

## Experiment No. 1

### AIM

Introduction to lab equipment & operational amplifier IC - 4A741C

### THEORY

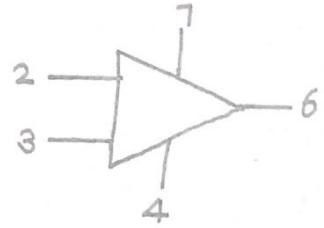
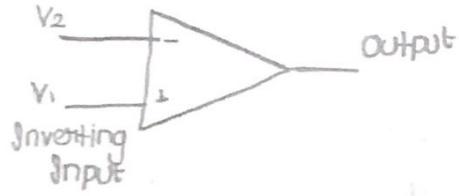
DSO or Digital Storage Oscilloscope is an instrument which stores a digital copy of the waveform in the digital memory which it analyses further using digital signal processing techniques. Signals are received, stored & then displayed.

Function Generator is usually a piece of electronic test equipment or software used to generate different types of electrical waveforms over a wide range of frequencies. Some of the most common waveforms produced by function generator are sine wave, square wave, triangular wave & sawtooth shapes.

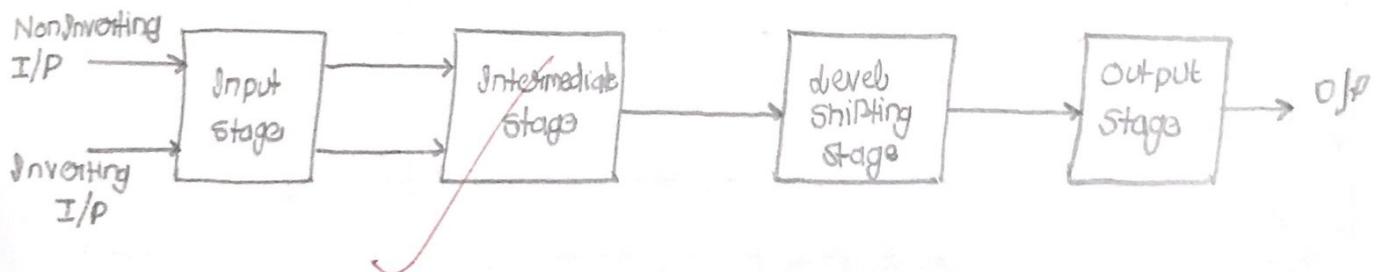
Dual Power Supply is a common equipment in electronics. The +ve polarity as well as -ve polarity & ground potential of op-amps need dual power sources.

Operational Amplifier is a direct coupled high gain voltage amplifier with very high impedance.

### Symbol of op-amp



### Block Diagram of Op-amp



### Ideal Values of 741C

$$A_v = \infty$$

$$R_i = \infty$$

$$R_o = 0$$

$$BW = \infty$$

$$CMRR = \infty$$

### Practical Values of 741C

$$A_v = 2 \times 10^5$$

$$R_i = 2 M\Omega$$

$$R_o = 75 \Omega$$

$$BW = 1 \text{ MHz}$$

$$CMRR = 90 \text{ dB}$$

### Different Types of Op-Amps

Each manufacturer uses a specific code & assigns a specific type number to the ICs. It produces -

1. Fairchild - 741, 741F
2. National Semiconductor - LM, LH, LF, TBA
3. Motorola - MG, MFA
4. Texas Instruments - SN
5. Signetics - N/S, NE/SE, SU

741 is a military grade op-amp (temp range  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ )

741C - Commercial grade op-amp ( $0^\circ\text{C}$  to  $70/75^\circ\text{C}$ )

741A & 741E are improved versions of 741 & 741C

74147 - Dual op-amp

LM32A      3 Quad op-amp  
LF 351/353

TL082 - JFET Dual op-amp

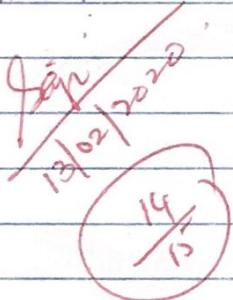
→ There are several different types of op-amps, some of which are -

- Voltage Follower
- Inverting op-amp
- Non-inverting op-amp
- Op-amp differentiator
- Differentiable amplifier
- Integrator



### RESULT

We studied & learned about lab equipment and uA741C amplifier.



## General purpose operational amplifier

 $\mu$ A741/ $\mu$ A741C/SA741C

## DESCRIPTION

The  $\mu$ A741 is a high performance operational amplifier with high open-loop gain, internal compensation, high common mode range and exceptional temperature stability. The  $\mu$ A741 is short-circuit-protected and allows for nulling of offset voltage.

## FEATURES

- Internal frequency compensation
- Short circuit protection
- Excellent temperature stability
- High input voltage range

## PIN CONFIGURATION

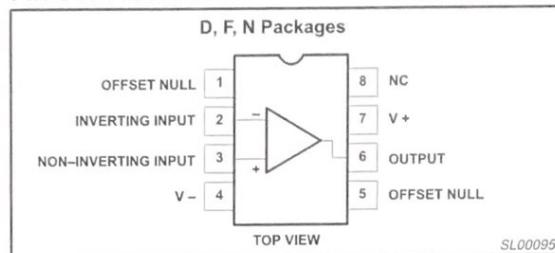


Figure 1. Pin Configuration

## ORDERING INFORMATION

DESCRIPTION	TEMPERATURE RANGE	ORDER CODE	DWG #
8-Pin Plastic Dual In-Line Package (DIP)	-55°C to +125°C	$\mu$ A741N	SOT97-1
8-Pin Plastic Dual In-Line Package (DIP)	0 to +70°C	$\mu$ A741CN	SOT97-1
8-Pin Plastic Dual In-Line Package (DIP)	-40°C to +85°C	SA741CN	SOT97-1
8-Pin Ceramic Dual In-Line Package (CERDIP)	-55°C to +125°C	$\mu$ A741F	0580A
8-Pin Ceramic Dual In-Line Package (CERDIP)	0 to +70°C	$\mu$ A741CF	0580A
8-Pin Small Outline (SO) Package	0 to +70°C	$\mu$ A741CD	SOT96-1

## ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	RATING	UNIT
$V_S$	Supply voltage $\mu$ A741C $\mu$ A741	$\pm 18$ $\pm 22$	V V
$P_D$	Internal power dissipation D package N package F package	780 1170 800	mW mW mW
$V_{IN}$	Differential input voltage	$\pm 30$	V
$V_{IN}$	Input voltage <sup>1</sup>	$\pm 15$	V
$I_{SC}$	Output short-circuit duration	Continuous	
$T_A$	Operating temperature range $\mu$ A741C SA741C $\mu$ A741	0 to +70 -40 to +85 -55 to +125	°C °C °C
$T_{STG}$	Storage temperature range	-65 to +150	°C
$T_{SOLD}$	Lead soldering temperature (10sec max)	300	°C

## NOTES:

- For supply voltages less than  $\pm 15$ V, the absolute maximum input voltage is equal to the supply voltage.

## General purpose operational amplifier

 $\mu$ A741/ $\mu$ A741C/SA741C

## DC ELECTRICAL CHARACTERISTICS

 $T_A = 25^\circ\text{C}$ ,  $V_S = \pm 15\text{V}$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	$\mu$ A741			$\mu$ A741C			UNIT
			Min	Typ	Max	Min	Typ	Max	
$V_{OS}$	Offset voltage	$R_S=10\text{k}\Omega$	1.0	5.0		2.0	6.0		mV
		$R_S=10\text{k}\Omega$ , over temp.	1.0	6.0		10	7.5		mV
			10						$\mu\text{V}/^\circ\text{C}$
$I_{OS}$	Offset current	Over temp.	20	200		20	200		nA
		$T_A=+125^\circ\text{C}$	7.0	200					nA
		$T_A=-55^\circ\text{C}$	20	500					nA
			200			200			$\text{pA}/^\circ\text{C}$
$I_{BIAS}$	Input bias current	Over temp.	80	500		80	500		nA
		$T_A=+125^\circ\text{C}$	30	500					nA
		$T_A=-55^\circ\text{C}$	300	1500					nA
			1			1			$\text{nA}/^\circ\text{C}$
$V_{OUT}$	Output voltage swing	$R_L=10\text{k}\Omega$	$\pm 12$	$\pm 14$		$\pm 12$	$\pm 14$		V
		$R_L=2\text{k}\Omega$ , over temp.	$\pm 10$	$\pm 13$		$\pm 10$	$\pm 13$		V
$A_{VOL}$	Large-signal voltage gain	$R_L=2\text{k}\Omega$ , $V_O=\pm 10\text{V}$	50	200		20	200		$\text{V}/\text{mV}$
		$R_L=2\text{k}\Omega$ , $V_O=\pm 10\text{V}$ , over temp.	25			15			$\text{V}/\text{mV}$
Offset voltage adjustment range				$\pm 30$			$\pm 30$		mV
PSRR	Supply voltage rejection ratio	$R_S \leq 10\text{k}\Omega$				10	150		$\mu\text{V}/\text{V}$
		$R_S \leq 10\text{k}\Omega$ , over temp.		10	150				$\mu\text{V}/\text{V}$
CMRR	Common-mode rejection ratio	Over temp.	70	90		70	90		dB
									dB
$I_{CC}$	Supply current	$T_A=+125^\circ\text{C}$		1.4	2.8		1.4	2.8	mA
		$T_A=-55^\circ\text{C}$		1.5	2.5				mA
				2.0	3.3				mA
$V_{IN}$	Input voltage range	( $\mu$ A741, over temp.)	$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		V
		$R_{IN}$	0.3	2.0		0.3	2.0		$\text{M}\Omega$
$P_D$	Power consumption	$T_A=+125^\circ\text{C}$		50	85		50	85	mW
		$T_A=-55^\circ\text{C}$		45	75				mW
				45	100				mW
$R_{OUT}$	Output resistance			75			75		$\Omega$
$I_{SC}$	Output short-circuit current		10	25	60	10	25	60	mA

## General purpose operational amplifier

 $\mu A741/\mu A741C/SA741C$ 

## DC ELECTRICAL CHARACTERISTICS

 $T_A = 25^\circ C$ ,  $V_S = \pm 15V$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	SA741C			UNIT
			Min	Typ	Max	
$V_{OS}$	Offset voltage	$R_S=10k\Omega$ $R_S=10k\Omega$ , over temp.		2.0	6.0	mV
$\Delta V_{OS}/\Delta T$				10	7.5	mV
$I_{OS}$	Offset current	Over temp.		20	200	nA
$\Delta I_{OS}/\Delta T$				200	500	pA/ $^\circ C$
$I_{BIAS}$	Input bias current	Over temp.		80	500	nA
$\Delta I_B/\Delta T$				1	1500	nA
$V_{OUT}$	Output voltage swing	$R_L=10k\Omega$ $R_L=2k\Omega$ , over temp.	$\pm 12$	$\pm 14$		V
		$R_L=2k\Omega, V_O=\pm 10V$	$\pm 10$	$\pm 13$		V
$A_{VOL}$	Large-signal voltage gain	$R_L=2k\Omega, V_O=\pm 10V$ , over temp.	20	200		V/mV
	Offset voltage adjustment range			15		V/mV
PSRR	Supply voltage rejection ratio	$R_S \leq 10k\Omega$		10	150	$\mu V/V$
CMRR	Common mode rejection ratio			70	90	dB
$V_{IN}$	Input voltage range	Over temp.	$\pm 12$	$\pm 13$		V
$R_{IN}$	Input resistance		0.3	2.0		M $\Omega$
$P_d$	Power consumption			50	85	mW
$R_{OUT}$	Output resistance			75		$\Omega$
$I_{SC}$	Output short-circuit current			25		mA

## AC ELECTRICAL CHARACTERISTICS

 $T_A=25^\circ C$ ,  $V_S = \pm 15V$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	$\mu A741, \mu A741C$			UNIT
			Min	Typ	Max	
$R_{IN}$	Parallel input resistance	Open-loop, f=20Hz	0.3			M $\Omega$
$C_{IN}$	Parallel input capacitance	Open-loop, f=20Hz		1.4		pF
	Unity gain crossover frequency	Open-loop		1.0		MHz
$t_R$	Transient response unity gain Rise time Overshoot Slew rate	$V_{IN}=20mV, R_L=2k\Omega, C_L \leq 100pF$		0.3		$\mu s$
SR		$C \leq 100pF, R_L \geq 2k\Omega, V_{IN}=\pm 10V$		5.0		%
				0.5		V/ $\mu s$

