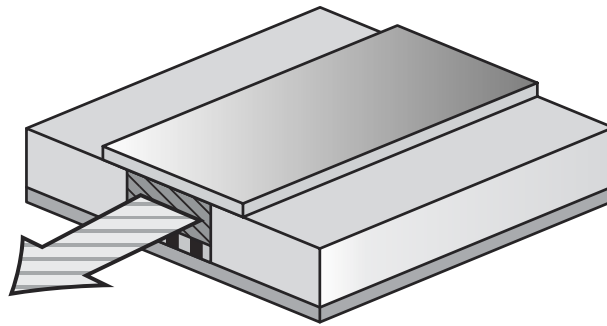


Optoelectronics and Lasers
Homework II



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Asymmetric Planar Dielectric Waveguide

In the present report we write the electromagnetic analysis for an Asymmetric Dielectric Planar Waveguide (ADPWG) in order to find the waveguided modes. Modes which are transverse electric (TE) and transverse magnetic (TM) with regards to their electric and magnetic components. Firstly, we follow the analysis of the waveguide depicted in Fig.1 and we end up to two equations, one for TE and one for TM, that shall be solved to find the effective transmission parameter β of every mode. Our notation and analysis is slightly different from the corresponding analysis in the Pochi Yeh's book, but is complete and fully understandable. Secondly, we implement our results on a computational example for specific values. Thirdly, in contrast to the Symmetric DPWG in ADPWG there is a cut-off wavelength for the TE₀ mode. Therefore, we make a thickness sweep over the waveguiding layer and we present the way that effective refractive index changes with respect to the specified thickness. At the end of the report, we prove the orthogonality between the TE₀ eigenmodes.

(a) Electromagnetic analysis

We write Maxwell's equations in a place without sources, or sources are at infinity, and assuming harmonic oscillation with $\exp(+j\omega t)$ in frequency domain as

$$\nabla \times \mathbf{E} = -j\omega \mathbf{B} \quad (1a)$$

$$\nabla \times \mathbf{B} = j\omega \mathbf{D} \quad (1b)$$

$$\nabla \cdot \mathbf{D} = 0 \quad (1c)$$

$$\nabla \cdot \mathbf{B} = 0. \quad (1d)$$

The above equations are accompanied with the relations which connects the electric field with dielectric displacement and magnetic field with magnetic induction

$$\mathbf{D} = \varepsilon \mathbf{E} \quad (2a)$$

$$\mathbf{B} = \mu \mathbf{H}. \quad (2b)$$

Also, it is known that dielectric constant $\varepsilon = \varepsilon_r \varepsilon_0$ is related to refractive index as $\varepsilon_r = n^2$. Regarding the magnetic constant $\mu = \mu_r \mu_0$, in our problem we can assume that is constant and $\mu_r = 1$ and it is actually consistent with the experimental results.

Taking the curl of the eq.(1a-b) and using the equations (2b) one shall create the wave equation in vector form (also known as Helmholtz equation) for the electric and magnetic field as

$$\nabla^2 \times \mathbf{E} - \omega^2 \mu \varepsilon \mathbf{E} = 0 \quad (3a)$$

$$\nabla^2 \times \mathbf{H} - \omega^2 \mu \varepsilon \mathbf{H} = 0, \quad (3b)$$

where $\nabla^2 = \partial^2/\partial x^2 + \partial^2/\partial y^2 + \partial^2/\partial z^2$ the Laplacian operator. We shall solve these equations for the case of a planar dielectric waveguide with different refractive indices at the substrate and cladding layer, see Fig.1 .

