

Assignment 3: Fourier Series & Transforms of Continuous & Discrete Time Signals

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Book Questions

Problem 4.1

Consider the full-wave rectified sinusoid in Figure 1
(a) - Determine its spectrum $X_a(F)$.

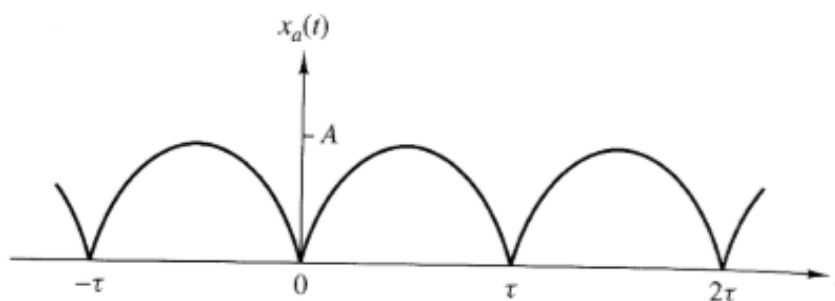


Figure 1: A full-wave rectified sinusoid.

We have a continuous, periodic function $x_a(t) = |A \sin(\frac{\pi}{\tau}t)|$, where $N = 2\tau$, $f = \frac{1}{2\tau}$. To determine the spectrum, we must find the Fourier Series of the signal. There are three methods to find this: The trigonometric, harmonic and exponential Continuous-Time Fourier Series. For this problem, we will use the exponential Fourier Series:

$$c_k = 1/T_p \int_0^{T_p} x(t) e^{-j\Omega_k t} dt, \Omega_k = 2\pi k F_p, F_p = 1/T_p$$

let $T_p = \tau$, or $\Omega_k = \frac{2\pi k}{\tau}$ I'm choosing this value because if you look at the graph, the pattern repeats every τ rather than every 2τ . If we had $\sin()$ instead

of $|\sin()|$ then our $T_p = 2\tau$. The math is easier with a period of τ since we can have $x(t) = A\sin(\frac{2\pi}{\tau}t)$ integrated from $0 \leq t \leq \tau$, and we can toss out the absolute value operator.

Before computing the Fourier Series coefficients, we need to determine a value k , which is the number of series (or the k th harmonic). The higher our k value, the more refined our spectrum will turn out. Let's start with $k = 0$ first:

$$\begin{aligned} c_o &= \frac{1}{\tau} \int_0^\tau A\sin(\frac{2\pi}{\tau}t)e^0 dt \\ c_o &= \frac{A}{\tau} \int_0^\tau \sin(\frac{2\pi}{\tau}t) dt \\ c_o &= \frac{A}{\tau} \left[\frac{-\tau}{2\pi} \cos(\frac{2\pi}{\tau}(\tau)) - \frac{-\tau}{2\pi} \cos(\frac{2\pi}{\tau}(0)) \right] \\ c_o &= \frac{A}{\tau} \frac{-\tau}{2\pi} [\cos(2\pi) - \cos(0)] \end{aligned}$$

Notice here: had we let $T_p = 2\tau$ instead of τ we would end up with the same result in the next step, since $\cos(2\pi) - \cos(0) = \cos(4\pi) - \cos(0) = 0$.

$$c_o = 0$$

The next coefficient is similar. We will still use the exponential CTFS method to find c_1 :

$$c_1 = \frac{1}{\tau} \int_0^\tau A\sin(\frac{2\pi}{\tau}t)e^{-j\frac{2\pi}{\tau}t} dt$$

In this case, I did use a calculator to solve the integral and simplified the expression to the following:

$$\begin{aligned} c_1 &= \frac{1}{\tau} \left(-\frac{j\tau}{2} \right) \\ c_1 &= -\frac{j}{2} \end{aligned}$$

For c_2 I'll use the trigonometric CTFS and find coefficients a_2 and b_2 , where $c_2 = \frac{1}{2}(a_2 - jb_2)$

$$\begin{aligned} a_2 &= \frac{2}{\tau} \int_0^\tau x(t)\cos(\Omega_k t) dt \\ a_2 &= \frac{2}{\tau} \int_0^\tau A\sin(\frac{2\pi}{\tau}t)\cos(\frac{2\pi}{\tau}(2)t) dt \\ a_2 &= \frac{2A}{\tau} \int_0^\tau \sin(\frac{2\pi}{\tau}t)\cos(\frac{4\pi}{\tau}t) dt \end{aligned}$$