## General purpose operational amplifier

 $\mu$ A741/ $\mu$ A741C/SA741C

## EQUIVALENT SCHEMATIC

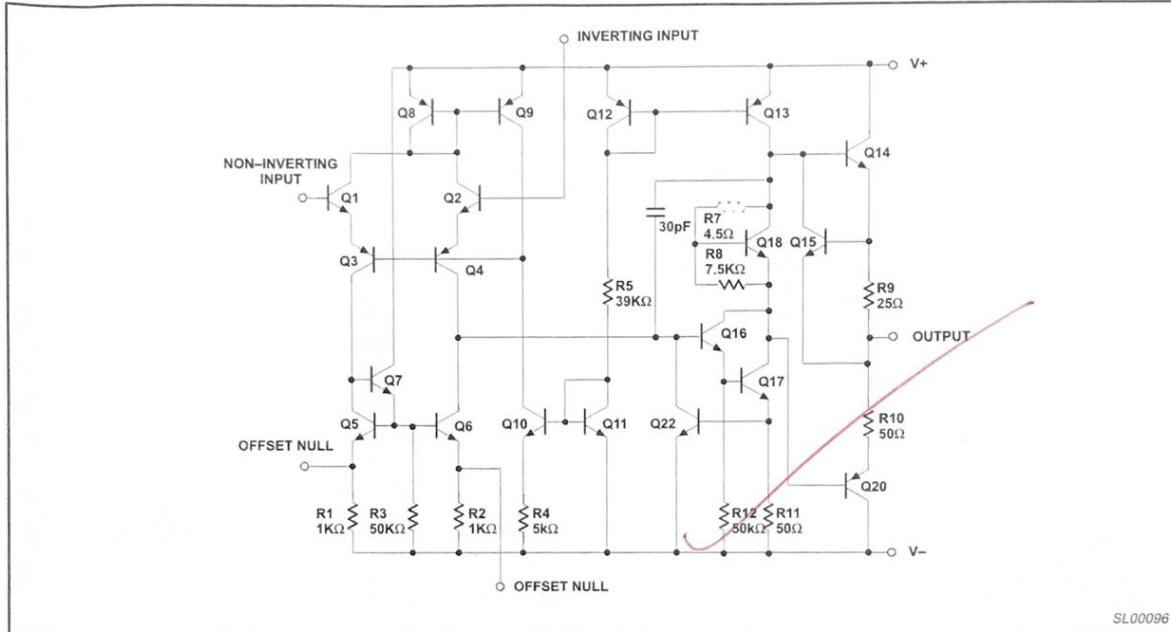


Figure 2. Equivalent Schematic

SL00096

## Experiment No. 2

### AIM

To construct & test performance of op-amp in open loop configuration as  
① Inverting Amplifier ② Non-Inverting Amp.  
③ Differential Amplifier using uA741C

### APPARATUS REQUIRED

DSO Tekhniko (40 MHz), Function Generator (25 MHz), op-amp uA741C,  
Dual power supply (Scientech 4075), Breadboard, connecting  
wires, probes

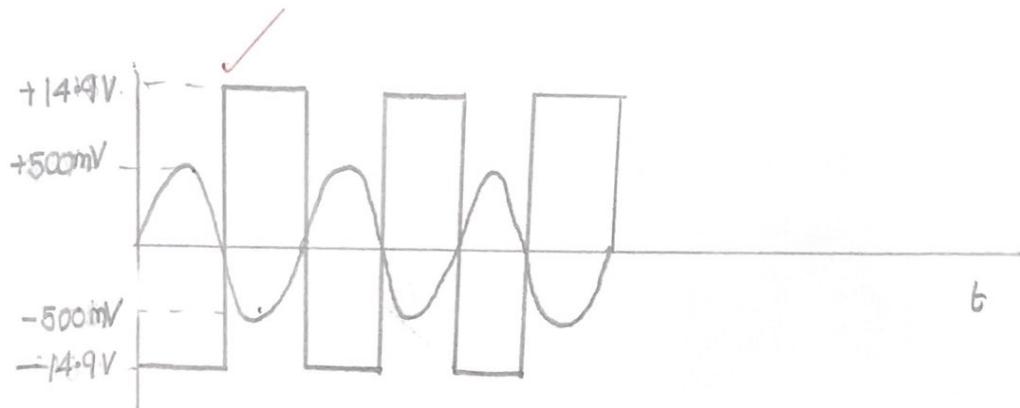
### THEORY

In case of amplifiers the term open loop gain indicates that no connection, either direct or via another network exists between the output & input terminals. That is, the output signal is not fed back in any form as part of input signal, and the loop that would have been formed with feedback is open.

When connected in open-loop configuration, the op-amp simply functions as a high gain amplifier. There are three open loop op-amp configurations

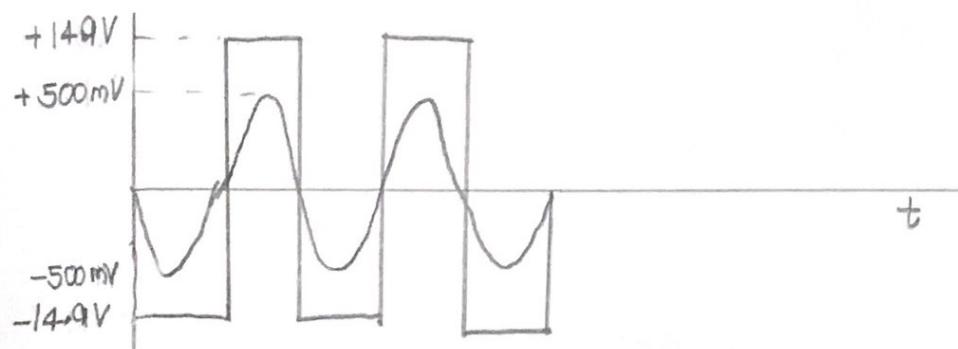
## Inverting Amplifier

	Freq.	Pk-Pk (V)	Max (v)	Min (v)
S/P	1kHz	1V	+500mV	-500mV
O/P	1kHz	29.8V	+14.9V	-14.9V

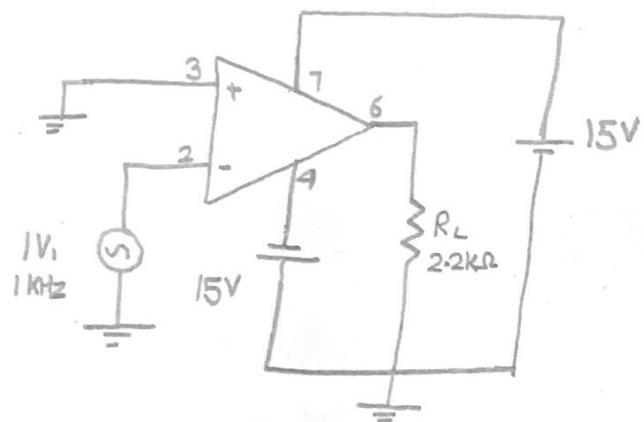


## Non-Inverting Amplifier

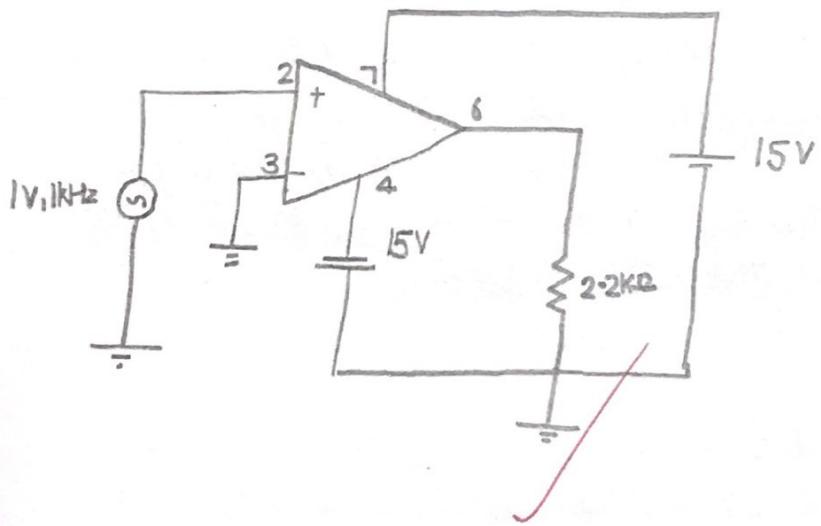
	Freq.	Pk-Pk (V)	Max (v)	Min (v)
S/P	1kHz	1V	+500mV	-500mV
O/P	1kHz	29.8V	+14.9V	-14.9V



## Inverting Amplifier



## Non-Inverting Amplifier



### 1. Inverting Amplifier :

In the inverting amplifier only one input is applied & that is to the inverting input terminal. The non inverting input terminal is grounded.

$$\text{Give } V_1 = 0V \text{ & } V_2 = V_{in}$$

$$V_o = -AV_{in}$$

The negative sign indicates that output voltage is out of phase with respect to input by  $180^\circ$  or is of opposite polarity. Thus in the inverting amplifier the input signal is amplified by gain A & is also inverted at the output.

### 2. Non inverting Amplifier :

In this configuration the input is applied to non inverting input terminal & the inverting terminal is connected to ground.

$$V_1 = V_{in} \text{ and } V_2 = 0V$$

$$\text{So } V_o = AV_{in}$$

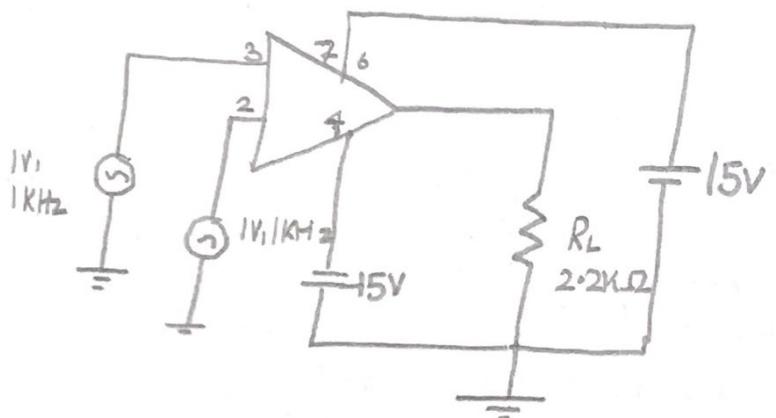
Output voltage is larger than input voltage by gain A & is in phase with the input signal.

### 3. Differential Amplifiers :

Since the op-amp amplifies the difference between the two input signals. This config. is called diff'nl amplifier.

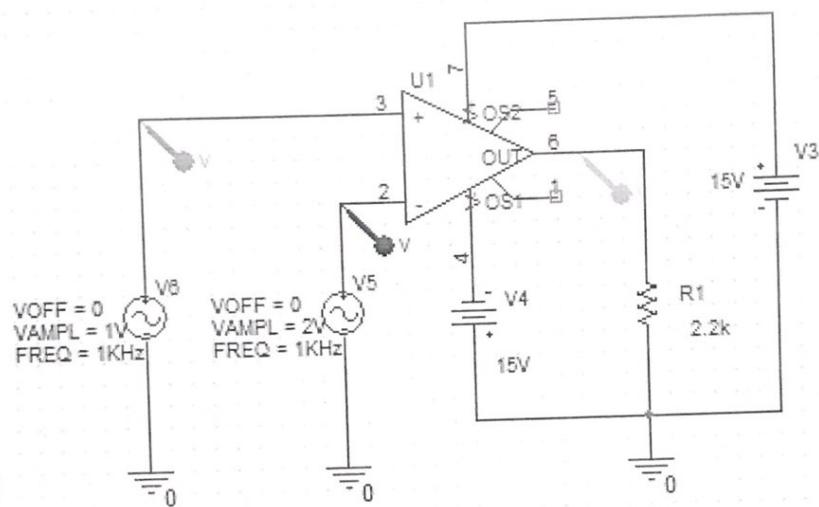
The op-amp is a versatile device because it amplifies both ac & dc input signals. The source resistances  $R_{in1}$  &  $R_{in2}$  are normally negligible compared to input resistance  $R_i$ . So the voltage

## Differential Amplifier

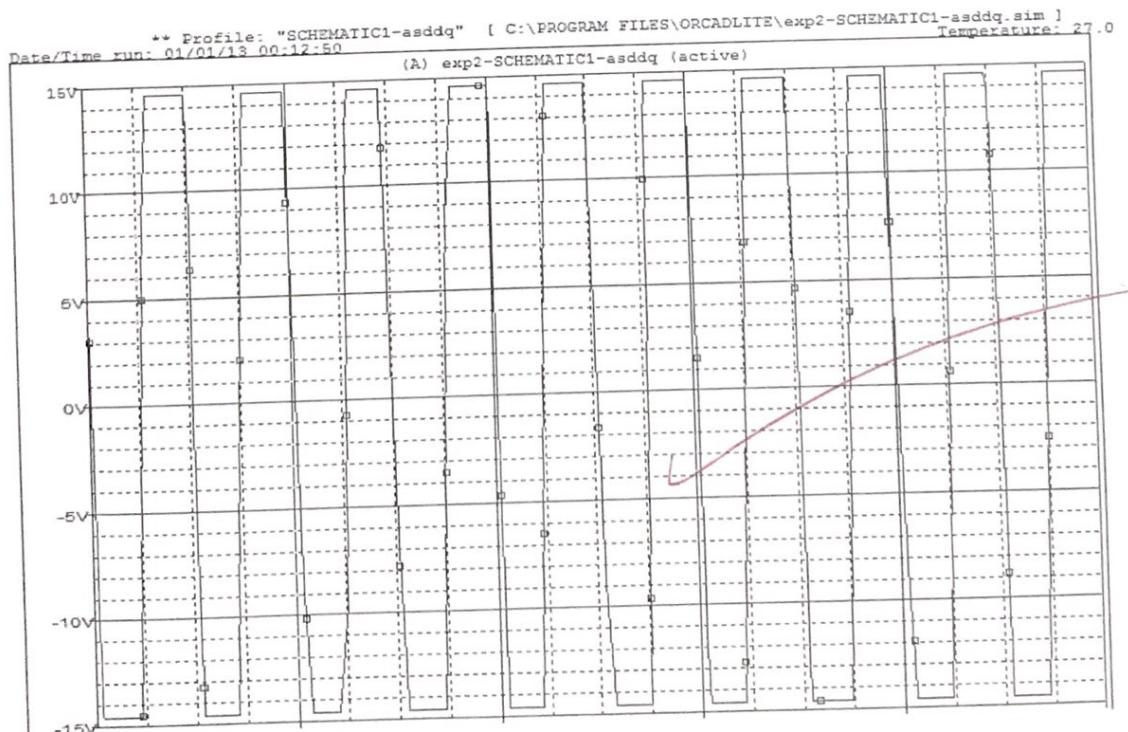


## Experiment No. 2

### Differential Amplifier



### Output



Drop across resistors can be assumed to be zero.

$$V_i = V_{in1} \text{ & } V_2 = V_{in2}$$

$$V_o = A(V_{in1} - V_{in2})$$

In open loop configurations, gain A is commonly referred to as openloop gain

### RESULT

Construction & test performance of op-amp in open loop configuration as ① Inverting Amp. ② Non Inverting Amp. & ③ Differential Amp. using UA741G has been done.


