



# BDD-Based Logic Decomposition

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Sources:

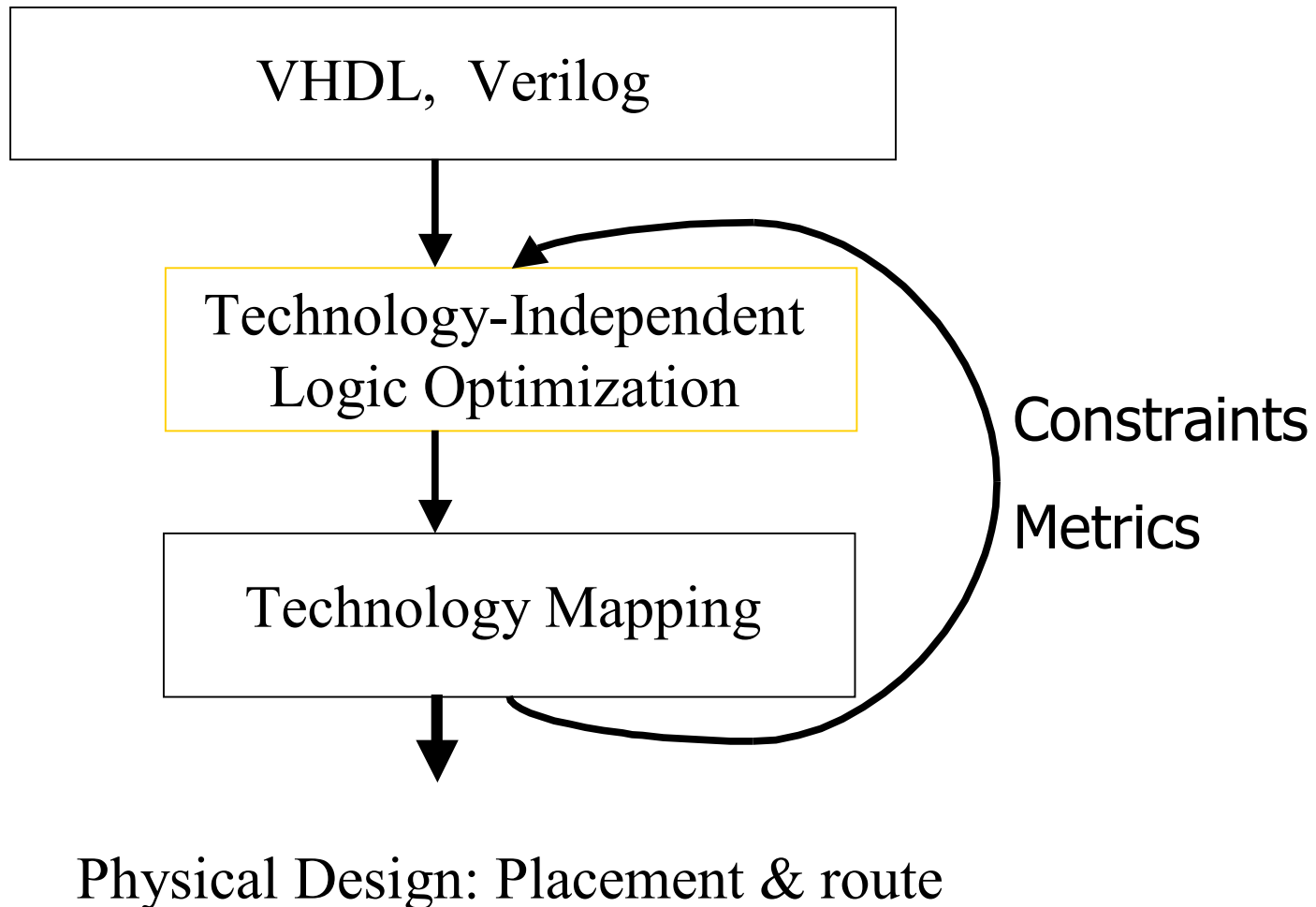
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F. Brewer, J. Roy, I. Markov



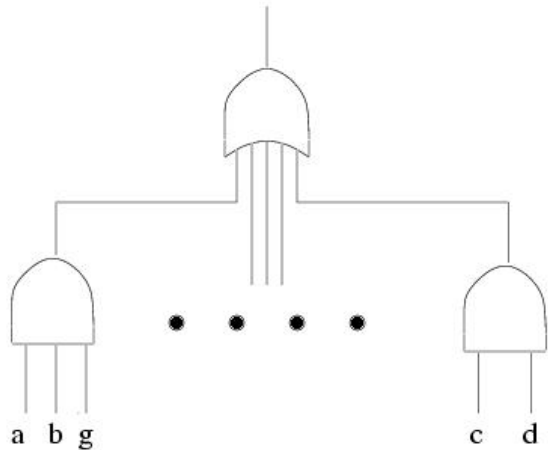
# Logic Synthesis Flow

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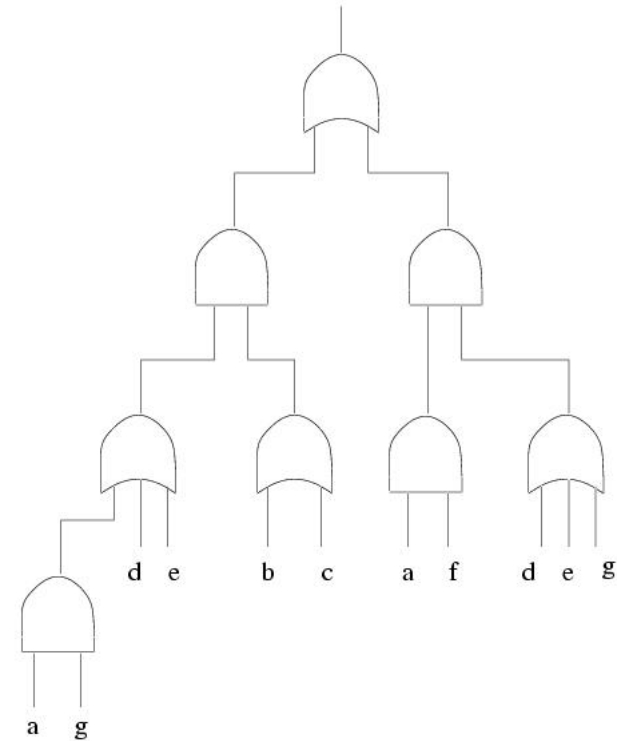
# Multi-Level Logic Synthesis

$$abg + acg + adf + aef + afg + bd + ce + be + cd$$



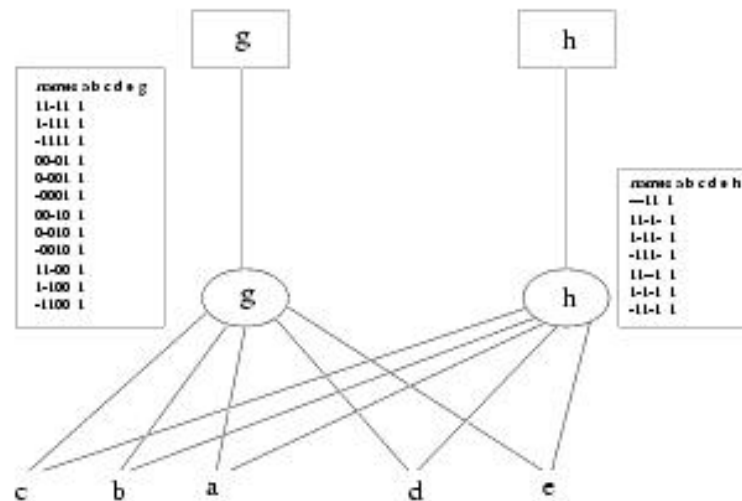
Minimizing # of product terms  
Quine-McCluskey method (minterm)  
Espresso (SOP)

$$(b+c)(d+e+ag) + (d+e+g)af$$

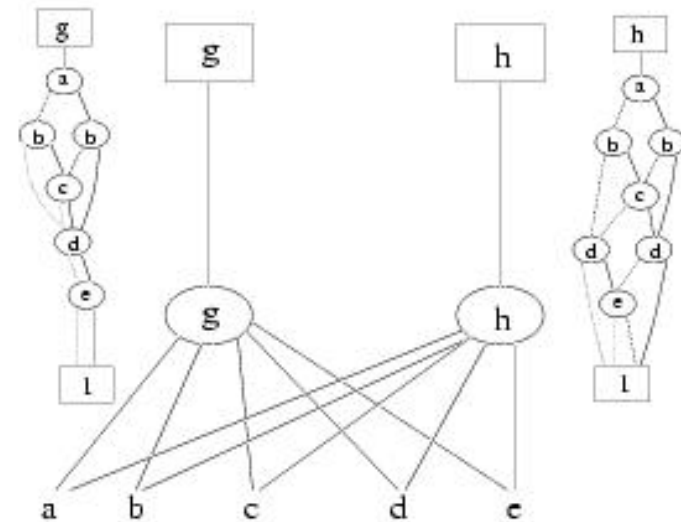


Minimizing # literals, gates, delay.  
Algebraic factorization.  
SIS (factored form)

# Global Representation of Boolean Network

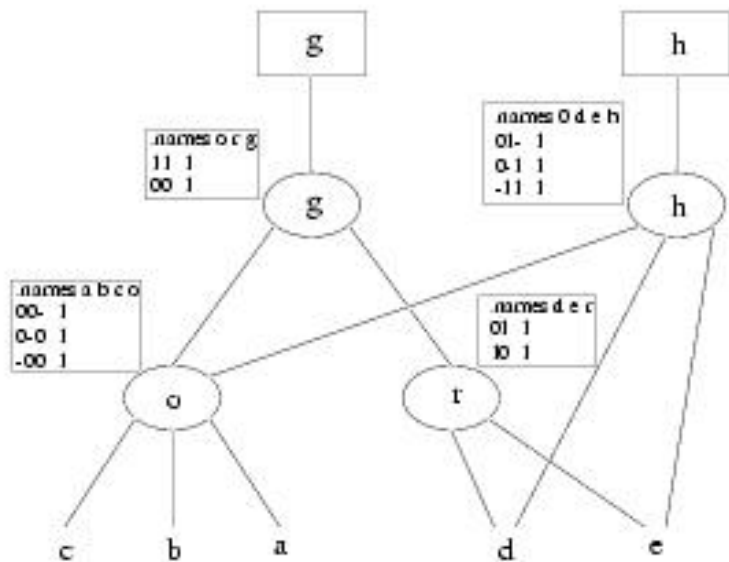


(i) Two-level SOP representation

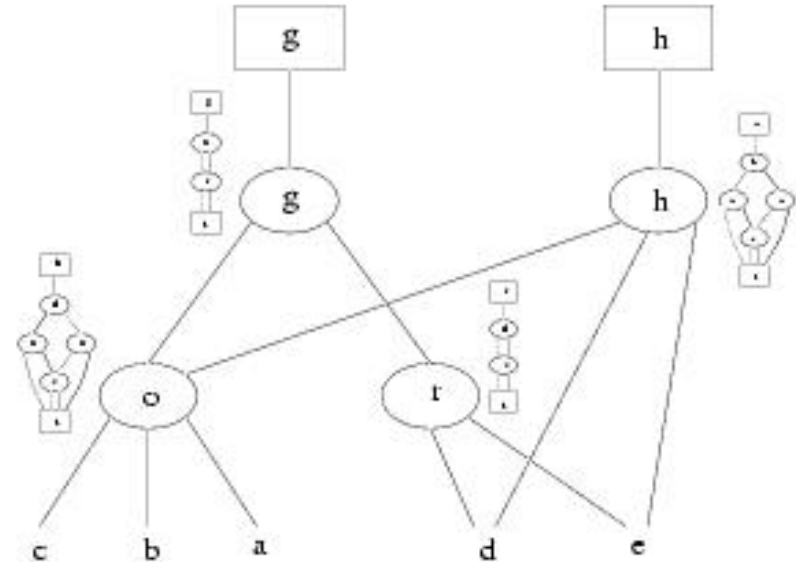


(j) Global BDD representation

# Boolean Network



(g) Multi-level SOP representation



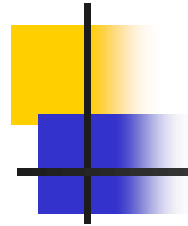
(h) Local BDD representation

# Boolean Decomposition

## – previous work

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- Ashenhurst [1959], Curtis [1962]
  - Tabular method based on cut: bound/free variables
  - BDD implementation:
    - Lai et al. [1993, 1996], Chang et al. [1996]
    - Stanion et al. [1995]
- Roth, Karp [1962]
  - Similar to Ashenhurst, but using cubes, covers
  - Also used by SIS
- Factorization based
  - SIS, algebraic factorization using cube notation
  - Bertacco et al. [1997], BDD-based recursive bidecomp.



# Functional Decomposition

- Decompose Network into Composed Functions
  - Minimize Depth and Redundancy

$$f(x_1, x_2 \dots x_n) = g(h(x_1, x_2 \dots x_m), x_{m+1} \dots x_n)$$

- Decomposition is disjoint if support of g and h are.
- Disjoint decomposition exists iff each

$$f|_{x_{m+1} \dots x_n}(x_1, x_2 \dots x_m) = f_1(x_1, x_2 \dots x_m) \vee f_2(x_1, x_2 \dots x_m)$$

- I.e. high degree of internal symmetry
- Non-disjoint decompositions are more common



# Ashenhurst Decomposition

Goal:  $f(X) = F(\Phi(Y), Z)$ .

Example:  $F = w'x'z' + wx'z + w'yz + wyz' = f(\Phi(w, z), x, y) = \Phi x' + \Phi' y$ .

		w z			
		0	1	2	3
x y	0	1	0	0	1
	1	1	1	1	1
	2	0	0	0	0
	3	0	1	1	0

(a)  $\mu = 2$

		y z			
		0	1	2	3
x w	0	1	0	1	1
	1	0	1	1	1
	2	0	0	0	1
	3	0	0	1	0

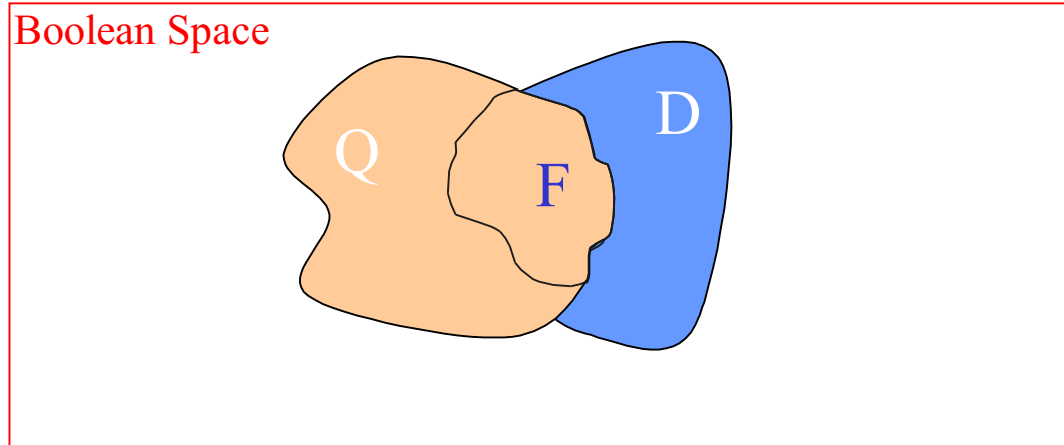
(b)  $\mu > 2$

$u$  is the number of distinct columns



# Boolean AND Decomposition

Goal : find  $D, Q$  such that  $F = Q * D$ .



$$F = e + bd; \quad Q = e + b; \quad D = e + d$$



# Boolean Division

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- $G$  is a **Boolean divisor** of  $F$  if there exist  $H$  and  $R$  such that
$$F = G H + R, \text{ and } G H \neq 0.$$

- $G$  is said to be a **factor** of  $F$  if, in addition,  $R=0$ , that is:
$$F = G H.$$

where  $H$  is the **quotient**,  $R$  is the **remainder**.

**Note:**  $H$  and  $R$  are not unique.

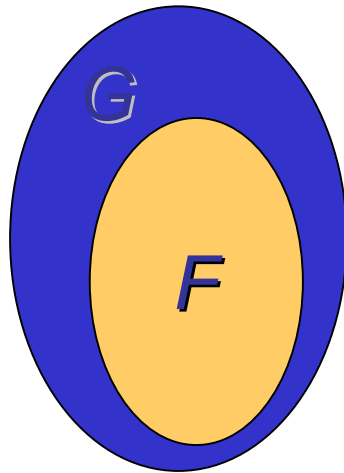
# Many Boolean Factors!

$G$  is a **Boolean factor** of  $F$  iff  $F \subseteq G$   
(i.e.  $FG' = 0$ , or  $G' \subseteq F'$ ).

