_{n_t}, DQS_{n_c} : n=0,1,2,3) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. tJIT(per)), as the setup and hold are relative to the data strobe signal crossing that latches the data. Regardless of clock jitter values, these values shall be met.

tDSS, tDSH

These parameters are measured from a data strobe signal (DQS_{x_t}, DQS_{x_c}) crossing to its respective clock signal (CK_t/CK_c) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. tJIT(per)), as the setup and hold of the data strobes are relative to the corresponding clock signal crossing. Regardless of clock jitter values, these values shall be met.

tDQSS

This parameter is measured from a data strobe signal (DQS_{x_t}, DQS_{x_c}) crossing to the subsequent clock signal (CK_t/CK_c) crossing. When the device is operated with input clock jitter, this parameter needs to be de-rated by the actual period jitter tJIT(per),act of the input clock in excess of the allowed period jitter tJIT(per),allowed.

tDQSS(min,derated) can be caculated by the formula shown below:

$$tDQSS(\text{min, derated}) = 0.75 - \frac{tJIT(\text{per, act,min}) - tJIT(\text{per, allowed,min})}{tCK(\text{avg})}$$

tDQSS(max,derated) can be caculated by the formula shown below:

$$tDQSS(\text{max, derated}) = 1.25 - \frac{tJIT(\text{per, act,max}) - tJIT(\text{per, allowed,max})}{tCK(\text{avg})}$$

Refresh Requirements

Density	Symbol	8Gb	Unit
Number of Banks		8	-
Refresh Window Tcase <= 85'C	tREFW	32	ms
Refresh Window 1/2-Rate Refresh	tREFW	16	ms
Refresh Window 1/4-Rate Refresh	tREFW	8	ms
Required number of REFRESH commands (min)	R	8,192	-
average time between REFRESH commands (for reference only) Tcase <= 85'C	REFab	tREFI	3.9
	REFpb	tREFIpB	0.4875
Refresh Cycle time	tRFCab	210	ns
Per Bank Refresh Cycle time	tRFCpb	90	ns
Burst Refresh Window = 4 x 8 x tRFCab	tREFBW	6.72	us

LPDDR3 Read and Write Latencies

Parameter	LPDDR3								Unit
	333	800	1066	1200	1333	1466	1600	1866	
Max. Clock Frequency	166	400	533	600	667	733	800	933	NHz
Max. Data Rate	333	800	1066	1200	1333	1466	1600	1866	MT/s
Average Clock Period	6	2.5	1.875	1.667	1.5	1.364	1.25	1.071	ns
Read Latency	3	6	8	9	10	11	12	14	tCK(avg)
Write Latency (Set A)	1	3	4	5	6	6	6	8	tCK(avg)
Write Latency (Set B)	1	3	4	5	8	9	9	11	tCK(avg)

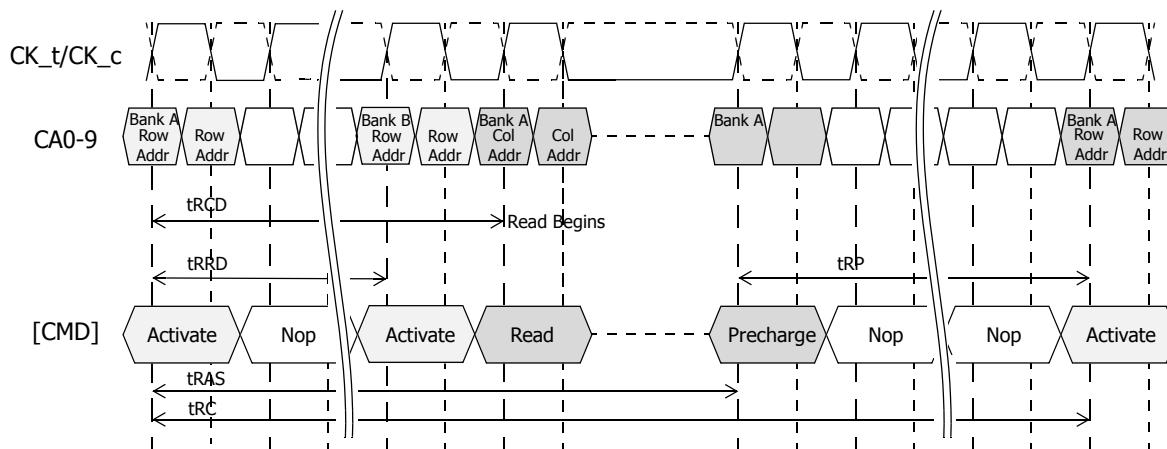
Note:

1. RL=3/WL=1 setting is an optional feature. Refer to MR0 OP<7>.
2. Write Latency (Set B) support is an optional feature. Refer to MR0 OP<6>.

Command Definitions

Activate command

The ACTIVATE command is issued by holding CS_n LOW, CA0 LOW, and CA1 HIGH at the rising edge of the clock. The bank addresses BA0 to BA2 are used to select the desired bank. Row addresses are used to determine which row to activate in the selected bank. The ACTIVATE command must be applied before any READ or WRITE operation can be executed. The device can accept a READ or WRITE command at tRCD after the ACTIVATE command is issued. After a bank has been activated it must be precharged before another ACTIVATE command can be applied to the same bank. The bank active and precharge times are defined as tRAS and tRP, respectively. The minimum time interval between successive ACTIVATE commands to the same bank is determined by the RAS cycle time of the device (tRC). The minimum time interval between ACTIVATE commands to different banks is tRRD.



Note:

1. A PRECHARGE-all command uses tRPab timing, while a single-bank PRECHARGE command uses tRPpb timing. In this figure, tRP is used to denote either an all-bank PRECHARGE or a single-bank PRECHARGE.

Figure. Activate command

Certain restrictions on operation of the 8-bank LPDDR3 devices must be observed. There are two rules: One rule restricts the number of sequential ACTIVATE commands that can be issued; the other provides more time for RAS pre-charge for a PRECHARGE ALL command. The rules are as follows:

- 8 bank device Sequential Bank Activation Restriction: No more than 4 banks may be activated (or refreshed, in the case of REFpb) in a rolling tFAW window. The number of clocks in a tFAW period is dependent upon the clock frequency, which may vary. If the clock frequency is not changed over this period, converting clocks is done by dividing tFAW[ns] by tCK[ns], and rounding up to the next integer value. As an example of the rolling window, if RU(tFAW/tCK) is 10 clocks, and an ACTIVATE command is issued in clock n, no more than three further ACTIVATE commands can be issued at or between clock n + 1 and n + 9. REFpb also counts as bank activation for purposes of tFAW. If the clock frequency is changed during the tFAW period, the rolling tFAW window may be calculated in clock cycles by adding up the time spent in each clock period. The tFAW requirement is met when the previous n clock cycles exceeds the tFAW time.
- 8 bank device Precharge All Allowance: tRP for a PRECHARGE ALL command must equal tRPab, which is greater than tRPpb.

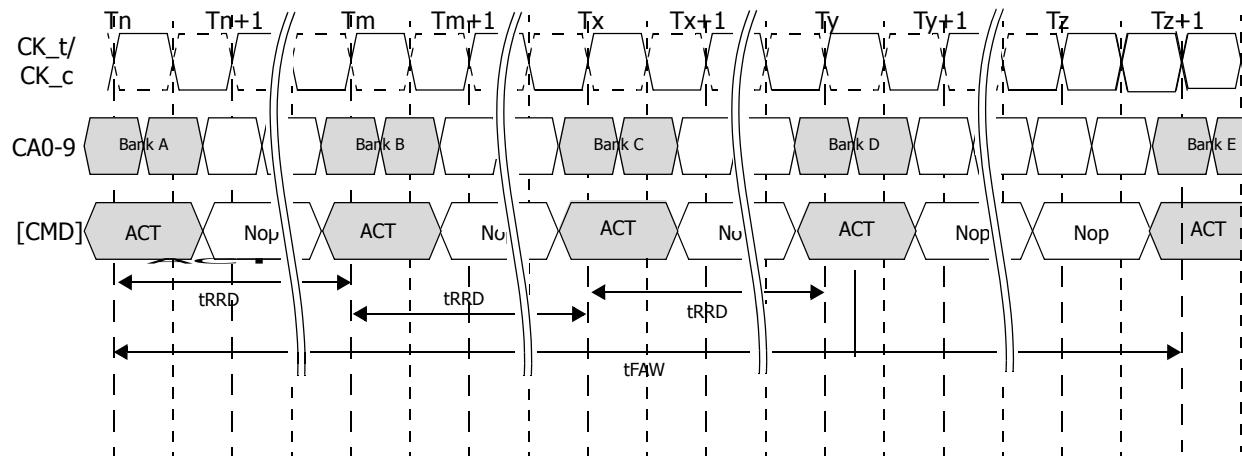
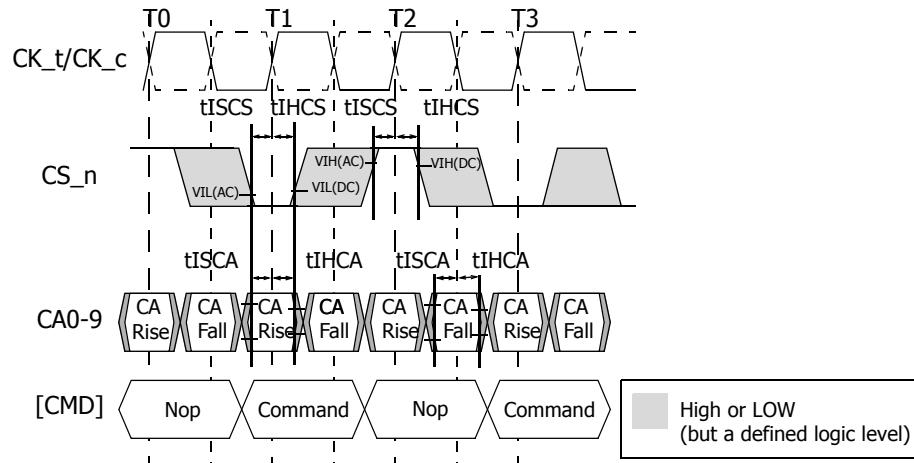


Figure. tFAW timing

Command Input Setup and Hold Timing

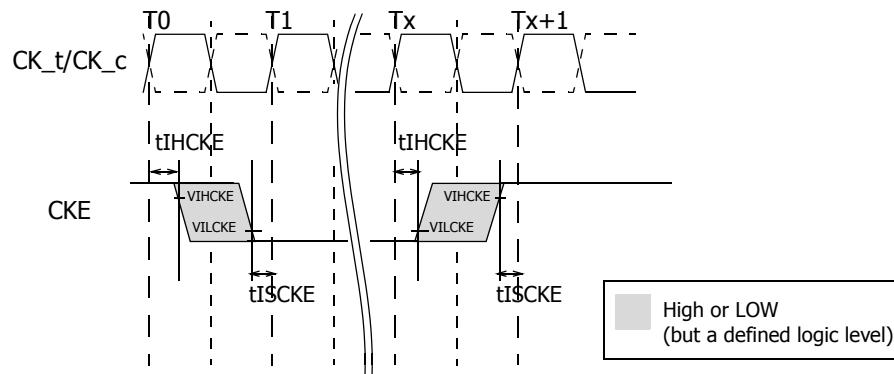


Note:

1. Setup and hold conditions also apply to the CKE pin. See section related to power down for timing diagrams related to the CKE pin.

Figure. Command Input Setup and Hold timing

CKE Input Setup and Hold Timing



Note:

1. After CKE is registered LOW, CKE signal level shall be maintained below VILCKE for tCKE specification (LOW pulse width).
2. After CKE is registered HIGH, CKE signal level shall be maintained above VIHCKE for tCKE specification (HIGH pulse width).

Figure. CKE Input Setup and Hold timing

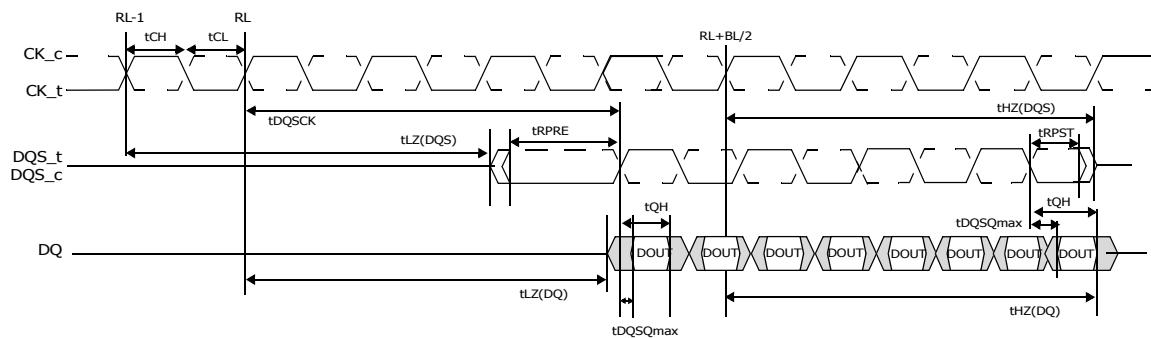
Read and Write access modes

After a bank has been activated, a read or write cycle can be executed. This is accomplished by setting CS_n LOW, CA0 HIGH, and CA1 LOW at the rising edge of the clock. CA2 must also be defined at this time to determine whether the access cycle is a read operation (CA2 HIGH) or a write operation (CA2 LOW).

The LPDDR3 SDRAM provides a fast column access operation. A single Read or Write Command will initiate a burst read or write operation on successive clock cycles. Burst interrupts are not allowed.

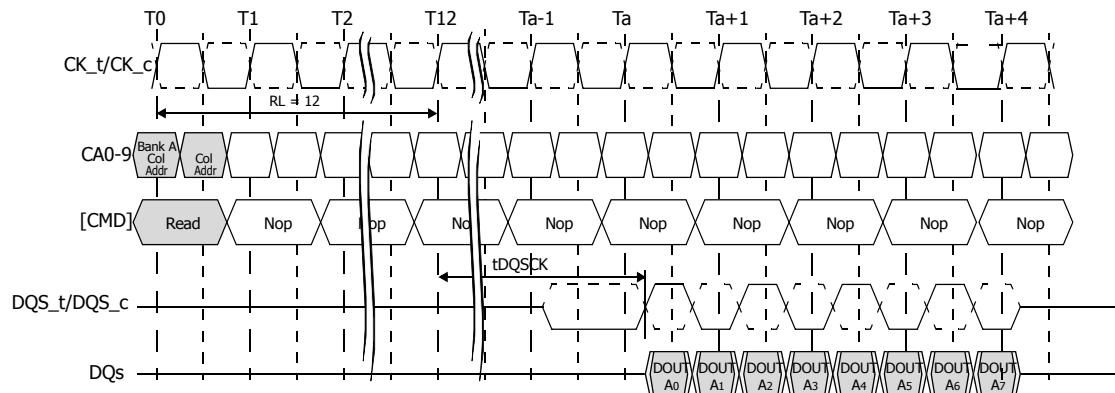
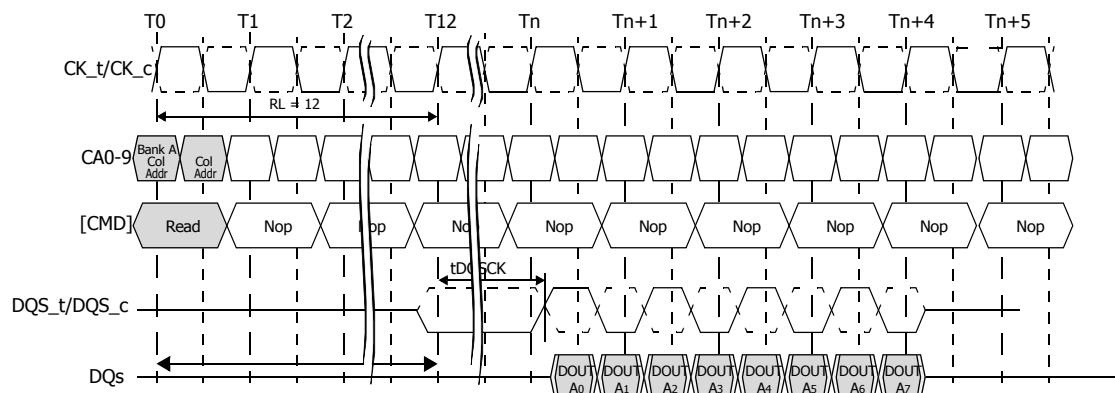
Burst read command

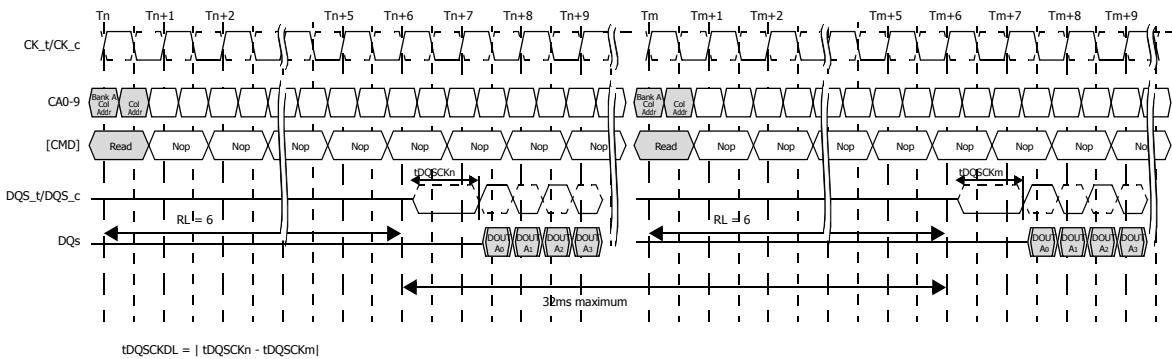
The burst READ command is initiated with CS_n LOW, CA0 HIGH, CA1 LOW, and CA2 HIGH at the rising edge of the clock. The command address bus inputs CA5r-CA6r and CA1f-CA9f determine the starting column address for the burst. The read latency (RL) is defined from the rising edge of the clock on which the READ command is issued to the rising edge of the clock from which the tDQSCK delay is measured. The first valid data is available $RL \times tCK + tDQSCK + tDQSQ$ after the rising edge of the clock when the READ command is issued. The data strobe output is driven LOW tRPRE before the first valid rising strobe edge. The first bit of the burst is synchronized with the first rising edge of the data strobe. Each subsequent data-out appears on each DQ pin, edge-aligned with the data strobe. The RL is programmed in the mode registers. Pin timings for the data strobe are measured relative to the crosspoint of DQS_t and its complement, DQS_c.