 $\frac{14}{13}$

$f_{op}$   
 $1310^2$   
 $2^{20}$

## Experiment No. 3

### AIM

To construct & test performance of closed loop op-amp in  
① Inverting Configuration ② Non Inverting Amplifier &  
③ Voltage Amplifier

### APPARATUS

DSO Techtronics (40 MHz), Function Generator (25 MHz), opamp 4A741G, Dual Power Supply (Scientech 4075), Breadboard, Connecting Wires & probes.

### THEORY

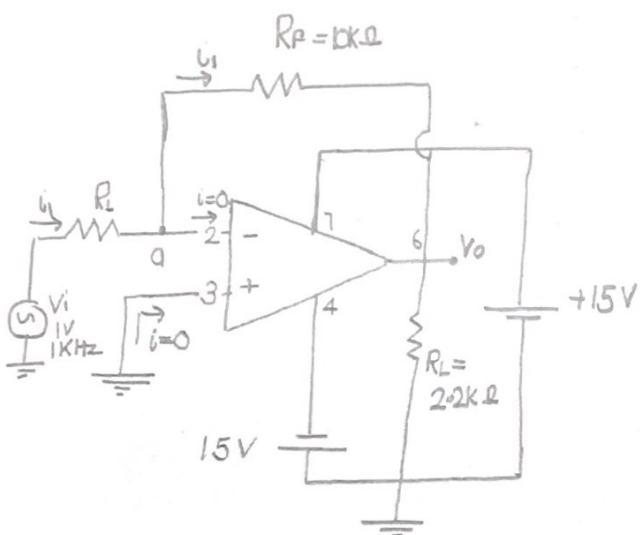
Gain of open loop opamp is very high as it makes the amplifier both unstable & hard to control as smallest of input signals would be enough to cause the output voltage to saturate & swing towards one or the other of voltage supply rails losing complete control of the output.

#### Feedback in Ideal Op-Amp.

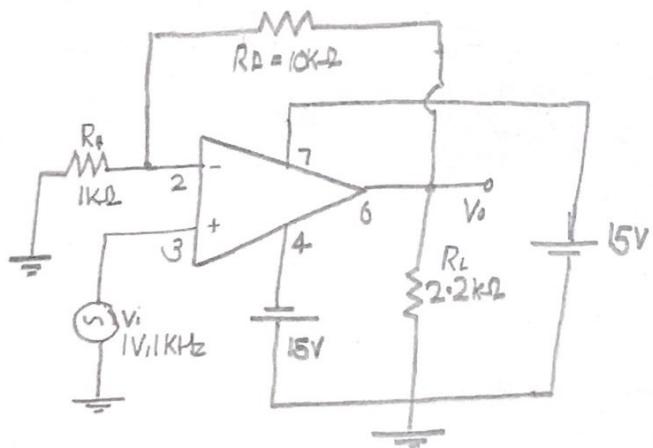
The utility of an op-amp can be greatly increased by providing negative feedback. The output in this case isn't driven into saturation & the circuit behaves in a linear manner.

#### # Inverting Amplifier

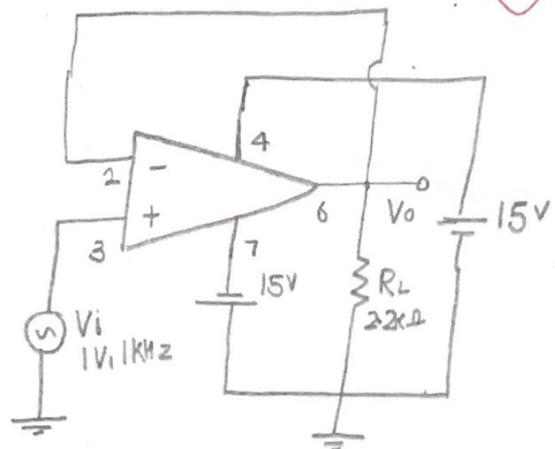
This is perhaps most widely used of all the op-amp



Inverting Amplifier



Non Inverting Amplifier



Voltage Follower

Circuits. The output voltage  $V_o$  is feedback to the inverting input terminal through the  $R_P - R_L$  network where,  $R_P$  is the feedback resistor. Input signal  $V_i$  is applied to the inverting input terminal through  $R_L$  & non inverting input terminal op-amp is grounded.

Gain of an inverting Amplifier is  $A_{ab} = -\frac{R_P}{R_L}$

#### # Non Inverting Amplifier

If a signal is applied to non-inverting input terminal and feedback, the circuit amplifies without inverting the input signal. Such a circuit is called non inverting amplifier. As the differential voltage  $V_d$  at the input terminal of op-amp is zero, the voltage at node 'a' is  $V_i$ , same as the input voltage applied to non inverting input terminals.

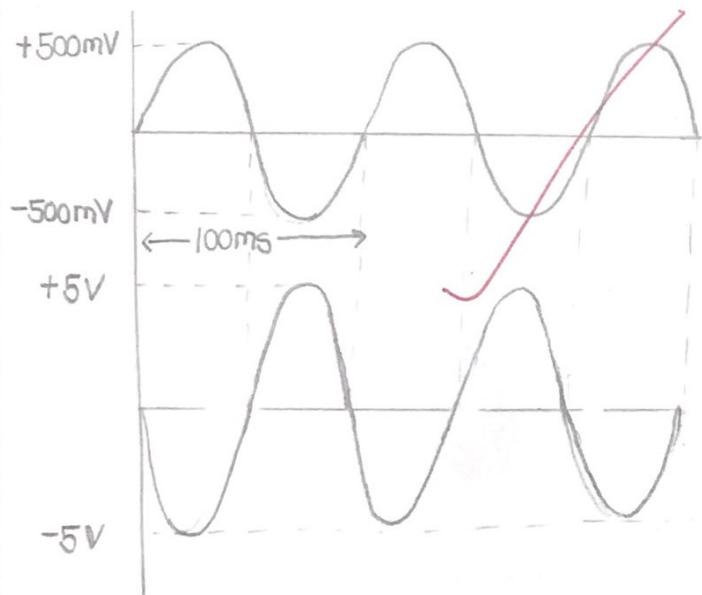
$$\text{Gain } A_{CL} = \frac{V_o}{V_i} = 1 + \frac{R_P}{R_L}$$

#### # Voltage Follower :

In the non inverting amplifier if  $R_P = 0$  and  $R_L = \infty$ , we get modified circuit &  $V_o = V_i$ .

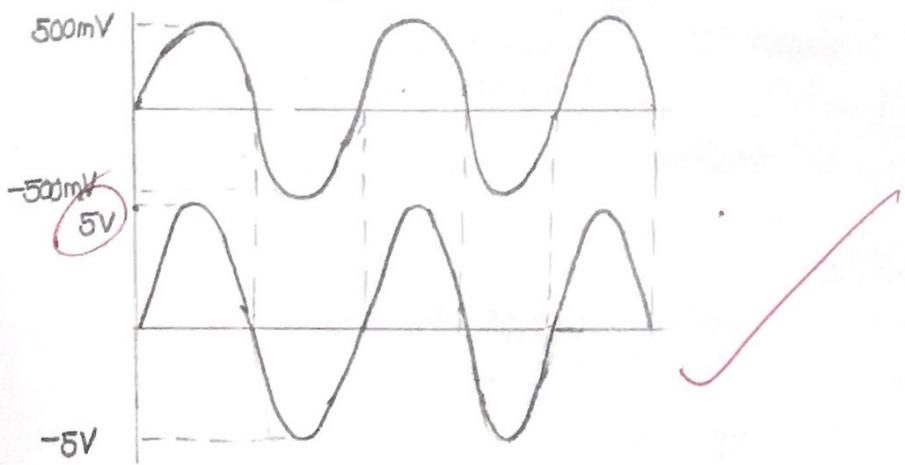
That is, the output voltage is equal to input voltage, both in magnitude & phase. In other words we can also say that the output voltage follows the input voltage exactly. Hence, the circuit is voltage follower.

## Inverting Amplifier



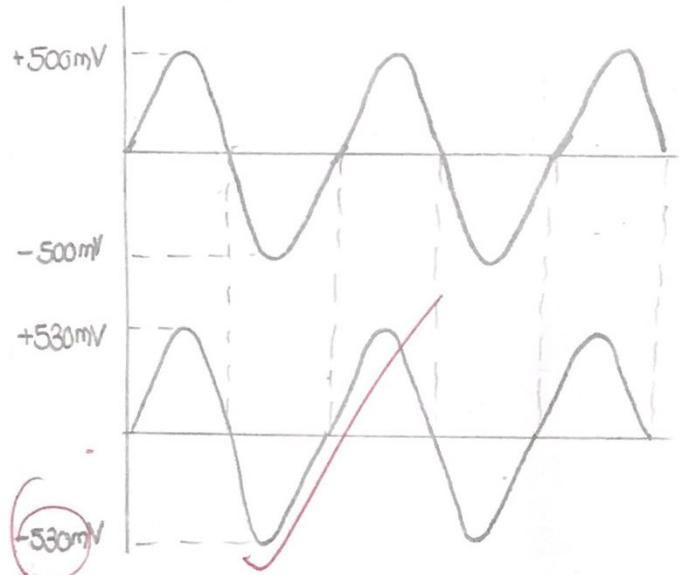
	Frequency	Pk-Pk (V)	Max	Min
Input	1KHz	1V	+500mV	-500mV
Output	998.0Hz	10V	+5V	-5V

## Non Inverting Amplifier



	Frequency	Pk-Pk(V)	Max	Min
Input	1KHz	1V	+500mV	-500mV
Output	1KHz	11V	+5.5V	-5.5V

## Voltage Follower:



	Frequency	Pk-Pk (V)	Max	Min
Input	1KHz	1V	+500mV	-500mV
Output	998.0Hz	1.06V	+530mV	-530mV

## RESULT

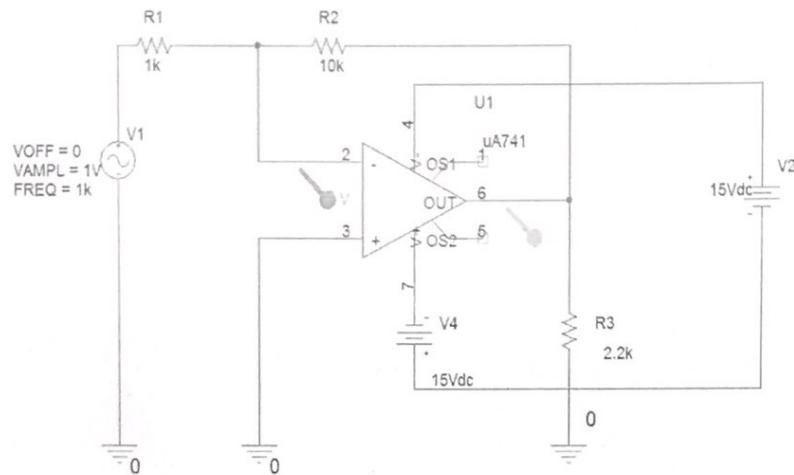
Construction & test performance of closed loop op-amp in Inverting configuration, Non Inverting Amplifier & voltage Follower using uA741C has been done.

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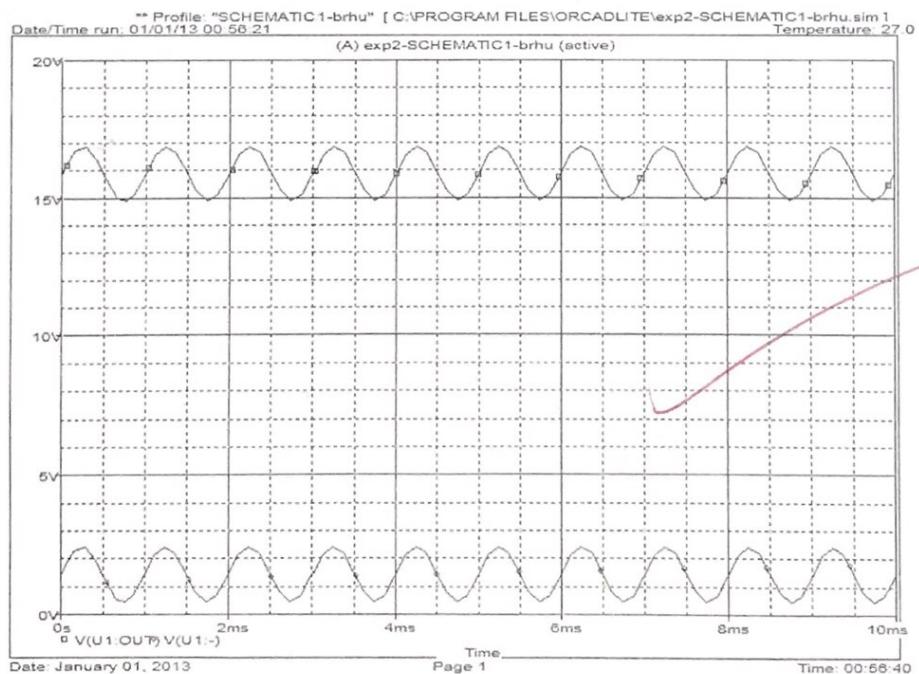
## EXPERIMENT-3

### Closed Loop Inverting

Circuit Diagram



Output



## Experiment No. 4

### AIM

To construct & test the performance of an Integrator using IC uA741G

### APPARATUS & SPECIFICATIONS

Dso Tektronic (40 MHz), Function Generator (25 MHz), op-amp uA741G, Dual Power Supply (Scientech 4075), Breadboard, Connecting Wires, probes

