

10 points for each problem except for Spice problem that is 60 points (total 200 points)

**6.1** The terminal voltages of various *npn* transistors are measured during operation in their respective circuits with the following results:

Case	E	B	C	Mode
1	0	0.7	0.7	
2	0	0.8	0.1	
3	-0.7	0	1.0	
4	-0.7	0	-0.6	
5	1.3	2.0	5.0	
6	0	0	5.0	

In this table, where the entries are in volts, 0 indicates the reference terminal to which the black (negative) probe of the voltmeter is connected. For each case, identify the mode of operation of the transistor.

**6.7** Consider an *npn* transistor whose base-emitter drop is 0.76 V at a collector current of 5 mA. What current will it conduct at  $v_{BE} = 0.70$  V? What is its base-emitter voltage for  $i_C = 5 \mu\text{A}$ ?

Assume "active"

**6.16** When operated in the active mode, a particular *npn* BJT conducts a collector current of 1 mA and has  $v_{BE} = 0.70$  V and  $i_B = 10 \mu\text{A}$ . Use these data to create specific transistor models of the form shown in Fig. 6.5(a) to (d).

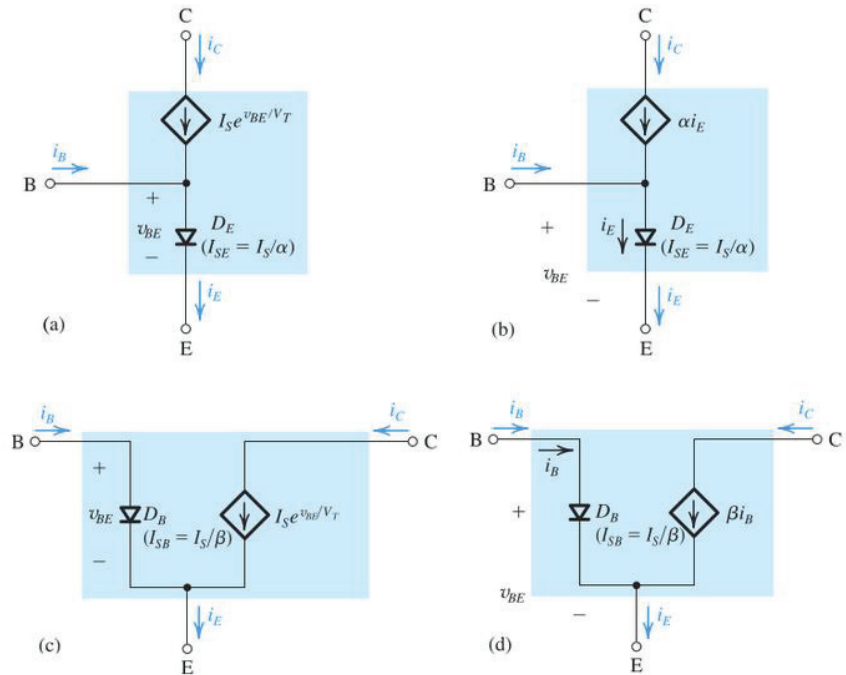


Figure 6.5 Large-signal equivalent-circuit models of the *npn* BJT operating in the forward active mode.

**6.28** For the circuits in Fig. P6.28, assume that the transistors have very large  $\beta$ . Some measurements have been made on these circuits, with the results indicated in the figure. Find the values of the other labeled voltages and currents.

only (a) and (d)

