

PRINCIPLES OF COMMUNICATIONS

ASSIGNMENT 04

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D10A 01

PCOM assignment 4

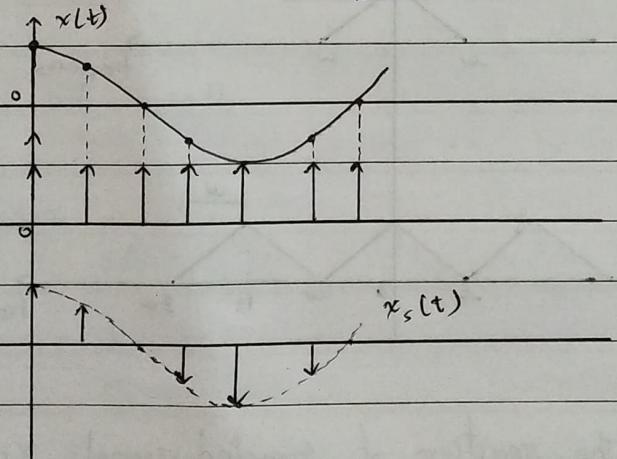
i) State and prove sampling theorem for low pass band limited signal

→ Sampling theorem states that: "If finite energy signal $x(t)$ contains no frequencies higher than ' w ' Hz (i.e. it is a band of limiting signal) then, it is completely determined by specifying its values at instant of time, which are spaced ($1/2 w$) seconds apart." If a finite energy signal $x(t)$ contains no frequency component higher than ' w ' Hz then it may be completely recovered from its samples which are spaced ($1/2 w$) seconds apart.

Proof:

Step 1:- Represent the sampling function $s(t)$ mathematically the spacing between the adjacent unit impulses is T_s seconds. Therefore the frequency of sampling function is equal to the sampling frequency f_s . The sampled signal is denoted by $x_s(t)$

$$\therefore \text{Sample function } s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$$



Step 2:- Represent the sampled signal $x_s(t)$ mathematically $x_s(t)$ is obtained by multiplying $x(t)$ and $s(t)$

$$\therefore x_s(t) = x(t) \times s(t) = x(nT_s) \times s(t)$$

$$\therefore x_s(t) = \sum_{n=-\infty}^{\infty} x(nT_s) \cdot \delta(t - nT_s)$$

Step 3: Obtain fourier transform of sampled signal.

$$\therefore S(f) = f_s \sum_{n=-\infty}^{\infty} \delta(f - n f_s)$$

$$\text{and } x(t) = f_0 \sum_{n=-\infty}^{\infty} \delta(t - n f_0)$$

$$\therefore X_s(f) = f_s \sum_{n=-\infty}^{\infty} x(t) * \delta(f - n f_s)$$

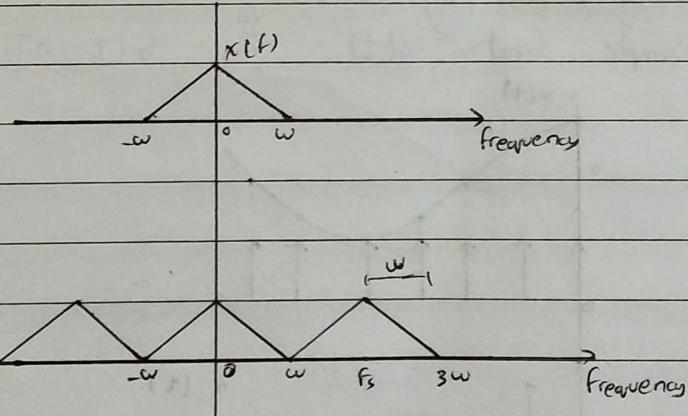
$$\therefore \text{conv}(x(t)), \delta(f - n f_s) = x(f - n f_s)$$

$$\therefore X_s(f) = f_0 \sum_{n=-\infty}^{\infty} x(f - n f_s)$$

Step 4:- Sampled signal completely represents $x(t)$

$x(f - n f_s)$ represents the shifted version of $x(f)$ of original signal $x(t)$

$$\therefore X(f - n f_s) = X(f) \text{ at } f=0, \pm f_s, \pm 2f_s$$



As seen above the spectrum of sampled signal $x_s(t)$ has frequency $f_s = 2\omega$. Thus sampling theorem is proved.

2) What do you mean by aliasing? How can it be avoided?

→ Aliasing is a phenomenon of a high frequency in the spectrum of original signal $x(t)$, taking on the identity of lower frequency in the spectrum of the sampled signal $x_s(t)$. Aliasing can be avoided by doing the following:

- i) Use a band limiting low pass filter and pass the signal $x(t)$ through it before sampling, this filter is also called anti aliasing or prealias filter.
- ii) Increase the sampling frequency f_s greater than $2W$.

3) The signal $v(t) = \cos 5\pi t + 0.5 \cos 10\pi t$ is instantaneously sampled. Calculate the Nyquist interval.

$$f_1 = \frac{5\pi}{2\pi} = 2.5 \text{ Hz}$$

$$f_2 = \frac{10\pi}{2\pi} = 5 \text{ Hz}$$

$$\therefore \text{Nyquist rate} = 2f_2 = 10 \text{ Hz}$$

$$\therefore \text{Nyquist interval} = \frac{1}{2f_2}$$

$$= \frac{1}{10}$$

$$= \underline{\underline{0.1 \text{ seconds}}}$$

4) A Bandpass signal has a spectral range that extends from 30 kHz - 75 kHz. Calculate sampling frequency f_s .

$$\Rightarrow \text{Bandwidth} = f_2 - f_1 = 75 - 30 = 45 \text{ kHz}$$

$$\text{Sampling frequency } f_s = \frac{2f_{\max}}{B} = \frac{2 \times 75}{45}$$

$$\therefore \underline{f_s = 3.333 \text{ kHz}}$$

5) Differentiate between analog and digital pulse modulation techniques.

