



# GRLIB IP Library User's Manual

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# 1 Introduction

## 1.1 Overview

The GRLIB IP Library is an integrated set of reusable IP cores, designed for *system-on-chip* (SOC) development. The IP cores are centered around a common on-chip bus, and use a coherent method for simulation and synthesis. The library is vendor independent, with support for different CAD tools and target technologies. A unique plug&play method is used to configure and connect the IP cores without the need to modify any global resources.

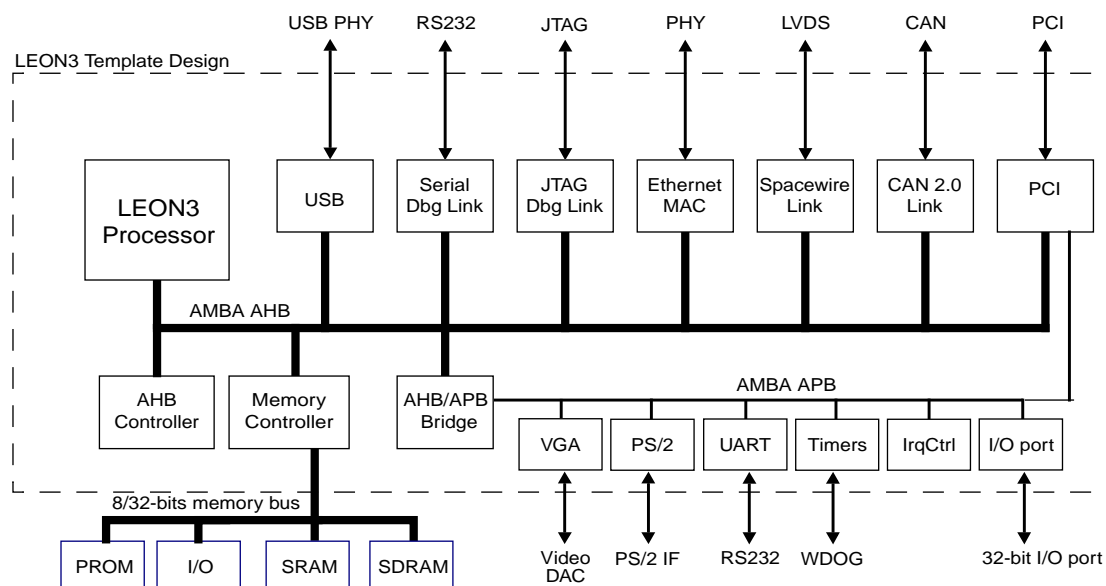
## 1.2 Library organization

GRLIB is organized around VHDL libraries, where each major IP (or IP vendor) is assigned a unique library name. Using separate libraries avoids name clashes between IP cores and hides unnecessary implementation details from the end user. Each VHDL library typically contains a number of packages, declaring the exported IP cores and their interface types. Simulation and synthesis scripts are created automatically by a global makefile. Adding and removing of libraries and packages can be made without modifying any global files, ensuring that modification of one vendor's library will not affect other vendors. A few global libraries are provided to define shared data structures and utility functions.

GRLIB provides automatic script generators for the Modelsim, Ncsim, Aldec, Sonata and GHDL simulators, and the Synopsys, Synplify, Cadence, Mentor, Actel, Altera, Lattice, eASIC and Xilinx implementation tools. Support for other CAD tools can be easily be added.

## 1.3 On-chip bus

The GRLIB is designed to be 'bus-centric', i.e. it is assumed that most of the IP cores will be connected through an on-chip bus. The AMBA-2.0 AHB/APB bus has been selected as the common on-chip bus, due to its market dominance (ARM processors) and because it is well documented and can be used for free without license restrictions. The figure below shows an example of a LEON3 system designed with GRLIB:



## 1.4 Distributed address decoding

Adding an IP core to the AHB bus is unfortunately not as straight-forward as just connecting the bus signals. The address decoding of AHB is centralized, and a shared address decoder and bus multiplexer must be modified each time an IP core is added or removed. To avoid dependencies on a global

resource, distributed address decoding has been added to the GRLIB cores and AMBA AHB/APB controllers.

## 1.5 Interrupt steering

GRLIB provides a unified interrupt handling scheme by adding 32 interrupt signals to the AHB and APB buses. An AMBA module can drive any of the interrupts, and the unit that implements the interrupt controller can monitor the combined interrupt vector and generate the appropriate processor interrupt. In this way, interrupts can be generated regardless of which processor or interrupt controller is being used in the system, and does not need to be explicitly routed to a global resource. The scheme allows interrupts to be shared by several cores and resolved by software.

## 1.6 Plug&Play capability

A broad interpretation of the term ‘plug&play’ is the capability to detect the system hardware configuration through software. Such capability makes it possible to use software application or operating systems which automatically configure themselves to match the underlying hardware. This greatly simplifies the development of software applications, since they do not need to be customized for each particular hardware configuration.

In GRLIB, the plug&play information consists of three items: a unique IP core ID, AHB/APB memory mapping, and used interrupt vector. This information is sent as a constant vector to the bus arbiter/decoder, where it is mapped on a small read-only area in the top of the address space. Any AHB master can read the system configuration using standard bus cycles, and a plug&play operating system can be supported.

To provide the plug&play information from the AMBA units in a harmonized way, a configuration record for AMBA devices has been defined (figure 1). The configuration record consists of 8 32-bit words, where four contain configuration words defining the core type and interrupt routing, and four contain so called ‘bank address registers’ (BAR), defining the memory mapping.

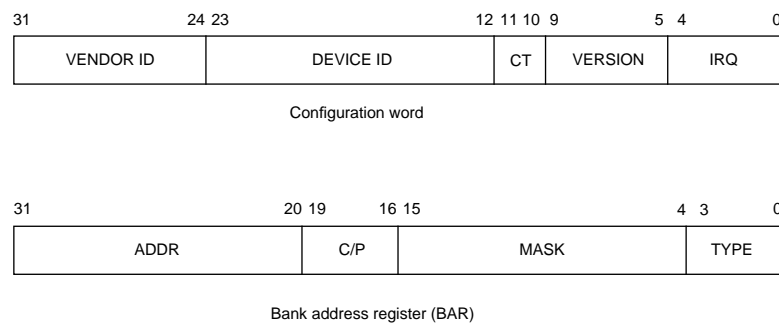


Figure 1. AMBA configuration record

The configuration word for each device includes a vendor ID, device ID, version number, and interrupt routing information. A configuration type indicator is provided to allow for future evolution of the configuration word. The BARs contain the start address for an area allocated to the device, a mask defining the size of the area, information whether the area is cacheable or pre-fetchable, and a type declaration identifying the area as an AHB memory bank, AHB I/O bank or APB I/O bank. The configuration record can contain up to four BARs and the core can thus be mapped on up to four distinct address areas.

## 1.7 Portability

GRLIB is designed to be technology independent, and easily implemented on both ASIC and FPGA technologies. Portability support is provided for components such as single-port RAM, two-port RAM, dual-port RAM, single-port ROM, clock generators and pads. The portability is implemented by means of virtual components with a VHDL generic to select the target technology. In the architecture of the component, VHDL generate statements are used to instantiate the corresponding macro cell from the selected technology library. For RAM cells, generics are also used to specify the address and data widths, and the number of ports.

## **1.8 Available IP cores**

The library includes cores for AMBA AHB/APB control, the LEON3 and LEON4 SPARC processors, IEEE-754 floating-point unit, AHB/ABH bridge, 32-bit PC133 SDRAM controller, DDR1/2 controllers, 32-bit PCI bridge with DMA, 10/100/1000 Mbit ethernet MAC, CAN-2.0 controller, USB-2.0 host and device controllers, 8/16/32-bit PROM/SRAM controller, 32-bit SSRAM controller, 32-bit GPIO port, timer unit, interrupt controller, PS/2 interface, VGA controller and many other legacy cores. Memory generators are available for Actel, Altera, Atmel, Eclipse, Lattice, UMC, Artisan, Virage and Xilinx.

## **1.9 Licensing**

The main infra-structure of GRLIB is released in open-source under the GNU GPL license. This means that designs based on the GPL version of GRLIB must be distributed in full source code under the same license. For commercial applications where source-code distribution is not desirable or possible, Aeroflex Gaisler offers low-cost commercial IP licenses. Contact [sales@gaisler.com](mailto:sales@gaisler.com) for more information or visit <http://www.gaisler.com/>.

## 2 Installation

### 2.1 Installation

GRLIB is distributed as a gzipped tar-file and can be installed in any location on the host system:

```
gunzip -c grlib-gpl-1.1.0-bxxxx.tar.gz | tar xf -
```

or

```
unzip grlib-gpl-1.1.0-bxxxx.zip
```

NOTE: Do NOT use unzip on the .tar.gz file, this will corrupt the files during extraction!

The distribution has the following file hierarchy:

bin	various scripts and tool support files
boards	support files for FPGA prototyping boards
designs	template designs
doc	documentation
lib	VHDL libraries
netlists	Vendor specific mapped netlists
software	software utilities and test benches
verification	test benches

GRLIB uses the GNU ‘make’ utility to generate scripts and to compile and synthesis designs. It must therefore be installed on a unix system or in a ‘unix-like’ environment. Tested hosts systems are Linux and Windows with Cygwin.

### 2.2 Directory organization

GRLIB is organized around VHDL libraries, where each IP vendor is assigned a unique library name. Each vendor is also assigned a unique subdirectory under grlib/lib in which all vendor-specific source files and scripts are contained. The vendor-specific directory can contain subdirectories, to allow for further partitioning between IP cores etc.

The basic directories delivered with GRLIB under grlib-1.0.x/lib are:

grlib	packages with common data types and functions
gaisler	Aeroflex Gaisler’s components and utilities
tech/*	target technology libraries for gate level simulation
techmap	wrappers for technology mapping of marco cells (RAM, pads)
work	components and packages in the VHDL work library

Other vendor-specific directories are also delivered with GRLIB, but are not necessary for the understanding of the design concept. Libraries and IP cores are described in detail in separate documentation.

### 2.3 Host platform support

GRLIB is design to work with a large variety of hosts. The paragraphs below outline the hosts tested by Aeroflex Gaisler. Other unix-based hosts are likely to work but are not tested. As a baseline, the following host software must be installed for the GRLIB configuration scripts to work:

- Bash shell
- GNU make
- GCC
- Tcl/tk-8.4
- patch utility



### **2.3.1 Linux**

The make utility and associated scripts should work on most linux distribution. GRLIB is primarily developed on linux hosts, and linux is the preferred platform.

### **2.3.2 Windows with Cygwin**

The make utility and associated scripts will work, although somewhat slow. Note that gcc and the make utility must be selected during the Cygwin installation. Warning: some versions of Cygwin are know to fail due to a broken 'make' utility. In this case, try to use a different version of Cygwin or update to a newer make.

## 3 LEON3 quick-start guide

### 3.1 Introduction

This chapter will provide a simple quick-start guide on how to implement a leon3 system using GRLIB, and how to download and run software on the target system. Refer to chapters 3 - 6 for a deeper understanding of the GRLIB organization.

### 3.2 Overview

Implementing a leon3 system is typically done using one of the template designs on the designs directory. For this tutorial, we will use the LEON3 template design for the GR-XC3S-1500 board. Implementation is typically done in three basic steps:

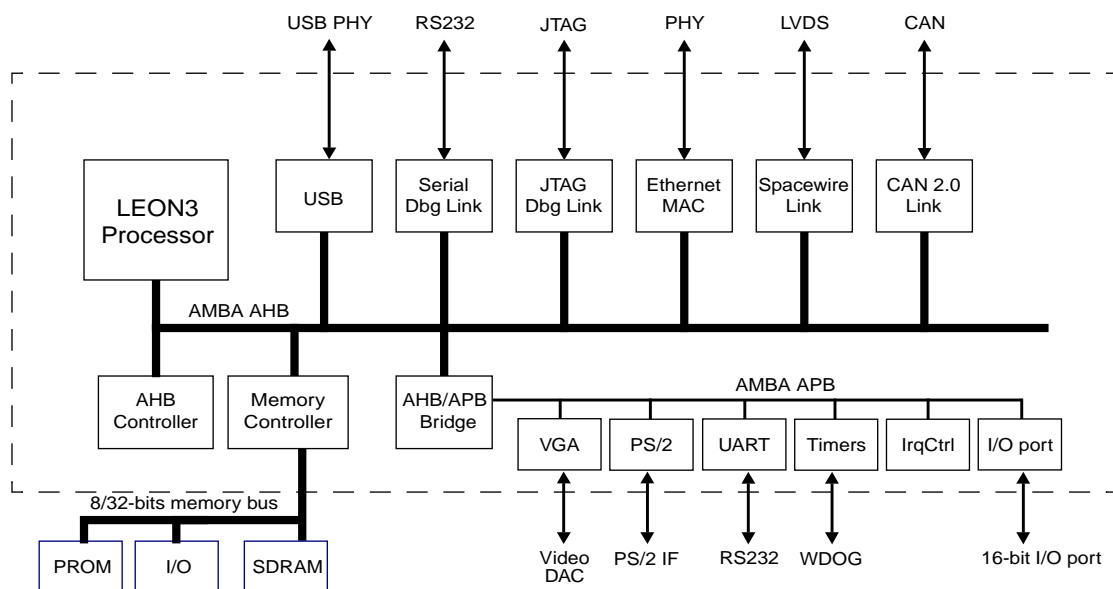
- Configuration of the design using xconfig
- Simulation of design and test bench
- Synthesis and place&route

The template design is located in `designs/leon3-gr-xc3s-1500`, and is based on three files:

- *config.vhd* - a VHDL package containing design configuration parameters. Automatically generated by the xconfig GUI tool.
- *leon3mp.vhd* - contains the top level entity and instantiates all on-chip IP cores. It uses *config.vhd* to configure the instantiated IP cores.
- *testbench.vhd* - test bench with external memory, emulating the GR-XC3S-1500 board.