**Figure. Read Output Timing**

Note:

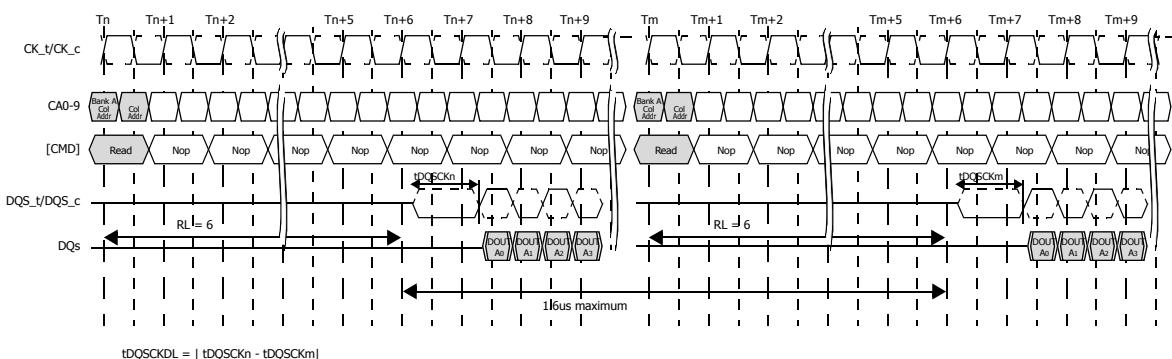
1. tDQSCK can span multiple clock periods.
2. An effective Burst Length of 8 is shown.

**Figure. Burst Read : RL=12, BL=8, tDQSCK>tCK****Figure. Burst Read : RL=12, BL=8, tDQSCK<tCK**


Figure. LPDDR3 : tDQSCKDL timing

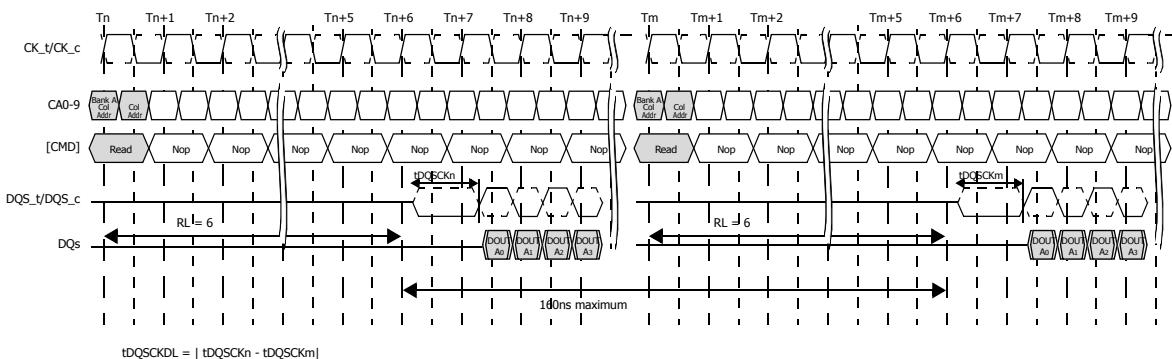
Note:

1. tDQSCKDLmax is defined as the maximum of ABS(tDQSCKn - tDQSCKm) for any {tDQSCKn ,tDQSCKm} pair within any 32ms rolling window.


Figure. LPDDR3 : tDQSCKDM timing

Note:

1. tDQSCKDMmax is defined as the maximum of ABS(tDQSCKn - tDQSCKm) for any {tDQSCKn ,tDQSCKm} pair within any 1.6us rolling window.


Figure. LPDDR3 : tDQSCKDS timing

Note:

1. tDQSCKDSmax is defined as the maximum of ABS(tDQSCKn - tDQSCKm) for any {tDQSCKn ,tDQSCKm} pair within any 160ns rolling window.

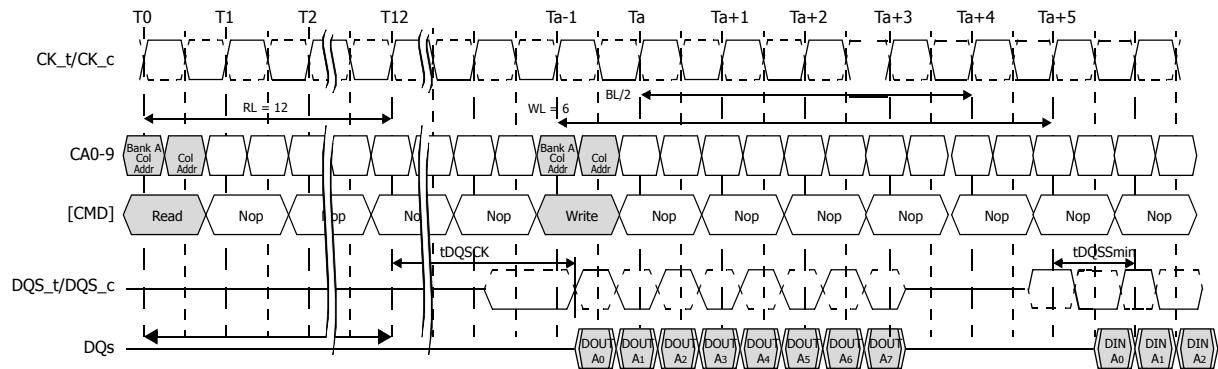


Figure. Burst READ Followed by Burst WRITE: RL=12, WL=6, BL=8

The minimum time from the burst READ command to the burst WRITE command is defined by the read latency (RL) and the burst length (BL). Minimum READ-to-WRITE latency is $RL + RU(tDQSCK(\text{MAX})/tCCK) + BL/2 + 1 - WL$ clock cycles.

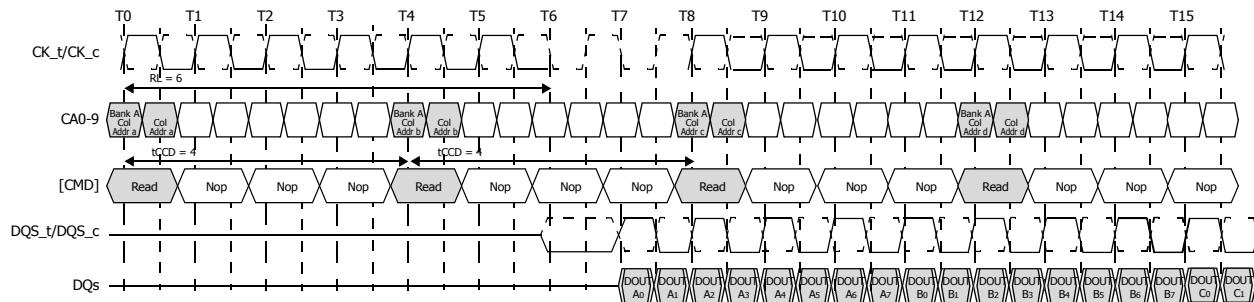


Figure. Seamless Burst READ: RL = 6, BL = 8, tCCD = 4

The seamless burst READ operation is supported by enabling a READ command at every fourth clock cycle for BL = 8 operation. This operation is supported as long as the banks are activated, whether the accesses read the same or different banks.

Burst write operation

The burst WRITE command is initiated with CS_n LOW, CA0 HIGH, CA1 LOW, and CA2 LOW at the rising edge of the clock. The command address bus inputs, CA5r-CA6r and CA1f-CA9f, determine the starting column address for the burst. Write latency (WL) is defined from the rising edge of the clock on which the WRITE command is issued to the rising edge of the clock from which the tDQSS delay is measured. The first valid data must be driven WL × tCK + tDQSS from the rising edge of the clock from which the WRITE command is issued. The data strobe signal (DQS) must be driven for time tWPRE prior to data input. The burst cycle data bits must be applied to the DQ pins tDS prior to the associated edge of the DQS and held valid until tDH after that edge. Burst data is sampled on successive edges of the DQS until the 8-bit burst length is completed. After a burst WRITE operation, tWR must be satisfied before a PRECHARGE command to the same bank can be issued. Pin input timings are measured relative to the crosspoint of DQS_t and its complement, DQS_c.

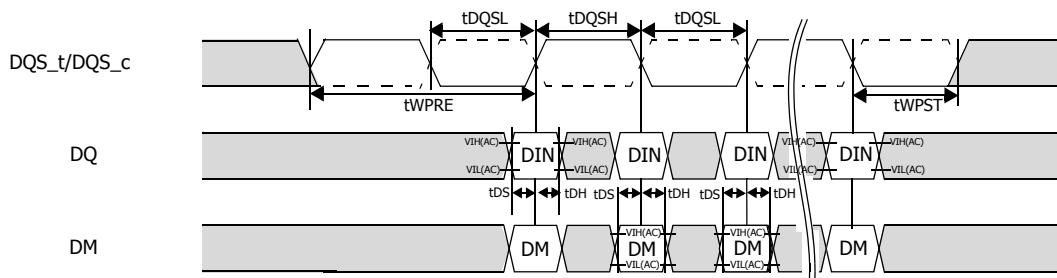


Figure. Data input (Write) timing

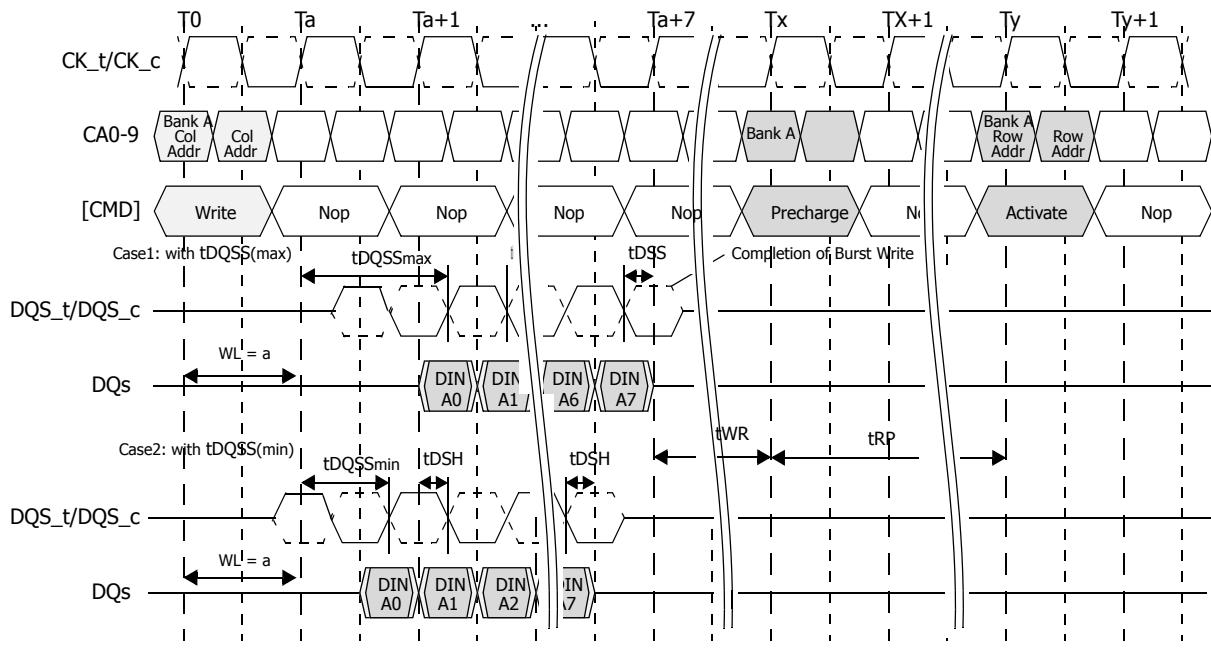


Figure. Burst write

tWPRE Calculation

The method for calculating tWPRE is shown in the following figure.

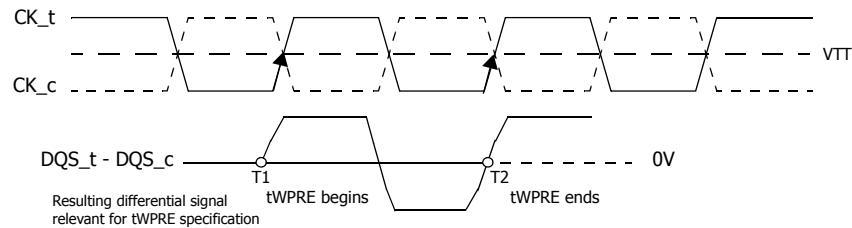


Figure. Method for Caculating tWPRE Transitions and Endpoints

tWPST Calculation

The method for calculating tWPST is shown in the following figure.

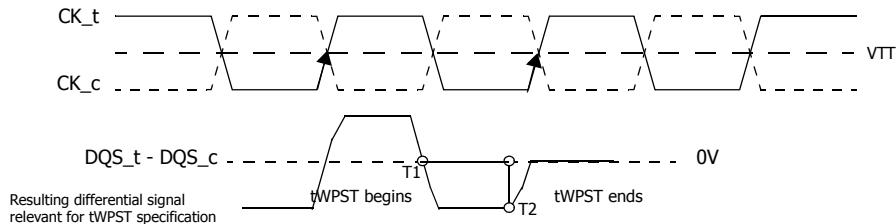
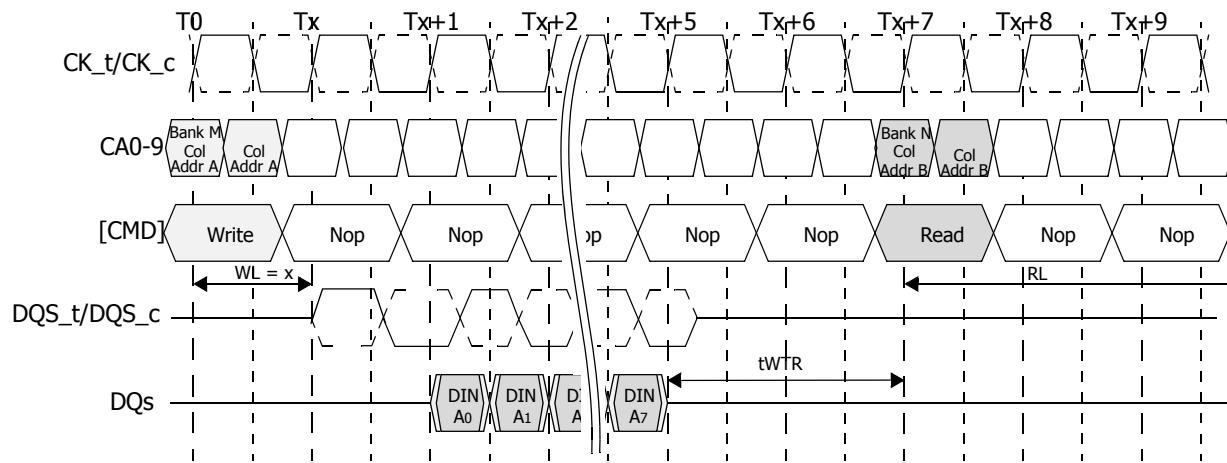


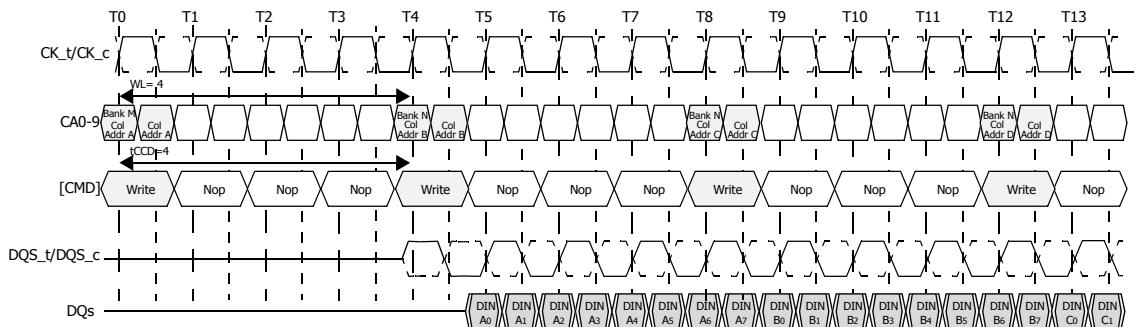
Figure. Method for Caculating tWPRE Transitions and Endpoints



Note:

1. The minimum number of clock cycles from the burst WRITE command to the burst READ command for any bank is $[WL + 1 + BL/2 + RU(tWTR/tCCK)]$.
2. tWTR starts at the rising edge of the clock after the last valid input datum.

Figure. Burst write followed by burst Read



Note:

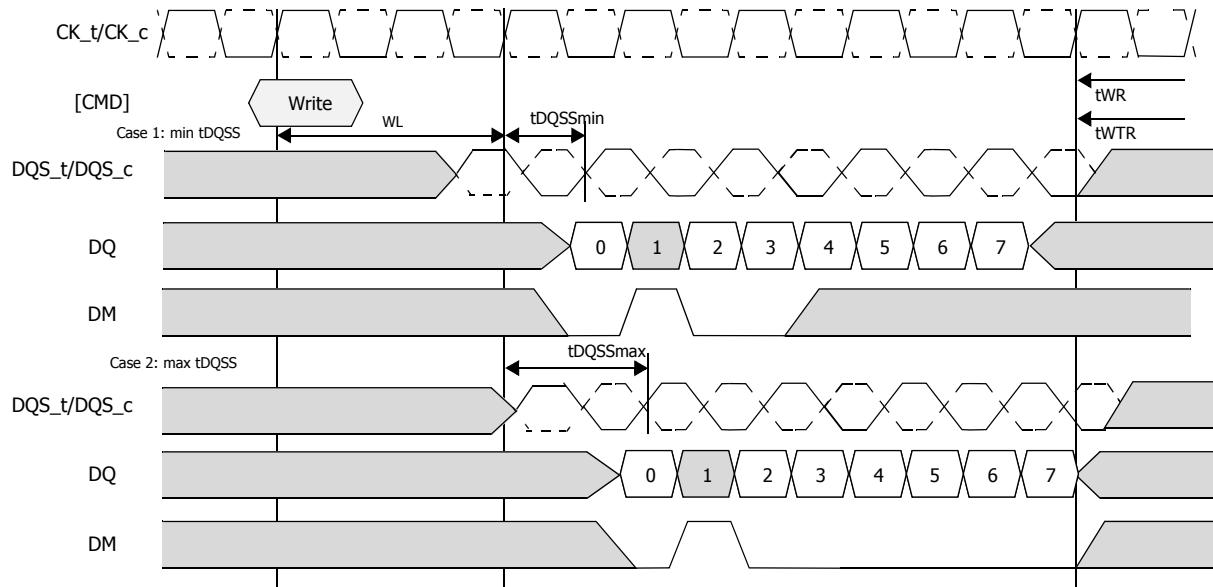
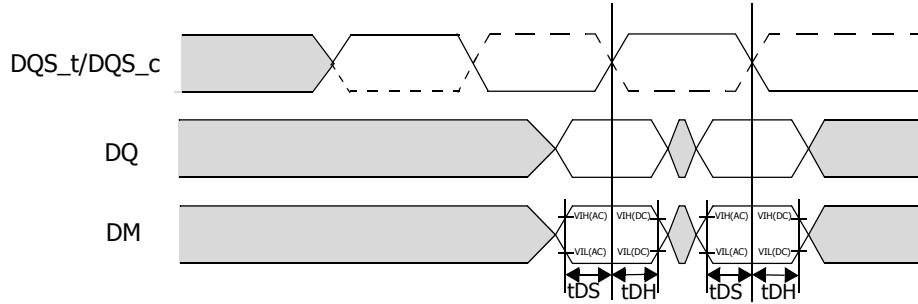
1. The seamless burst WRITE operation is supported by enabling a write command every four clocks for $BL = 8$ operation. This operation is supported for any activated bank.

Figure. Seamless Burst WRITE

Write data mask

One write data mask (DM) pin for each data byte (DQ) is supported, consistent with the implementation on LPDDR3 SDRAM. Each DM can mask its respective DQ for any given cycle of the burst. Data mask timings match data bit timing, but are inputs only. Internal data-mask loading is identical to data-bit loading to ensure matched system timing.