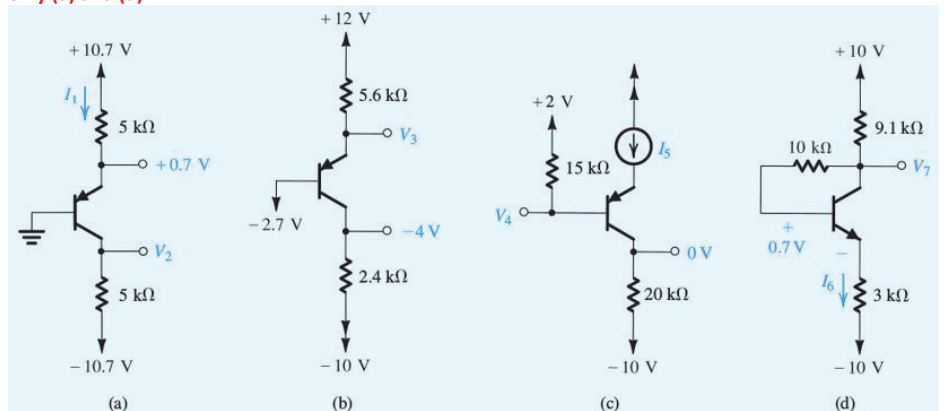


Figure P6.28

**6.35** For each of the circuits shown in Fig. P6.35, find the emitter, base, and collector voltages and currents. Use  $\beta = 50$ , but assume  $|V_{BE}| = 0.8 \text{ V}$  independent of current level.

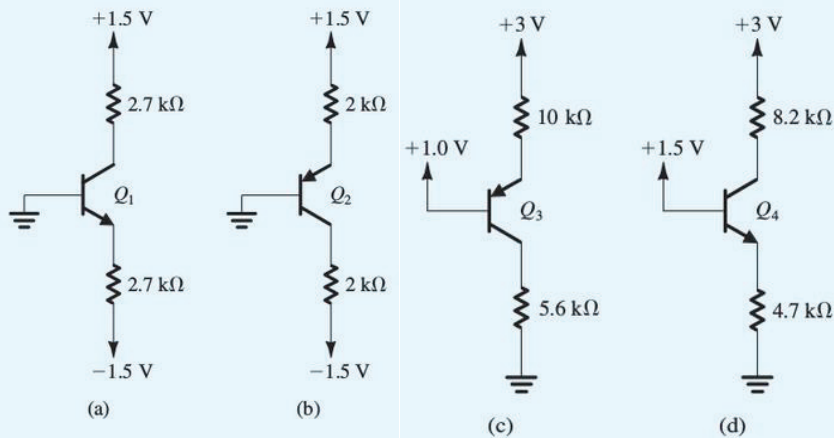


Figure P6.35

**6.59** In the circuit shown in Fig. P6.58, the transistor has  $\beta = 50$ . Find the values of  $V_B$ ,  $V_E$ , and  $V_C$ , and verify that the transistor is operating in the active mode. What is the largest value that  $R_C$  can have while the transistor remains in the active mode?

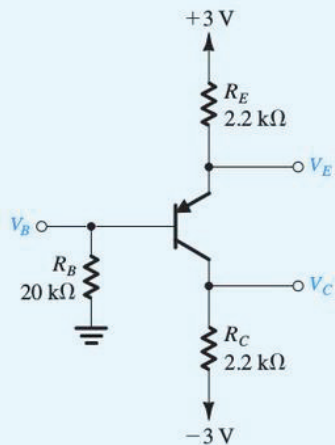


Fig. P6.58

**\*6.66** For the circuit shown in Fig. P6.66, find the labeled node voltages for:

(a)  $\beta = \infty$

(b)  $\beta = 100$

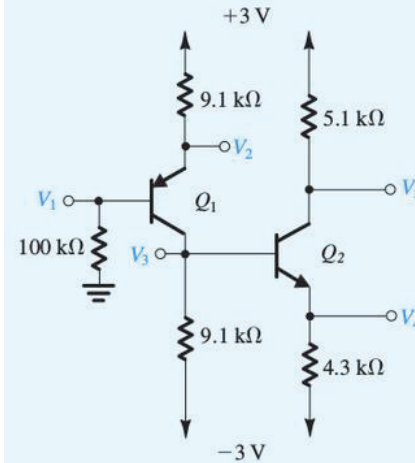


Fig. P6.66

**7.51** Figure P7.51 shows a transistor with the collector connected to the base. The bias arrangement is not shown. Since a zero  $v_{BC}$  implies operation in the active mode, the BJT can be replaced by one of the small-signal models of Figs. 7.24 and 7.26. Use the model of Fig. 7.26(b) and show that the resulting two-terminal device, known as a diode-connected transistor, has a small-signal resistance  $r$  equal to  $r_e$ .