Each core in the template design is configurable using VHDL generics. The value of these generics is assigned from the constants declared in *config.vhd*, created with the xconfig GUI tool.

LEON3 GR-XC3S-1500 Template Design



### 3.3 Configuration

Change directory to `designs/leon3-gr-xc3s-1500`, and issue the command ‘make xconfig’ in a bash shell (linux) or cygwin shell (windows). This will launch the xconfig GUI tool that can be used to modify the leon3 template design. When the configuration is saved and xconfig is exited, the *config.vhd* is automatically updated with the selected configuration.

### 3.4 Simulation

The template design can be simulated in a test bench that emulates the prototype board. The test bench includes external PROM and SDRAM which are pre-loaded with a test program. The test program will execute on the LEON3 processor, and tests various functionality in the design. The test program will print diagnostics on the simulator console during the execution.

The following command should be give to compile and simulate the template design and test bench:

```
make vsim
vsim testbench
```

A typical simulation log can be seen below.

```
$ vsim testbench
```

```
VSIM 1> run -a
# LEON3 GR-XC3S-1500 Demonstration design
# GRLIB Version 1.0.15, build 2183
# Target technology: spartan3 , memory library: spartan3
# ahbctrl: AHB arbiter/multiplexer rev 1
# ahbctrl: Common I/O area disabled
# ahbctrl: AHB masters: 4, AHB slaves: 8
# ahbctrl: Configuration area at 0xfffff000, 4 kbyte
# ahbctrl: mst0: Gaisler Research      Leon3 SPARC V8 Processor
# ahbctrl: mst1: Gaisler Research      JTAG Debug Link
# ahbctrl: mst2: Gaisler Research      SpaceWire Serial Link
# ahbctrl: mst3: Gaisler Research      SpaceWire Serial Link
# ahbctrl: slv0: European Space Agency Leon2 Memory Controller
# ahbctrl:      memory at 0x00000000, size 512 Mbyte, cacheable, prefetch
# ahbctrl:      memory at 0x20000000, size 512 Mbyte
# ahbctrl:      memory at 0x40000000, size 1024 Mbyte, cacheable, prefetch
# ahbctrl: slv1: Gaisler Research      AHB/APB Bridge
# ahbctrl:      memory at 0x80000000, size 1 Mbyte
# ahbctrl: slv2: Gaisler Research      Leon3 Debug Support Unit
# ahbctrl:      memory at 0x90000000, size 256 Mbyte
# apbctrl: APB Bridge at 0x80000000 rev 1
# apbctrl: slv0: European Space Agency Leon2 Memory Controller
# apbctrl:      I/O ports at 0x80000000, size 256 byte
# apbctrl: slv1: Gaisler Research      Generic UART
# apbctrl:      I/O ports at 0x80000100, size 256 byte
# apbctrl: slv2: Gaisler Research      Multi-processor Interrupt Ctrl.
# apbctrl:      I/O ports at 0x80000200, size 256 byte
# apbctrl: slv3: Gaisler Research      Modular Timer Unit
# apbctrl:      I/O ports at 0x80000300, size 256 byte
# apbctrl: slv8: Gaisler Research      General Purpose I/O port
# apbctrl:      I/O ports at 0x80000800, size 256 byte
# apbctrl: slv12: Gaisler Research     SpaceWire Serial Link
# apbctrl:      I/O ports at 0x80000c00, size 256 byte
# apbctrl: slv13: Gaisler Research     SpaceWire Serial Link
# apbctrl:      I/O ports at 0x80000d00, size 256 byte
# grspw13: Spacewire link rev 0, AHB fifos 2x64 bytes, rx fifo 16 bytes, irq 11
# grspw12: Spacewire link rev 0, AHB fifos 2x64 bytes, rx fifo 16 bytes, irq 10
# grgpio8: 18-bit GPIO Unit rev 0
# gptimer3: GR Timer Unit rev 0, 8-bit scaler, 2 32-bit timers, irq 8
# irqmp: Multi-processor Interrupt Controller rev 3, #cpu 1
# apbuart1: Generic UART rev 1, fifo 1, irq 2
# ahbjtag: AHB Debug JTAG rev 0
# dsu3_2: LEON3 Debug support unit + AHB Trace Buffer, 2 kbytes
# leon3_0: LEON3 SPARC V8 processor rev 0
# leon3_0: icache 1*8 kbyte, dcache 1*4 kbyte
# clkgen_spartan3e: spartan3/e sdram/pci clock generator, version 1
# clkgen_spartan3e: Frequency 50000 KHz, DCM divisor 4/5
#
# **** GRLIB system test starting ****
# Leon3 SPARC V8 Processor
#   CPU#0 register file
#   CPU#0 multiplier
#   CPU#0 radix-2 divider
#   CPU#0 floating-point unit
```

```
# CPU#0 cache system
# Multi-processor Interrupt Ctrl.
# Generic UART
# Modular Timer Unit
# timer 1
# timer 2
# chain mode
# Test passed, halting with IU error mode
# ** Failure: *** IU in error mode, simulation halted ***
# Time: 1104788 ns Iteration: 0 Process: /testbench/iuerr File: testbench.vhd
# Stopped at testbench.vhd line 338
VSIM 2>
```

The test program executed by the test bench consists of two parts, a simple prom boot loader (prom.S) and the test program itself (systest.c). Both parts can be re-compiled using the ‘make soft’ command. This requires that the BCC tool-chain is installed on the host computer.

Note that the simulation is terminated by generating a VHDL failure, which is the only way of stopping the simulation from inside the model. An error message is then printed:

```
# Test passed, halting with IU error mode
# ** Failure: *** IU in error mode, simulation halted ***
# Time: 1104788 ns Iteration: 0 Process: /testbench/iuerr File: testbench.vhd
# Stopped at testbench.vhd line 338
```

This error can be ignored.

### 3.5 Synthesis and place&route

The template design can be synthesized with either Synplify, Precision or ISE/XST. Synthesis can be done in batch or interactively. To use synplify in batch mode, use the command:

```
make synplify
```

To use synplify interactively, use:

```
make synplify-launch
```

The corresponding command for ISE are:

```
make ise-map
and
make ise-launch
```

To perform place&route for a netlist generated with synplify, use:

```
make ise-synp
```

For a netlist generated with XST, use:

```
make ise
```

In both cases, the final programming file will be called ‘leon3mp.bit’. See the GRLIB User’s Manual chapter 3 for details on simulation and synthesis script files.

### 3.6 Simulation of post-synthesis netlist

If desired, it is possible to simulate the synthesized netlist in the test bench. The synplify synthesis tool generates a VHDL netlist in the file synplify/leon3mp.vhm. To re-run the test bench with the netlist, do as follows:

```
vcom synplify/leon3mp.vhm
vsim -c testbench
vsim> run -all
```

### 3.7 Board re-programming

The GR-XC3S-1500 FPGA configuration PROMs can be programmed from the shell window with the following command:

```
make ise-prog-prom
```

For interactive programming, use Xilinx Impact software. See the GR-XC3S-1500 Manual for details on which configuration PROMs to specify.

A pre-compiled FPGA bit file is provided in the bitfiles directory, and the board can be re-programmed with this bit file using:

```
make ise-prog-prom-ref
```

### 3.8 Running applications on target

To download and debug applications on the target board, GRMON debug monitor is used. GRMON can be connected to the target using RS232, JTAG, ethernet or USB. The most convenient way is probably to use JTAG. GRMON can be connected to the target using the Xilinx parallel port cable programming cable and or the Platform USB cable. See the GRMON manual for details. To connect using the parallel port cable, do:

```
grmon -jtag -u
```

This should print the configuration of the target board:

```
initialising .....
detected frequency: 40 MHz
```

Component	Vendor
LEON3 SPARC V8 Processor	Gaisler Research
AHB Debug UART	Gaisler Research
AHB Debug JTAG TAP	Gaisler Research
SVGA frame buffer	Gaisler Research
GR Ethernet MAC	Gaisler Research
AHB ROM	Gaisler Research
AHB/APB Bridge	Gaisler Research
LEON3 Debug Support Unit	Gaisler Research
DDR266 Controller	Gaisler Research
Generic APB UART	Gaisler Research
Multi-processor Interrupt Ctrl	Gaisler Research
Modular Timer Unit	Gaisler Research
Keyboard PS/2 interface	Gaisler Research
Keyboard PS/2 interface	Gaisler Research

To download an application, use the 'load' command. To run it, use 'run' :

```
load stanford.exe
run
```

The console output will occur in the grmon window if grmon was started with -u, otherwise it will be sent to the RS232 connector of the board.

### 3.9 Flash PROM programming

The GR-XC3S-1500 board has a 64 Mbit (8Mx8) Intel flash PROM for LEON3 application software. A PROM image is typically created with the sparc-elf-mkprom utility provided with the BCC tool chain. The suitable mkprom parameters for the GR-XC3S-1500 board are:

```
sparc-elf-mkprom -romws 4 -freq 40 -col 9 -nosram -sdram 64 -msoft-float -baud 38400
```

Note that the -freq option should reflect the selected processor frequency, which depends on the clock generator settings. If the processor includes an FPU, the -msoft-float switch can be omitted.

Once the PROM image has been created, the on-board flash PROM can be programmed through GRMON. The procedure is described in the GRMON manual, below is the required GRMON command sequence:

```
flash erase all
flash load prom.out
```

## 4 Implementation flow

### 4.1 Introduction

The following sections will describe how simulation and synthesis is performed using the GRLIB make system. It is recommended to try out the various commands on one of the template designs, such as designs/leon3mp.

### 4.2 Using Makefiles and generating scripts

GRLIB consists of a set of VHDL libraries from which IP cores are instantiated into a local design. GRLIB is designed to reside in a global location and to be used in read-only mode. All compilation, simulation and synthesis is done in a local design directory, using tool-specific scripts. The GRLIB IP cores (components) are instantiated in the local design by the inclusion of various GRLIB packages, declaring the components and associated data types.

A design typically contains of one or more VHDL files, and a local makefile:

```
bash$ ls -g mydesign
-rw-r--r--      1 users          1776 May 25 10:37 Makefile
-rw-r--r--      1 users        12406 May 25 10:46 mydesign.vhd
```

The GRLIB files are accessed through the environment variable GRLIB. This variable can either be set in the local shell or in a local makefile, since the ‘make’ utility is used to automate various common tasks. A GRLIB-specific makefile is located in bin/Makefile. To avoid having to specify the GRLIB makefile using the -f option, the local makefile should include the GRLIB makefile:

```
GRLIB=../../grib
include $(GRLIB)/bin/Makefile
```

Running ‘make help’ with this makefile will print a short menu:

```
$ make help
```

```
interactive targets:
```

```
make avhdl-launch      : start active-hdl gui mode
make riviera-launch    : start riviera
make vsim-launch       : start modelsim
make ncsim-launch      : compile design using ncsim
make sonata-launch     : compile design using sonata
make actel-launch-synp : start Actel Designer for current project
make ise-launch        : start ISE project navigator for XST project
make ise-launch-synp   : start ISE project navigator for synplify project
make quartus-launch    : start Quartus for current project
make quartus-launch-synp : start Quartus for synplify project
make synplify-launch   : start synplify
make vivado-launch     : start Vivado project navigator
make planAhead-launch  : start PlanAhead project navigator
make xgrlib            : start grlib GUI
```

```
batch targets:
```

```
make avhdl            : compile design using active-hdl gui mode
make vsimsa           : compile design using active-hdl batch mode
make riviera          : compile design using riviera
make sonata           : compile design using sonata
make vsim             : compile design using modelsim
make ncsim            : compile design using ncsim
make ghdl             : compile design using GHDL
make actel            : synthesize with synplify, place&route Actel Designer
make ise              : synthesize and place&route with Xilinx ISE
make ise-map          : synthesize design using Xilinx XST
make ise-prec         : synthesize with precision, place&route with Xilinx ISE
make ise-synp         : synthesize with synplify, place&route with Xilinx ISE
make isp-synp         : synthesize with synplify, place&route with ISPLever
make quartus          : synthesize and place&route using Quartus
make quartus-map      : synthesize design using Quartus
make quartus-synp     : synthesize with synplify, place&route with Quartus
make precision        : synthesize design using precision
make synplify         : synthesize design using synplify
make scripts          : generate compile scripts only
make vivado           : synthesize and place&route with Xilinx Vivado
make planAhead        : synthesize and place&route with Xilinx PlanAhead
make clean            : remove all temporary files except scripts
make distclean        : remove all temporary files
```

Generating tool-specific compile scripts can be done as follows:

```
$ make scripts
$ ls compile.*
compile.dc  compile.ncsim  compile.synp  compile.vsim  compile.xst  compile.ghdl
```

The local makefile is primarily used to generate tool-specific compile scripts and project files, but can also be used to compile and synthesize the current design. To do this, additional settings in the makefile are needed. The makefile in the design template `grib/designs/leon3mp` can be seen as an example:

```
$ cd grlib/designs/leon3mp
$ cat Makefile
GRLIB=../..
TOP=leon3mp
BOARD=gr-pci-xc2v
include $(GRLIB)/boards/$(BOARD)/Makefile.inc
DEVICE=$(PART)-$(PACKAGE)$(SPEED)
UCF=$(GRLIB)/boards/$(BOARD)/$(TOP).ucf
QSF=$(BOARD).qsf
EFFORT=1
VHDSYNFILES=config.vhd leon3mp.vhd
VHDSIMFILES=testbench.vhd
SIMTOP=testbench
SDCFIL=$(GRLIB)/boards/$(BOARD)/default.sdc
BITGEN=$(GRLIB)/boards/$(BOARD)/default.ut
CLEAN=local-clean
include $(GRLIB)/bin/Makefile
```

The table below summarizes the common (target independent) ‘make’ targets:

**TABLE 1. Common make targets**

Make target	Description
scripts	Generate GRLIB compile scripts for all supported tools
xconfig	Run the graphic configuration tool (leon3 designs)
clean	Remove all temporary files except scripts and project files
distclean	Remove all temporary files
xgrib	Run the graphical implementation tool (see page 29)

Simulation, synthesis and place&route of GRLIB designs can also be done using a graphical tool called **xgrib**. This tool is described further in chapter “XGrib graphical implementation tool” on page 39.