Analog pulse modulation

1) Amplitude, width or position of pulsed carrier is varied with message signal

1) Message is sampled, quantized and encoded into codes

2) Output is a train of analog pulses

2) Output is train of digital bits that forms the codeword

3) It is simple and less expensive

3) It is complex and expensive

4) Repeaters can not be used

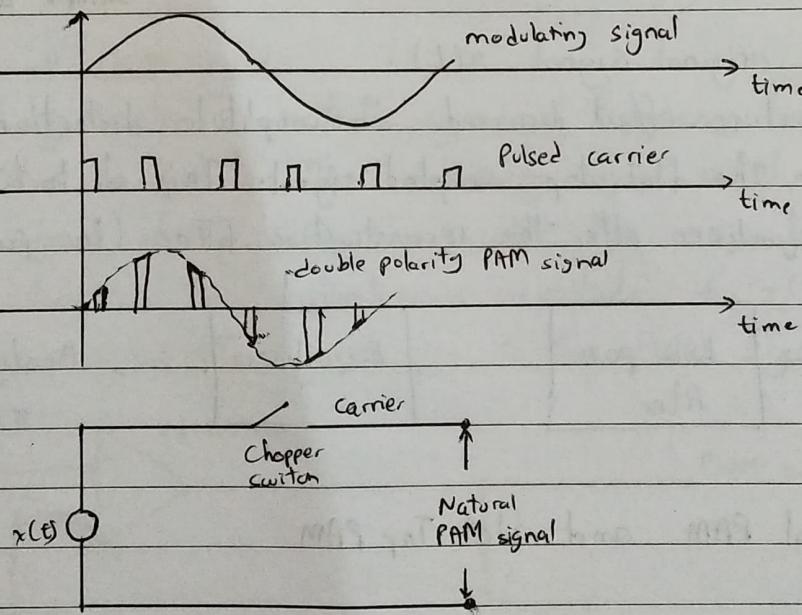
4) Repeaters can be used

6) Explain with neat diagram and waveform generation and detection of natural and flat top PAM

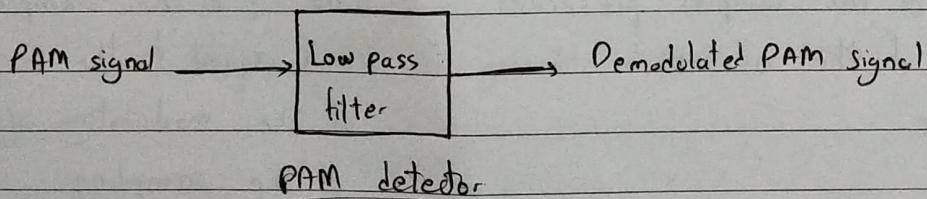
→ In PAM system, the amplitude of pulsed carrier is changed in proportion with the instantaneous amplitude of modulating signal $x(t)$

i) Natural PAM

The continuous modulating signal $x(t)$ is passed through a low pass filter. It will band limit this signal to f_m . The pulse generator generates a pulse train at a frequency f_s such that $f_s \geq 2f_m$. Sampling takes place in the multiplier, to generate the PAM signal.



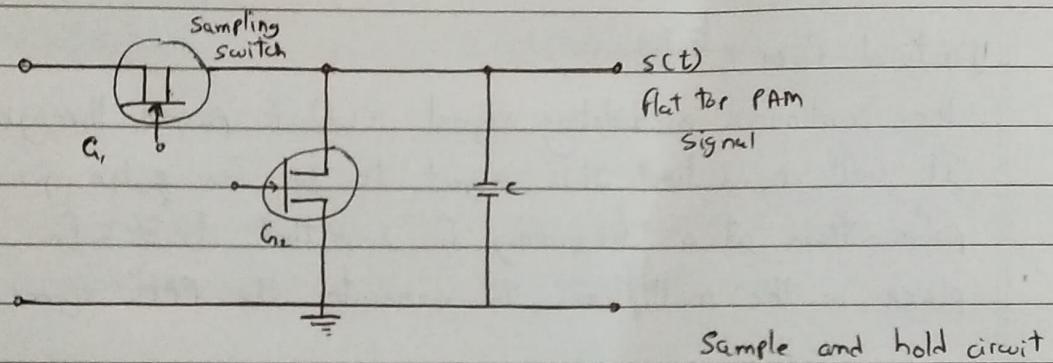
This PAM signal can be detected by passing it through a low pass filter.



Flat top PAM :-

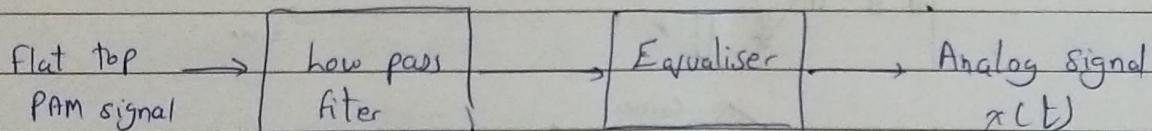
The sample and hold circuit consists of two FET switches and a capacitor. The analog signal $x(t)$ is applied at the input of this circuit and the flat topped PAM signal $s(t)$ is obtained across the capacitor.