### THEORY

A circuit in which the output voltage waveform is integral of input voltage waveform is the integrator or the integration amplifier

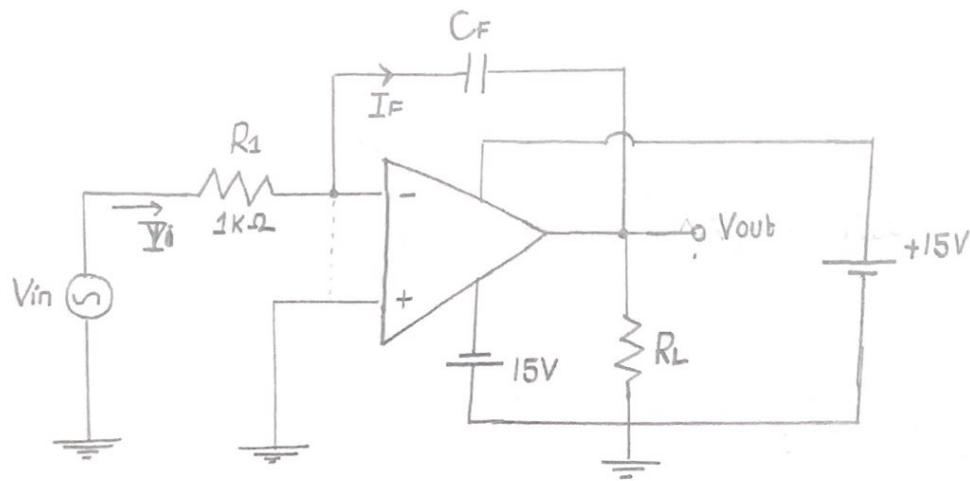
$I_o = I_o$  (due to presence of virtual ground)

We know that  $I_o = C \frac{dV_o}{dt}$

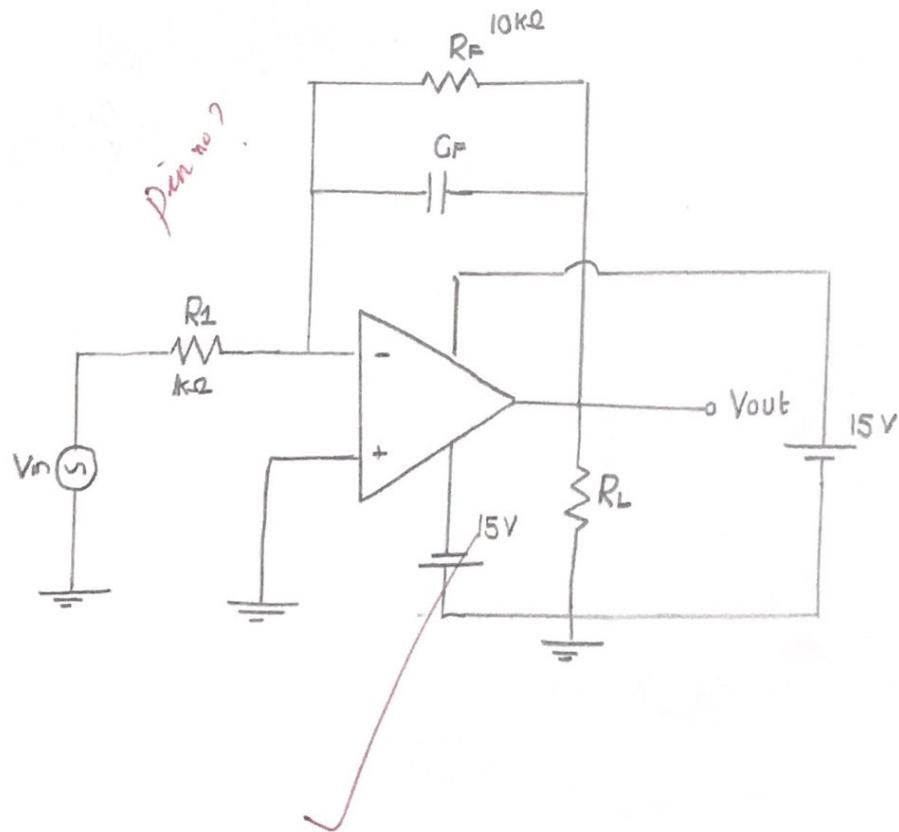
$$\therefore \frac{V_{in} - O}{R_1} = G_F \frac{d(O - V_{out})}{dt}$$

$$\frac{d}{dt} V_{out} = -\frac{1}{R_1 G_F} V_{in}$$

$$V_{out} = -\frac{1}{R_1 G_F} \int V_{in}(t) dt + C$$



Ideal Integrator



① Input : Sine Wave

Output : Cosine Wave

S.No.		Freq.	PK-PK	Max	Min
1	I/P	1 KHz	1V	+500mV	-500mV
	O/P	1 KHz	4.20V	+2.10V	-2.10V
2	I/P	10 KHz	1 V	+500mV	-500mV
	O/P	10.20 KHz	580mV	+290mV	-290mV
3	I/P	50 KHz	1V	+500mV	-500mV
	O/D	50.25 KHz	660mV	+330mV	-330mV
4	I/P	100 KHz	1V	+500mV	-500mV
	O/P	100 KHz	680mV	+340mV	-340mV

② Input : Square Wave

Output : Triangular Wave

	Frequency	PK-PK	Max	Min
I/P	1 KHz	1V	+500mV	-500mV
	994.7 Hz	6.40V	+3.20V	-3.20V

③ Input : Triangular Wave

Output : Parabolic Wave

	Frequency	PK-PK	Max	Min
I/P	1 KHz	1V	+500mV	-500mV
	1 KHz	3.36V	+1.18V	-1.18V

The output voltage is directly proportional to negative integral of input voltage & inversely proportional to time constant  $R_C F$ . For sine input the integrator gives cosine output.

When  $V_{in}=0$ , integrator works in open loop config. CF acts as an open circuit to the input offset voltage  $V_{io}$ . The input off  $V_{io}$  & the part of input current charging capacitor CF produces an error voltage at the output.

∴ In practical integrator we use a resistor  $R_F$  connected across feedback capacitor  $CF$ . Thus  $R_F$  limits low frequency gain & hence minimizes the variations in the output voltages.

Both stability & roll off problems can be corrected by addition of resistor  $R_F$ .

#### Applications :

Waveshaping Circuits

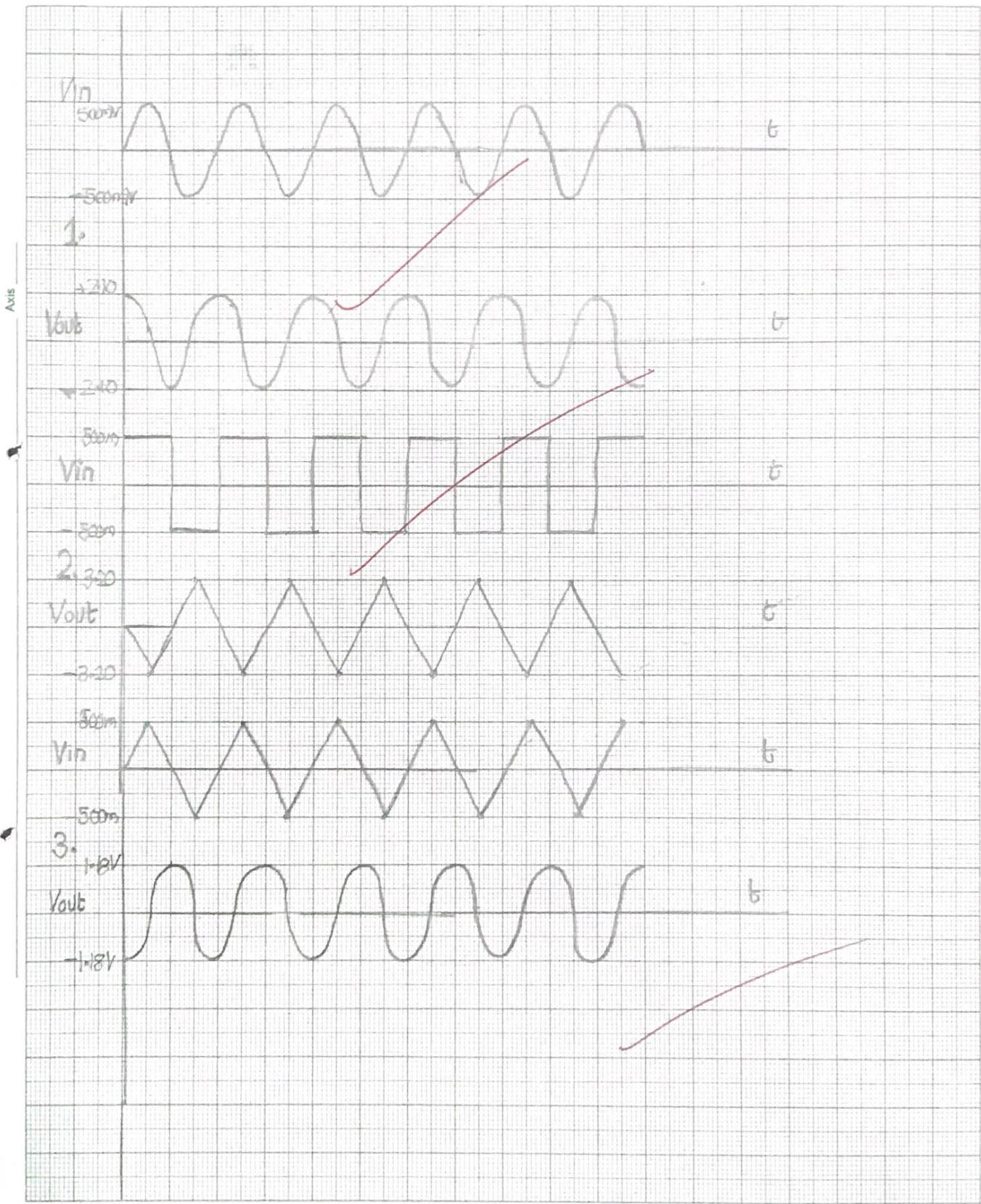
Analog Computers

Analog to Digital Converter

#### RESULT

Construction & test performance of an integrator using IC uA741G has been done

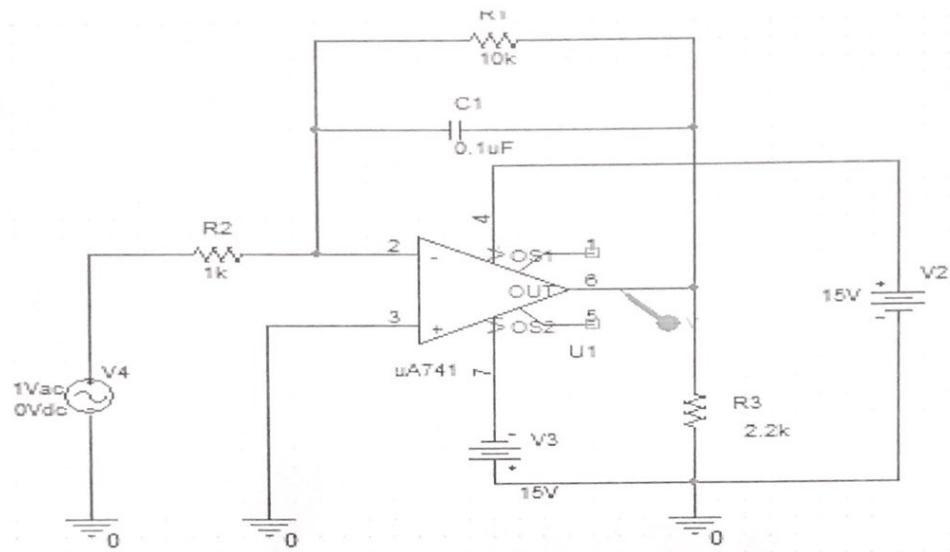
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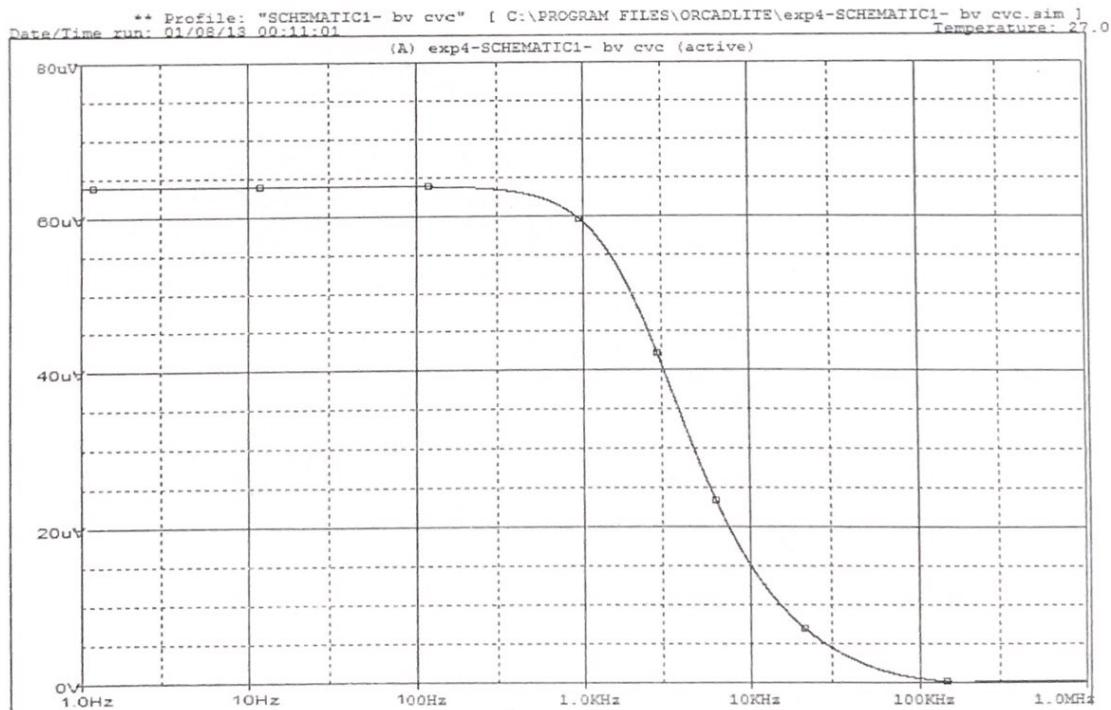
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## Experiment No. 4

### Integrator using $\mu$ A741c



### Output



## Experiment No. 5

**Aim :**

To construct & test the performance of the differentiator using IC uA741C.