Figure. Data Mask Timing



Note.

1. For the data mask function, BL=8 is shown; the second data bit is masked.

Precharge

The PRECHARGE command is used to precharge or close a bank that has been activated. The PRECHARGE command is initiated with CS_n LOW, CA0 HIGH, CA1 HIGH, CA2 LOW, and CA3 HIGH at the rising edge of the clock. The PRECHARGE command can be used to precharge each bank independently or all banks simultaneously. The AB flag and the bank address bits BA0, BA1, and BA2 are used to determine which bank(s) to precharge. The precharged bank(s) will be available for subsequent row access tRPab after an all-bank PRECHARGE command is issued, or tRPpb after a single-bank PRECHARGE command is issued.

To ensure that LPDDR3 devices can meet the instantaneous current demand required to operate, the row-precharge time for an all-bank PRECHARGE (tRPab) will be longer than the row PRECHARGE time for a single-bank PRECHARGE (tRPpb).

Table. Bank selection for Precharge by address bits

AB (CA4r)	BA2 (CA9r)	BA1 (CA8r)	BA0 (CA7r)	Precharged Bank(s) 8-bank device
0	0	0	0	Bank 0 only
0	0	0	1	Bank 1 only
0	0	1	0	Bank 2 only
0	0	1	1	Bank 3 only
0	1	0	0	Bank 4 only
0	1	0	1	Bank 5 only
0	1	1	0	Bank 6 only
0	1	1	1	Bank 7 only
1	Don't Care	Don't Care	Don't Care	All Banks

Burst read operation followed by Precharge

For the earliest possible precharge, the PRECHARGE command can be issued BL/2 clock cycles after a READ command. A new bank ACTIVATE command can be issued to the same bank after the row PRECHARGE time (t_{RP}) has elapsed. A PRECHARGE command cannot be issued until after t_{RAS} is satisfied. The minimum READ-to-PRECHARGE time must also satisfy a minimum analog time from the rising clock edge that initiates the last 8-bit prefetch of a READ command. t_{RTP} begins BL/2 - 4 clock cycles after the READ command.

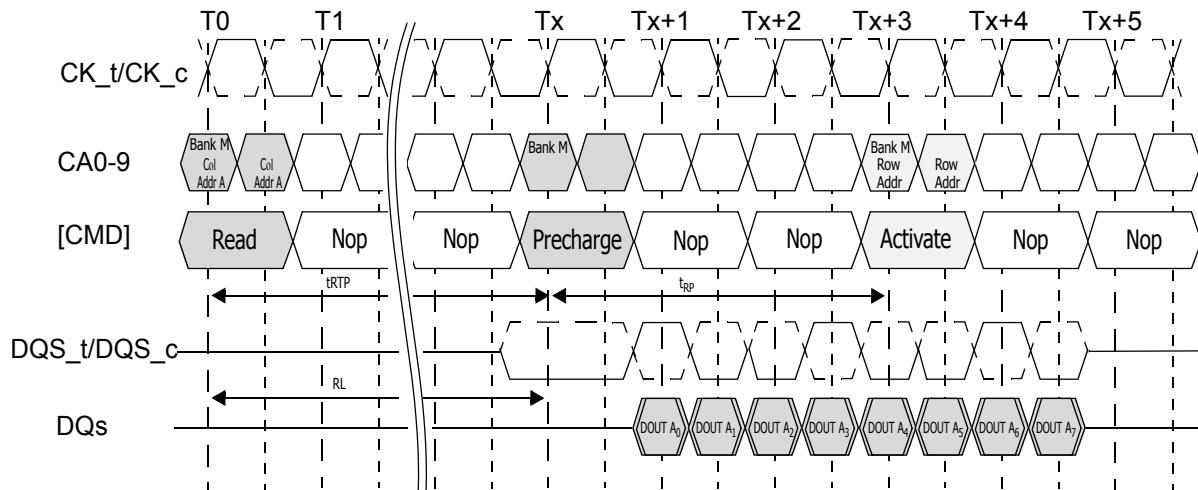


Figure. Burst read followed by Precharge

Burst write followed by precharge

For WRITE cycles, a WRITE recovery time (t_{WR}) must be provided before a PRECHARGE command can be issued. This delay is referenced from the last valid burst input data to the completion of the burst WRITE. PRECHARGE command must not be issued prior to the t_{WR} delay.

LPDDR3 devices write data to the array in prefetch multiples(prefetch = 8). An internal WRITE operation can only begin after a prefetch group has been completely latched, so t_{WR} starts at prefetch boundaries. The minimum WRITE-to-PRECHARGE time for commands to the same bank is $WL + BL/2 + 1 + RU(t_{WR}/tCK)$ clock cycles.

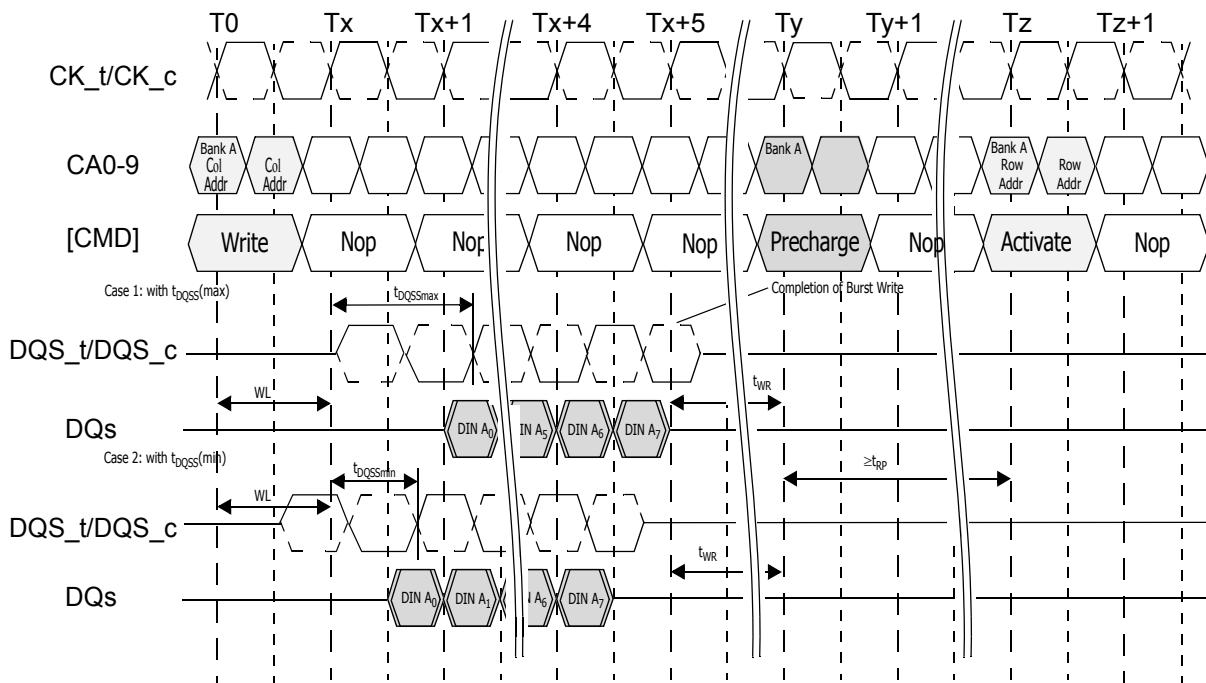


Figure. Burst write followed by Precharge

Auto Precharge Operation

Before a new row can be opened in an active bank, the active bank must be precharged using either the PRECHARGE command or the auto precharge function. When a READ or a WRITE command is issued to the device, the AP bit (CA0f) can be set to enable the active bank to automatically begin precharge at the earliest possible moment during the burst READ or WRITE cycle.

If AP is LOW when the READ or WRITE command is issued, then normal READ or WRITE burst operation is executed and the bank remains active at the completion of the burst.

If AP is HIGH when the READ or WRITE command is issued, the auto precharge function is engaged. This feature enables the PRECHARGE operation to be partially or completely hidden during burst READ cycles (dependent upon READ or WRITE latency) thus improving system performance for random data access.

Burst Read with Auto Precharge

If AP (CA0f) is HIGH when a READ command is issued, the READ with auto-precharge function is engaged. LPDDR3 devices start an auto-precharge operation on the rising edge of the clock BL/2 or BL/2 - 4+ RU(tRTP/tCK) clock cycles later than the READ with auto precharge command, whichever is greater. For LPDDR3 auto-precharge calculations see the table in the next page. Following an auto-precharge operation, an ACTIVATE command can be issued to the same bank if the following two conditions are satisfied simultaneously:

- The RAS precharge time (tRP) has been satisfied from the clock at which the auto- precharge begins.
- The RAS cycle time (tRC) from the previous bank activation has been satisfied.

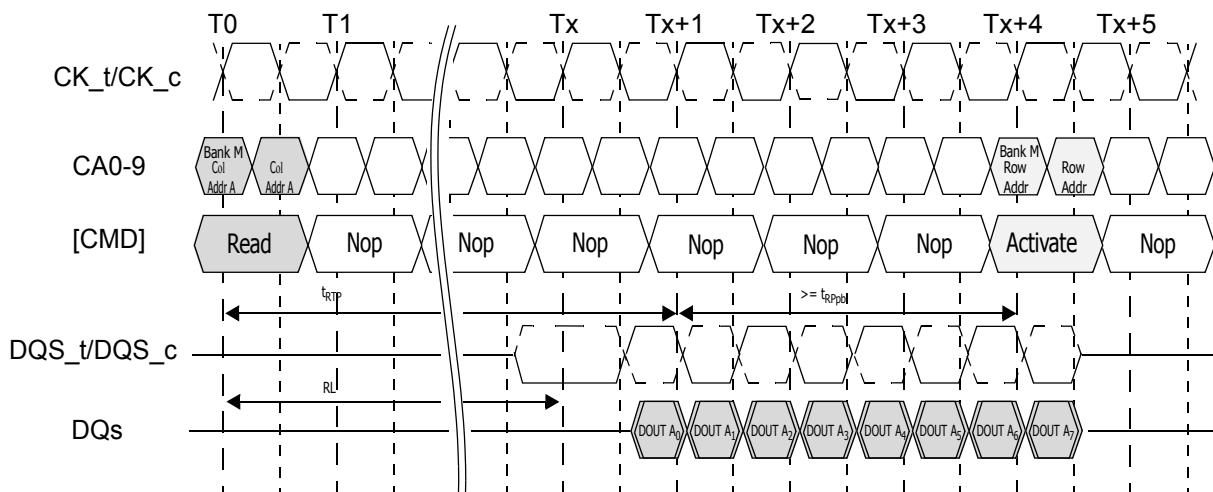


Figure. Burst read with Auto-Precharge

Burst Write with Auto Precharge

If AP (CA0f) is HIGH when a WRITE command is issued, the WRITE with auto precharge function is engaged. The device starts an auto precharge on the rising edge t_{WR} cycles after the completion of the burst WRITE. Following a WRITE with auto precharge, an ACTIVATE command can be issued to the same bank if the following two conditions are met:

The RAS precharge time (t_{RP}) has been satisfied from the clock at which the auto-precharge begins.

The RAS cycle time (t_{RC}) from the previous bank activation has been satisfied.

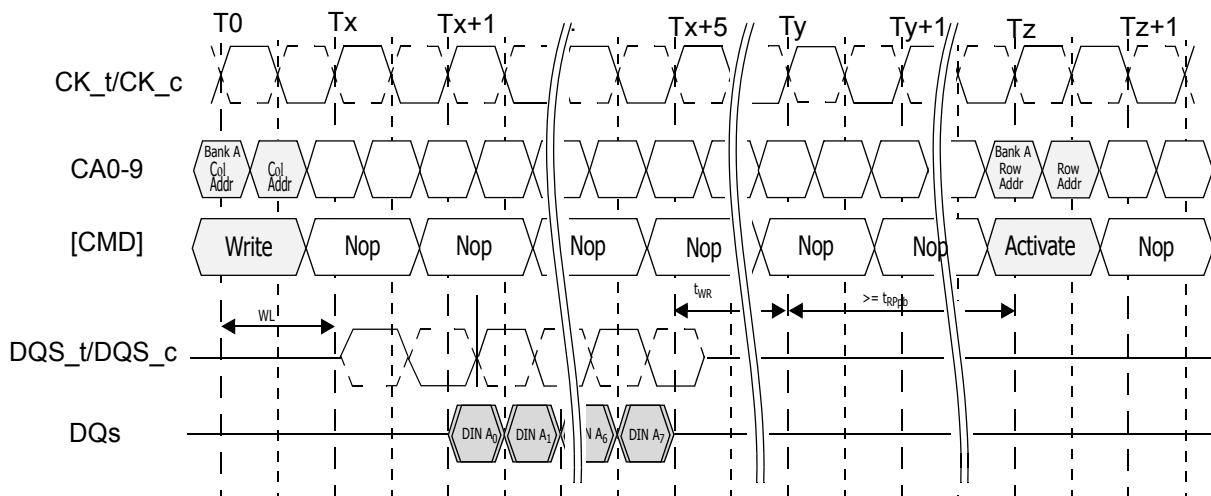


Figure. Burst write with Auto Precharge

Table. Precharge and auto precharge clarification

From command	To command	Minimum delay between From command to To command	Unit	Notes
Read	Precharge (to same bank as read)	BL/2 + max(4, RU (tRTP/tCK)) - 4	CLK	1
	Precharge All	BL/2 + max(4, RU (tRTP/tCK)) - 4	CLK	1
Read w/ AP	Precharge (to same bank as read w/ AP)	BL/2 + max(4, RU (tRTP/tCK)) - 4	CLK	1
	Precharge All	BL/2 + max(4, RU (tRTP/tCK)) - 4	CLK	1
	Activate (to same bank as read w/ AP)	BL/2 + max(4, RU (tRTP/tCK)) - 4 + RU(tRPpb/tCK)	CLK	1
	Write or Write w/AP (same bank)	Illegal	CLK	3
	Write or Write w/AP (different bank)	RL + BL/2 + RU(tDQSCKmax/tCK) -WL + 1	CLK	3
	Read or Read w/AP (same bank)	Illegal	CLK	3
Write	Read or Read w/AP (different bank)	BL/2	CLK	3
	Precharge (to same bank as write)	WL + BL/2 + RU (tWR/tCK) + 1	CLK	1
Write w/ AP	Precharge All	WL + BL/2 + RU (tWR/tCK) + 1	CLK	1
	Precharge (to same bank as write w/ AP)	WL + BL/2 + RU (tWR/tCK) + 1	CLK	1
	Precharge All	WL + BL/2 + RU (tWR/tCK) + 1	CLK	1
	Activate (to same bank as write w/ AP)	WL + BL/2 + RU (tWR/tCK) + 1 + RU(tRPpb/tCK)	CLK	1
	Write or Write w/AP (same bank)	Illegal	CLK	3
	Write or Write w/AP (different bank)	BL/2	CLK	3
Precharge	Read or Read w/AP (same bank)	Illegal	CLK	3
	Read or Read w/AP (different bank)	WL + BL/2 + RU(tWTR/tCK) + 1	CLK	3
Precharge All	Precharge (to same bank as precharge)	1	CLK	1
	Precharge All	1	CLK	1
Precharge All	Precharge	1	CLK	1
	Precharge All	1	CLK	1

Note:

- For a given bank, the precharge period should be counted from the latest precharge command, either one bank precharge or precharge all, issued to that bank. The precharge period is satisfied after tRP depending on the latest precharge command issued to that bank.
- Any command issued during the minimum delay time as specified in Table above is illegal.
- After Read with AP, seamless read operations to different banks are supported. After Write with AP, seamless write operations to different banks are supported. Read w/AP and Write w/AP may not be interrupted or truncated.

Refresh Command

The REFRESH command is initiated with CS_n LOW, CA0 LOW, CA1 LOW, and CA2 HIGH at the rising edge of the clock. Per-bank REFRESH is initiated with CA3 LOW at the rising edge of the clock. All-bank REFRESH is initiated with CA3 HIGH at the rising edge of the clock.

A per-bank REFRESH command (REFpb) performs a per-bank REFRESH operation to the bank scheduled by the bank counter in the memory device. The bank sequence for per-bank REFRESH is fixed to be a sequential round-robin: 0-1-2-3-4-5-6-7-0-1-.... The bank count is synchronized between the controller and the SDRAM by resetting the bank count to zero. Synchronization can occur upon issuing a RESET command or at every exit from self refresh. Bank addressing for the per-bank REFRESH count is the same as established for the single-bank PRECHARGE command. A bank must be idle before it can be refreshed. The controller must track the bank being refreshed by the per-bank REFRESH command.

The REFpb command must not be issued to the device until the following conditions are met:

- tRFCab has been satisfied after the prior REFab command
- tRFCpb has been satisfied after the prior REFpb command
- tRP has been satisfied after the prior Precharge command to that given bank

tRRD has been satisfied after the prior ACTIVATE command (if applicable, for example after activating a row in a different bank than the one affected by the REFpb command).

The target bank is inaccessible during per-bank REFRESH cycle time (tRFCpb), however, other banks within the device are accessible and can be addressed during the cycle. During the REFpb operation, any of the banks other than the one being refreshed can be maintained in an active state or accessed by a READ or a WRITE command. When the per-bank REFRESH cycle has completed, the affected bank will be in the idle state.

After issuing REFpb, these conditions must be met:

- tRFCpb must be satisfied before issuing a REFab command
- tRFCpb must be satisfied before issuing an ACTIVATE command to the same bank
- tRRD must be satisfied before issuing an ACTIVATE command to a different bank
- tRFCpb must be satisfied before issuing another REFpb command

An all-bank REFRESH command (REFab) issues a REFRESH command to all banks. All banks must be idle when REFab is issued (for instance, by issuing a PRECHARGE-all command prior to issuing an all-bank REFRESH command). REFab also synchronizes the bank count between the controller and the SDRAM to zero. The REFab command must not be issued to the device until the following conditions have been met:

- tRFCab has been satisfied after the prior REFab command
- tRFCpb has been satisfied after the prior REFpb command
- tRP has been satisfied after prior Precharge commands

When an all-bank refresh cycle has completed, all banks will be idle. After issuing REFab:

- tRFCab latency must be satisfied before issuing an ACTIVATE command
- tRFCab latency must be satisfied before issuing a REFab or REFpb command

Table. Command Scheduling Separations related to Refresh

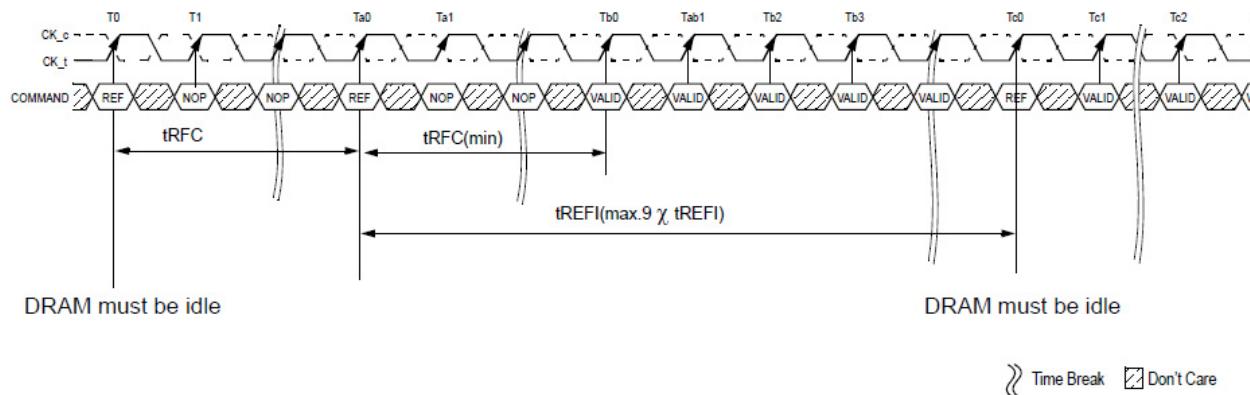
Symbol	minimum delay from	to	Notes
tRFCab	REFab	REFab	
		Activate command to any bank	
		REFpb	
tRFCpb	REFpb	REFab	
		Activate command to same bank as REFpb	
		REFpb	
tRRD	REFpb	Activate command to different bank than REFpb	
	Activate	REFpb	1
		Activate command to different bank than prior Activate command	

Note:

1. A bank must be in the idle state before it is refreshed, so following an ACTIVATE command REFab is prohibited; REFpb is supported only if it affects a bank that is in the idle state.