### 4.3 Simulating a design

The ‘make scripts’ command will generate compile scripts and/or project files for the modelsim, ncsim and ghdl simulators. This is done by scanning GRLIB for simulation files according to the method described in “GRLIB organisation” on page 69. These scripts are then used by further make targets to build and update a GRLIB-based design and its test bench. The local makefile should set the VHDSYNFILES to contain all synthesizable VHDL files of the local design. Likewise, the VHDL-SIMFILES variable should be set to contain all local design files to be used for simulation only. The variable TOP should be set to the name of the top level design entity, and the variable SIMTOP should be set to the name of the top level simulation entity (e.g. the test bench).

```
VHDSYNFILES=config.vhd ahbrom.vhd leon3mp.vhd
VHDLSIMFILES=testbench.vhd
TOP=leon3mp
SIMTOP=testbench
```

The variables must be set before the GRLIB makefile is included, as in the example above.

All local design files are compiled into the VHDL work library, while the GRLIB cores are compiled into their respective VHDL libraries.

The following simulator are currently supported by GRLIB:

**TABLE 2. Supported simulators**

Simulator	Comments
GNU VHDL (GHDL)	version 0.25, VHDL only
Aldec Active-HDL	batch and GUI
Aldec Riviera	
Mentor Modelsim version	version 6.1e or later
Cadence NcSim	IUS-5.8-sp3 and later
Synphony-EDA Sonata	verison 3.1 or later, VHDL only
Xilinx ISIM	ISE-13 or later



## 4.4 Synthesis and place&route

The **make scripts** command will scan the GRLIB files and generate compile and project files for all supported synthesis tools. For this to work, a number of variables must be set in the local makefile:

```
TOP=leon3mp
TECHNOLOGY=virtex2
PART=xc2v3000
PACKAGE=fg676
SPEED=-4
VHDL SYNFILES=config.vhd ahbrom.vhd leon3mp.vhd
SDCFILE=
XSTOPT=-resource_sharing no
DEVICE=xc2v3000-fg676-4
UCF=default.ucf
EFFORT=std
BITGEN=default.ut
```

The TOP variable should be set to the top level entity name to be synthesized. TECHNOLOGY, PART, PACKAGE and SPEED should indicate the target device parameters. VHDL SYNFILES should be set to all local design files that should be used for synthesis. SDCFILE should be set to the (optional) Synplify constraints file, while XSTOPT should indicate additional XST synthesis options. The UCF variable should indicate the Xilinx constraint file, while QSF should indicate the Quartus constraint file. The EFFORT variable indicates the Xilinx place&route effort and the BITGEN variable defines the input script for Xilinx bitfile generation.

The technology related variables are often defined in a makefile include file in the board support packages under GRLIB/boards. When a supported board is targeted, the local makefile can include the board include file to make the design more portable:

```
BOARD=gr-pci-xc2v
include $(GRLIB)/boards/$(BOARD)/Makefile.inc
SDCFILE=$(GRLIB)/boards/$(BOARD)/$(TOP).sdc
UCF=$(GRLIB)/boards/$(BOARD)/$(TOP).ucf
DEVICE=$(PART)-$(PACKAGE)-$(SPEED)
```

The following synthesis tools are currently supported by GRLIB:

**TABLE 3. Supported synthesis and place&route tools**

Synthesis and place&route tool	Recommended version
Actel Designer/Libero	version 91.
Altera Quartus	version 10
Cadence RTL	version 6.1
Lattice Diamond	version 1.3
Mentor Leonardo Precision	2011a.61
Synopsys DC	2010.12
Synplify	2010.03
Xilinx ISE/XST	version 13.3
Xilinx Vivado	2012.3
Xilinx PlanAhead	version 14.3

## 4.5 Skipping unused libraries, directories and files

GRLIB contains a large amount of files, and creating scripts and compiling models might take some time. To speed up this process, it is possible to skip whole libraries, directories or individual files from being included in the tool scripts. Skipping VHDL libraries is done by defining the constant LIBSKIP in the Makefile of the current design, before the inclusion of the GRLIB global Makefile.

To skip a directory in a library, variable DIRSKIP should be used. All directories with the defined names will be excluded when the tool scripts are built. In this way, cores which are not used in the current design can be excluded from the scripts. To skip an individual file, the variable FILESKIP should be set to the file(s) that should be skipped. Below is an example from the leon3-digilent-xc3s1000 template design. All target technology libraries except unisim (Xilinx) are skipped, as well as cores such as PCI, DDR and Spacewire. Care has to be taken to skip all dependent directories when a library is skipped.

```
LIBSKIP = core1553bbc core1553brm core1553brt gr1553 corePCIF \
          tmtc openchip micron hynix cypress ihp opencores spw
DIRSKIP = b1553 pcif leon2 leon2ft crypto satcan pci leon3ft ambatest \
          spacewire ddr can usb ata
FILESKIP = grcan.vhd

include $(GRLIB)/bin/Makefile
```

By default, all technology cells and mapping wrappers are included in the scripts and later compiled. To select only one or a sub-set of technologies, the variable TECHLIBS can be set in the makefile:

```
TECHLIBS = unisim
```

The table below shows which libraries should added to TECHLIBS for each supported technology.

**TABLE 4. TECHLIB settings for various target technologies**

Technology	TECHLIBS defines
Xilinx (All)	unisim
Altera Stratix-II	altera altera_mf stratixii
Altera Cyclone-III	altera altera_mf cycloneiii
Altera Stratix-III	altera altera_mf stratixiii
Altera others	altera altera_mf
Actel Axcelerator	axcelerator
Actel Axcelerator DSP	axcelerator
Actel Proasic3	proasic3
Actel Proasic3e	proasic3e
Actel Proasic3l	proasic3l
Actel Fusion	fusion
Lattice	ec
Quicklogic	eclipsee
Atmel ATC18	atc18 virage
Atmel ATC18RHA	atc18rha_cell
eASIC 90 nm	nextreme
eASIC 45 nm	nextreme2
IHP 0.25	ihp25
IHP 0.25 RH	sgb25vrh
Aeroflex 0.25 RH	ut025crh
Aeroflex 0.13 RH	ut130hbd
Ramon 0.18 RH	rh_lib18t
UMC 0.18 um	umc18
UMC 0.18 um DARE	dare
TSMC 90 nm	tsmc90

Note that some technologies are not available in the GPL version. Contact Aeroflex Gaisler for details.

## 4.6 Encrypted RTL

GRLIB supports encrypted script generation to include encrypted RTL files. The information in this section is applicable if you have purchased GRLIB IP cores that are delivered as encrypted RTL. The open source (GPL) release of GRLIB does not include any encrypted RTL.

There are several different solutions for IP protection available from the EDA vendors. Standardisation work is ongoing but at the time of writing it is not possible to generate one encrypted RTL file that can be used with tools from all vendors. Because of this, encrypted RTL is delivered in several versions. All versions contain the same RTL but in different containers to be used with a specific EDA tool.

Currently the GRLIB script generation supports IP protection (encrypted RTL) for the following tools:

- Cadence tools supporting Cadence IP protection (proprietary and IEEE-P1735)

- Mentor Graphics tools with support for IEEE-P1735

- Synopsys Design Compiler with support for IEEE-P1735

- Synopsys Synplify with support for IEEE-P1735

The list above may become outdated. Please contact Aeroflex Gaisler to ensure that your EDA tools are capable of working with GRLIB and encrypted RTL. Specify which tools you will use at the time of order when placing an order for IP cores that are delivered as encrypted RTL.

The RTL source is not available for viewing and simulator views are restricted when using components that are delivered as encrypted RTL.

## 4.7 Tool-specific usage

### 4.7.1 GNU VHDL (GHDL)

GHDL is the GNU VHDL compiler/simulator, available from <http://ghdl.free.fr/>. It is used mainly on linux hosts, although a port to windows/cygwin has recently been reported.

The complete GRLIB as well as the local design are compiled by **make ghdl**. The simulation models will be stored locally in a sub-directory (`./gnu`). A `ghdl.path` file will be created automatically, containing the proper VHDL library mapping definitions. A sub-sequent invocation of **make ghdl** will re-analyze any outdated files in the WORK library using a makefile created with '`ghdl --gen-makefile`'. GRLIB files will not be re-analyzed without a **make ghdl-clean** first.

GHDL creates an executable with the name of the SIMTOP variable. Simulation is started by directly executing the created binary:

```
$ ./testbench
```

**TABLE 5. GHDL make targets**

Make target	Description
ghdl	Compile or re-analyze local design
ghdl-clean	Remove compiled models and temporary files
ghdl-run	Run test bench in batchmode

**TABLE 6. GHDL scripts and files**

File	Description
compile.ghdl	Compile script for GRLIB files
make.ghdl	Makefile to rebuild local design
gnu	Directory with compiled models
SIMTOP	Executable simulation model of test bench

### 4.7.2 Cadence ncsim

The complete GRLIB as well as the local design are compiled and elaborated in batch mode by **make ncsim**. The simulation models will be stored locally in a sub-directory (`./xncsim`). A `cds.lib` file will be created automatically, containing the proper VHDL library mapping definitions, as well as an empty `hdl.var`. Simulation can then be started by using **make ncsim-launch**.

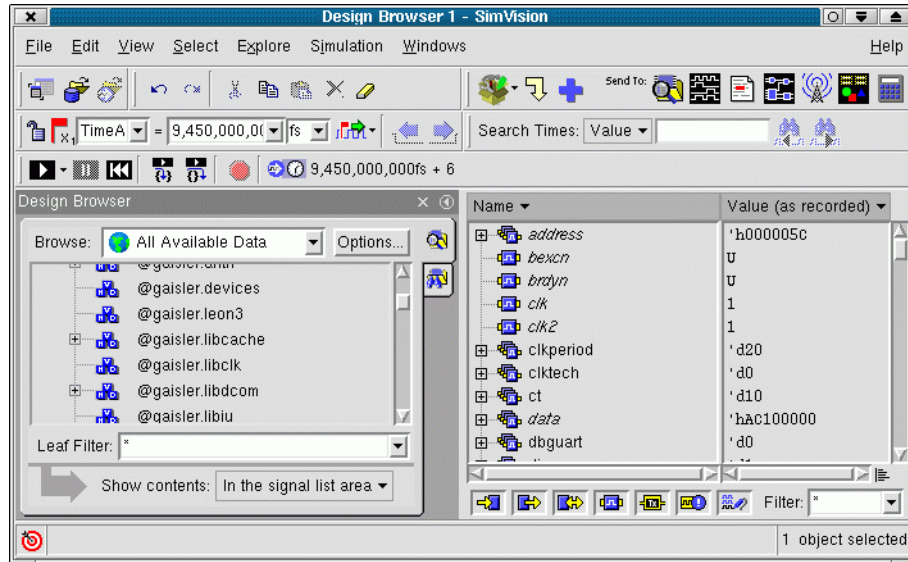


Figure 2. Ncsim graphical user interface

To rebuild the local design, run **make ncsim** again. This will use the `ncupdate` utility to rebuild out-of-date files. The tables below summarize the make targets and the files created by make scripts.

TABLE 7. Ncsim make targets

Make target	Description
<code>ncsim</code>	Compile or re-analyze GRLIB and local design
<code>ncsim-clean</code>	Remove compiled models and temporary files
<code>ncsim-launch</code>	Start modelsim GUI on current test bench
<code>ncsim-run</code>	Run test bench in batchmode

TABLE 8. Ncsim scripts and files

File	Description
<code>compile.ncsim</code>	Compile script for GRLIB files
<code>make.ncsim</code>	Makefile to rebuild GRLIB and local design
<code>xncsim</code>	Directory with compiled models

### 4.7.3 Mentor ModelSim

The complete GRLIB as well as the local design are compiled by **make vsim**. The compiled simulation models will be stored locally in a sub-directory (./modelsim). A modelsim.ini file will be created automatically, containing the necessary VHDL library mapping definitions. Running **make vsim** again will then use a vmake-generated makefile to check dependencies and rebuild out of date modules..

An other way to compile and simulate the library with modelsim is to use a modelsim project file. When doing **make scripts**, a modelsim project file is created. It is then possible to start vsim with this project file and perform compilation within vsim. In this case, vsim should be started with **make vsim-launch**. In the vsim window, click on the build-all icon to compile the complete library and the local design. The project file also includes one simulation configuration, which can be used to simulate the test bench (see figure below).

