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Agenda



Introduction

Observability

Full Order Observer

Observer Design

Observer Based Control

Curriculum for Reguleringsteknik (REG)



Matematiske og grafiske metoder til syntese af lineære tidsinvariante systemer:

- ► diskret og kontinuert tilstandsbeskrivelse
- analyse i tid og frekvens
- stabilitet, reguleringshastighed, følsomhed og fejl
- ► digitale PI, PID, LEAD og LAG regulatorer (serieregulatorer)
- ► tilstandsregulering, pole-placement og tilstands-estimering (observer)
- ► optimal regulering (least squares) og optimal tilstands-estimation (Kalman-filter)

Færdigheder:

Efter gennemførelse af kurset kan den succesfulde studerende:

 kunne analysere, dimensionere og implementere såvel kontinuert som tidsdiskret regulering af lineære tidsinvariante og stokastiske systemer

Kompetencer:

Efter gennemførelse af kurset kan den succesfulde studerende:

 anvende og implementere klassiske og moderne reguleringsteknikker for at kunne styre og regulere en robot hurtig og præcist

¹ Based on https://fagbesk.sam.sdu.dk/?fag_id=39673



The twelve lectures of the course are

- ► Lecture 1: Introduction to Linear Time-Invariant Systems
- ► Lecture 2: Stability and Performance Analysis
- ► Lecture 3: Introduction to Control
- ► Lecture 4: Design of PID Controllers
- ► Lecture 5: Root Locus
- ► Lecture 6: The Nyquist Plot
- ► Lecture 7: Dynamic Compensators and Stability Margins
- ► Lecture 8: Implementation
- ► Lecture 9: State Feedback
- ► Lecture 10: Observer Design
- ► Lecture 11: Optimal Control (Linear Quadratic Control)
- ► Lecture 12: The Kalman Filter



Introduction

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A continuous time system

$$\dot{x}(t) = Ax(t), \quad y(t) = Cx(t)$$

is said to be *observable* iff $y(t) \equiv 0 \Rightarrow x(t) \equiv 0$.

A discrete time system

$$x_{k+1} = \Phi x_k \,, \quad y_k = C x_k$$

is said to be *observable* iff $y_k \equiv 0 \Rightarrow x_k \equiv 0$.



We consider the discrete time system

$$x_{k+1} = \Phi x_k \,, \quad y_k = C x_k \,, \quad x_0 = x_0$$

$$x_0 = x_0 \qquad y_0 = Cx_0$$



We consider the discrete time system

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We consider the discrete time system

$$x_{k+1} = \Phi x_k$$
, $y_k = C x_k$, $x_0 = x_0$

$$x_0 = x_0 y_0 = Cx_0$$
 $x_1 = \Phi x_0 y_1 = C\Phi x_0$
 $x_2 = \Phi^2 x_0 y_2 = C\Phi^2 x_0$
 \vdots
 $x_{n-1} = \Phi x_{n-2}$



We consider the discrete time system

$$x_{k+1} = \Phi x_k$$
, $y_k = C x_k$, $x_0 = x_0$

$$x_0 = x_0 y_0 = Cx_0$$
 $x_1 = \Phi x_0 y_1 = C\Phi x_0$
 $x_2 = \Phi^2 x_0 y_2 = C\Phi^2 x_0$
 \vdots
 $x_{n-1} = \Phi x_{n-2}$



We consider the discrete time system

$$x_{k+1} = \Phi x_k$$
, $y_k = C x_k$, $x_0 = x_0$

$$\begin{array}{rclrcl}
x_0 & = & x_0 & y_0 & = & Cx_0 \\
x_1 & = & \Phi x_0 & y_1 & = & C\Phi x_0 \\
x_2 & = & \Phi^2 x_0 & y_2 & = & C\Phi^2 x_0 \\
& \vdots & & & & & \\
x_{n-1} & = & \Phi^{n-1} x_0 & y_{n-1} & = & C\Phi^{n-1} x_0
\end{array}$$



Writing the equations

$$y_k = C\Phi^k x_0, k = 0, \dots, n-1$$

in matrix form we obtain:

Condition for Observability (2)



Writing the equations

$$y_k = C\Phi^k x_0, k = 0, \dots, n-1$$

in matrix form we obtain:

$$\begin{bmatrix}
C \\
C\Phi \\
\vdots \\
C\Phi^{n-1}
\end{bmatrix} x_0 = \begin{bmatrix}
0 \\
0 \\
\vdots \\
0
\end{bmatrix}$$

Observability matrix

Condition for Observability (2)



Writing the equations

$$y_k = C\Phi^k x_0, k = 0, \dots, n-1$$

in matrix form we obtain:

$$\begin{bmatrix} C \\ C\Phi \\ \vdots \\ C\Phi^{n-1} \end{bmatrix} \quad x_0 = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

When is this equation solvable for some $x_0 \neq 0$?