The flat top PAM can be obtained by using the sample and hold circuit as shown below



Reconstruction of original signal $x(t)$:

Due to the aperture effect discussed, an amplitude distortion as well as a delay is introduced in the flat top sampled signal. This distortion can be corrected an equalizer after the reconstruction filter (low pass filter).



7) Compare natural PAM and flat Top PAM

Natural PAM

FLAT TOP PAM

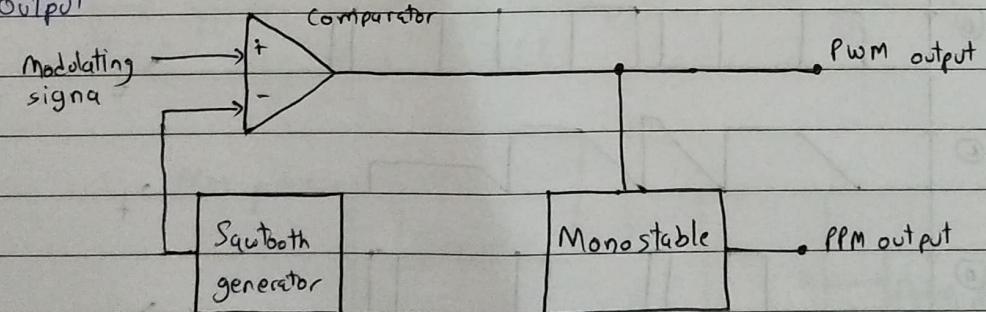
i) The top of the samples preserves the shape of the waveform amplitude.

ii) The top of the sample remains constant and equal to the instantaneous value of the modulating signal, at the start of the sampling.

- 2) SNR is worse than that of flat top FDM signal 2) Better SNR due to increased signal power
 3) Reconstruction of original signal requires only a low pass filter. 3) Reconstruction of original signal requires lowpass filter and equaliser.
 4) Very little distortion 4) Aperture effect introduces distortion
- 8) Draw the block diagram of PWM signal generator and detector. Explain the working, giving waveform at the output of each block.

Generation of PWM signal

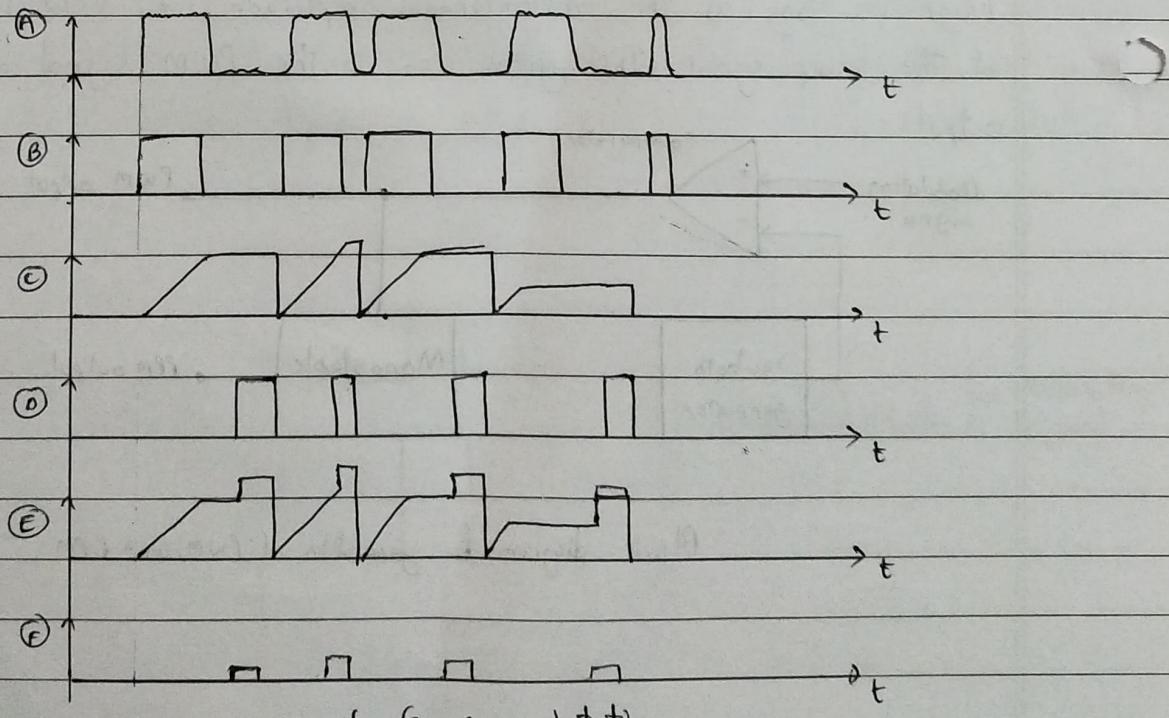
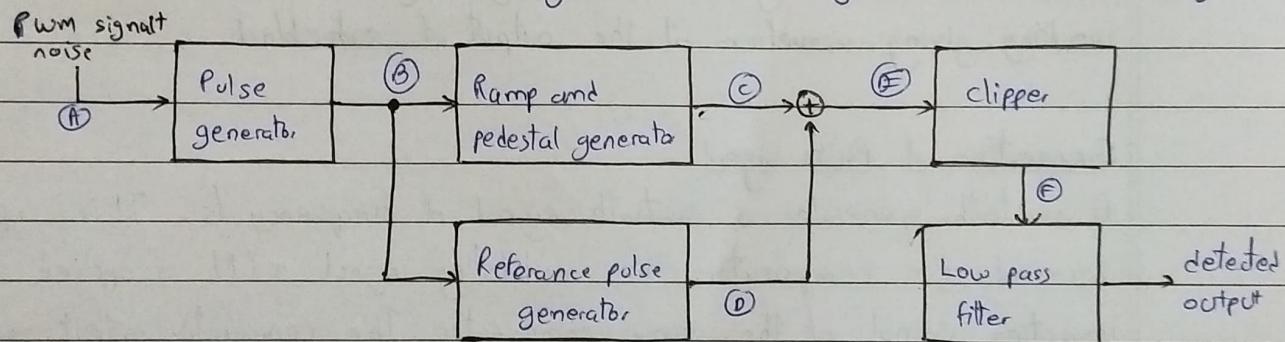
A sawtooth generates a sawtooth signal of frequency f_s . It is applied to inverting terminal of a comparator. The modulating signal $x(t)$ is applied to the non inverting signal of the same comparator. The comparator output will be remain high as long as the instantaneous amplitude of $x(t)$ is higher than that of the ramp signal. This gives rise to the PWM signal at the comparator output