I could use a calculator, but I need to brush up on integration by parts. Let $u = \sin(\frac{2\pi}{\tau}t)$ and $du = -\frac{\tau}{2\pi}\cos(\frac{2\pi}{\tau}t)dt$, then let $dv = \cos(\frac{4\pi}{\tau}t)dt$ and $v = \frac{-\tau}{4\pi}\sin(\frac{4\pi}{\tau}t)$

$$\int_0^\tau u dv = uv - \int_0^\tau v du$$

$$\int_0^\tau \sin(\frac{2\pi}{\tau}t)\cos(\frac{4\pi}{\tau}t)dt = \frac{-\tau}{4\pi}\sin(\frac{2\pi}{\tau}t)\sin(\frac{4\pi}{\tau}t) - \int_0^\tau \frac{-\tau}{4\pi}\sin(\frac{4\pi}{\tau}t)\frac{-\tau}{2\pi}\cos(\frac{2\pi}{\tau}t)$$

Ok, maybe this is way too much for the assignment. Next time, convert the sinusoidal functions into exponentials using Euler's rules. But long story short, you would do integration by parts again on the rhs of the equation. From there you subtract the integrals out of the equation and stuff... but the calculator shows:

$$a_2 = 0$$

For b_2 we will also use the calculator:

$$b_2 = \frac{2}{\tau} \int_0^\tau x(t)\sin(\Omega_k t)dt$$

$$b_2 = 0$$

Solve for c_2 , even though we know the answer:

$$c_2 = \frac{1}{2}(a_2 - jb_2) = 0$$

Now that we have stepped through the process we will just plug in our respective c_k values to the calculator. I'm using the TI-89 calculator and using the following command:

$$1/x*\text{integral}(A*\sin(2\pi/x*t)*e^{(-j2\pi/x*k*t)},t,0,x)$$

Where $x = \tau$, and I insert k for each c_k . Our Fourier Series coefficients are:

$$c_k = [0, 0, 0, 0, 0, \frac{jA}{2}, 0, \frac{-jA}{2}, 0, 0, 0, 0, 0]$$

This may seem really simple, but let's plot this and maybe this will look familiar:

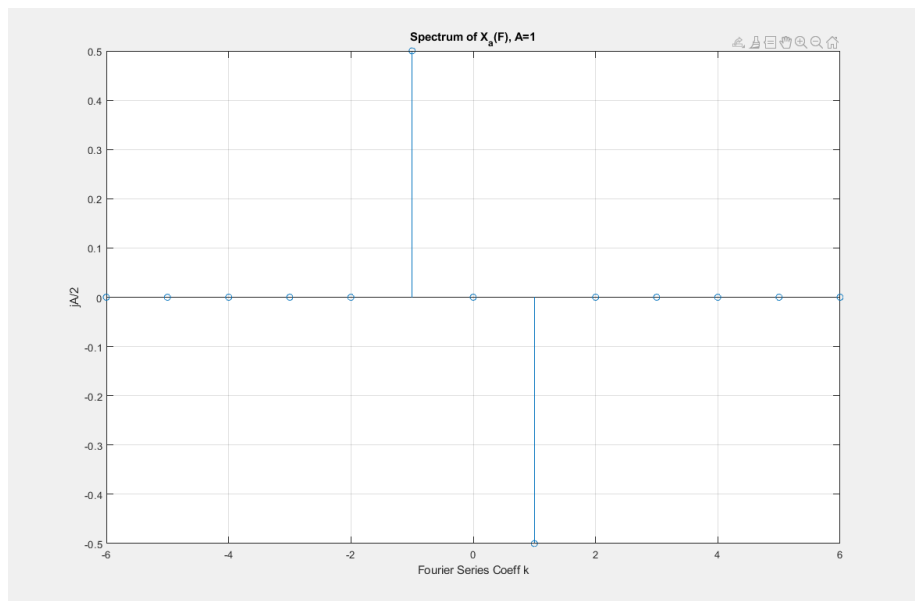


Figure 2: Spectrum plot of $X_a(F)$. Note that this looks like the Fourier transform of a sine wave.

(b) - Compute the power of the signal.

$$P_x = \frac{1}{T_p} \int_0^\tau |A \sin(\frac{2\pi}{\tau}t)|^2 dt$$

$$P_x = \frac{|A|^2}{T_p} \int_0^\tau |\sin(\frac{2\pi}{\tau}t)|^2 dt$$

$x(t)$ is real with positive magnitude for interval $0 \leq t \leq \tau$

$$P_x = \frac{A^2}{T_p} \int_0^\tau \sin^2(\frac{2\pi}{\tau}t) dt$$

Using integration by parts we eventually end up with:

$$P_x = \frac{A^2}{2}$$

(c) - Plot the power spectral density

To do this we take our Fourier Series Coefficients and create a plot of $|c_k|^2$. See Figure 3 for the plot, and the MATLAB section in the appendix for the code.

