

# Bayesian Dictionary Learning for EEG Source Identification

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Mathematical Engineering, MATTEK

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# Preface

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Aalborg University, January 22, 2020

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# Contents

<b>Preface</b>	<b>v</b>
<b>Introduction</b>	<b>3</b>
<b>1 Motivation</b>	<b>5</b>
1.1 EEG Measurements . . . . .	5
1.2 Modelling . . . . .	7
1.3 Related Work and Our Objective . . . . .	8
<b>2 Problem Statement</b>	<b>11</b>
<b>3 Modelling of Model</b>	<b>13</b>
3.1 Single Measurement Vector Model . . . . .	13
3.2 Multiple Measurement Vector Model . . . . .	14
3.3 Solution Methods . . . . .	15
<b>4 Covariance-Domain Dictionary Learning</b>	<b>17</b>
4.1 Introduction to Compressive Sensing . . . . .	18
4.2 K-SVD . . . . .	20
4.3 Covariances domain representation . . . . .	23
4.4 Determination of the Dictionary . . . . .	24
<b>5 Multiple Sparse Bayesian Learning</b>	<b>29</b>
5.1 Maximum a Posterior Estimation . . . . .	29
5.2 Empirical Bayesian Estimation . . . . .	30
<b>Bibliography</b>	<b>35</b>
<b>A Extended ICA Algorithms</b>	<b>37</b>
A.1 Fixed-Point Algorithm - FastICA . . . . .	37





# Introduction

The problem addressed throughout this thesis arise from the increasing use of electroencephalographic measurements for a wide range of scientific purposes, especially within the medical field. An electroencephalography captures electric signals caused by activity within the brain. The signals from the brain is recorded over time by multiple sensors placed on the scalp. One essential issue concerning an electroencephalography is to extract the exact sources of the captured brain activity. This is of interest when studying correlation among activities in different parts of the brain, referred to as functional integration. The recorded signal from one sensor is basically a mixture of electric signals released from a various number of active neurons within the brain, forming one or several sources. Furthermore, this mixture is distorted as it travels through the scalp. The need for source extraction is confirmed by studies showing how analysis performed on electroencephalographic measurements differs significantly from similar analysis performed directly on the original source[12].

Considering this issue of source extraction from a mathematical perspective the electroencephalographic measurements can be modelled by a linear system of equations, from which it is possible to extract a limited number of sources under certain conditions. However, it is a general acknowledged issue that the true number of sources is unknown. The task complexity of extracting the sources from the linear system is increased in cases where the number of sources exceeds the number of electrodes providing measurements.

This thesis explores the state of the art mathematical method for source extraction embracing the case of more sources than sensors. Overall this method, published in 2015, consist of two steps, that is finding receptively the mixture the signals have undergone and then extracting the source signals. The two steps originates from two different approaches considering the mathematical orientation. The main goal of the thesis is to explore and unite the necessary theory into one algorithm. The practical aspect will include an implementation of the algorithm to be tested on new electroencephalographic measurements with the purpose of supporting(?) the current results. Furthermore the problem is connected to an current application within the hearing aid industry. Here the intention is to reduce the amount of energy spent by the hearing aid user. Basically, this is attempted by identifying the listening direction

intended by the user, from analysis of the active sources measured on the user. In this thesis the number of active sources are sought related to the amount of energy used by the hearing aid user this includes considerations upon the issue of the true number of active sources being unknown.

(følgende kan skrives bedre hvis ikke det skal være et andet sted)The thesis consist of a motivational part introducing electroencephalography and the potential use within research especially in the hearing aid industry. Furthermore, existing literature considering different mathematical approaches for source extraction are examined. The Motivational part is concluded by the problem statement specifying the objective of the thesis. Next is the theoretical part. The system model is specified and the solution approach are presented. The necessary theory are introduced leading to the state of the art algorithms for source extraction. The theoretical part is followed by implementation and test of the algorithm for verification. Next is.. Finally discussion and conclusion upon the achieved results are presented followed by a consideration upon further studies.

# Chapter 1

## Motivation

This chapter accounts for the motivation behind source extraction from an Electroencephalography (EEG). The concept of EEG is introduced along with its application. The potential and importance of source extraction are considered and related to the hearing aid industry. The commonly applied mathematical model for EEG measurements is presented. Currently applied methods for source extraction are considered leading to a presentation of the current state of the art methods which succeeds to overcome the limitations of previous methods. Lastly the contribution provided in this thesis is presented.

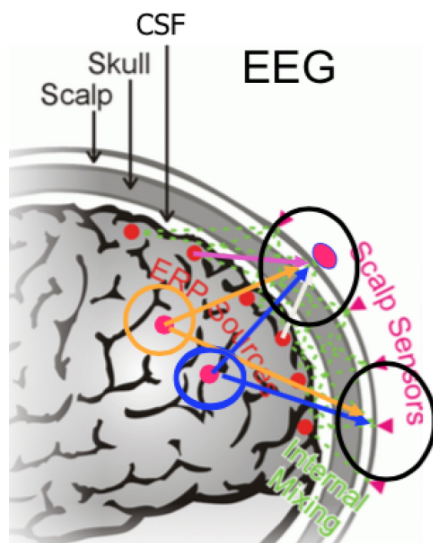
### 1.1 EEG Measurements

EEG is an imaging technique used within the medical field. EEG is measuring electric signals on the scalp, caused by brain activity. The human brain consist of an enormous amounts of cells, called neurons, which are mutually connected in neural nets. When a neuron is activated, for instance by a physical stimuli, local current flows are produced among the neurons. This is considered as neural interaction across different parts of the human brain[21].

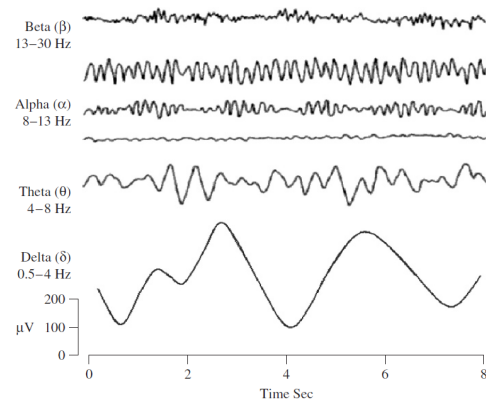
EEG measurements are provided by a number of metal electrodes, referred to as sensors, carefully placed on the human scalp. Each sensor reads the present electrical signals over time. It takes a large amount of simultaneous active neurons to generate an electrical signal that is recordable on the scalp. Such signal is considered as one *source* of neural activity. The current has to penetrate the skull, skin and several other thin layers causing a reduced signal to reach the sensor. It is possible that the measurement of one sensor is a sum of multiple signals from different sources. Nor is the range of a single sensor separated from the other sensors. Thus the same signal can easily be measured by two or more sensors. Furthermore, interfering signals can occur in the measurements resulting from physical movement of e.g. eyes and jawbone [21]. Lastly, the transmission of the electric signal through the biological

tissue to the sensor has an unknown mixing effect on the signal. This process is called volume conduction [18, p. 68] [19]. This clarifies the mixture of fluctuating electrical signals with noise that forms the EEG measurements. The concept is sought illustrated on figure 1.1. The resulting EEG measurements are classified within four groups according to the dominant frequency. The delta wave (0.5 – 4 Hz) is observed from infants and sleeping adults, the theta wave (4 – 8 Hz) is observed from children and sleeping adults, the alpha wave (8 – 13 Hz) is the most extensively studied brain rhythm, which is induced by an adult laying down with closed eyes. Lastly, the beta wave (13 – 30 Hz) is considered the normal brain wave for adults, associated with active thinking, active attention or solving concrete problems [18, p. 11]. An example of EEG measurements within the four categories is illustrated by figure 1.2.

tilføj beskrivelse af stationarit



**Figure 1.1:** Illustration of volume conduction, source: [6](we will make our own figure here instead)



**Figure 1.2:** Example of time dependent EEG measurements within the four defined categories, source: [18]

EEG performed on humans and animals have a great number of applications with both clinical and research purposes. Examples of clinical applications covers diagnosis and management of neurological disorders such as epilepsy and monitor alertness regarding coma or brain death. EEG capitalizes on the procedure being non-invasive and fast. Neural activity can be measured within fractions of a second after a stimuli has been provided. These advantages contributes to the wide range of applications within research of the neural processes involved in or resulting from actions, emotions or cognition. Today such neural research are use in many different fields[21, p. 4]. The hearing aid industry is one example where this research is highly prioritized. At Eriksholm research center, which is a part of the hearing aid manufacturer Oticon, cognitive hearing science is a research area within fast

development [20]. One main purpose at Eriksholm is to make it possible for a hearing aid to identify the user-intended sound source from real time EEG measurements and thereby exclude noise from elsewhere [2] [7]. It is essentially the well known but unsolved cocktail problem which is sought improved by use of EEG. This is where EEG and occasionally so called in-ear EEG is interesting. In conjunction with the technology of beamforming it is possible for a hearing aid to receive only signals from a specific direction.

Over the past two decades, functional integration has become an area of interest regarding EEG research [11]. Within neurobiology functional integration refers to the study of the correlation among activities in different regions of the brain. In other words, how do different parts of the brain work together to process information and conduct a response [12]. For this purpose separation and localization of the original sources which contribute to the EEG measurement is of interest. An article from 2016 [19] points out the importance of performing analysis regarding functional integration at source level rather than at EEG level. It is argued through experiments that analysis at EEG level does not allow interpretations about the interaction between sources. This emphasize a potential for improving results within a wide range of EEG research, if the original active sources can be extracted from a specific EEG measurements.