Block diagram for generation of PWM and PPM

Detection of PWM signal

The PWM signal received at input of detection circuit contains noise, hence it is first passed through pulse generator, to remove noise. Hence it is applied to a reference pulse generator which produces a train of constant amplitude, constant width pulse. The regenerated pulse are also applied to the ramp generator the output of which is a constant slope ramp for the duration of pulse. At the end of the pulse a sample and hold amplifier retains the final ramp voltage. The output pulses of reference pulse generator are then added to the ramp signal. The output of adder is then clipped off by a clipper. A low pass filter is used to recover the original modulation signal back from the signal.

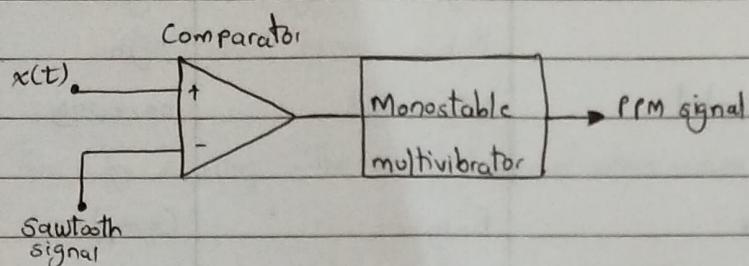


waveform for PWM detection

g) With the help of neat block diagram, and waveform, explain the working of PPM generation and degeneration.

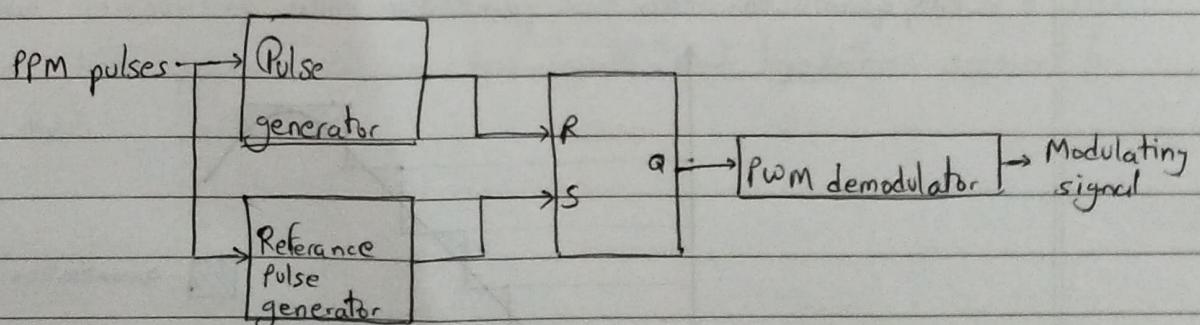
→ Generation of PPM signal:

The PWM pulses obtained at comparator output are applied to the monostable multivibrator. The monostable is negative edge triggered. Corresponding to each training edge of PWM signal, the output goes higher for a fixed time.



Demodulation of PPM signal

Noise corrupted PPM signal is fixed by the pulse generator, and reference pulse generator which output the respective pulses at SR flip flop. Due to the set and reset signal applies to the flip f, we get a PWM signal at its output which is demodulated using PWM demodulator.