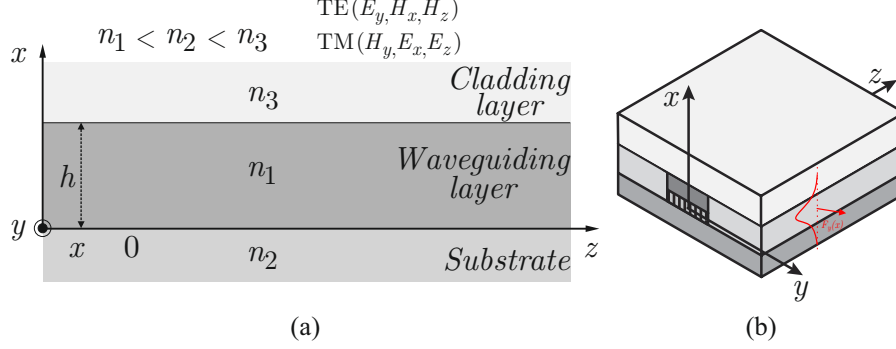


Figure 1: (a) Two-dimensional schematic of the Asymmetric Planar Dielectric Waveguide supporting TE and TM modes with transmission along z axis. (b) Three-dimensional representation of the APDWG with a typical feature of the localized TE_0 mode.

Transverse Electric modes (TE)

In this subsection we examine the case in which the electric field has only one component and especially along axis y , i.e. $\mathbf{E} = (0, E_y, 0)$ with $E_y \neq 0$. However, the afore mentioned component is a function of x , $E_y = E_y(x)$, and that because there is a variation of the refractive index along this axis. Hence, from eq.(3a) we write

$$\nabla^2 E_{yi}(x, z) + n_i^2 k_0^2 E_{yi}(x, z) = 0, \quad i = 1, 2, 3 \quad (4)$$

$$k_0 = \omega \sqrt{\mu_0 \epsilon_0}.$$

We follow the well known method to assume solutions in the form of

$$E_{yi}(x, y) = E_i(x) \exp(-j\beta z), \quad (5)$$

where total solution is a product of two coefficients which are functions of x and z , respectively. Specifically, an exponential with the transmission constant β which describes the change of phase while the field is transmitted along z axis and a function $E_i(x)$ (generally $E_i(x, y)$) which depicts the modes' profile.

We substitute eq.(5) to eq.(4) and we get

$$\frac{\partial^2 E_i(x)}{\partial x^2} + (n_i^2 k_0^2 - \beta^2) E_i(x) = 0, \quad i = 1, 2, 3. \quad (6)$$

At this point we shall assume that the field under discussion is retained mainly in the waveguiding layer with the mode profile we are about to reveal, while outside of this layer is decaying exponentially, i.e. in substrate and cladding layer. Hence, we write $E_{yi}(x)$ for every region along x axis

$$\text{WG layer } E_1(x) = A_1 \cos(\kappa x - \phi), \quad \kappa = \sqrt{n_1^2 k_0^2 - \beta^2} \quad (7a)$$

$$\text{Substrate } E_2(x) = A_2 \exp(\gamma x), \quad \gamma = \sqrt{\beta^2 - n_2^2 k_0^2} \quad (7b)$$

$$\text{Cladding } E_3(x) = A_3 \exp(-\delta(x - h)), \quad \delta = \sqrt{\beta^2 - n_3^2 k_0^2}. \quad (7c)$$

To determine the coefficients A_1, A_2, A_3, ϕ but also κ, γ, δ one needs to implement the boundary conditions. We demand the continuity of the tangential components of the electric and magnetic field at the borderlines $x = 0$ and $x = h$. From the eq.(1a) we get

$$\begin{aligned} \nabla \times \mathbf{E} &= -j\omega\mu (\hat{x}H_x + \hat{z}H_z) \implies \\ H_x &= -\frac{\beta}{\omega\mu} E_y \end{aligned} \quad (8a)$$

$$H_z = -j \frac{1}{\omega\mu} \frac{\partial E_y}{\partial x}. \quad (8b)$$

One can easily see that $H_x \propto E_y$ and $H_z \propto \partial E_y / \partial x$, thus the boundary conditions demand the continuity of E_y and $\partial E / \partial x$ at $x = 0, h$.

