

Physics 2. Electrical Engineering Week 10 Dynamic Circuits Analysis 2



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Objectives



The main objectives of today's lecture are:

- Review transient response of first-order circuits
- Study the transient response of second-order circuits
- Draw analogies between electrical and mechanical dynamic systems

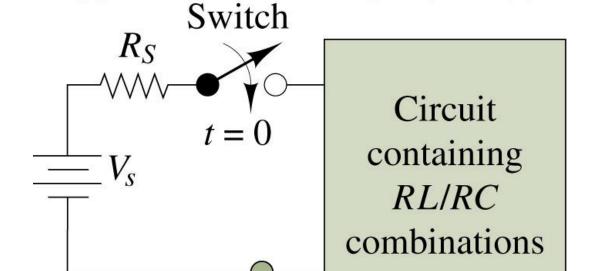
Transient Response of First-Order Circuits: Review

Introduction to Transient Analysis



The object of transient analysis is to describe the behavior of a voltage or a current during the transition between two distinct steady-state conditions.

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Transient Analysis



The previous equation can be rewritten as

$$a_1 \frac{dx(t)}{dt} + a_0 x(t) = b_0 f(t)$$

$$\frac{a_1}{a_0} \frac{dx(t)}{dt} + x(t) = \frac{b_0}{a_0} f(t)$$
or
$$\tau \frac{dx(t)}{dt} + x(t) = K_S f(t)$$

The constants are

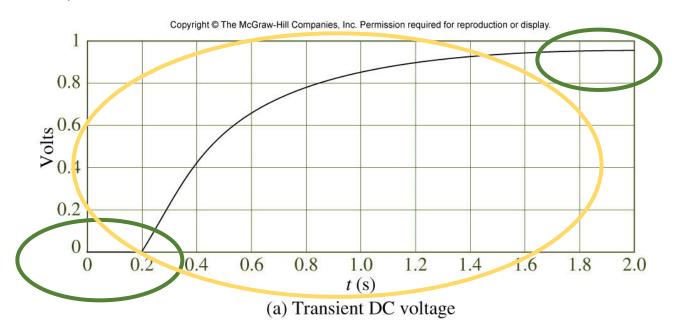
- τ time constant
- K_S DC gain.

Transient Response (1)



The transient response of a circuit consists of three parts:

- 1. The steady-state response prior to the transient.
- 2. The transient response.
- 3. The steady-state response after the end of the transient.



Transient Response (3)



The steps to find a first-order transient response are:

- 1. Solve for the steady-state response x(0) of the circuit before the switch changes state $(t = 0^-)$ and after the transient has died out $(x(\infty), t \to \infty)$.
- 2. Identify the initial conditions for the circuit.
- 3. Write the differential equation of the circuit for $t=0^+$.
- 4. Solve for the time constant of the circuit: $\tau = RC$ or $\tau = L/R$.
- 5. Write the complete solution for the circuit in the form

$$x(t) = x(\infty) + [x(0) - x(\infty)]e^{-t/\tau}$$

Transient Response: Example (1)

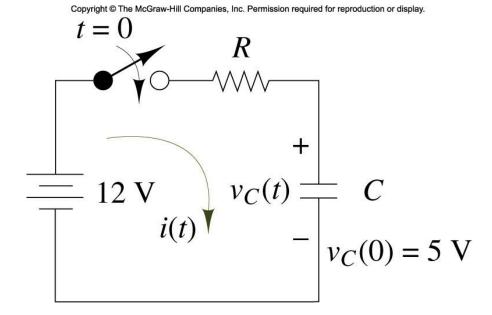


Determine an expression for the capacitor voltage in the circuit shown.

Given:
$$v_C(t = 0^-) = 5 \text{ V}$$
, $v_B = 12 \text{ V}$.