In general, an all bank refresh command needs to be issued to the LPDDR3 SDRAM regularly every tREFI interval. To allow for improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided for postponing and pulling-in refresh command. A maximum of 8 Refresh commands can be postponed during operation of the LPDDR3 SDRAM, meaning that at no point in time more than a total of 8 Refresh commands are allowed to be postponed. In case that 8 Refresh commands are postponed in a row, the resulting maximum interval between the surrounding Refresh commands is limited to $9 \times tREFI$ (see Figure 1). A maximum of 8 additional Refresh commands can be issued in advance ("pulled in"), with each one reducing the number of regular Refresh commands required later by one. Note that pulling in more than 8, depending on Refresh mode, Refresh commands in advance does not further reduce the number of regular Refresh commands required later, so that the resulting maximum interval between two surrounding Refresh commands is limited to $9 \times tREFI$. At any given time, a maximum of 16 REF commands can be issued within $2 \times tREFI$.

And for per bank refresh, a maximum 8×8 per bank refresh commands can be postponed or pulled in for scheduling efficiency. At any given time, a maximum of $2 \times 8 \times 8$ per bank refresh commands can be issued within $2 \times tREFI$.



Note.

1. Only NOP commands allowed after Refresh command registered until tRFC(min) expires.
2. Time interval between two Refresh commands may be extended to a maximum of 9 X tREFI.

Figure. Refresh Command Timing

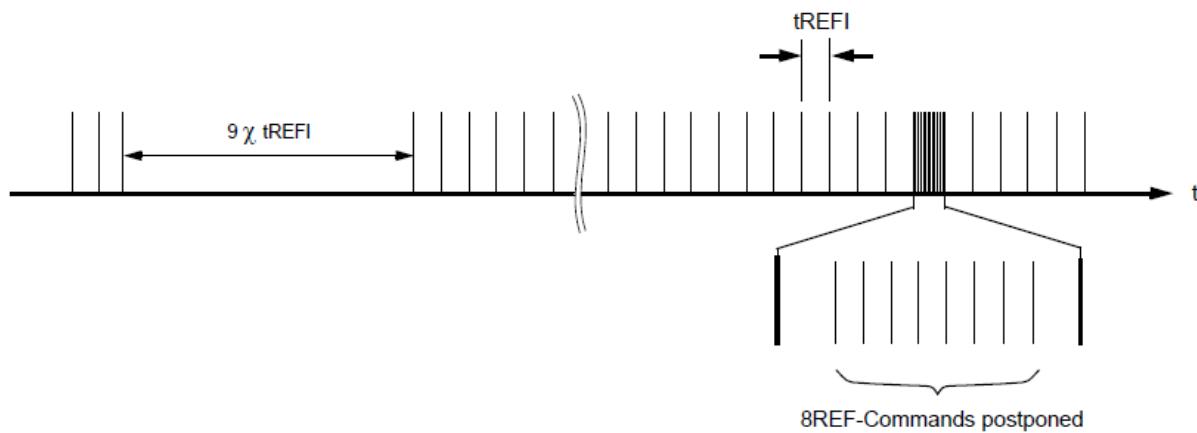


Figure. Postponing Refresh Commands

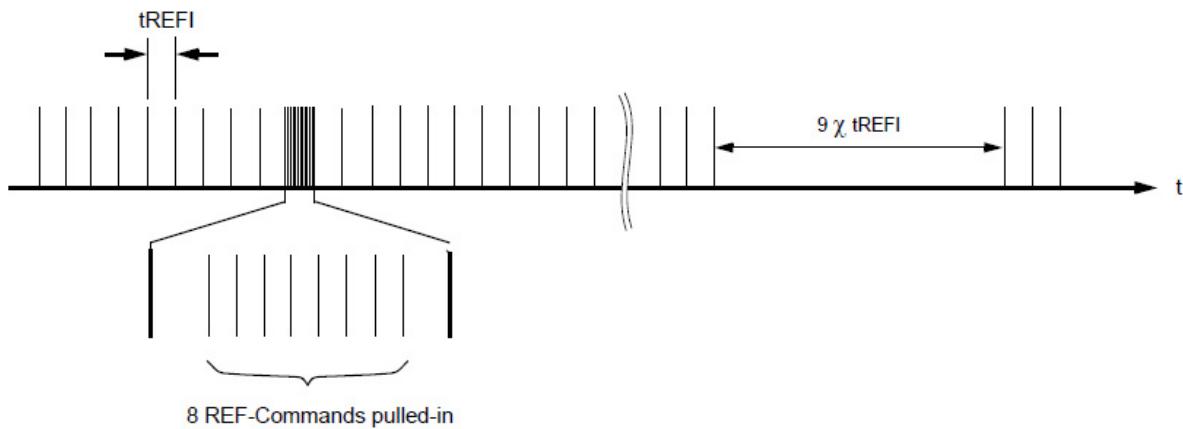


Figure. Pulling-in Refresh Commands

Refresh Requirements

1. Minimum number of REFRESH commands

LPDDR3 requires a minimum number, R, of REFRESH (REFab) commands within any rolling refresh window ($t_{REFW} = 32$ ms @ MR4[2:0] = 011 or $TC \leq 85^{\circ}\text{C}$). For t_{REFW} and t_{REFI} refresh multipliers at different MR4 settings, refer to the MR4 definition.

When using per-bank REFRESH, a REFab command can be replaced by a full cycle of eight REFpb commands.

2. Burst REFRESH limitation

Self refresh mode may be entered with a maximum of eight refresh commands being postponed. After exiting self-refresh mode with one or more refresh commands postponed, additional refresh commands may be postponed to the extent that the total number of postponed refresh commands (before and after the self refresh) will never exceed eight. During self-refresh mode, the number of postponed or pulled-in REF commands does not change.

The use of self refresh mode introduces the possibility that an internally timed refresh event can be missed when CKE is raised for exit from self refresh mode. Upon exit from self refresh, the LPDDR3 SDRAM requires a minimum of one extra refresh command before it is put back into self refresh mode.

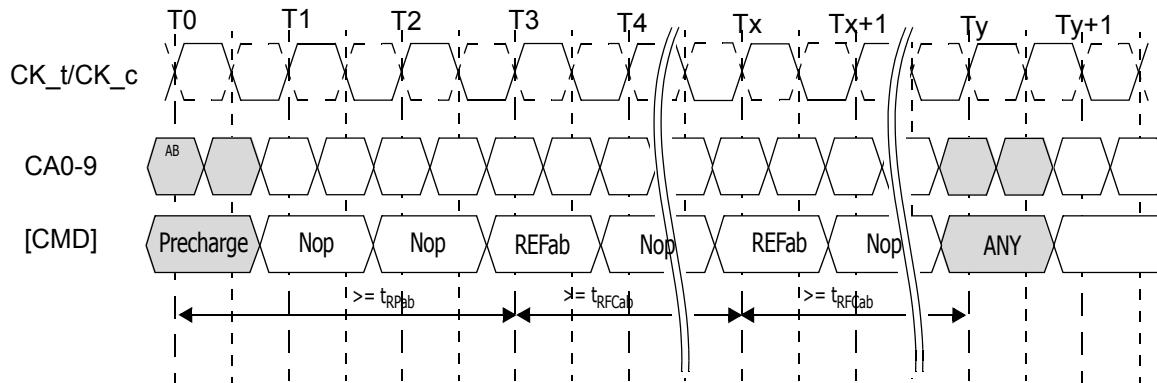
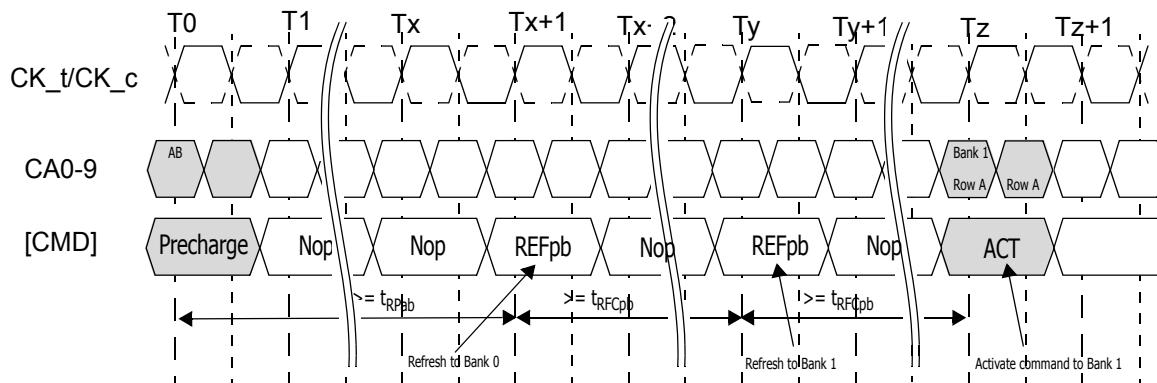


Figure. All Bank Refresh Operation



Note:

1. In the beginning of this example, the REFpb bank is pointing to Bank 0.
2. Operations to other banks than the bank being refreshed are allowed during the t_{RFCpb} period.

Figure. Per Bank Refresh Operation

Self refresh operation

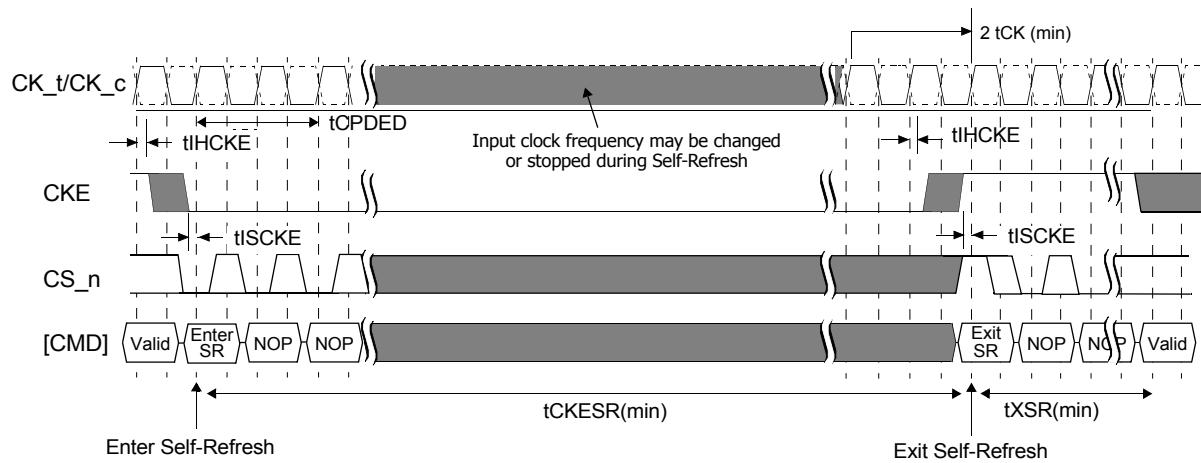
The Self Refresh command can be used to retain data in the LPDDR3 SDRAM, even if the rest of the system is powered down. When in the Self Refresh mode, the SDRAM retains data without external clocking. The device has a built-in timer to accommodate Self Refresh operation. The Self Refresh Command is defined by having CKE LOW, CS_n LOW, CA0 LOW, CA1 LOW, and CA2 HIGH at the rising edge of the clock. CKE must be HIGH during the previous clock cycle. CKE must not go LOW while MRR, MRW, READ, or WRITE operations are in progress. To ensure that there is enough time to account for internal delay on the CKE signal path, two NOP commands are required after CKE is driven LOW, this timing period is defined as tCPDED. CKE LOW will result in deactivation of input receivers after tCPDED has expired. Once the command is registered, CKE must be held LOW to keep the device in Self Refresh mode.

LPDDR3 SDRAM devices can operate in Self Refresh in both the standard or elevated temperature ranges. LPDDR3 devices will also manage Self Refresh power consumption when the operating temperature changes, lower at low temperatures and higher at high temperatures.

Once the SDRAM has entered Self Refresh mode, all of the external signals except CKE, are "don't care". For proper self refresh operation, power supply pins (VDD1, VDD2, and VDDCA) must be at valid levels. VDDQ may be turned off during Self-Refresh. Prior to exiting Self-Refresh, VDDQ must be within specified limits. VrefDQ and VrefCA may be at any level within minimum and maximum levels (see Absolute Maximum DC Ratings). However prior to exiting Self-Refresh, VrefDQ and VrefCA must be within specified limits (see Recommended DC Operating Conditions). The SDRAM initiates a minimum of one all-bank refresh command internally within tCKESR period once it enters Self Refresh mode. The clock is internally disabled during Self Refresh Operation to save power. The minimum time that the SDRAM must remain in Self Refresh mode is tCKESR. The user may change the external clock frequency or halt the external clock one clock after Self Refresh entry is registered; however, the clock must be restarted and stable before the device can exit Self Refresh operation.

The procedure for exiting Self Refresh requires a sequence of commands. First, the clock shall be stable and within specified limits for a minimum of 2 tCK prior to the positive clock edge that registers CKE HIGH. Once Self Refresh Exit is registered, a delay of at least tXSR must be satisfied before a valid command can be issued to the device to allow for any internal refresh in progress. CKE must remain HIGH for the entire Self Refresh exit period tXSR for proper operation except for self refresh re-entry. NOP commands must be registered on each positive clock edge during the Self Refresh exit interval tXSR.

The use of Self Refresh mode introduces the possibility that an internally timed refresh event can be missed when CKE is raised for exit from Self Refresh mode. Upon exit from Self Refresh, it is required that at least one Refresh command (8 per-bank or 1 all-bank) is issued before entry into a subsequent Self Refresh.



Note:

1. Input clock frequency may be changed or stopped during self-refresh, provided that upon exiting self-refresh, a minimum of 2 clocks of stable clock are provided and the clock frequency is between the minimum and maximum frequency for the particular speed grade.
2. Device must be in the "All banks idle" state prior to entering Self Refresh mode.
3. tXSR begins at the rising edge of the clock after CKE is driven HIGH.
4. A valid command may be issued only after tXSR is satisfied. NOPs shall be issued during tXSR.

Figure. Self Refresh Operation

Partial Array Self Refresh: Bank Masking

LPDDR3 SDRAM has 8 banks (additional banks may be required for higher densities). Each bank of LPDDR3 SDRAM can be independently configured whether a self refresh operation is taking place. One mode register unit of 8 bits accessible via MRW command is assigned to program the bank masking status of each bank up to 8 banks. For bank masking bit assignments see Mode Register 16.

The mask bit to the bank controls a refresh operation of entire memory within the bank. If a bank is masked via MRW, a refresh operation to the entire bank is blocked and data retention by a bank is not guaranteed in self refresh mode. To enable a refresh operation to a bank, a coupled mask bit has to be programmed, "unmasked". When a bank mask bit is unmasked, a refresh to a bank is determined by the programmed status of segment mask bits, which is described in the following chapter.

Partial Array Self Refresh: Segment Masking

Segment masking scheme may be used in lieu of or in combination with bank masking scheme in LPDDR3 SDRAM. LPDDR3 devices utilize 8 segments per bank. For segment masking bit assignments, see Mode Register 17.

For those refresh-enabled banks, a refresh operation to the address range which is represented by a segment is blocked when the mask bit to this segment is programmed, "masked". Programming of segment mask bits is similar to the one of bank mask bits. With LPDDR3, 8 segments are used as listed in Mode Register 17. One mode register unit is used for the programming of segment mask bits up to 8 bits. One more mode register unit may be reserved for future use. Programming of bits in the reserved registers has no effect on the device operation.

Table: Example of Bank and Segment Masking use in LPDDR3 devices

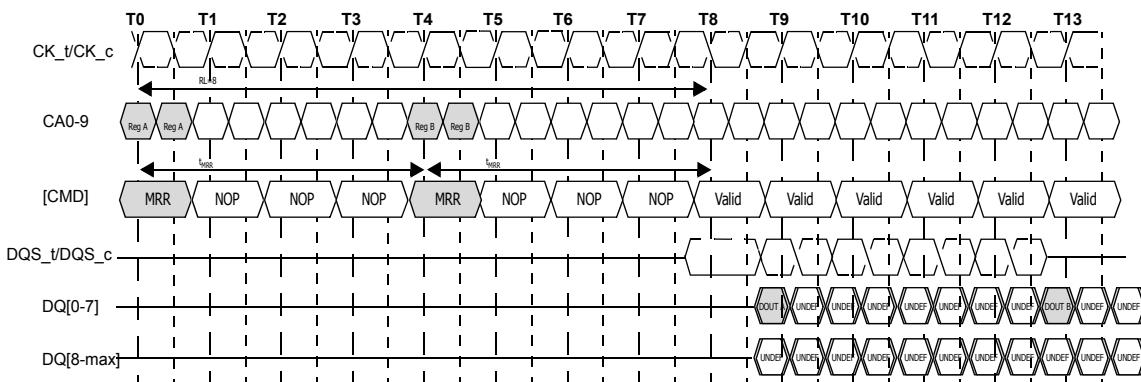
Bank Mask (MR16)	Segment Mask (MR17)	Bank 0	Bank 1	Bank 2	Bank 3	Bank 4	Bank 5	Bank 6	Bank 7
	0	0	1	0	0	0	0	0	1
Segment 0	0		M						M
Segment 1	0		M						M
Segment 2	1	M	M	M	M	M	M	M	M
Segment 3	0		M						M
Segment 4	0		M						M
Segment 5	0		M						M
Segment 6	0		M						M
Segment 7	1	M	M	M	M	M	M	M	M

Note: 1. This table illustrates an example of an 8-bank LPDDR3 device, when a refresh operation to bank 1 and bank 7, as well as segment 2 and segment 7 are masked.

Mode Register Read Command

The MRR command is used to read configuration and status data from SDRAM mode registers. The MRR command is initiated with CS_n LOW, CA0 LOW, CA1 LOW, CA2 LOW, and CA3 HIGH at the rising edge of the clock. The mode register is selected by CA1f-CA0f and CA9r-CA4r. The mode register contents are available on the first data beat of DQ[7:0] after RL × tCK + tDQSCK + tDQSQ following the rising edge of the clock where MRR is issued. Subsequent data beats contain valid but undefined content, except in the case of the DQ calibration function, where subsequent data beats contain valid content as described in the DQ Calibration specification. All DQS are toggled for the duration of the mode register READ burst.