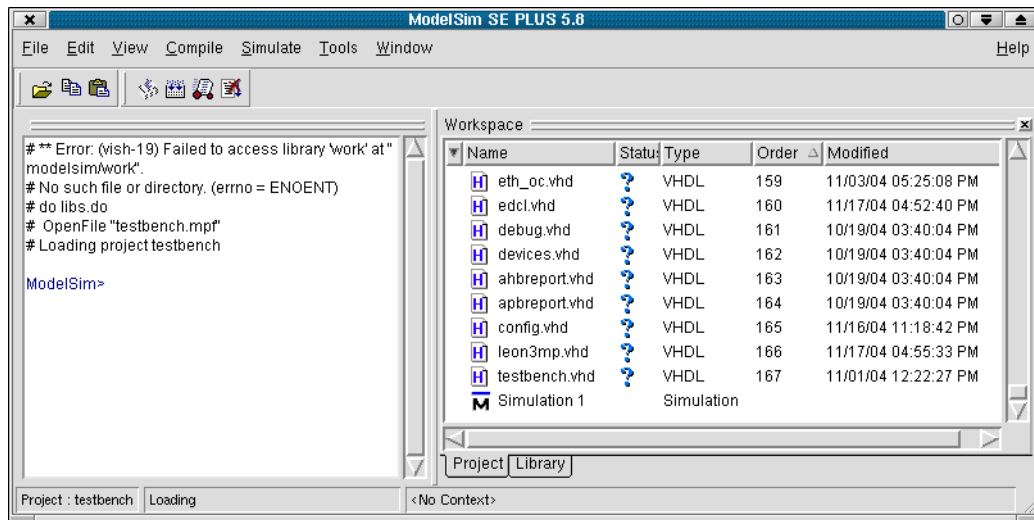


Figure 3. Modelsim simulator window using a project file

TABLE 9. Modelsim make targets

Make target	Description
vsim	Compile or re-analyze local design
vsim-clean	Remove compiled models and temporary files
vsim-launch	Start modelsim GUI on current test bench
vsim-fix	Run after <b>make vsim</b> to fix problems with make in CygWin
vsim-run	Run test bench in batchmode

TABLE 10. Modelsim scripts and files

File	Description
compile.vsim	Compile script for GRLIB files
make.work	Makefile to rebuild GRLIB and local design
modelsim	Directory with compiled models
SIMTOP.mpf	Modelsim project file for compilation and simulation

#### 4.7.4 Aldec Active-HDL

The Active-HDL tool from Aldec can be used in the standalone batch mode (`vsimsa.bat`) and in the GUI mode (`avhdl.exe`, or started from Windows icon/menu).

The batch mode does not support waveforms and is generally not directly transferable to the GUI mode. The batch mode uses ModelSim compatible command line names such as `vlib` and `vcom`. To use the batch mode, one must ensure that these commands are visible in the shell to be used. Note that the batch mode simulator requires a separate license from Active-HDL.

In batch mode, the completed GRLIB as well as the local design are compiled by `make vsimsa`. The compiled simulation models will be stored locally in a sub-directory (`./activehdl`). A `vsimsa.cfg` file will be created automatically, containing the necessary VHDL library mapping definitions. The simulation can then be started using the Active-HDL `vsimsa.bat` or `vsim` command. The simulation can also be started with `make vsimsa-run`.

Another way to compile and simulate the library is with the Active-HDL GUI using a `tcl` command file. When doing `make avhdl`, the `tcl` command file is automatically created for GRLIB and the local design files. The file can then be executed within Active-HDL with `do avhdl.tcl`, creating all necessary libraries and compiling all files. The compiled simulation models will be stored locally in a sub-directory (`./work`). Note that only the local design files are directly accessible from the design browser within Active-HDL. The compilation and simulation can also be started from the cygwin command line with `make avhdl-launch`.

Note that it is not possible to use both batch and GUI mode in the same design directory.

**TABLE 11. Active-HDL make targets**

Make target	Description
<code>vsimsa</code>	Compile GRLIB and local design
<code>vsimsa-clean</code>	Remove compiled models and temporary files
<code>vsim-run</code>	Run test bench in batch mode (must be compiled first)
<code>avhdl</code>	Setup GRLIB and local design
<code>avhdl-clean</code>	Remove compiled models and temporary files
<code>avhdl-launch</code>	Compile and Run test bench in GUI mode (must be setup first)

**TABLE 12. Active-HDL scripts and files**

File	Description
<code>compile.asim</code>	Compile script for GRLIB files (batch mode)
<code>make.asim</code>	Compile script for GRLIB files and local design (batch mode)
<code>activehdl</code>	Directory with compiled models (batch mode)
<code>work</code>	Directory with compiled models (GUI mode)
<code>avhdl.tcl</code>	Active-HDL <code>tcl</code> file for compilation and simulation (GUI mode)



### 4.7.5 Aldec Riviera

The Riviera tool from Aldec can be used in the standalone batch mode and in the GUI mode. The two modes are compatible, using the same compiled database.

In both modes, the completed GRLIB as well as the local design are compiled by **make riviera**. The compiled simulation models will be stored locally in a sub-directory (`./riviera`). A `vsimsa.cfg` file will be created automatically, containing the necessary VHDL library mapping definitions.

The standalone batch mode simulation can be started with **make riviera-run**. The GUI mode simulation can be started with **make riviera-launch**.

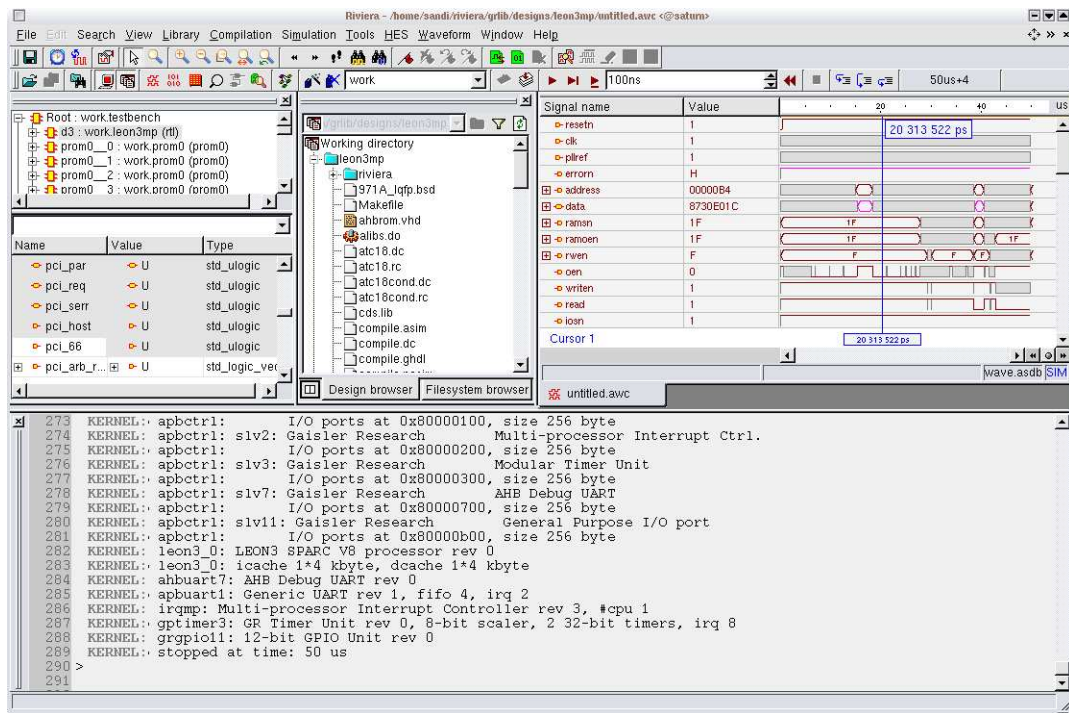


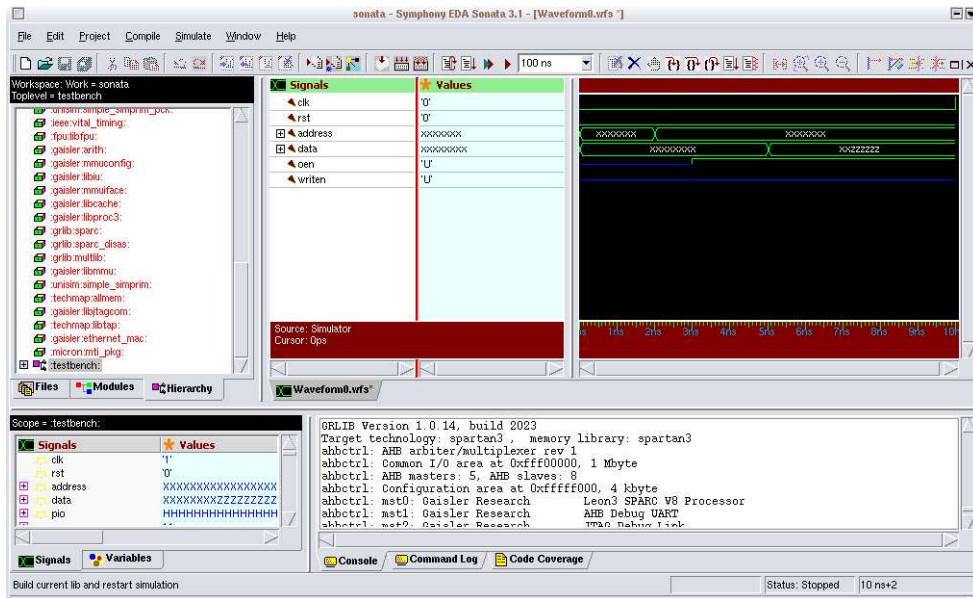
TABLE 13. Riviera make targets

Make target	Description
<code>riviera</code>	Compile GRLIB and local design
<code>riviera-clean</code>	Remove compiled models and temporary files
<code>riviera-run</code>	Run test bench in batch mode (must be compiled first)
<code>riviera-launch</code>	Run test bench in GUI mode (must be compiled first)

TABLE 14. Riviera scripts and files

File	Description
<code>riviera</code>	Directory with compiled models
<code>riviera.do</code>	Riviera script file for simulation (GUI mode)

The complete GRLIB as well as the local design are compiled by **make sonata**. The compiled simulation models will be stored locally in a sub-directory (`./sonata`). A `symphony.ini` file will be created automatically, containing the necessary VHDL library mapping definitions. To run the Sonata simulator in GUI mode, do **make sonata-launch** or start Sonata using the created `sonata.sws` project file. Sonata can also be run in batch mode, with **make sonata-run**. The VHDL work library will be mapped on library 'sonata', as 'work' is reserved and cannot be used.



Make target	Description
sonata	Compile GRLIB and local design
sonata-clean	Remove compiled models and temporary files
sonata-run	Compile GRLIB and run test bench in batch mode
sonata-launch	Compiler GRLib and run test bench in GUI mode

File	Description
sonata	Directory with compiled models
symphony.ini	Sonata library mapping for batch simulation
sonata.sws	Sonata project file for GUI version

### 4.7.7 Synthesis with Synplify

The **make scripts** command will create a `compile.synp` file which contains Synplify tcl commands for analyzing all GRLIB files and a synplify project file called `TOP_synplify.prj`, where TOP will be replaced with the name of the top level entity.

Synthesizing the design in batch mode using the generated project file can be done in one step using **make synplify**. All synthesis results will be stored locally in a sub-directory (`./synplify`). Running Synplify in batch requires that it supports the `-batch` option (Synplify Professional). If the installed Synplify version does not support `-batch`, first create the project file and then run Synplify interactively. By default, the synplify executable is called `'synplify_pro'`. This can be changed by supplying the SYNPLIFY variable to `'make'`:

```
make synplify SYNPLIFY=synplify_pro.exe
```

The synthesis script will set the following mapping option by default:

```
set_option -symbolic_fsm_compiler 0
set_option -resource_sharing 0
set_option -use_fsm_explorer 0
set_option -write_vhdl 1
set_option -disable_io_insertion 0
```

Additional options can be set through the SYNPOPT variable in the Makefile:

```
SYNPOPT="set_option -pipe 0; set_option -retiming 1"
```

Note that the Synplify tool does have some bugs, which can cause the generation of corrupt netlist for large designs. Currently, the most stable version seems to be 8.9.

**TABLE 17. Synplify make targets**

Make target	Description
synplify	Synthesize design in batch mode
synplify-clean	Remove compiled models and temporary files
synplify-launch	Start synplify interactively using generated project file

**TABLE 18. Synplify scripts and files**

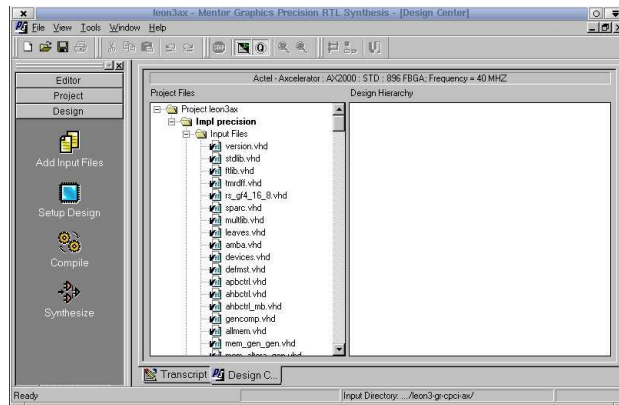
File	Description
compile.synp	Tcl compile script for all GRLIB files
TOP_synplify.prj	Synplify project file
synplify	Directory with netlist and log files

### 4.7.8 Synthesis with Mentor Precision

The **make scripts** command will create a `TOP_precision.tcl` file which contains tcl script to create a Precision project file. The project file (`TOP_precision.psp`) is created on the first invocation of Precision, but can also be created manually with **precision -shell -file TOP\_precision.tcl**.