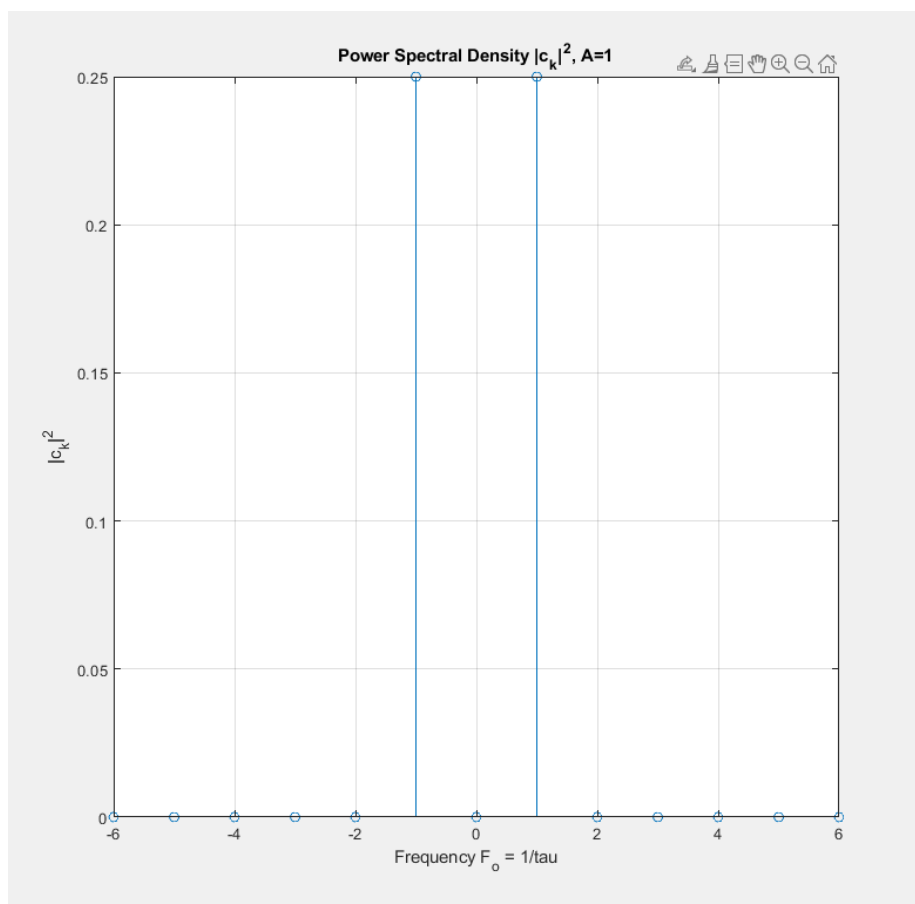


Figure 3: Power Spectral Density $|c_k|^2$.

(d) - Check the validity of Parseval's relation for this signal.
Parseval's relation is defined as follows:

$$P_x = \frac{1}{T_p} \int_{T_p} |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |c_k|^2$$

We obtained $P_x = \frac{A^2}{2}$, and the sum of our coefficients is:

$$\begin{aligned} \sum_{k=-\infty}^{\infty} |c_k|^2 &= \left| \frac{jA}{2} \right|^2 + \left| \frac{-jA}{2} \right|^2 \\ \sum_{k=-\infty}^{\infty} |c_k|^2 &= \frac{A^2}{2} = P_x \end{aligned}$$

Problem 4.2

Compute and sketch the magnitude and phase spectra for the following signals ($a > 0$):

$$(a) - x_a(t) = \begin{cases} Ae^{-at}, & \text{for } t \geq 0 \\ 0, & \text{for } t < 0. \end{cases}$$

This is an aperiodic, continuous signal. Remember that in order for there to be a Fourier transform of a signal $x_a(t)$, Dirichlet conditions must be satisfied. These are (1) the signal has a finite number of finite discontinuities, (2) the signal has a finite number of maxima and minima, and (3) the signal is absolutely integrable. This exponential equation satisfies those conditions.

To find the magnitude and phase spectra, we need to find the spectrum $X(F)$ of $x_a(t)$, which could very likely be complex in value such that:

$$X(F) = |X(F)|e^{j\Theta(F)}$$

Where $|X(F)|$ is the magnitude spectra and $\Theta(F)$ is the phase spectrum. Let's find $X(F)$:

$$\begin{aligned} X(F) &= \int_{-\infty}^{\infty} x(t)e^{-j2\pi Ft} dt \\ X(F) &= \int_0^{\infty} Ae^{-at}e^{-j2\pi Ft} dt \\ X(F) &= A \int_0^{\infty} e^{-j2\pi Ft - at} dt \\ X(F) &= A \int_0^{\infty} e^{(-j2\pi F - a)(t)} dt \\ X(F) &= A \left[\frac{1}{(-j2\pi F - a)} (e^{(-j2\pi F - a)(\infty)} - e^{(-j2\pi F - a)(0)}) \right] \end{aligned}$$