Continuity at $x = 0$.

$$\left. \begin{aligned} A_2 &= A_1 \cos(\phi) \\ \gamma A_2 &= \kappa A_1 \sin(\phi) \end{aligned} \right\} \implies \tan(\phi) = \frac{\gamma}{\kappa}. \quad (9)$$

Continuity at $x = h$.

$$\left. \begin{aligned} A_3 &= A_1 \cos(\kappa \cdot h - \phi) \\ \delta A_3 &= \kappa A_1 \sin(\kappa \cdot h - \phi) \end{aligned} \right\} \implies \tan(\kappa \cdot h - \phi) = \frac{\delta}{\kappa}. \quad (10)$$

Combining the two equations above with the identity $\tan(a \pm b) = \frac{\tan(a) \pm \tan(b)}{1 \mp \tan(a) \tan(b)}$ we get

$$\tan(\kappa \cdot h) = \frac{\kappa(\gamma + \delta)}{\kappa^2 - \gamma\delta}. \quad (11)$$

Given that κ, γ, δ are functions of β , it is obvious that eq.(11) is an equation that engages transmission constant β , angular frequency ω and the waveguide characteristics (thickness h , and refractive indices). Also, eq.(11) is identical to equation 11.2-5 that we were asked to prove.

Transverse Magnetic modes (TM)

In this subsection we shall follow almost the same steps for the magnetic field and the eq.(3b), but we shall mention any differences when we meet them. The scheme now is that we magnetic field has only one component and it is parallel to axis y . Hence, we write eq.(3b) in the form

$$\begin{aligned} \nabla^2 H_{yi}(x, z) + n_i^2 k_0^2 H_{yi}(x, z) &= 0, \quad i = 1, 2, 3 \\ k_0 &= \omega \sqrt{\mu_0 \varepsilon_0}. \end{aligned} \quad (12)$$

As in the previous case, we assume a product of two coefficients to be a solution in the form of

$$H_{yi}(x, y) = H_i(x) \exp(-j\beta z). \quad (13)$$

Hence, eq.(12) takes the form

$$\frac{\partial^2 H_i(x)}{\partial x^2} + (n_i^2 k_0^2 - \beta^2) H_i(x) = 0, \quad i = 1, 2, 3. \quad (14)$$

For the same reasons we mentioned in the previous section, we want the field to be guided in the the waveguiding layer. Thus, we write the solution for each layer as

$$\text{WG layer } H_1(x) = A_1 \cos(\kappa x - \phi), \quad \kappa = \sqrt{n_1^2 k_0^2 - \beta^2} \quad (15a)$$

$$\text{Substrate } H_2(x) = A_2 \exp(\gamma x), \quad \gamma = \sqrt{\beta^2 - n_2^2 k_0^2} \quad (15b)$$

$$\text{Cladding } H_3(x) = A_3 \exp(-\delta(x - h)), \quad \delta = \sqrt{\beta^2 - n_3^2 k_0^2}, \quad (15c)$$

where A_1, A_2, A_3 , are different from the former ones, but unless someone wants to calculate the profiles of the modes these amplitudes will not concern us. With regards to boundary conditions, the tangential components of electric E_z and magnetic field H_y must be continuous at $x = 0$ and $x = h$. From eq.(1b) we get

$$\nabla \times \mathbf{B} = -j\omega\varepsilon(\hat{x}E_x + \hat{z}E_z) \implies \quad (16a)$$

$$E_{xi}(x) = \frac{\beta}{\omega\varepsilon_0 n_i^2} H_y(x) \quad (16b)$$

$$E_{zi}(x) = j \frac{1}{\omega\varepsilon_0 n_i^2} \frac{\partial H_y}{\partial x}. \quad (16c)$$

Therefore, we demand the continuity of $E_{zi} \propto \frac{1}{n_i^2} \partial H_y / \partial x$ and $H \propto H_y$ at points $x = 0$ and $x = h$.