Example:

$$F = a + bc; \quad G = (a+c)$$

$$F = (a+b)(a+c); \quad R=0$$



Proof:

$\Rightarrow$ :  $G$  is a Boolean factor of  $F$ . Then  $\exists H$  s.t.  $F = GH$ ;  
Hence,  $F \subseteq G$  (as well as  $F \subseteq H$ ).

$\Leftarrow$ :  $F \subseteq G \Rightarrow F = GF = G(F + R) = GH$ .  
(Here  $R$  is any function  $R \subseteq G'$ )

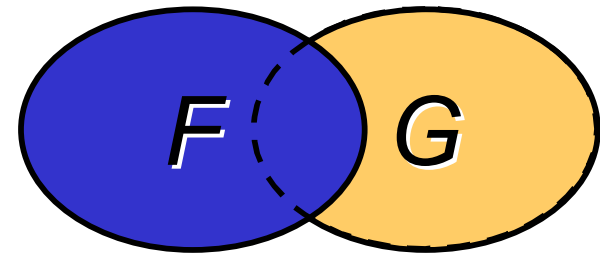
# More Boolean Divisors

$G$  is a **Boolean divisor** of  $F$  if and only if  $FG \neq 0$ .

Example:

$$F = ab + ac + bc; G = b + c$$

$$F = a(b + c) + bc; R = bc$$



Proof:

$$\Rightarrow: F = GH + R, GH \neq 0 \Rightarrow FG = GH + GR. \text{ Since } GH \neq 0, FG \neq 0.$$

$$\Leftarrow: \text{ Assume that } FG \neq 0. F = FG + FG' = G(F + K) + FG'. \text{ (Here } K \subseteq G'.)$$

$$\text{ Then } F = GH + R, \text{ with } H = F + K, R = FG'. \text{ Since } GH = FG \neq 0, \text{ then } GH \neq 0.$$



# Algebraic Factorization

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- Objective : Obtain a minimized factored form
- $x$  and  $x'$  treated as different variables
- Boolean algebra rules:  $x + x' = 1$ ;  $x x' = 0$ , not applied.
- A Boolean formula treated as polynomial formula
- Two-level techniques for Boolean simplification.
- Polynomial representation is the bottleneck in Boolean factorization



# Boolean Division

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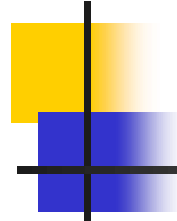
- $F = abg + acg + adf + aef + afg + bd + ce + be + cd.$  (23 lits)
- **Algebraic** :  $F = (b + c)(d + e + ag) + (d + e + g)af.$  (11 lits)
- **Boolean**:  $F = (af + b + c)(ag + d + e).$  (8 lits)



# Algebraic Logic Synthesis Issues

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- Weak Boolean factorization
- Difficult to perform XOR and MUX decomposition.
- Separate platforms for Boolean operations and factorization.
- Goal: use BDD platform for both Boolean operations and Factorization
  - BDD structure used to find good factors
  - Maybe form *non*-conjunctive decompositions



# BDD – Synthesis

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- BDD is canonical
  - Variable reordering implicit logic simplification
  - Some redundancy removed
- BDD is efficient factored form
  - Reduced representation should lead to efficient matching
- Problem : How to translate a BDD into a multi-level (circuit) representation?





# Shannon Expansion

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A Boolean function  $f(x_1, x_2, \dots, x_i, \dots, x_n)$  can be expressed as,

$$\begin{aligned} f(\dots, x_i, \dots) &= x_i f(\dots, 1, \dots) + x'_i f(\dots, 0, \dots) \\ &= x_i f_i^1 + x'_i f_i^0; \end{aligned}$$

$$\begin{aligned} f(\dots, x_i, \dots) &= (x_i + f(\dots, 0, \dots))(x'_i + f(\dots, 1, \dots)) \\ &= (x_i + f_i^0)(x'_i + f_i^1); \end{aligned}$$

$f_i^1$  : positive cofactor;  $f_i^0$  : negative cofactor



# Orthogonal Expansion

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- Shannon Expansion is just a special case of linear orthonormal expansion:

Given set of functions:  $g_i(x_1..x_n)$

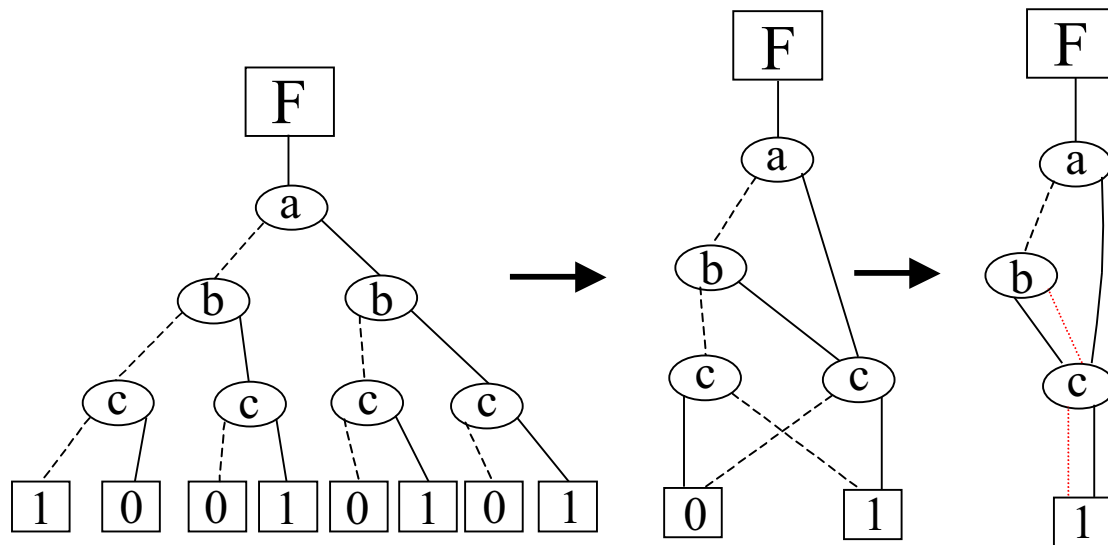
Where  $\bigcup_i g_i = T$  and  $\forall_{i,j,i \neq j} g_i g_j = 0$

$$f(x_1..x_n) = \bigcup_i g_i(x_1..x_n) f|_g$$

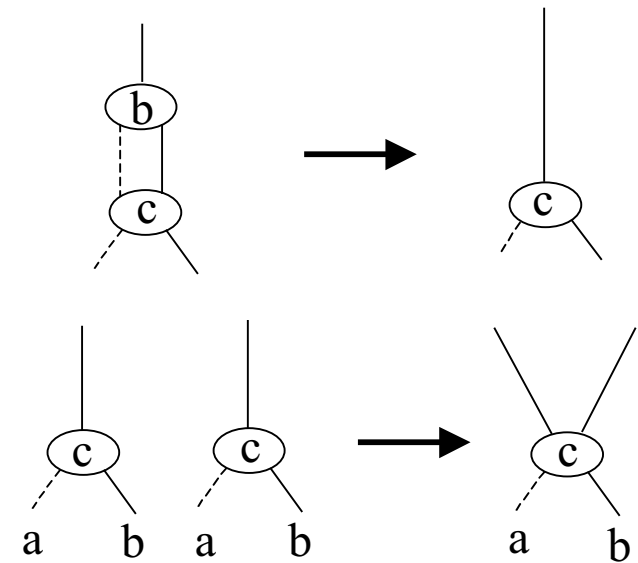
$f|_g(x_1..x_n)$  is the generalized co-factor of  $f$  by  $g$ .

# BDD Reduction

Example:  $F = ac + bc + a'b'c'$ ;

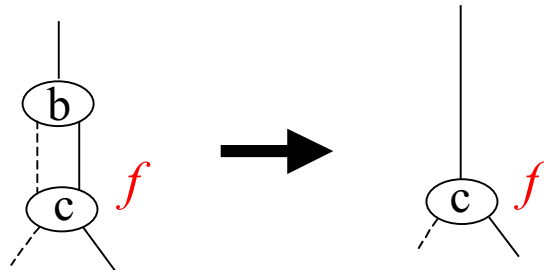


Reduction Rules :

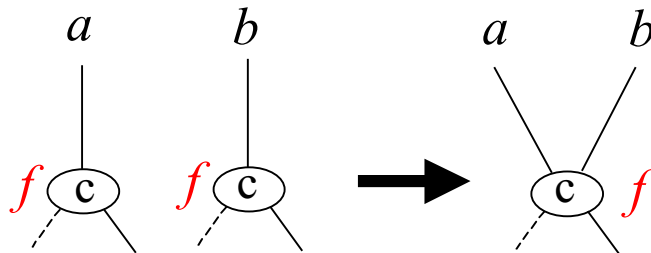


# BDD -- Synthesis

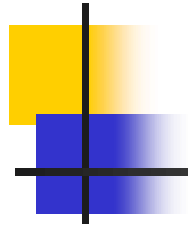
- A BDD is an implicitly factored representation



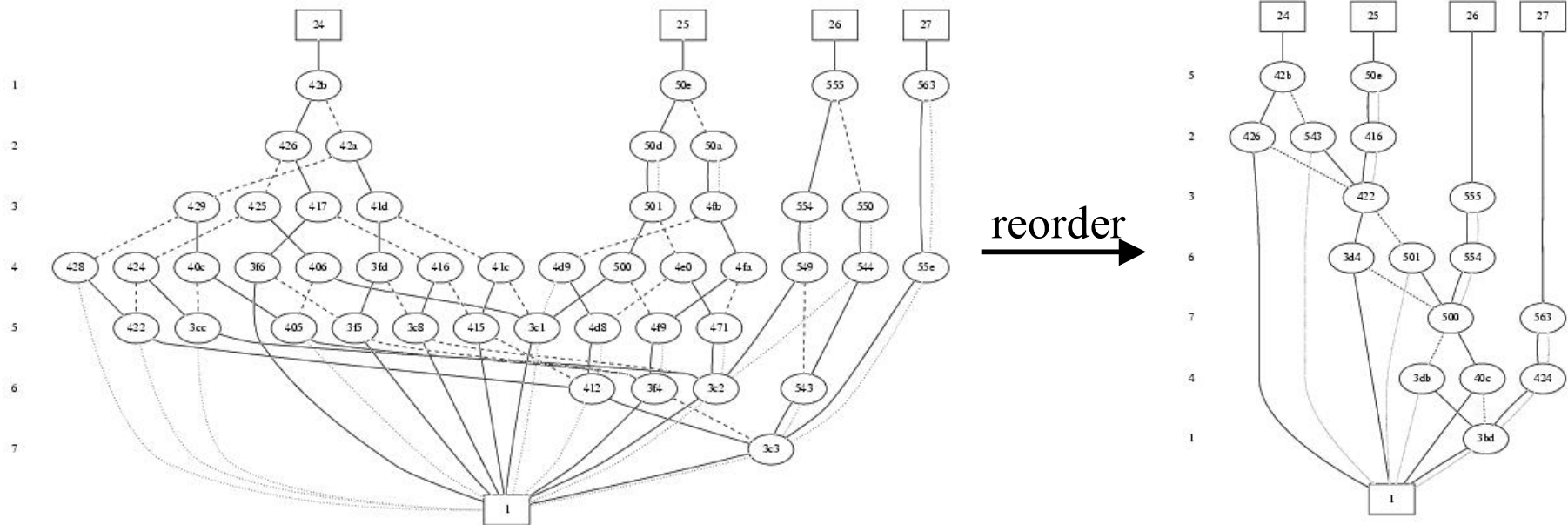
$$b'f + bf = f$$



$$af + bf = (a + b)f$$

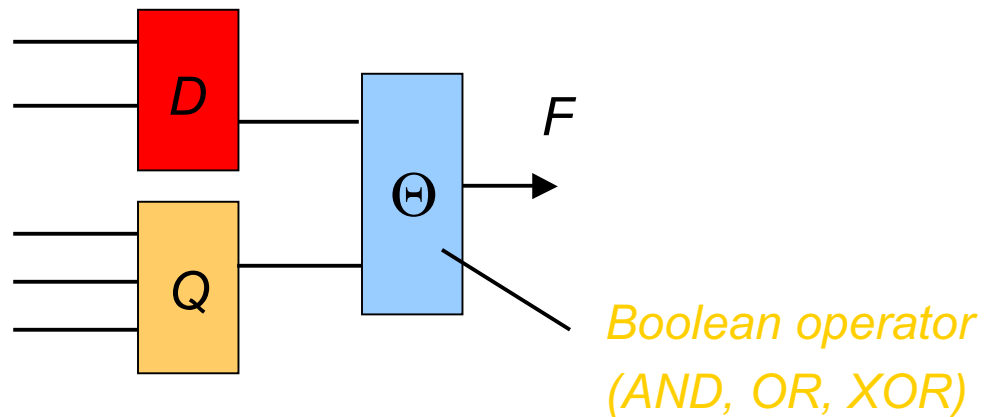
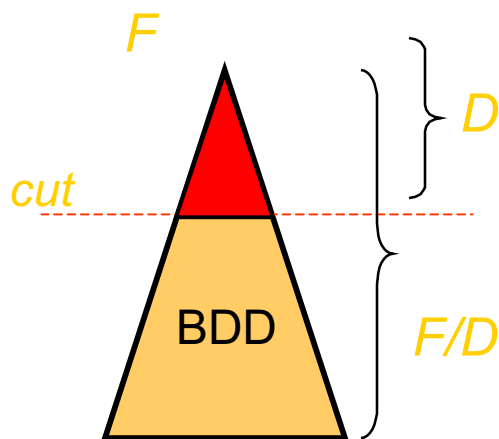


# Variable Ordering



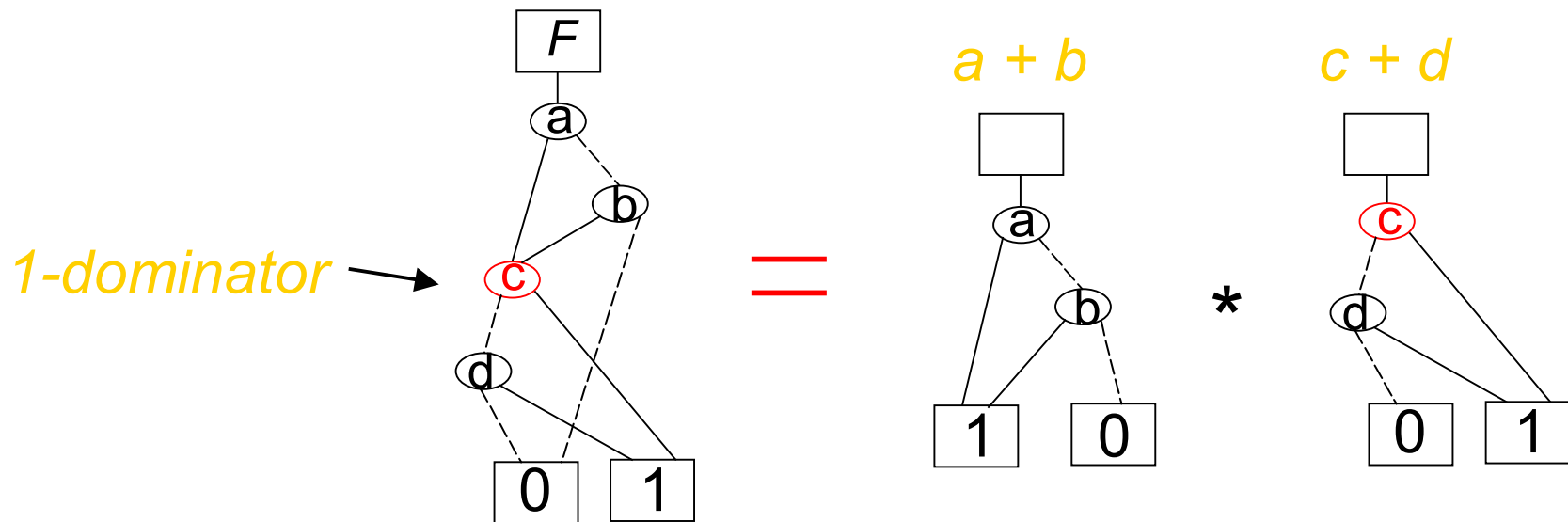
# Non-Disjoint Decomposition

- Bi-decomposition:  $F = D \oplus Q$ 
  - uses BDD cut
    - *Not a partition of input support variables!*
  - The top set defines a divisor  $D$
  - The bottom set defines the quotient  $Q$  (sort of ...)