Synthesizing the design in batch mode can be done in one step using **make precision**. All synthesis results will be stored locally in a sub-directory (`./precision`). Precision can also be run interactively by issuing **make precision-launch**. By default, the Precision executable is called with 'precision'. This can be changed by supplying the `PRECISION` variable to 'make':

```
make precision PRECISION=/usr/local/bin/precision
```



**TABLE 19. Precision make targets**

Make target	Description
precision	Synthesize design in batch mode
precision-clean	Remove compiled models and temporary files
precision-launch	Start Precision interactively using generated project file

**TABLE 20. Precision scripts and files**

File	Description
<code>TOP_precision.tcl</code>	Tcl compile script to create Precision project file
<code>TOP_precision.psp</code>	Precision project file
<code>precision</code>	Directory with netlist and log files

#### 4.7.9 Actel Designer

Actel Designer is used to place&route designs targeting Actel FPGAs. It does not include a synthesis engine, and the design must first be synthesized with synplify.

The `make scripts` command will generate a tcl script to perform place&route of the local design in batch mode. The tcl script is named `TOP_designer.tcl`, where `TOP` is replaced with the name of the top entity.

The command `make actel` will place&route the design using the created tcl script. The design database will be placed in `actel/TOP.adb`. The command `make actel-launch` will load the edif netlist of the current design, and start Designer in interactive mode.

GRLIB includes a leon3 design template for the GR-CPCI-AX board from Pender/Gaisler. The template design is located in `designs/leon3-gr-cpci-ax`. The local design file uses board settings from the `boards/gr-cpci-ax` directory. The leon3-gr-cpci-ax design can be used as a template for other AX-based projects.

GRLIB also includes a leon3 template design for the Actel CoreMP7 board (Proasic3-1000). It is located in `designs/leon3-actel-coremp7`.

A template design can specify the variable `DESIGNER_LAYOUT_OPT` to override the switches passed to the `layout` command.

**TABLE 21. Actel Designer make targets**

Make target	Description
actel	Place&route design in batch mode
actel-clean	Remove compiled models and temporary files
actel-launch	Start Designer interactively using synplify netlist

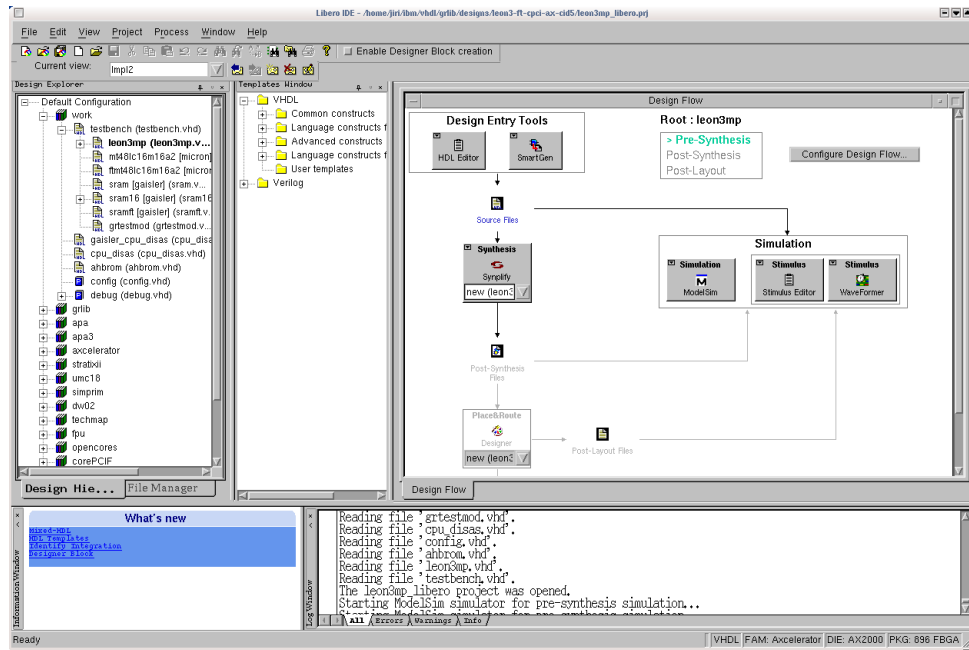
**TABLE 22. Actel Designer scripts and files**

File	Description
<code>TOP_designer.tcl</code>	Batch script for Actel Designer place&route

#### 4.7.10 Actel Libero

Actel Libero is an integrated design environment for implementing Actel FPGAs. It consists of Actel-specific versions of Synplify and Modelsim, together with the Actel Designer back-end tool.

Using Libero to implement GRLIB designs is possible using Libero-8.1 and later versions. The **make scripts** command will create a Libero project file called `TOP_libero.prj`. Libero can then be started with `libero TOP_libero.prj`, or by the command `make libero-launch`. Implementation of the design is done using the normal Libero flow.



Note that when synplify is launched from Libero the first time, the constraints file defined in the local Makefile is not included in the project, and must be added manually. Before simulation is started first time, the file `testbench.vhd` in the template design should be associated as stimulus file.

TABLE 23. Libero make targets

Make target	Description
<code>scripts</code>	Created libero project file
<code>libero-launch</code>	Create project file and launch libero

TABLE 24. Libero scripts and files

File	Description
<code>TOP_libero.prj</code>	Libero project file

#### 4.7.11 Altera Quartus

Altera Quartus is used for Altera FPGA targets, and can be used to both synthesize and place&route a design. It is also possible to first synthesize the design with synplify and then place&route with Quartus.

The **make scripts** command will generate two project files for Quartus, one for an EDIF flow where a netlist has been created with synplify and one for a Quartus-only flow. The project files are named TOP.qpf and TOP\_synplify.qpf, where TOP is replaced with the name of the top entity.

The command **make quartus** will synthesize and place&route the design using a quartus-only flow in batch mode. The command **make quartus-synp** will synthesize with synplify and run place&route with Quartus. Interactive operation is achieved through the command **make quartus-launch** (quartus-only flow), or **make quartus-launch-synp** (EDIF flow). Quartus can also be started manually with **quartus TOP.qpf** or **quartus TOP\_synplify.qpf**.

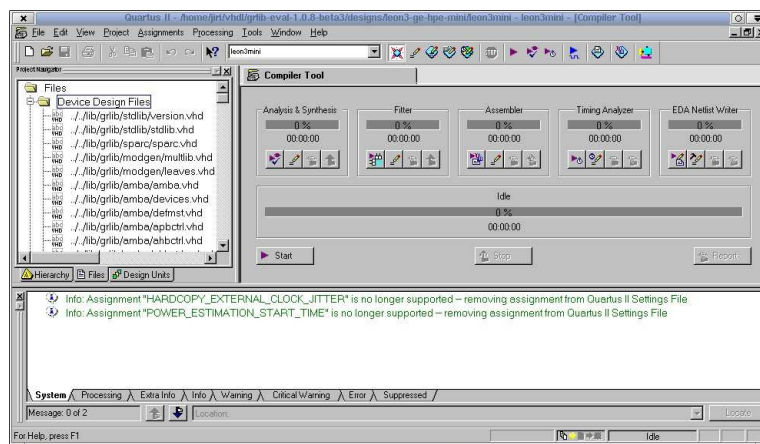


TABLE 25. Altera Quartus make targets

Make target	Description
quartus	Synthesize and place&route design with Quartus in batch mode
quartus-clean	Remove compiled models and temporary files
quartus-launch	Start Quartus interactively using Quartus-only flow
quartus-launch-synp	Start Quartus interactively using EDIF flow
quartus-map	Synthesize design with Quartus in batch mode
quartus-synp	Synthesize with synplify and place&route with Quartus in batch mode
quartus-prog-fpga	Program FPGA in batch mode

TABLE 26. Altera Quartus scripts and files

File	Description
TOP.qpf	Project file for Quartus-only flow
TOP_synplify.qpf	Project file for EDIF flow

#### 4.7.12 Xilinx ISE

Xilinx ISE is used for Xilinx FPGA targets, and can be used to simulate, synthesize and place&route a design. It is also possible to first synthesize the design with synplify and the place&route with ISE. It is generally recommended to use the latest version of ISE. Simulation of grlib template design using ISIM is supported as of ISE-13.2. The simulator is launched from the project navigator GUI.

The **make scripts** command will create an XML project file (TOP.xise), useful with ISE-11 and above. When executing **make ise-launch**, this XML will be used to launch the ISE project manager. Synthesis and place&route can also be run in batch mode (preferred option) using **make ise** for the XST flow and **make ise-synp** for synplify flow.

Many Xilinx FPGA boards are supported in GRLIB, and can be re-programmed using **make ise-prog-fpga** and **make ise-prog-prom**. The first command will only re-program the FPGA configuration, while the second command will reprogram the configuration prompts (if available). Programming will be done using the ISE Impact tool in batch mode.

When simulating designs that depends on Xilinx macro cells (RAM, PLL, pads), a built-in version of the Xilinx UNSIM simulation library will be used. The built-in library has reduced functionality, and only contains the cells used in grlib. The full Xilinx UNISIM library can be installed using **make install-unisim**. This will copy the UNISIM files from ISE into grlib. A **make distclean** must first be given before the libraries can be used. It is possible to revert to the built-in UNISIM libraries by issuing **make remove-unisim**. Note: to install the Xilinx UNISIM files, the variable **XILINX** must point to the installation path of ISE. The variable is normally set automatically during installation of ISE. To simulate designs using the Xilinx MIG memory controllers, the secureIP library must first be installed using **make install-secureip**.

**TABLE 27. Xilinx ISE make targets**

Make target	Description
ise	Synthesize and place&route design with XST in batch mode
ise-prec	Synthesize and place&route design with Precision in batch mode
ise-synp	Synthesize and place&route design with Synplify in batch mode
ise-launch	Start project navigator interactively using XST flow
ise-launch-synp	Start project navigator interactively using EDIF flow
ise-map	Synthesize design with XST in batch mode
ise-prog-fpga	Program FPGA on target board using JTAG
ise-prog-fpga-ref	Program FPGA on target board with reference bit file
ise-prog-prom	Program configuration prompts on target board using JTAG
ise-prog-prom-ref	Program configuration prompts with reference bit file
install-unisim	Install Xilinx UNISIM libraries into grlib
remove-unisim	Remove Xilinx UNISIM libraries from grlib
install-secureip	Install Xilinx SecureIP files into grlib
remove-secureIP	Remove Xilinx SecureIP files from grlib

**TABLE 28. Xilinx ISE scripts and files**

File	Description
compile.xst	XST synthesis include script for all GRLIB files
TOP.xst	XST synthesis script for local design
TOP.npl	ISE 8 project file for XST flow
TOP.ise	ISE 9/10project file for XST flow
TOP.xise	ISE 11/12/13 XML project file for XST flow
TOP_synplify.npl	ISE 8 project file for EDIF flow



**Old and deprecated ISE versions:**

The **make scripts** command also generates .npl project files for the ISE-8 project navigator, for both EDIF flow where a netlist has been created with synplify and for ISE/XST flow. The project navigator can be launched with **make ise-launch-synp** for the EDIF flow, and with **make ise-launch8** for the XST flow. The project navigator can also be started manually with **ise TOP.npl** or **ise TOP\_synplify.npl**. The .npl files are intended to be used with ISE 6 - 8.

For ISE-9 and ISE-10, an .ise file will be generated using xtclsh when **make ise-launch** is given, or by **make TOP.ise**. The Xilinx xtclsh operate very slowly, so generation of the .ise file will take 10 - 20 minutes (!).

#### 4.7.13 Xilinx PlanAhead

Xilinx PlanAhead is supported for Xilinx devices and prototype boards to improve runtime and performance. The GRLIB environment allows the user to experiment with different implementation options to improve design results via runtime option specified in `$(GRLIB)/boards/$(BOARD)/Makefile.inc`. The Xilinx PlanAhead flow should be seen as an extension of GRLIB Xilinx ISE flow.

The **make scripts** command will create compile scripts for the PlanAhead tool, useful with ISE-14 and above. When executing **make planAhead-launch**, the compile scripts will be used to launch the PlanAhead project manager. Synthesis and place&route can also be run in batch mode (preferred option) using **make planAhead**.

Many Xilinx FPGA boards are supported in GRLIB, and can be re-programmed using **make ise-prog-fpga** and **make ise-prog-prom**. The first command will only re-program the FPGA configuration, while the second command will reprogram the configuration prompts (if available). Programming will be done using the ISE Impact tool in batch mode.

**TABLE 29. Xilinx PlanAhead specific make targets**

Make target	Description
planAhead	Synthesize and place&route design with PlanAhead in batch mode
planAhead-launch	Start project navigator interactively using planAhead flow
planAhead-clean	Remove all planAhead generated project files

**TABLE 30. Xilinx PlanAhead scripts and files**

File	Description
compile.planAhead	PlanAhead synthesis include script for all GRLIB files
planAhead.tcl	PlanAhead script for creating a PlanAhead project and to build the project.

#### 4.7.14 Xilinx Vivado

Xilinx Vivado is the build flow for Xilinx 7 series devices and prototype boards. The GRLIB environment allows the user to experiment with different implementation options to improve design results via runtime option specified in `$(GRLIB)/boards/$(BOARD)/Makefile.inc`.

The **make scripts** command will create compile scripts for the Vivado tool, useful with ISE-14.2 and above. When executing **make vivado-launch**, the compile scripts will be used to launch the Vivado project manager. Synthesis and place&route can also be run in batch mode (preferred option) using **make vivado**.

Many Xilinx FPGA boards are supported in GRLIB, and can be re-programmed using **make ise-prog-fpga** and **make ise-prog-prom**. The first command will only re-program the FPGA configuration, while the second command will reprogram the configuration prompts (if available). Programming will be done using the ISE Impact tool in batch mode.

**TABLE 31. Xilinx Vivado specific make targets**

Make target	Description
vivado	Synthesize and place&route design with Vivado in batch mode
vivado-launch	Start project navigator interactively using Vivado flow
vivado-clean	Remove all Vivado generated project files
vivado-prog-fpga	Optional program target for faster programming of the FPGA Device. This target needs Xilinx EDK/SDK to be installed.