The MRR command has a burst length of eight. MRR operation (consisting of the MRR command and the corresponding data traffic) must not be interrupted.

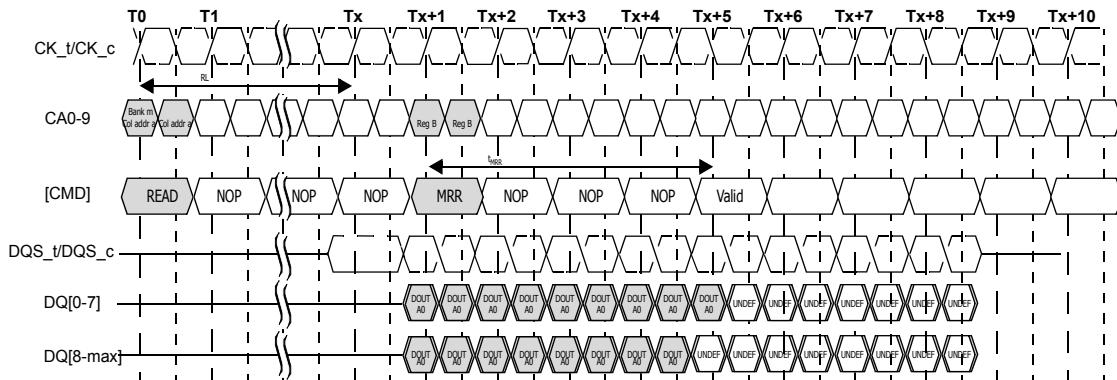


Note:

1. MRRs to DQ calibration registers MR32 and MR40 are described in DQ calibration section.
2. Only the NOP command is supported during tMRR.
3. Mode register data is valid only on DQ[7:0] on the first beat. Subsequent beats contain valid but undefined data. DQ[MAX:8] contain valid but undefined data for the duration of the MRR burst.
4. Minimum Mode Register Read to write latency is $RL + RU(tDQSCKmax/tCK) + 8/2 + 1 - WL$ clock cycles.
5. Minimum Mode Register Read to Mode Register Write latency is $RL + RU(tDQSCKmax/tCK) + 8/2 + 1$ clock cycles.
6. In this example, $RL = 8$ for illustration purposes only.

Figure. Mode Register Read Timing

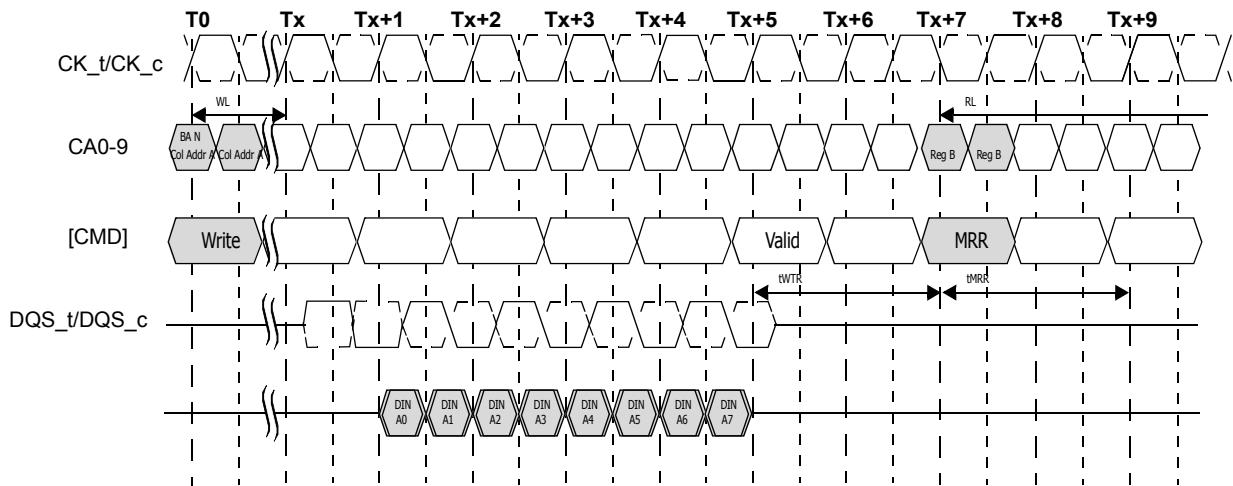
After a prior READ command, the MRR command must not be issued earlier than BL/2 clock cycles, or WL + 1 + BL/2 + RU(tWTR/tCK) clock cycles after a prior WRITE command, as READ bursts and WRITE bursts must not be truncated by MRR.



Note:

1. Only the NOP command is supported during tMRR.
2. The minimum number of clock cycles from the burst READ command to the MRR command is BL/2.

Figure. Read to MRR timing



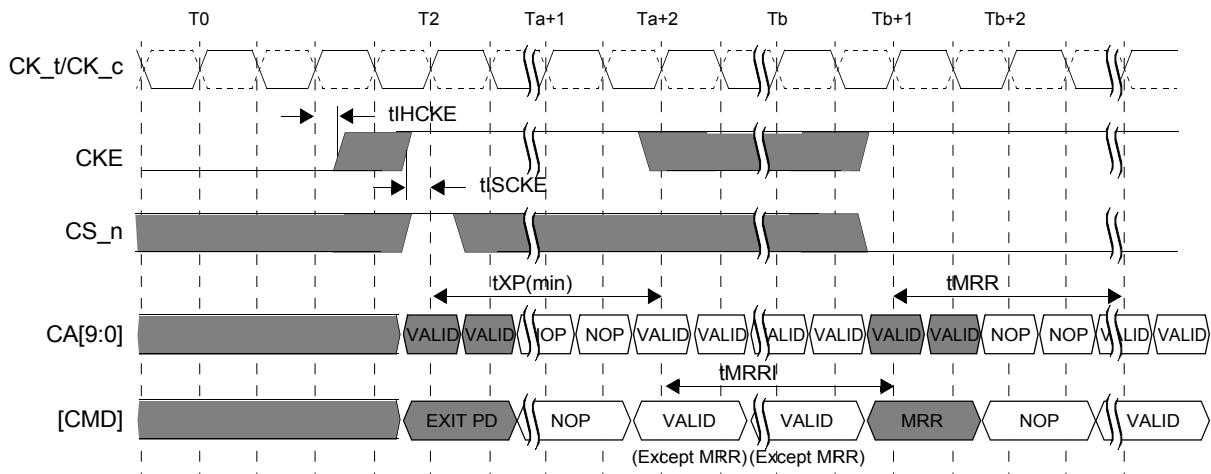
Note:

1. The minimum number of clock cycles from the burst WRITE command to the MRR command is [WL + 1 + BL/2 + RU(tWTR/tCK)].
2. Only the NOP command is supported during tMRR.

Figure. Burst Write Followed by MRR

MRR Following Idle Power-Down State

Following the idle power-down state, an additional time, tMRRI, is required prior to issuing the mode register read (MRR) command. This additional time (equivalent to tRCD) is required in order to be able to maximize power-down current savings by allowing more power-up time for the MRR data path after exit from standby, idle power-down mode.



Note:

1. Any valid command from the idle state except MRR.
2. tMRRI = tRCD.

Figure. MRR Following Power-Down Idle State

Temperature Sensor

LPDDR3 devices feature a temperature sensor whose status can be read from MR4. This sensor can be used to determine an appropriate refresh rate, determine whether AC timing de-rating is required in the elevated temperature range, and/or monitor the operating temperature. Either the temperature sensor or the device TOPER (See Operating Temperature Range) may be used to determine whether operating temperature requirements are being met.

LPDDR3 devices shall monitor device temperature and update MR4 according to tTSI. Upon exiting self-refresh or power-down, the device temperature status bits shall be no older than tTSI.

When using the temperature sensor, the actual device case temperature may be higher than the TOPER specification (See Operating Temperature Range) that applies for the standard or elevated temperature ranges. For example, TCASE may be above 85°C when MR4[2:0] equals 011B. LPDDR3 devices shall allow for a 2°C temperature margin between the point at which the device temperature enters the elevated temperature range and point at which the controller re-configures the system accordingly. In the case of tight thermal coupling of the memory device to external hot spots, the maximum device temperature might be higher than what is indicated by MR4.

To assure proper operation using the temperature sensor, applications should consider the following factors:

- TempGradient is the maximum temperature gradient experienced by the memory device at the temperature of interest over a range of 2°C.
- ReadInterval is the time period between MR4 reads from the system.
- TempSensorInterval (tTSI) is maximum delay between internal updates of MR4.
- SysRespDelay is the maximum time between a read of MR4 and the response by the system.

In order to determine the required frequency of polling MR4, the system shall use the maximum TempGradient and the maximum response time of the system using the following equation:

$$\text{TempGradient} \times (\text{ReadInterval} + \text{tTSI} + \text{SysRespDelay}) \leq 2^\circ\text{C}$$

Paramter	Sysmbol	Min/Max	Value	Unit	Note
System Temperature Gradient	TempGradient	Max	System Dependent	°C/s	
MR4 Read Interval	ReadInterval	Max	System Dependent	ms	
Temperature Sensor Interval	tTSI	Max	32	ms	
System Response Delay	SysRespDelay	Max	System Dependent	ms	
MR4 Temp Margin	TempMargin	Max	2	°C	

For example, if TempGradient is 10°C/s and the SysRespDelay is 1ms:

$$10^\circ\text{C/s} \times (167\text{ms} + 32\text{ms} + 1\text{ms}) \leq 2^\circ\text{C}$$

In this case, ReadInterval shall be no greater than 167ms.

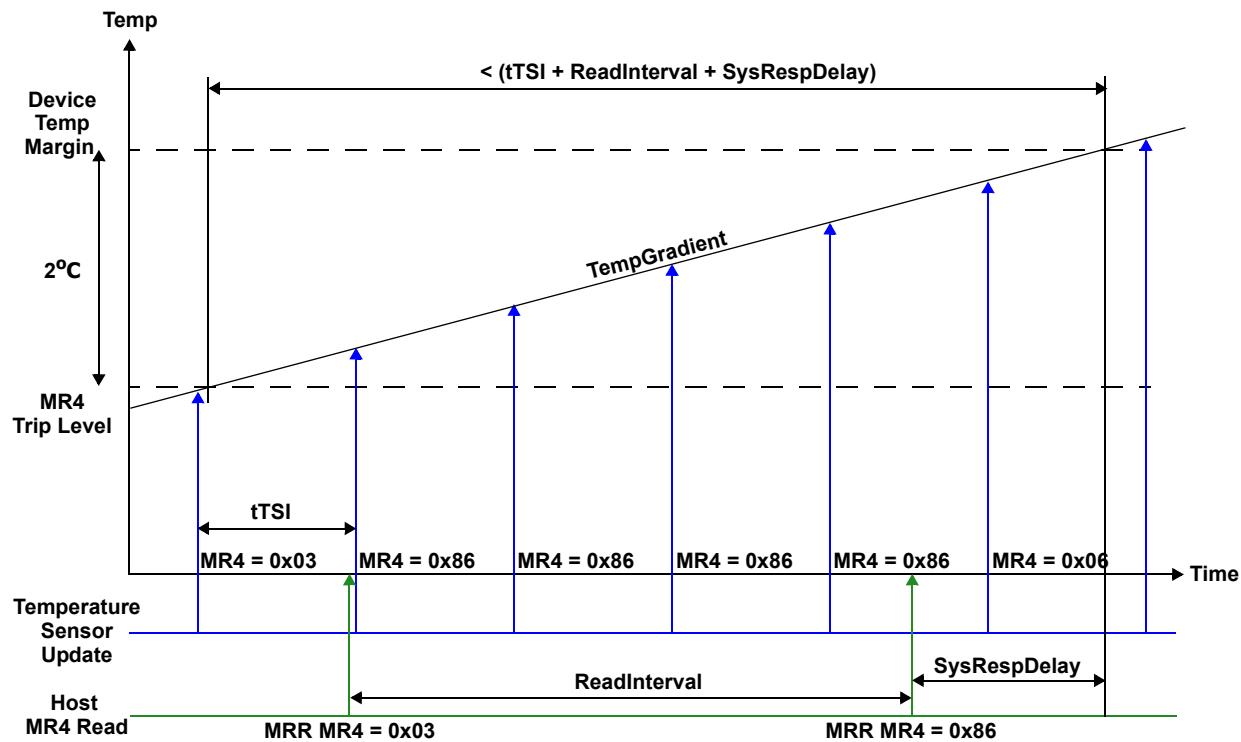


Figure. Temp Sensor Timing

DQ Calibration

LPDDR3 device features a DQ calibration function that outputs one of two predefined system timing calibration patterns. A Mode Register Read to MR32 (Pattern "A") or MR40 (Pattern "B") will return the specified pattern on DQ[0] and DQ[8] for X16 devices, and DQ[0], DQ[8], DQ[16], and DQ[24] for X32 devices. For X16 devices, DQ[7:1] and DQ[15:9] may optionally drive the same information as DQ[0] or may drive 0b during the MRR burst. For X32 devices, DQ[7:1], DQ[15:9], DQ[23:17], and DQ[31:25] may optionally drive the same information as DQ[0] or may drive 0b during the MRR burst.

Table. Data Calibration Pattern Description

	Bit Time 0	Bit Time 1	Bit Time 2	Bit Time 3	Bit Time 4	Bit Time 5	Bit Time 6	Bit Time 7
Pattern A (MR32)	1	0	1	0	1	0	1	0
Pattern B (MR40)	0	0	1	1	0	0	1	1

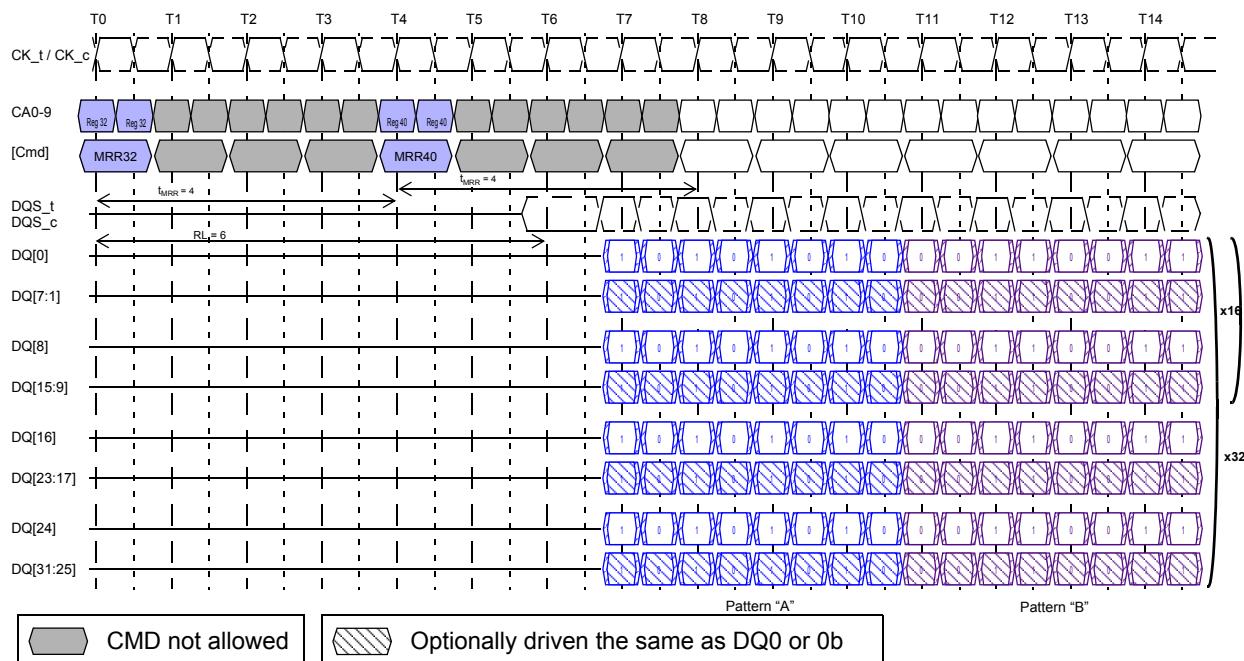


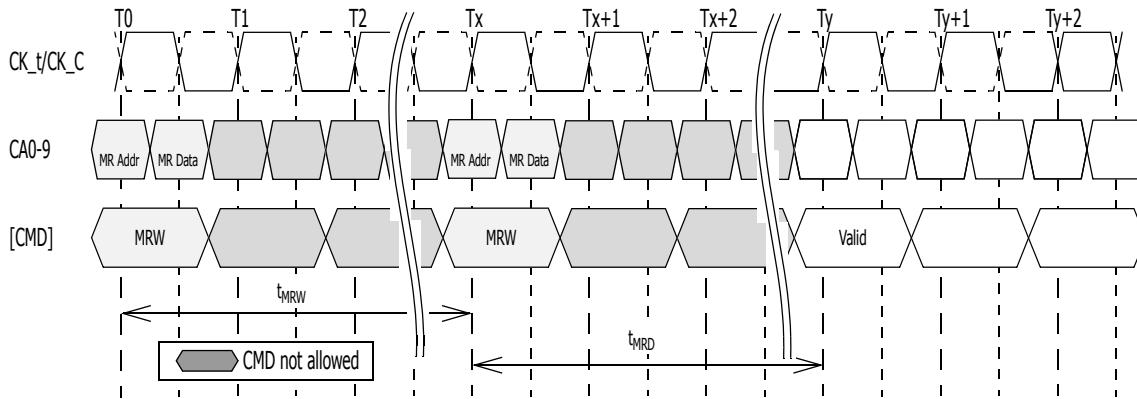
Figure. MR32 and MR40 DQ Calibration timing example

Note:

1. Mode Register Read has a burst length of eight.
2. Mode Register Read operation shall not be interrupted.
3. Mode Register Reads to MR32 and MR40 drive valid data on DQ[0] during the entire burst. For X16 devices, DQ[8] shall drive the same information as DQ[0] during the burst. For X32 devices, DQ[8], DQ[16] and DQ[24] shall drive the same information as DQ[0] during the burst.
4. For X16 devices, DQ[7:1] and DQ[15:9] may optionally drive the same information as DQ[0] or they may drive 0b during the burst. For X32 devices, DQ[7:1], DQ[15:9], DQ[23:17] and DQ[31:25] may optionally drive the same information as DQ[0] or they may drive 0b during the burst.

Mode Register Write Command

The MRW command is used to write configuration data to mode registers. The MRW command is initiated with CS_n LOW, CA0 LOW, CA1 LOW, CA2 LOW, and CA3 LOW at the rising edge of the clock. The mode register is selected by CA1f-CA0f, CA9r-CA4r. The data to be written to the mode register is contained in CA9f-CA2f. The MRW command period is defined by tMRW. Mode register WRITES to read-only registers have no impact on the functionality of the device.