## 1.2 Modelling

Consider the issue of extracting the activated sources from EEG measurements. A known approach is to model the observed data by a linear system

$$\mathbf{Y} = \mathbf{A}\mathbf{X}.$$

$\mathbf{Y} \in \mathbb{R}^{M \times L}$  is the EEG measurements where  $L$  is the number of samples over time with each sample consisting of  $M$  measurements, one for each sensor.  $\mathbf{X} \in \mathbb{R}^{N \times L}$  makes  $L$  samples of  $N$  possible sources within the brain. The non-zero entries of a column in  $\mathbf{X}$  represent the active sources at the time of the sample.  $\mathbf{A} \in \mathbb{R}^{M \times N}$  is an unknown mixing matrix, resembling the volume conduction. The  $i^{th}$  column of  $\mathbf{A}$  represents the relative projection weights from the  $i^{th}$  source to every sensor [6]. Due to the input  $\mathbf{Y}$  being a matrix of multiple samples vectors of measurements the linear system is in general referred to as a multiple measurement vector model. Only the measurements  $\mathbf{Y}$  are known hence it is not possible to solve the linear system with respect to  $\mathbf{X}$  using basic linear algebra. The task in this case is to identify both  $\mathbf{A}$  and then  $\mathbf{X}$ , given the measurements  $\mathbf{Y}$ . This problem is referred to as the inverse problem of EEG. Finding  $\mathbf{X}$  from the inverse problem is referred to as source separation and localization. Separation is to find the signal from each active source and localization is to place each active source signal at the right position within the

source vector of dimension  $N$ , where  $N$  is the maximum number of sources to be active.

Independent Component Analysis (ICA) is one commonly applied method to solve the inverse problem of EEG [15], [14]. ICA is a technique to find the matrix  $A$  such that the column wise elements of  $X$  is statistically independent. Thus statistical independence between the active sources is the essential assumption, which in the case of EEG are considered valid due to the volume conduction being effectively instantaneous [14, p. 3]. Application of ICA has shown great results regarding source separation of high-density EEG. However, a significant flaw to this method is that the EEG measurements are only separated into a number of sources that is equal to or less than the number of sensors [4]. Meaning that the EEG inverse problem can not form an under-determined system, which is the case when the maximum number of unknown sources  $N$  exceeds the number of sensors  $M$ . Such assumption undermines the reliability and usability of ICA, as the number of active sources easily exceed the number of sensors [6]. This is especially a drawback when low-density EEG are considered. Low-density EEG measurements are collected from equipment with less than 32 sensors, increasing the changes of  $M$  being less than  $N$ . However, improved capabilities of low-density EEG devices are desirable due to their relative low cost, mobility and ease to use.

This argues the importance of considering the inverse problem of EEG in the under-determined case where  $N > M$ . In the next section existing work considering the under-determined inverse problem of EEG is investigated further.

### 1.3 Related Work and Our Objective

As mentioned above ICA is a solid method for source separation in the case where separation into a number of sources equal to the number of sensors is adequate. The issue occurs in cases where the number of sources  $N$  exceeds the number of sensors  $M$ . To overcome this issue an extension of ICA was suggested, referred to as the ICA mixture model [4]. Instead of identifying one overcomplete mixing matrix  $\mathbf{A} \in \mathbb{R}^{M \times N}$  this approach learns  $N_{\text{model}}$  different mixing matrices  $\mathbf{A}_i \in \mathbb{R}^{M \times M}$ , to make computations more tractable. This method was further adapted into the Adaptive Mixture ICA (AMICA) which showed successful results regarding identification of more sources than sensors [17]. However, the successful results relies on the assumption that no more than  $M$  out of  $N$  possible sources is simultaneously active. That is explicit that the source vector of dimension  $N$  has at most  $M$  non-zero entries. This assumption is still an essential limitation to the frame work, especially when considering low-density EEG. Other types of ICA algorithms for under-determined systems have been proposed, without overcoming the limitation of jointly active sources exceeding the number of sensors.

In 2015 O. Balkan et. al. suggested a new approach also targeting the iden-

tification of more active sources than sensors regarding EEG measurements. One method is proposed for learning  $\mathbf{A}$  from  $\mathbf{Y}$  [4] and a different method is proposed for finding  $\mathbf{X}$  given  $\mathbf{Y}$  and  $\mathbf{A}$  [5].

To learn  $\mathbf{A}$  the suggested method, referred to as Cov-DL, is a covariance-domain based dictionary learning algorithm. The method is based upon theory of dictionary learning and compressive sensing, dictating a framework for solving an under-determined system when  $\mathbf{X}$  is sufficiently sparse. The point is to transfer the EEG measurements into the covariance domain, for which higher dimensionality can be achieved compared to the original EEG sensor domain of size  $M$ . The transformation can be done when assuming the volume conduction is linear and that sources are uncorrelated. As a result the theory of compressive sensing is found to apply, allowing to learn  $\mathbf{A}$  through dictionary learning and an inverse transformation. The Cov-DL algorithm stands out from other straight forward dictionary learning methods as it does not rely on the sparsity of active sources. This is an essential advantage when low-density EEG is considered. Cov-DL was tested and found to outperform AMICA [4]. As mentioned, the Cov-DL algorithm only learns the mixing matrix  $\mathbf{A}$ , resembling the volume conduction, and not the explicit source activity  $\mathbf{X}$ .

For the purpose of recovering  $\mathbf{X}$ , from  $\mathbf{Y}$  and  $\mathbf{A}$ , a multiple measurement sparse Bayesian learning (M-SBL) algorithm is proposed. This method is also targeting the case of more active sources than sensors. ... (lidt nærmere tilkendegivelse af grundlæggende teori)... The method was proven to outperform the previously used algorithms, even when the defined recovery conditions regarding the found mixing matrix  $\mathbf{A}$  was not fulfilled [5].

One drawback, which is not fully covered in the referred literature, is that the two methods rely on the number of active sources being known. In practise this is not the case. Hence an estimation of the number of active sources has to be considered for the algorithm to be useful in practice. To address this issue a simple approach is to optimise the result with respect to the number active source, provided that some prior assumption of the expected result can be made.

The two state of the art methods resulting in source separation and localization makes the foundation of this thesis. The aim of this thesis is to investigate and fully understand the two methods in order to implement and test a joint algorithm. Secondary it is of interest to consider the practical application of the results within a hearing aid. As described in section 1.1 it is of interest to reduce the amount of energy it takes to listen to a specific sound source surrounded by noise. For this purpose we want to relate the found number of active sources to the level of concentration that the test person is experiencing. As described this will include an estimation of the number of active sources through

test lige kode igen hvor forkert k angives

kan vi undersøge dette yderligere ovenfor først





## Chapter 2

# Problem Statement

From the motivation and related work described in chapter 1 it is concluded that EEG measurement of the brain activity has great potential to contribute within the hearing aid industry, regarding the development of hearing aids with improved performance in situations as the cocktail party problem. By solving the overcomplete EEG inverse problem, in order to localize the sources of the brain activity, the results could be used to guide and adapt the performance of a hearing aid – e.g. by moving the microphone beam in the direction of interest. This leads to the following problem statement.

*How can sources of activation within the brain be recovered from the EEG inverse problem, in the over-complete case of less sensors than sources, and how can this recovery process be implemented as a real-time application providing feedback to improve the listening experience?*

From the problem statement some clarifying sub-questions have been made.

- How can Cov-DL be used to estimate the mixing matrix  $\mathbf{A}$  from the overcomplete EEG inverse problem?
- How can M-SBL be used to estimate the source matrix  $\mathbf{X}$  from the overcomplete EEG inverse problem?
- How can the above methods be implemented as one application performing source recovery from EEG measurements in real-time.
- How can the real-time feedback of the system be analysed and used to control the microphone beam of a simulated hearing aid.



## Chapter 3

# Modelling of Model

Through this chapter a model representing the EEG measurements will be introduced. Along the model different terminologies will be introduced and described for further use in this master thesis.

At last some solution methods will be described and will be the ones this master thesis will investigate and use to estimate the parameters of for EEG measurements model.

### 3.1 Single Measurement Vector Model

Some measurement vector  $\mathbf{y} \in \mathbb{R}^M$  can be described as a linear combination of a coefficient matrix  $\mathbf{A} \in \mathbb{R}^{M \times N}$  and some vector  $\mathbf{x} \in \mathbb{R}^N$  such that

$$\mathbf{y} = \mathbf{A}\mathbf{x}. \quad (3.1)$$

The measurement vector  $\mathbf{y}$  consists of  $M$  measurements,  $\mathbf{x}$  is an unknown vector of  $N$  elements, and the coefficient matrix  $\mathbf{A}$  models the linear measurement process column-wise. 3.1 makes a system of linear equations with  $M$  equations and  $N$  unknowns and will be referred to as a linear system for the rest of the master thesis. In other literature, the linear system is also referred to as a single measurement vector (SMV) model.

To solve the linear system one must look at the different cases that occur for the number of  $M$  equations and  $N$  unknowns. In the case where the number of equations equals the number of unknowns,  $M = N$ , the coefficient matrix  $\mathbf{A}$  becomes a square matrix. A solution for the linear system,  $\mathbf{x}^*$ , can be found, provided that a solution exists, if the square coefficient matrix  $\mathbf{A}$  has full rank –  $\mathbf{A}$  consists of linearly independent columns or rows – as the square matrix can be inverted and the solution is achieved by

$$\mathbf{x}^* = \mathbf{A}^{-1}\mathbf{y}.$$

Such linear system, when  $M = N$ , is called determined and there will exist a unique solution, provided that a solution exist. In the cases where  $M > N$  and  $M < N$  the linear system is called over-determined and under-determined, respectively. For an over-determined system there does not exist a solution and for under-determined systems there exist infinitely many solutions, provided that one solution exist [8, p. ix].