THEOREM. A system

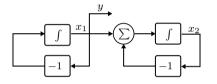
continuous time	discrete time
$\Sigma : \left\{ \begin{array}{l} \dot{x}(t) &= Ax(t) \\ y(t) &= Cx(t) \end{array} \right.$	$\Sigma : \left\{ \begin{array}{rcl} x_{k+1} & = & \Phi x_k \\ y_k & = & C x_k \end{array} \right.$

where $x \in \mathbb{R}^n$, $y \in \mathbb{R}^p$, is observable if and only if

$$\operatorname{rank} \mathcal{O} = \operatorname{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} = n$$

Example: Series Connection (1)





State and output equations:

$$\left\{
\begin{array}{ccc}
\dot{x}_1 & = & -x_1 \\
\dot{x}_2 & = & -x_2 + x_1 \\
y & = & x_1
\end{array}
\right\}$$

State space model:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$
$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Example: Series Connection (2)



For the state space matrices:

$$A = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix}, \quad C = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

the observability matrix $\ensuremath{\mathcal{O}}$ becomes:

$$\mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -1 & 0 \end{bmatrix}$$

Example: Series Connection (2)



For the state space matrices:

$$A = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix}, \quad C = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

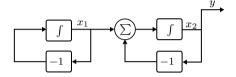
the observability matrix \mathcal{O} becomes:

$$\mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -1 & 0 \end{bmatrix}$$

 $\det \mathcal{O} = 0 \implies$ system is unobservable.

Example: Series Connection (3)





State and output equations:

$$\left\{
\begin{array}{ccc}
\dot{x}_1 & = & -x_1 \\
\dot{x}_2 & = & -x_2 + x_1 \\
y & = & x_2
\end{array}
\right\}$$

State space model:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$
$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Example: Series Connection (4)



For the state space matrices:

$$A = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix}, \quad C = \begin{bmatrix} 0 & 1 \end{bmatrix}$$

the observability matrix \mathcal{O} becomes:

$$\mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix}$$

Example: Series Connection (4)



For the state space matrices:

$$A = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix}, \quad C = \begin{bmatrix} 0 & 1 \end{bmatrix}$$

the observability matrix \mathcal{O} becomes:

$$\mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix}$$

 $\det \mathcal{O} = -1 \neq 0 \implies$ system is observable.

Full Order Observer



Introduction

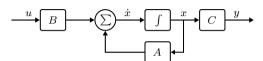
Observability

Full Order Observer

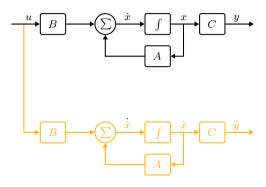
Observer Design

Observer Based Contro

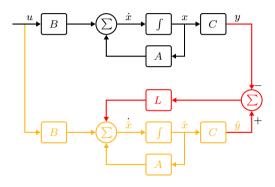














System:
$$\dot{x} = Ax + Bu$$

 $y = Cx$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$
 $\hat{y} = C\hat{x}$



System:
$$\dot{x} = Ax + Bu$$

 $y = Cx$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$

 $\dot{y} = C\hat{x}$

Error,
$$e = \hat{x} - x$$
:

$$\dot{e} = \frac{\dot{\hat{x}}}{\hat{x}} - \dot{x} = A\hat{x} + Bu + L(C\hat{x} - y) - (Ax + Bu)$$



System:
$$\dot{x} = Ax + Bu$$

 $y = Cx$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$

 $\dot{y} = C\hat{x}$

Error,
$$e = \hat{x} - x$$
:

$$\dot{e} = \dot{\hat{x}} - \dot{x} = A\hat{x} + Bu + L(C\hat{x} - y) - (Ax + Bu)$$
$$= A(\hat{x} - x) + L(C\hat{x} - Cx)$$



System:
$$\dot{x} = Ax + Bu$$

 $y = Cx$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$

 $\dot{y} = C\hat{x}$

Error,
$$e = \hat{x} - x$$
:

$$\dot{e} = \dot{\hat{x}} - \dot{x} = A\hat{x} + Bu + L(C\hat{x} - y) - (Ax + Bu)$$

$$= A(\hat{x} - x) + L(C\hat{x} - Cx)$$

$$= (A + LC)(\hat{x} - x) = (A + LC)e$$



THEOREM. A full order observer for the system

with observer gain L is stable, if and only if the eigenvalues of the matrix A + LC all have negative real part.