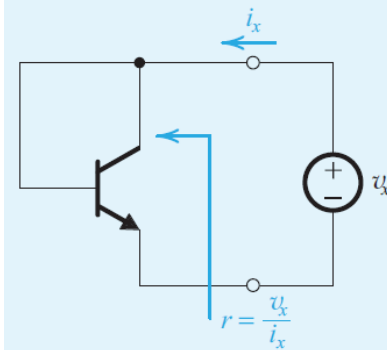


Figure P7.51

**7.54** In the circuit shown in Fig. P7.54, the transistor has a  $\beta$  of 200. What is the dc voltage at the collector? Replacing the BJT with one of the hybrid- $\pi$  models (neglecting  $r_o$ ), draw the equivalent circuit of the amplifier. Find the input resistances  $R_{ib}$  and  $R_{in}$  and the overall voltage gain ( $v_o/v_{sig}$ ). For an output signal of  $\pm 0.4$  V, what values of  $v_{sig}$  and  $v_b$  are required?

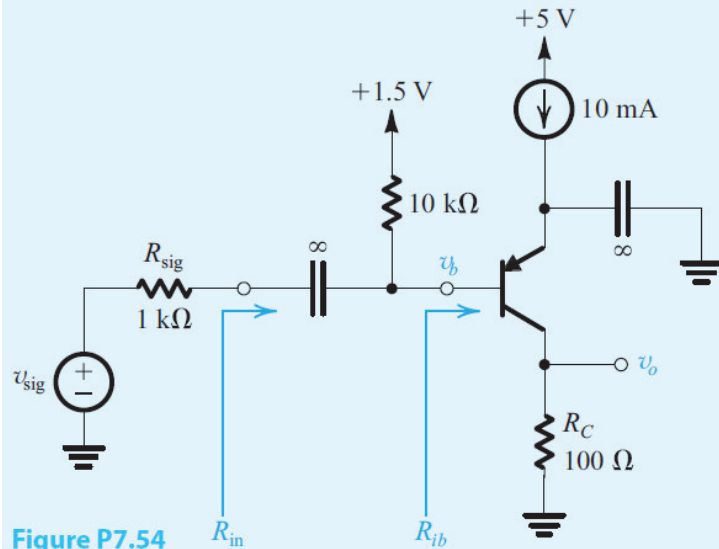


Figure P7.54

**8.49** Transistor  $Q_1$  in the circuit of Fig. P8.49 is operating as a CE amplifier with an active load provided by transistor  $Q_2$ , which is the output transistor in a current mirror formed by  $Q_2$  and  $Q_3$ . (Note that the biasing arrangement for  $Q_1$  is *not* shown.)

- Neglecting the finite base currents of  $Q_2$  and  $Q_3$  and assuming that their  $V_{BE} \simeq 0.7$  V and that  $Q_2$  has five times the area of  $Q_3$ , find the value of  $I$ .
- If  $Q_1$  and  $Q_2$  are specified to have  $|V_A| = 30$  V, find  $r_{o1}$  and  $r_{o2}$  and hence the total resistance at the collector of  $Q_1$ .

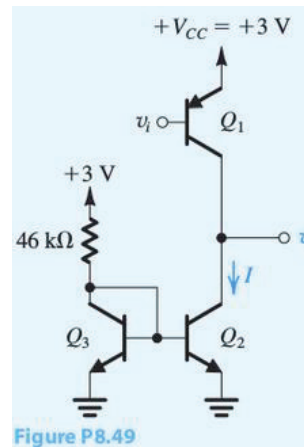


Figure P8.49

- Find  $r_{\pi 1}$  and  $g_{m1}$  assuming that  $\beta_1 = 50$ .
- Find  $R_{in}$ ,  $A_v$ , and  $R_o$ .

**D 8.85** (a) The circuit in Fig. P8.85 is a modified version of the Wilson current mirror. Here the output transistor is “split” into two matched transistors,  $Q_3$  and  $Q_4$ . Find  $I_{O1}$  and  $I_{O2}$  in terms of  $I_{REF}$ . Assume all transistors to be matched with current gain  $\beta$ .

