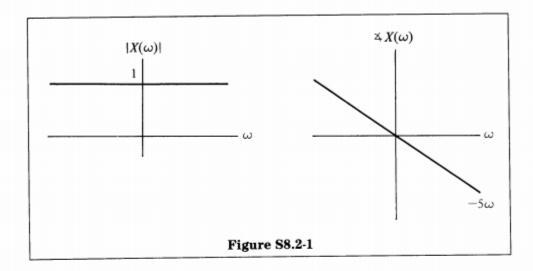
(a)
$$X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt = \int_{-\infty}^{\infty} \delta(t-5)e^{-j\omega t} dt = e^{-j5\omega} = \cos 5\omega - j\sin 5\omega$$
,

by the sifting property of the unit impulse.



(b)
$$X(\omega) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-j\omega t} dt = \int_{0}^{\infty} e^{-at} e^{-j\omega t} dt$$
$$= \int_{0}^{\infty} e^{-(a+j\omega)t} dt = \frac{-1}{a+j\omega} e^{-(a+j\omega)t} \Big|_{0}^{\infty}$$

Since $Re\{a\} > 0$, e^{-at} goes to zero as t goes to infinity. Therefore,

$$X(\omega) = \frac{-1}{a + j\omega} (0 - 1) = \frac{1}{a + j\omega},$$

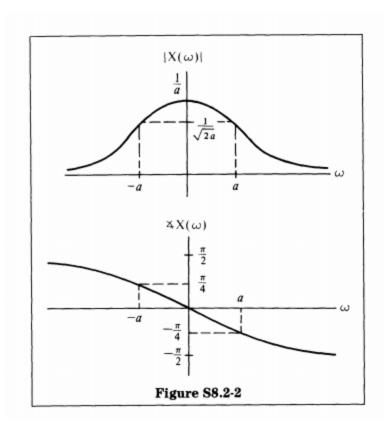
$$|X(\omega)| = [X(\omega)X^*(\omega)]^{1/2} = \left[\frac{1}{a + j\omega} \left(\frac{1}{a - j\omega}\right)\right]^{1/2} \frac{1}{\sqrt{a^2 + \omega^2}},$$

$$Re\{X(\omega)\} = \frac{X(\omega) + X^*(\omega)}{2} = \frac{a}{a^2 + \omega^2},$$

$$Im\{X(\omega)\} = \frac{X(\omega) - X^*(\omega)}{2} = \frac{-\omega}{a^2 + \omega^2},$$

$$A(\omega) = \tan^{-1}\left[\frac{Im\{X(\omega)\}}{Re\{X(\omega)\}}\right] = -\tan^{-1}\frac{\omega}{a}$$

The magnitude and angle of $X(\omega)$ are shown in Figure S8.2-2.



Ex. b

$$\chi(t) := \frac{b}{t^{2}+b}$$
If $\chi(j\omega) \approx the F.T. of $\chi(t)$, we have,

$$\chi(t) := \frac{1}{2\pi} \int_{-\infty}^{\infty} \chi(j\omega) e^{j\omega t} d\omega$$

$$\frac{d}{dt} \chi(t) := \frac{1}{2\pi} \int_{-\infty}^{\infty} j \omega . \chi(j\omega) e^{j\omega t} d\omega$$

$$:= \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega . \chi(j\omega) e^{j\omega t} d\omega$$

$$:= \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega . \chi(j\omega) e^{j\omega t} d\omega$$

$$:= \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega . \chi(j\omega) e^{j\omega t} d\omega$$

$$:= \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega . \chi(j\omega) e^{j\omega t} d\omega$$

$$:= \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega . \chi(j\omega) d\omega$$$

S20.5

There are two ways to solve this problem.

Method 1

This method is based on recognizing that the system input is a superposition of eigenfunctions. Specifically, the eigenfunction property follows from the convolution integral

$$y(t) = \int_{-\infty}^{+\infty} h(\tau)x(t-\tau) d\tau$$

Now suppose $x(t) = e^{at}$. Then

$$y(t) = \int_{-\infty}^{\infty} h(\tau)e^{a(t-\tau)} d\tau = e^{at} \int_{-\infty}^{\infty} h(\tau)e^{-a\tau} d\tau$$

Now we recognize that

$$\int_{-\infty}^{\infty} h(\tau)e^{-a\tau} d\tau = H(s)\bigg|_{s=a},$$

so that if $x(t) = e^{at}$, then

$$y(t) = \left[\left. H(s) \right|_{s=a} \right] e^{at},$$

i.e., e^{at} is an eigenfunction of the system.