Moreover, such an L always exists, if (A, C) is observable.

Full Order Observer Observable Canonical Form (1)



Any observable *single output* system can be written in the form:

$$\dot{x}_o = A_o x_o$$
, $y = C_o x_o$, $x_o \in \mathbb{R}^n$, $y \in \mathbb{R}$

where

$$A_o = \left[\begin{array}{c} a & I_{n-1} \\ \hline 0_{1\times(n-1)} \end{array} \right], \quad C_o = \left[\begin{array}{c} 1 & 0_{1\times(n-1)} \end{array} \right]$$

and where $a \in \mathbb{R}^{n \times 1}$, $a^T = \begin{bmatrix} a_1 & a_2 & \dots & a_n \end{bmatrix}$. It can be shown that

$$\det(\lambda I - A_o) = \lambda^n - a_1 \lambda^{n-1} - \dots - a_n$$

Full Order Observer Observable Canonical Form (2)



For n=3 the observable canonical form becomes:

$$A_o = \begin{bmatrix} a_1 & 1 & 0 \\ a_2 & 0 & 1 \\ a_3 & 0 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$

which is indeed observable:

$$\mathcal{O}_o = \begin{bmatrix} C_o \\ C_o A_o \\ C_o A_o^2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ a_1 & 1 & 0 \\ a_1^2 + a_2 & a_1 & 1 \end{bmatrix}$$

 $\det(\mathcal{O}) = 1 \neq 0 \Longrightarrow$ system is observable.

Full Order Observer Observable Canonical Form (3)



Consider a system:

$$\dot{x} = Ax$$
, $y = Cx$, $x \in \mathbb{R}^n$, $y \in \mathbb{R}$

For n=3, the observable canonical form for this system can be found through the following procedure:

1. Compute
$$t_3 = \mathcal{O}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$
 where $\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix}$

Observable Canonical Form (3)



1. Compute
$$t_3 = \mathcal{O}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$
 where $\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix}$

2. Compute $t_2 = At_3$, $t_1 = At_2$.

Observable Canonical Form (3)



1. Compute
$$t_3 = \mathcal{O}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$
 where $\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix}$

- **2.** Compute $t_2 = At_3$, $t_1 = At_2$.
- **3.** Define $T = \begin{bmatrix} t_1 & t_2 & t_3 \end{bmatrix}$

Full Order Observer Observable Canonical Form (3)



1. Compute
$$t_3 = \mathcal{O}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$
 where $\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix}$

- **2.** Compute $t_2 = At_3$, $t_1 = At_2$.
- **3.** Define $T = \begin{bmatrix} t_1 & t_2 & t_3 \end{bmatrix}$
- 4. The state space matrices for the observable canonical form are now given by $A_o=T^{-1}AT$, and $C_o=CT$.

Example: Observable Canonical Form (1)



We consider the system

$$\dot{x} = \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} x + \begin{bmatrix} 2 \\ 3 \end{bmatrix} u$$

$$y = \begin{bmatrix} -3 & 2 \end{bmatrix} x$$

having the observability matrix

$$\mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} -3 & 2 \\ 2 & -1 \end{bmatrix}, \quad \det(\mathcal{O}) = -1 \neq 0$$

Example: Observable Canonical Form (2)



We compute the columns of T by

$$t_2 = \mathcal{O}^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \begin{bmatrix} 1 & 2 \\ 2 & 3 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \begin{bmatrix} 2 \\ 3 \end{bmatrix}$$
$$t_1 = At_2 = \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} \begin{bmatrix} 2 \\ 3 \end{bmatrix} = \begin{bmatrix} -5 \\ -7 \end{bmatrix}$$