Continuity at $x = 0$.

$$(E_{z2}=E_{z1})|_{x=0} \implies \left. \begin{aligned} A_2 &= A_1 \cos(\phi) \\ \frac{\gamma}{n_2^2} A_2 &= \frac{\kappa}{n_1^2} A_1 \sin(\phi) \end{aligned} \right\} \implies \tan(\phi) = \frac{\gamma'}{\kappa}. \quad (17)$$

Continuity at $x = h$.

$$(E_{z3}=E_{z1})|_{x=h} \Rightarrow \left. \begin{aligned} A_3 &= A_1 \cos(k \cdot h - \phi) \\ \frac{\delta}{n_3^2} A_2 &= \frac{\kappa}{n_1^2} A_1 \sin(\kappa \cdot h - \phi) \end{aligned} \right\} \Rightarrow \tan(\kappa \cdot h - \phi) = \frac{\delta'}{\kappa}, \quad (18)$$

where $\gamma' = \gamma(n_1^2/n_2^2)$ and $\delta' = \delta(n_1^2/n_3^2)$. Now, we use the same identity for $\tan(x)$ and we get

$$\tan(\kappa \cdot h) = \frac{\kappa(\gamma' + \delta')}{\kappa^2 - \gamma'\delta'}. \quad (19)$$

The solutions of the above equation returns the effective transmission parameters for every TM mode. Also, we have ended up to the relation we had to prove.

(b)

Next we shall solve the transcendental equation of the TE modes, i.e. (11), for an APDWG with characteristics (see Fig.(1)):

- wavelength $\lambda_0 = 350$ nm
- thickness $h = 500$ nm
- refractive indices $n_1 = 2.6$, $n_2 = 2.4$, $n_3 = 1$

We solved this equation graphically and we reveal the results in Fig.(2) and in Table 1 for better precision. Specifically, we calculated the effective transmission constant β_{eff} and also the effective refractive index n_{eff} . Moreover, we found out that for the given parameters we have three TE modes which one feels different transmission constant. Obviously, we see that $\beta_{TE_0} > \beta_{TE_1} > \beta_{TE_2}$.

Table 1: Gathered results for TE modes of APDWG. This table contains all the information needed to answer (b) and (c).

Figure 2(a)	Transmission constant (rad/ μm)	Effective Refractive index	Minimum thickness (nm)
TE ₀	46.569	2.5829	64
TE ₁	45.714	2.5301	242
TE ₂	43.953	2.4442	413

For the sake sake of completeness, in Fig.3 we present the profiles of the three guided modes. More specifically, we see three different types of profiles with $(i + 1)$, $i = 0, 1, 2$ nodes and the way they decay outside of the waveguiding

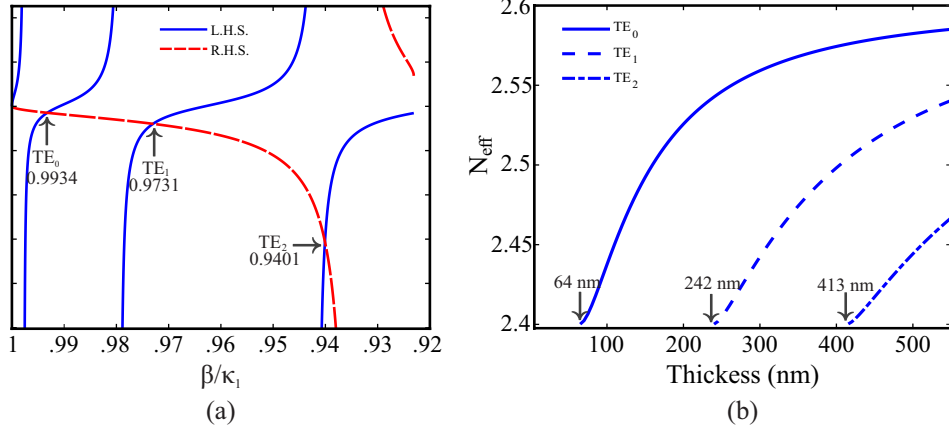


Figure 2: (a) Graphic solution for the eq.(11), where with blue (solid) line is the L.H.S. and with red (dashed) line is the R.H.S. In the x-axis is the normalized quantity β/κ_1 . Also, the points of intersection are labeled inside the figure. (b) Representation of the effective refractive index for every guided mode versus different thickness h . In contrast to the SPDWG, TE_0 in APDWG generates for thickness greater than 64 nm.

layer. The higher the mode is the slower it decays outside the guiding layer.