Using linearity and superposition, we recognize that if

$$x(t) = e^{-t/2} + 2e^{-t/3},$$

then

$$y(t) = e^{-t/2}H(s)\bigg|_{s=-1/2} + 2e^{-t/3}H(s)\bigg|_{s=-1/3}$$

so that

$$y(t) = 2e^{-t/2} + 3e^{-t/3}$$
 for all t.

Method 2

We consider the solution of this problem as the superposition of the response to two signals $x_1(t)$, $x_2(t)$, where $x_1(t)$ is the noncausal part of x(t) and $x_2(t)$ is the causal part of x(t). That is,

$$x_1(t) = e^{-t/2}u(-t) + 2e^{-t/3}u(-t),$$

$$x_2(t) = e^{-t/2}u(t) + 2e^{-t/3}u(t)$$

This allows us to use Laplace transforms, but we must be careful about the ROCs. Now consider $\mathcal{L}\{x_1(t)\}$, where $\mathcal{L}\{\cdot\}$ denotes the Laplace transform:

$$\mathcal{L}\{x_1(t)\} = X_1(s) = -\frac{1}{s+\frac{1}{2}} - \frac{2}{s+\frac{1}{3}}, \quad Re\{s\} < -\frac{1}{2}$$

Now since the response to $x_1(t)$ is

$$y_1(t) = \mathcal{L}^{-1}\{X_1(s)H(s)\},$$

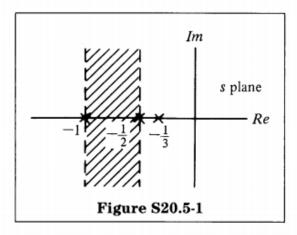
then

$$\begin{split} Y_{\mathrm{I}}(s) &= -\frac{1}{(s+1)(s+\frac{1}{2})} - \frac{2}{(s+\frac{1}{3})(s+1)}\,, \qquad -1 < Re\{s\} < -\frac{1}{2}\,, \\ &= \frac{2}{s+1} + \frac{-2}{s+\frac{1}{2}} + \frac{-3}{s+\frac{1}{3}} + \frac{3}{s+1}\,, \\ &= \frac{5}{s+1} - \frac{2}{s+\frac{1}{2}} - \frac{3}{s+\frac{1}{3}}\,, \end{split}$$

so

$$y_1(t) = 5e^{-t}u(t) + 2e^{-t/2}u(-t) + 3e^{-t/3}u(-t)$$

The pole-zero plot and associated ROC for $Y_1(s)$ is shown in Figure S20.5-1.



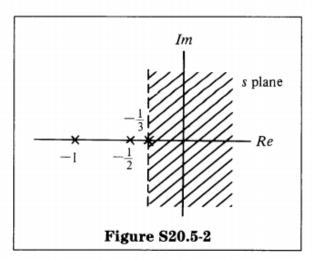
Next consider the response $y_2(t)$ to $x_2(t)$:

$$\begin{split} x_2(t) &= e^{-t/2}u(t) + 2e^{-t/3}u(t), \\ X_2(s) &= \frac{1}{s + \frac{1}{2}} + \frac{2}{s + \frac{1}{3}}, \quad Re\{s\} > -\frac{1}{3}, \\ Y_2(s) &= X_2(s)H(s) = \frac{1}{(s + \frac{1}{2})(s + 1)} + \frac{2}{(s + \frac{1}{3})(s + 1)}, \\ Y_2(s) &= \frac{2}{s + \frac{1}{2}} + \frac{-2}{s + 1} + \frac{3}{s + \frac{1}{2}} + \frac{-3}{s + 1}, \end{split}$$

so

$$y_2(t) = -5e^{-t}u(t) + 2e^{-t/2}u(t) + 3e^{-t/3}u(t)$$

The pole-zero plot and associated ROC for $Y_2(s)$ is shown in Figure S20.5-2.