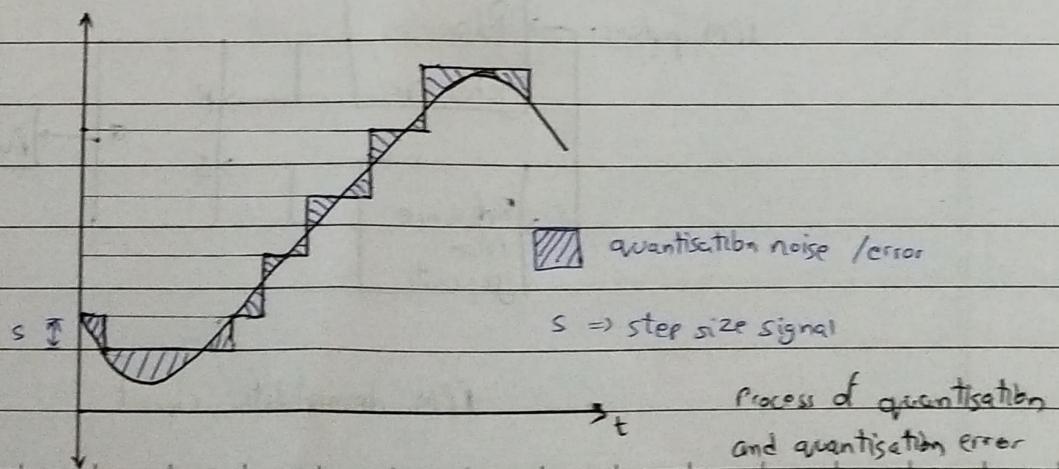
PPM demodulator circuit

10 Compare PAM, PWM and PPM

Parameter	PAM	PWM	PPM
i) Type of carrier	Train of pulses	Train of pulses	Train of pulses
ii) Variable characteristics of pulsed carrier	Amplitude	width	position
iii) B.W requirement	Low	High	High
iv) Noise immunity	Low	High	High
v) Need to transmit	Not needed	Not needed	necessary
vi) Complexity of generation	complex and detection	Easy	Complex
vii) Output waveform			

11) Describe quantisation noise or quantisation error

The difference between instantaneous value of the quantised signal and input signal is called quantisation error or quantisation noise.



(2) What is companding? Show how companding reduces the quantization error

→ Companding is the process of compression and then expansion of the signal. With companded signal. Wi the higher amplitude analog signals are compressed prior to transmission and then expanded in the receiver. Companding is non-uniform quantisation. Due to the inverse nature of compressor and expander characteristics, the compander is a straight line. This indicates that all boosted signals are brought back to their original amplitudes.

$$\text{Quantisation error } N_q = \frac{s^2}{12}$$

Thus by boosting, The SNR of quantisation noise is better

(3) Write A-law and u-law of companding

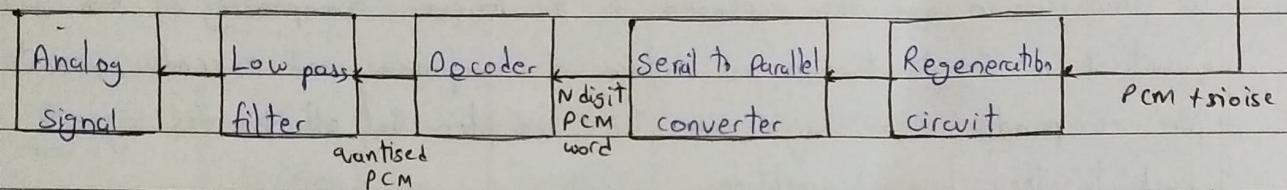
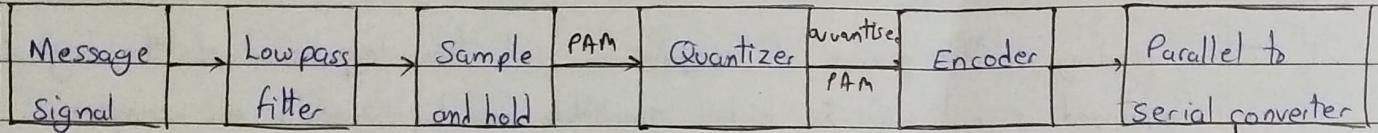
In the u-law companding, the compressor characteristic is continuous. It is approximately linear for smaller values of input levels and logarithmic for high levels of input. Mathematically,

$$z(x) = (\operatorname{sgn} x) \frac{\ln(1 + u|x|/x_{\max})}{\ln(1+u)}$$

where $0 \leq |x|/x_{\max} \leq 1$

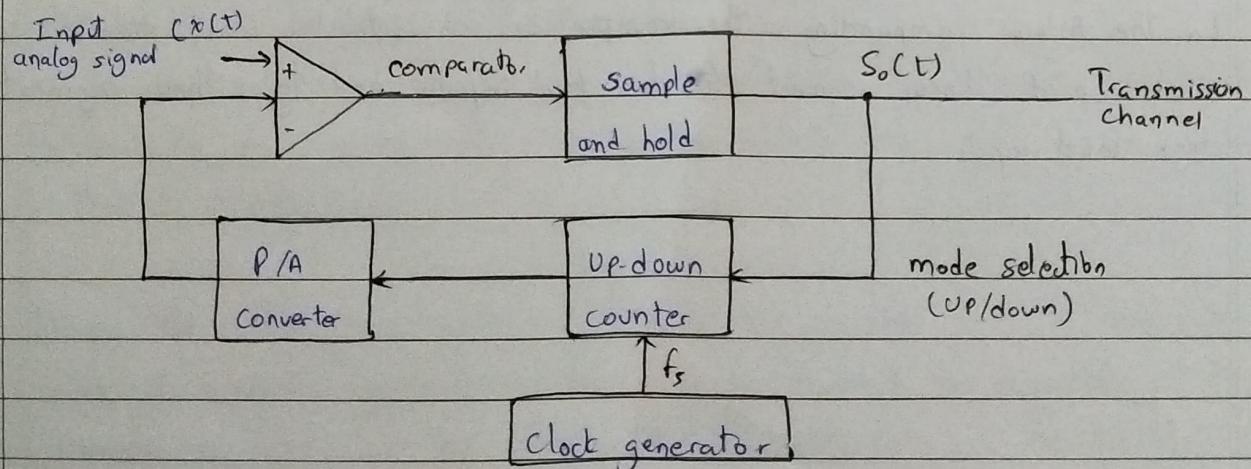
In the A-law companding, the compressor characteristic is of piecewise nature, made up of linear segment for low level inputs and logarithmic segment for high level inputs.