$$X(F) = A \left[\frac{-1}{(j2\pi F + a)} (0 - 1) \right]$$

$$X(F) = \frac{A}{(j2\pi F + a)}$$

$$X(F) = \frac{A}{(j2\pi F + a)} \frac{(-j2\pi F + a)}{(-j2\pi F + a)}$$

$$X(F) = \frac{A}{(2\pi F)^2 + a^2} (a - j2\pi F)$$

From here we can extract the magnitude and phase of the spectrum:

$$|X(F)| = \sqrt{real^2 + imag^2}$$

$$|X(F)| = \frac{A}{(2\pi F)^2 + a^2} \sqrt{(a)^2 + (2\pi F)^2}$$

$$|X(F)| = \frac{A}{(2\pi F)^2 + a^2} \sqrt{(a)^2 + (2\pi F)^2} \frac{\sqrt{(a)^2 + (2\pi F)^2}}{\sqrt{(a)^2 + (2\pi F)^2}}$$

$$|X(F)| = \frac{A}{\sqrt{(a)^2 + (2\pi F)^2}}$$

$$\Theta(F) = \tan^{-1}(imag/real)$$

$$\Theta(F) = -\tan^{-1}(-2\pi F/a)$$

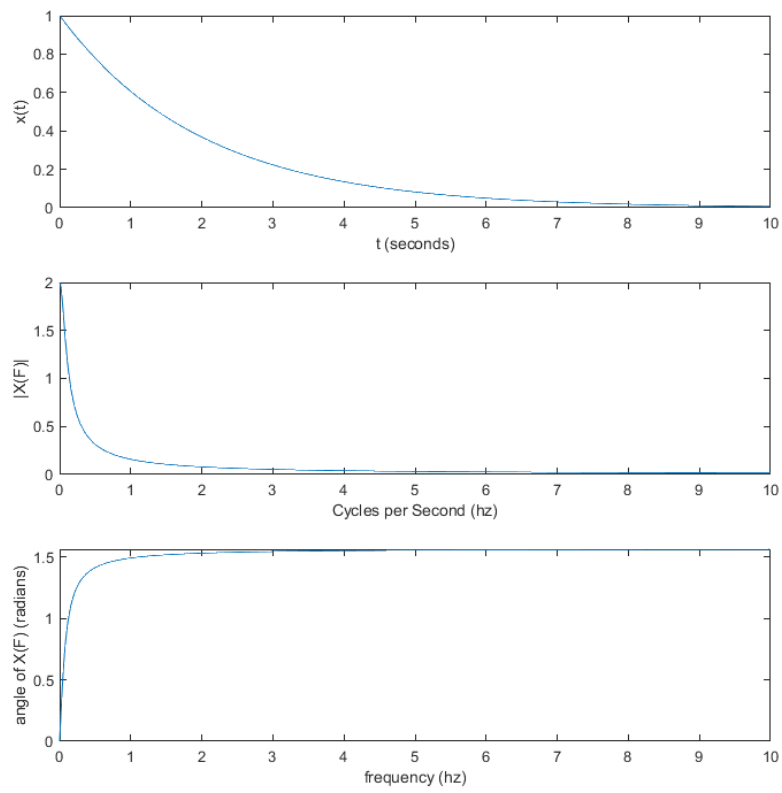


Figure 4: Plots of $x(t)$ and respective spectrum magnitude and phase plots.

$$(b) - x_a(t) = Ae^{-a|t|}$$

This is a decaying exponential function that is symmetric across the $x_a(t)$ axis. It satisfies the Dirichlet conditions just like the last problem. Because $Ae^{-a|t|}$ is symmetric about the axis, the spectrum can be represented as follows:

$$X(F) = \int_{-\infty}^0 Ae^{-at} e^{-j2\pi Ft} dt + \int_0^{\infty} Ae^{-at} e^{-j2\pi Ft} dt$$

$$X(F) = 2 \int_0^{\infty} Ae^{-at} e^{-j2\pi Ft} dt$$

$$X(F) = \frac{2A}{(j2\pi F + a)}$$

It's kinda convenient that this problem is similar to the last part. The magnitude and phase are as follows:

$$|X(F)| = \frac{2A}{\sqrt{(a)^2 + (2\pi F)^2}}$$

$$\Theta(F) = -\tan^{-1}(-2\pi F/a)$$

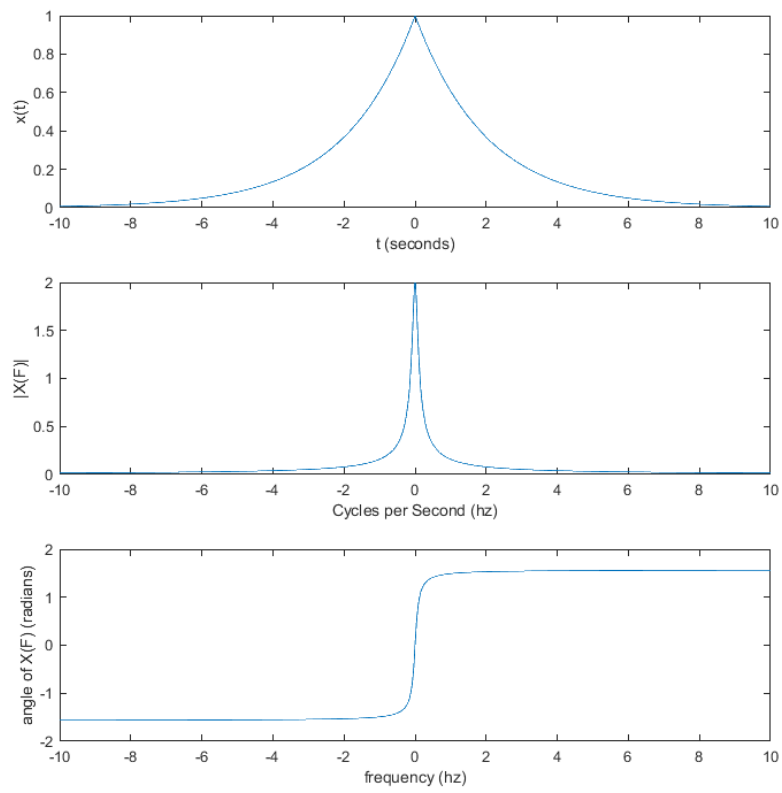


Figure 5: Plots of $x(t)$ and respective spectrum magnitude and phase plots.

Problem 4.4

Consider the following periodic signal:

$$x(n) = \dots, 1, 0, 1, 2, 3, 2, 1, 0, 1, \dots$$

(a) - Sketch the signal $x(n)$ and its magnitude and phase spectra.

Since we are dealing with a discrete signal, we can do this all in MATLAB. See the MATLAB section for the code used on this problem. Our plots are below in Figure 6:

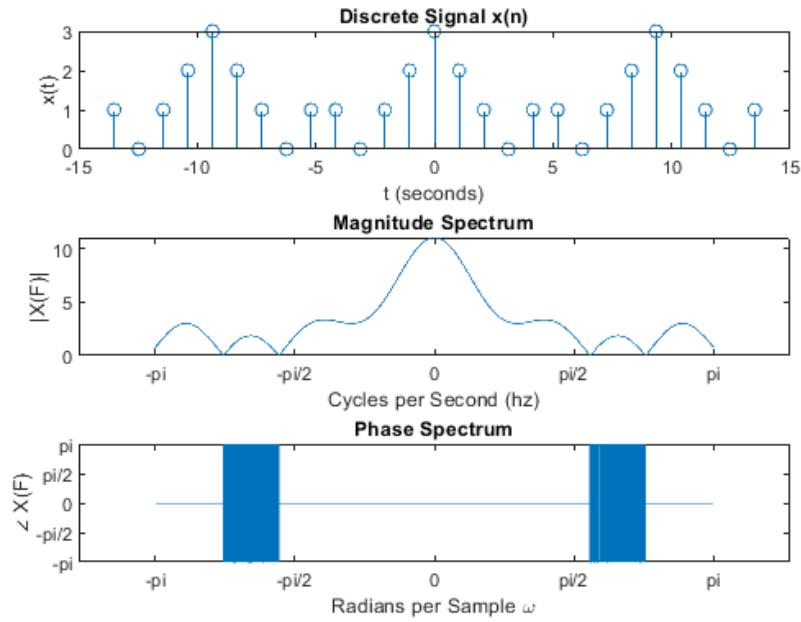


Figure 6: Plots of $x(n)$ and respective spectrum magnitude and phase plots.

(b) - Using the results in part (a), verify Parseval's relation by computing the power in the time and frequency domains.

Parseval's relation in discrete time is as follows (from eq 4.2.1 in the book):

$$\sum_{k=0}^{N-1} |c_k|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |x(n)|^2$$

In the MATLAB code below I found the series coefficients and found the power. I obtained an approximate value $P_x = 2.333$

Problem 4.6 (d)

Determine and sketch the magnitude and phase spectra of the following periodic signals:

(d) - $x(n) = \{\dots, -2, -1, 0, 1, 2, -2, -1, 0, 1, 2, \dots\}$

This is very similar to the last problem, so I created a MATLAB section specifically for this problem. Below in Figure

