

# Information Theory (III)

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07/02/2014

# 1 The Shannon Theorems

## 2 Summary

# Shannon Source Coding Theorem

- Source coding theorem determines a limit to possible data compression.
- Source entropy is related to the analysis of source coding theorem.
  - Assuming DMS emits a large number of symbols taken from an alphabet  $A = \{x_1, x_2, \dots, x_M\}$  in the form of a sequence of  $n_f$  symbols.
  - Priori probability of each symbol is  $P(x_i)$ ,  $i = 1, \dots, M$  and there is

$$\sum_i^M P(x_i) = 1.$$

- A particular sequence  $\mathbf{s} = s_1 s_2 \dots s_{n_f}$  with probability

$$P(s_1 s_2 \dots s_{n_f}) = P(s_1) P(s_2) \dots P(s_{n_f})$$

as the symbols are statistically independent from each other.

# Shannon Source Coding Theorem

- Consider a very long sequence  $\mathbf{s}$ . Typically, in sequence  $\mathbf{s}$  the symbol  $x_1$  will appear  $\approx n_f P(x_1)$  times, symbols  $x_2$  will appear  $\approx n_f P(x_2)$  times,  $\dots$ , symbols  $x_M$  will appear  $\approx n_f P(x_M)$  times.
- Hence, the probability of such **typical sequence** is roughly

$$P(\mathbf{s}) \approx P_{typ} = P(x_1)^{n_f P(x_1)} \dots P(x_M)^{n_f P(x_M)} = \prod_{i=1}^M [P(x_i)]^{n_f P(x_i)}$$

- It can prove that  $P(\mathbf{s}) \approx 2^{-n_f H(X)}$ . Hence there are  $2^{n_f H(X)}$  typical sequences.
- **Typical sequences** are those with the maximum probability of being emitted by the information source.
- **Non-typical sequences** are those with very low probability of occurrence.

# Shannon Source Coding Theorem

- Even though there are the total  $M^{n_f}$  possible sequences which can be emitted by information source alphabet  $A = \{x_1, x_2, \dots, x_M\}$ , ONLY  $2^{n_f H(X)}$  sequences have a significant probability of occurrence.
- When assuming that only  $2^{n_f H(X)}$  sequences are transmitted instead of the total possible number of them, an error of magnitude  $\varepsilon$  is made. The introduced error can be arbitrary small if  $n_f \rightarrow \infty$ .
- It means that the source information can be transmitted using a significant lower number of sequences than the total possible number of them.
- If only  $2^{n_f H(X)}$  sequences are to be transmitted and using a binary format of representation information, there will be  $n_f H(X)$  bits needed for representing this information.
- So each symbol can be represented by  $H(X)$  bits.

# Shannon Source Coding Theorem

- For a  $M$ -ary DMS emitting equally likely symbols, there is

$$H(X) = \log_2 M$$

- then

$$2^{n_f H(X)} = 2^{n_f \log_2 M} = M^{n_f}$$

- In this case, the number of the typical sequences for a DMS with equally likely symbols is equal to the maximum possible number of sequences that this source can emit.
- For a DMS with independent symbols, compression of the information is possible only if the symbols of this source are not equally likely.

# Source coding example

- **Example:** Let  $A$  be a 4-ary source  $\{a_0, a_1, a_2, a_3\}$ , we can use two binary digits to represent each source symbol. If we know that the probabilities of each symbol as follows. What is the entropy of the source?

$$P(a_0) = 0.5 \quad P(a_1) = 0.3 \quad P(a_2) = 0.15 \quad P(a_3) = 0.05$$

- **Solution:**

$$H(A) = \sum_{i=0}^3 P(a_i) * \log_2 \frac{1}{P(a_i)} = 1.6477$$

- The efficiency of the uncoded source is  $H(A)/2 = 0.82385$

# Source coding example

- Instead of using 2 bits for each symbol, we can encode the source by

$$P(a_0) = 0.5 \quad C(a_0) \rightarrow 0$$

$$P(a_1) = 0.3 \quad C(a_1) \rightarrow 10$$

$$P(a_2) = 0.15 \quad C(a_2) \rightarrow 110$$

$$P(a_3) = 0.05 \quad C(a_3) \rightarrow 111$$

What's the average number of bits in the new coded word?

- **Solution:**

$$\bar{L} = \sum_{i=0}^3 P(a_i) L(a_i) = .5(1) + .3(2) + .15(3) + .05(3) = 1.70$$

- The efficiency of the coded source is  $H(A)/\bar{L} = 0.96924$



# Source coding summary

- For a DMS emitting an alphabet  $A = \{x_1, x_2, \dots, x_M\}$ 
  - The arbitrary information sources can have a considerable range of possible entropies.  $0 \leq H(X) \leq \log_2(M)$ ;
  - The entropy of a source is the average information carried per symbol;
  - As there is a cost to transmit or store each symbol, it is desirable to use minimum number of symbols to represent the information.
  - It is possible to compress the information provided the source only if the symbols of this source are not equally likely, i.e.,  $H(X) < \log_2(M)$ .