Since $y(t) = y_1(t) + y_2(t)$, then

$$y(t) = 2e^{-t/2} + 3e^{-t/3}$$
 for all t

(b) Calculating the Laplace transform of both sides, we obtain

$$\left[s^{2}Y(s) - s \underbrace{y(0^{-})}_{=0} - \underbrace{\dot{y}(0^{-})}_{=0} \right] + 4\left[sY(s) - \underbrace{y(0^{-})}_{=0} \right] + 4Y(s) = \frac{1}{s}$$

or,
$$(s^2 + 4s + 4)Y(s) = \frac{1}{s}$$
 or $Y(s) = \frac{1}{s(s+2)^2} = \frac{A}{s} + \frac{B}{s+2} + \frac{C}{(s+2)^2}$,

where the partial fraction coefficients are calculated as

$$A = \frac{1}{(s+2)^2} \Big|_{s=0} = \frac{1}{4}$$
, and $C = \left[\frac{1}{s}\right]_{s=-2} = -\frac{1}{2}$

Expanding Y(s), and comparing the numerator of both sides, we get

$$1 = A(s+2)^{2} + Bs(s+2) + Cs$$
$$= (A+B)s^{2} + (4A+2B+C) + 4A$$

Comparing the coefficients of s^2 in both sides, we get (A + B) = 0 or B = -1/4.

In other words,
$$Y(s) = \frac{1}{s(s+2)^2} = \frac{1/4}{s} - \frac{1/4}{s+2} - \frac{1/2}{(s+2)^2}$$

Calculating the inverse Laplace transform of Y(s) yields

$$y(t) = \frac{1}{4} \left[1 - e^{-2t} - 2te^{-2t} \right] u(t) = \frac{1}{4} \left[1 - (2t+1)e^{-2t} \right] u(t)$$
.

(c) Calculating the Laplace transform of both sides, we obtain

$$\left[s^{2}Y(s) - s \underbrace{y(0^{-})}_{=1} - \underbrace{\dot{y}(0^{-})}_{=1} \right] + 6 \left[sY(s) - \underbrace{y(0^{-})}_{=1} \right] + 8Y(s) = \frac{1}{(s+3)^{2}}$$

or,
$$(s^2 + 6s + 8)Y(s) = \frac{1}{(s+3)^2} + (s+1+6)$$
 or $Y(s) = \frac{1}{(s+2)(s+3)^2(s+4)} + \frac{s+7}{(s+2)(s+4)}$.

Calculating the partial fraction expansion of the two terms separately, we obtain

$$\frac{1}{(s+2)(s+3)^2(s+4)} = \frac{1/2}{s+2} + \frac{0}{s+3} - \frac{1}{(s+3)^2} - \frac{1/2}{s+4}$$
and
$$\frac{s+7}{(s+2)(s+4)} = \frac{5/2}{s+2} - \frac{3/2}{s+4}$$

Expanding Y(s) as

$$Y(s) = \frac{1/2}{s+2} - \frac{1}{(s+3)^2} - \frac{1/2}{s+4} + \frac{5/2}{s+2} - \frac{3/2}{s+4} = \frac{3}{s+2} - \frac{1}{(s+3)^2} - \frac{2}{s+4}$$

Calculating the inverse Laplace transform of Y(s) yields

$$y(t) = \left(3e^{-2t} - te^{-3t} - 2e^{-4t}\right)u(t).$$

$$V_1(s) = \phi_1(s) - \theta(s)$$

$$V_2(s) = K_1G(s)V_1(s) = K_1G(s)[\phi_1(s) - \theta(s)].$$

Substituting $\theta(s) = V_2(s)/s$ in the above equation, we get

$$V_2(s) = K_1G(s)[\phi_1(s) - V_2(s)/s]$$

or,

$$V_2(s) = \frac{sK_1G(s)}{s + K_1G(s)}\phi_1(s)$$
.

The output

$$V(s) = K_2 V_2(s) = K_2 \frac{s K_1 G(s)}{s + K_1 G(s)} \phi_1(s) ,$$

which results in the transfer function $H(s) = \frac{V(s)}{\phi(s)} = K_1 K_2 \frac{sG(s)}{s + K_1 G(s)}$.

Differentiator: For the PLL to behave as an ideal differentiator, its transfer function H(s) = Ks, i.e.,

$$H(s) = Ks = K_1 K_2 \frac{sG(s)}{s + K_1 G(s)},$$

or,

$$K(s+K_1G(s))=K_1K_2G(s).$$

Solving in terms of G(s), we obtain

$$G(s) = \frac{K}{K_1(K_2 - K)} s.$$

Solutions

9

Another way of obtaining an ideal differentiator is to set $K_1 \to \infty$ in the transfer function as shown below:

$$\lim_{K_1 \to \infty} H(s) = \lim_{K_1 \to \infty} K_2 \frac{sG(s)}{s/K_1 + G(s)} = K_2 s.$$