Thus,

$$T = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \implies T^{-1} = \begin{bmatrix} -3 & 2 \\ -7 & 5 \end{bmatrix}$$

Example: Observable Canonical Form (3)



Eventually, we have

$$A_o = T^{-1}AT = \begin{bmatrix} -3 & 2 \\ -7 & 5 \end{bmatrix} \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix}$$
$$= \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}$$

Example: Observable Canonical Form (3)



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Example: Observable Canonical Form (3)



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$$= \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}$$

$$C_o = CT = \begin{bmatrix} -3 & 2 \end{bmatrix} \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

Example: Observable Canonical Form (3)



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Example: Observable Canonical Form (3)



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$$= \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix} \Rightarrow \det(\lambda I - A) = \lambda^2 + 3\lambda + 2$$

$$C_o = CT = \begin{bmatrix} -3 & 2 \end{bmatrix} \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

Example: Observable Canonical Form (3)



Eventually, we have

$$A_o = T^{-1}AT = \begin{bmatrix} -3 & 2 \\ -7 & 5 \end{bmatrix} \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix}$$
$$= \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix} \Rightarrow \det(\lambda I - A) = (\lambda + 1)(\lambda + 2)$$

$$C_o = CT = \begin{bmatrix} -3 & 2 \end{bmatrix} \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} = \begin{bmatrix} 1 & 0 \end{bmatrix}$$



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Observer Design Observer gain design (1)



For a single output system in observable canonical form, an observer state matrix takes a particular simple form:

$$A_o = \begin{bmatrix} a_1 & 1 & 0 \\ a_2 & 0 & 1 \\ a_3 & 0 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$

Applying the observer gain

$$L_o = \begin{bmatrix} \ell_1 \\ \ell_2 \\ \ell_3 \end{bmatrix}$$

Observer Design Observer Gain Design (2)



we obtain:

$$A_o + L_o C_o = \begin{bmatrix} a_1 & 1 & 0 \\ a_2 & 0 & 1 \\ a_3 & 0 & 0 \end{bmatrix} + \begin{bmatrix} \ell_1 \\ \ell_2 \\ \ell_3 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$
$$= \begin{bmatrix} a_1 + \ell_1 & 1 & 0 \\ a_2 + \ell_2 & 0 & 1 \\ a_3 + \ell_3 & 0 & 0 \end{bmatrix}$$



Thus, the characteristic polynomial has been changed from

$$\det(\lambda I - A_o) = \lambda^n - a_1 \lambda^{n-1} - \dots - a_n$$

to

$$\det(\lambda I - (A_o + L_o C_o)) = \lambda^n - (a_1 + \ell_1)\lambda^{n-1} - \dots - (a_n + \ell_n)$$

By choosing ℓ_1, \dots, ℓ_n appropriately, *any* observer pole configuration can be obtained. This is known as *observer pole assignment*.



Let $A \in \mathbb{R}^{n \times n}$, $C \in \mathbb{R}^{1 \times n}$ be given.

1. Choose desired observer polynomial $\det(\lambda I - (A + LC)) = \lambda^n - a_{\mathsf{obs},1}\lambda^{n-1} - \ldots - a_{\mathsf{obs},n}.$



Let $A \in \mathbb{R}^{n \times n}$, $C \in \mathbb{R}^{1 \times n}$ be given.

- 1. Choose desired observer polynomial $\det(\lambda I (A + LC)) = \lambda^n a_{\mathsf{obs},1}\lambda^{n-1} \ldots a_{\mathsf{obs},n}.$
- **2.** Determine T, such that $A_o = T^{-1}AT$ and $C_o = CT$ are in observable canonical form.

Observer Pole Assignment



Let $A \in \mathbb{R}^{n \times n}$, $C \in \mathbb{R}^{1 \times n}$ be given.

- 1. Choose desired observer polynomial $\det(\lambda I (A + LC)) = \lambda^n a_{\mathsf{obs},1}\lambda^{n-1} \ldots a_{\mathsf{obs},n}.$
- **2.** Determine T, such that $A_o = T^{-1}AT$ and $C_o = CT$ are in observable canonical form.
- 3. Determine open loop polynomial $det(\lambda I A) = \lambda^n a_1 \lambda^{n-1} \ldots a_n$



Let $A \in \mathbb{R}^{n \times n}$, $C \in \mathbb{R}^{1 \times n}$ be given.