### APPARATUS & SPECIFICATIONS.

DSO Tektronics (40 MHz), Function Generator (25 MHz), opamp uA741C, Dual Power Supply (Scientech 4075), breadboard, connecting wires, probes

### Theory :

The circuit performs the mathematical operation of differentiation, that is output waveform is derivative of the input waveform. The differentiator may be constructed from a basic inverting amplifier if an input resistor  $R_1$  is replaced by a capacitor.

The expression for the output voltage can be obtained from Kirchoff's Current Equation written at node  $V_2$  as follows -

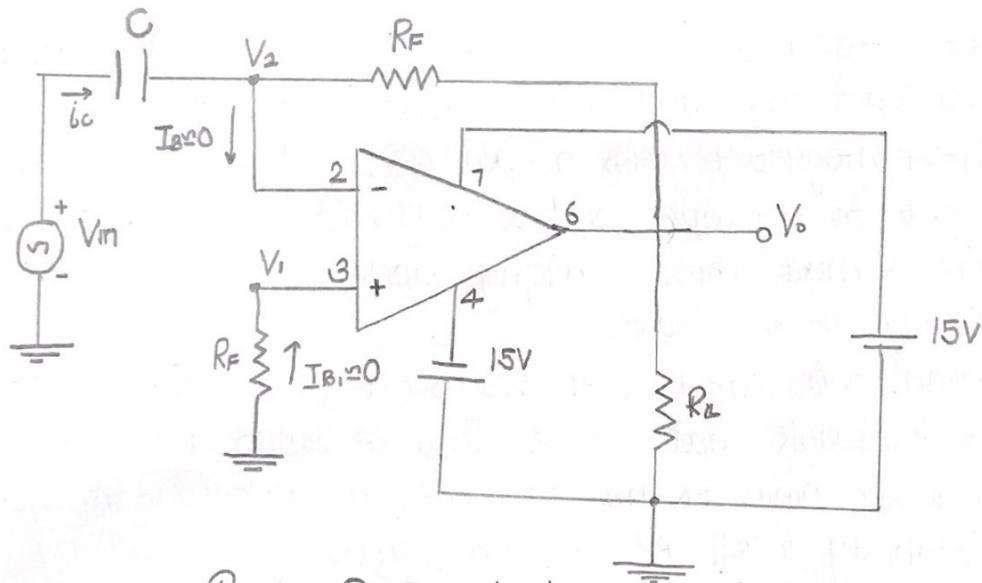
$$i_C = I_B + i_F \quad , \quad \because I_B \approx 0$$

$$i_C = i_F$$

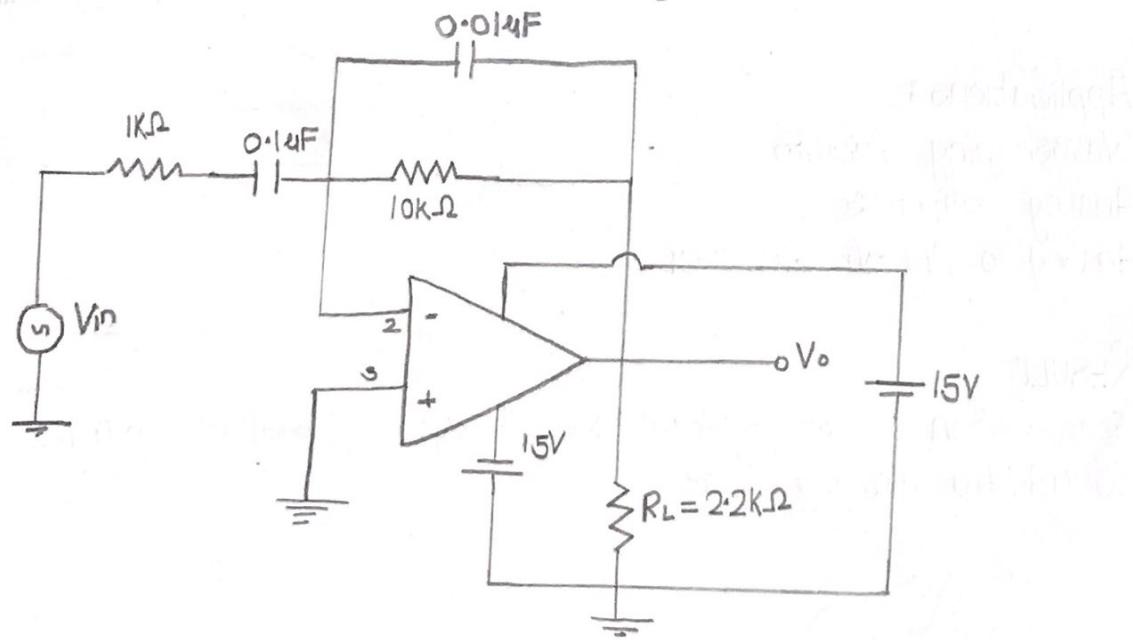
$$C_1 \frac{d}{dt} (V_1 - V_2) = \frac{V_2 - V_o}{R_F}$$

But  $V_1 = V_2 \approx 0V$  because  $A$  is very large. Therefore,

Circuit Diagram :



Basic Differentiator Circuit



Practical Differentiator Circuit

## Observations -

### (1) Input : Sine Wave

	Frequency	Pk-Pk	Max	Min
1.	Input 1KHz	1V	+500mV	-500mV
	Output 1KHz	6.28V	+3.14V	-3.14V
2.	Input 10KHz	1V	+500mV	-500mV
	Output 10.01KHz	3.18V	+1.59V	-1.59V
3.	Input 50KHz	1V	+500mV	-500mV
	Output 50KHz	780mV	+390mV	-390mV
4.	Input 100KHz	1V	+500mV	-500mV
	Output 100KHz	540mV	+270mV	-270mV

### (2) Input : Square Wave

	Frequency	Pk-Pk	Max	Min
(1)	Input 500 Hz	1V	+500mV	-500mV
	Output 500 Hz	6.88V	+3.44V	-3.44V
(2)	Input 1KHz	1V	+500mV	-500mV
	Output 1KHz	6.24V	+3.12V	-3.12V
(3)	Input 10KHz	1V	+500mV	-500mV
	Output 10KHz	2V	+1V	-1V
(4)	Input 50KHz	1V	+500mV	-500mV
	Output 50KHz	880mV	+440mV	-440mV

(3) Input : Ramp Wave

	Frequency	Pk-Pk	Max	Min
1.	Input 1KHz	1V	+500mV	-500mV
	Output 1KHz	3.92V	+1.91V	-1.91V
2.	Input 10KHz	1V	+500mV	-500mV
	Output 10KHz	968mV	+484mV	-484mV
3.	Input 100KHz	1V	+500mV	-500mV
	Output 100KHz	672mV	+336mV	-336mV
4.	Input 150KHz	1V	+500mV	-500mV
	Output 150KHz	640mV	+320mV	-320mV

$$C_1 \frac{dV_{in}}{dt} = -\frac{V_o}{R_F}$$

$$V_o = -R_F C_1 \frac{dV_{in}}{dt}$$

The output  $V_o$  is equal to  $R_F C_1$  times the negative instantaneous rate of change of the input voltage  $V_{in}$  with time. Since the diff. performs the reverse of integrator function. However, this differentiator will have some practical problems. The gain of the circuit ( $R_F / X_C$ ) increases with increase in Frequency at a rate of 20 dB/decade. It makes circuit unstable. Also the input impedance  $X_C$  decreases with increase in Frequency, which makes the circuit very susceptible to high Frequency noise. Both the stability & high Frequency noise problems can be constructed by addition of two components  $R_i$  &  $C_f$ . This circuit is practical differentiator.

Some of its applications are -

- (1) Waveshaping Circuits
- (2) Rate of change detector in FM Modulator

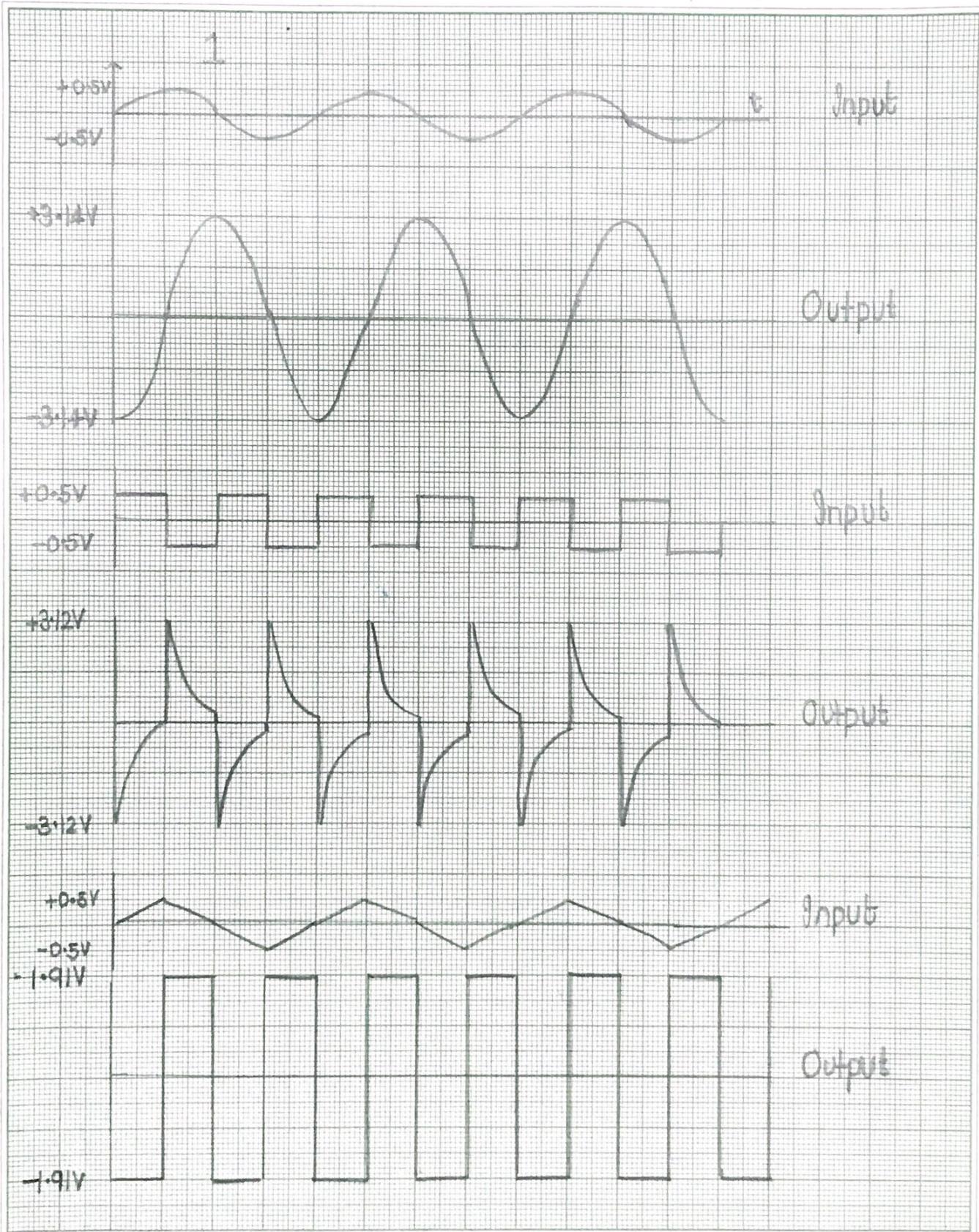
Results

Construction & test performance of the differentiator using IC 4040 has been done.