# 1-dominator: Karplus [1988]:

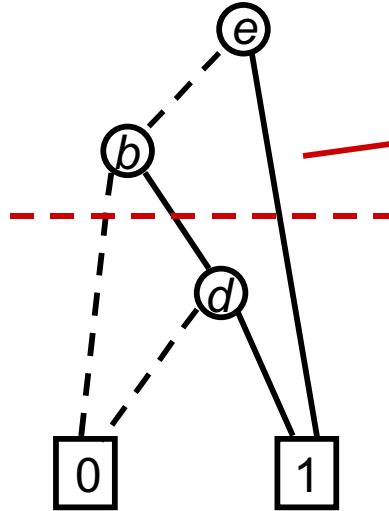
- 1-dominator* is a node that belongs to every path from root to 1-terminal.



- 1-dominator* defines algebraic conjunctive (AND) decomposition:  $F = (a + b)(c + d)$ .

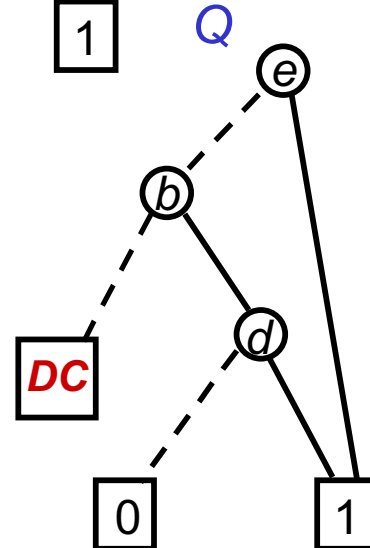
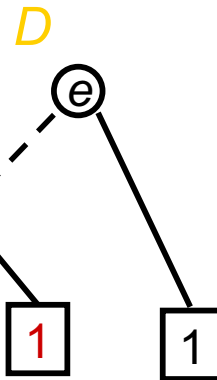
# Eg. AND Decomposition: $F = D Q$

$$F = e + bd$$

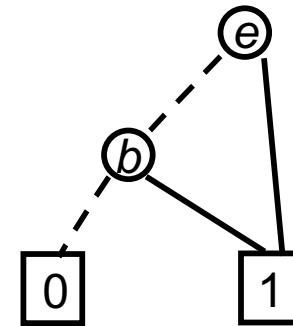


$Q$

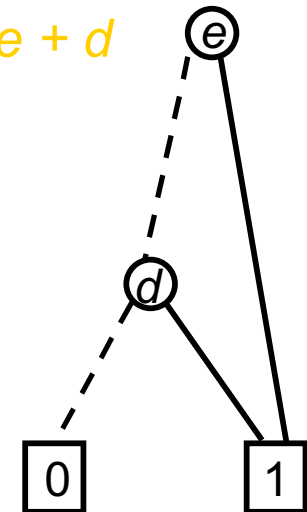
	$eb$	00	01	11	10
$d$	0	DC		1	1
	1	DC	1	1	1



$$D = e + b,$$



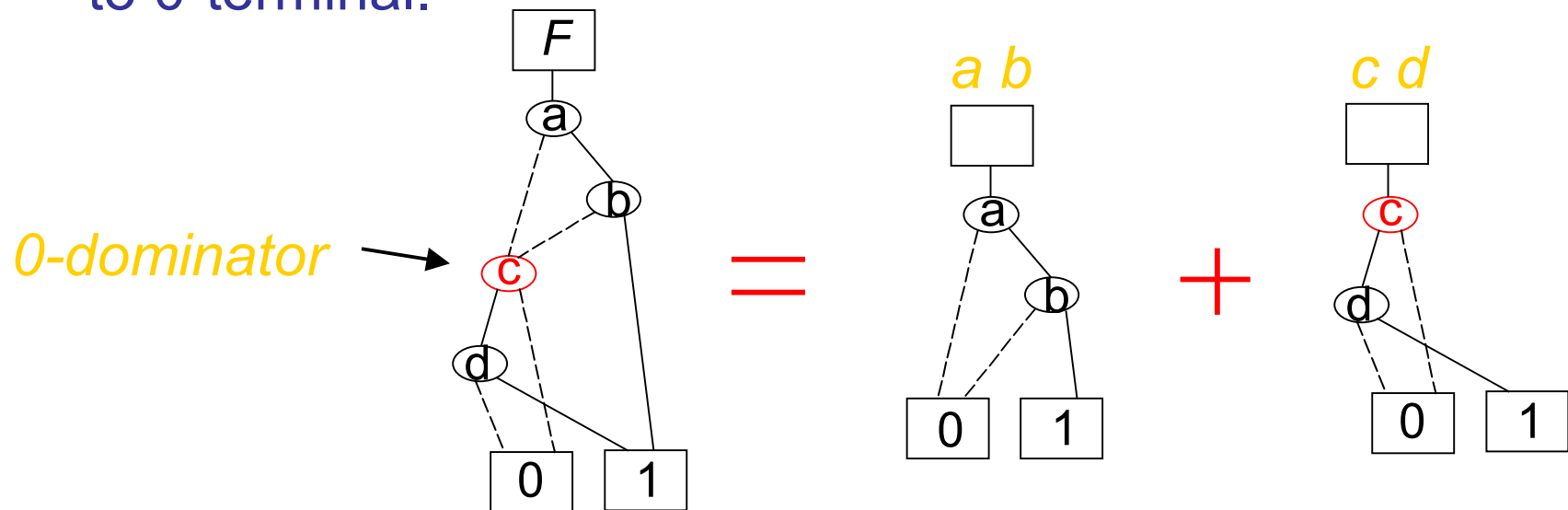
$$Q = e + d$$





# 0-dominator

*0-dominator* is a node belonging to every path from root to 0-terminal.



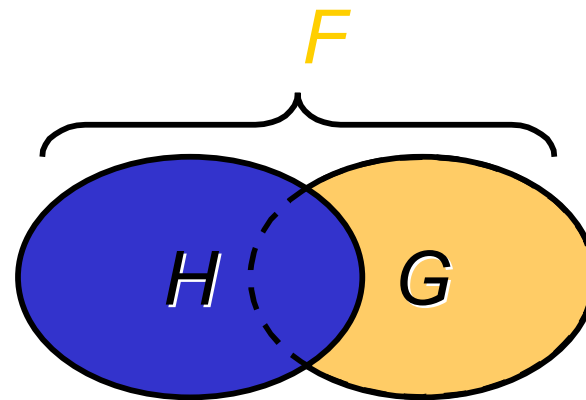
- 0-dominator* defines algebraic disjunctive (OR) decomposition:  $F = ab + cd$ .

# Disjunctive Decomposition

- Disjunctive (OR) decomposition:  $F = G + H$ .

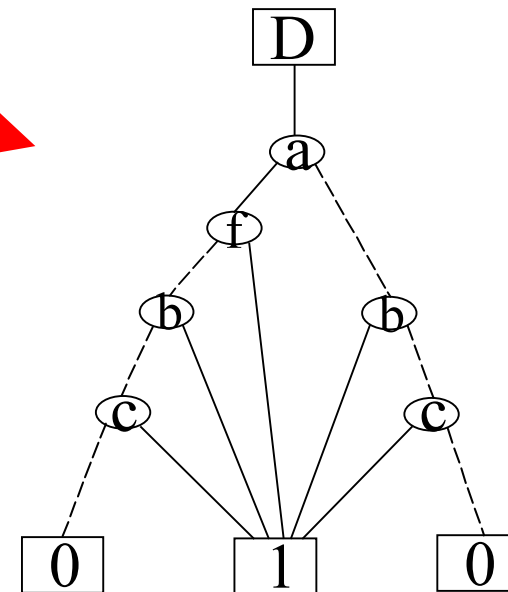
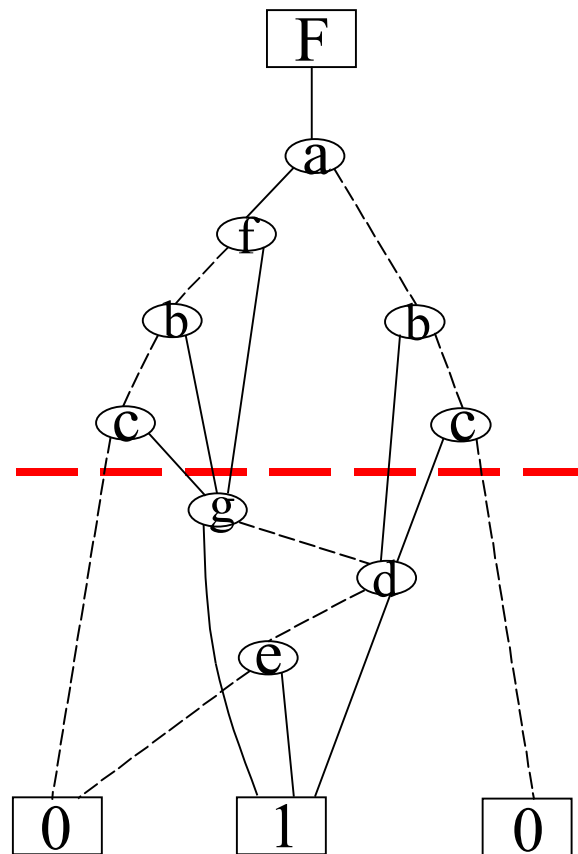
Boolean function  $F$  has disjunctive decomposition  
iff  $F \supseteq G$ . For a given  $G$ ,  $H$  must satisfy:

$$F' \subseteq H' \subseteq F' + G.$$



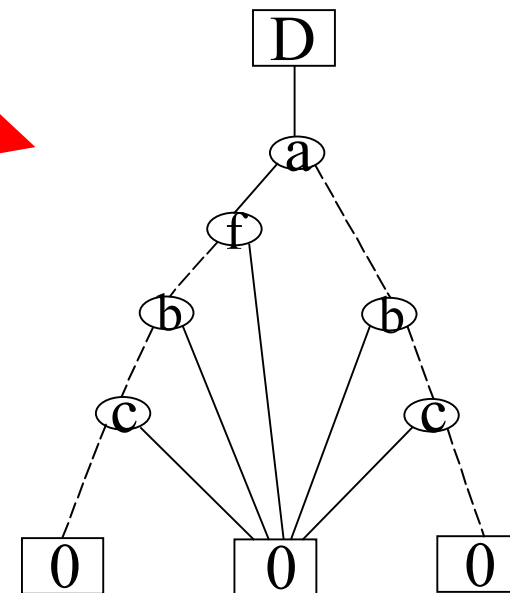
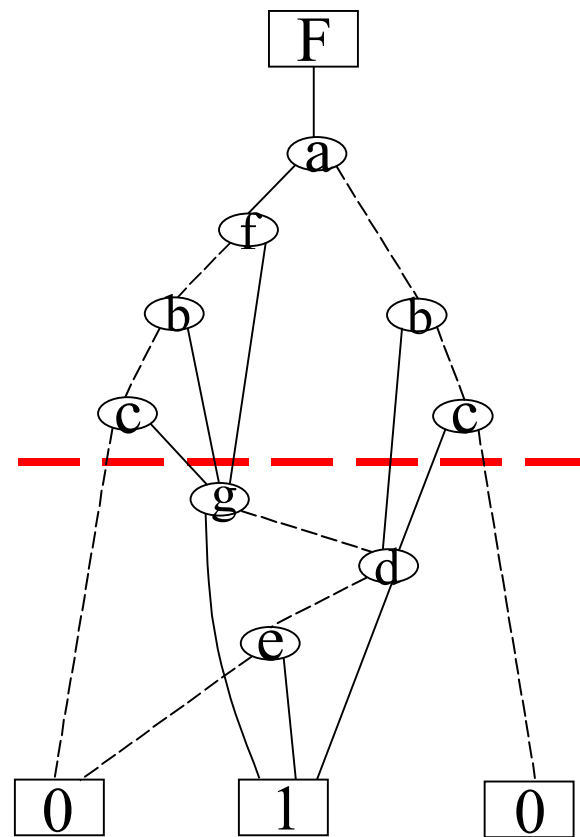
- For a given  $(F, G)$ , this provides a recipe for  $H$ .

# Generalized Dominator (And)



Boolean divisor

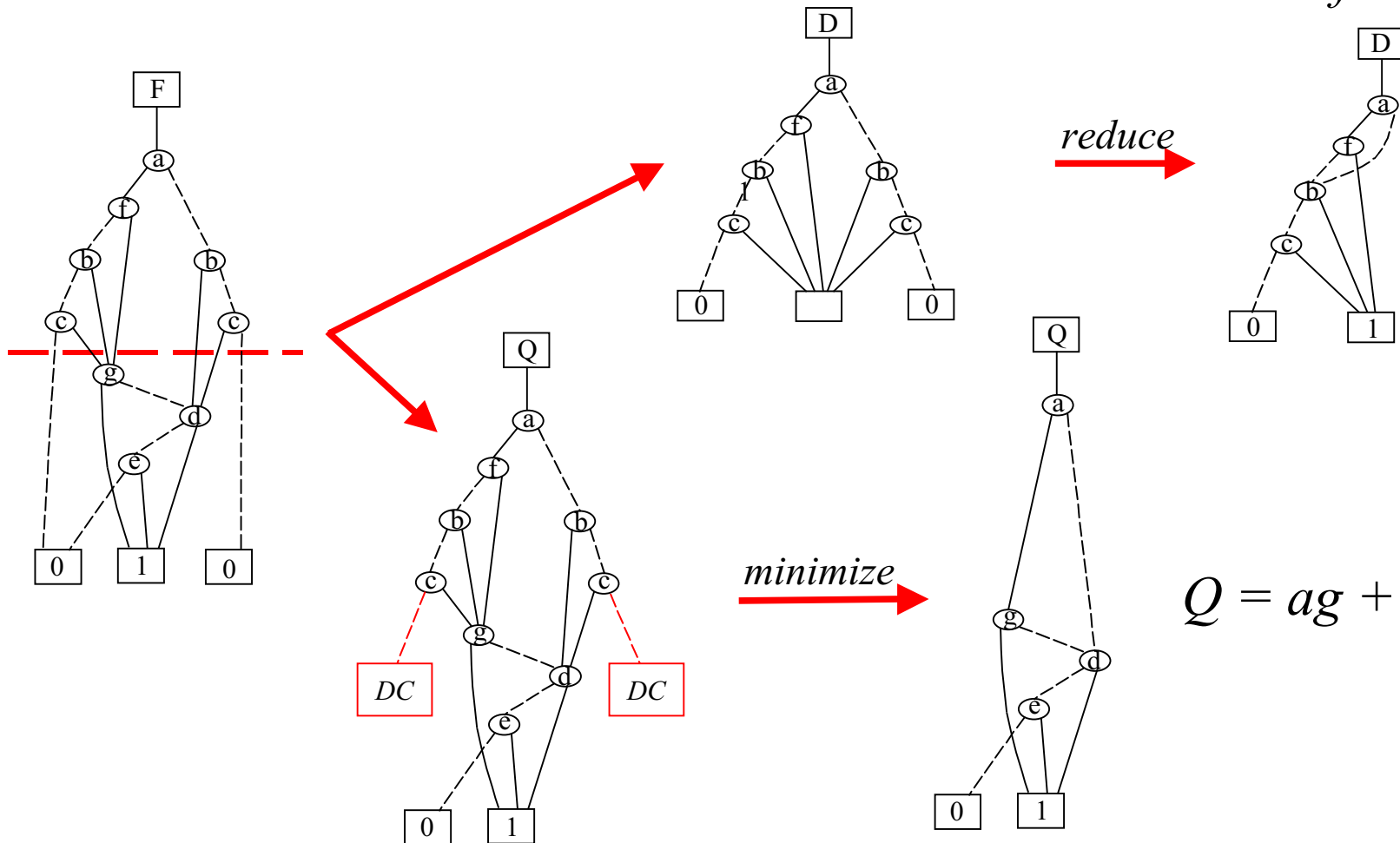
# Generalized Dominator (or)