When the switch has been closed for a long time, the capacitor current becomes zero (capacitor = open circuit). Hence,

$$v_C(\infty) = V_B = 12V$$



Transient Response: Example (2)



Using KVL, we can write

$$V_B - Ri_C(t) - v_C(t) = V_B - RC \frac{dv_C(t)}{dt} - v_C(t) = 0$$

$$RC \frac{dv_C(t)}{dt} + v_C(t) = V_B, \qquad t \ge 0$$

- Time constant is $\tau = RC$
- Hence,

$$v_C(t) = v_C(\infty) + [v_C(0) - v_C(\infty)]e^{-t/\tau}$$

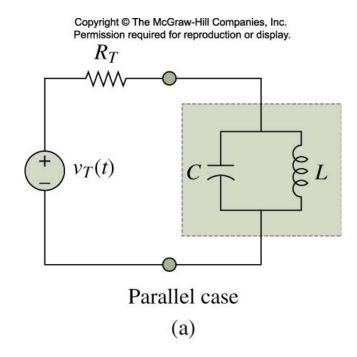
Copyright © The McGraw-Hill Companies, Inc. Permission required for reproduction or display. Complete, transient and steady-state response of RC circuit 15 10 -vc(t)vcT(t)---- vcSS(t)Volts -5-100.2 0.4 0.6 0.8 1.2 1.4 1.6 1.8 Time(s) (a)

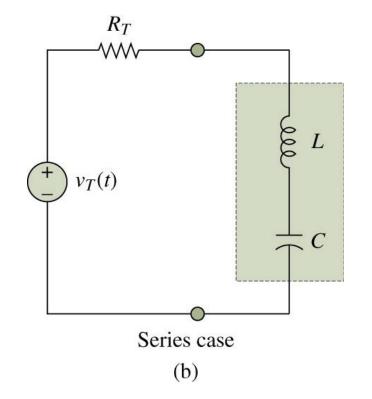
Transient Response of Second-Order Circuits

Second-Order Circuits (1)



A simple second-order circuit is a circuit containing R-L-C elements.





Second-Order Circuits: Parallel Case (1)



Let us consider the parallel case first.

Starting with KVL around the left-hand loop,

$$v_T(t) - R_T i_S(t) - v_C(t) = 0$$

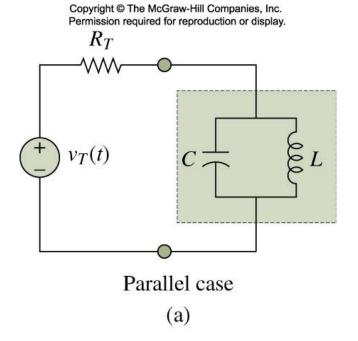
- Then apply KCL to the top node to obtain $i_S(t) i_C(t) i_L(t) = 0$
- We know that

$$v_C(t) = v_L(t)$$

Hence

$$\frac{v_T(t) - v_C(t)}{R_T} - C \frac{dv_C(t)}{dt} - i_L(t) = 0,$$

$$v_C(t) = L \frac{di_L(t)}{dt}$$



Second-Order Circuits: Parallel Case (2)

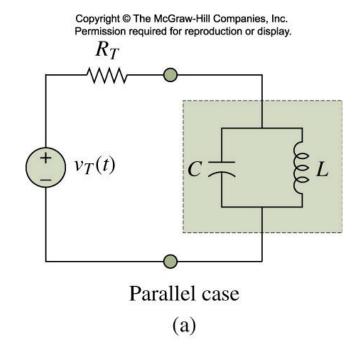


Finally, we obtain

$$\frac{1}{R_T}v_T(t) - \frac{L}{R_T}\frac{di_L(t)}{dt} = LC\frac{d^2i_L(t)}{dt^2} + i_L(t)$$
or
$$LC\frac{d^2i_L(t)}{dt^2} + \frac{L}{R_T}\frac{di_L(t)}{dt} + i_L(t) = \frac{1}{R_T}v_T(t)$$

• This equation can be written in the following form:

$$\frac{1}{\omega_n^2} \frac{d^2 x(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{dx(t)}{dt} + x(t) = K_S f(t)$$



Second-Order Circuits (2)