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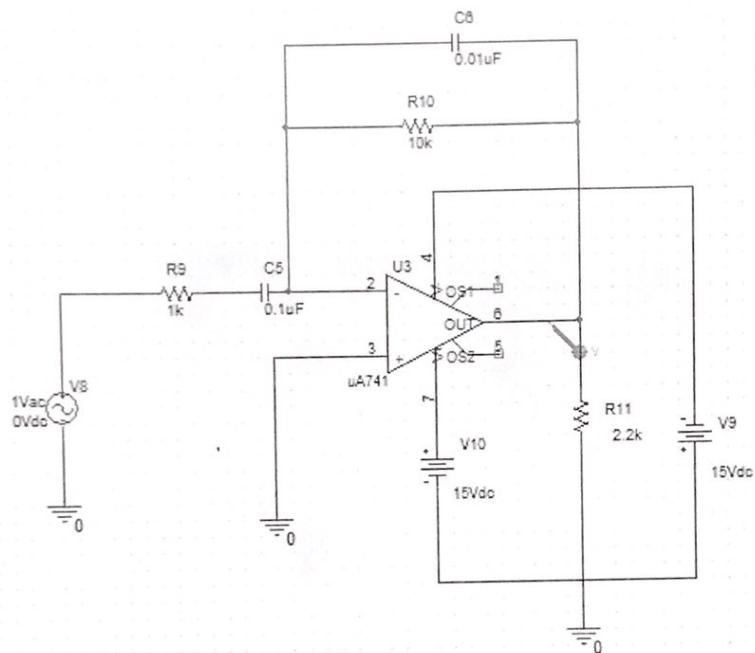
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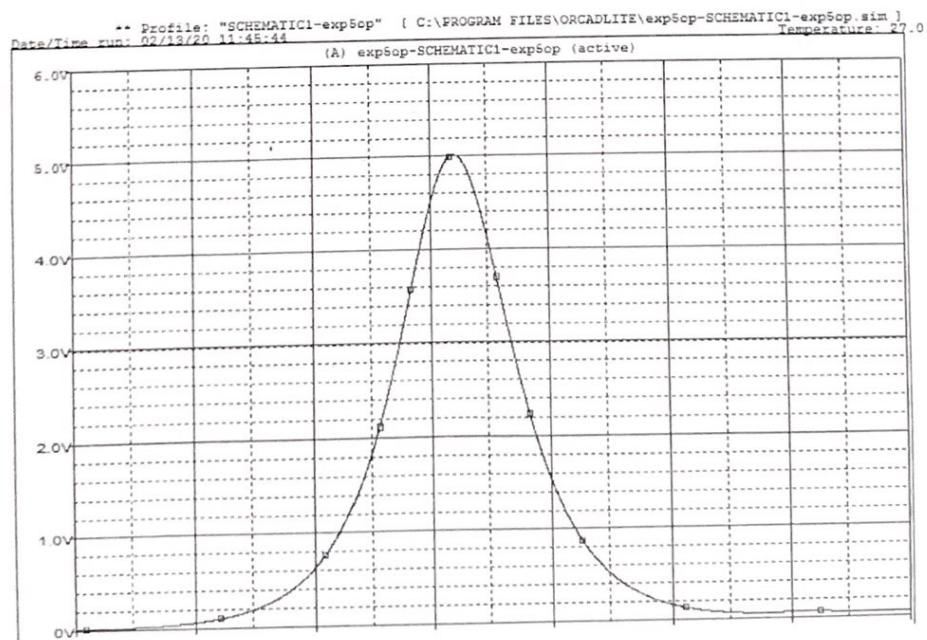


## Experiment No. 5

### Differentiator using $\mu$ A741c



### Output



## Experiment No. 6

Aim :

To design & verify operation of active low pass filter using IC uA741C

Software Used :

Pspice

Theory :

A low pass filter is a filter which only allows frequencies with lower magnitude as compared to ~~specific~~ Frequency to pass through and block the higher frequencies. The limit of the frequencies up to which the circuit allows is known as ~~critical~~ Frequency.

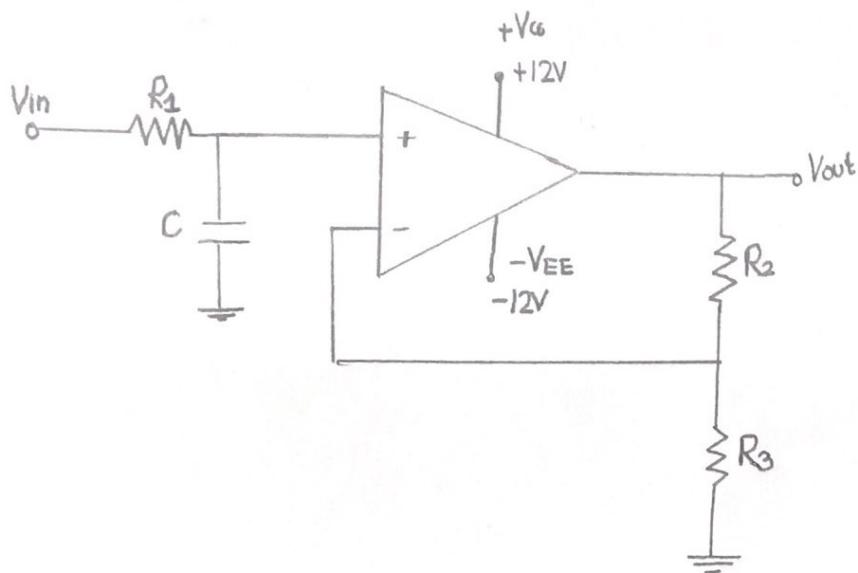
The Frequency is present at -3dB or the maximum magnitude or 0.707 or 70.7% the maximum magnitude.

Calculations :

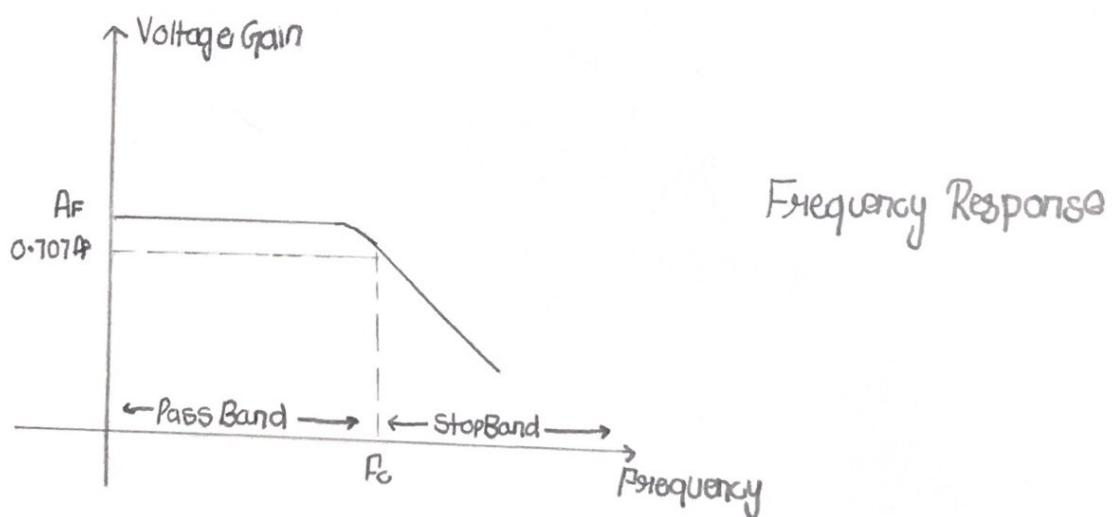
Voltage gain of Low Pass Filter

$$Av = \frac{V_{out}}{V_{in}} = \frac{A_F}{\sqrt{1 + \left(\frac{f}{f_c}\right)^2}}$$

Cutoff Frequency  $f_c = \frac{1}{2\pi RC}$



First Order  
Low Pass Circuit



Topic \_\_\_\_\_

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Let us design a Low Pass Filter with  $f_0 = 10\text{kHz}$   
Let  $C = 1\mu\text{F}$ ,  $R$  is found using equation  $f_0 = 1/2\pi RC$

To find  $R_1$  and  $R_2$

$$\text{Gain} = 1 + \left( \frac{R_1}{R_2} \right)$$

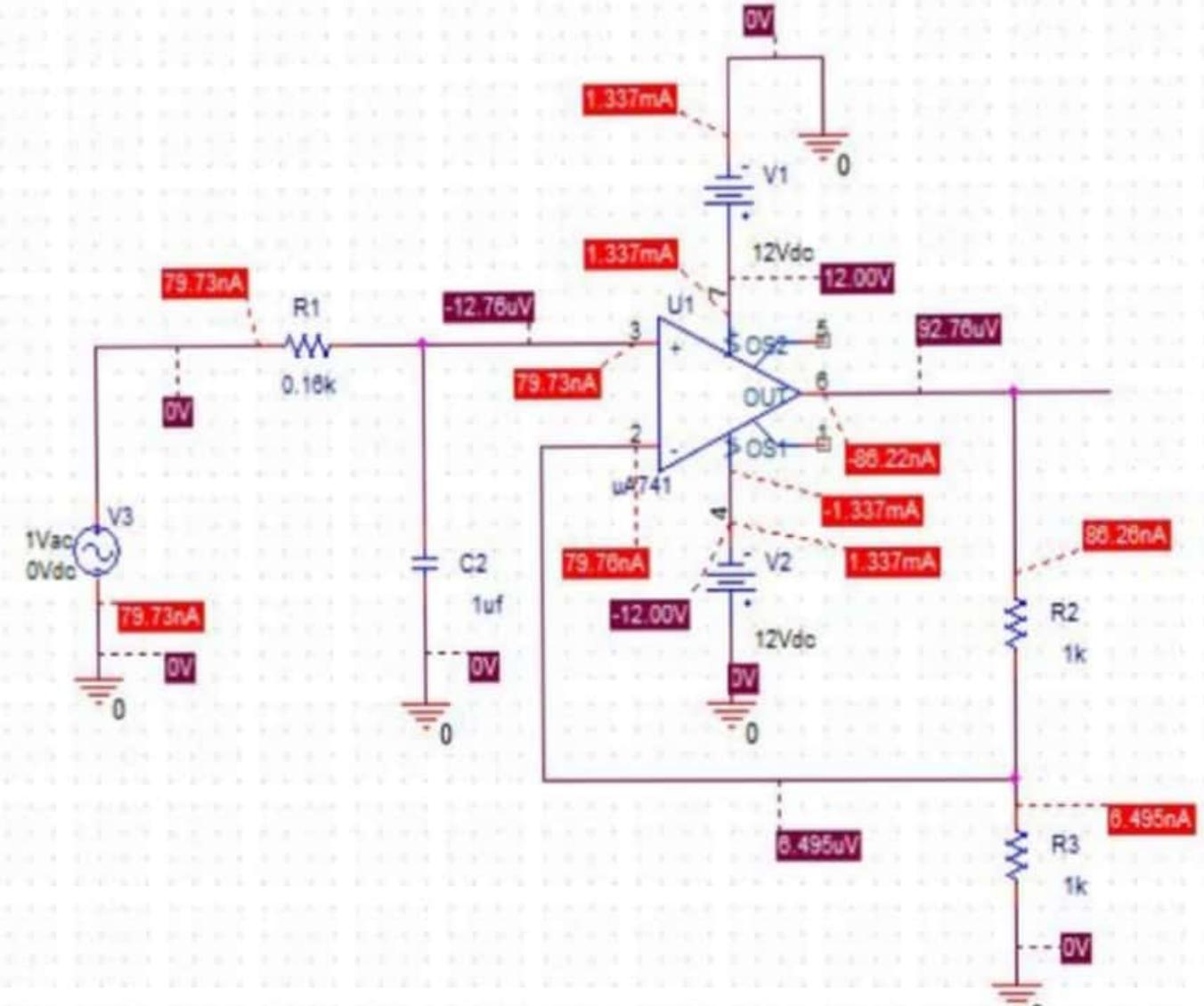
For  $A_v = 2$

$$\frac{R_1}{R_2} = 1$$

If  $R_1 = 1\text{k}\Omega$  then  $R_2 = 1\text{k}\Omega$

Results

Design & verification of active low pass filter using IC 4A741  
has been done



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Date: April 12, 2020

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## Experiment No. 7

Aim:

To design & construct a Schmitt trigger using IC 40106

Tool:

Orcad Pspice

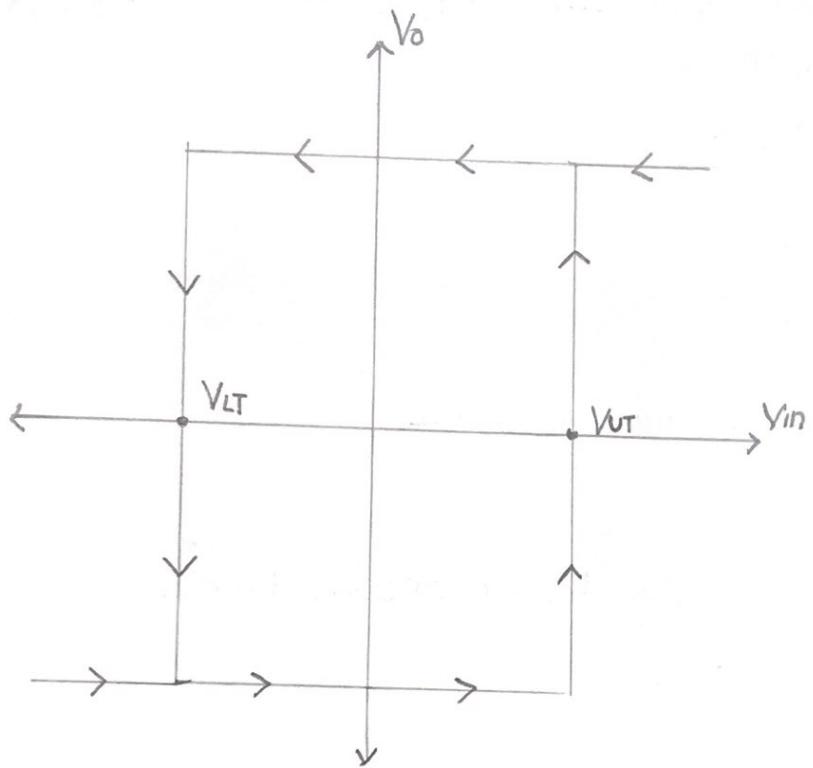
Theory:

Schmitt Trigger converts an irregular shaped waveform to a square wave or pulse. Here, the input voltage triggers the output voltage every time it exceeds certain voltage levels called the upper threshold voltage  $V_{UP}$  and lower threshold voltage  $V_{LP}$ . The input voltage is applied to inverting input. Because the feedback voltage is aiding the input voltage, the feedback is positive. A comparator using positive feedback is usually called Schmitt Trigger. Schmitt Trigger is used as a squaring circuit, amplitude comparator etc.