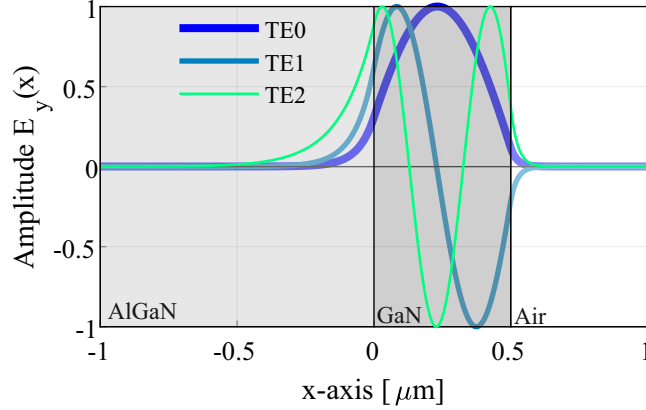


Figure 3: Mode profiles for the asymmetric planar/slab dielectric waveguide. Modes depicted: TE_0 blue(bold) line, TE_1 light-blue (solid) line and TE_2 cyan (thin) line. Lower ordered modes are more localized in the waveguiding material, i.e. the material with the higher refractive index.

(c)

Although we calculated, numerically the minimum thickness needed for at least one mode at this point we find the lower thickness via the analytical

way. As we know, effective transmission constant β_{eff} satisfies the inequality

$$\beta_2 \leq \beta_{\text{eff}} \leq \beta_1, \quad (20)$$

where $\beta_{1,2}$ the transmission constants in materials with the two higher refractive indices, and a mode can be guided at the limit where $\beta_{\text{eff}} = \beta_2$. This is the lower limit, because for $\beta_{\text{eff}} < \beta_1$ we have an imaginary part which forces the mode to decay exponentially.

Hence,

$$\begin{aligned} \beta = \beta_2 \Rightarrow \tan \left(k_0 h \sqrt{n_1^2 - n_2^2} \right) &= \frac{k_0^2 \sqrt{n_1^2 - n_2^2} \sqrt{n_2^2 - n_3^2}}{k_0^2 \sqrt{n_1^2 - n_2^2}} \Rightarrow \\ h &= \frac{1}{k_0 \sqrt{n_1^2 - n_2^2}} \left(m\pi + \tan^{-1} \left(\frac{n_2^2 - n_3^2}{n_1^2 - n_2^2} \right) \right). \end{aligned} \quad (21)$$

Therefore, for our parameters we get

$$h = 63.58 \text{ nm}. \quad (22)$$

A result that verifies the thickness we found through the computational way, see Table 1.

(d)

Last but not least we have to prove the orthogonality between the eigenmodes of the above mentioned eigenvalue problem. We begin from the general form of eq.(4) for two different TE modes we write

$$\nabla_t^2 \mathbf{E}_1 + (k_0^2 n^2(x, y) - \beta_1^2) \mathbf{E}_1 = 0 \quad (23a)$$

$$\nabla_t^2 \mathbf{E}_2 + (k_0^2 n^2(x, y) - \beta_2^2) \mathbf{E}_2 = 0, \quad (23b)$$

where $\beta_{1,2}$ the transmission constant of each mode and $\nabla_t^2 = \partial^2/\partial x^2 + \partial^2/\partial y^2$ the Laplacian operator on the transverse plane. We have to mention that field vectors refer to the transverse plane. We multiply the above equations with \mathbf{E}_2 and \mathbf{E}_1 respectively and we subtract them to get