From chapter 1 the linear system of interest consists of a observed EEG measurement vector  $\mathbf{y} \in \mathbb{R}^M$  with  $M$  sensors and a unknown source vector  $\mathbf{x} \in \mathbb{R}^N$  with  $N$  sources. The matrix  $\mathbf{A} \in \mathbb{R}^{M \times N}$  is the mixing matrix. As described in chapter 1, the case of interest is an under-determined linear system,  $M < N$ , as the EEG measurement have less sensors than sources – hence a solution has to be found within the infinite solution set. For the source vector  $\mathbf{x}$  the activation of the sources is denoted by  $k$  and there are at most  $k \leq N$  activations in the EEG measurements. To find the solution of the linear system (3.1) the number of activations  $k$  must also be included in the recovering.

The number of  $k$  active source in the source vector  $\mathbf{x} \in \mathbb{R}^N$  can be view as non-zeros – the non-activations are described by zeros. To count non-zeros entries of a vector the  $\ell_0$ -norm can be used:

$$\|\mathbf{x}\|_0 := \text{card}(\text{supp}(\mathbf{x})).$$

The function  $\text{card}(\cdot)$  gives the cardinality of the input and the support of  $\mathbf{x}$ , consisting of the non-zeros entries, is given as

$$\text{supp}(\mathbf{x}) = \{j \in [N] : x_j \neq 0\},$$

where  $[N]$  is a set of integers  $\{1, 2, \dots, N\}$  [10, p. 41].

### 3.2 Multiple Measurement Vector Model

As EEG measurements vary over time multiple EEG measurement vectors are achieved. Therefore, to provided a practical use of the linear system the system (3.1) is expanded to include the multiple measurement vectors, such that the linear system now becomes a multiple measurement vector (MMV) model:

$$\mathbf{Y} = \mathbf{A}\mathbf{X} + \mathbf{E}, \quad (3.2)$$

where  $\mathbf{Y} \in \mathbb{R}^{M \times L}$  is the observed measurement matrix,  $\mathbf{X} \in \mathbb{R}^{N \times L}$  is the source matrix, and  $\mathbf{A} \in \mathbb{R}^{M \times N}$  is the mixing matrix. By expanding the system to include multiple measurement vectors some noise will occur. This is added to the linear system as a noise matrix  $\mathbf{E} \in \mathbb{R}^{M \times L}$ .  $L$  denotes the number of observed EEG measurement vectors each consisting of  $M$  measurements, that is  $L$  samples given. For  $L = 1$  the

Her skal vi være opmærksom på at dette ikke altid gælder. At der godt kan være en løsning i over-determined tilfælde (I følge Rasmus) - Laura

Nu nævner vi jo herover at vi skal finde en løsning i det uendelige løsningsrum men vi går egentligt ikke videre med det da vi i stedet introducer k og MMV. Går der for lang til fra at det er nævnt til at man kommer med et bud? Vi kommer jo først til at nævne det i næste sektion "solution method".

linear system (3.2) will just be the SMV model (3.1). As for the linear system (3.1) the MMV model (3.2) is under-determined with  $M < N$  [8, p. 42].

As the EEG measurements are non-stationary it can be hard to say something about the activations in the source matrix  $\mathbf{X}$  as the sources fluctuates between be active and non-active – depending on the situation the EEG measurements are measured in. Therefore it is of interest that the EEG measurements becomes stationary. This could possibly be done by segmentation of the EEG measurements as the measurements becomes stationary in short time period .

Let  $f$  be the sample frequency of the observed EEG measurements  $\mathbf{Y}$  and let  $s$  denoted a segment index . As such the observed EEG measurements can be divided into segments  $\mathbf{Y}_s \in \mathbb{R}^{M \times t_s f}$ , possibly overlapping, where  $t_s$  is the length of the segments in seconds. For each segment the MMV model (3.2) still holds and is rewritten into

$$\mathbf{Y}_s = \mathbf{A}\mathbf{X}_s + \mathbf{E}_s, \quad \forall s. \quad (3.3)$$

As the segmented EEG measurement matrix  $\mathbf{Y}_s$  now consists of stationary measurement, the segmented source matrix  $\mathbf{X}_s$ , consisting of the vectors  $\mathbf{x}_{s1}, \dots, \mathbf{x}_{sL}$  which have been stacked column-wise, will now consists of at most  $k$  non-zero rows – the activations of sources. The support of the segmented source matrix  $\mathbf{X}_s$  denotes the index set of non-zero rows of  $\mathbf{X}_s$ . As mentioned before,  $\mathbf{X}_s$  has at most  $k$  non-zero rows, the non-zero values occur in common location for all columns. By using this joint information it is possible to recover  $\mathbf{X}_s$  from fewer measurements [8, p. 43].

### 3.3 Solution Methods

For the EEG measurements the MMV model is an under-determined linear system with less sensors than sources,  $M < N$ , and therefore a solution must be found in the infinitely solution space, provided one solution exists – simple linear algebra can not be used.

O. Balkan [4] proposed in 2015 a method which can recover a mixing matrix  $\mathbf{A}$  from a under-determined segmented MMV model. The method is called covariance-domain dictionary learning (Cov-DL). The background of this method take a part of compressive sensing. Compressive sensing is method which can recover the unknown signals and mixing matrix from under-determined linear systems. With this method a new term is introduce as sparsity. Sparsity can be view as the number of activation within the sources  $k$  – the number of non-zeros indices in  $\mathbf{x}$ . One thing which is necessary for compressive sensing is the the number of activations  $k$  is limited by the number of sensors,  $k \leq M$ , to ensure uniquely recovery of  $\mathbf{x}$ . This is not always the case in the EEG measurements as more activations than sensors occurs when using low-density equipment to measure the EEG measurements. To overcome this problem O. Balkan transform the compressive sensing problem to covariance domain

ref eller er det almen  
kendt

$f$  er samples pr sek.,  $L$   
er antal sampels i alt,  
 $Ls$  er antal samples pr  
segment og  $t_s$  er længen  
pr segment i sekunder

to increase the dimensionality of the problem – instead the number of activations  $k$  is now limited by  $k \leq \frac{M(M+1)}{2}$  which allows us to have  $k \geq M$ .

O. Balkan [5] did also, in 2014, proposed another method which could identify the sources of a non-segmented MMV model. This method is called multiple sparse bayesian learning (M-SBL). The method take advantage of a Bayesian approach as O. Balkan constructs a variance dependent log-likelihood which as been induce by a empirical prior that ensure sparsity of the likelihood. By minimisation of the variance of the log-likelihood an estimate for the source matrix can be drawn from the log-likelihood.

## Chapter 4

# Covariance-Domain Dictionary Learning

Through this chapter the method covariance-domain dictionary learning to recover the mixing matrix  $\mathbf{A}$  from the multiple measurement vector (MMV) model (3.2) will be presented along with the necessary theory such as compressive sensing, dictionary learning and the dictionary learning method K-SVD.

Covariance-domain dictionary learning (Cov-DL) is an algorithm proposed by O. Balkan [4], claiming to successfully identify more active sources  $k$  than available measurements  $M$  by transforming the segmented multiple measurement vector model

$$\mathbf{Y}_s = \mathbf{A}_s \mathbf{X}_s + \mathbf{E}_s,$$

into the covariance domain and the bypass the limitation of compressive sensing.

$\mathbf{Y}_s \in \mathbb{R}^{M \times t_s f}$  is the observed segmented EEG measurement matrix,  $\mathbf{X}_s \in \mathbb{R}^{N \times t_s f}$  the segmented source matrix and  $\mathbf{E}_s \in \mathbb{R}^{M \times t_s f}$  is the additional segmented noise matrix.

Cov-DL takes advantage of the covariance domain where the dimensionality is increased allowing for an enlarged number of sources to be active while the dictionary remains recoverable. An important aspect of this method is the prior assumption that the sources within one segment are uncorrelated, that is the rows of  $\mathbf{X}_s$  being mutually uncorrelated. From the assumption of uncorrelated sources it can be assumed that the sample covariance of  $\mathbf{X}_s$  becomes nearly diagonal. This is of importance when the system is transformed to the covariance domain.

The Cov-DL do only recover the mixing matrix  $\mathbf{A}$  given the measurements  $\mathbf{Y}$ . Given  $\mathbf{A}$  the source matrix  $\mathbf{X}$  is to be recovered by use of the Multiple Sparse Bayesian Learning algorithm, this is described in section 5

## 4.1 Introduction to Compressive Sensing

Compressive sensing is the theory of efficient recovery of a signal from a minimal number of observed measurements. It is build upon empirical observations assuring that many signals can be approximated by remarkably sparser signals. Assume linear acquisition of the observed measurements, then the relation between the measurements and the signal to be recovered can be modelled by the multiple measurement vector (MMV) model (3.2) [10].

Through this section the introduction of the theory behind compressive sensing will be presented for one measurement vector of (3.2),  $\mathbf{y}$ , such that the theory is based on the linear system (3.1). This will be done for simplicity but the theory will still apply for the extend linear system.

In compressive sensing terminology,  $\mathbf{x} \in \mathbb{R}^N$  is the signal of interest which is sought recovered from the EEG measurement  $\mathbf{y} \in \mathbb{R}^M$  by solving the linear system (3.1). In the typical compressive sensing case the system is under-determined,  $M < N$ , and there will therefore exist infinitely many solutions, provided that one solution exist. However, by enforcing certain sparsity constraints it is possible to recover the wanted signal, hence the term sparse signal recovery [10]. The sparsity constraints are the ones presented in 3.1 where the  $\ell_0$  is introduced to count the non-zeros of the signal of interest, the source vector  $\mathbf{x}$ . The number of non-zeros (active sources)  $k$  describe how sparse the source vector is.