**TABLE 32. Xilinx Vivado scripts and files**

File	Description
compile.vivado	Vivado synthesis include script for all GRLIB files
vivado.tcl	Vivado script for creating a PlanAhead project and to build the project.

#### 4.7.15 Lattice ISP Tools

Implementing GRLIB design on Lattice FPGAs is supported with Synplify for synthesis and the Lattice ISP Lever for place&route. The **make isp-synp** command will automatically synthesize and place&route a Lattice design. The associated place&route script is provided in `bin/route_lattice`, and can be modified if necessary. Supported FPGA families are EC and ECP. The template design `leon3-hpe-mini-lattice` is a Lattice ECP design which can be used to test the implementation flow. On linux, it might be necessary to source the ISP setup script in order to set up necessary paths:

```
source $ISPLEVER_PATH/ispcpd/bin/setup_lv.sh
```

**TABLE 33. Lattice ISP make targets**

Make target	Description
isp-synp	Synthesize and place&route design with Sunplify in batch mode
isp-clean	Remove compiled models and temporary files
isp-prom	Create FPGA prom

#### 4.7.16 Synthesis with Synopsys Design Compiler

The **make scripts** command will create a `compile.dc` file which contains Design Compiler commands for analyzing all GRLIB files. The `compile.dc` file can be run manually using `'dc_shell -f compile.dc'`. A script for the local design is created automatically and called *TOP\_dc.tcl* where *TOP* is the top entity name:

```
$ cat leon4mp_dc.tcl
sh mkdir synopsys
set objects synopsys
#set trans_dc_max_depth 1
#set hdlin_seqmap_sync_search_depth 1
#set hdlin_nba_rewrite false
set hdlin_ff_always_sync_set_reset true
set hdlin_ff_always_async_set_reset false
#set hdlin_infer_complex_set_reset true
#set hdlin_translate_off_skip_text true
set suppress_errors VHDL-2285
#set hdlin_use_carry_in true
source compile.dc
analyze -f VHDL -library work config.vhd
analyze -f VHDL -library work ahbrom.vhd
analyze -f VHDL -library work clkgate.vhd
analyze -f VHDL -library work qmod.vhd
analyze -f VHDL -library work qmod_prect.vhd
analyze -f VHDL -library work leon4mp.vhd
elaborate leon4mp
```

The script can be run with `dc_shell-xg-t` via the command **make dc**. The created script will analyze and elaborate the local design. Compilation and mapping will not be performed, the script should be seen as a template only.

#### 4.7.17 Synthesis with Cadence RTL Compiler

The **make scripts** command will create a `compile.rc` file which contains RTL Compiler commands for analyzing all GRLIB files. The `compile.rc` file can be run manually using `rc -files compile.rc` or through **make rc**. A script to analyze and synthesize the local design is created automatically and called *TOP.rc* where *TOP* is the top entity name:

```
$ cat netcard.rc
set_attribute input_pragma_keyword "cadence synopsys g2c fast ambit pragma"
include compile.rc
read_hdl -vhdl -lib work netcard.vhd
elaborate netcard
write_hdl -generic > netcard_gen.v
```

The created script will analyze and elaborate the local design, and save it to a Verilog file. Compilation and mapping will not be performed, the script should be seen as a template only.

#### 4.7.18 eASIC eTools

GRLIB support for eTools with eASIC Nextreme technology was discontinued in GRLIB version 1.1.0-b4109.

Support for the Nextreme2 technology and eTools 9 can be requested from Aeroflex Gaisler but is not included in any of the default GRLIB distributions. To work with eTools 9 the environment variable ETOOLS\_N2X\_HOME must be set to the eTools installation directory.

**TABLE 34. eASIC Nextreme2 make targets**

Make target	Description
import-easic-n2x	Imports eASIC RTL and IP libraries from eTools into GRLIB. Requires that the environment variable.
remove-easic-n2x	Removes eASIC RTL and IP libraries from GRLIB.
etools-n2x-init	Creates a eTools project file. Makes use of the environment variables TOP, DEVICE, PACKAGE, PNC, SDCFILE, and GRLIB_NHCPU. The last variable defines the number of available host CPUs.
etools-n2x-launch	Launch eTools DesignNavigator for the current project
etools-n2x-launch-no_iu	LauncheTools DesignNavigator for the current project in CLI mode.

The GRLIB technology map for eASIC Nextreme2 makes extensive use of eASIC's RAM and pad generators, and also of wrappers for the DDR2 PHY. When eASIC's IP library has been imported into GRLIB (via the *import-easic-n2x* make target), the normal technology map components (pads, memory, DDR2 PHY) can be used.

The GRLIB SYNCRAM\* components map to both rFiles and bRAMs. The conditions for selecting between these RAM types may need to be adjusted for each design in order to not over-utilize one or the other. The selection between rFiles and bRAMs is made with the function *n2x\_use\_rfile(..)* that is defined in the file *lib/techmap/nextreme2/memory\_n2x\_package.vhd*.

The technology map also includes a clock generator map for eASIC PLLs. However it is strongly recommended to use eASIC's IP generators instead and directly instantiate the Nextreme2 PLLs in the design.

## 4.8 XGrlib graphical implementation tool

### 4.8.1 Introduction

XGrlib serves as a graphical front-end to the makefile system described in the previous chapters. It is written in tcl/tk, using the Visual-tcl (vtcl) GUI builder. XGrlib allows to select which CAD tools will be used to implement the current design, and how to run them. XGrlib should be started in a directory with a GRLIB design, using `make xgrlib`. Other make variables can also be set on the command line, as described earlier:

```
make xgrlib SYNPLIFY=synplify_pro GRLIB="../../"
```

Since XGrlib uses the make utility, it is necessary that all used tools are in the execution path of the used shell. The tools are divided into three categories: simulation, synthesis and place&route. All tools can be run in batch mode with the output directed to the XGrlib console, or launched interactively through each tool's specific GUI. Below is a figure of the XGrlib main window:

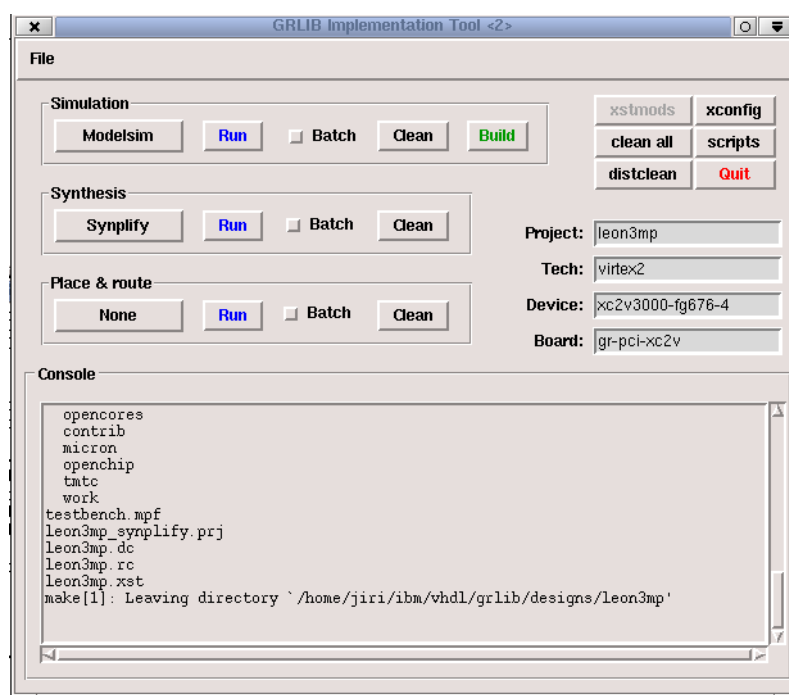


Figure 4. XGrlib main window

### 4.8.2 Simulation

The simulator type can be selected through the left menu button in the frame marked 'Simulation'. There are seven options available: modelsim, ncsim, GHDL, libero, riviera, active-hdl, and active-hdl batch. Once the simulator has been selected, the design can be compiled by pressing the green 'Build' button. The simulator can then be launched interactively by pressing the 'Run' button. If the 'Batch' check-button has been set, the 'Run' button will run the default test bench in batch mode with the output displayed in the console frame. The 'Clean' button will remove all generated file for the selected tool.

Note: on windows/cygwin platforms, launching modelsim interactively can fail due to conflict of cygwin and modelsim tcl/tk libraries. We are trying to resolve this issue.

### 4.8.3 Synthesis

The synthesis tool is selected through the menu button in the frame labeled with 'Synthesis'. There are five possibilities: Synplify, Altera Quartus, Xilinx ISE/XST, Mentor Precision and Actel Libero. The 'Batch' check-button defines if synthesis will be run in batch mode or if the selected tool will be launched interactively. The selected tool is started through the 'Run' button.

If a tool is started interactively, it automatically loads a tool-specific project file for the current design. It is then possible to modify the settings for the project before synthesis is started. Only one tool should be started at a time to avoid I/O conflicts. The 'Clean' button in the 'Synthesis' frame will remove all generated file for the selected synthesis tool.

Note that the Libero tool actually performs both simulation, synthesis and place&route. It has been added to the 'Synthesis' menu for convenience.

#### **4.8.4 Place & Route**

Place & route is supported for three FPGA tool-chains: Actel Designer, Altera Quartus and Xilinx ISE. Selecting the tool-chain is done through the menu button in the frame labeled 'Place & Route'. Again, the 'Batch' check-button controls if the tool-chain will be launched interactively or run in batch mode. Note that the selection of synthesis tool affects on how place&route is performed. For instance: if synplify has been selected for synthesis and the Xilinx ISE tool is launched, it will use a project file where the edif netlist from synplify is referenced. If the XST synthesis tool has been selected instead, the .ngc netlist from XST would have been used.

The 'Clean' button in the 'Place&Route' frame will remove all generated file for the selected place&route tool.

#### **4.8.5 Additional functions**

##### **Cleaning**

The 'Clean' button in each of the three tool frames will remove all generated files for selected tool. This make it possible to for instance clean and rebuild a simulation model without simultaneously removing a generated netlist. Generated files for all tools will be removed when the 'clean all' button is pressed. This will however not removed compile scripts and project files. To remove these as well, use the 'distclean' button.

##### **Generating compile scripts**

The compile scripts and project files are normally automatically generated by the make utility when needed by a tool. They can also be created directly through the 'scripts' button.

##### **Xconfig**

If the local design is configured through xconfig (leon3 systems), the xconfig tool can be launched by pressing the 'xconfig' button. The configuration file (config.vhd) is automatically generated if xconfig is exited by saving the new configuration.

##### **FPGA prom programming**

The button 'PROM prog' will generate FPGA prom files for the current board, and program the configuration proms using JTAG. This is currently only supported on Xilinx-based boards. The configuration prom must be reloaded by the FPGA for the new configuration to take effect. Some boards has a special reload button, while others must be power-cycled.



## 5 GRLIB Design concept

### 5.1 Introduction

GRLIB is a collection of reusable IP cores, divided on multiple VHDL libraries. Each library provides components from a particular vendor, or a specific set of shared functions or interfaces. Data structures and component declarations to be used in a GRLIB-based design are exported through library specific VHDL packages.

GRLIB is based on the AMBA AHB and APB on-chip buses, which is used as the standard interconnect interface. The implementation of the AHB/APB buses is compliant with the AMBA-2.0 specification, with additional ‘sideband’ signals for automatic address decoding, interrupt steering and device identification (a.k.a. plug&play support). The AHB and APB signals are grouped according to functionality into VHDL records, declared in the GRLIB VHDL library. The GRLIB AMBA package source files are located in lib/grlib/amba.

All GRLIB cores use the same data structures to declare the AMBA interfaces, and can then easily be connected together. An AHB bus controller and an AHB/APB bridge are also available in the GRLIB library, and allows to assemble quickly a full AHB/APB system.

The following sections will describe how the AMBA buses are implemented and how to develop a SOC design using GRLIB.

### 5.2 AMBA AHB on-chip bus

#### 5.2.1 General

The AMBA Advanced High-performance Bus (AHB) is a multi-master bus suitable to interconnect units that are capable of high data rates, and/or variable latency. A conceptual view is provided in figure 5. The attached units are divided into master and slaves, and controlled by a global bus arbiter.

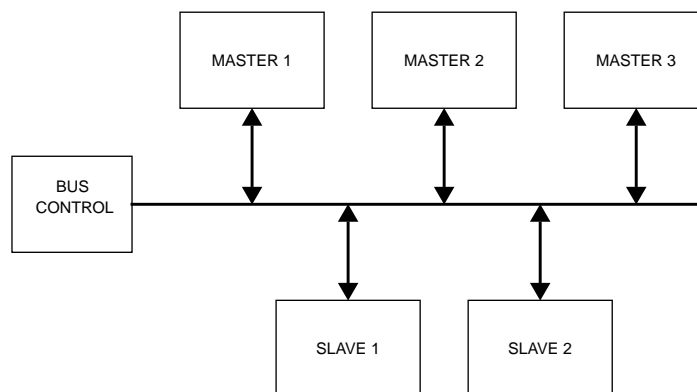


Figure 5. AMBA AHB conceptual view

Since the AHB bus is multiplexed (no tristate signals), a more correct view of the bus and the attached units can be seen in figure 6. Each master drives a set of signals grouped into a VHDL record called `ahbmo`. The output record of the current bus master is selected by the bus multiplexers and sent to the input record (`ahbsi`) of all AHB slaves. The output record (`ahbso`) of the active slave is selected by the bus multiplexer and forwarded to all masters. A combined bus arbiter, address decoder and bus multiplexer controls which master and slave are currently selected.