- 14) Draw the block diagram of PCM system, and explain function of each block

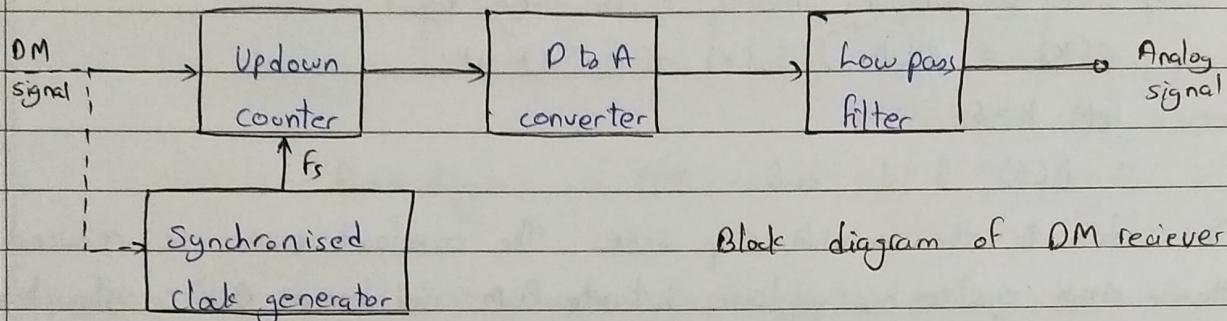


Analog signal $x(t)$ is passed through low pass filter having a cut off frequency $f_c = w$ Hz. This is then passed through sample and hold circuit where it is sampled at high sampling rate. These samples are then subjected to quantisation process in quantizer. The quantized PAM pulses are applied to encoder. The regeneration circuit is a part of the receiver and it separates the PCM pulses from noise. This clean PCM signal is passed through a serial parallel converter, which is then applied to a decoder, which is D to A converter. The quantized PAM obtained from decoder is passed through a low passed filter to recover the analog signal $x(t)$

- 15) Explain the working of delta modulator transmitter and receiver with neat block dia



$x(t)$ and its quantised version $x'(t)$ are applied to the comparator, which results in either 1 or 0. The sample and hold circuit will hold this level for entire clock cycle period, and its output is transmitted as the output of DM transmitter. The transmitted signal is also used to decide the mode of operation of an up/down counter.



Block diagram of DM receiver

The block diagram of DM receiver is shown above, on comparing it with the transmitter block, we find that it's identical to the chain of block producing signal $x'(t)$. The original modulating can be recovered by passing this signal through low pass filter.

- Q6) What are the limitations of linear delta modulator? How do they overcome in adaptive delta modulator.

Disadvantages of linear Delta Modulators

- The two distortions, namely slope overload error and granular noise are present.
- Practically, the signalling rate with no slope overload error will be much higher than that of PCM.

During the working of adaptive delta modulator, in order k th k^{th} form pulse setting edge, the processor generates a step which is equal to the magnitude to the step generator in response to the previous $(k-1)^{\text{th}}$ clock edge. if the direction of both

steps is same, the processor will increase in magnitude and decrease it otherwise
 s_o i.e. output of ADM is given as

$$s_o(t) = 1 \text{ if } (x(t) > x'(t)) \text{ before } k^{\text{th}} \text{ edge}$$

$$\text{and } s_o(t) = -1 \text{ if } x(t) < x'(t) \text{ before } k^{\text{th}} \text{ edge}$$

\therefore Step size at sampling instant k is given by

$$s(k) = s(k-1) s_o(k) + \delta s_o(k-1)$$

$$\text{let } k=6$$

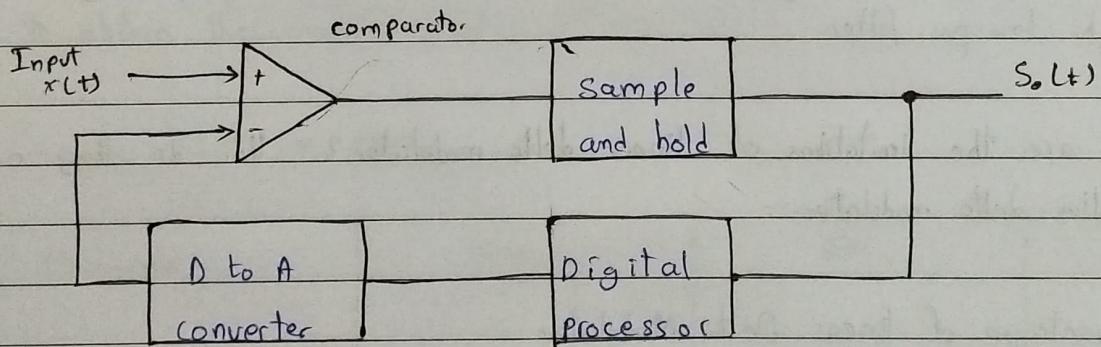
$$\therefore \delta(6) = \delta + f = 2\delta$$

Thus, due to the variable step size, the overload error is reduced.