To find a  $k$ -sparse solution to the linear system (3.1) it can be viewed as the following optimisation problem.

$$\mathbf{x}^* = \arg \min_{\mathbf{x} \in \mathbb{C}} \|\mathbf{x}\|_0 \quad \text{subject to} \quad \mathbf{A}\mathbf{x} = \mathbf{y}.$$

Unfortunately, this optimisation problem is non-convex due to the definition of the  $\ell_0$ -norm and is therefore difficult to solve – it is an NP-hard problem. Instead, by replacing the  $\ell_0$ -norm with the  $\ell_1$ -norm, the optimisation problem can be approximated and hence becomes computationally feasible [8, p. 27]

$$\mathbf{x}^* = \arg \min_{\mathbf{x} \in \mathbb{C}} \|\mathbf{x}\|_1 \quad \text{subject to} \quad \mathbf{A}\mathbf{x} = \mathbf{y}. \quad (4.1)$$

With this optimisation problem the best  $k$ -sparse solution  $\mathbf{x}^*$  can be found. The optimisation problem is referred to as  $\ell_1$  optimisation problem or Basis Pursuit.

The following theorem justifies that the  $\ell_1$  optimisation problem finds a sparse solution [10, p. 62-63].

### Theorem 4.1.1

A mixing matrix  $\mathbf{A} \in \mathbb{R}^{M \times N}$  is defined with columns  $\mathbf{A} = [\mathbf{a}_1, \dots, \mathbf{a}_N]$ . By assuming uniqueness of a solution  $\mathbf{x}^*$  to

$$\min_{\mathbf{x} \in \mathbb{R}^N} \|\mathbf{x}\|_1 \quad \text{subject to} \quad \mathbf{A}\mathbf{x} = \mathbf{y},$$

4.1: skal vi indføre  $z$  som en approximation til  $x$ , og så et nyt omega eller? eller kan vi lade  $x^*$  være løsningen til både P0 og P1



the system  $\{\mathbf{a}_j, j \in \text{supp}(\mathbf{x}^*)\}$  is linearly independent, and in particular

$$\|\mathbf{x}^*\|_0 = \text{card}(\text{supp}(\mathbf{x}^*)) \leq M.$$

To prove this theorem one needs to realise that the set  $\{\mathbf{a}_j, j \in S\} \leq M$ , with  $S = \text{supp}(\mathbf{x}^*)$ , can not have more than  $M$  linearly independence columns. This will be done by a contradiction. So when  $M \ll N$  a sparse signal is automatically achieved.

**Proof**

Assume that the set  $\{\mathbf{a}_l, l \in S\}$  of  $l$  columns from matrix  $\mathbf{A} \in \mathbb{R}^{M \times N}$  is linearly dependent with the support  $S = \text{supp}(\mathbf{x}^*)$ . Thus a non-zero vector  $\mathbf{v} \in \mathbb{R}^N$  supported on  $S$  exists such that  $\mathbf{A}\mathbf{v} = \mathbf{0}$  – the system is linear dependent. The unique solution  $\mathbf{x}^*$  can then be written as, for any  $t \neq 0$ ,

$$\|\mathbf{x}^*\|_1 < \|\mathbf{x}^* + t\mathbf{v}\|_1 = \sum_{l \in S} |x_l^* + tv_l| = \sum_{l \in S} \text{sgn}(x_l^* + tv_l)(x_l^* + tv_l). \quad (4.2)$$

For a small  $|t|$

$$|t| < \min_{l \in S} \frac{|x_l^*|}{\|\mathbf{v}\|_\infty},$$

then the sign function become

$$\text{sgn}(x_l^* + tv_l) = \text{sgn}(x_l^*), \quad \forall l \in S.$$

By including this result in (4.2) and remembering  $t \neq 0$ :

$$\|\mathbf{x}^*\|_1 < \sum_{l \in S} \text{sgn}(x_l^*)(x_l^* + tv_l) = \sum_{l \in S} \text{sgn}(x_l^*)x_l^* + t \sum_{l \in S} \text{sgn}(x_l^*)v_l = \|\mathbf{x}^*\|_1 + t \sum_{l \in S} \text{sgn}(x_l^*)v_l.$$

From this it can be seen that it is always possible to choose  $t \neq 0$  small enough such that

$$t \sum_{l \in S} \text{sgn}(x_l^*)v_l \leq 0,$$

which contradicts that  $\mathbf{v}$  make the columns of  $\mathbf{A}$  linear dependent. Therefore, the set  $\{\mathbf{a}_l, l \in S\}$  must be linearly independent. ■

An alternative solution method to the  $\ell_1$  optimisation includes greedy algorithms such as the Orthogonal Matching Pursuit (OMP) [10, P. 65]. The OMP algorithm is an iteration process where an index set  $S$  is updated – at each iteration – by adding indices corresponding to the columns of  $\mathbf{A}$  which describe the residual best possible, hence greedy. The vector  $\mathbf{x}$  is then updated by a vector supported on  $S$  which minimise the residual, that is the orthogonal projection of  $\mathbf{y}$  onto the span $\{\mathbf{a}_l \mid l \in S\}$ .

### 4.1.1 Dictionary Learning

As clarified from the proof of theorem 4.1.1 the estimation of the mixing matrix  $\mathbf{A}$  is essential to achieve the best recovery of the sparse source vector  $\mathbf{x}$  from the EEG measurement  $\mathbf{y}$ . For the estimation one could choose a dictionary as choice for the mixing matrix  $\mathbf{A}$ .

Pre-constructed dictionaries do exist which in many cases results in simple and fast algorithms for reconstruction of  $\mathbf{x}$  [9]. Pre-constructed dictionaries are typically fitted to a specific kind of data. For instance the discrete Fourier transform or the discrete wavelet transform are used especially for sparse representation of images [9]. Hence the results of using such dictionaries depend on how well they fit the data of interest, which is creating a certain limitation. An alternative is to consider an adaptive dictionary based on a set of training data that resembles the data of interest. For this purpose learning methods are considered to empirically construct a fixed dictionary which can take part in the application. Different dictionary learning algorithms exist. One is the K-SVD which is to be elaborated in this chapter. The K-SVD algorithm was presented in 2006 by Elad et al. and found to outperform pre-constructed dictionaries when computational cost is of secondary interest [1].

For the K-SVD algorithm several measurements vectors  $\mathbf{y}$  must be used to learn the dictionary  $\mathbf{A}$ . Therefore, the K-SVD algorithm will be used on the multiple measurement vector (MMV) model as described in (3.2).

## 4.2 K-SVD

Consider  $\mathbf{Y} = [\mathbf{y}_1, \dots, \mathbf{y}_L]$ ,  $\mathbf{y}_j \in \mathbb{R}^M$  as a training database, created by  $\mathbf{y}_j = \mathbf{A}\mathbf{x}_j$  for which one wants to learn the best suitable dictionary  $\mathbf{A}$  and sparse representation  $\mathbf{X} = [\mathbf{x}_1, \dots, \mathbf{x}_L]$ ,  $\mathbf{x}_j \in \mathbb{R}^N$ . For a known sparsity constraint  $k$  this can be defined by an optimisation problem similar to the general compressive sensing problem of multiple measurements [9]

$$\min_{\mathbf{A}, \mathbf{X}} \sum_{j=1}^L \|\mathbf{y}_j - \mathbf{A}\mathbf{x}_j\|_2^2 \quad \text{subject to} \quad \|\mathbf{x}_j\|_1 \leq k, \quad 1 \leq j \leq L. \quad (4.3)$$

The learning consists of jointly solving the optimization problem on  $\mathbf{X}$  and  $\mathbf{A}$ . The uniqueness of  $\mathbf{A}$  depends on the recovery sparsity condition. As clarified earlier recovery is only possible if  $k < M$  [6].

### 4.2.1 K-SVD Algorithm

The dictionary learning algorithm K-SVD provides an updating rule which is applied to each column of  $\mathbf{A}_0 = [\mathbf{a}_0, \dots, \mathbf{a}_N]$  where  $\mathbf{A}_0$  being a random initial dictionary matrix. Updating first  $\mathbf{a}_j$  and then the corresponding coefficients in  $\mathbf{X}$  which it is

gør opmærksom på at i forhold til det tidligere defineret P1 problem har vi nu støj derfor er de byttet rund

multiplied with the  $i$ -th row in  $\mathbf{X}$  denoted by  $\mathbf{x}_{i\cdot}$ .

Let  $\mathbf{a}_{j_0}$  be the column to be updated and let the remaining columns be fixed. By rewriting the objective function in (4.3) using matrix notation it is possible to isolate the contribution from  $\mathbf{a}_{j_0}$ .

$$\begin{aligned}\|\mathbf{Y} - \mathbf{A}\mathbf{X}\|_F^2 &= \left\| \mathbf{Y} - \sum_{j=1}^N \mathbf{a}_j \mathbf{x}_{i\cdot} \right\|_F^2 \\ &= \left\| \left( \mathbf{Y} - \sum_{j \neq j_0}^N \mathbf{a}_j \mathbf{x}_{i\cdot} \right) - \mathbf{a}_{j_0} \mathbf{x}_{i_0\cdot} \right\|_F^2,\end{aligned}\quad (4.4)$$

Tjek nedenstående udledning. a og x er ikke lige lange da  $a_j$  er M lang mens  $x_{i\cdot}$  er L lang

where  $i = j$ ,  $i_0 = j_0$  and where  $F$  is the Frobenius norm that works on matrices

$$\|\mathbf{A}\|_F = \sqrt{\sum_{i=1}^M \sum_{j=1}^N |a_{ij}|^2}.$$

In (4.4) the term in the parenthesis is denoted by  $\mathbf{E}_{j_0}$ , an error matrix, and hence by minimising (4.4) with respect to  $\mathbf{a}_{j_0}$  and  $\mathbf{x}_{i_0\cdot}$  leads to the optimal contribution from  $j_0$

$$\min_{\mathbf{a}_{j_0}, \mathbf{x}_{i_0\cdot}} \|\mathbf{E}_{j_0} - \mathbf{a}_{j_0} \mathbf{x}_{i_0\cdot}\|_F^2. \quad (4.5)$$