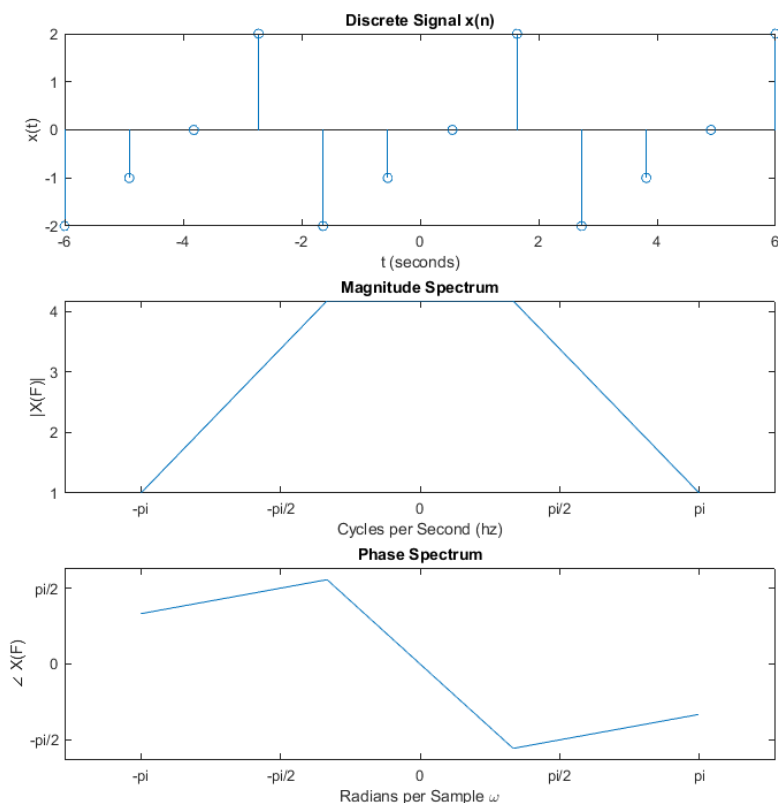


Figure 7: Plots of $x(n)$ and respective spectrum magnitude and phase plots.

Problem 4.9 (d)

Compute the Fourier transform of the following signals:

(d) - $x(n) = (\alpha^n \sin \omega_o n) u(n)$, $|\alpha| < 1$ Since $x(n)$ approaches 0 as $n = \infty$, and the signal is aperiodic, we can use the analysis equation given below:

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}$$

Where ω is our radians per sample ranging from $[-\pi, \pi]$. Note that this is the DFT instead of the DTFT. We can insert our function and get the following:

$$X(\omega) = \sum_{n=-\infty}^{\infty} (\alpha^n \sin \omega_o n) u(n) e^{-j\omega n}$$

$$X(\omega) = \sum_{n=0}^{\infty} (\alpha^n \sin \omega_o n) e^{-j\omega n}$$

It would be easiest to compute this in MATLAB and then plot $X(\omega)$ since this is a discrete Fourier transform. Plotting the results will yield a visual of the Fourier transform (see Figure 8). Another trick we can use is the convolution property, where multiplication in the time domain is convolution in the frequency domain. Let's divide $x(n)$ such that:

$$x_1(n) = \alpha^n u(n), |\alpha| < 1 \text{ and } x_2(n) = \sin(\omega_o n)$$

$$X_2(\omega) = \sum_{n=-\infty}^{\infty} \sin(\omega_o n) e^{-j\omega n}$$

There is an identity for this, but also look at the CTFT of $\sin(ax)$:

$$X_2(\omega) = \frac{\pi}{j} [\delta(\omega - \omega_o) + \delta(\omega + \omega_o)]$$

$$X_1(\omega) = \sum_{n=0}^{\infty} \alpha^n e^{-j\omega n}$$

$$X_1(\omega) = 1 + \alpha e^{-j\omega} + \alpha^2 e^{-j\omega 2} + \alpha^3 e^{-j\omega 3} + \dots + 0$$

$$X_1(\omega) = \frac{1}{1 - \alpha e^{-j\omega}} \text{ (see pg 291)}$$

Now we convolve X_0 with X_1 :

$$X(\omega) = X_0(\omega) * X_1(\omega)$$

$$X(\omega) = \frac{\pi}{j} \left[\frac{1}{1 - \alpha e^{-j(-\omega_o)}} + \frac{1}{1 - \alpha e^{-j(\omega_o)}} \right]$$