(b) Use this idea to design a circuit that generates currents of 0.1 mA, 0.2 mA, and 0.4 mA, using a reference current source of 0.7 mA. What are the actual values of the currents generated for  $\beta = 50$ ?

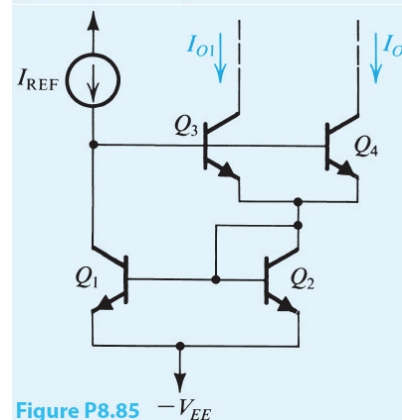


Figure P8.85

**8.96** The transistors in the circuit of Fig. P8.96 have  $\beta = 100$  and  $V_A = 50$  V.



- (a) Find  $R_{in}$  and the overall voltage gain.  
 (b) What is the effect of increasing the bias currents by a factor of 10 on  $R_{in}$ ,  $G_v$ , and the power dissipation?

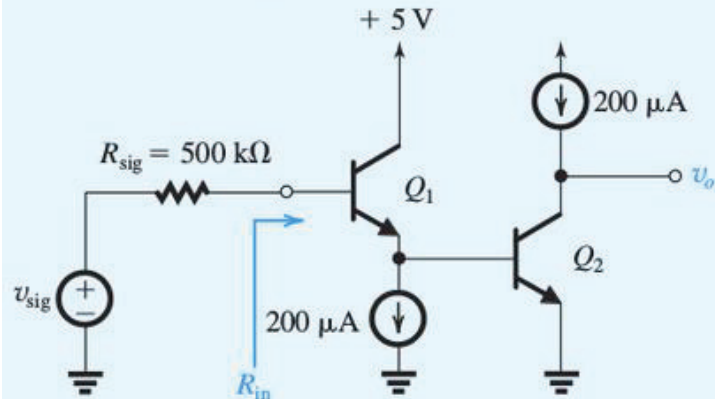


Figure P8.96

**8.98** The BJTs in the Darlington follower of Fig. P8.98 have  $\beta = 100$ . If the follower is fed with a source having a 100-k $\Omega$  resistance and is loaded with 1 k $\Omega$ , find the input resistance and the output resistance (excluding the load). Also find the overall voltage gain, both open-circuited and with load. Neglect the Early effect.

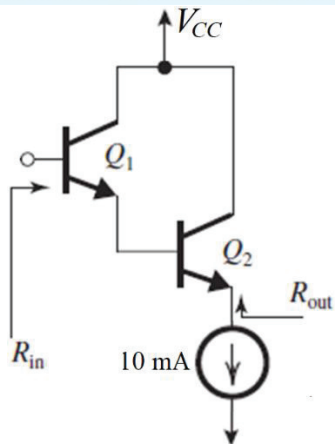


Fig. P8.98

**\*\*8.101** In each of the six circuits in Fig. P8.101, let  $\beta = 100$ , and neglect  $r_o$ . Calculate the overall voltage gain.

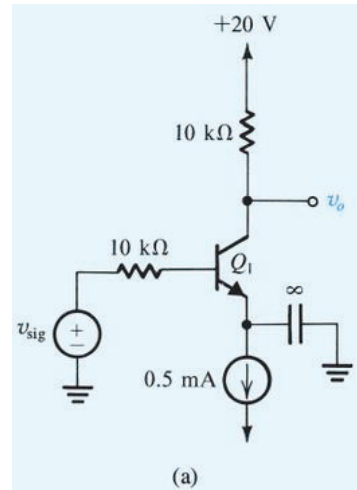
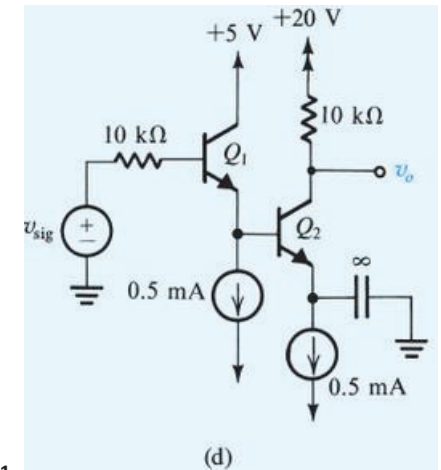