Note:

1. At time Ty, the device is in the idle state.
2. Only the NOP command is supported during tMRW.

Figure. Mode Register Write timing example

Mode Register Write

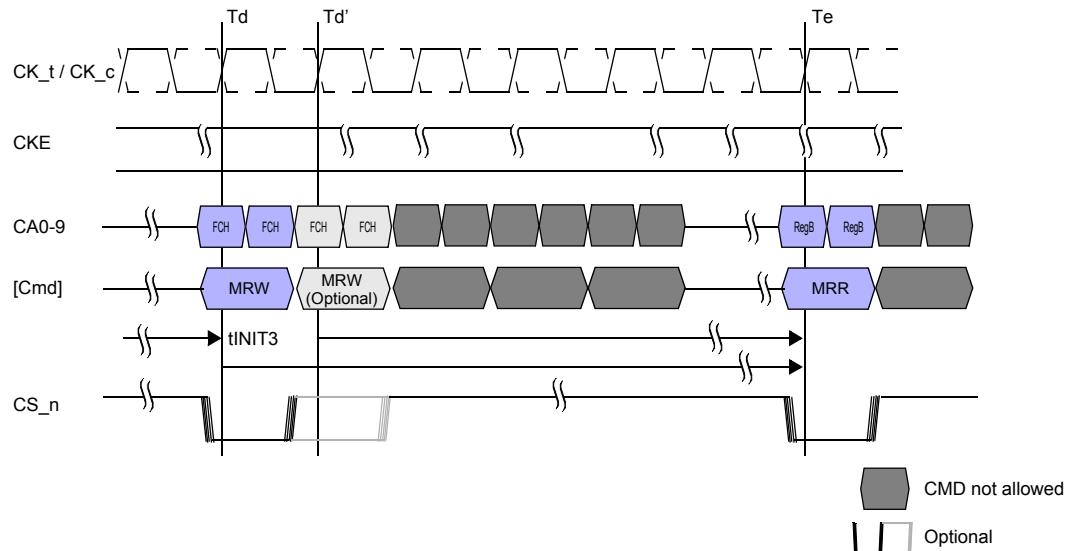
MRW can only be issued when all banks are in the idle precharge state. One method of ensuring that the banks are in this state is to issue a PRECHARGE-ALL command.

MRW Reset

The MRW RESET command brings the device to the device auto-initialization (resetting) state in the power-on initialization sequence. The MRW RESET command can be issued from the idle state. This command resets all mode registers to their default values. After MRW RESET, boot timings must be observed until the device initialization sequence is complete and the device is in the idle state. Array data is undefined after the MRW RESET command.

If the initialization is to be performed at-speed (greater than the recommended boot clock frequency), then CA Training may be necessary to ensure setup and hold timings. Since the MRW RESET command is required prior to CA Training an alternate MRW RESET command with an op-code of 0xFCh should be used. This encoding ensures that no transitions occur on the CA bus. Prior to CA Training, it is recommended to hold the CA bus stable for one cycle prior to, and one cycle after, the issuance of the MRW RESET command to ensure setup and hold timings on the CA bus.

Current State	Command	Intermediate State	Next State
All Banks Idle	MRR	Mode Register Reading (All Banks Idle)	All Banks Idle
	MRW	Mode Register Writing (All Banks Idle)	All Banks Idle
	MRW (RESET)	Resetting (Device Auto-Init)	All Banks Idle
Bank(s) Active	MRR	Mode Register Reading (Bank(s) Active)	Bank(s) Active
	MRW	Not Allowed	Not Allowed
	MRW (RESET)	Not Allowed	Not Allowed



Note:

1. Optional MRW RESET command and optional CS_n assertion are allowed, When optional MRW RESET command is used, t_{INIT4} starts at T_d' .

Figure. Mode Register Write Timing for MRW RESET

Mode Register Write ZQ Calibration Command

The MRW command is used to initiate the ZQ calibration command. This command is used to calibrate the output driver impedance across process, temperature, and voltage. LPDDR3 devices support ZQ calibration.

There are four ZQ calibration commands and related timings: tZQINIT, tZQRESET, tZQCL, and tZQCS. tZQINIT is for initialization calibration; tZQRESET is for resetting ZQ to the default output impedance; tZQCL is for long calibration(s); and tZQCS is for short calibration(s).

The initialization ZQ calibration (ZQINIT) must be performed for LPDDR3. ZQINIT provides an output impedance accuracy of ± 15 percent. After initialization, the ZQ calibration long (ZQCL) can be used to recalibrate the system to an output impedance accuracy of ± 15 percent. A ZQ calibration short (ZQCS) can be used periodically to compensate for temperature and voltage drift in the system.

The ZQ reset command (ZQRESET) resets the output impedance calibration to a default accuracy of ± 30 percent across process, voltage, and temperature. This command is used to ensure output impedance accuracy to ± 30 percent when ZQCS and ZQCL commands are not used.

One ZQCS command can effectively correct at least 1.5% (ZQ correction) of output impedance errors within tZQCS for all speed bins, assuming the maximum sensitivities specified are met. The appropriate interval between ZQCS commands can be determined from using these tables and system-specific parameters.

LPDDR3 devices are subject to temperature drift rate (TdriftrateE) and voltage drift rate (Vdriftrate) in various applications. To accommodate drift rates and calculate the necessary interval between ZQCS commands, apply the following formula:

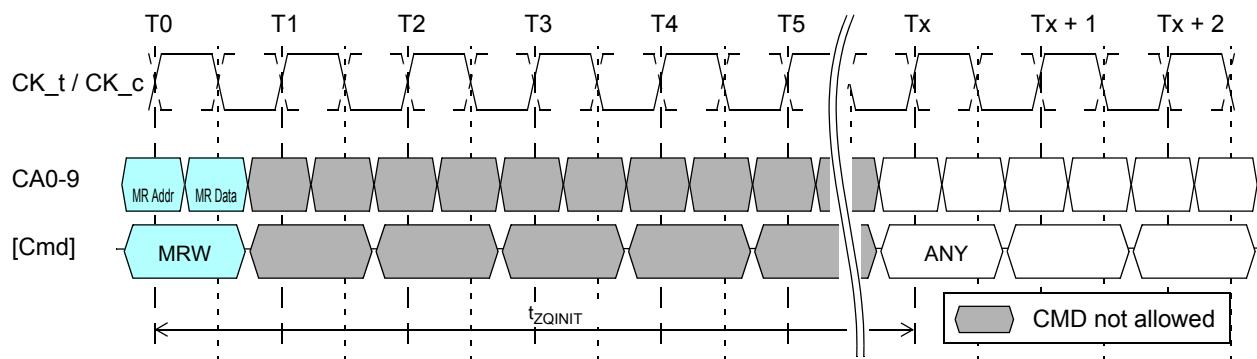
$$\frac{ZQ Correction}{(TSens \times Tdriftrate) + (VSens \times Vdriftrate)}$$

where TSens = max(dRONdT) and VSens = max(dRONdV) define the LPDDR3 temperature and voltage sensitivities.

For example, if TSens = 0.75% / °C, VSens = 0.20% / mV, Tdriftrate = 1°C / sec and Vdriftrate = 15 mV / sec, then the interval between ZQCS commands is calculated as:

$$\frac{1.5}{(0.75 \times 1) + (0.20 \times 15)} = 0.4s$$

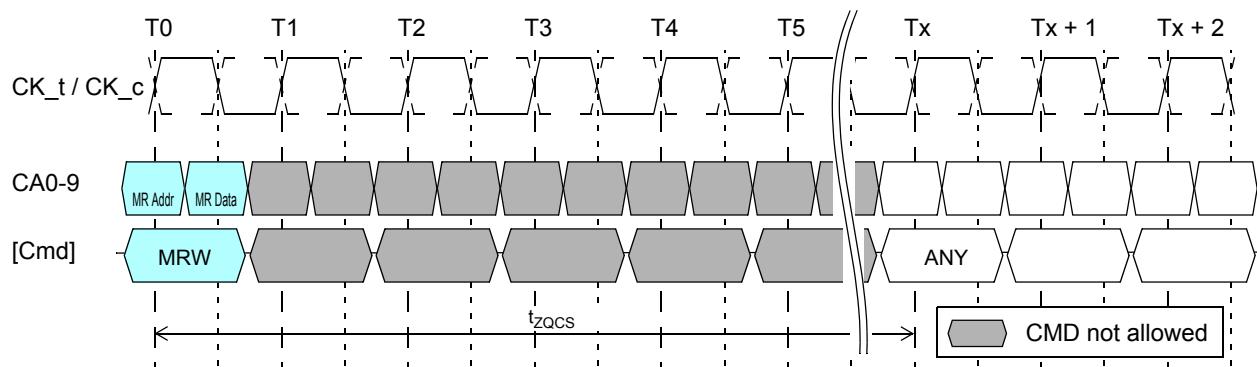
A ZQ calibration command can only be issued when the device is in the idle state with all banks precharged. ODT shall be disabled via the mode register or the ODT pin prior to issuing a ZQ calibration command. No other activities can be performed on the data bus and the data bus shall be un-terminated during calibration periods (tZQINIT, tZQCL, or tZQCS). The quiet time on the data bus helps to accurately calibrate output impedance. There is no required quiet time after the ZQ RESET command. If multiple devices share a single ZQ resistor, only one device can be calibrating at any given time. After calibration is complete, the ZQ ball circuitry is disabled to reduce power consumption. In systems sharing a ZQ resistor between devices, the controller must prevent tZQINIT, tZQCS, and tZQCL overlap between the devices. ZQ RESET overlap is acceptable.



Note:

1. Only the NOP command is supported during ZQ calibration.
2. CKE must be registered HIGH continuously during the calibration period.
3. All devices connected to the DQ bus should be High-Z during the calibration process

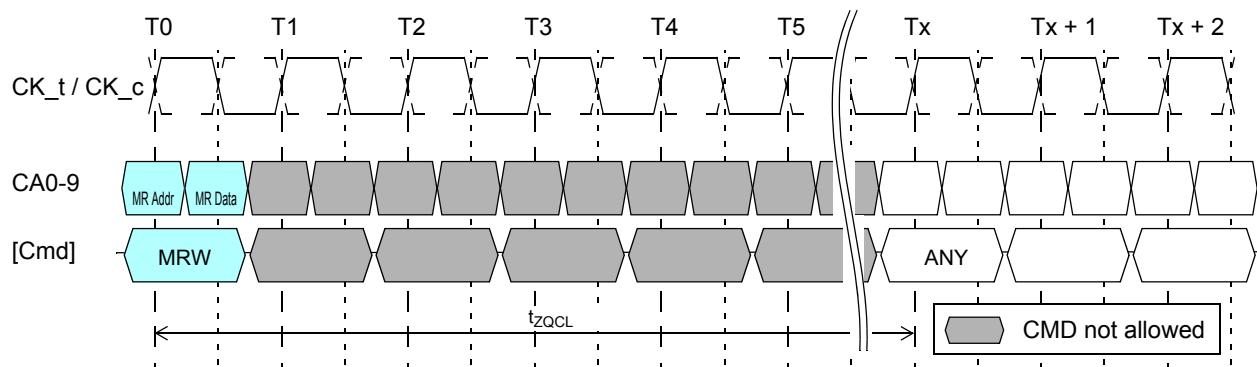
Figure. ZQ Calibration Initialization timing



Note:

1. Only the NOP command is supported during ZQ calibration.
2. CKE must be registered HIGH continuously during the calibration period.
3. All devices connected to the DQ bus should be High-Z during the calibration process.

Figure. ZQ Calibration Short timing



Note:

1. Only the NOP command is supported during ZQ calibration.
2. CKE must be registered HIGH continuously during the calibration period.
3. All devices connected to the DQ bus should be High-Z during the calibration process.

Figure. ZQ Calibration Long timing

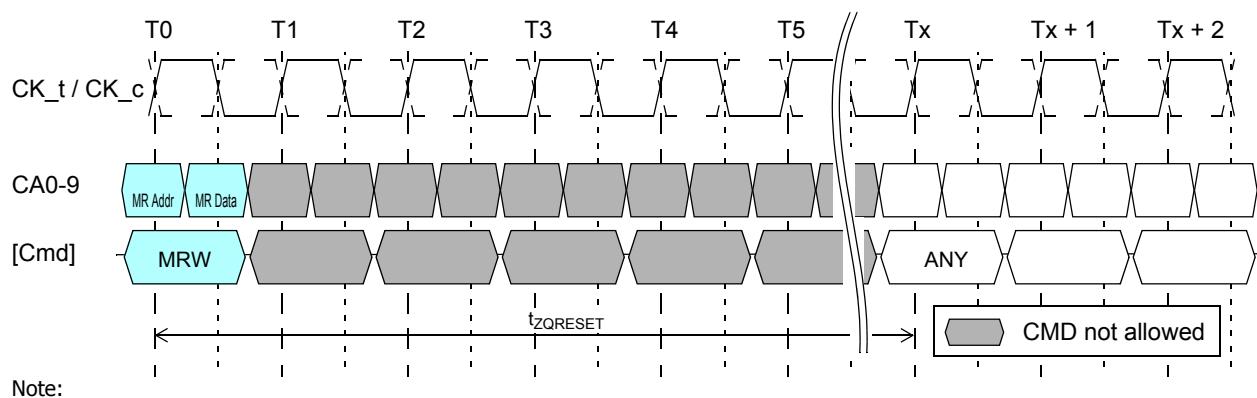


Figure. ZQ Calibration Reset timing example

ZQ External Resistor Value, Tolerance and Capacitive Loading

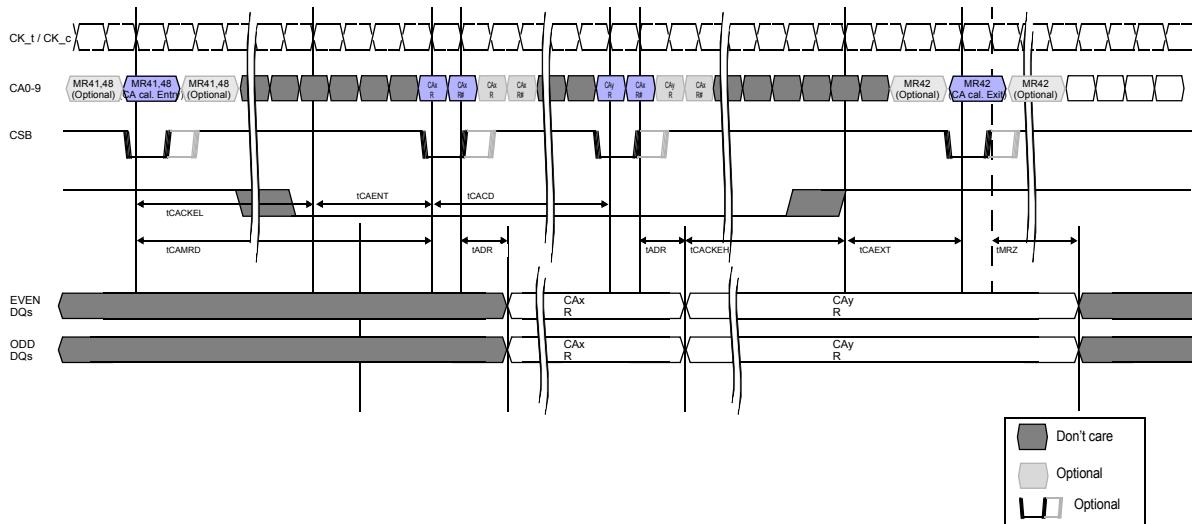
To use the ZQ calibration function, an $R_{ZQ} \pm 1\%$ tolerance external resistor must be connected between the ZQ pin and ground. A single resistor can be used for each device or one resistor can be shared between multiple devices if the ZQ calibration timings for each device do not overlap. The total capacitive loading on the ZQ pin must be limited (see Pin Capacitance table).

Mode Register Write - CA Training Mode

Because CA inputs are Double Data Rate, it may be difficult for memory controller to satisfy CA input setup/hold timings at higher frequency. CA Training mechanism is provided.

CA Training Sequence

- CA Training mode entry: Mode Register Write to MR#41
- CA Training session
 - Calibrate CA0, CA1, CA2, CA3, CA5, CA6, CA7 and CA8 (see the table "CA to DQ mapping (... #MR41)")
- CA to DQ mapping change: Mode Register Write to MR#48
- Additional CA Training session
 - Calibrate remaining CA pins (CA4 and CA9) (see the table "CA to DQ mapping (... #MR48)")
- CA Training mode exit: Mode Register Write to MR#42



Note:

- Unused DQ must be valid HIGH or LOW during data output period. Unused DQ may transition at the same time as the active DQ. DQS must remain static and not transition.
- CA to DQ mapping change via MR #48 omitted here for clarity of the timing diagram. Both MR41 and MR48 training sequences must be completed before exiting the training mode (MR42). To enable a CA to DQ mapping change, CKE must be driven HIGH prior to issuance of the MRW 48 command. For details, please refer to Sequence section.
- Because data out control is asynchronous and will be an analog delay from when all the CA data is available, tADR and tMRZ are defined from CK_t falling edge.
- It is recommended to hold the CA bus stable for one cycle prior to and one cycle after the issuance of the MRW Entry Command to ensure setup and hold timings on the CA bus.
- Clock phase may be adjusted in mode while CS_n is high and CKE is low resulting in an irregular clock with shorter/longer periods and pulse widths.**
- Optional MRW 41, 48, 42 command and CA calibration command are allowed. To complement these optional commands, optional CS_n assertions are also allowed. All timing must comprehend these optional CS_n assertions:
 - tADR starts at the falling clock edge after the last registered CS_n assertion.
 - tCADCD, tCACKEL, tCAMRD start with the rising clock edge of the last CS_n assertion.
 - tCAENT, tCAEXT need to be met by the first CS_n assertion.
 - tMRZ will be met after the falling clock edge following the first CS_n assertion with exit (MR42) command.

The LPDDR3 SDRAM may not properly recognize Mode Register Write command at normal operation frequency before CA Training is finished. Special encodings are provided for CA Training mode enable/disable.

MR#41 and MR#42 encodings are selected so that rising edge and falling edge values are the same. The LPDDR3 SDRAM will recognize MR#41 and MR#42 at normal operation frequency even before CA timing adjustment is finished.

Calibration data will be output through DQ pins. CA to DQ mapping is described in the table below.

After timing calibration with MR#41 is finished, users will issue MRW to MR#48 and calibrate remaining CA pins (CA4 and CA9) using DQ0,1,8,9 as calibration data output pins in the table below.

Table. CA Training mode enable (MR#41(29H, 0010 1001B), OP=A4H(1010 0100B))

	CA0	CA1	CA2	CA3	CA4	CA5	CA6	CA7	CA8	CA9
Rising Edge	L	L	L	L	H	L	L	H	L	H
Falling Edge	L	L	L	L	H	L	L	H	L	H

Table. CA Training mode disable (MR#42(2AH, 0010 1010B), OP=A8H(1010 1000B))

	CA0	CA1	CA2	CA3	CA4	CA5	CA6	CA7	CA8	CA9
Rising Edge	L	L	L	L	L	H	L	H	L	H
Falling Edge	L	L	L	L	L	H	L	H	L	H

Table. CA to DQ mapping (CA Training mode enabled with MR#41)

CA0	CA1	CA2	CA3	CA5	CA6	CA7	CA8	Clock edge
DQ0	DQ2	DQ4	DQ6	DQ8	DQ10	DQ12	DQ14	CK_t rising edge
DQ1	DQ3	DQ5	DQ7	DQ9	DQ11	DQ13	DQ15	CK_t falling edge

Table. CA Training mode enable (MR#48(30H, 0011 0000B), OP=C0H(1100 0000B))

	CA0	CA1	CA2	CA3	CA4	CA5	CA6	CA7	CA8	CA9
Rising Edge	L	L	L	L	L	L	L	L	H	H
Falling Edge	L	L	L	L	L	L	L	L	H	H

Table. CA to DQ mapping (CA Training mode enabled with MR#48)

CA4	CA9	Clock edge
DQ0	DQ8	CK_t rising edge
DQ1	DQ9	CK_t falling edge

Note: Other DQs must have valid output (either HIGH or LOW)

Mode Register Write - WR Leveling Mode

In order to provide for improved signal integrity performance, the LPDDR3 SDRAM provides a write leveling feature to compensate for timing skew, affecting timing parameters such as tDQSS, tDSS, and tDSH.