The optimal solution to (4.5) is known to be the rank-1 approximation of  $\mathbf{E}_{j_0}$ . This comes from the Eckart–Young–Mirsky theorem [?] saying that a partial single value decomposition (SVD) makes the best low-rank approximation of a matrix such as  $\mathbf{E}_{j_0}$ . The SVD is given as

$$\mathbf{E}_{j_0} = \mathbf{U}\mathbf{\Sigma}\mathbf{V}^T \in \mathbb{R}^{M \times N},$$

with  $\mathbf{U} \in \mathbb{R}^{M \times M}$  and  $\mathbf{V} \in \mathbb{R}^{N \times N}$  being unitary matrices<sup>1</sup> and  $\mathbf{\Sigma} = \text{diag}[\sigma_1, \dots, \sigma_M] \in \mathbb{R}^{M \times N}$  a diagonal matrix.  $\sigma_j$  are the non-negative singular values of  $\mathbf{E}_{j_0}$ . The best  $k$ -rank approximation to  $\mathbf{E}_{j_0}$ , with  $k < \text{rank}(\mathbf{E}_{j_0})$  is then given by :

kilde

$$\mathbf{E}_{j_0}^{(k)} = \sum_{j=1}^k \sigma_j \mathbf{u}_j \mathbf{v}_j^T.$$

Since the outer product always have rank-1 letting  $\mathbf{a}_{j_0} = \mathbf{u}_1$  and  $\mathbf{x}_{i_0\cdot} = \sigma_1 \mathbf{v}_1^T$  solves the optimisation problem (4.5). However in order to preserve the sparsity in  $\mathbf{X}$  while optimising, only the non-zero entries in  $\mathbf{x}_{i_0\cdot}$  are allowed to vary. For this purpose only a subset of columns in  $\mathbf{E}_{j_0}$  is considered, those which correspond to the non-zero

<sup>1</sup>Unitary matrix:  $\mathbf{U}^T \mathbf{U} = \mathbf{U} \mathbf{U}^T = \mathbf{I}$

entries of  $\mathbf{x}_{i_0\cdot}$ . A matrix  $\mathbf{P}_{i_0}$  is defined to restrict  $\mathbf{x}_{i_0\cdot}$  to only contain the non-zero-rows corresponding to  $N_{j_0}$  non-zero rows:

$$\mathbf{x}_{i_0\cdot}^{(R)} = \mathbf{x}_{i_0\cdot} \mathbf{P}_{i_0}$$

where  $R$  denoted the restriction. By applying the SVD to the error matrix which has been restricted  $\mathbf{E}_{j_0}^{(R)} = \mathbf{E}_{j_0} \mathbf{P}_{i_0}$  and updating  $\mathbf{a}_{j_0}$  and  $\mathbf{x}_{i_0\cdot}^{(R)}$  the rank-1 approximation is found and the original representation vector is updated as  $\mathbf{x}_{i_0\cdot} = \mathbf{x}_{i_0\cdot}^{(R)} \mathbf{P}_{i_0}^T$ .

The main steps of K-SVD is described in algorithm 1.

---

**Algorithm 1** K-SVD

---

```

1:  $k = 0$ 
2: Initialize random  $\mathbf{A}_{(0)}$ 
3: Initialize  $\mathbf{X}_{(0)} = \mathbf{0}$ 
4:
5: procedure K-SVD( $\mathbf{A}_{(0)}$ )
6:   Normalize columns of  $\mathbf{A}_{(0)}$ 
7:   while error  $\geq$  limit do
8:      $j = j + 1$ 
9:     for  $j \leftarrow 1, 2, \dots, L$  do  $\triangleright$  updating each col. in  $\mathbf{X}_{(k)}$ 
10:       $\hat{\mathbf{x}}_j = \min_{\mathbf{x}} \|\mathbf{y}_j - \mathbf{A}_{(k-1)} \mathbf{x}_j\|$  subject to  $\|\mathbf{x}_j\| \leq k$   $\triangleright$  use Basis Pursuit
11:    end for
12:     $\mathbf{X}_{(k)} = \{\hat{\mathbf{x}}_j\}_{j=1}^L$ 
13:    for  $j_0 \leftarrow 1, 2, \dots, N$  do
14:       $\Omega_{j_0} = \{j \mid 1 \leq j \leq L, \mathbf{X}_{(k)}[j_0, j] \neq 0\}$ 
15:      From  $\Omega_{j_0}$  define  $\mathbf{P}_{i_0}$ 
16:       $\mathbf{E}_{j_0} = \mathbf{Y} - \sum_{j \in \Omega_{j_0}} \mathbf{a}_j \mathbf{x}_j$ 
17:       $\mathbf{E}_{j_0}^{(R)} = \mathbf{E}_{j_0} \mathbf{P}_{i_0}$ 
18:       $\mathbf{E}_{j_0}^{(R)} = \mathbf{U} \mathbf{\Sigma} \mathbf{V}^T$   $\triangleright$  perform SVD
19:       $\mathbf{a}_{j_0} \leftarrow \mathbf{u}_1$   $\triangleright$  update the  $j_0$  col. in  $\mathbf{A}_{(k)}$ 
20:       $(\mathbf{x}_{i_0\cdot})^{(R)} \leftarrow \sigma_1 \mathbf{v}_1$ 
21:       $\mathbf{x}_{i_0\cdot} \leftarrow (\mathbf{x}_{i_0\cdot})^{(R)} \mathbf{P}_{i_0}^T$   $\triangleright$  update the  $i_0$  row in  $\mathbf{X}_{(k)}$ 
22:    end for
23:    error =  $\|\mathbf{Y} - \mathbf{A}_{(k)} \mathbf{X}_{(k)}\|_F^2$ 
24:  end while
25: end procedure

```

---

The dictionary learning algorithm K-SVD is a generalisation of the well known K-means clustering also referred to as vector quantization. In K-means clustering a set of  $K$  vectors is learned referred to as mean vectors. Each signal sample is then

represented by its nearest mean vector. That corresponds to the case with sparsity constraint  $k = 1$  and the representation reduced to a binary scalar  $x = \{1, 0\}$ . Further instead of computing the mean of  $K$  subsets the K-SVD algorithm computes the SVD factorisation of the  $K$  different sub-matrices that correspond to the  $K$  columns of  $\mathbf{A}$ . The section is inspired by chapter 3 in [6] and the article [4]. INTRO..

### 4.3 Covariances domain representation

Consider the covariance of a vector  $\mathbf{x}_i$

$$\mathbf{\Sigma} = \mathbb{E}[(\mathbf{x}_i - \mathbb{E}[\mathbf{x}_i])(\mathbf{x}_i - \mathbb{E}[\mathbf{x}_i])^T].$$

Assume that all samples has zero mean and the same distribution within one segment. The observed measurements  $\mathbf{Y}_s \in \mathbb{R}^{M \times L}$  can be described in the covariance domain by the sample covariance  $\hat{\mathbf{\Sigma}}$  which is defined as the covariance among the  $M$  measurements across the  $L_s$  samples. That is a  $M \times M$  matrix  $\mathbf{\Sigma}_{\mathbf{Y}_s} = [\sigma_{jk}]$  with entries

$$\sigma_{jk} = \frac{1}{L} \sum_{i=1}^L y_{ji} y_{ki}^T.$$

Using matrix notation the sample covariance of  $\mathbf{Y}_s$  can be written as

$$\hat{\mathbf{\Sigma}}_{\mathbf{Y}_s} = \frac{1}{L} \mathbf{Y}_s \mathbf{Y}_s^T.$$

Similar the source matrix  $\mathbf{X}_s$  can be described in the covariance domain by the sample covariance matrix

$$\hat{\mathbf{\Sigma}}_{\mathbf{X}_s} = \frac{1}{L} \mathbf{X}_s \mathbf{X}_s^T = \mathbf{\Lambda}_s + \varepsilon$$

From the assumption of uncorrelated sources within  $\mathbf{X}_s$  the sample covariance matrix is expected to be nearly diagonal, thus it can be written as  $\mathbf{\Lambda}_s + \varepsilon$  where  $\mathbf{\Lambda}_s$  is a diagonal matrix consisting of the diagonal entries of  $\hat{\mathbf{\Sigma}}_{\mathbf{X}_s}$  and  $\varepsilon$  is the estimation error[4].