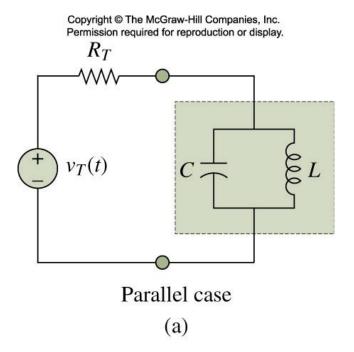


The equations of this type are called second-order linear ordinary differential equations

$$\frac{1}{\omega_n^2} \frac{d^2 x(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{dx(t)}{dt} + x(t) = K_S f(t)$$

The parameters are:

- ω_n the natural frequency
- ζ the damping ratio
- K_S DC gain.



Second-Order Circuits: Series Case



Now, let us derive the equations for the series RLC circuit.

Starting with KVL around the loop,

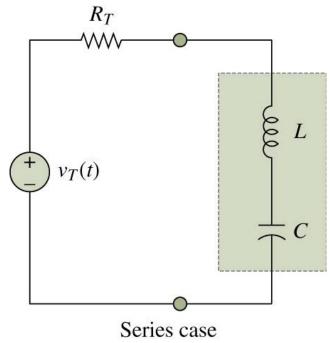
$$v_T - R_T i_L - v_C - v_L = 0$$

Recalling the equations for capacitor and inductor,

$$v_T - R_T i_L - v_C - L \frac{di_L}{dt} = 0$$
, $i_L = i_C = C \frac{dv_C}{dt}$

• Hence, writing for v_c yields

$$LC\frac{d^2v_C}{dt^2} + R_TC\frac{dv_C}{dt} + v_C = v_T \rightarrow \frac{1}{\omega_n^2}\frac{d^2x(t)}{dt^2} + \frac{2\zeta}{\omega_n}\frac{dx(t)}{dt} + x(t) = K_Sf(t)$$



(b)

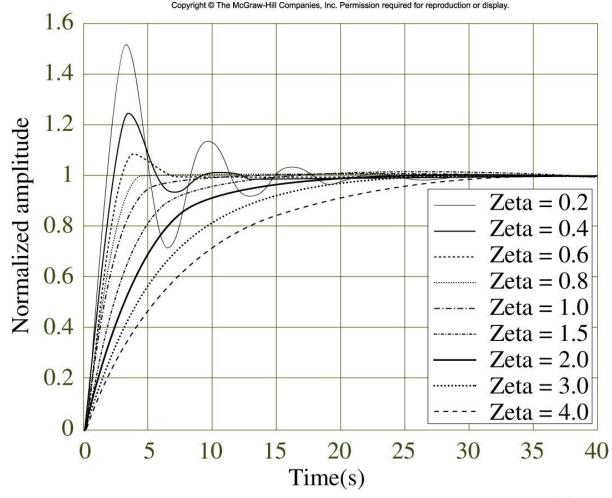
Second-Order Circuits (3)



A typical response of switched second-order system is shown here for different values of ζ .

We immediately note three important characteristics of the response:

- 1. The response asymptotically tends to a constant value.
- The response oscillates.
- 3. The oscillations decay (and eventually disappear) as time progresses.



Second-Order Circuits: Natural Response



The natural response is found by setting the excitation equal to zero.

$$\frac{1}{\omega_n^2} \frac{d^2 x(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{dx(t)}{dt} + x(t) = K_S f(t)$$

Thus, we solve the equation

$$\frac{1}{\omega_n^2} \frac{d^2 x_N(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{d x_N(t)}{dt} + x_N(t) = 0$$

- The solution is known to be of exponential form: $x_N(t) = \alpha e^{st}$
- Substituting into the initial equation yields

$$\frac{1}{\omega_n^2} s^2 \alpha e^{st} + \frac{2\zeta}{\omega_n} s \alpha e^{st} + \alpha e^{st} = 0 \quad \Rightarrow \frac{s^2}{\omega_n^2} + \frac{2\zeta}{\omega_n} s + 1 = 0$$