Boolean subtractor

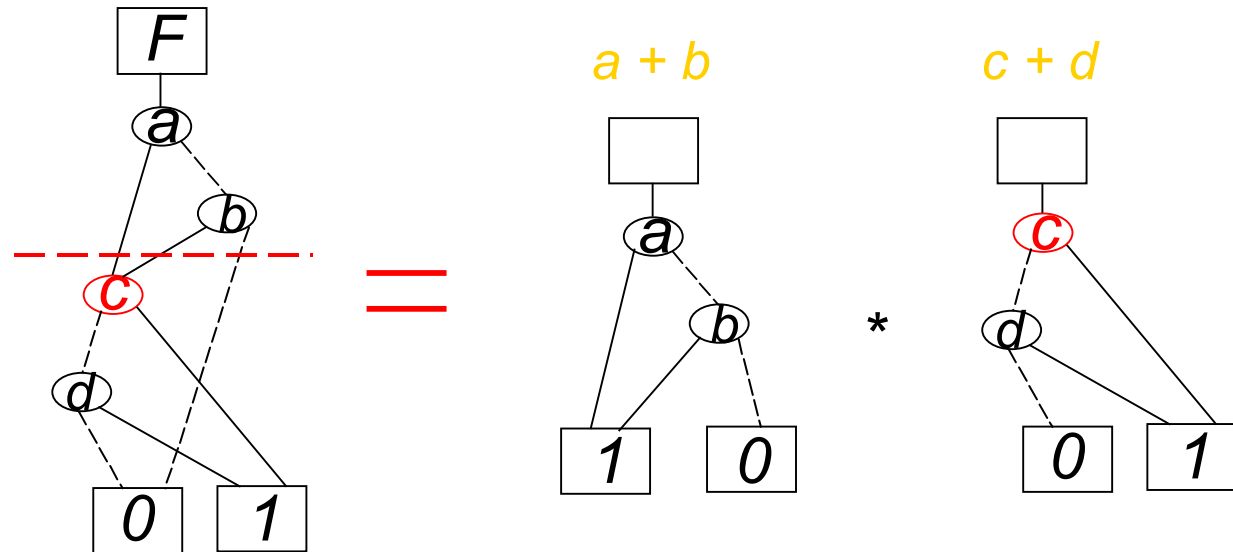
# Boolean Division via Dominator

$$D = af + b + c$$



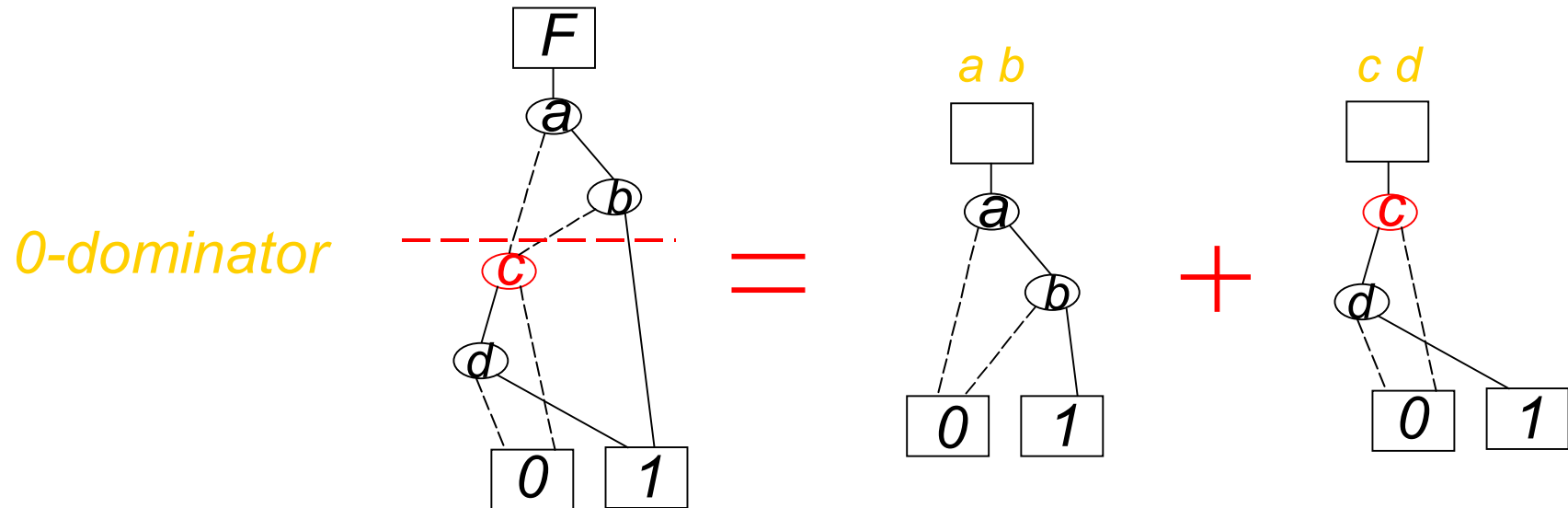
# Special Case: 1-dominator

1-dominator



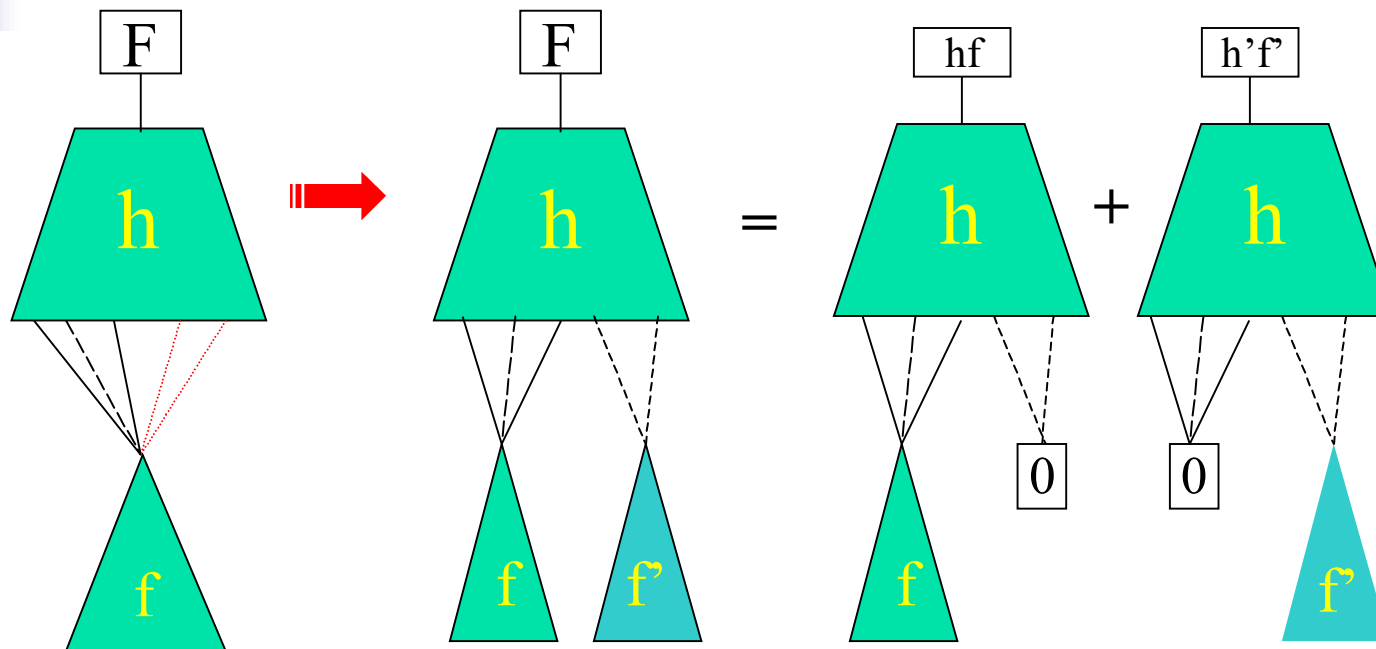
$$F = (a+b)(c+d)$$

# Special Case: 0-dominator



$$F = ab + cd$$

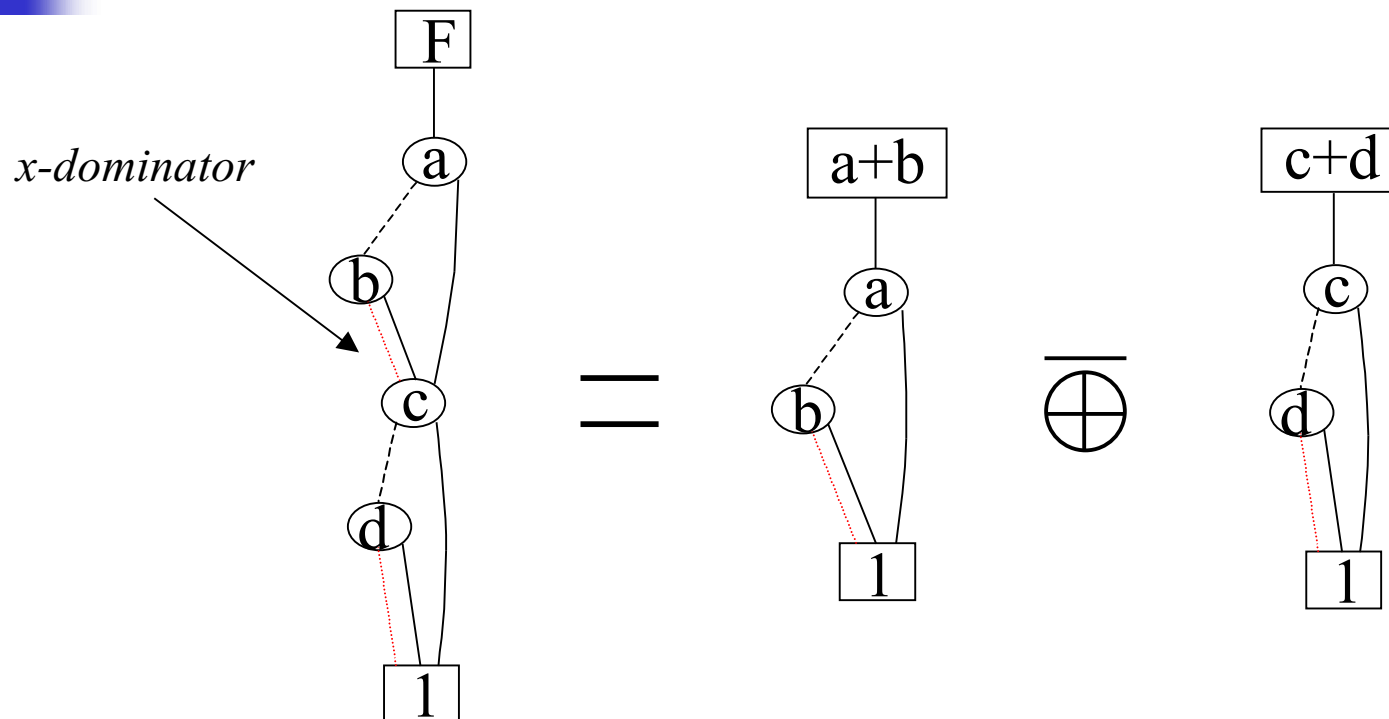
# Algebraic XOR Decomposition



..... = complement edge  
 ————— = 1-edge edge  
 - - - - - = 0-edge edge



# Algebraic XOR Decomposition: x-dominator

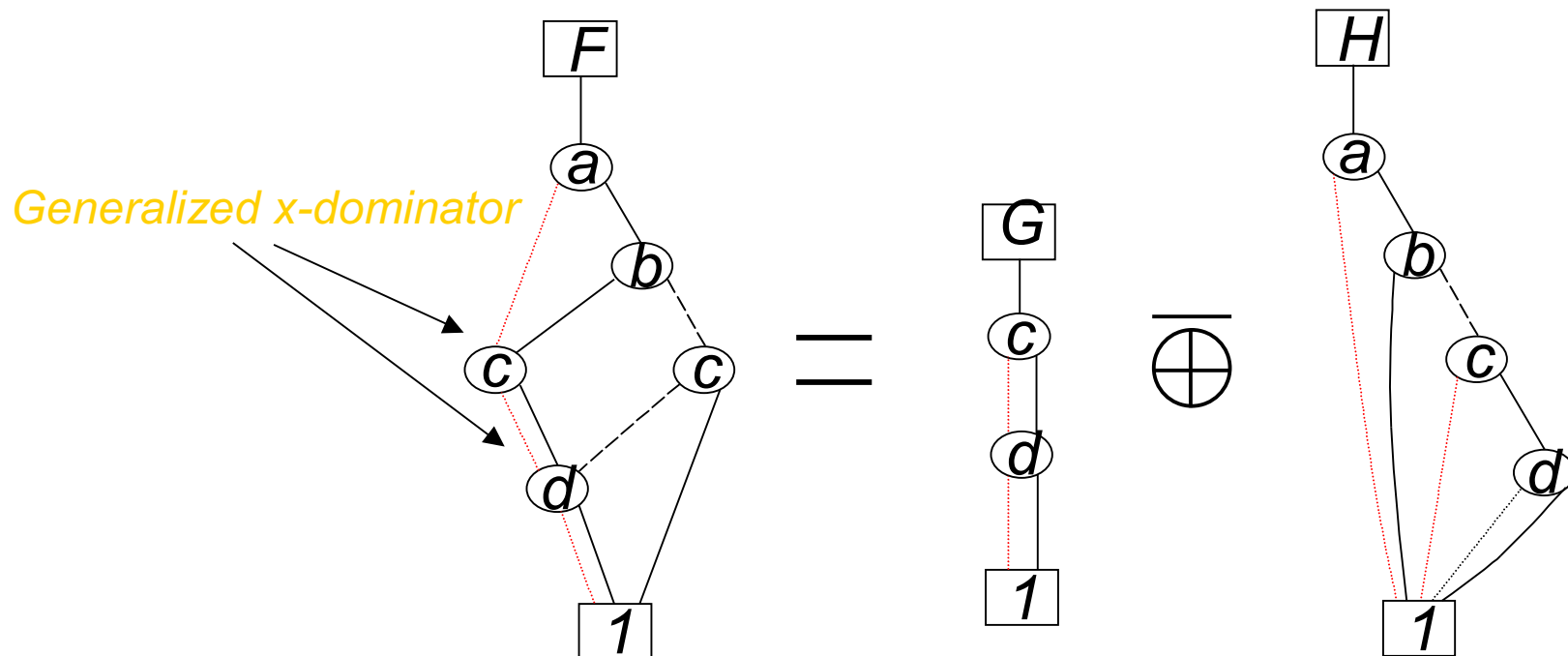


- X-dominator nodes contain *every* path in BDD
  - Denotes equivalence of upper and lower parts
  - Complement Edges are the key to finding such nodes

# Boolean XOR Decomposition: Generalized x-dominators

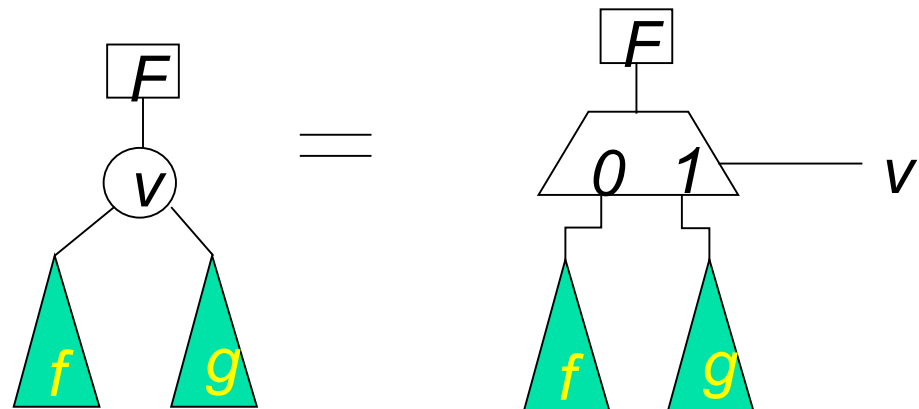
Generalized X-dom: any node with incoming true and complement edges.