The memory controller uses the write leveling feature to receive feedback from the SDRAM allowing it to adjust the clock to data strobe signal relationship for each DQS_t/DQS_c signal pair. The memory controller performing the leveling must have adjustable delay setting on DQS_t/DQS_c signal pair to align the rising edge of DQS signals with that of the clock signal at the DRAM pin. The DRAM asynchronously feeds back CLK, sampled with the rising edge of DQS signals. The controller repeatedly delays DQS signals until a transition from 0 to 1 is detected. The DQS signals delay established through this exercise ensures the tDQSS specification can be met.

All DQS signals may have to be leveled independantly. During Write Leveling operations each DQS signal latches the clock with a rising strobe edge and drives the result on all DQ[n] of its respective byte.

The LPDDR3 SDRAM enters into write leveling mode when mode register MR2[7] is set HIGH. When entering write leveling mode, the state of the DQ pins is undefined. During write leveling mode, only NOP commands are allowed, or MRW command to exit write leveling operation. Upon completion of the write leveling operation, the DRAM exits from write leveling mode when MR2[7] is reset LOW.

The controller will drive DQS_t LOW and DQS_c HIGH after a delay of tWLQSEN. After time tWLMDR, the controller provides DQS signal input which is used by the DRAM to sample the clock signal driven from the controller. The delay time tWLMDR(max) is controller dependent. The DRAM samples the clock input with the rising edge of DQS and provides asynchronous feedback on all the DQ bits after time tWLO. The controller samples this information and either increment or decrement the DQS_t and/or DQS_c delay settings and launches the next DQS/DQS# pulse. The sample time and trigger time is controller dependent. Once the following DQS_t/DQS_c transition is sampled, the controller locks the strobe delay settings, and write leveling is achieved for the device. The figure below describes the timing for the write leveling operation.

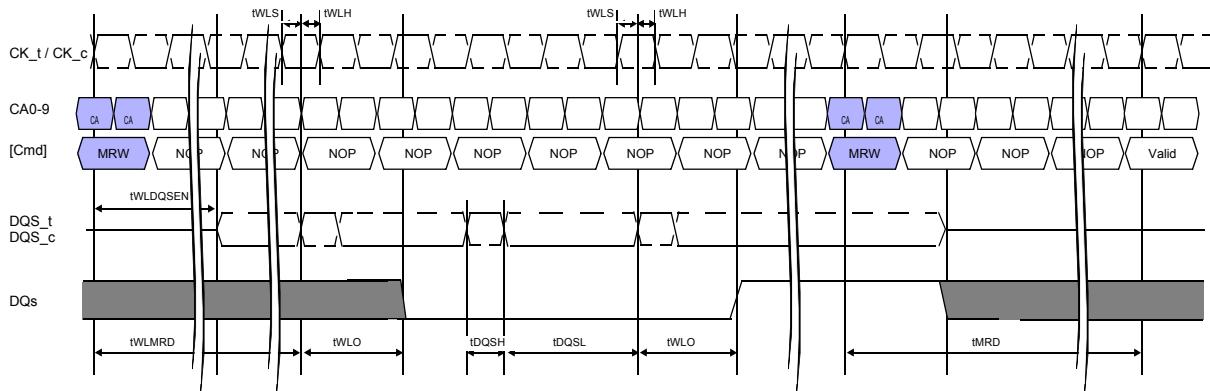


Figure. Write Leveling Timing diagram

On Die Termination (ODT)

ODT Functional Description

ODT (On-Die Termination) is a feature of the LPDDR3 SDRAM that allows the DRAM to turn on/off termination resistance for each DQ, DQS_t, DQS_c and DM via the ODT control pin. The ODT feature is designed to improve signal integrity of the memory channel by allowing the DRAM controller to independently turn on/off termination resistance for any or all DRAM devices. Unlike other command inputs, the ODT pin directly controls ODT operation and is not sampled by the clock.

The ODT feature is turned off and not supported in Self-Refresh mode. ODT operation can optionally be enabled during CKE Power Down via a mode register. Note that if ODT is enabled during Power Down mode VDDQ may not be turned off during Power Down. The DRAM will also disable termination during read operations.

A simple functional representation of the DRAM ODT feature is shown in the Figure "Functional Representation of ODT".

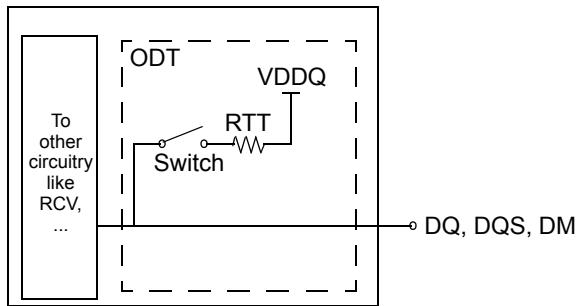


Figure. Functional Representation of ODT

The switch is enabled by the internal ODT control logic, which uses the external ODT pin and other mode register control information. The value of RTT is determined by the settings of Mode Register bits. The ODT pin will be ignored if the Mode Register MR11 is programmed to disable ODT, in self-refresh, in CKE power down (mode register option) and during read operations.

ODT Mode Register

The ODT Mode is enabled if MR11 OP<1:0> are non zero. In this case, the value of RTT is determined by the settings of those bits. The ODT Mode is disabled if MR11 OP<1:0> are zero.

MR11 OP<2> determines whether ODT, if enabled through MR11 OP<1:0>, will operate during CKE power down.

Asynchronous ODT

The ODT feature is controlled asynchronously based on the status of the ODT pin, except ODT is off when:..

- ODT is disabled through MR11 OP<1:0>
- DRAM is performing a read operation (RD or MRR)
- DRAM is in CKE Power Down and MR11 OP<2> is zero
- DRAM is in Self-Refresh mode
- DRAM is in CA Training Mode

In asynchronous ODT mode, the following timing parameters apply when ODT operation is controlled by the ODT pin: tODTon,min,max, tODToff,min,max.

Minimum RTT turn-on time (tODTonmin) is the point in time when the device termination circuit leaves high impedance state and ODT resistance begins to turn on. Maximum RTT turn on time (tODTonmax) is the point in time when the ODT resistance is fully on. tODTonmin and tODTonmax are measured from ODT pin high.

Minimum RTT turn-off time (tODToffmin) is the point in time when the device termination circuit starts to turn off the ODT resistance. Maximum ODT turn off time (tODToffmax) is the point in time when the on-die termination has reached high impedance. tODToffmin and tODToffmax are measured from ODT pin low.

ODT During Read Operations (RD or MRR)

During read operations, LPDDR3 SDRAM will disable termination and disable ODT control through the ODT pin. After read operations are completed, ODT control is resumed through the ODT pin (if ODT Mode is enabled).

ODT During Power Down

When MR11 OP<2> is zero, termination control through the ODT pin will be disabled when the DRAM enters CKE power down. After a power down command is registered, termination will be disabled within a time window specified by tOD-Td,min,max. After a power down exit command is registered, termination will be enabled within a time window specified by tODTe,min,max.

Minimum RTT disable time (tODTd,min) is the point in time when the device termination circuit will no longer controlled by the ODT pin. Maximum ODT disable time (tODTd,max) is the point in time when the on-die termination will be in high impedance.

Minimum RTT enable time (tODTe,min) is the point in time when the device termination circuit will no longer be in high impedance. The ODT pin shall control the device termination circuit after maximum ODT enable time (tODTe,max) is satisfied.

When MR11 OP<2> is enabled and MR11 OP<1:0> are non zero, ODT operation is supported during CKE power down with ODT control through the ODT pin.

ODT During Self Refresh

LPDDR3 SDRAM disables the ODT function during self refresh. After a self refresh command is registered, termination will be disabled within a time window specified by tODTd,min,max. After a self refresh exit command is registered, termination will be enabled within a time window specified by tODTe,min,max.

ODT During CA Training and Write Leveling

During CA Training Mode, LPDDR3 SDRAM will disable on-die termination and ignore the state of the ODT control pin. For ODT operation during Write Leveling mode, refer to the DRAM Termination Function In Write Leveling Mode Table for termination activation and deactivation for DQ and DQS_t/DQS_c.

ODT pin	DQS_t/DQS_c termination	DQ termination
de-asserted	OFF	OFF
asserted	ON	OFF

Table. DRAM Termination Function In Write Leveling Mode

If ODT is enabled, the ODT pin must be high, in Write Leveling mode.

	Write	Read/DQ cal	ZQ cal	CA Training	Write Leveling
DQ Termination	Enabled	Disabled	Disabled	Disabled	Disabled
DQS Termination	Enabled	Disabled	Disabled	Disabled	Enabled

Table. ODT States Truth Table

Note:

1. ODT is enabled with MR11[1:0]=01b, 10b or 11b and ODT pin HIGH. ODT is disabled with MR11[1:0]=00b or ODT pin LOW.

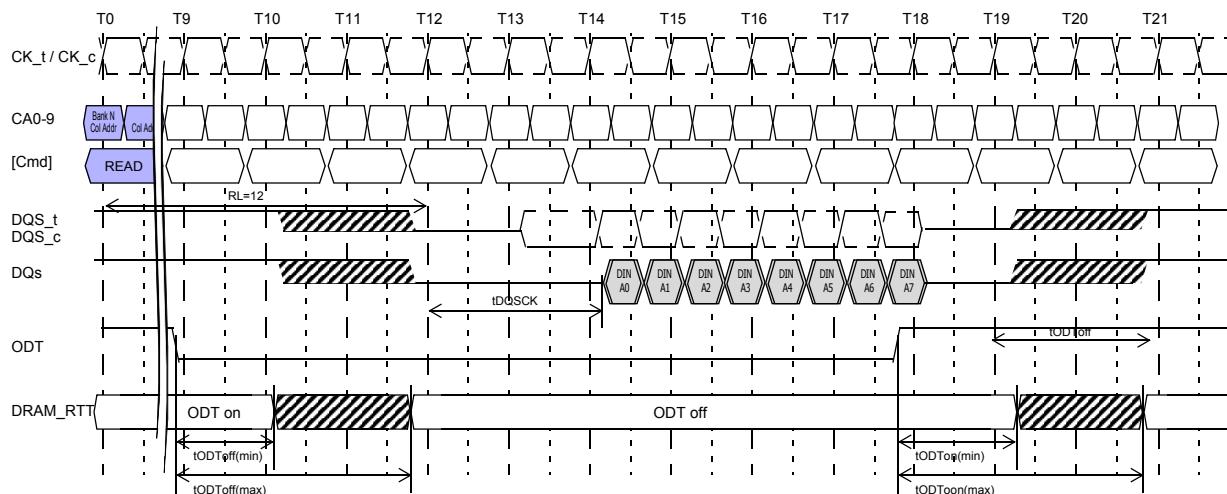


Figure. ASYNCHRONOUS ODT TIMING EXAMPLE FOR RL=12

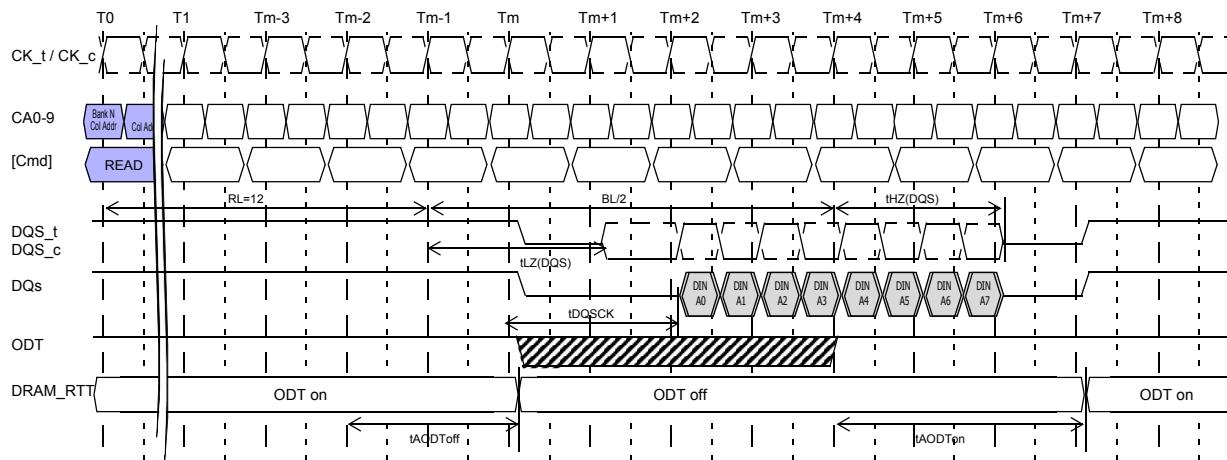


Figure. AUTOMATIC ODT TIMING DURING READ OPERATION EXAMPLE FOR RL=m

Note:

1. The automatic RTT turn-off delay, t_{AODT off}, is referenced from the rising edge of "RL-2" clock at Tm-2.
2. The automatic RTT turn-on delay, t_{AODT on}, is referenced from the rising edge of "RL+ BL/2" clock at Tm+4.

Power-down

Power-down is entered synchronously when CKE is registered LOW and CS_n is HIGH at the rising edge of clock. CKE must not go LOW while MRR, MRW, READ, or WRITE operations are in progress. CKE can go LOW while any other operations such as row activation, PRECHARGE, auto precharge, or REFRESH are in progress, but the power-down IDD specification will not be applied until such operations are complete. Power-down entry and exit are shown in following figures.

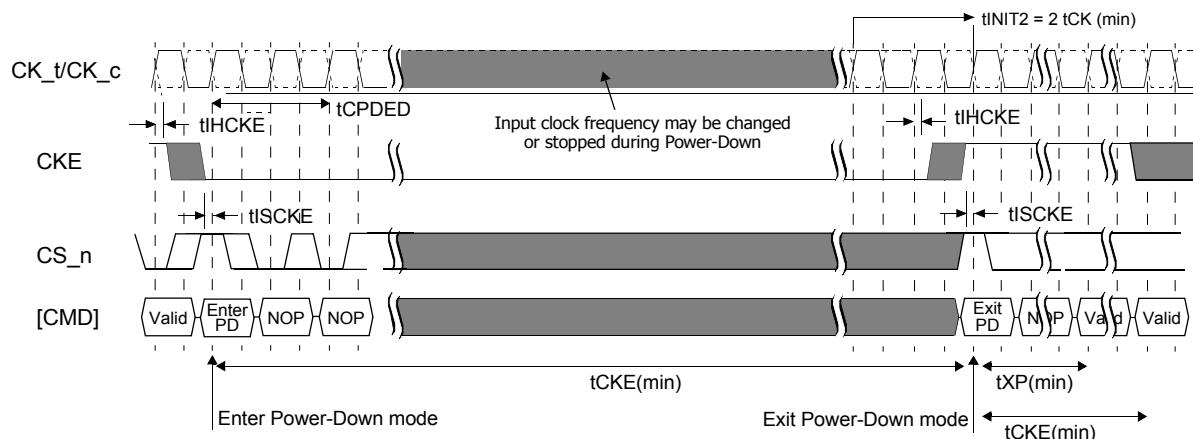
Entering power-down deactivates the input and output buffers, excluding CKE. To ensure that there is enough time to account for internal delay on the CKE signal path, two NOP commands are required after CKE is driven LOW, this timing period is defined as tCPDED. CKE LOW will result in deactivation of input receivers after tCPDED has expired. In power-down mode, CKE must be held LOW; all other input signals are "Don't Care." CKE LOW must be maintained until tCKE,min is satisfied. VREFCA must be maintained at a valid level during power-down.

VDDQ can be turned off during power-down. If VDDQ is turned off, VREFDQ must also be turned off. Prior to exiting power-down, both VDDQ and VREFDQ must be within their respective minimum/maximum operating ranges.

No refresh operations are performed in power-down mode. The maximum duration in power-down mode is only limited by the refresh requirements outlined in the Refresh command section.

The power-down state is exited when CKE is registered HIGH. The controller must drive CS_n HIGH in conjunction with CKE HIGH when exiting the power-down state. CKE HIGH must be maintained until tCKE is satisfied. A valid, executable command can be applied with power-down exit latency tXP after CKE goes HIGH. Power-down exit latency is defined in the AC timing parameter table.

If power-down occurs when all banks are idle, this mode is referred to as idle power-down; if power-down occurs when there is a row active in any bank, this mode is referred to as active power-down. For the description of ODT operation and specifications during power-down entry and exit, see section "On-Die Termination".



Note:

1. Input clock frequency can be changed or the input clock stopped during power-down, provided that the clock frequency is between the minimum and maximum specified frequencies for the speed grade in use, and that prior to power-down exit, a minimum of 2 stable clocks complete.

Figure. Basic power down entry and exit timing diagram

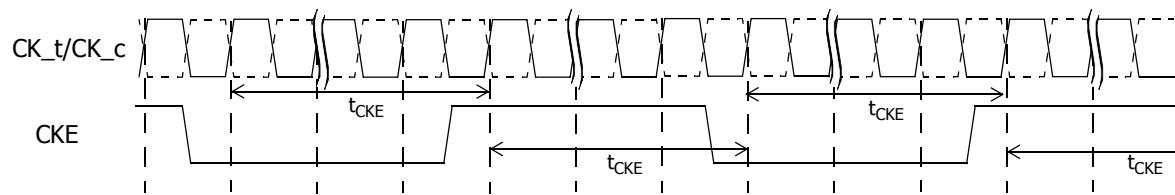
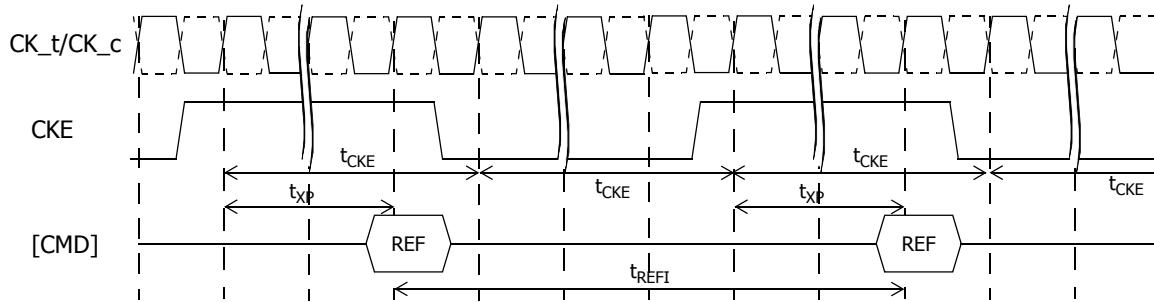


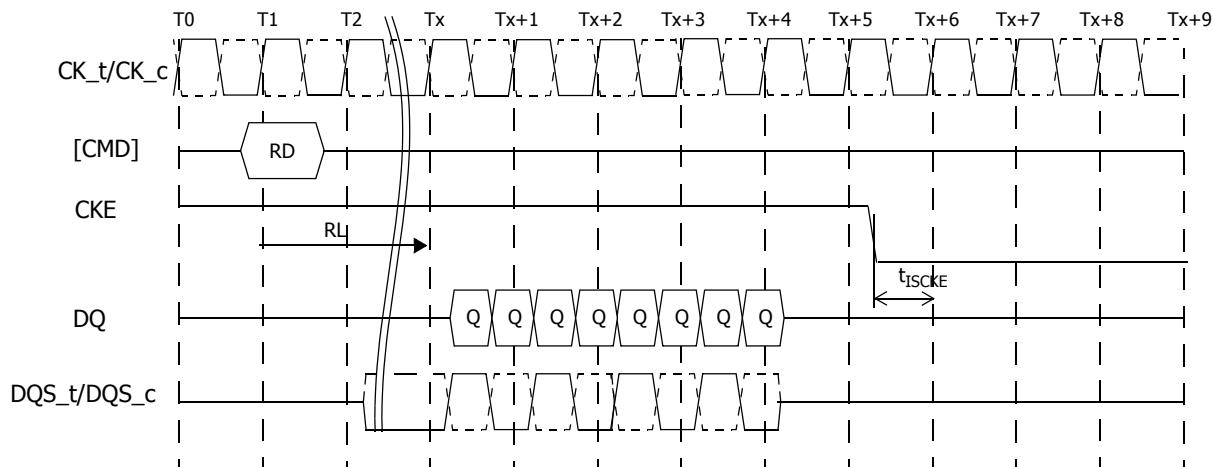
Figure. CKE intensive environment



Note:

1. The pattern shown can repeat over an extended period of time. With this pattern, all AC and DC timing and voltage specifications with temperature and voltage drift are ensured.