$$\begin{aligned} \mathbf{E}_2 \nabla_t^2 \mathbf{E}_1 - \mathbf{E}_1 \nabla_t^2 \mathbf{E}_2 &= (\beta_1^2 - \beta_2^2) \mathbf{E}_1 \cdot \mathbf{E}_2 \Rightarrow \\ \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \mathbf{E}_2 \nabla_t^2 \mathbf{E}_1 - \mathbf{E}_1 \nabla_t^2 \mathbf{E}_2 \, dx dy &= (\beta_1^2 - \beta_2^2) \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \mathbf{E}_1 \cdot \mathbf{E}_2 \, dx dy \xrightarrow{*} \\ \lim_{C \rightarrow \infty} \oint_C (\mathbf{E}_2 \cdot \nabla_t) \mathbf{E}_1 - (\mathbf{E}_1 \cdot \nabla_t \mathbf{E}_2) \, d\bar{\ell} &= (\beta_1^2 - \beta_2^2) \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \mathbf{E}_1 \cdot \mathbf{E}_2 \, dx dy \rightarrow \\ (\beta_1^2 - \beta_2^2) \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \mathbf{E}_1 \cdot \mathbf{E}_2 \, dx dy &= 0. \end{aligned} \quad (24)$$

At the last step we used the fact that the fields are localized inside the waveguide layer, as we shown above, and they decays exponentially at infinity. Thus the contour integral is equal to zero. Also, at \star we used the relation below accompanied with Gauss' theorem,

$$\mathbf{E}_2 \nabla_t^2 \mathbf{E}_1 - \mathbf{E}_1 \nabla_t^2 \mathbf{E}_2 = \nabla_t ((\mathbf{E}_1 \cdot \nabla_t) \mathbf{E}_2 - (\mathbf{E}_2 \cdot \nabla_t) \mathbf{E}_1), \quad (25)$$

because as we can see from eq.(1b)

$$\nabla_t \cdot \mathbf{E} = \nabla \cdot \nabla \times \mathbf{H} = 0, \quad (26)$$

because in TE modes electric field \mathbf{E} does not have a componet along z axis. Gathering all the above information we shall conclude that eq.(24) leads to

$$\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \mathbf{E}_1 \cdot \mathbf{E}_2 \, dx \, dy = 0, \quad \text{when } \beta_1 \neq \beta_2. \quad (27)$$

Consequently, two different eigenmodes are orthogonal to each other and, thus, if we normalize them they can form an orthonormal basis in the Hilbert space. Obviously, if $\beta_1 = \beta_2$ we get the zero result from the parenthesis outside the integral, while the integral returns a finite value.

Code

In this section we attach the MATLAB code that has been used to produce the above mentioned results. For comments, mistakes, clarifications etc. please send a message to dakisfilippos@gmail.com.

```

1 % Filippas Tzimkas-Dakis UoC October 2020
2 % Homework 2 ----- Optoelectronics and Lasers
3 % This is a script to calculate the waveguided eigenmodes TM of
  a field in
4 % a dielectric slab waveguide.
5 % Also, this script calculates all the transmission constants \
  beta for
6 % specific characteristics (refractive indexes n1,n2,n3, width/
  thickness
7 %h, frequency or wave length f,\lambda).
8 %% Solving for TE modes
9 close all
10 clear all
11 clc
12 c = 3*10^8; % speed of light in vacuum
13 n1 = 2.6; % refractive index of the waveguide layer / core
14 n2 = 2.4; % refractive inde of the substrate
15 n3 = 1; % refractive inde of the cladding layer
16
17 d = 500*1e-9; % thickness of the eaveguide layer (
  meters)