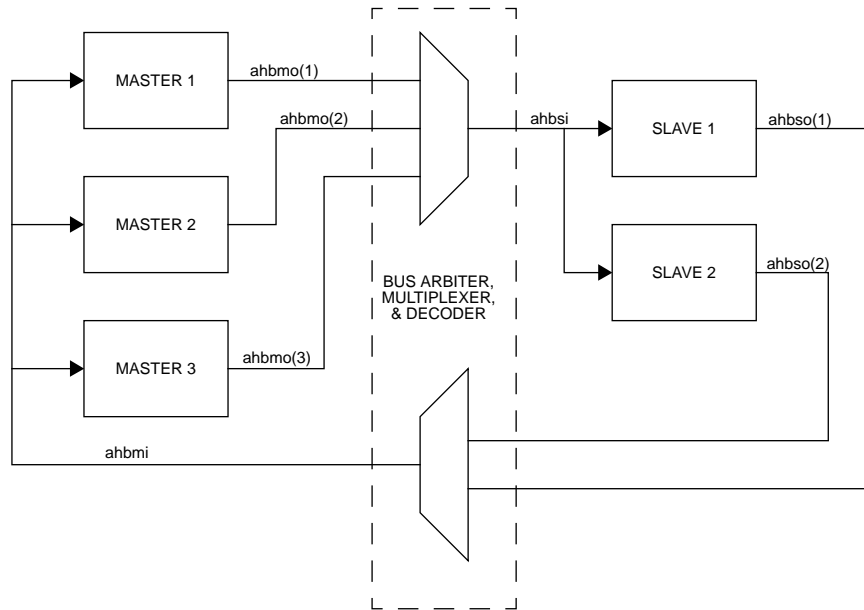


Figure 6. AHB inter-connection view

### 5.2.2 AHB master interface

The AHB master inputs and outputs are defined as VHDL record types, and are exported through the AMBA package in the GRLIB library:

```
-- AHB master inputs
type ahb_mst_in_type is record
  hgrant  : std_logic_vector(0 to NAHBMST-1);    -- bus grant
  hready  : std_ulogic;                          -- transfer done
  hresp   : std_logic_vector(1 downto 0);        -- response type
  hrdata  : std_logic_vector(31 downto 0);       -- read data bus
  hirq    : std_logic_vector(NAHBIRQ-1 downto 0); -- interrupt result bus
end record;

-- AHB master outputs
type ahb_mst_out_type is record
  hbusreq : std_ulogic;                          -- bus request
  hlock   : std_ulogic;                          -- lock request
  htrans  : std_logic_vector(1 downto 0);        -- transfer type
  haddr   : std_logic_vector(31 downto 0);       -- address bus (byte)
  hwrite  : std_ulogic;                          -- read/write
  hsize   : std_logic_vector(2 downto 0);        -- transfer size
  hburst  : std_logic_vector(2 downto 0);        -- burst type
  hprot   : std_logic_vector(3 downto 0);        -- protection control
  hwdata  : std_logic_vector(31 downto 0);       -- write data bus
  hirq    : std_logic_vector(NAHBIRQ-1 downto 0); -- interrupt bus
  hconfig : ahb_config_type;                    -- memory access reg.
  hindex  : integer range 0 to NAHBMST-1;       -- diagnostic use only
end record;
```

The elements in the record types correspond to the AHB master signals as defined in the AMBA 2.0 specification, with the addition of three sideband signals: HIRQ, HCONFIG and HINDEX. A typical AHB master in GRLIB has the following definition:

```
library grlib;
use grlib.amba.all;
library ieee;
use ieee.std_logic.all;

entity ahbmaster is
```

```

generic (
  hindex : integer := 0);          -- master bus index
port (
  reset   : in  std_ulogic;
  clk     : in  std_ulogic;
  ahbmi   : in  ahb_mst_in_type;   -- AHB master inputs
  ahbmo   : out ahb_mst_out_type  -- AHB master outputs
);
end entity;

```

The input record (AHBMI) is routed to all masters, and includes the bus grant signals for all masters in the vector AHBMI.HGRANT. An AHB master must therefore use a generic that specifies which HGRANT element to use. This generic is of type integer, and typically called HINDEX (see example above).

### 5.2.3 AHB slave interface

Similar to the AHB master interface, the inputs and outputs of AHB slaves are defined as two VHDL records types:

```

-- AHB slave inputs
type ahb_slv_in_type is record
  hsel       : std_logic_vector(0 to NAHBSLV-1);  -- slave select
  haddr      : std_logic_vector(31 downto 0);    -- address bus (byte)
  hwrite     : std_ulogic;                       -- read/write
  htrans     : std_logic_vector(1 downto 0);     -- transfer type
  hsize      : std_logic_vector(2 downto 0);     -- transfer size
  hburst     : std_logic_vector(2 downto 0);     -- burst type
  hwdata     : std_logic_vector(31 downto 0);    -- write data bus
  hprot      : std_logic_vector(3 downto 0);     -- protection control
  hready     : std_ulogic;                       -- transfer done
  hmaster    : std_logic_vector(3 downto 0);     -- current master
  hmastlock  : std_ulogic;                       -- locked access
  hbssel     : std_logic_vector(0 to NAHBCFG-1); -- bank select
  hirq       : std_logic_vector(NAHBIRQ-1 downto 0); -- interrupt result bus
end record;

-- AHB slave outputs
type ahb_slv_out_type is record
  hready     : std_ulogic;                       -- transfer done
  hresp      : std_logic_vector(1 downto 0);     -- response type
  hrdata     : std_logic_vector(31 downto 0);    -- read data bus
  hsplit     : std_logic_vector(15 downto 0);    -- split completion
  hirq       : std_logic_vector(NAHBIRQ-1 downto 0); -- interrupt bus
  hconfig    : ahb_config_type;                 -- memory access reg.
  hindex     : integer range 0 to NAHBSLV-1;    -- diagnostic use only
end record;

```

The elements in the record types correspond to the AHB slaves signals as defined in the AMBA 2.0 specification, with the addition of four sideband signals: HSEL, HIRQ, HCONFIG and HINDEX. A typical AHB slave in GRLIB has the following definition:

```

library grlib;
use grlib.amba.all;
library ieee;
use ieee.std_logic.all;

entity ahbslave is
  generic (
    hindex : integer := 0);          -- slave bus index
  port (
    reset   : in  std_ulogic;
    clk     : in  std_ulogic;
    abhsi   : in  ahb_slv_in_type;   -- AHB slave inputs
    abhso   : out ahb_slv_out_type  -- AHB slave outputs
  );
end entity;

```

The input record (abhssi) is routed to all slaves, and include the select signals for all slaves in the vector abhssi.hsel. An AHB slave must therefore use a generic that specifies which hsel element to use. This generic is of type integer, and typically called HINDEX (see example above).

### 5.2.4 AHB bus control

GRLIB AMBA package provides a combined AHB bus arbiter (AHBCTRL), address decoder and bus multiplexer. It receives the ahbmo and ahbso records from the AHB units, and generates ahbmi and abhsi as indicated in figure 6. The bus arbitration function will generate which of the ahbmi.hgrant elements will be driven to indicate the next bus master. The address decoding function

will drive one of the `ahbsi.hsel` elements to indicate the selected slave. The bus multiplexer function will select which master will drive the `ahbsi` signal, and which slave will drive the `ahbmo` signal.

### 5.2.5 AHB bus index control

The AHB master and slave output records contain the sideband signal `HINDEX`. This signal is used to verify that the master or slave is driving the correct element of the `ahbso/ahbmo` buses. The generic `HINDEX` that is used to select the appropriate `hgrant` and `hsel` is driven back on `ahbmo.hindex` and `ahbso.hindex`. The AHB controller then checks that the value of the received `HINDEX` is equal to the bus index. An error is issued during simulation if a mismatch is detected.

### 5.2.6 Support for wide AHB data buses

#### 5.2.6.1 Overview

The cores in GRLIB and the GRLIB infrastructure can be configured to support an AMBA AHB data bus width of 32, 64, 128, or 256 bits. The default AHB bus width is 32 bits and AHB buses with data vectors having widths over 32 bits will in this section be referred to as wide AHB buses.

Changing the AHB bus width can increase performance, but may also increase the area requirements of a design, depending on the synthesis tool used and the type of cores instantiated. Manual modification of the GRLIB CONFIG package is required to enable support for wide AHB buses. Alternatively, a local version of the GRLIB CONFIG package can be placed in the current template design, overriding the settings in the global GRLIB CONFIG package.

When modifying the system's bus width, care should be taken to verify that all cores have been instantiated with the correct options with regards to support for wide buses.

Note that the APB bus in GRLIB will always be 32-bits, regardless of the AHB data bus width.

#### 5.2.6.2 Implementation of support for wide AHB buses

To support wide buses, the AHB VHDL records that specify the GRLIB AMBA AHB interface have their data vector lengths defined by a constant, `CFG_AHBDW`, defined in the GRLIB CONFIG VHDL package.

Using a wide AHB bus places additional requirements on the cores in a design; The cores should drive the extra positions in the AHB data vector in order to minimize the amount of undriven signals in the design, and to allow synthesis tool optimisations for cores that do not support AMBA accesses larger than word accesses. The cores are also required to select and drive the applicable byte lanes, depending on access size and address.

In order to minimize the amount of undriven signals, all GRLIB AHB cores drive their AHB data vector outputs via a subprogram, `ahbdrivedata(..)`, defined in the GRLIB AMBA VHDL package. The subprogram replicates its input so that the whole AHB data vector is driven. Since data is present on all byte lanes, the use of this function also ensures that data will be present on the correct byte lanes.

The AMBA 2.0 Specification requires that cores select their data from the correct byte lane. For instance, when performing a 32-bit access in a system with a 64-bit wide bus, valid data will be on positions 63:32 of the data bus if bit 2 of the address is 0, otherwise the valid data will be on positions 31:0. In order to ease adding support for variable buses, the GRLIB AMBA VHDL package includes subprograms, `ahbread*(...)`, for reading the AMBA AHB data vectors, hereafter referred to as AHB read subprograms. These subprograms exist in two variants; The first variant takes an address argument so that the subprogram is able to select the valid byte lanes of the data vector. This functionality is not always enabled, as will be explained below. The second variant does not require the address argument, and always returns the low slice of the AHB data vector.

Currently the majority of the GRLIB AHB cores use the functions without the address argument, and therefore the cores are only able to read the low part of the data vector. The cores that only read the low part of the AHB data vector are not fully AMBA 2.0 compatible with regard to wide buses. However, this does not affect the use of a wide AHB bus in a GRLIB system, since all GRLIB cores place valid data on the full AHB data vector. As adoption of wide buses become more widespread, the cores will be updated so that they are able to select the correct byte lanes.

The GRLIB AHB controller core, `AHBCTRL`, is a central piece of the bus infrastructure. The AHB controller includes a multiplexer of the width defined by the AMBA VHDL package constant `AHBDW`. The core also has a generic that decides if the controller should perform additional AMBA data multiplexing. Data multiplexing is discussed in the next section.

### 5.2.6.3 AMBA AHB data multiplexing

Almost all GRLIB cores drive valid data on all lanes of the data bus, some exceptions exist, such as the cores in the AMBA Test Framework). Since the *ahbdrivedata(..)* subprogram duplicates all data onto the wider bus, all cores will be compliant to the AMBA standard with regards to placing valid data on the correct lane in the AHB data vector.

As long as there are only GRLIB cores in a design, the cores can support wide AHB buses by only reading the low slice of the AHB data vectors, which is the case for most cores, as explained in the section above. However, if a core that only drives the required part of the data vector is introduced in a design there is a need for support to allow the GRLIB cores to select the valid part of the data.

The current implementation has two ways of accomplishing this:

Set the ACDM generic of AHBCTRL to 1. When this option is enabled the AHB controller will check the size and address of each access and propagate the valid part of the data on the entire AHB data bus. The smallest portion of the slice to select and duplicate is 32-bits. This means that valid data for a a byte or halfword access will not be present on all byte lanes, however the data will be present on all the required byte lanes.

Set the CFG\_AHB\_ACDM constant to 1 in the GRLIB CONFIG VHDL package. This will make the AHB read subprograms look at the address and select the correct slice of the incoming data vector. If a core uses one of the AHB read subprograms that does not have the address argument there will be a failure asserted. If CFG\_AHB\_ACDM is 0, the AHB read subprograms will return the low slice of the data vector. With CFG\_AHB\_ACDM set to 1, a core that uses the subprograms with the correct address argument will be fully AMBA compliant and can be used in non-GRLIB environments with bus widths exceeding 32 bits.

Note that it is unnecessary to enable both of these options in the same system.

### 5.2.6.4 Modified cores

Several cores in the IP library make use of the wide buses, see the core documentation in the GRLIB IP Cores User's Manual to determine the state of wide bus support for specific cores. All cores in GRLIB can be used in a system with wide AHB buses, however they do not all exploit the advantages of a wider bus.

### 5.2.6.5 GRLIB CONFIG Package

The location of the global GRLIB CONFIG package is in lib/grlib/stdlib/config.vhd. This file contains the settings for the wide buses as described above, and some additional global parameters. This package can be replaced by a local version by setting the variable GRLIB\_CONFIG in the Makefile of a template design to the location of an alternative version. When the simulation and synthesis scripts are built, the alternative CONFIG package will be used instead of the global one. The the variable GRLIB\_CONFIG is modified, the scripts have to be re-built for the new value to take effect.