Each segment is then modelled in the covariance domain as

$$\begin{aligned} \hat{\mathbf{\Sigma}}_{\mathbf{Y}_s} &= \frac{1}{L_s} \mathbf{Y}_s \mathbf{Y}_s^T = \frac{1}{L_s} (\mathbf{A} \mathbf{X}_s + \mathbf{E}_s) (\mathbf{A} \mathbf{X}_s + \mathbf{E}_s)^T \\ \mathbf{Y}_s \mathbf{Y}_s^T &= (\mathbf{A} \mathbf{X}_s) (\mathbf{A} \mathbf{X}_s)^T + \mathbf{E}_s \mathbf{E}_s^T + \mathbf{E}_s (\mathbf{A} \mathbf{X}_s)^T + \mathbf{A} \mathbf{X}_s \mathbf{E}_s^T \\ &= \mathbf{A} \mathbf{X}_s \mathbf{X}_s^T \mathbf{A}^T + \mathbf{E}_s \mathbf{E}_s^T + \mathbf{E}_s \mathbf{X}_s^T \mathbf{A}^T + \mathbf{A} \mathbf{X}_s \mathbf{E}_s^T \\ &= \mathbf{A} (\mathbf{\Lambda}_s + \varepsilon) \mathbf{A}^T + \mathbf{E}_s \mathbf{E}_s^T + \mathbf{E}_s \mathbf{X}_s^T \mathbf{A}^T + \mathbf{A} \mathbf{X}_s \mathbf{E}_s^T \\ &= \mathbf{A} \mathbf{\Lambda}_s \mathbf{A}^T + \mathbf{A} \varepsilon \mathbf{A}^T + \mathbf{E}_s \mathbf{E}_s^T + \mathbf{E}_s \mathbf{X}_s^T \mathbf{A}^T + \mathbf{A} \mathbf{X}_s \mathbf{E}_s^T \quad (4.6) \\ &= \mathbf{A} \mathbf{\Lambda}_s \mathbf{A}^T + \tilde{\mathbf{E}} \quad (4.7) \\ & \quad (4.8) \end{aligned}$$

From (4.6) to (4.7) all terms where noise is included are defined as a united noise term  $\tilde{\mathbf{E}}$ . By vector notation (4.7) is rewritten and then vectorised. Because the covariance matrix  $\hat{\Sigma}_{\mathbf{Y}_s}$  is symmetric it is sufficient to vectorize only the lower triangular parts, including the diagonal. For this the function  $\text{vec}(\cdot)$  is defined to map a symmetric  $M \times M$  matrix into a vector of size  $\frac{M(M+1)}{2}$  making a row-wise vectorization of its upper triangular part. Furthermore, let  $\text{vec}^{-1}(\cdot)$  be the inverse function for de-vectorisation. This results in the following model

$$\begin{aligned}
 \Sigma_{\mathbf{Y}_s} &= \sum_{i=1}^N \Lambda_{s_{ii}} \mathbf{a}_i \mathbf{a}_i^T + \tilde{\mathbf{E}} \\
 \text{vec}(\Sigma_{\mathbf{Y}_s}) &= \sum_{i=1}^N \Lambda_{s_{ii}} \text{vec}(\mathbf{a}_i \mathbf{a}_i^T) + \text{vec}(\tilde{\mathbf{E}}) \\
 &= \sum_{i=1}^N \mathbf{d}_i \Lambda_{s_{ii}} + \text{vec}(\tilde{\mathbf{E}}) \\
 &= \mathbf{D} \boldsymbol{\delta}_s + \text{vec}(\tilde{\mathbf{E}}), \quad \forall s.
 \end{aligned} \tag{4.9}$$

Here  $\boldsymbol{\delta}_s \in \mathbb{R}^N$  contains the diagonal entries of the source sample-covariance matrix  $\Lambda_s$  and the matrix  $\mathbf{D} \in \mathbb{R}^{M(M+1)/2 \times N}$  consists of the columns  $\mathbf{d}_i = \text{vec}(\mathbf{a}_i \mathbf{a}_i^T)$ . Note that  $\mathbf{D}$  and  $\boldsymbol{\delta}_s$  are unknown while  $\text{vec}(\Sigma_{\mathbf{Y}_s})$  is known from the observed data. By this transformation to the covariance domain one segment is now represented as the single measurement model with  $M(M+1)/2$  "measurements". It has been shown that this model allow for identification of  $k \leq M(M+1)/2$  active sources [16], which is a much weaker sparsity constraint than the original sparsity constraint  $k \leq M$ . The purpose of the Cov-DL algorithm is to leverage this model to find the dictionary  $\mathbf{A}$  from  $\mathbf{D}$  and then still allow for  $k \leq M(M+1)/2$  active sources to be identified. That is the number of active sources are allowed to exceed the number of observations as intended.

## 4.4 Determination of the Dictionary

The goal is now to learn first  $\mathbf{D}$  and then the associated mixing matrix  $\mathbf{A}$ . Two methods are considered relying on the relation of  $M$  and  $N$ .

### Under-determined $\mathbf{D}$

In the case of  $N > \frac{M(M+1)}{2}$   $\mathbf{D}$  becomes under-determined. This is similar to the original system being under-determined when  $N > M$ . Thus, it is again possible to solve the under-determined system if certain sparsity is withhold. Namely  $\boldsymbol{\delta}_s$  being  $\frac{M(M+1)}{2}$ -sparse. Assuming the sufficient sparsity on  $\boldsymbol{\delta}_s$  is withhold it is possible to learn the dictionary matrix of the covariance domain  $\mathbf{D}$  by traditional dictionary learning methods applied to the observations represented in the covariance domain

er yderligere argumentation nødvendig her?

$\text{vec}(\mathbf{\Sigma}_{\mathbf{Y}_s})$  for all  $s$ . For this K-SVD algorithm, described in section 4.2 is used. Note here that the number of samples that are used to learn the dictionary is remarkable reduces as one segment effectively corresponds to one sample in the covariance domain.

When  $\mathbf{D}$  is learned it is possible to find the original mixing matrix  $\mathbf{A}$  that generated  $\mathbf{D}$  through the relation  $\mathbf{d}_j = \text{vec}(\mathbf{a}_j \mathbf{a}_j^T)$ . Here each column is found by the optimisation problem

$$\min_{\mathbf{a}_j} \|\text{vec}^{-1}(\mathbf{d}_j) - \mathbf{a}_j \mathbf{a}_j^T\|_2^2,$$

for which the global minimizer is  $\mathbf{a}_j^* = \sqrt{\lambda_j} \mathbf{b}_j$ . Here  $\lambda_j$  is the largest eigenvalue of  $\text{vec}^{-1}(\mathbf{d}_j)$ ,

$$\text{vec}^{-1}(\mathbf{d}_j) = \begin{bmatrix} d_{11} & d_{12} & \cdots & d_{1N} \\ d_{21} & d_{22} & \cdots & d_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ d_{N1} & d_{N2} & \cdots & d_{NN} \end{bmatrix}, \quad i \in [N]$$

and  $\mathbf{b}_j$  is the corresponding eigenvector.

redegørelse for resultatet her skal laves

### Over-determined $\mathbf{D}$

In the case of  $N < \frac{M(M+1)}{2}$  an over-determined system is achieved and it is not possible to find  $\mathbf{D}$  by dictionary learning methods.

By assuming that  $\frac{M(M+1)}{2}$  will be close to  $N$ , because  $N > M$  is given, then the measurements in the covariance domain  $\text{vec}(\mathbf{\Sigma}_{\mathbf{Y}_s})$  will live on or near a subspace of dimension  $N$ . This subspace is spanned by the columns of  $\mathbf{D}$ , and is denoted as  $\mathcal{R}(\mathbf{D})$ . To learn  $\mathcal{R}(\mathbf{D})$  without having to impose any sparsity constraint on  $\delta_s$  it is possible to use Principal Component Analysis(PCA). By use of PCA a set of basis vectors  $\mathbf{U}$  is achieved such that  $\mathcal{R}(\mathbf{U}) = \mathcal{R}(\mathbf{D})$ . This however do not imply that  $\mathbf{D} = \mathbf{U}$ .

how else can they live on the same space??

evt. teoretisk beskrivelse af PCA i appendix?

In the case of two sets of basis vectors span the same space, namely  $\mathcal{R}(\mathbf{U}) = \mathcal{R}(\mathbf{D})$ , the projection operator of the given subset must be unique. Which is true if and only if  $\mathbf{D}(\mathbf{D}^T \mathbf{D})^{-1} \mathbf{D}^T = \mathbf{U}(\mathbf{U}^T \mathbf{U})^{-1} \mathbf{U}^T$ . Remember from the above derivation the condition that  $\mathbf{d}_i = \text{vec}(\mathbf{a}_i \mathbf{a}_i^T)$ . From this it is possible to obtain  $\mathbf{A}$  through the optimisation problem

kilde foruden phd p. 51?

$$\begin{aligned} \min_{\mathbf{a}_i} \|\mathbf{D}(\mathbf{D}^T \mathbf{D})^{-1} \mathbf{D}^T - \mathbf{U}(\mathbf{U}^T \mathbf{U})^{-1} \mathbf{U}^T\|_F^2 \\ \text{s.t. } \mathbf{d}_i = \text{vec}(\mathbf{a}_i \mathbf{a}_i^T) \end{aligned} \quad (4.10)$$

where  $\mathbf{U}$  is learned by use of PCA performed on  $\text{vec}(\mathbf{\Sigma}_{\mathbf{Y}_s})$ .