Given  $F$  and  $G$ , there exists  $H : F = G \otimes H; \quad H = F \otimes G.$

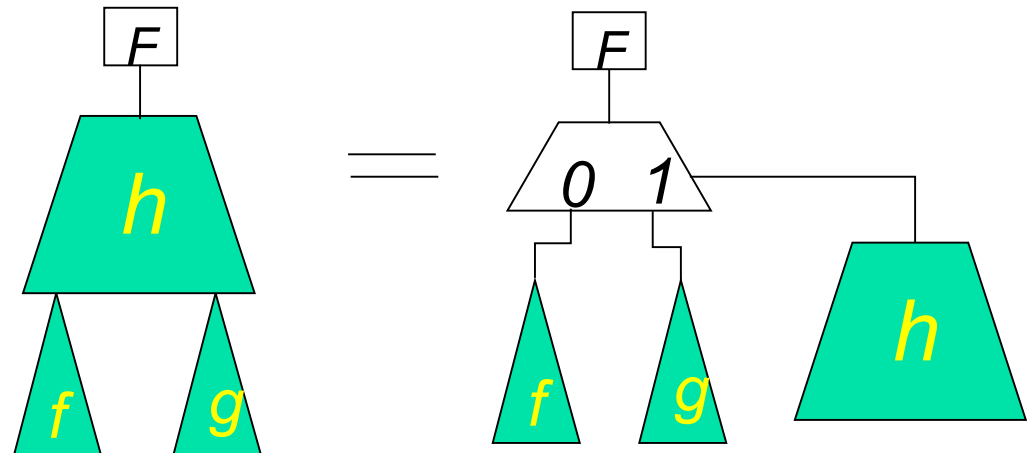


# MUX Decomposition

- Simple MUX decomposition

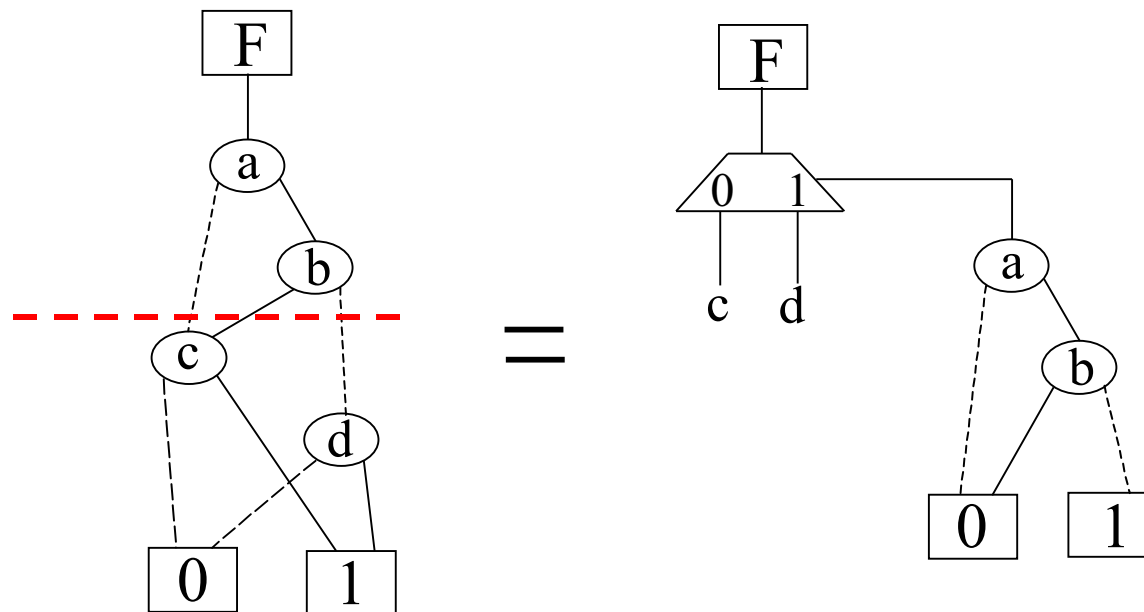


- Complex MUX decomposition

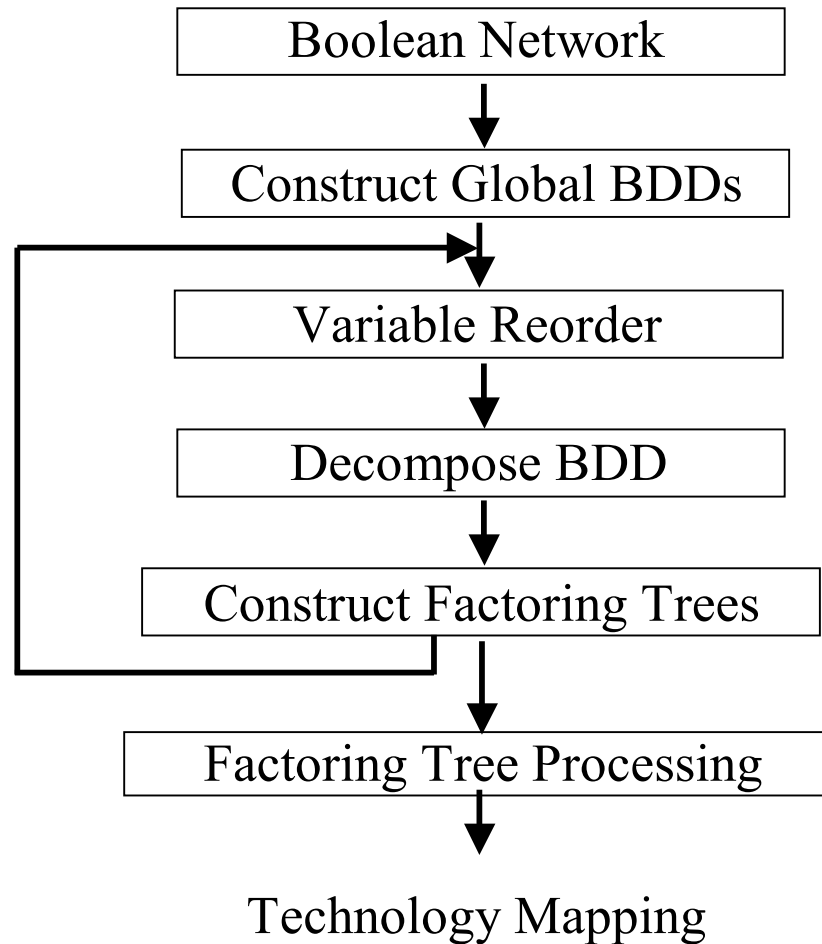


# Functional MUX Decomposition

- If two nodes  $u$  and  $v$  cover all paths in a BDD we can decompose  $f = hu + h'v$  where  $h$  comes from forcing  $u=1$  and  $v=0$  in  $F$ .
  - Useful if  $u$  and  $v$  share little



# Synthesis Flow of BDDlopt



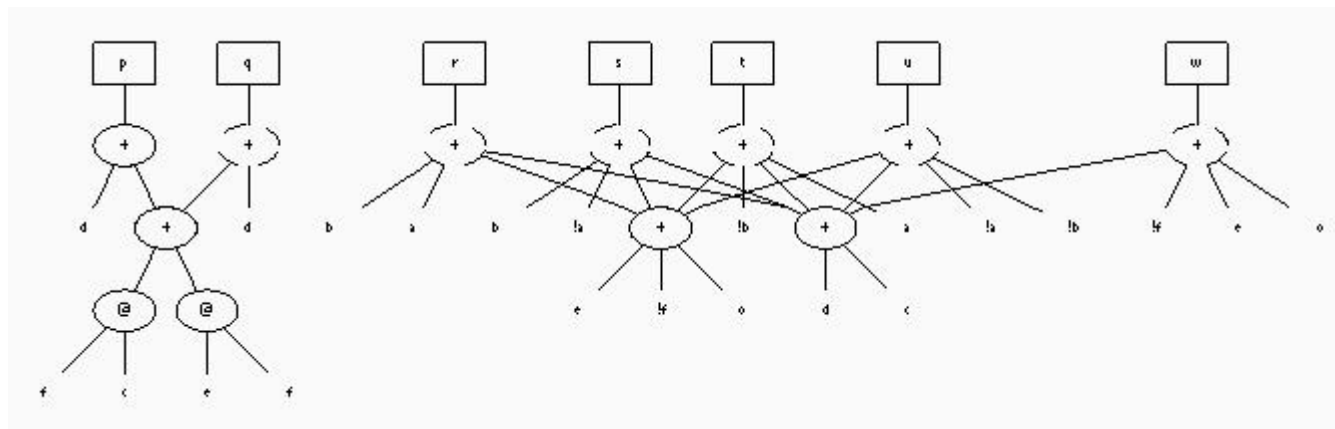
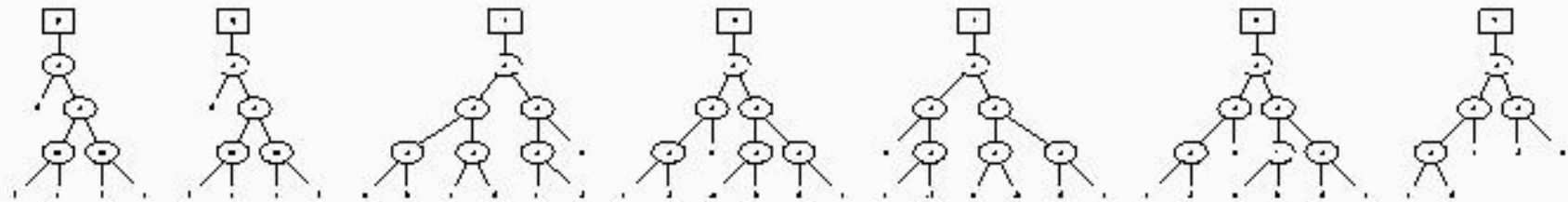


# Summary BDD Decomposition

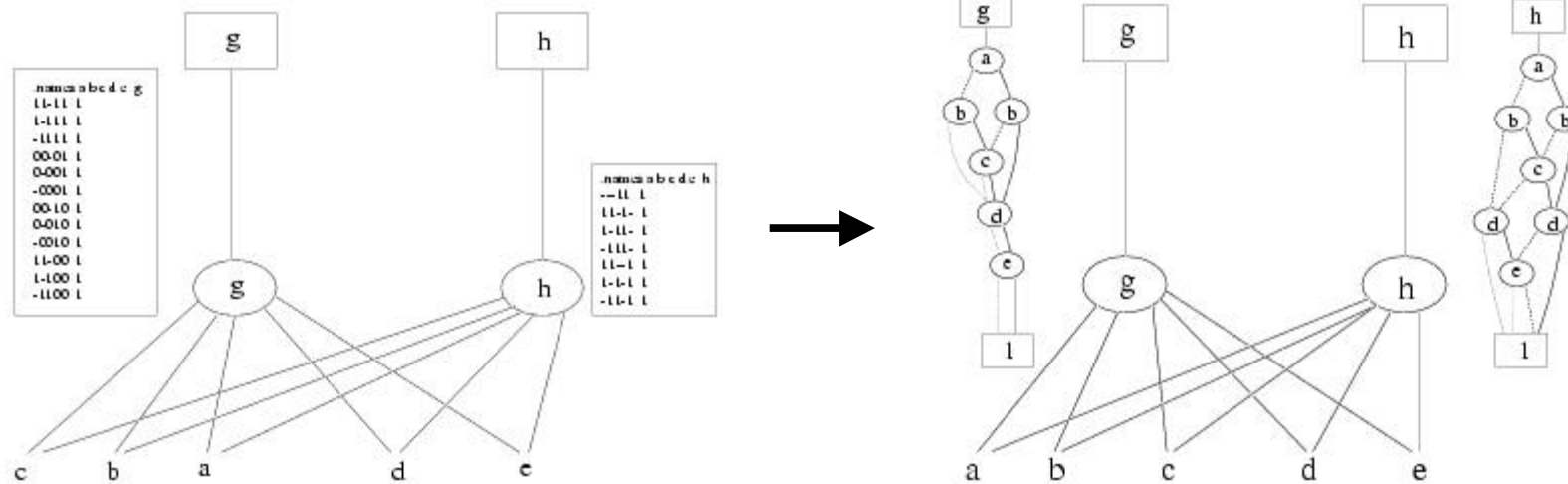
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- Ordering allows diagnosis of good cut points
- Cut of BDD Sponsored by redirecting edges of cofactors to constants to minimize the cut portion(s)
- Redirected edges create don't cares for remaining BDD nodes
- Don't care minimization by Restrict, Constrain, or GFRC
- Linear Decomposition theorem allows non-binary decompositions as a generalization of cuts.