Second-Order Circuits: Natural Response



This polynomial in the variable s is called the characteristic polynomial of the differential equation, and it gives rise to two characteristic roots, s_1 and s_2 .

$$\frac{s^2}{\omega_n^2} + \frac{2\zeta}{\omega_n}s + 1 = 0$$

$$x_N(t) = \alpha_1 e^{S_1 t} + \alpha_2 e^{S_2 t}$$

$$s_{1,2} = -\zeta \omega_n \pm \omega_n \sqrt{\zeta^2 - 1}$$

For the roots of second-order systems, three cases exist:

- 1. Real and distinct roots ($\zeta > 1$). This leads to an overdamped response
- 2. Real and repeated roots ($\zeta = 1$) critically damped response
- 3. Complex conjugate roots ($0 < \zeta < 1$) underdamped response

$$s_{1,2} = -\zeta \omega_n = -\omega_n$$

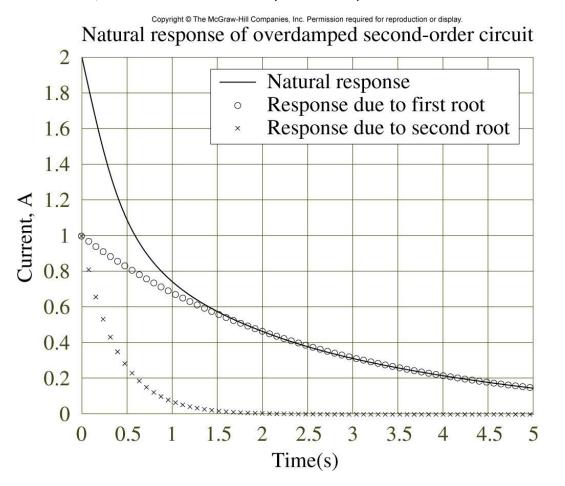
$$s_{1,2} = -\zeta \omega_n \pm j\omega_n \sqrt{1 - \zeta^2}$$

 $s_{1,2} = -\zeta \omega_n \pm \omega_n \sqrt{\zeta^2 - 1}$

Second-Order Circuits: Overdamped



1. Real and distinct roots ($\zeta > 1$) - overdamped response



$$s_{1,2} = -\zeta \omega_n \pm \omega_n \sqrt{\zeta^2 - 1}$$

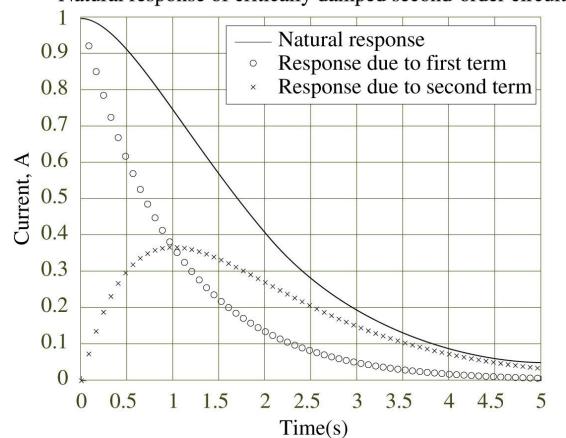
$$x_N(t) = \alpha_1 e^{s_1 t} + \alpha_2 e^{s_2 t}$$

Second-Order Circuits: Critically Damped



2. Real and repeated roots ($\zeta = 1$) - critically damped response

Natural response of critically damped second-order circuit



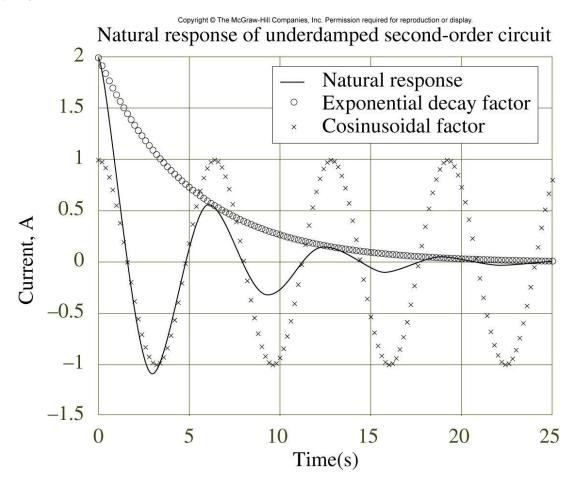
$$s_{1,2} = -\zeta \omega_n = -\omega_n$$