Hence ADM system has a lower bit rate PCM and hence BW is also less

- 17) Draw the block diagram of ADM transmitter and receiver and explain working

ADM transmitter

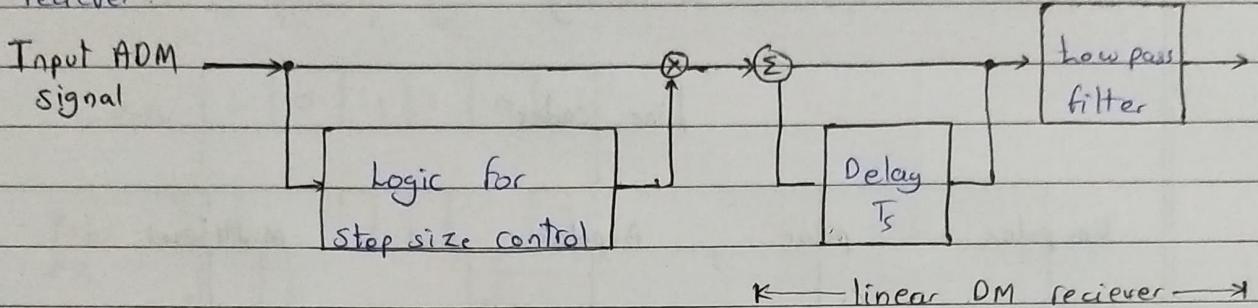


Block diagram of ADM transmitter

Operation:- On comparing this diagram, with that of delta modulator, we find that, except for the counter replaced by digital processor, the remaining blocks are identical. In response to the k^{th} clock pulse trailing edge, the processor generates a step which is equal in magnitude to the step generated in response to previous $(k-1)^{\text{th}}$ clock edge. If direction of both step is same, then there is processor will

will increase the magnitude of the present step by α

ADM receiver:



Block diagram of ADM receiver

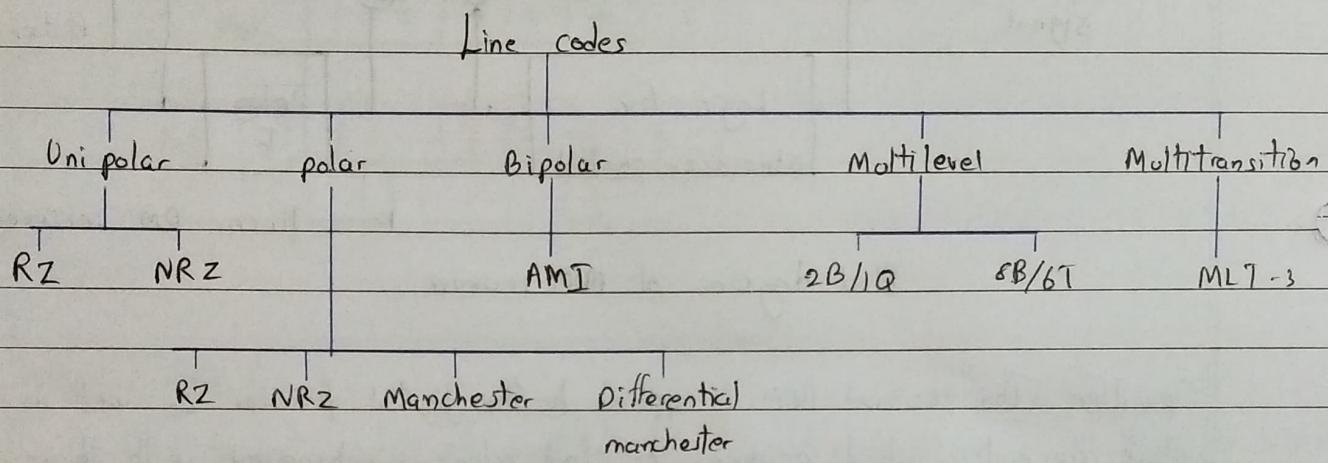
Operation: The received ADM is applied to the digital processor as well as synchronised clock generator, which produces the clock pulses synchronised with those at the transmitter. The digital processor produces a digital signal in parallel form which is converted into an analog signal by P to A converter. The output of DAC is in the form of quantised PAM which is passed through a low pass filter to obtain the analog signal.

18) Compare PCM, DM, ADM

Parameter	PCM	DM	ADM
i) No. of bits per sample	$N = 4, 8, 16, 32 \dots$	$N = 1$	$N = 1$
ii) Step size	Depends on number of α levels	Step size is fixed	Step size is variable
iii) Distortion / errors	Quantization error	Slope overload and granular noise	Granular noise
iv) System complexity	Complex	Simple	Simple
v) Feedback from output	No feedback	Feedback present	Feedback present

19) State and explain different types of line codes.