# BDD Cut selection

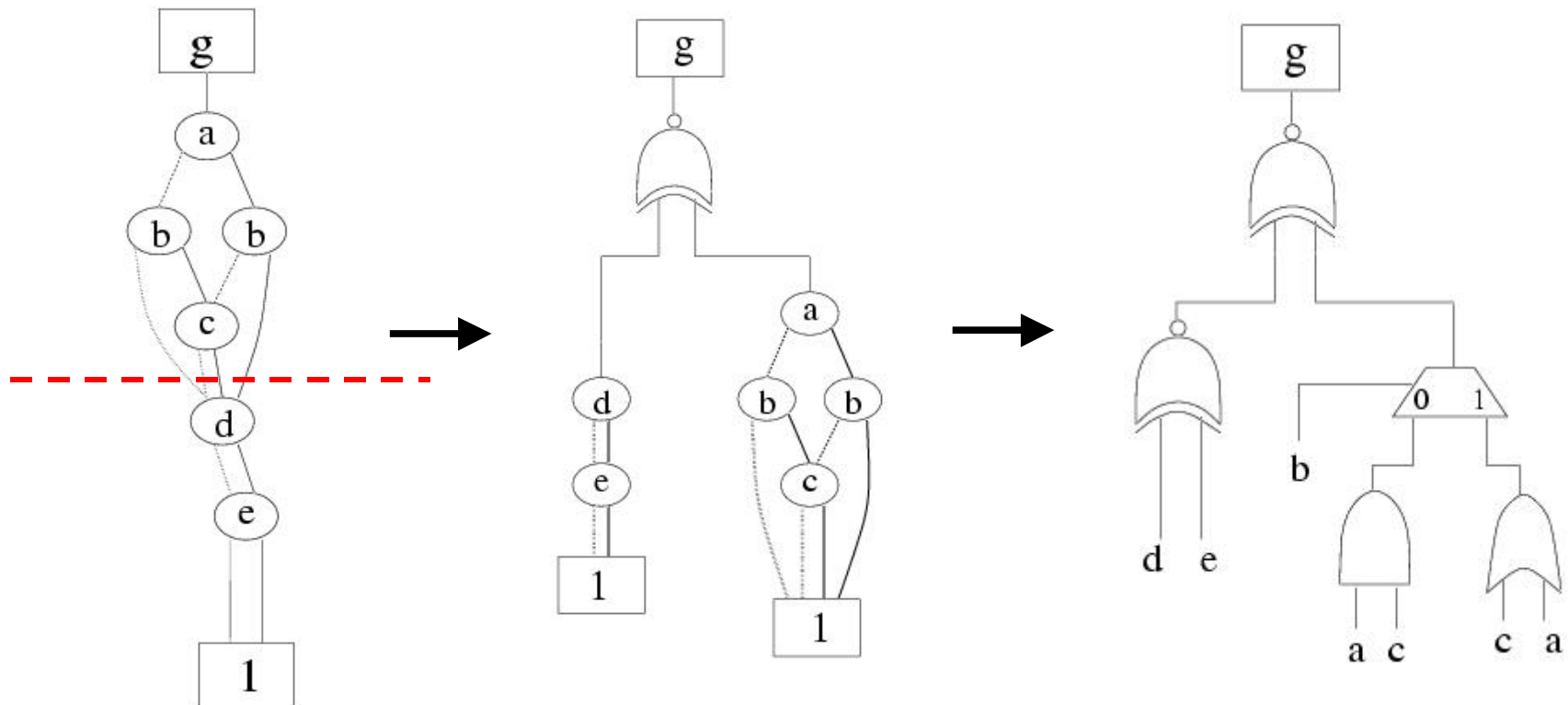


# Synthesis Example I

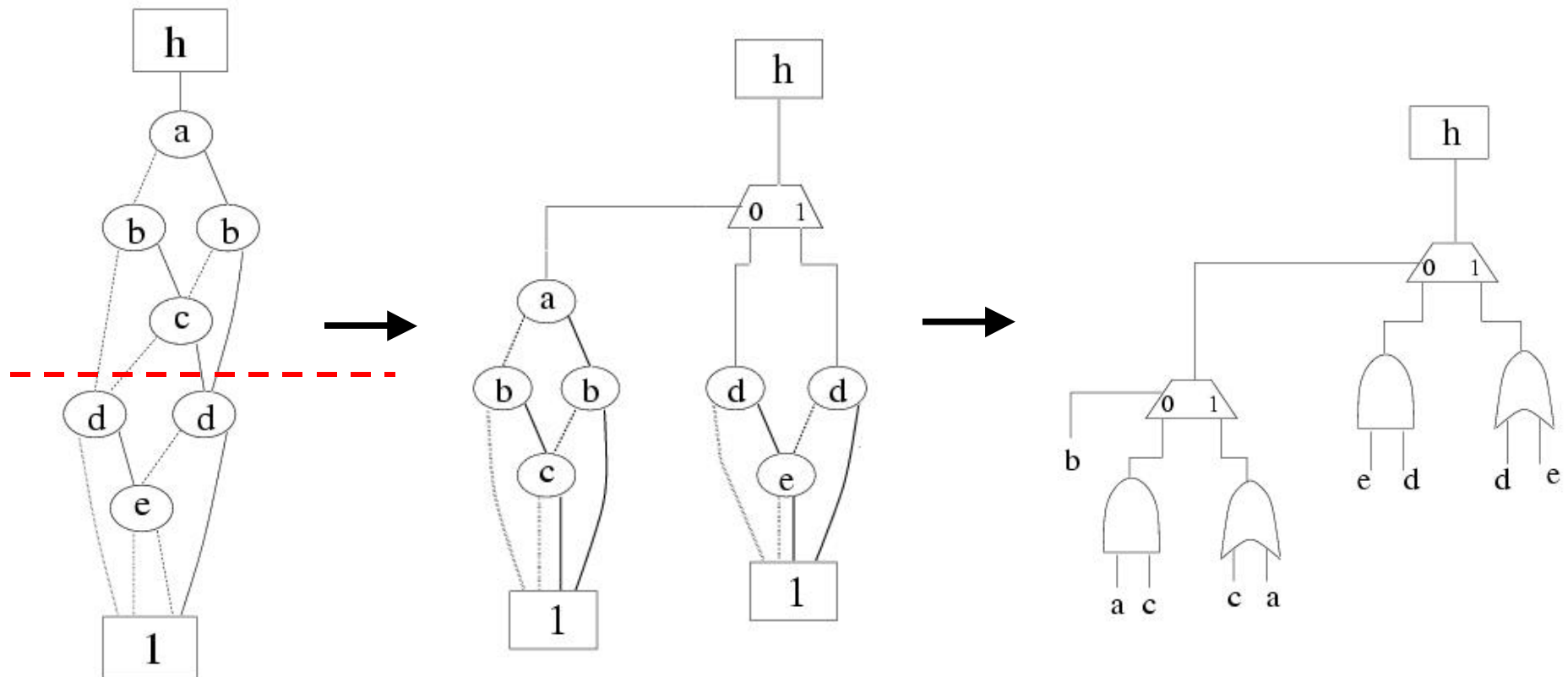




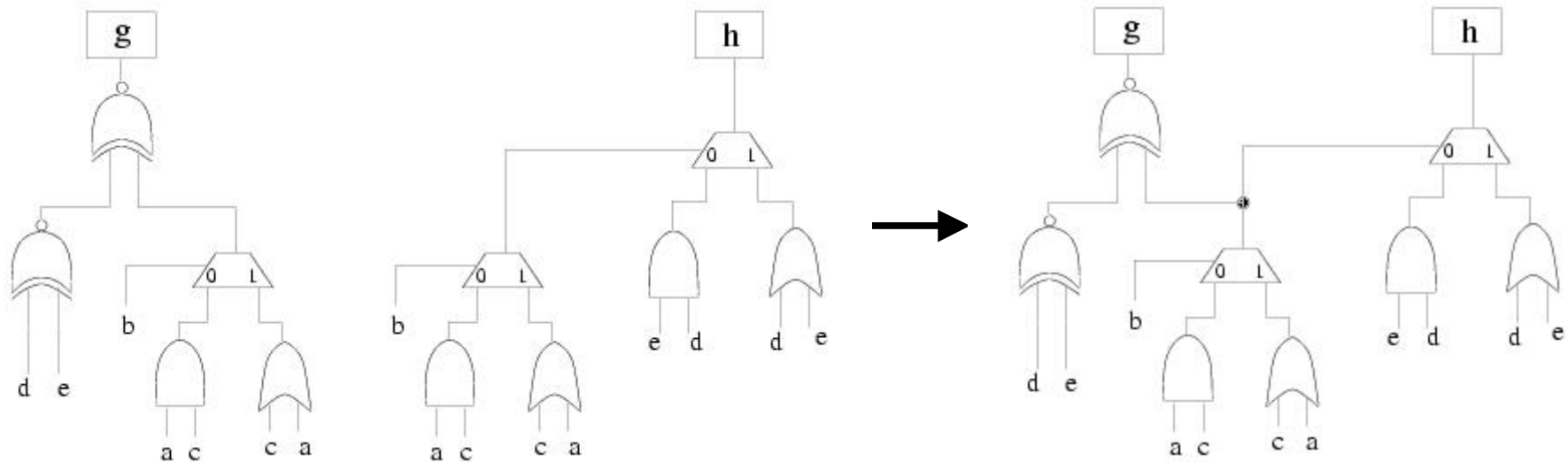
# Synthesis Example II (g)



# Synthesis Example III (h)

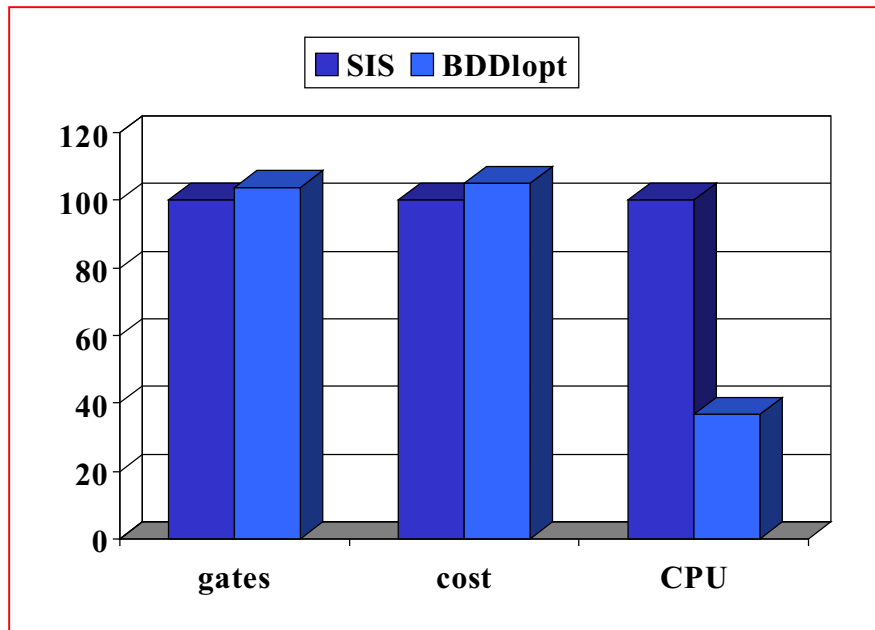


# Synthesis Example IV (share)

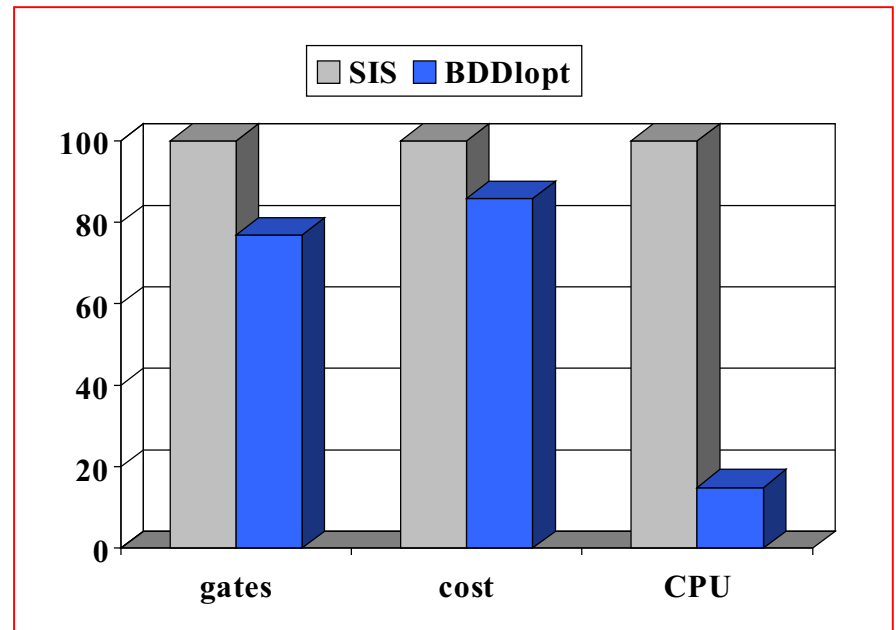


- Keep hash of BDD fragments
  - Equivalence and inverse are trivial since hashes are equal
  - Bias recursive cuts to enhance chance of reuse

# Experimental Results



Average ratio of AND/OR-intensive



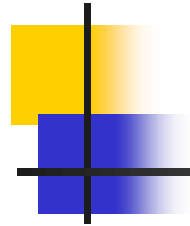
Average ratio of XOR-intensive



# Lessons from BDDlopt

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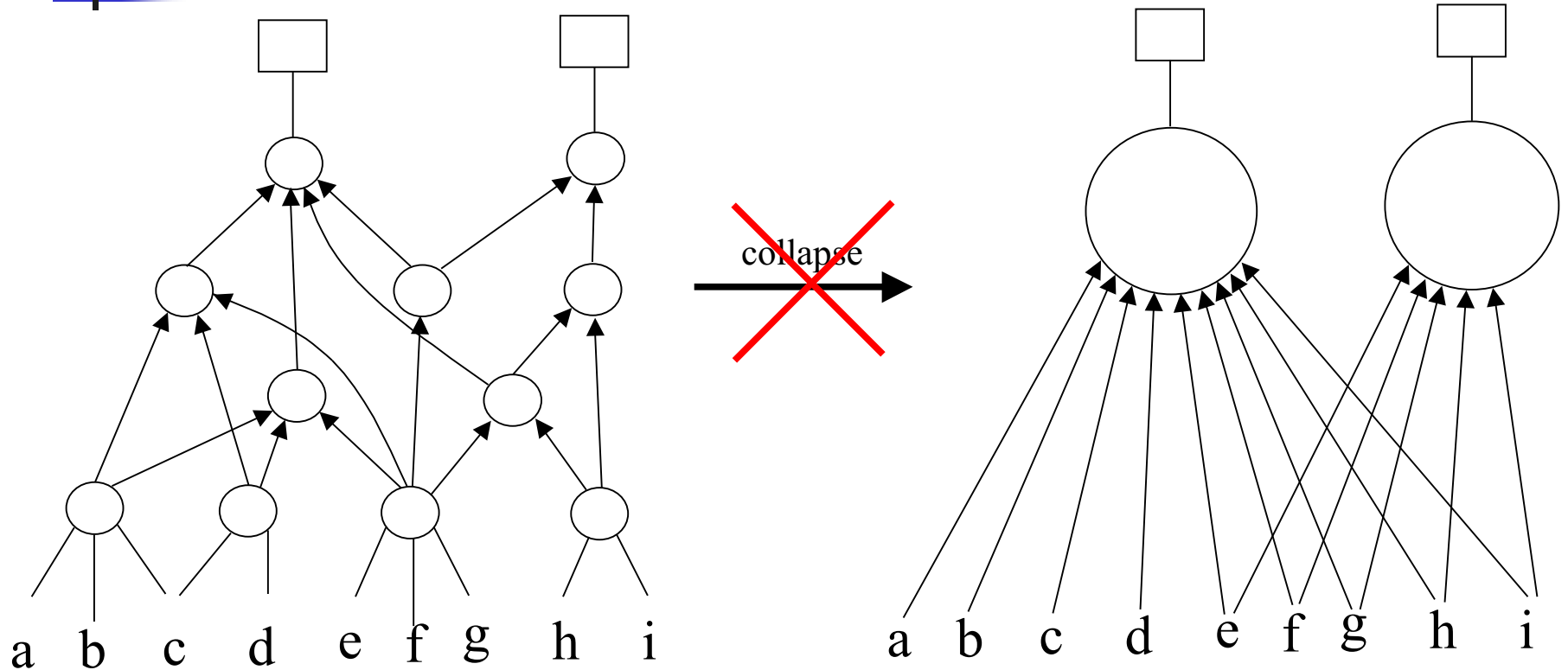
- BDD decomposition is a good alternative to traditional logic optimization.
- BDD decomposition-based logic optimization is fast.
- Stand-alone BDD decomposition is not amenable to large circuits.
  - SCALABILITY



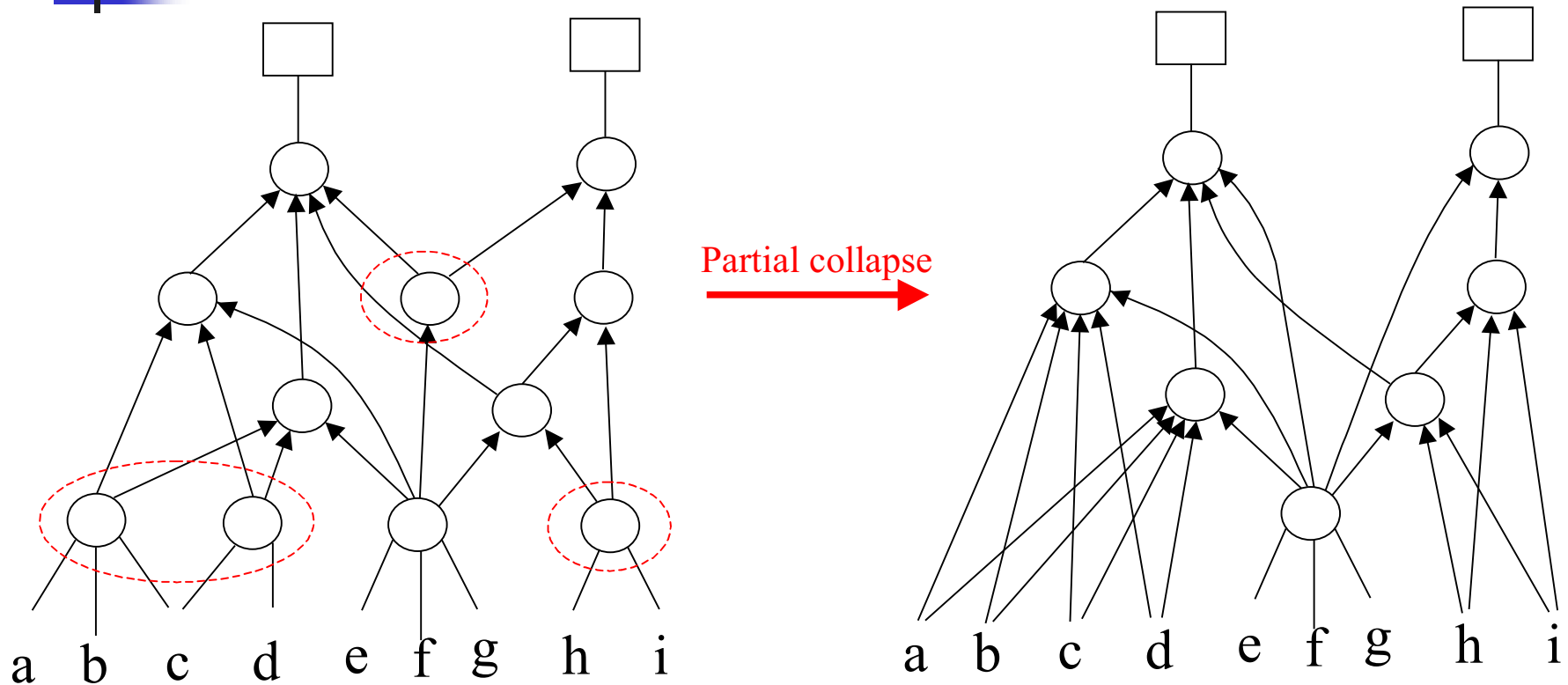
# Local vs. Global BDD Representation

Circuits	Global BDDs	Local BDDs
C1355	33450	893
C1908	6734	1229
C2670	5554	1712
C3540	25828	2326
C432	1226	283
C499	26890	341
C5315	2942	3516
C7552	19322	5012
C880	15004	601
pair	4940	1808
rot	7340	934

# Trade-off Between Local and Global Representation



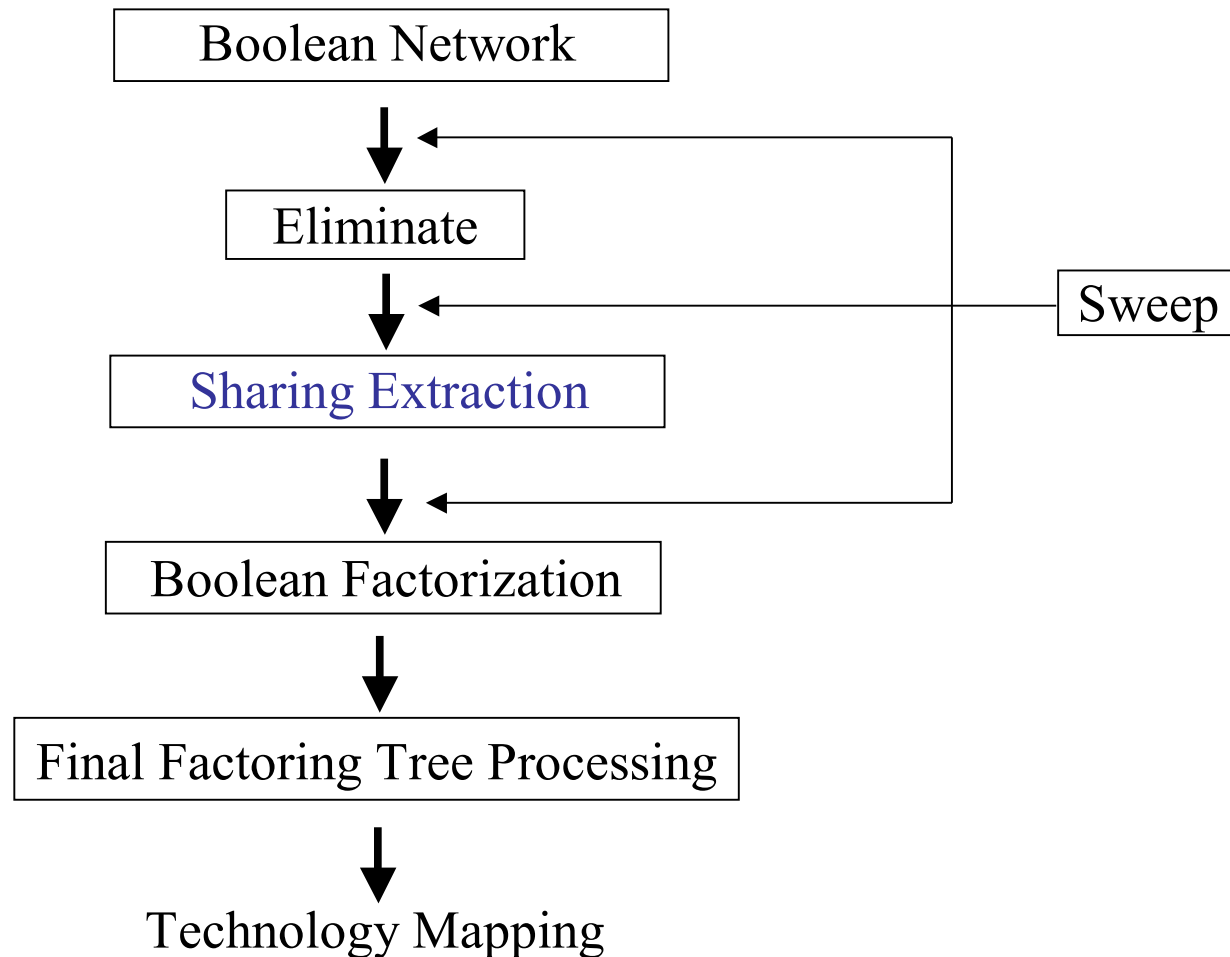
# Trade-off Between Local and Global Representation