# Prefix codes and instantaneous Decoding

- Let's look at this sequence of letters:
  - IFIWANTEDTOPICKONE
  - IF I WANTED TO PICK ONE vs. IF I WANT ED TO PICK ONE
- The English language is not generally self-punctuating.
- **Prefix code** is a code that has the property of being self-punctuating.
  - It has punctuation built into the structure.
  - It is accomplished by designing the code such that no codeword is a prefix of another (longer) codeword.
  - It is instantaneously decodable.

- **Example:** Let the encoded map pairs of symbols into the codewords shown below. Please decode the sequence: 1000001111111011101, assuming the codewords are transmitted bit serially from left to right.

$\langle x_i, x_j \rangle$	$P(x_i, x_j)$	$b_m$	$\langle x_i, x_j \rangle$	$P(x_i, x_j)$	$b_m$
$x_1 x_1$	.25	00	$x_3 x_1$	.075	1101
$x_1 x_2$	.15	100	$x_3 x_2$	.045	0111
$x_1 x_3$	.075	1100	$x_3 x_3$	.0225	111110
$x_1 x_4$	.025	11100	$x_3 x_4$	.0075	1111110
$x_2 x_1$	.15	101	$x_4 x_1$	.025	11101
$x_2 x_2$	.09	010	$x_4 x_2$	.015	111101
$x_2 x_3$	.045	0110	$x_4 x_3$	.0075	11111110
$x_2 x_4$	.015	111100	$x_4 x_4$	.0025	11111111

- **Solution:**

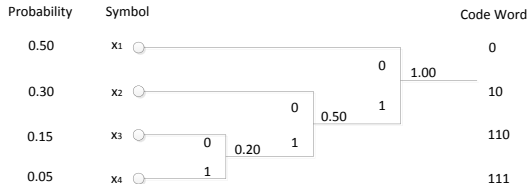
- 100, 00, 0111, 111110, 11101;
- which decodes as  $x_1 x_2 x_1 x_1 x_3 x_2 x_3 x_3 x_4 x_1$ .

# Construct a Huffman code example

- **Example:** Construct a Huffman code for the 4-ary source alphabet  $x_1, x_2, x_3, x_4$  with probability

$$P(x_1) = 0.5 \quad P(x_2) = 0.3 \quad P(x_3) = 0.15 \quad P(x_4) = 0.05$$

- The constructed Huffman tree:
  - Start from small probabilities
  - Form a tree
  - Assign 0 to higher branch, 1 to lower branch



# Huffman coding

- Huffman codes are lossless data compression codes;
- Huffman codes are widely used in data communications, speech coding, and video or graphical image compression;
- Huffman codes can deliver codeword sequences which asymptotically approach the source entropy.
- Huffman codes generally have variable length codewords.
- Huffman codes belong to prefix codes.

# Preview of Channel coding theorem

- Taking BSC channel as example here. The input  $\mathbf{X}$  and the output  $\mathbf{Y}$  are the sequences of  $n$  symbols.
- With error probability  $p$  of BSC, the output  $\mathbf{Y}$  will differ in  $np$  positions comparing with  $\mathbf{X}$ .
- The number of sequence of  $n$  symbols with difference in  $np$  position is  $\binom{n}{np} \approx 2^{n\Omega(p)}$ .
- This result indicates that for each input sequence of  $n$  symbols, there exists  $2^{n\Omega(p)}$  possible output sequences as a result of the errors introduced by the channel.
- The total number of the typical output sequences of this channel  $2^{nH(Y)}$ .
- Use only uniquely distinguishable input sequences, i.e., divide the set of  $2^{nH(Y)}$  sequences into disjoint sets of size  $2^{n\Omega(p)}$ , the maximum number of such sets is

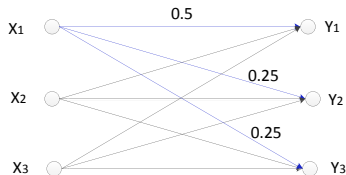
$$\frac{2^{nH(Y)}}{2^{n\Omega(p)}} = 2^{n[H(Y)-\Omega(p)]} = 2^{nI(X;Y)}$$

- This derivation outlines an upper bound on the capacity.

# Channel coding theorem

- **Channel Coding Theorem:** If the transmission rate  $R$  fits the condition  $R < C$ , then for an arbitrary value  $\varepsilon > 0$ , there exists a code with block length  $n$  that makes the error probability of the transmission be less than  $\varepsilon$ . If  $R > C$  then there is no guarantee of reliable transmission; namely there is no guarantee that the arbitrary value of  $\varepsilon$  is a bound for the error probability.