- 1. Choose desired observer polynomial $\det(\lambda I (A + LC)) = \lambda^n a_{\text{obs},1}\lambda^{n-1} \ldots a_{\text{obs},n}$.
- **2.** Determine T, such that $A_o = T^{-1}AT$ and $C_o = CT$ are in observable canonical form.
- **3.** Determine open loop polynomial $det(\lambda I A) = \lambda^n a_1 \lambda^{n-1} \ldots a_n$

4. Define
$$L_o = \begin{bmatrix} a_{\mathsf{obs},1} - a_1 \\ \vdots \\ a_{\mathsf{obs},n} - a_n \end{bmatrix}$$
.

Observer Design Observer Pole Assignment



Let $A \in \mathbb{R}^{n \times n}$, $C \in \mathbb{R}^{1 \times n}$ be given.

- 1. Choose desired observer polynomial $\det(\lambda I (A + LC)) = \lambda^n a_{\mathsf{obs},1}\lambda^{n-1} \ldots a_{\mathsf{obs},n}.$
- **2.** Determine T, such that $A_o = T^{-1}AT$ and $C_o = CT$ are in observable canonical form.
- **3.** Determine open loop polynomial $det(\lambda I A) = \lambda^n a_1 \lambda^{n-1} \ldots a_n$

4. Define
$$L_o = \begin{bmatrix} a_{\mathsf{obs},1} - a_1 \\ \vdots \\ a_{\mathsf{obs},n} - a_n \end{bmatrix}$$
.

5. Compute resulting observer gain $L = TL_o$.

Example: Observer Pole Assignment (1)



We consider again the system

$$\dot{x} = \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} x + \begin{bmatrix} 2 \\ 3 \end{bmatrix} u$$

$$y = \begin{bmatrix} -3 & 2 \end{bmatrix} x$$

for which we would like to assign observer poles to $\{-4, -5\}$, i.e. to design L such that A + LC has eigenvalues in $\{-4, -5\}$.

Observer Design Example: Observer Pole Assignment (2)



1. Desired closed loop polynomial: $\lambda^2 + 9\lambda + 20$

Example: Observer Pole Assignment (2)



1. Desired closed loop polynomial: $\lambda^2 + 9\lambda + 20$

2.
$$T = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \Rightarrow A_o = \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

Example: Observer Pole Assignment (2)



1. Desired closed loop polynomial: $\lambda^2 + 9\lambda + 20$

2.
$$T = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \Rightarrow A_o = \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

3. Open loop polynomial: $\lambda^2 + 3\lambda + 2$

Example: Observer Pole Assignment (2)



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$$T = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \Rightarrow A_o = \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

3. Open loop polynomial: $\lambda^2 + 3\lambda + 2$

4.
$$L_o = \begin{bmatrix} -9 - (-3) \\ -20 - (-2) \end{bmatrix} = \begin{bmatrix} -6 \\ -18 \end{bmatrix}$$

Example: Observer Pole Assignment (2)



1. Desired closed loop polynomial: $\lambda^2 + 9\lambda + 20$

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$$T = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \Rightarrow A_o = \begin{bmatrix} -3 & 1 \\ -2 & 0 \end{bmatrix}, C_o = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

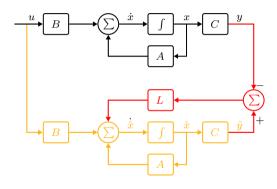
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$$L_o = \begin{bmatrix} -9 - (-3) \\ -20 - (-2) \end{bmatrix} = \begin{bmatrix} -6 \\ -18 \end{bmatrix}$$

5.
$$L = TL_o = \begin{bmatrix} -5 & 2 \\ -7 & 3 \end{bmatrix} \begin{bmatrix} -6 \\ -18 \end{bmatrix} = \begin{bmatrix} -6 \\ -12 \end{bmatrix}$$