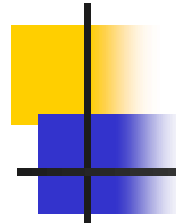






# BDD-Based Logic Synthesis System (BDS)





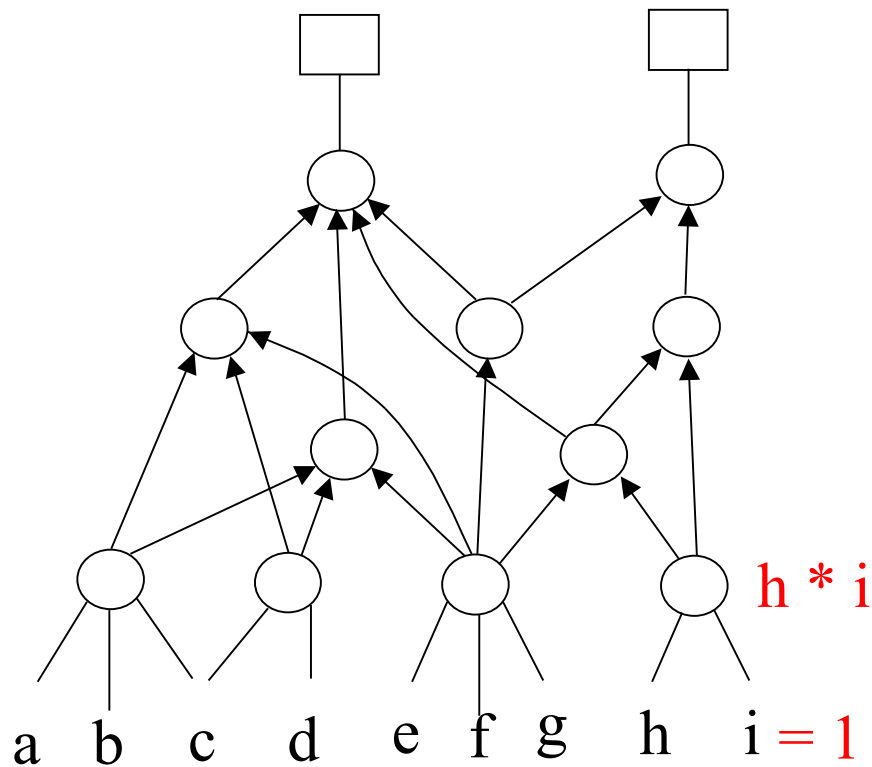
# Sweep Boolean Network

---

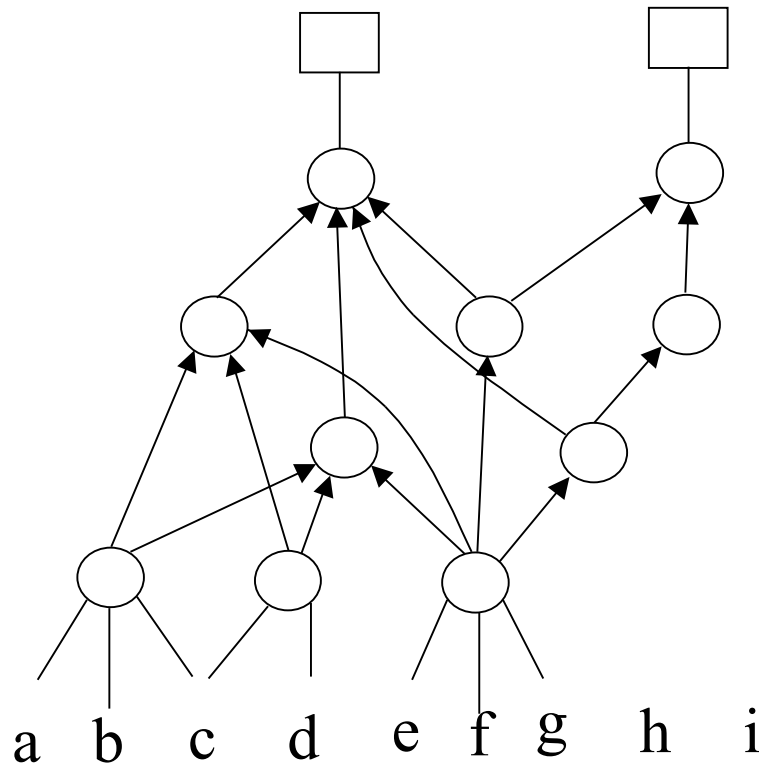
Goal : Network preprocessing.

- Constant propagation.  $a + 1 = 1$ ;  $a + a' = 1$ ; etc
- Remove single-input Boolean nodes.
- Remove functionally duplicated Boolean nodes.

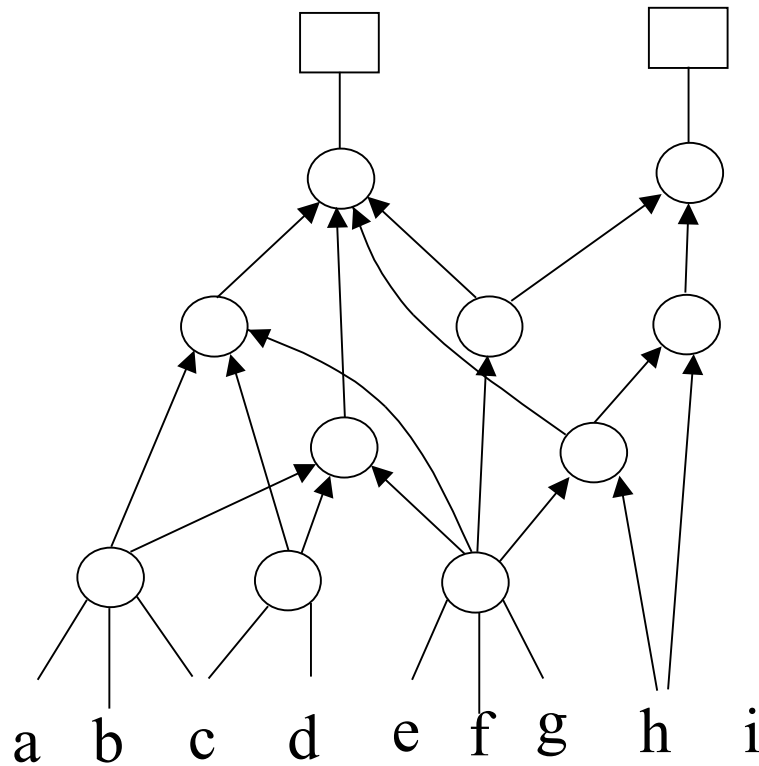
# Sweep Boolean Network (e.g.)