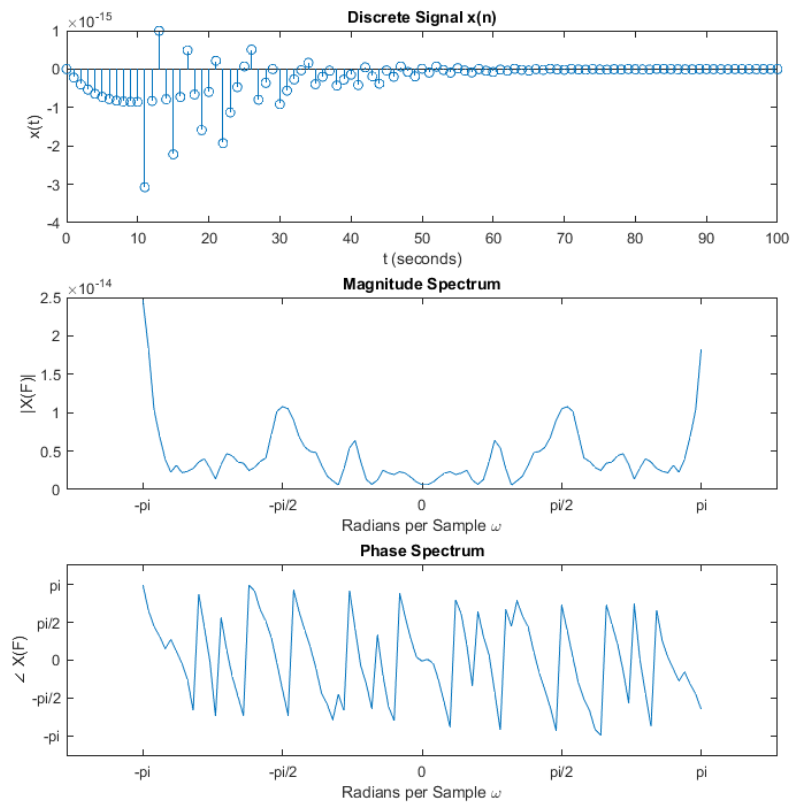


Figure 8: Plots of $x(n)$ and respective spectrum magnitude and phase plots.

Matlab Code - Problem 1

```

1 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
2 %% Problem 4.1
3 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
4 A = 1;
5 c_k = [0,0,0,0,0,A*1/2,0,A*(-1/2),0,0,0,0];
6 n = (-6:6);
7 stem(n,c_k);
8 grid on;
9 title('Spectrum of X_a(F), A=1');
10 xlabel('Fourier Series Coeff k');
11 ylabel('jA/2');
12
13 % part c - power spectral density
14 power_spectral = abs(c_k).^2;
15 figure(2);
16 stem(n,power_spectral);
17 grid on;
18 title('Power Spectral Density |c_k|^2, A=1');
19 xlabel('Frequency F_o = 1/tau');
20 ylabel('|c_k|^2');
21
22 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
23 %% Problem 4.2 - Aperiodic Fourier Txfrm
24 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
25 N = 10;
26 k = 0.01; % sets how refined the plot is
27 F = (0:k:N);
28 A = 1;
29 a = 0.5;
30 duration = 10;
31
32
33 % part A functions
34 t = (0:k:duration);
35 x = A*exp(-a.*t);
36 X = A ./ (a^2+(F.*(2*pi)).^2).^(0.5);
37 TH = -atan((-2*pi).*F./a);
38 % TH = -atan2((-2*pi).*F./a);
39
40 % part B functions
41 % t = (-duration:k:duration);
42 % F = (-N:k:N);
43 % x = A*exp(-a.*abs(t));

```



```

44 % X = A ./ (a^2+(F.*(2*pi)).^2).^0.5);
45 % TH = -atan((-2*pi).*F./a);
46
47 subplot(3,1,1)
48 plot(t,x);
49 xlabel('t (seconds)')
50 ylabel('x(t)')
51
52
53 subplot(3,1,2)
54 plot(F,X);
55 xlabel('Cycles per Second (hz)')
56 ylabel('|X(F)|')
57
58 subplot(3,1,3)
59 plot(F,TH);
60 xlabel('frequency (hz)')
61 ylabel('angle of X(F) (radians)')
62 set(gca,'YTick',0:pi/8:pi/2)
63 set(gca,'YTickLabel',{'0','pi/8','pi/4','3pi/8','pi/2'})
64
65 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
66 %% Problem 4.4
67 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
68
69 % -----
70 % part a – Magnitude and Phase Spectrum
71 % -----
72 x = [1,0,1,2,3,2,1,0,1];
73 k = 10000; % number of points for refinement
74 n = linspace(-length(x)/2,length(x)/2,length(x)); % the
    length() command is useful for dynamic code
75 w = linspace(-pi,pi,length(n));
76 % find the spectrum by calculating the dtft of periodic x
77 X = dtft(x,n,w);
78 N = (0:length(X)-1);
79 W = linspace(-pi,pi,length(X)); % use for even spacing of
    n items
80 Xmag = abs(X);
81 Xang = angle(X);
82
83 % -----
84 % part b – Power and Parseval's
85 % -----
86 % our Spectrum X is a sum of our coefficients c_k
87 % finding c_k