```

```

18 lamda0 = 350*1e-9; % wavelength in vacuum
19 k0 = 2*pi/lamda0; % wave-number in vacuum
20 k1 = n1 *k0; % wave-number in core
21 k2 = n2 *k0; % wave-number in substrate
22 k3 = n3 *k0; % wave-number in cladding layer
23
24 xi = (n2^2-n3^2)/(n1^2-n2^2); % APDWG "asymmetry" factor (0==
    SPDWG)
25 V = k0*d/2*sqrt( n1^2-n2^2 ); % V-number, same definition as
    for SPDWG
26 VasosTE = atan( sqrt(xi) )/2; % V-number "asymmetry-offset" for
    TE-modes
27 TE_modes = ceil( (V-VasosTE)/(pi/2) ); % Number of modes
    expected
28
29 %===== The waveguide looks like this =====
30 %| | y?
31 %| n3 , k3 |
32 % ----- x?--> z
33 %| n1, k1 , thickness d |
34 %| |
35 % -----
36 %| n2 , k2 |
37 %| |
38 %=====
39
40 b = linspace(k1,max(k2,k3),2000); % creates the \beta vector
41 % this loop calculates only the cases where the R.H.S. is real
42
43 kapa = sqrt(k1^2 - b.^2); % variable of the Charecteristic
    Equation
44 gama = sqrt(b.^2 - k2^2); % variabla of the Ch.Eq.
45 delta = sqrt(b.^2 - k3^2); % variabla of the Ch.Eq.
46
47 % L.H.S of equation (15)
48 f1 = tan(kapa*d);
49 % R.H.S of equation (15)
50 f2 = (kapa.*(gama + delta))./(kapa.^2 - delta.*gama);
51 %% Finding the TE modes
52 % Plots the two parts of the Ch.Eq. in order to find the
    specific points of
53 % intersection
54 figure
55 hold on
56 plot(b/k1,f1/max(f2),'Color','blue','Linewidth',1.5) % L.H.S.
57 plot(b/k1,f2/max(f2),'Color','red','Linewidth',1.5) % R.H.S.
58 xlabel('\beta /\kappa_1')
59 ylabel('A.U.')
60 legend('L.H.S.', 'R.H.S.')
61 title('TE modes')
62 set(gca, 'XDir','reverse')
63 %%
64 % solver's options
65 options = optimset('Display','off');

```

```

66 % Characteristic equation
67 fun1 = @(t)tan(sqrt(k1^2 - t^2)*d) - (sqrt(k1^2 - t^2)*...
68     (sqrt(t^2 - k2^2) + sqrt(t^2 - k3^2)))/(k1^2 - t^2 - sqrt(t
    ^2 - k3^2)*sqrt(t^2 - k2^2));
69 neff = NaN*ones(1,TE_modes);
70 % Calculates the infinities of Tan(kappa*d)
71 for n = 0:TE_modes-1
72     k(n+1) = sqrt(k1^2 - ((pi/2/d)^2)*(2*n + 1)^2);
73     neff(n+1) = fsolve(fun1,k(n+1)*(0.99999),options)/k0;
74 end
75 % Prints the results. Effective Refr. Index
76 % and transmission constant
77 fprintf('\nThe results from the above figure are \n \n')
78 for i = 1:TE_modes
79     fprintf('***TE%d n_eff = %6.4f    beta = %6.2f (rad/um)***
    \n',...
80         i-1,neff(i),neff(i)*k0*10^-6);
81 end
82 fprintf('\n')
83
84 %% Calculates N_eff of every guided mode
85 di = (550:-5:1)*1e-9; % span of thickness to be tested
86 xi = (n2^2-n3^2)/(n1^2-n2^2); % APDWG "asymmetry" factor (0==
    SPDWG)
87 V = k0*max(di)/2*sqrt(n1^2-n2^2); % V-number, same
    definition as for SPDWG
88 VasosTE = atan(sqrt(xi))/2; % V-number "asymmetry-offset"
    for TE-modes
89 TE_modes = ceil((V-VasosTE)/(pi/2)); % Number of modes
    expected
90 out = NaN*ones(TE_modes,length(di)); % initialization
91
92 for u = 1:length(di)
93     d = di(u); % thickness tested
94     V = k0*d/2*sqrt(n1^2-n2^2); % V-number, same definition
    as for SPDWG
95     TE_modes = ceil((V-VasosTE)/(pi/2)); % Number of modes
    expected
96     % solver's options
97     options = optimset('Display','off');
98     % Characteristic equation to be solved for any d
99     fun1 = @(t)tan(sqrt(k1^2 - t^2)*d*k1_) - (sqrt(k1^2 - t^2)
    *...
100         (sqrt(t^2 - k2^2) + sqrt(t^2 - k3^2)))/(k1^2 - t^2 -
    sqrt(t^2 - k3^2)*sqrt(t^2 - k2^2));
101     % Calculate the infinities of Tan(kappa * d)
102     for n = 0:TE_modes-1
103         k(n+1) = sqrt(k1_^2 - ((pi/2/d)^2)*(2*n + 1)^2)/k1_;
104     end
105     % Solves the Ch.Eq. and calculates transmission constant \
    beta for
106     % every guided TE mode. Also, a recursive method is used
    when is needed
107     for n = 1:TE_modes