To solve this optimization problem the cost function is minimized by use of quasi-Newton optimization methods. Several specific quasi-Newton methods exist but the

basic principal will be presented here. The Newton optimization method is a multidimensional gradient method. The method is based on a quadratic approximation of the optimization problem by use of the Taylor series, which is elaborated in [3, p. 29]. Let  $f(\mathbf{x})$  be the cost function and  $\boldsymbol{\delta}$  be the change in  $\mathbf{X}$ . By differentiating the Taylor approximation of  $f(\mathbf{x} + \boldsymbol{\delta})$  and setting it equal to zero, the optimal change in  $\mathbf{x}$  is found to be  $\boldsymbol{\delta} = -\mathbf{H}^{-1}\mathbf{g}$ . Where  $\mathbf{g}$  is the gradient and  $\mathbf{H}$  is the Hessian. The quasi-Newton methods deviate from the basic Newton method by letting the direction search be based on a positive semi-definite matrix  $\mathbf{S}$  which is generated from available data in order to approximate  $\mathbf{H}^{-1}$ . Details of the method is found in [3, p. 175]



#### 4.4.1 Pseudo Code of the Cov-DL Algorithm

---

**Algorithm 2** Cov-DL
 

---

```

1: procedure COV-DL( $\mathbf{Y}_s$ )
2:   for  $s \leftarrow 1, \dots, n\_seg$  do
3:     compute sample covariance matrix  $\widehat{\Sigma}_{\mathbf{Y}_s}$ 
4:      $\mathbf{y}_{cov_s} = \text{vec}(\widehat{\Sigma}_{\mathbf{Y}_s})$ 
5:   end for
6:    $\mathbf{Y}_{cov} = \{\mathbf{y}_{cov_s}\}_{s=1}^{n\_seg}$ 
7:   if  $N > \frac{M(M+1)}{2}$  then
8:     procedure K-SVD( $\mathbf{Y}_{cov}$ )
9:       returns  $\mathbf{D} \in \mathbb{R}^{M(M+1)/2 \times N}$ 
10:    end procedure
11:    for  $j \leftarrow 1, \dots, N$  do
12:       $\mathbf{T} = \text{vec}^{-1}(d_j)$ 
13:       $\lambda_j \leftarrow \max\{\text{eigenvalue}(\mathbf{T})\}$ 
14:       $\mathbf{b}_j \leftarrow \text{eigenvector}(\lambda_j)$ 
15:       $\mathbf{a}_j \leftarrow \sqrt{\lambda_j} \mathbf{b}_j$ 
16:    end for
17:     $\mathbf{A} = \{\mathbf{a}_j\}_{j=1}^N$ 
18:  end if
19:
20:  if  $N < \frac{M(M+1)}{2}$  then
21:    procedure PCA( $\text{vec}(\Sigma_{\mathbf{Y}_s})$ )
22:      returns  $\mathbf{U} \in \mathbb{R}^{M(M+1)/2 \times N}$ 
23:    end procedure
24:    procedure QUASI-NEWTON(problem (4.10))
25:      returns  $\mathbf{A} = \{\mathbf{a}_j\}_{j=1}^N$ 
26:    end procedure
27:  end if
28: end procedure

```

---

#### 4.4.2 Remarks to be considered and investigated in the code

- the number of samples for dictionary learning is reduced remarkably when Cov-DL is used, this is mentioned above. increasing overlap will improve this.
- the effect of the length of segments  $L_s$ . it will effect the diagonality of  $cov(X)$
- the values of the individual sources must not be constant over time, that is the power of the sources, it will not be a problem for EEG data.

- for Cov-DL2 the solution tends to be unique when  $M < N < M(M+1)/2$ , that is the found cost function may have a local minima. for this it is recommended to use several random initial point.

## Chapter 5

# Multiple Sparse Bayesian Learning

....  
INTRODUCTION  
....

Consider the multiple measurement vector model (MMV)

$$\mathbf{Y} = \mathbf{A}\mathbf{X} + \mathbf{E},$$

where the unknown matrices  $\mathbf{A} \in \mathbb{R}^{M \times N}$  and  $\mathbf{X} \in \mathbb{R}^{N \times L}$  are wished recovered from the known measurement matrix  $\mathbf{Y} \in \mathbb{R}^{M \times L}$  in the case of  $M < N$ . In chapter 4 the mixing matrix  $\mathbf{A}$  was found. In this chapter, a method to recover the source matrix  $\mathbf{X}$  is sought. The approach is to find the support set  $S$  of  $\mathbf{X}$  providing the non-zeros rows of  $\mathbf{X}$  which corresponds to localisation of the active sources. By ... the active sources are identified in order to fully recover  $\mathbf{X}$ .

The chapter is inspired by [22] and the articles [24], [5].

### 5.1 Maximum a Posterior Estimation

With the knowledge of  $\mathbf{A}$  and  $\mathbf{Y}$  it is possible to find maximum likelihood estimate of  $\mathbf{X}$ . By maximising the likelihood  $p(\mathbf{Y}|\mathbf{X})$  an estimate of  $\mathbf{X}$  can be achieved in the case of more sensors than sources,  $M > N$ . But in the desired case where  $M < N$  the estimation becomes complicate as the MMV model becomes under-determined and potentially an infinitely number of solutions exist with equal likelihoods.

As the optimisation problem of the MMV model is NP-hard another estimation method must be used.

Bayesian probability ...

indsæt her hvad tanken  
bag Bayesian egentlig er

Within this Bayesian framework the source matrix  $\mathbf{X}$  is seen as a variable which is drawn from some distribution  $p(\mathbf{X})$  such that it is possible to narrow down the infinitely solution space. Assuming a prior belief that  $\mathbf{Y}$  is generated from a sparse coefficient matrix, that is the distribution from where  $\mathbf{X}$  is drawn has a sharp, possibly infinite, spike at zero surrounded by fat tails.

By applying an exponential function  $\exp(-(\cdot))$  transformation onto our optimisation problem a Gaussian likelihood function  $p(\mathbf{Y}|\mathbf{X})$  with a  $\lambda$ -dependent variance is achieved:

$$p(\mathbf{Y}|\mathbf{X}) \propto \exp\left(-\frac{1}{\gamma}\|\mathbf{Y} - \mathbf{A}\mathbf{X}\|_2^2\right),$$

with a prior distribution  $p(\mathbf{X}) \propto \exp(-\|\mathbf{X}\|_0)$  [22, p. 137]. The optimisation problem is then rewritten by Bayers formula

$$\begin{aligned}\hat{\mathbf{X}} &= \arg \max_{\mathbf{X}} p(\mathbf{Y}|\mathbf{X})p(\mathbf{X}) \\ &= \arg \max_{\mathbf{X}} \frac{p(\mathbf{Y}|\mathbf{X})p(\mathbf{X})}{p(\mathbf{Y})} \quad (\text{Bayers Formular}) \\ &= \arg \max_{\mathbf{X}} p(\mathbf{X}|\mathbf{Y}).\end{aligned}$$

From this it is possible to view the optimization problem as a MAP estimation challenge.

## 5.2 Empirical Bayesian Estimation

Different MAP estimation approaches exist separated by the choice of sparsity inducing prior and optimization method. Some problems have shown to occur when using a fixed and algorithm-dependent prior as the posterior is not sparse enough if a prior is not as sparse leading to a non-recovery. Another issue is that a combinatorial number of suboptimal local solutions can occur.

By use of automatic relevance determination (ARD) the problems related to the sparse prior can be avoided. ARD is a method where a prior is introduced to determine the relevance of a parameter. This prior is modulated by a vector of hyperparameters affecting the prior variance of each row in  $\mathbf{X}$ . This can also be viewed as a regularisation of the solution space which is narrowed to consist only of relevant information [23].

An empirical prior can be used with ARD as the empirical prior is flexible and depends on the unknown hyperparameter  $\gamma$  and therefore more data-dependent – such prior can be controlled to induce sparsity.

Let the likelihood  $p(\mathbf{Y}|\mathbf{X})$  be Gaussian, with known noise variance  $\sigma^2$ . Then for each column in  $\mathbf{Y}$  and  $\mathbf{X}$  the likelihood is written as

$$\begin{aligned} p(\mathbf{y}_j|\mathbf{x}_j) &= \mathcal{N}(\mathbf{A}\mathbf{x}_j, \sigma^2\mathbf{I}) \\ &= (2\pi\sigma^2)^{-N/2} \exp\left(-\frac{1}{2\sigma^2}\|\mathbf{y}_j - \mathbf{A}\mathbf{x}_j\|_2^2\right). \end{aligned}$$

With the use of ARD the  $i$ -th row of the sources matrix  $\mathbf{X}$ ,  $\mathbf{x}_i$ , is assigned an  $L$ -dimensional independent Gaussian prior with zero mean and a variance controlled by  $\gamma_i$  which is unknown:

$$p(\mathbf{x}_i; \gamma_i) = \mathcal{N}(0, \gamma_i\mathbf{I}).$$

By combining the row priors

$$p(\mathbf{X}; \gamma) = \prod_{j=1}^L p(\mathbf{x}_i; \gamma_i),$$

a full prior of  $\mathbf{X}$  is achieved modulated by the hyperparameter vector  $\gamma = [\gamma_1, \dots, \gamma_M]^T$ . By combining the full prior and the likelihood  $p(\mathbf{y}_j|\mathbf{x}_j)$  the posterior of the  $j$ -th column of the source matrix  $\mathbf{X}$  becomes

$$p(\mathbf{x}_j|\mathbf{y}_j; \gamma) = \frac{p(\mathbf{x}_j, \mathbf{y}_j; \gamma)}{\int p(\mathbf{x}_j, \mathbf{y}_j; \gamma) d\mathbf{x}_j} = \mathcal{N}(\boldsymbol{\mu}_j, \boldsymbol{\Sigma}),$$

with mean and covariance given as

$$\begin{aligned} \boldsymbol{\Sigma} &= \text{Cov}(\mathbf{x}_j|\mathbf{y}_j; \gamma) = \boldsymbol{\Gamma} - \boldsymbol{\Gamma}\mathbf{A}^T\boldsymbol{\Sigma}_y^{-1}\mathbf{A}\boldsymbol{\Gamma}, \quad \forall j = 1, \dots, L \\ \mathcal{M} &= [\boldsymbol{\mu}_1, \dots, \boldsymbol{\mu}_L] = \mathbb{E}[\mathbf{X}|\mathbf{Y}; \gamma] = \boldsymbol{\Gamma}\mathbf{A}^T\boldsymbol{\Sigma}_y^{-1}\mathbf{Y}, \end{aligned} \tag{5.1}$$

where  $\boldsymbol{\Gamma} = \text{diag}(\gamma)$  and  $\boldsymbol{\Sigma}_y = \sigma^2\mathbf{I} + \mathbf{A}\boldsymbol{\Gamma}\mathbf{A}^T$ . Let now the posterior mean serve as the point estimate for  $\mathbf{X}$  without involving the support set  $S$ .