$$x_N(t) = \alpha_1 e^{s_1 t} + \alpha_2 e^{s_2 t}$$

Second-Order Circuits: Underdamped



3. Complex conjugate roots ($0 < \zeta < 1$) - underdamped response



$$s_{1,2} = -\zeta \omega_n \pm j \omega_n \sqrt{1 - \zeta^2}$$

$$x_N(t) = \alpha_1 e^{s_1 t} + \alpha_2 e^{s_2 t}$$

Second-Order Circuits: Complete Response

The forced response is the solution of the equation

$$\frac{1}{\omega_n^2} \frac{d^2 x(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{dx(t)}{dt} + x(t) = K_S f(t)$$

After the transient settles,

$$x_F(t) = K_S f(t)$$

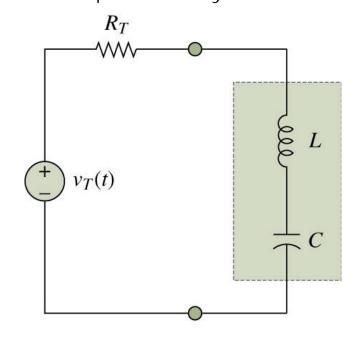
The complete response is then

$$x(t) = x_N(t) + x_F(t)$$

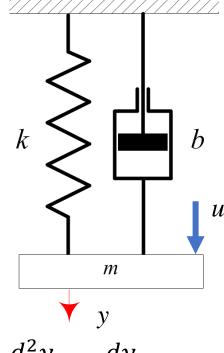
Second-Order Systems: Analogies (1)



Now, let us compare two systems: electrical and mechanical



$$L\frac{d^2v_C}{dt^2} + R_T\frac{dv_C}{dt} + \frac{1}{C}v_C = \frac{1}{C}v_T$$



$$m\frac{d^2y}{dt^2} + b\frac{dy}{dt} + ky = u$$

Second-Order Systems: Analogies (2)



Upon comparing the equations, one can readily identify the analogies between the electrical and mechanical worlds:

Electrical System	Mechanical System
Inductance L	Mass m
Resistance R	Damping coeff. b
Capacitance <i>C</i>	Compliance 1/k

$$L \frac{d^2 v_C}{dt^2} + R_T \frac{dv_C}{dt} + \frac{1}{C} v_C = \frac{1}{C} v_T$$

$$m \frac{d^2 y}{dt^2} + b \frac{dy}{dt} + k y = u$$

Second-Order Systems: Oscillations



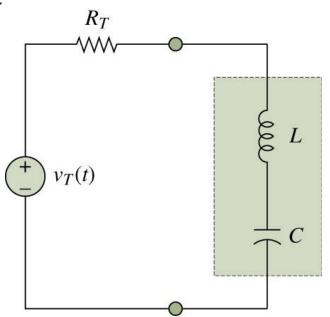
Lastly, let us return to an RLC circuit we have started with. Recall the corresponding differential equations:

$$LC\frac{d^2v_C}{dt^2} + R_TC\frac{dv_C}{dt} + v_C = v_T$$

$$\rightarrow \frac{1}{\omega_n^2} \frac{d^2 x(t)}{dt^2} + \frac{2\zeta}{\omega_n} \frac{dx(t)}{dt} + x(t) = K_S f(t)$$

Q: Given that $\zeta = 0$ creates a purely oscillatory response,

how to turn this circuit into an oscillator?





Thank you for your attention!



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