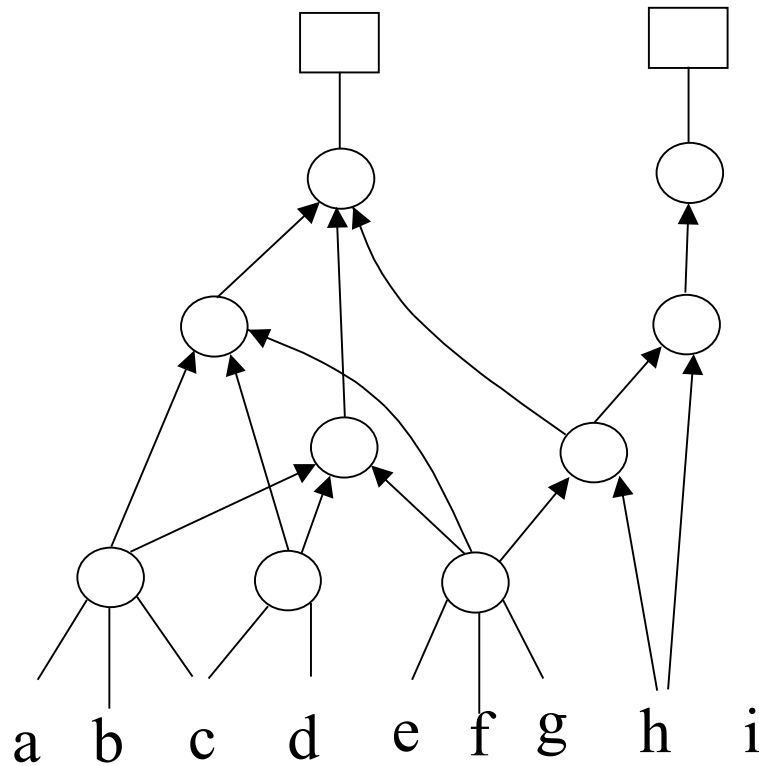
# Sweep Boolean Network



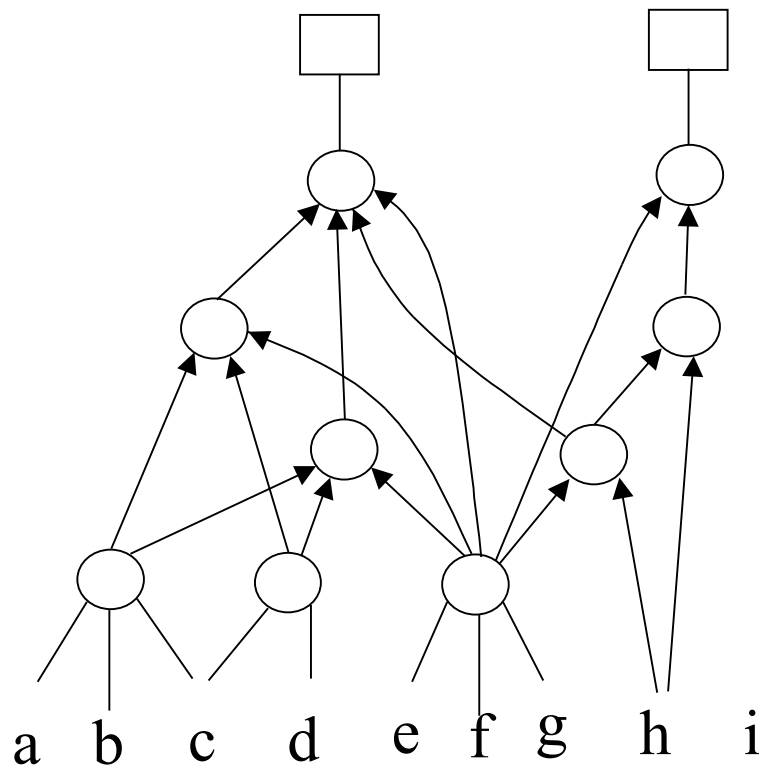
# Sweep Boolean Network

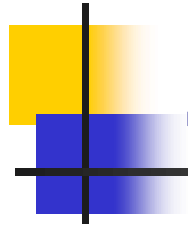


# Sweep Boolean Network



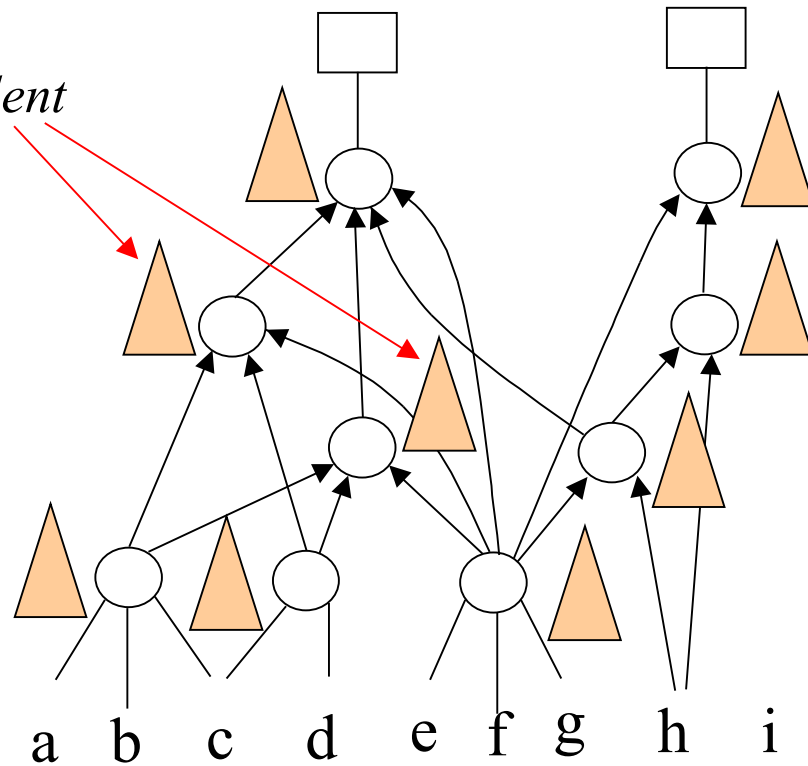
# Sweep Boolean Network



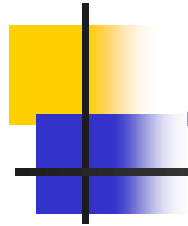


# Sweep Boolean Network

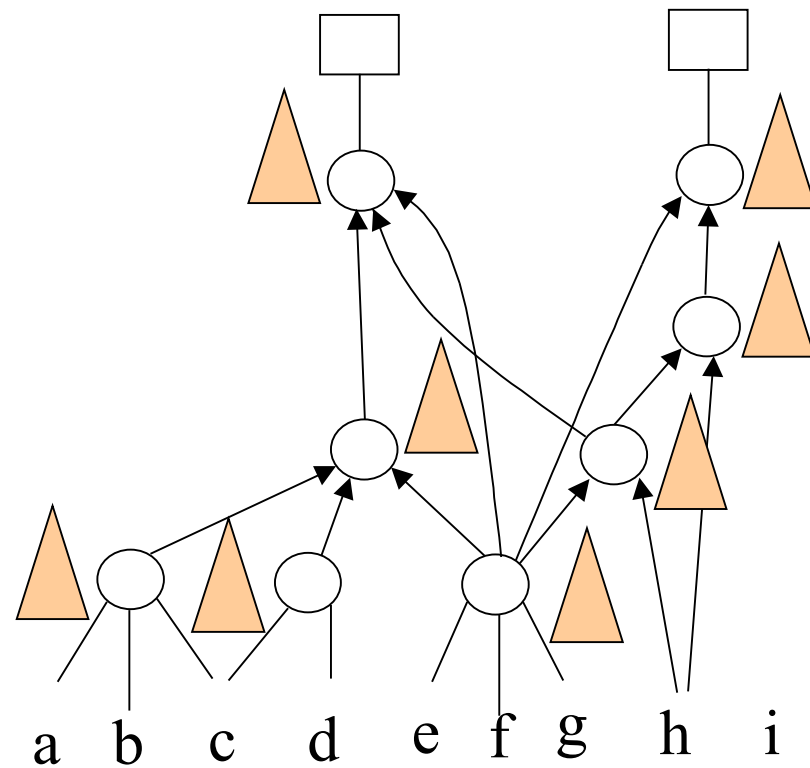
*Equivalent*







# Sweep Boolean Network



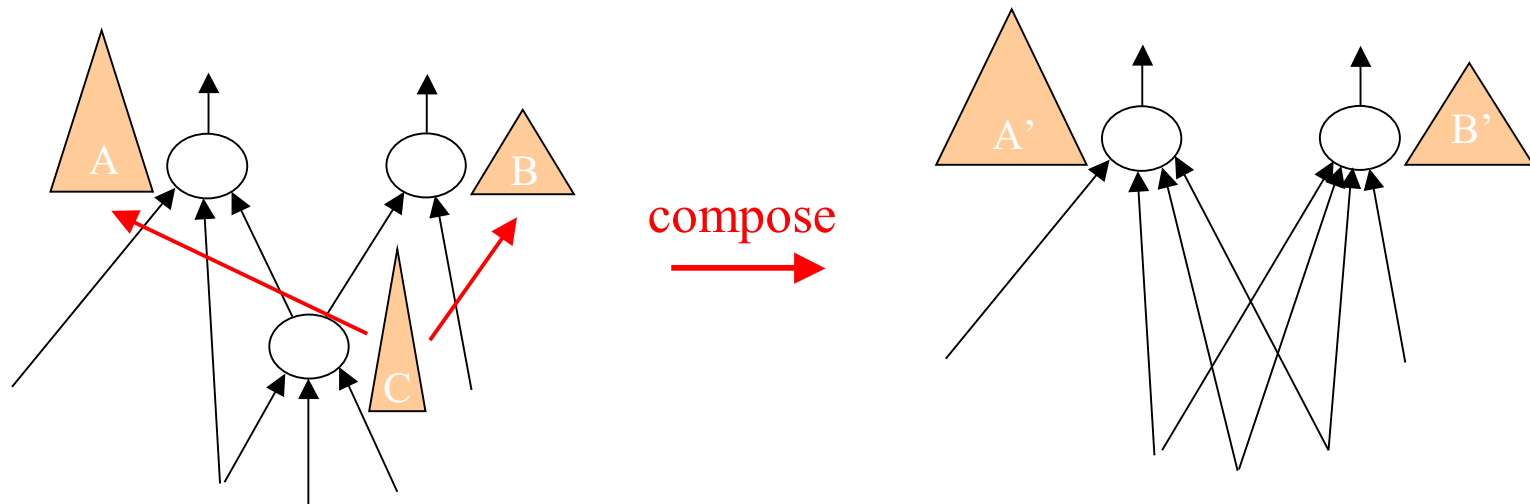


# Sweep Duplicated Nodes

---

Circuits	Total Nodes	Duplicated Nodes
C1908	441	118
C2670	787	72
C3540	956	247
C5315	1467	197
C6228	2353	30
C7552	2165	355
C880	302	10
dalv	985	249
i8	1183	186
i9	329	22
i10	1634	84
pair	830	16
vda	123	3

# Boolean Node Elimination



$$\text{If } (\triangle A' + \triangle B') - (\triangle A + \triangle B + \triangle C) < \textit{threshold}$$



# Eliminate Algorithm

---

## Conventional

```
candidate = collect(bddmgr, network);  
while(candidate) {  
    execute(bddmgr, candidate);  
    reorder(bddmgr);  
    candidate = collect(bddmgr, network);  
}
```

## BDS

```
candidate = collect(bddmgr, network);  
while(candidate) {  
    execute(bddmgr, candidate);  
    newbddmgr = bddMapping(bddmgr, network);  
    reorder(newbddmgr);  
    free(bddmgr);  
    bddmgr = newbddmgr;  
    candidate = collect(bddmgr, network);  
}
```



# Iterative Eliminate Results

---

Circuits	Chaudhry <i>et al</i> [42]		BDS	
	BDD nodes	CPU (s)	BDD nodes	CPU (s)
C1355	211	270	207	0.3
C1908	310	25.4	276	0.6
C2670	615	197	527	1.7
C3540	974	101.7	901	3.2
C432	181	4.5	183	0.5
C499	196	2.4	228	0.2
C5315	1008	307.6	918	4.0
C6288	1677	540.7	1507	4.4
C7552	1592	382.1	1227	6.4
C880	298	7.5	300	0.4
Total	7066	1838.9	6274	21.7

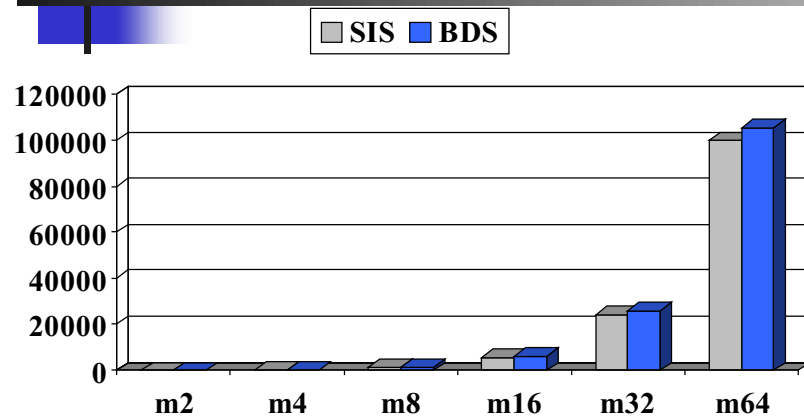


# BDS Experimental Results

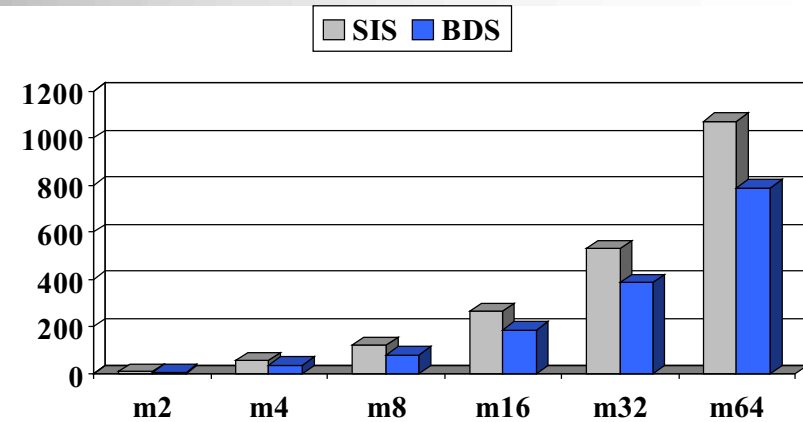
Circuits	SIS				BDS			
	Area	Delay	CPU (s)	Mem (M)	Area	Delay	CPU (s)	Mem (M)
C1355	689	39.40	6.6	1.2	711	45.60	0.4	1.0
C1908	695	68.60	8.1	1.2	730	65.00	0.8	1.0
C3540	1695	81.40	16.1	3.3	1713	81.20	3.6	1.9
C432	290	75.90	46.1	0.7	357	78.40	0.2	0.5
C499	689	39.40	6.8	0.9	708	43.60	0.6	0.5
C5315	2286	68.60	10.2	3.1	2402	70.50	5.3	3.0
C6288	4631	237.8	21.8	4.1	4677	178.3	3.8	1.1
C7552	3038	115.70	54.2	4.9	3112	83.30	4.2	4.8
C880	567	56.10	1.9	1.0	563	43.20	0.7	0.8
pair	2274	74.30	16.1	2.5	2466	52.60	2.1	2.0
rot	965	51.60	4.5	2.0	1025	51.90	1.0	0.9
Total	17819	908.8	192.4	24.9	18464	793.6	22.7	17.5

# Synthesis of Multipliers

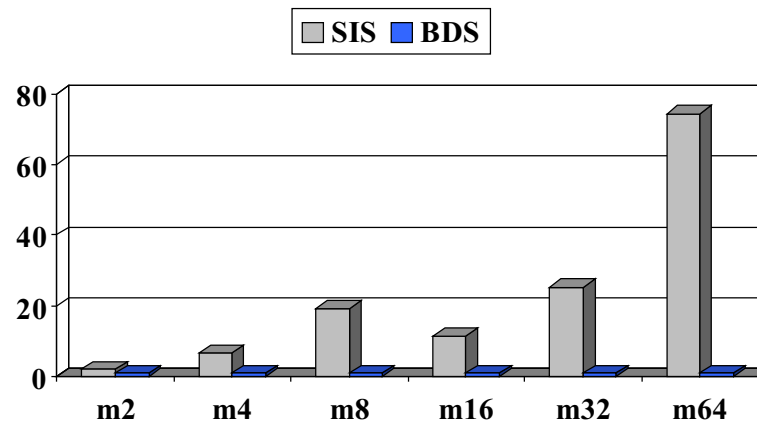
Cost



Delay (ns)

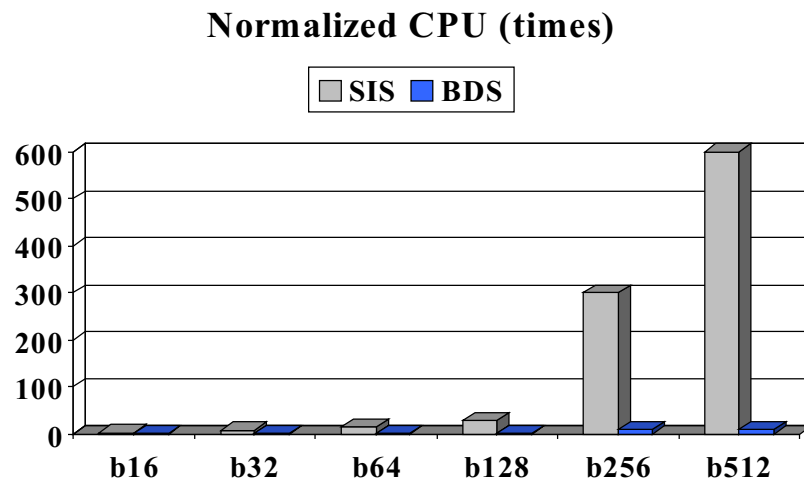
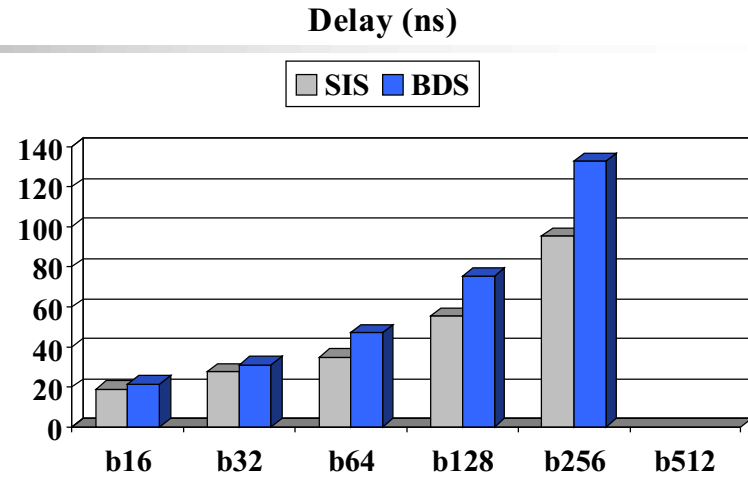
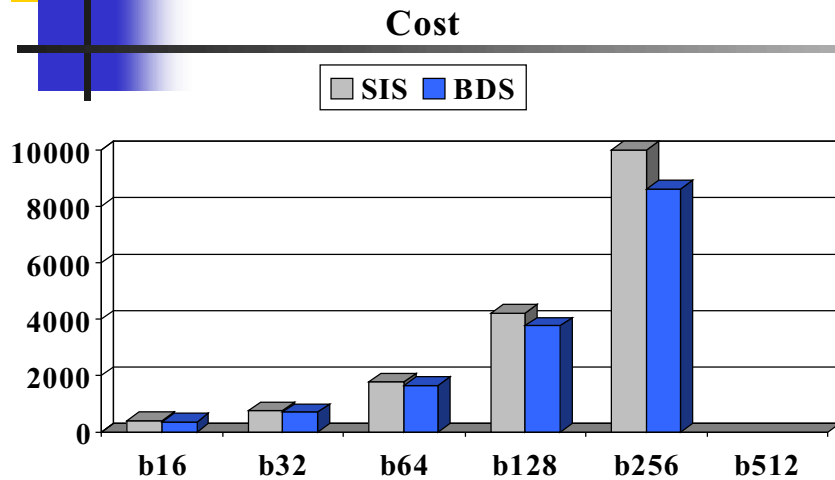


Normalized CPU (times)



Result of  $m64X64$   
Size : ~ 41K  
CPU (SIS) : 6.6 hrs  
CPU (BDS): 5.3 minutes

# Synthesis Results for bshift



Result of *bshift512*  
Size : ~ 7K  
CPU (SIS) : > 15 hrs  
CPU (BDS): 1.5 minutes





# Synthesis for Mixed CMOS/PTL Logic

---

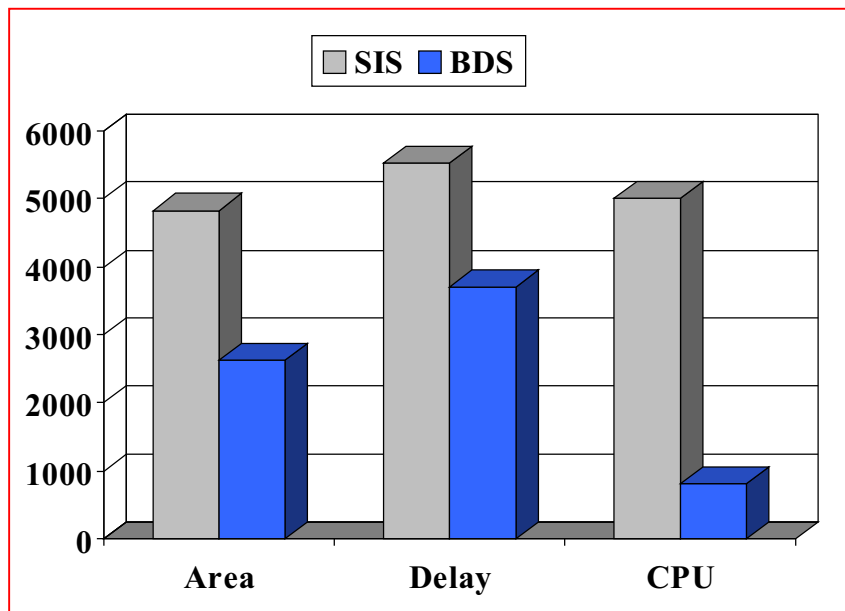
## CMOS

- Concise AND, OR logic
  - High noise immunity
  - Level-restoring capability
- 
- Inefficient MUX, XOR
  - Higher power

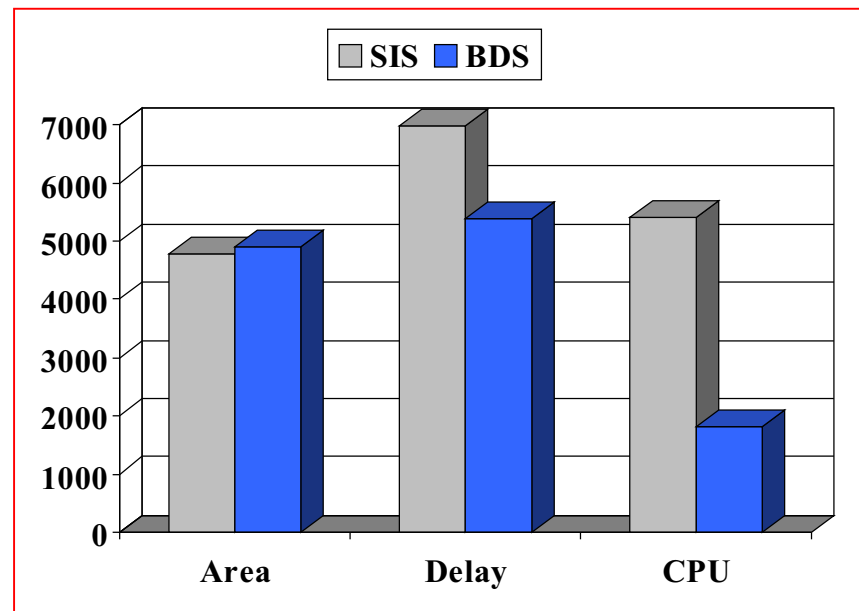
## PTL

- Concise MUX, XOR logic
  - Low power
  - Small area (faster)
- 
- Inefficient AND, OR
  - Low noise immunity

# Results of Mixed CMOS/PTL Synthesis



XOR-intensive functions



AND/OR-intensive functions



# Conclusions

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- New theory of BDD decomposition. For the first time BDD is used to perform Boolean AND/OR, XOR decompositions.
- First unified approach to both AND/OR- and XOR-intensive Boolean functions.
- Implementation of first working BDD-based multi-level logic synthesis system (BDS).
- Efficient BDD manipulation techniques.
- First systematic approach to mixed CMOS/PTL synthesis