From Circuit diagram using Voltage divider circuit rule,

$$V_{UP} = \frac{R_2}{R_1 + R_2} (+V_{sat})$$

$$V_{LP} = \frac{R_2}{R_1 + R_2} (-V_{sat})$$



Hysteresis Curve

If the threshold voltages  $V_{UT}$  &  $V_{LT}$  are made larger than input noise voltages, the positive feedback will eliminate the false output transitions. Also the positive feedback, because of regenerative reaction, will make  $V_o$  switch faster between  $+V_{sat}$  &  $-V_{sat}$ .

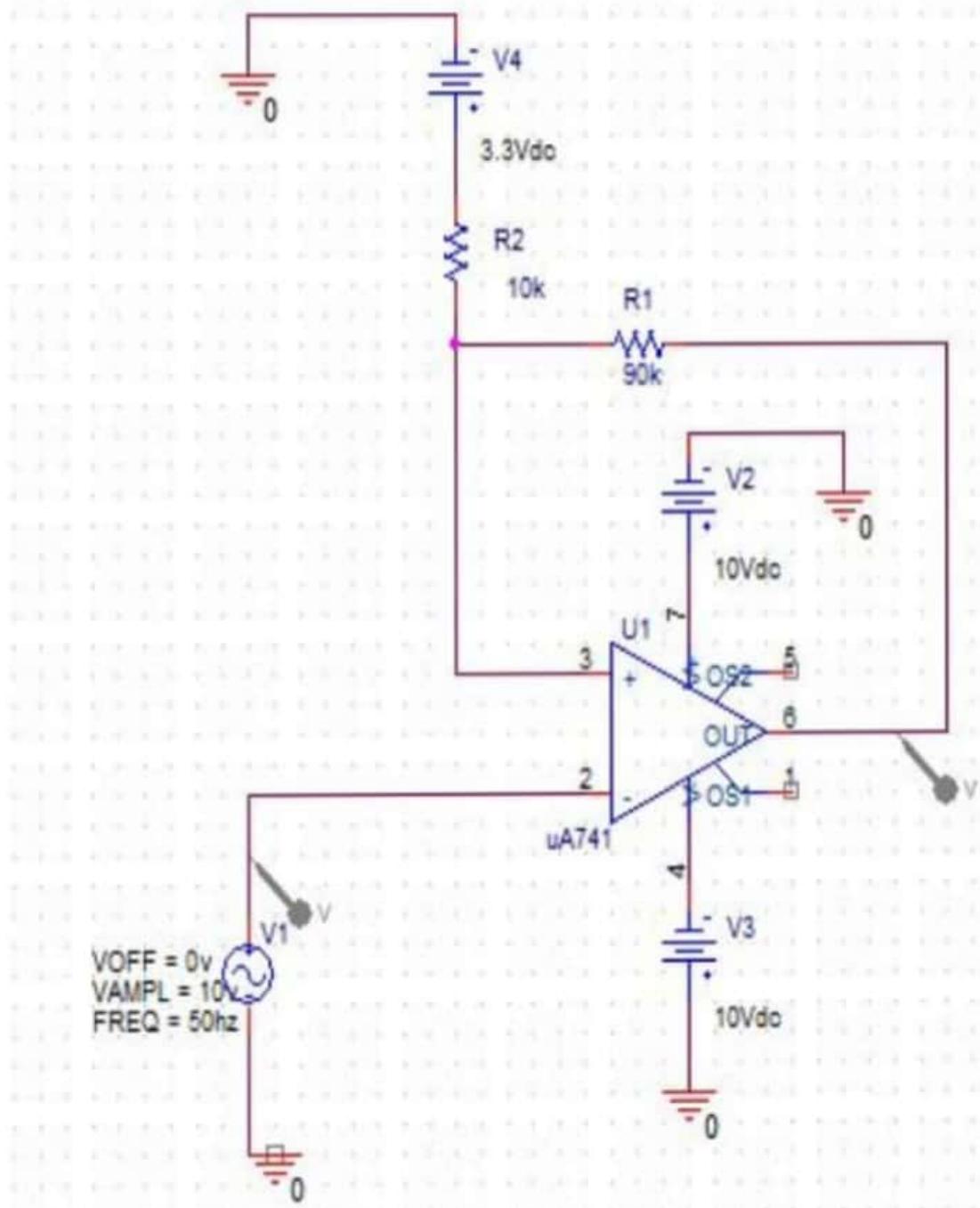
Calculations :

$$V_{UT} = \frac{R_2}{R_1+R_2} (+V_{sat}) = \frac{10}{10+90} \times 10 = +1V$$

$$V_{LT} = \frac{R_2}{R_1+R_2} (-V_{sat}) = \frac{10}{10+90} (-10) = -1V$$

Result :

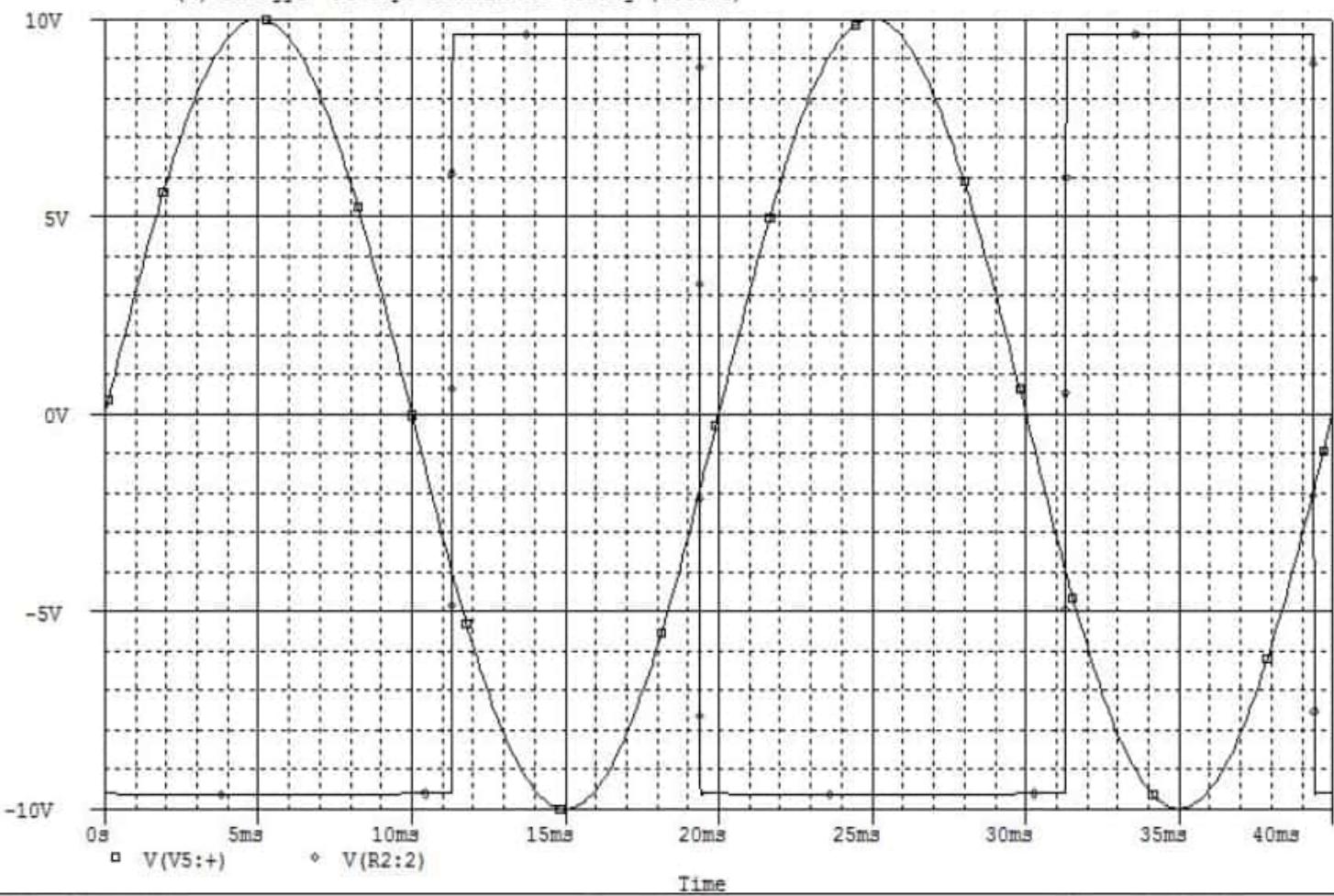
Design & Construction of Schmitt trigger using IC 401741 has been done.



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Date: March 22, 2020

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Time: 12:37:40

## Experiment No. 8

Aim:

To design & test Wein Bridge Oscillator using IC 4A741C

Tools:

PSpice

Theory:

In Wein bridge oscillator, Wein bridge circuit is connected between the amplifier input terminals & output terminals.

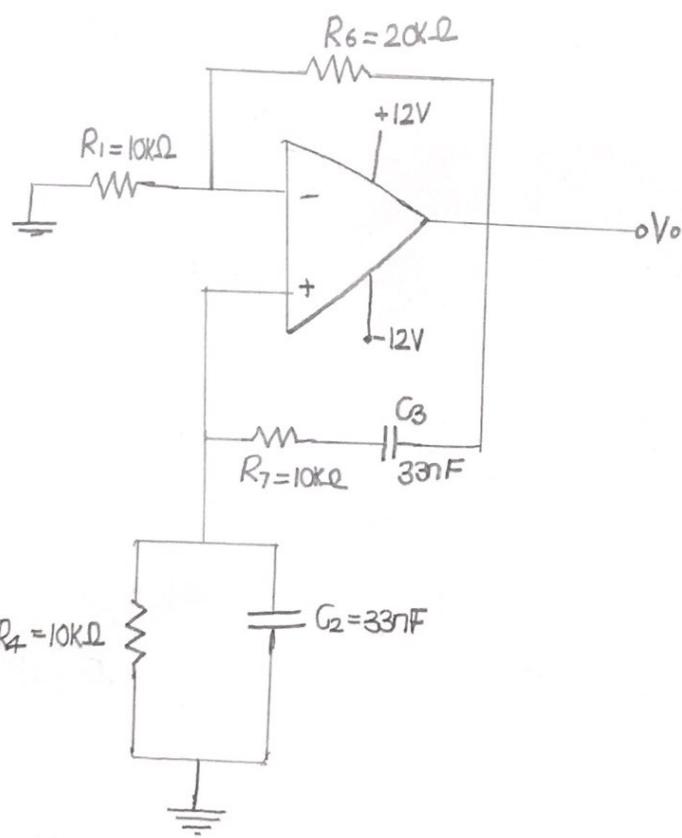
The bridge has a series RC network in one arm & parallel network in adjoining arm. In the remaining 2 arms of the bridge resistors  $R_i$  &  $R_f$  are connected.

To maintain oscillations total phase shift around the circuit must be zero & loop gain unity. First condition occurs only when the bridge is balanced. Assuming that the resistors & capacitors are equal in value, the Resonant Frequency of balanced bridge is given by,

$$f_o = \frac{1}{2\pi R C}$$

Conditions For Oscillation:

1. The gain required for sustained oscillations is given by



From the waveform the value of Time period ( $T$ ) is 2ms

$$A_F = 1 + \frac{R_6}{R_1}$$

$$\text{hence, } 3 = 1 + \frac{R_6}{R_1} \Rightarrow R_6 = 2R_1$$

Let  $R_1 = 10\text{ k}\Omega$  then  $R_6 = 20\text{ k}\Omega$

2. Oscillating Frequency,  $f_0 = 1/2\pi RC$

Assuming the values of some of the resistors and capacitors the value of frequency is calculated

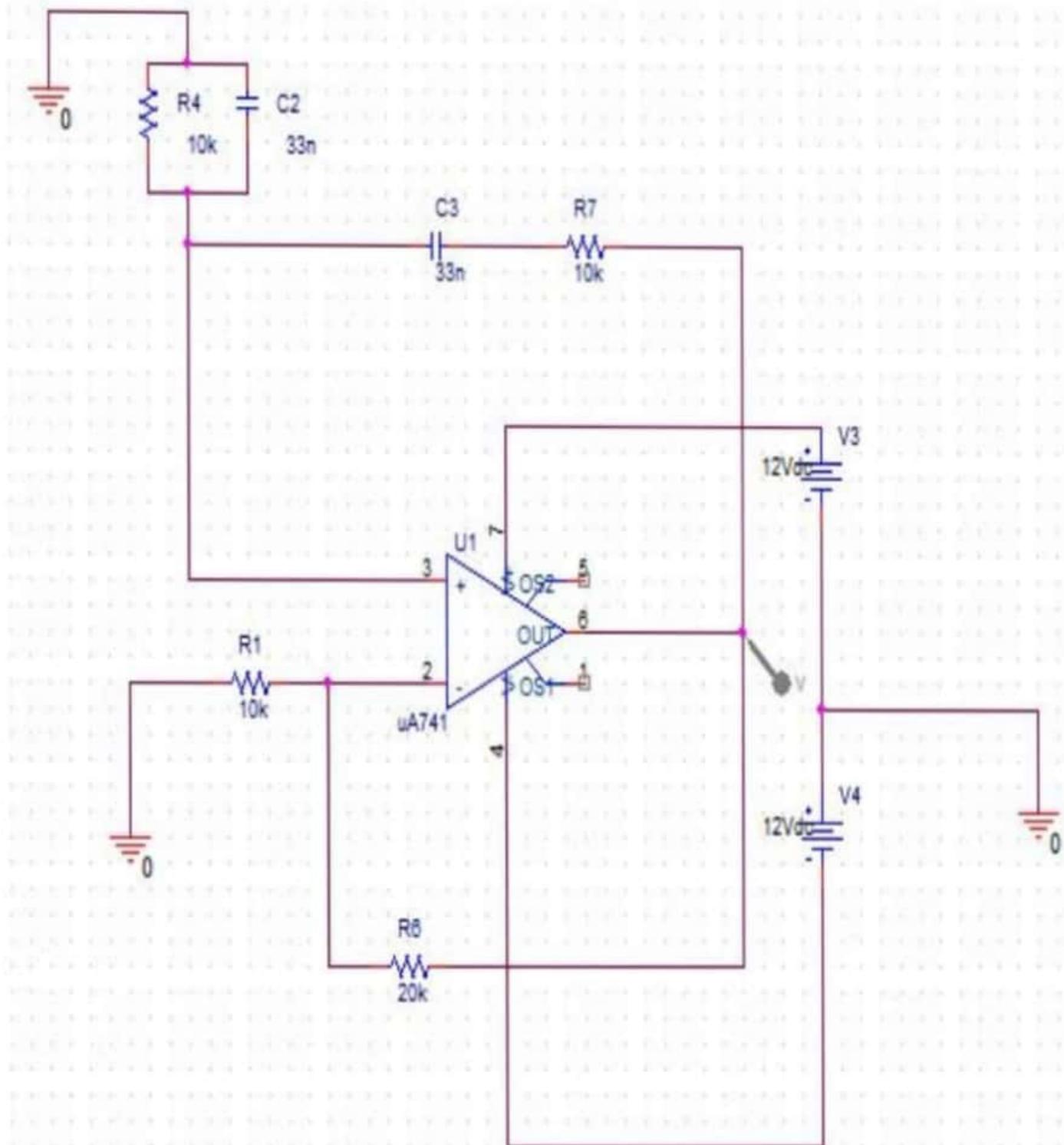
$$\text{Theoretical Value: } f = 1/(2 \times 3.14 \times 10 \times 10^3 \times 33 \times 10^{-9})$$