```

```

108     ii = -2; % computational parameter
109     out(n,u) = fsolve(fun1,k(n)*(0.99999),options);
110     if (~isreal(out(n,u)) || out(n,u)<=k2 || out(n,u)>=k1)
111         zz = 1;% computational parameter
112         flag = 0;% computational parameter
113         while ((out(n,u)<=k2 || out(n,u)>=k1 || flag<1e-3)
114             && ii<=-1)
115             switch zz
116                 case 1,
117                     if (u==1)
118                         out(n,u) = real(out(n,u));
119                     elseif(isreal(out(n,u-1)))
120                         out(n,u) = fsolve(fun1,out(n,u-1)
121                             *0.999,options);
122                     else
123                         out(n,u) = fsolve(fun1,k(n)
124                             *(1+0.1*10^ii),options);
125                     end
126                 case 2,
127                     flag = out(n,u);
128                     if(isreal(out(n,u)))
129                         out(n,u)= (out(n,u) + fsolve(fun1,
130                             out(n,u)*(1+10*eps),options))/2;
131                     else
132                         out(n,u)= (out(n,u) + fsolve(fun1,k
133                             (n)*(1+0.1*10^ii),options))/2;
134                     end
135                     flag = abs(out(n,u) - flag);
136                 end
137                 zz = 2;% computational parameter
138                 ii= ii + 0.1;% computational parameter
139             end
140         end
141     end
142 end
143 % Corrects the computational faults, if something went wrong.
144 % Usually, this happens when the thickness step is too small
145 for n = 1:TE_modes
146     for u = 2:length(di)-1
147         if( ~isnan(out(n,u-1)*out(n,u+1)))
148             if out(n,u)<out(n,u+1)
149                 mid = real((out(n,u-1) + out(n,u+1)))/2;
150                 out(n,u) = mid;
151             end
152         end
153     end
154 end
155 n_eff = out*k1_/k0; % effective refractive index for every
    guided mode
156
157 %% Plot the results    N_eff vs Thickness d
158 figure
159 hold on; axis on; box on;
160 % TE_0

```

```

156 plot(di*10^9,real(n_eff(1,:)), 'Color','blue','LineWidth',2);
157 % TE_1
158 plot(di*10^9,real(n_eff(2,:)), 'Color','blue','LineWidth',2, '
    LineStyle','--');
159 % TE_2
160 plot(di*10^9,real(n_eff(3,:)), 'Color','blue','Linewidth',2, '
    LineStyle','-');
161
162 xlabel('Thickness (nm)')
163 ylabel('N_{eff}')
164 legend('TE_0','TE_1','TE_2')
165 ylim([n2*0.999 n1])
166 xlim([min(di) max(di)]*10^9)

```