Fig. P8.101



### Spice Problem (60%)

Q1-Q3 are all identical with a BJT model whose Spice parameters are given below

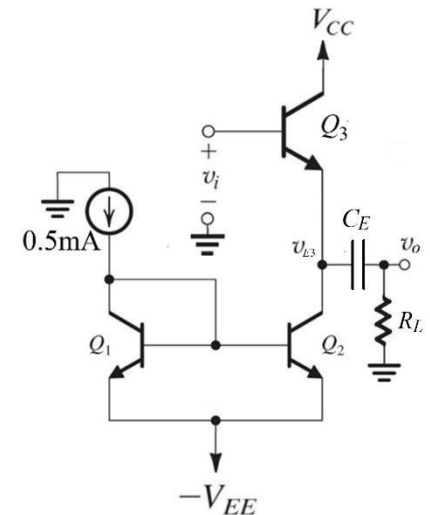
$IS=1fA$ ,  $NF=1.1$ ,  $NR = 1.5$ ,  $BF=50$ ,  $BR = 10$ ,  $VA_F=20V$ ,  $VAR = 10V$ ,  $CJE=0.3nf$ ,  $CJC=0.2nf$ ,  $CJS=0.5nf$ ,  $TR=1ns$  and  $TF = 1ns$

where the ideality factors for BE and BC junctions are NF and NR, Early voltage  $V_A = VAF$ , the common emitter current gain  $\beta \approx BF$

$V_{CC}=V_{EE} = 3.5V$ ,  $R_L = 10k\Omega$  and  $C_E = 1\mu F$

1. Perform Spice simulation with an ac input and use .ac to sweep the frequency from 10Hz to 100MHz (100meg in Spice). Plot  $v_o/v_i$ . Answer the following questions

- (a) Theoretically calculate the voltage gain compared to the midband gain observed in Spice simulation, where in the midband the gain is nearly constant and the phase difference between  $v_o$  and  $v_i$  is nearly  $0^\circ$ .



- (b) Change  $C_E$  to 10 $\mu$ F and plot  $v_o/v_i$  using the same scale/division as that in (a) on the vertical and horizontal axes. Discuss why these 2 Bode plots are different in the low frequency? Use the  $R_{eq}C_E$  time constant to estimate the lower corner frequency  $f_{3dB}$  compared with the observed value in Spice,  $R_{eq}C_E = \omega_{3dB} \approx 2\pi f_{3dB}$  for a single pole circuit.
2. Perform Spice simulation with  $C_E = 1\mu$ F and input voltage  $v_i(t) = 2V\sin(\omega t)$  at 10kHz. Plot  $v_i$ ,  $v_C$  and  $v_o$ . Answer the following questions
- (a) Explain how this follower works. Why doesn't  $v_C$  follows  $v_i$ ?  
Theoretically find the relation between  $v_C$  follows  $v_i$ , compared with the observation in Spice.
- (b) Change  $C_E$  to 0.01 $\mu$ F and observe the change(s). explain your findings qualitatively.

### SPICE information on BJTs

- The BJT element statement in SPICE:  
**qxxx C B E modelname**  
qxxx is the BJT device name that must start with q. xxx could be any character or number. *modelname* is the name of the BJT model that appears in the model statement. C, B, E are the node numbers for the collector, base, and emitter, respectively
- The npn BJT model statement  
**.model modelname npn values of parameters**
- The pnp BJT model statement  
**.model modelname pnp values of parameters**  
*modelname* is the name of the BJT model that also appears in the BJT element statement. Many transistors may use the same model name, meaning these BJT are identical