# Example 1.11: Calculate Channel capacity



- **Example 1.11:** Determine the channel capacity of the channel whose model is shown above. Assuming the input symbols are equally likely, and the channel transition probabilities are listed below.

$$\begin{aligned}
 P(y_1/x_1) &= P(y_2/x_2) = P(y_3/x_3) = 0.5 \\
 P(y_1/x_2) &= P(y_1/x_3) = 0.25 \\
 P(y_2/x_1) &= P(y_2/x_3) = 0.25 \\
 P(y_3/x_1) &= P(y_3/x_2) = 0.25
 \end{aligned}$$

Calculate the channel capacity.



# Shannon's Channel Capacity Theorem

- Channel Capacity in bits per second of additive Gaussian noise (AWGN) channel:

$$C = B \log_2 \left( 1 + \frac{S}{N_0 B} \right)$$

- where  $B$  is the bandwidth,  $S$  is the signal power,  $N_0$  is the power density of the noise;
- The channel capacity,  $C$  increases as the available bandwidth  $B$  increases and as the signal to noise ratio  $\frac{S}{N}$  increases (improves).
- In a single formula it highlights the interplay between 3 key system parameters:
  - channel bandwidth
  - average received signal power
  - noise power at the channel output

# Shannon Limit

- Channel Capacity in bits per second of additive Gaussian noise (AWGN) channel:

$$C = B \log_2 \left( 1 + \frac{S}{N_0 B} \right)$$

- where  $B$  is the bandwidth,  $S$  is the signal power,  $N_0$  is the power density of the noise;
- Rewrite the above equation using  $E_b$  average energy per bit:

$$\frac{C}{B} = \log_2 \left( 1 + \frac{E_b}{N_0} \frac{C}{B} \right)$$

- There is  $\lim_{x \rightarrow 0} (1 + x)^{1/x} = e$ ;
- Based on  $\log_2(1 + x) = x \frac{1}{x} \log_2(1 + x) = x \log_2 [(1 + x)^{1/x}]$ ,

$$\begin{aligned} \frac{C}{B} &= \frac{E_b}{N_0} \frac{C}{B} \log_2 \left( 1 + \frac{E_b}{N_0} \frac{C}{B} \right)^{N_0 B / C E_b} \\ \Rightarrow 1 &= \frac{E_b}{N_0} \log_2 \left( 1 + \frac{E_b}{N_0} \frac{C}{B} \right)^{N_0 B / C E_b} \end{aligned}$$

- If  $\frac{C}{B} \rightarrow 0$ ,

$$\frac{E_b}{N_0} = \frac{1}{\log_2(e)} = 0.6931$$

$$\left( \frac{E_b}{N_0} \right)_{dB} = 10 \log_{10}(0.6931) = -1.59 \text{ dB}$$

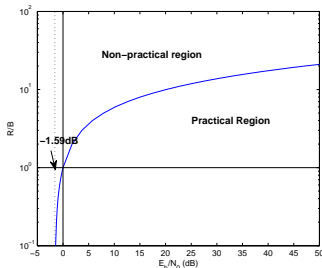
- -1.59dB is Shannon limit.
- It means that if the ratio  $E_b/N_0$  is kept slightly higher than -1.59dB, it is possible to have error-free transmission by sophisticated coding technique.
- In other words, if the ratio  $E_b/N_0$  is below -1.59dB. It is impossible to have error-free transmission, no matter how sophisticated coding is used.

$$C = B \log_2 \left( 1 + \frac{E_b}{N_0} \frac{C}{B} \right)$$

Rewrite the equation into:

$$2^{C/B} = 1 + \frac{E_b}{N_0} \left( \frac{C}{B} \right)$$

Let  $R = C$ ,



- The curve shows when  $R/B \rightarrow 0$ ,  $E_b/N_0$  reaches Shannon limit.
- For each value of  $R/B$ , there exists a corresponding bound which can be read from the curve.

# Summary of Information Theory

- Information,  $I_i$ .
- Entropy,  $H(X)$ .
- Information channels, e.g., BSC, BEC. And their corresponding channel model.
- Different probabilities, i.e.,  $p(x_i)$ ,  $p(y_j)$ ,  $p(x_i, y_j)$ ,  $p(y_j/x_i)$ ,  $p(x_i/y_j)$ .
- Different Entropies, i.e.,  $H(X)$ ,  $H(Y)$ ,  $H(X, Y)$ ,  $H(Y/X)$ ,  $H(X/Y)$ .
- Mutual information  $I(X; Y)$  and its properties.
- Shannon source coding theorem.
- Shannon channel coding theorem
- Shannon limit.