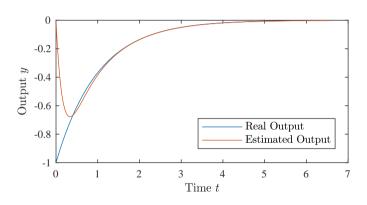
Observer Design The Full Order Observer





Observer Design Example: Observer Pole Assignment





Observer Based Control



Introduction

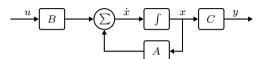
Observability

Full Order Observer

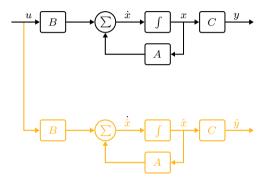
Observer Design

Observer Based Control

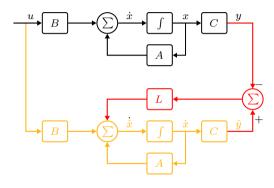




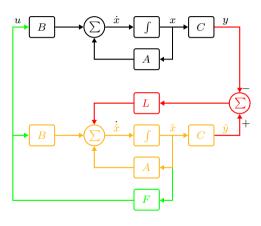














System:
$$\begin{array}{cccc} \dot{x} & = & Ax & + & Bu \\ y & = & Cx \end{array}$$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$

 $\hat{y} = C\hat{x}$

Feedback:
$$u = F\hat{x}$$



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$$= \frac{A\hat{x}}{\hat{x}} + \frac{BF\hat{x}}{\hat{x}} + \frac{L(C\hat{x} - y) - (Ax + BF\hat{x})}{\hat{x}}$$



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$$= A(\hat{x} - x) + \frac{L(C\hat{x} - Cx)}{\hat{x}}$$



System:
$$\begin{array}{cccc} \dot{x} & = & Ax & + & Bu \\ u & = & Cx \end{array}$$

Observer:
$$\dot{\hat{x}} = A\hat{x} + Bu + L(C\hat{x} - y)$$

 $\dot{\hat{y}} = C\hat{x}$

Feedback:
$$u = F\hat{x}$$

Error, $e = \hat{x} - x$:

$$\begin{split} \dot{e} &= \dot{\hat{x}} - \dot{x} \\ &= A\hat{x} + BF\hat{x} + L(C\hat{x} - y) - (Ax + BF\hat{x}) \\ &= A(\hat{x} - x) + L(C\hat{x} - Cx) \\ &= (A + LC)(\hat{x} - x) = (A + LC)e \end{split}$$

Observer Based Control The Separation Principle (1)



Combining the two equations:

$$\dot{x} = Ax + Bu = Ax + BF\hat{x} = Ax + BF(e+x)$$
$$= (A+BF)x + BFe$$

and

$$\dot{e} = (A + \mathbf{L}C)e$$

gives:

$$\begin{bmatrix} \dot{x} \\ \dot{e} \end{bmatrix} = \begin{bmatrix} A + BF & BF \\ 0 & A + \mathbf{L}C \end{bmatrix} \begin{bmatrix} x \\ e \end{bmatrix}$$

Observer Based Control The Separation Principle (2)



THEOREM. An observer based controller for the system

$$\begin{array}{lclcrcl} \dot{x} & = & Ax & + & Bu & , & x \in \mathbb{R}^n \\ y & = & Cx & & & \end{array}$$

with observer gain L and feedback gain F results in 2n closed loop poles, coinciding with the eigenvalues of the two matrices:

$$A + BF$$
 and $A + LC$

Observer Based Control

Example: Observer Based Control (1)



We consider again the system

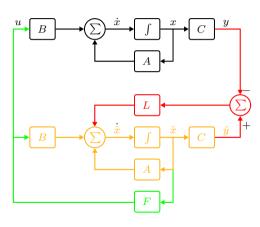
$$\dot{x} = \begin{bmatrix} 2 & -3 \\ 4 & -5 \end{bmatrix} x + \begin{bmatrix} 2 \\ 3 \end{bmatrix} u$$

$$y = \begin{bmatrix} -3 & 2 \end{bmatrix} x$$

for which we apply an observer based controller with

$$\mathbf{L} = \begin{bmatrix} -6 \\ -12 \end{bmatrix} \quad \text{and} \quad F = \begin{bmatrix} 42 & -30 \end{bmatrix}$$





Observer Based Control

Example: Observer Based Control (2)



The transfer function of the controller becomes:

$$K(s) = -F (sI - A - BF - LC)^{-1} L$$
$$= -108 \frac{s + \frac{7}{3}}{s^2 + 15s + 74}$$

The closed loop transfer function becomes:

$$G(s) (I - K(s)G(s))^{-1} = \frac{s^2 + 15s + 74}{(s+5)^2(s+4)^2}$$

Observer Based Control Example: Observer Based Control (3)