$$f = 482.5\text{ Hz}$$

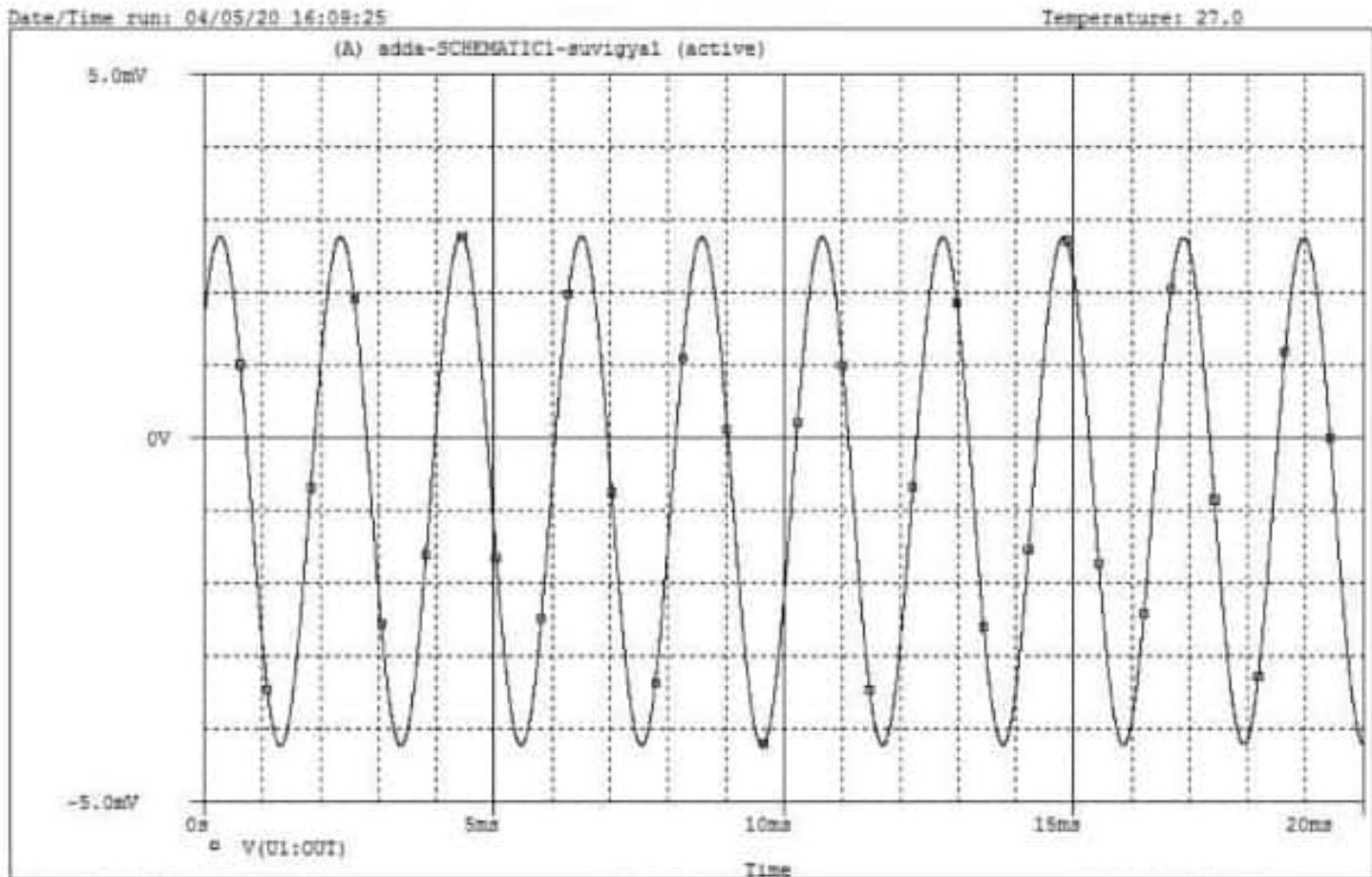
$$T = \frac{1}{f} = 2.07 \text{ ms}$$

Result:

Wein Bridge oscillator has been studied, the output waveform traced & theoretical time period & time period from output waveform were approximately equal.



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## Experiment No. 9

Aim :

To design & test astable multivibrator using IC 555

Tools:

Pspice

Theory :

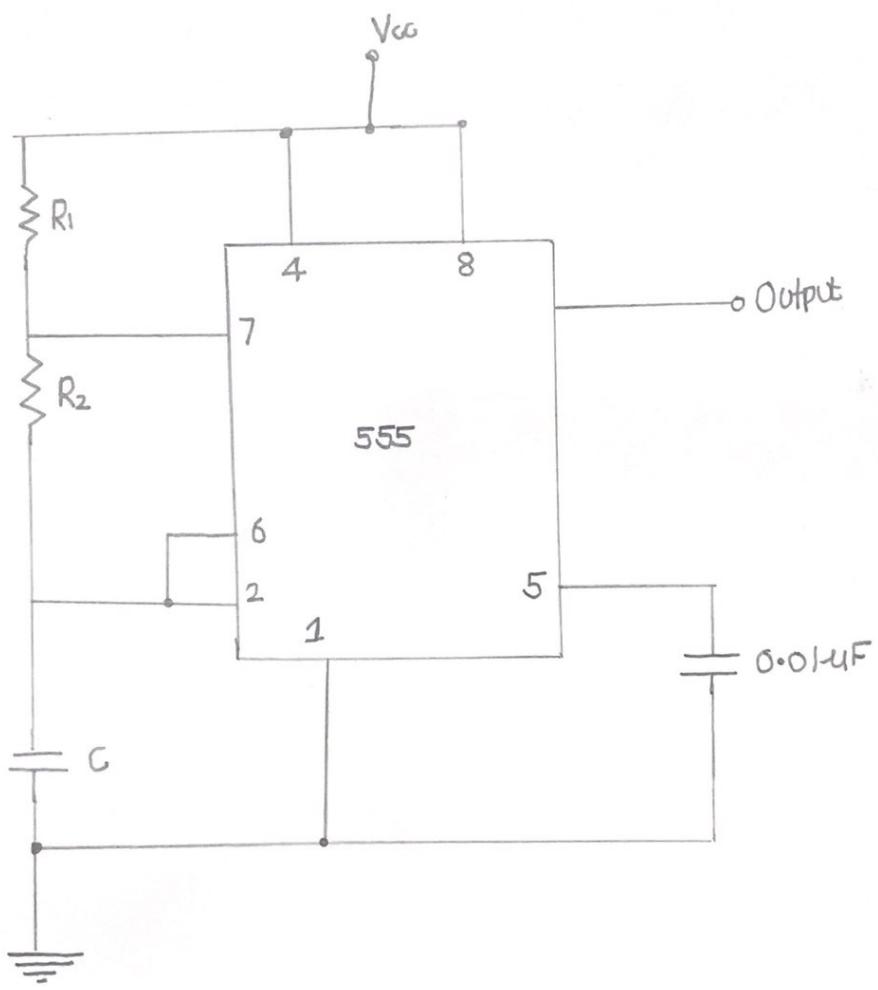
The 555 timer IC can be connected either in its monostable mode thereby producing a precision timer of a fixed time duration or its bistable mode to produce a Flip Flop type switching action. But we can also connect the 555 timer IC in an Astable mode to produce a very stable 555 oscillator circuit for generating highly accurate free running waveforms whose output frequency can be adjusted by means of an externally connected RC tank circuit consisting of just two resistors & a capacitor.

The individual times required completing one charge & discharge cycle of output is given as:

$$t_1 = 0.693(R_1 + R_2)C$$

$$t_2 = 0.693R_2C$$

$$T = t_1 + t_2$$



Calculations :

$$t_2 = 0.693 s$$

$$t_1 = 0.699 s$$

$$C = 10 \mu F$$

$$t_2 = 0.693 R_2 C$$

$$0.693 = 0.693 R_2 C$$

$$R_2 = 100 k\Omega$$

$$t_1 = 0.693 (R_1 + 100) C$$

$$0.699 = 0.693 (R_1 + 100) 10 \times 10^{-6}$$

$$R_1 = 1 k\Omega$$

$$T = t_1 + t_2$$

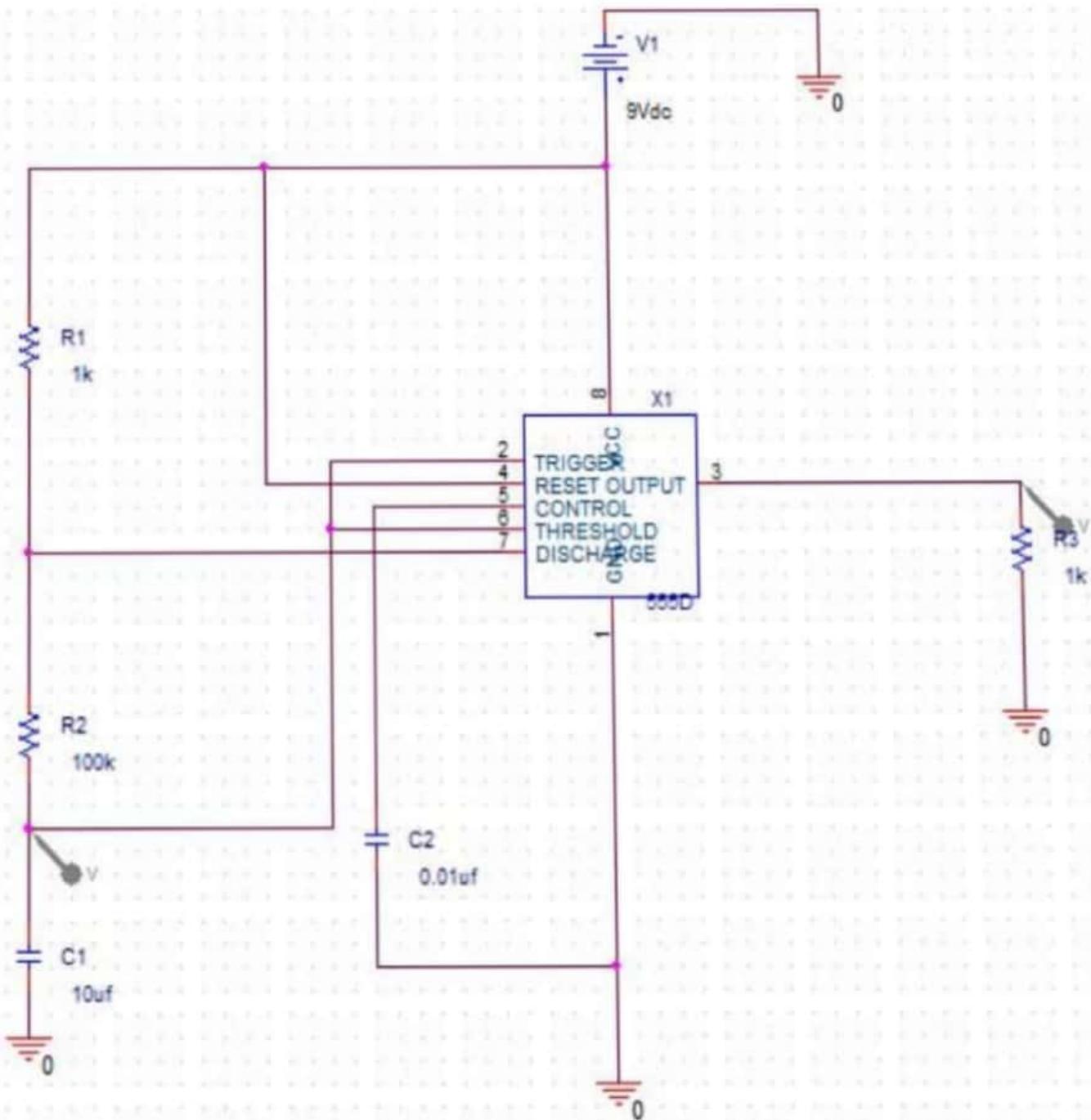
$$= (0.699 + 0.693)$$

$$= 1.392 s$$

From the waveform time period obtained is approximately equal to the theoretical value of time period.

Result:

Hence Astable vibrator using 555 is studied & its output waveform is traced.



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Temperature: 27.0