```

```

88 N_c = length(x);
89 for k=1:N_c
90     ck(k) = 0;
91     for m=1:N_c
92         ck(k) = ck(k) + 1/N_c * x(m)*exp((-1i*2*pi*k/N_c)
          *n(m));
93     end
94 end
95 P_ck = sum(abs(ck).^2);
96 P_x = sum(x.*conj(x))/length(x);
97
98
99
100 % -----
101 % Plots of part a
102 % -----
103
104
105
106
107 % plotting repeated x
108 P = 3; % The number of times we repeat the signal
109 x = x' * ones(1,P);
110 x = x(:); % long column vector
111 x = x'; % transpose to long row vector
112 n = linspace(-length(x)/2,length(x)/2,length(x));
113 W = linspace(-pi,pi,length(X));
114
115
116 subplot(3,1,1)
117 stem(n,x);
118 title('Discrete Signal x(n)')
119 xlabel('t (seconds)')
120 ylabel('x(t)')
121
122
123 subplot(3,1,2)
124 plot(W,Xmag);
125 xlabel('Cycles per Second (hz)')
126 title('Magnitude Spectrum')
127 ylabel('|X(F)|')
128 set(gca,'XTick',-pi:pi/2:pi)
129 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
130
131
132 subplot(3,1,3)

```

```

133 plot(W,Xang);
134 title('Phase Spectrum')
135 xlabel('Radians per Sample \omega')
136 ylabel('\angle X(F)')
137 set(gca,'YTick',-pi:pi/2:pi)
138 set(gca,'YTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
139 set(gca,'XTick',-pi:pi/2:pi)
140 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
141
142
143 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
144 %% Problem 4.6 (d)
145 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
146 % -----
147 % part a – Magnitude and Phase Spectrum
148 % -----
149 x = [-2,-1,0,2];
150 k = 10000; % number of points for refinement
151 n = linspace(-length(x)/2,length(x)/2,length(x)); % the
    length() command is useful for dynamic code
152 % n = (-5:5);
153 % n = n-length(x);
154 w = linspace(-pi,pi,length(n));
155 % find the spectrum by calculating the dtft of periodic x
156 X = dtft(x,n,w);
157 N = (0:length(X)-1);
158 W = linspace(-pi,pi,length(X)); % use for even spacing of
    n items
159 Xmag = abs(X);
160 Xang = angle(X);
161
162 % -----
163 % part b – Power and Parseval's
164 % -----
165 % our Spectrum X is a sum of our coefficients c_k
166 % P_X = sum(Xmag.^2);
167 % P_x = sum(abs(x).^2)/length(x);
168
169
170
171 % -----
172 % Plots of part a
173 % -----
174
175
176

```

```

177
178 % plotting repeated x
179 P = 3; % The number of times we repeat the signal
180 x = x' * ones(1,P);
181 x = x(:); % long column vector
182 x = x'; % transpose to long row vector
183 n = linspace(-length(x)/2,length(x)/2,length(x));
184 W = linspace(-pi,pi,length(X));
185
186
187 subplot(3,1,1)
188 stem(n,x);
189 title('Discrete Signal x(n)')
190 xlabel('t (seconds)')
191 ylabel('x(t)')
192
193
194 subplot(3,1,2)
195 plot(W,Xmag);
196 xlabel('Cycles per Second (hz)')
197 title('Magnitude Spectrum')
198 ylabel('|X(F)|')
199 set(gca,'XTick',-pi:pi/2:pi)
200 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
201
202
203 subplot(3,1,3)
204 plot(W,Xang);
205 title('Phase Spectrum')
206 xlabel('Radians per Sample \omega')
207 ylabel('\angle X(F)')
208 set(gca,'YTick',-pi:pi/2:pi)
209 set(gca,'YTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
210 set(gca,'XTick',-pi:pi/2:pi)
211 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
212
213
214
215 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
216 %% Problem 4.9 (d)
217 %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
218 a = 0.9;
219 n=(0:100); % remember that x(n) has a unit step function
           component, hence 0
220 f = 1; % cycles per sample
221 w = 2*pi*f; % radians per sample conversion

```

```

222 x = a.^n.*sin(w.*n);
223 X = DFT(x,length(n));
224 W = linspace(-pi,pi,length(X));
225
226 subplot(3,1,1)
227 stem(n,x);
228 title('Discrete Signal x(n)')
229 xlabel('t (seconds)')
230 ylabel('x(t)')
231
232
233 subplot(3,1,2)
234 plot(W,abs(X));
235 xlabel('Radians per Sample \omega')
236 title('Magnitude Spectrum')
237 ylabel('|X(F)|')
238 set(gca,'XTick',-pi:pi/2:pi)
239 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
240
241
242 subplot(3,1,3)
243 plot(W,angle(X));
244 title('Phase Spectrum')
245 xlabel('Radians per Sample \omega')
246 ylabel('\angle X(F)')
247 set(gca,'YTick',-pi:pi/2:pi)
248 set(gca,'YTickLabel',{'-pi','-pi/2','0','pi/2','pi'})
249 set(gca,'XTick',-pi:pi/2:pi)
250 set(gca,'XTickLabel',{'-pi','-pi/2','0','pi/2','pi'})

```