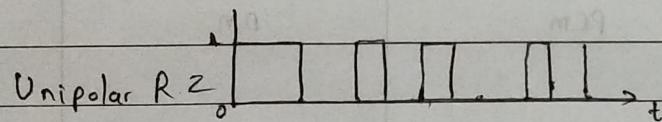
Line codes are classified as following



i) Unipolar RZ format

In this, each "0" is represented by an off pulse (0) and each "1" by an one pulse with amplitude (A) and duration of $T_b/2$ followed by a return to zero level

Data | 1 | 0 | 1 | 1 | 0 | 1 |

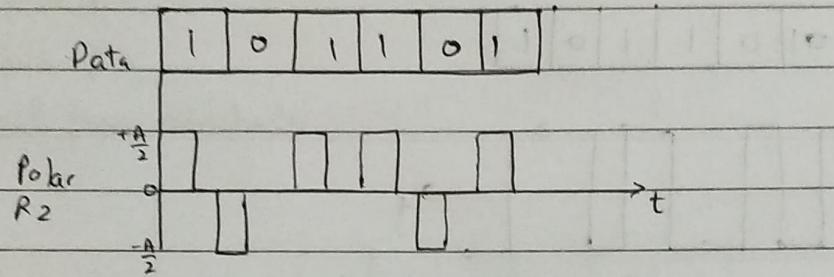


ii) Unipolar NRZ format

In this format a logic "1" is represented by a pulse of full bit duration T_b and amplitude A while a logic "0" is represented by an off pulse or zero amplitude

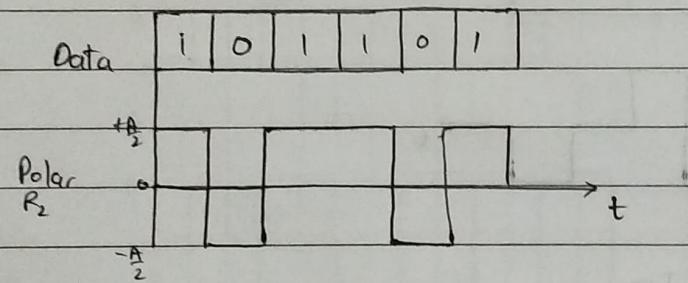
iii) Polar RZ format

In this format, the signal shows opposite polarity pulse of amplitude $\pm \frac{A}{2}$ that are used to represent logic "1" and "0"



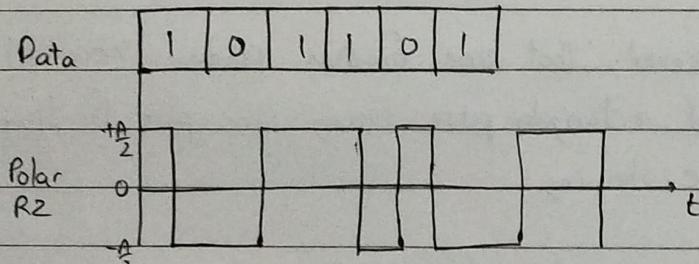
iv) Polar NRZ format

In this format, pulse of amplitude $\pm \frac{A}{2}$ of duration T_b is used to represent a logic "1" and pulse of amplitude $-\frac{A}{2}$ of same duration represented by "0"



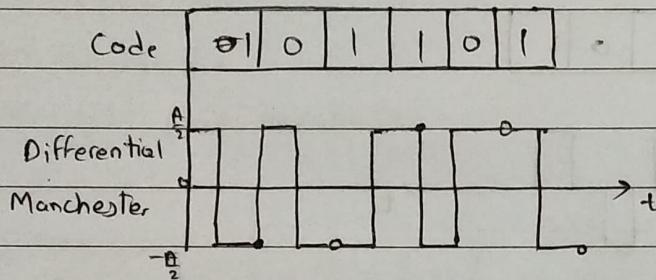
v) Split phase manchester format

In this format, logic "1" is represented by transmitting a positive pulse $\frac{A}{2}$ for one half of the symbol duration, followed by a negative pulse of amplitude $-\frac{A}{2}$ for remaining duration. For logic "0" these places are transmitted in reverse order



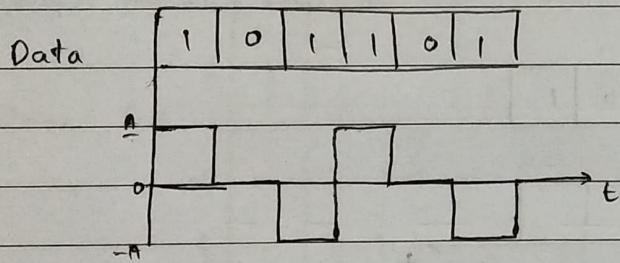
vii) Differential Manchester format

In this code, there is always a transition in middle of a bit interval, the binary zero has an additional transition at the beginning of bit interval



viii) Bipolar NRZ format (AMI)

In this format the successive "1"s are represented by pulses which alternate polarity, and no pulse is transmitted for logic "0"



20) What do you mean by ISI

In communication system when data is being transmitted in the form of pulses the pit produced at the receiver due to the other bits or symbols interfere with output produced by the desired bit. This is called inter-symbol interference or ISI

Remedies to reduce ISI

Since it has been proved that sinc function produces zero ISI, we transmit a sinc pulse instead of rectangular pulse. Using sinc pulse for transmission is known as Nyquist pulse shaping

2(b) The binary data 1 1 0 1 0 1 0 1 is transmitted over a baseband channel draw the waveform for transmitted data using the following data formats

i) Unipolar-

ii) Unipolar RZ

iii) Bipolar RZ

iv) Split phase manchester.