It is clear that row sparsity is achieved whenever  $\gamma_i = 0$ . From this the posterior must satisfy the following

$$P(\mathbf{x}_{i\cdot} = \mathbf{0}|\mathbf{Y}; \gamma_i = 0) = 1,$$

which ensure that the posterior mean  $\mathcal{M}$  of the  $i$ -th row,  $\boldsymbol{\mu}_{i\cdot}$ , will be zero. Now, instead of estimating the support set/sparsity profile of our source matrix  $\mathbf{X}$  it is sufficient to estimate the hyperparameters  $\gamma_i$  [22, p. 147].

Each different hyperparameter  $\gamma$  correspond to different hypothesis for the prior distribution of the underlying generation of  $\mathbf{Y}$ . Therefore the determination of  $\gamma$  is seen as a model selection for which an empirical Bayesian strategy can be used. By the empirical Bayesian strategy the unknown weights, making the source matrix  $\mathbf{X}$ , are treated as nuisance parameters and are integrated out.

By integrating the likelihood of  $\mathbf{Y}$  with respect to the unknown sources  $\mathbf{X}$  the marginal likelihood of the observed mixed data  $\mathbf{Y}$ ,  $p(\mathbf{Y}; \gamma)$  is achieved [22, p. 146]. By applying the  $-2\log(\cdot)$  transformation the marginal likelihood function is transformed to a cost function

$$\begin{aligned} (\gamma) &= -2\log\left(\int p(\mathbf{Y}|\mathbf{X})p(\mathbf{X}; \gamma) d\mathbf{X}\right) \\ &= -2\log(p(\mathbf{Y}; \gamma)) \\ &= \log(|\Sigma_y|) + \frac{1}{L} \sum_{j=1}^L \mathbf{y}_j^T \Sigma_y^{-1} \mathbf{y}_j \end{aligned}$$

To minimise the marginal log likelihood  $(\gamma)$  with respect to  $\gamma$  the evidence maximisation (EM) algorithm can be used. The E-step of the EM algorithm is to compute the posterior moments using (5.1) while the M-step is the following update rule of  $\gamma_i$ :

$$\gamma_i^{(k+1)} = \frac{1}{L} \|\boldsymbol{\mu}_{i\cdot}\|_2^2 + \Sigma_{ii}, \quad \forall i = 1, \dots, M.$$

The M-step is very slow on large data. Instead one could use a fixed point update to fasten the convergence on large data, however convergence is no longer ensured. The fixed point updating step is achieved by taking the derivative of the marginal log likelihood  $(\gamma)$  with respect to  $\gamma$  and equating it with zero. This lead to the following update equation which can replace the above M-step in the EM-algorithm:

$$\gamma_i^{(k+1)} = \frac{\frac{1}{L} \|\boldsymbol{\mu}_{i\cdot}\|_2^2}{1 - \gamma_i^{-1(k)} \Sigma_{ii}}, \quad \forall i = 1, \dots, M.$$

Empirically this alternative update rule have shown use full in highly under-determined large scale cases by driving many hyper parameters toward zero allowing for the corresponding weight in the source matrix to be discarded. For simultaneous sparse approximation problems this is the process referred to as multiple sparse Bayesian learning, M-SBL.

From the resulting  $\gamma^*$  the support set  $S$  of the source matrix  $\mathbf{X}$  can be extracted,

$$S = \{i | \hat{\gamma}_i \neq 0\},$$

concluding the localisation of active sources within  $\mathbf{X}$ . In practise some arbitrary small threshold can be used such that that any sufficiently small hyperparameter is discarded.

For identification of the active sources the estimate of the source matrix  $\mathbf{X}$  is given as  $\mathbf{X}^* = \mathcal{M}^* \approx \mathbf{X}$ , with  $\mathcal{M}^* = \mathbb{E}[\mathbf{X}|\mathbf{Y}; \gamma^*]$ . This leads to the following estimate

$$\mathbf{X}^* = \begin{cases} \mathbf{x}_{i\cdot} = \boldsymbol{\mu}_{i\cdot}^*, & i \in S \\ \mathbf{x}_{i\cdot} = \mathbf{0}, & i \notin S \end{cases}$$

---

**Algorithm 3** M-SBL
 

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1. Given  $\mathbf{Y}$  and a dictionary matrix  $\mathbf{A}$ .
  2. Initialise  $\gamma$ , e.g.  $\gamma = \mathbf{1}$ .
  3. Compute the posterior moments  $\Sigma$  and  $\mathcal{M}$ .
  4. Update  $\gamma$  using EM or fixed-point.
  5. Repeat step 3 and 4 until convergence to a fixed point  $\gamma^*$ .
  6. Update the posterior moments  $\Sigma^*$  and  $\mathcal{M}^*$  with  $\gamma^*$ .
  7. Extract support set  $S$
  8. Set source matrix estimate  $\hat{\mathbf{X}}^* = \mathcal{M}_S^*$
-





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# Appendix A

## Extended ICA Algorithms

This appendix provide an extension to the basic algorithm for ICA regarding the measure of non-Gaussianity and the computation method. This extended algorithm is referred to as fast ICA and is more commonly used for source separation. This is the algorithm used to apply ICA on EEG measurements for comparison within the thesis.

### A.1 Fixed-Point Algorithm - FastICA

An advantage of gradient algorithms is the possibility of fast adoption in non-stationary environments due the use of all input,  $\mathbf{y}$ , at once. A disadvantage of the gradient algorithm is the resulting slow convergence, depending on the choice of  $\gamma$  for which a bad choice in practise can disable convergence. A fixed-point iteration algorithm to maximise the non-Gaussianity is an alternative that could be used.

Consider the gradient step derived in section ?? . In the fixed point iteration the sequence of  $\gamma$  is omitted and replaced by a constant. This builds upon the fact that for a stable point of the gradient algorithm the gradient must point in the direction of  $\mathbf{b}_j$ , hence be equal to  $\mathbf{b}_j$ . In this case adding the gradient to  $\mathbf{b}_j$  does not change the direction and convergence is achieved.

Letting the gradient given in (??) be equal to  $\mathbf{w}$  and considering the same simplifications again suggests the new update step as [13, p. 179]

$$\mathbf{b}_j \leftarrow \mathbb{E}[\mathbf{y}(\mathbf{b}_j^T \mathbf{y})^3] - 3\mathbf{b}_j.$$

After the fixed point iteration  $\mathbf{b}_j$  is again divided by its norm to withhold the constraint  $\|\mathbf{b}_j\| = 1$ . Instead of  $\gamma$  the fixed-point algorithm compute  $\mathbf{b}_j$  directly from previous  $\mathbf{b}_j$ .

The fixed-point algorithm is referred to as FastICA. The algorithm has shown to converge fast and reliably, then the current and previous  $\mathbf{w}$  laid in the same direction [13, p. 179].

wiki: The fixed point is stable if the absolute value of the derivative of  $\mathbf{w}$  at the point is strictly less than 1?

### A.1.1 Negentropy

An alternative measure of non-Gaussianity is the negentropy, which is based on the differential entropy. The differential entropy  $H$  of a random vector  $\mathbf{y}$  with density  $p_y(\boldsymbol{\eta})$  is defined as

$$H(\mathbf{y}) = - \int p_y(\boldsymbol{\eta}) \log(p_y(\boldsymbol{\eta})) d\boldsymbol{\eta}.$$

The entropy describes the information that a random variable gives. The more unpredictable and unstructured a random variable is higher is the entropy, e.g. Gaussian random variables have a high entropy, in fact the highest entropy among the random variables of the same variance [13, p. 182].

Negentropy is a normalised version of the differential entropy such that the measure of non-Gaussianity is zero when the random variable is Gaussian and non-negative otherwise. The negentropy  $J$  of a random vector  $\mathbf{y}$  is defined as

$$J(\mathbf{y}) = H(\mathbf{y}_{\text{gaus}}) - H(\mathbf{y}),$$

with  $\mathbf{y}_{\text{gaus}}$  being a Gaussian random variable of the same covariance and correlation as  $\mathbf{y}$  [13, p. 182].

As the kurtosis is sensitive for outliers the negentropy is instead difficult to compute computationally as the negentropy require a estimate of the pdf. As such an approximation of the negentropy is needed.

To approximate the negentropy it is common to use the higher order cumulants including the kurtosis. The following approximation is stated without further elaboration, the derivation can be found in [13, p. 182].

### A.1.2 Fixed-Point Algorithm with Negentropy

Maximization of negentropy by use of the fixed-point algorithm is now presented, for derivation of the fixed point iteration see [13, p. 188]. Algorithm 4 show Fast ICA using negentropy, this is the algorithm which is implemented for comparison with the source separation methods which are tested in this thesis.

---

**Algorithm 4** Fast ICA – with negentropy

---

```

1: procedure PRE-PROCESSING( $\mathbf{y}$ )
2:   Center measurements  $\mathbf{y} \leftarrow \mathbf{y} - \bar{\mathbf{y}}$ 
3:   Whitening  $\mathbf{y} \leftarrow \mathbf{y}_{white}$ 
4: end procedure
5:
6: procedure FASTICA( $\mathbf{y}$ )
7:    $k = 0$ 
8:   Initialise random vector  $\mathbf{b}_{j(k)}$   $\triangleright$  unit norm
9:   for  $j \leftarrow 1, 2, \dots, N$  do
10:    while convergence critia not meet do
11:       $k = k + 1$ 
12:       $\mathbf{b}_{j(k)} \leftarrow \mathbb{E}[\mathbf{y}(\mathbf{b}_j^T \mathbf{y})] - \mathbb{E}[g'(\mathbf{b}_j^T \mathbf{y})]\mathbf{b}_j$   $\triangleright g$  defined in [13, p. 190]
13:       $\mathbf{b}_{j(k)} \leftarrow \mathbf{b}_j / \|\mathbf{b}_j\|$ 
14:    end while
15:     $x_j = \mathbf{b}_j^T \mathbf{y}$ 
16:  end for
17: end procedure

```

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