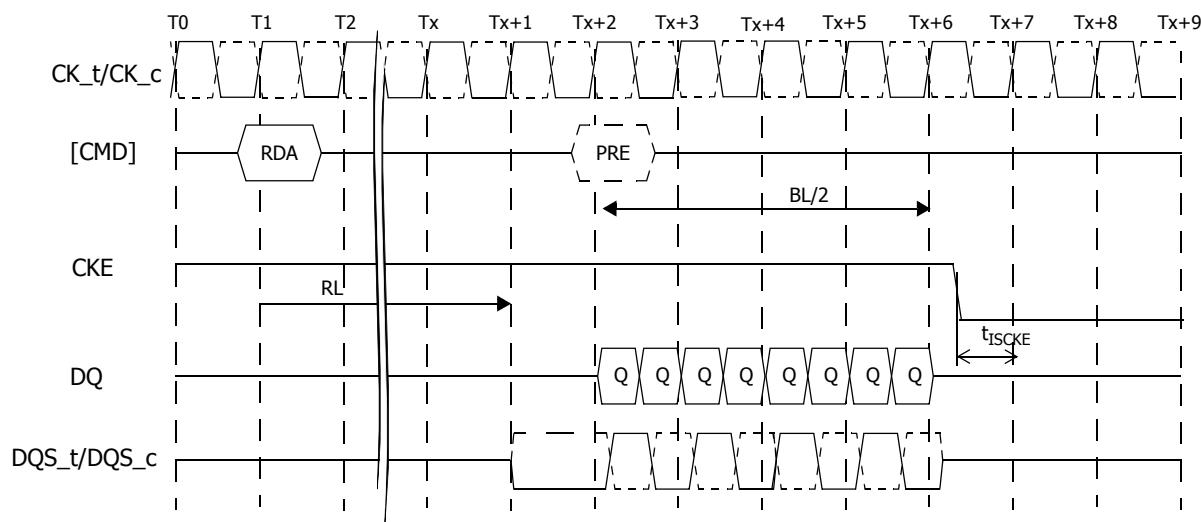
Figure. REFRESH to REFRESH timing with CKE intensive environment



Note:

1. CKE must be held HIGH until the end of the burst operation.
2. CKE can be registered LOW at $RL + RU(tDQSCK(MAX)/tCK) + BL/2 + 1$ clock cycles after the clock on which the READ command is registered.

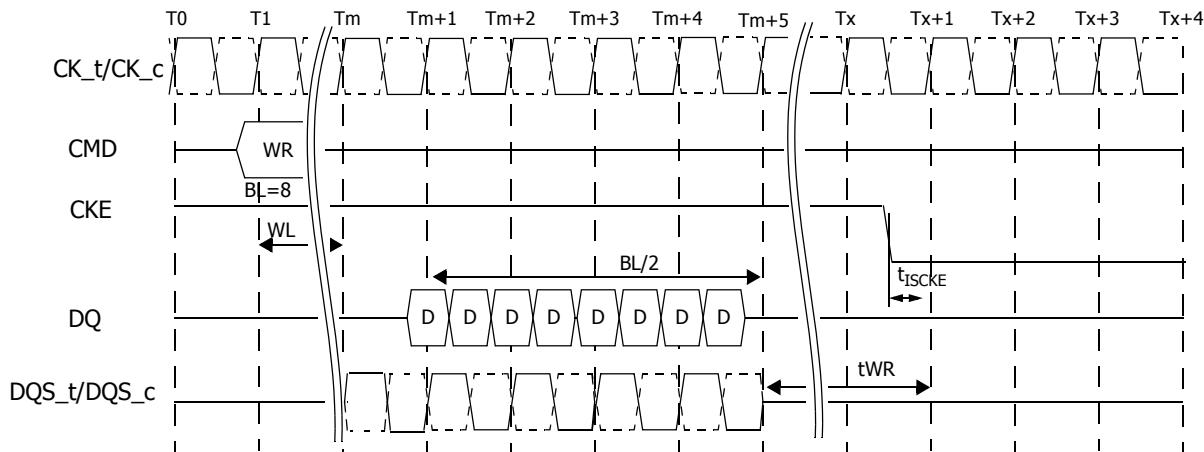
Figure. Read to power-down entry



Note:

1. CKE must be held HIGH until the end of the burst operation.
2. CKE can be registered LOW at $t_{DQSCK}/tCK + BL/2 + 1$ clock cycles after the clock on which the READ command is registered.
3. $BL/2$ with $t_{RTP} = 7.5\text{ns}$ and $t_{RAS} (\text{MIN})$ is satisfied.
4. Start internal PRECHARGE.

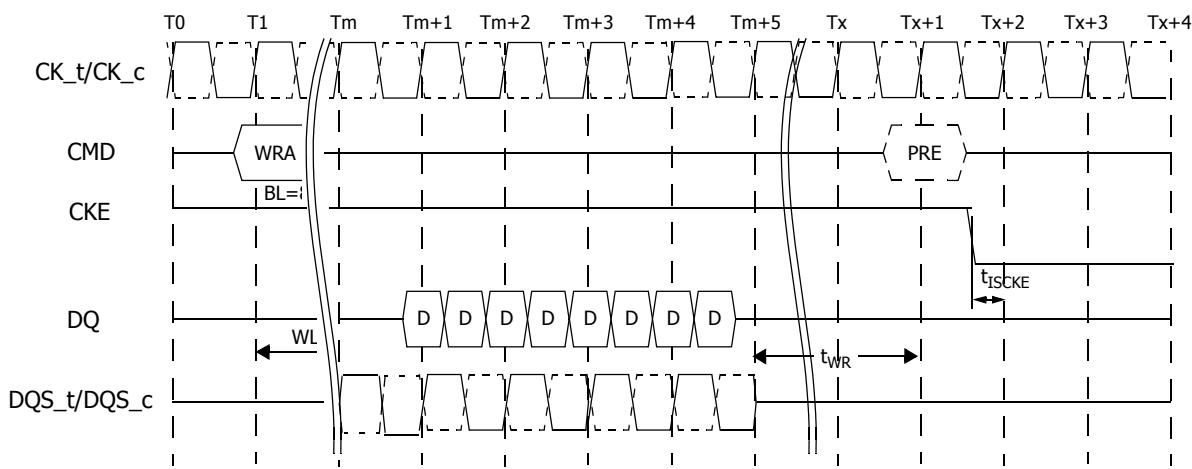
Figure. Read with Auto Precharge to power-down entry



Note:

1. CKE can be registered LOW at $WL + 1 + BL/2 + RU(t_{WR}/tCK)$ clock cycles after the clock on which the WRITE command is registered.

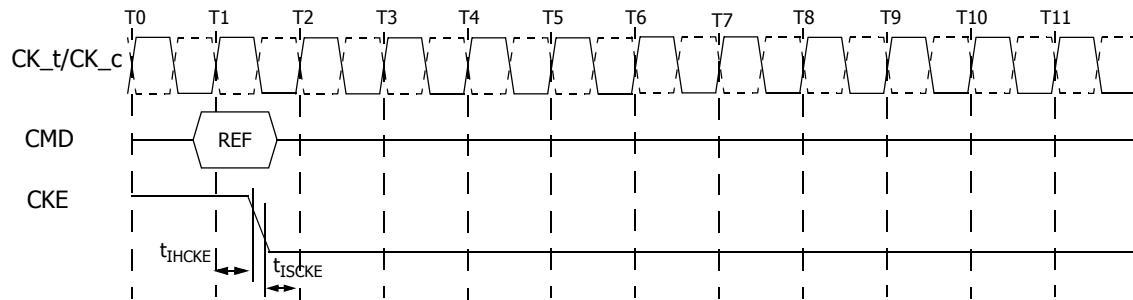
Figure. Write to power-down entry



Note:

1. CKE can be registered LOW at WL + 1 + BL/2 + RU(tWR/tCK) + 1 clock cycles after the WRITE command is registered.
2. Start internal PRECHARGE.

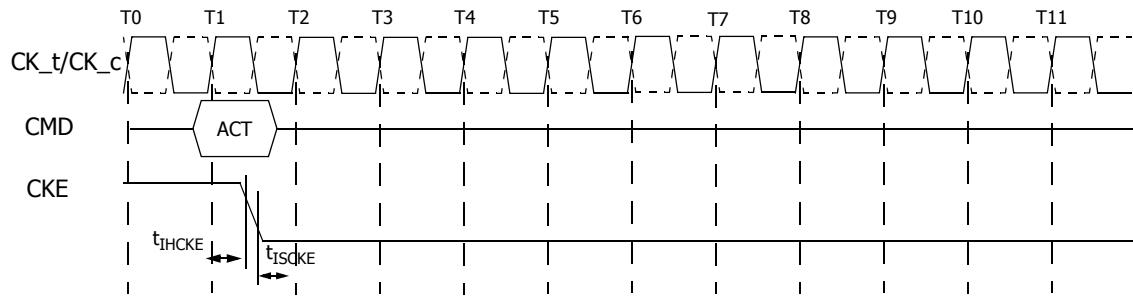
Figure. Write with auto precharge to power-down entry



Note.

1. CKE may go LOW tIHCKE after the clock on which the Refresh command is registered.

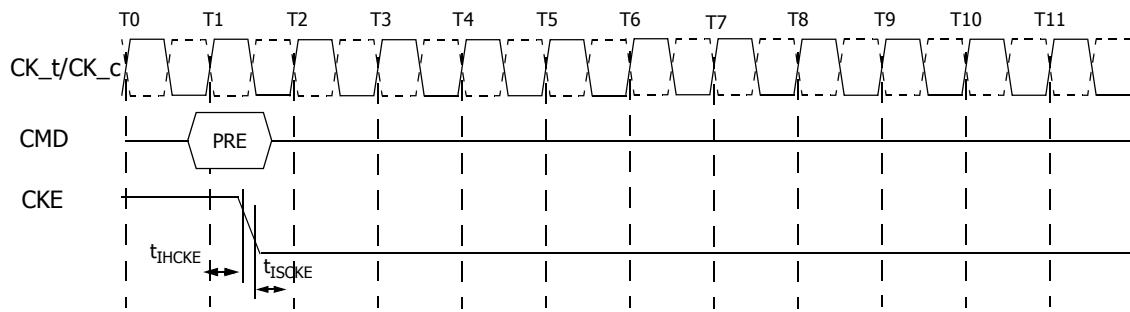
Figure. Refresh command to power-down entry



Note.

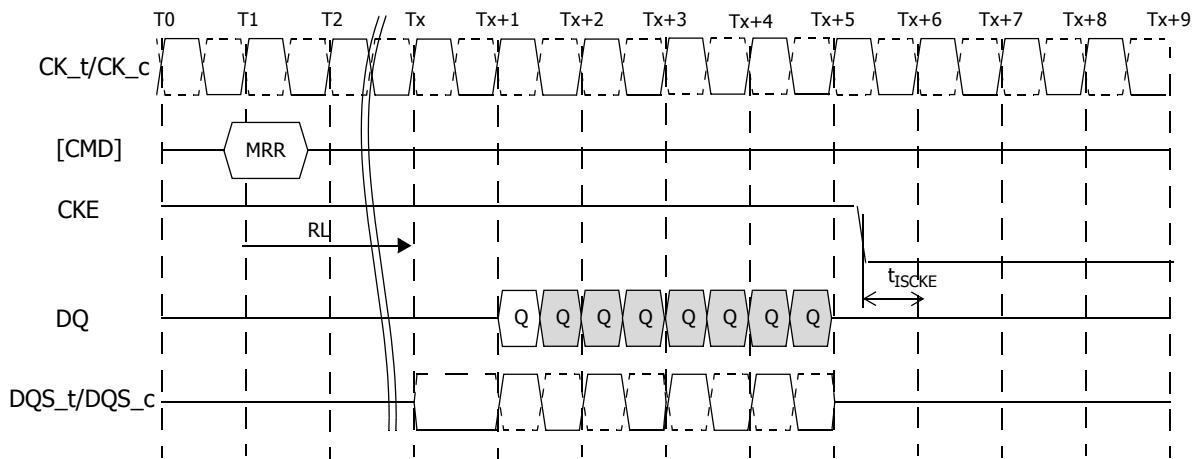
1. CKE may go LOW tIHCKE after the clock on which the Activate command is registered.

Figure. Activate command to power-down entry



Note. 1. CKE can go LOW t_{IHCKE} after the clock on which the PRECHARGE command is registered.

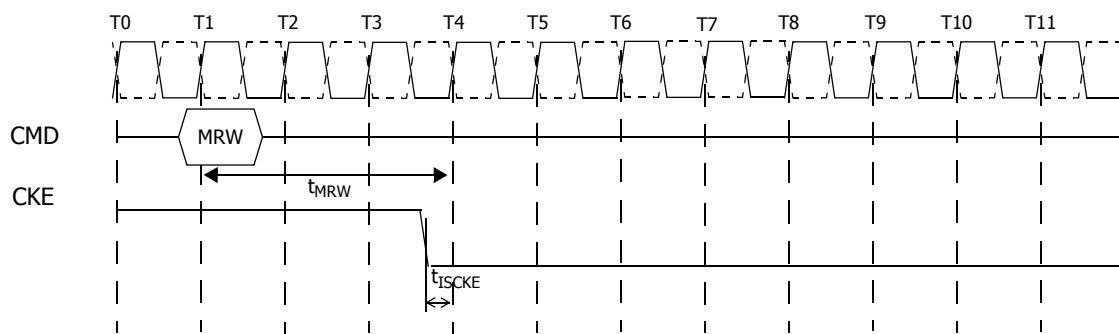
Figure. Precharge command to power-down entry



Note. 1. CKE can be registered LOW $RL + RU(t_{DQSCK}/tCK) + BL/2 + 1$ clock cycles after the clock on which the MRR command is registered.

2. CKE should be held high until the end of the burst operation.

Figure. MRR to power-down entry



Note. 1. CKE may be registered LOW t_{MRW} after the clock on which the Mode Register Write command is registered.

Figure. MRW command to power-down entry

Input clock stop and frequency change

LPDDR3 devices support input clock frequency change during CKE LOW under the following conditions:

- tCK(abs)min is met for each clock cycle;
- Refresh Requirements apply during clock frequency change;
- During clock frequency change, only REFab or REFpb commands may be executing;
- Any Activate or Precharge commands have executed to completion prior to changing the frequency;
- The related timing conditions (tRCD, tRP) have been met prior to changing the frequency;
- The initial clock frequency shall be maintained for a minimum of 3 clock cycles after CKE goes LOW;
- The clock satisfies tCH(abs) and tCL(abs) for a minimum of 2 clock cycles prior to CKE going HIGH.

After the input clock frequency is changed and CKE is held HIGH, additional MRW commands may be required to set the WR, RL etc. These settings may need to be adjusted to meet minimum timing requirements at the target clock frequency.

LPDDR3 devices support clock stop during CKE LOW under the following conditions:

- CK_t is held LOW and CK_c is held HIGH during clock stop;
- Refresh Requirements apply during clock stop;
- During clock stop, only REFab or REFpb commands may be executing;
- Any Activate or Precharge commands have executed to completion prior to stopping the clock;
- The related timing conditions (tRCD, tRP) have been met prior to stopping the clock;
- The initial clock frequency shall be maintained for a minimum of 3 clock cycles after CKE goes LOW;
- The clock satisfies tCH(abs) and tCL(abs) for a minimum of 2 clock cycles prior to CKE going HIGH.

LPDDR3 devices support input clock frequency change during CKE HIGH under the following conditions:

- tCK(abs)min is met for each clock cycle;
- Refresh Requirements apply during clock frequency change;
- Any Activate, Read, Write, Precharge, Mode Register Write, or Mode Register Read commands must have executed to completion, including any associated data bursts prior to changing the frequency;
- The related timing conditions (tRCD, tWR, tRP, tMRW, tMRR, etc.) have been met prior to changing the frequency;
- CS_n shall be held HIGH during clock frequency change;
- During clock frequency change, only REFab or REFpb commands may be executing;
- The LPDDR3 device is ready for normal operation after the clock satisfies tCH(abs) and tCL(abs) for a minimum of 2tCK + tXP.

After the input clock frequency is changed, additional MRW commands may be required to set the WR, RL etc. These settings may need to be adjusted to meet minimum timing requirements at the target clock frequency.

LPDDR3 devices support clock stop during CKE HIGH under the following conditions:

- CK_t is held LOW and CK_c is held HIGH during clock stop;
- CS_n shall be held HIGH during clock stop;
- Refresh Requirements apply during clock stop;
- During clock stop, only REFab or REFpb commands may be executing;
- Any Activate, Read, Write, Precharge, Mode Register Write, or Mode Register Read commands must have executed to completion, including any associated data bursts prior to stopping the clock;
- The related timing conditions (tRCD, tWR, tRP, tMRW, tMRR, etc.) have been met prior to stopping the clock;
- The LPDDR3 device is ready for normal operation after the clock is restarted and satisfies tCH(abs) and tCL(abs) for a minimum of 2tCK + tXP.

No Operation command

The purpose of the No Operation command (NOP) is to prevent the LPDDR3 device from registering any unwanted command between operations. Only when the CKE level is constant for clock cycle N-1 and clock cycle N, a NOP command may be issued at clock cycle N. A NOP command has two possible encodings:

1. CS_n HIGH at the clock rising edge N.
2. CS_n LOW and CA0, CA1, CA2 HIGH at the clock rising edge N.

The No Operation command will not terminate a previous operation that is still executing, such as a burst read or write cycle.