### 5.2.6.6 Issues with wide AHB buses

A memory controller may not be able to respond all access sizes. With the current scheme the user of the system must keep track of which areas that can be accessed with accesses larger then word accesses. For instance, if SVGACTRL is configured to use 4WORD accesses and the designs has a DDR2SPA core and a MCTRL core in the system, the SVGACTRL will only receive correct data if the framebuffer is placed in the DDR2 memory area.

Special care must be taken when using wide buses so that the core specific settings for wider buses matches the intended use for the cores. Most cores are implemented so that they include support for handling access sizes up to AHBDW.

## 5.3 AHB plug&play configuration

### 5.3.1 General

The GRLIB implementation of the AHB bus includes a mechanism to provide plug&play support. The plug&play support consists of three parts: identification of attached units (masters and slaves), address mapping of slaves, and interrupt routing. The plug&play information for each AHB unit consists of a configuration record containing eight 32-bit words. The first word is called the identification register and contains information on the device type and interrupt routing. The last four words are called bank address registers, and contain address mapping information for AHB slaves. The remaining three words are currently not assigned and could be used to provide core-specific configuration information.

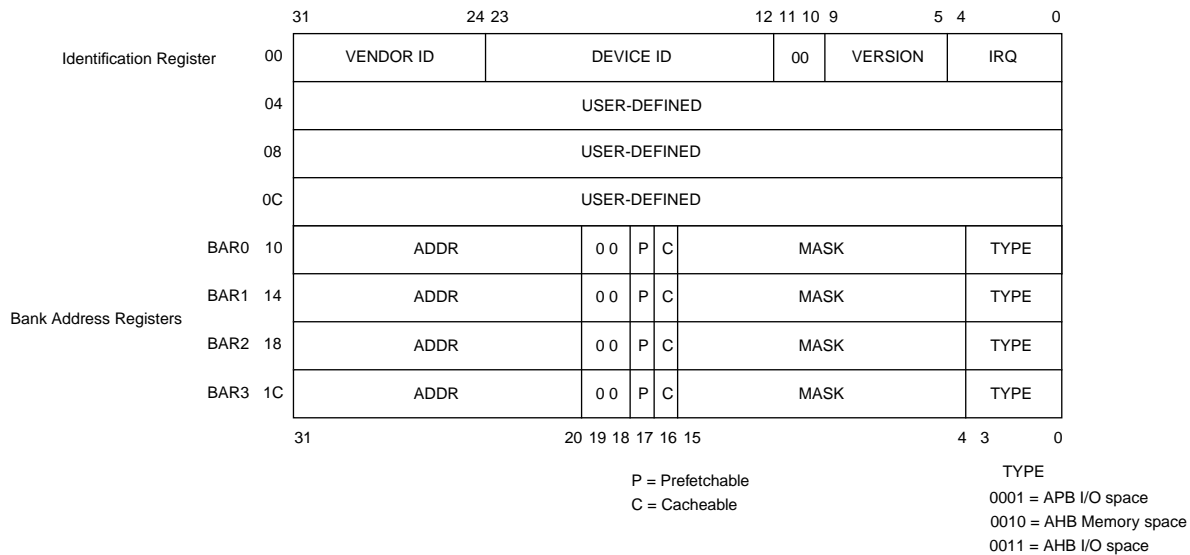


Figure 7. AHB plug&play configuration layout

The plug&play information for all attached AHB units appear as a read-only table mapped on a fixed address of the AHB, typically at 0xFFFFF000. The configuration records of the AHB masters appear in 0xFFFFF000 - 0xFFFFF800, while the configuration records for the slaves appear in 0xFFFFF800 - 0xFFFFF8FC. Since each record is 8 words (32 bytes), the table has space for 64 masters and 64 slaves. A plug&play operating system (or any other application) can scan the configuration table and automatically detect which units are present on the AHB bus, how they are configured, and where they are located (slaves).

The top four words of the plug&play area (0xFFFFFFF0 - 0xFFFFFFF7) may contain device specific information such as GRLIB build ID and a (SoC) device ID. If present, this information shadows the bank address registers of the last slave record, limiting the number of slaves on one bus to 63. All systems that use the GRLIB AHB controller have the library's build ID in the most significant half-word, and a (SoC) device ID in the least significant half-word, of the word at address 0xFFFFFFF0. The contents of the top four words is described in the AHB controller's IP core manual.

The configuration record from each AHB unit is sent to the AHB bus controller via the HCONFIG signal. The bus controller creates the configuration table automatically, and creates a read-only memory area at the desired address (default 0xFFFFF000). Since the configuration information is fixed, it can be efficiently implemented as a small ROM or with relatively few gates. A debug module (ahbreport) in the WORK.DEBUG package can be used to print the configuration table to the console during simulation, which is useful for debugging. A typical example is provided below:

```

VSIM 1> run
.
.
# LEON3 Actel PROASIC3-1000 Demonstration design
# GRLIB Version 1.0.16, build 2460
# Target technology: proasic3 , memory library: proasic3
# ahbctrl: AHB arbiter/multiplexer rev 1
# ahbctrl: Common I/O area disabled
# ahbctrl: AHB masters: 2, AHB slaves: 8
# ahbctrl: Configuration area at 0xfffff000, 4 kbyte
# ahbctrl: mst0: Gaisler Research      Leon3 SPARC V8 Processor
# ahbctrl: mst1: Gaisler Research      AHB Debug UART
# ahbctrl: slv0: European Space Agency Leon2 Memory Controller
# ahbctrl:      memory at 0x00000000, size 512 Mbyte, cacheable, prefetch
# ahbctrl:      memory at 0x20000000, size 512 Mbyte
# ahbctrl:      memory at 0x40000000, size 1024 Mbyte, cacheable, prefetch
# ahbctrl: slv1: Gaisler Research      AHB/APB Bridge
# ahbctrl:      memory at 0x80000000, size 1 Mbyte
# ahbctrl: slv2: Gaisler Research      Leon3 Debug Support Unit
# ahbctrl:      memory at 0x90000000, size 256 Mbyte
# apbctrl: APB Bridge at 0x80000000 rev 1
# apbctrl: slv0: European Space Agency Leon2 Memory Controller
# apbctrl:      I/O ports at 0x80000000, size 256 byte
# apbctrl: slv1: Gaisler Research      Generic UART
# apbctrl:      I/O ports at 0x80000100, size 256 byte
# apbctrl: slv2: Gaisler Research      Multi-processor Interrupt Ctrl.
# apbctrl:      I/O ports at 0x80000200, size 256 byte
# apbctrl: slv3: Gaisler Research      Modular Timer Unit
# apbctrl:      I/O ports at 0x80000300, size 256 byte
# apbctrl: slv7: Gaisler Research      AHB Debug UART
# apbctrl:      I/O ports at 0x80000700, size 256 byte
# apbctrl: slv11: Gaisler Research     General Purpose I/O port
# apbctrl:      I/O ports at 0x80000b00, size 256 byte
# grgpio11: 8-bit GPIO Unit rev 0
# gptimer3: GR Timer Unit rev 0, 8-bit scaler, 2 32-bit timers, irq 8
# irqmp: Multi-processor Interrupt Controller rev 3, #cpu 1
# apbuart1: Generic UART rev 1, fifo 1, irq 2
# ahbuart7: AHB Debug UART rev 0
# dsu3_2: LEON3 Debug support unit + AHB Trace Buffer, 1 kbytes
# leon3_0: LEON3 SPARC V8 processor rev 0
# leon3_0: icache 1*2 kbyte, dcache 1*2 kbyte

```

### 5.3.2 Device identification

The Identification Register contains three fields to identify uniquely an attached AHB unit: the vendor ID, the device ID, and the version number. The vendor ID is a unique number assigned to an IP vendor or organization. The device ID is a unique number assigned by a vendor to a specific IP core. The device ID is not related to the core's functionality. The version number can be used to identify (functionally) different versions of the unit.

The vendor IDs are declared in a package in each vendor library, usually called DEVICES. Vendor IDs are provided by Aeroflex Gaisler. The following ID's are currently assigned:

Vendor	ID
Gaisler Research	0x01
Pender Electronic Design	0x02
European Space Agency	0x04
Astrium EADS	0x06
OpenChip.org	0x07
OpenCores.org	0x08
DLR	0x0A
Eonic BV	0x0B
Radionor	0x0F
Gleichmann Electronics	0x10
Menta	0x11

TABLE 35. Vendor ID assignment

Vendor	ID
Sun Microsystems	0x13
Movidia	0x14
Orbita	0x17
Siemens AG	0x1A
Actel Corporation	0xAC
Caltech	0xCA
Embeddit	0xEA

TABLE 35. Vendor ID assignment

Vendor ID 0x00 is reserved to indicate that no core is present. Unused slots in the configuration table will have Identification Register set to 0.

### 5.3.3 Address decoding

The address mapping of AHB slaves in GRLIB is designed to be distributed, i.e. not rely on a shared static address decoder which must be modified as soon as a slave is added or removed. The GRLIB AHB bus controller, which implements the address decoder, will use the configuration information received from the slaves on HCONFIG to automatically generate the slave select signals (HSEL). When a slave is added or removed during the design, the address decoding function is automatically updated without requiring manual editing.

The AHB address range for each slave is defined by its Bank Address Registers (BAR). Address decoding is performed by comparing the 12-bit ADDR field in the BAR with part of the AHB address (HADDR). There are two types of banks defined for the AHB bus: AHB memory bank and AHB I/O bank. The AHB address decoding is done differently for the two types.

For AHB memory banks, the address decoding is performed by comparing the 12-bit ADDR field in the BAR with the 12 most significant bits in the AHB address (HADDR(31:20)). If equal, the corresponding HSEL will be generated. This means that the minimum address range occupied by an AHB memory bank is 1 MByte. To allow for larger address ranges, only the bits set in the MASK field of the BAR are compared. Consequently, HSEL will be generated when the following equation is true:

$$((\text{BAR.ADDR} \text{ xor } \text{HADDR}[31:20]) \text{ and } \text{BAR.MASK}) = 0$$

As an example, to decode a 16 MByte AHB memory bank at address 0x24000000, the ADDR field should be set to 0x240, and the MASK to 0xFF0. Note: if MASK = 0, the BAR is disabled rather than occupying the full AHB address range.

For AHB I/O banks, the address decoding is performed by comparing the 12-bit ADDR field in the BAR with 12 bits in the AHB address (HADDR(19:8)). If equal, the corresponding HSEL will be generated. This means that the minimum address range occupied by an AHB I/O bank is 256 Byte. To allow for larger address ranges, only the bits set in the MASK field of the BAR are compared. Consequently, HSEL will be generated when the following equation is true:

$$((\text{BAR.ADDR} \text{ xor } \text{HADDR}[19:8]) \text{ and } \text{BAR.MASK}) = 0$$

The 12 most significant bits in the AHB address (HADDR(31:20)) are always fixed to 0xFFFF, effectively placing all AHB I/O banks in the 0xFFFF0000-0xFFFFFEFFF address space. As an example, to decode an 4 kByte AHB I/O bank at address 0xFFFF2400, the ADDR field should be set to 0x240, and the MASK to 0xFF0. Note: if MASK = 0, the BAR is disabled rather than occupying the full AHB I/O address range.

The AHB slaves in GRLIB define the value of their ADDR and MASK fields through generics. This allows to choose the address range for each slave when it is instantiated, without having to modify a central decoder or the slave itself. Below is an example of a component declaration of an AHB RAM memory, and how it can be instantiated:

```
component ahbram
  generic (
    hindex : integer := 0;           -- AHB slave index
    haddr  : integer := 0;
    hmask  : integer := 16#fff#);
  port (
    rst    : in  std_ulogic;
    clk    : in  std_ulogic;
```



```

    ahbsi  : in  ahb_slv_in_type;          -- AHB slave input
    ahbso  : out ahb_slv_out_type);       -- AHB slave output
end component;

ram0 : ahbmem
  generic map (hindex => 1, haddr => 16#240#, hmask => 16#FF0#)
  port map (rst, clk, ahbsi, ahbso(1));

```

An AHB slave can have up to four address mapping registers, thereby decode four independent areas in the AHB address space. HSEL is asserted when any of the areas is selected. To know which particular area was selected, the ahbsi record contains the additional bus signal HBSEL(0:3). The elements in HBSEL(0:3) are asserted if the corresponding to BAR(0-3) caused HSEL to be asserted. HBSEL is only valid when HSEL is asserted. For example, if BAR1 caused HSEL to be asserted, the HBSEL(1) will be asserted simultaneously with HSEL.

### 5.3.4 Cacheability

In processor-based systems without an MMU, the cacheable areas are typically defined statically in the cache controllers. The LEON processors build the cacheability table automatically during synthesis, using the cacheability information in the AHB configuration records. In this way, the cacheability settings always reflect the current configuration.

For systems with an MMU, the cacheability information can be read out by from the configuration records through software. This allows the operating system to build an MMU page table with proper cacheable-bits set in the page table entries.

### 5.3.5 Interrupt steering

GRLIB provides a unified interrupt handling scheme by adding 32 interrupt signals (HIRQ) to the AHB bus, both as inputs and outputs. An AHB master or slave can drive as well as read any of the interrupts.

The output of each master includes all 32 interrupt signals in the vector ahbmo.hirq. An AHB master must therefore use a generic that specifies which HIRQ element to drive. This generic is of type integer, and typically called HIRQ (see example below).

```

component ahbmaster is
  generic (
    hindex : integer := 0;          -- master index
    hirq   : integer := 0);        -- interrupt index
  port (
    reset  : in  std_ulogic;
    clk    : in  std_ulogic;
    hmsti  : in  ahb_mst_in_type;   -- AHB master inputs
    hmsto  : out ahb_mst_out_type;  -- AHB master outputs
  );
end component;

master1 : ahbmaster
  generic map (hindex => 1, hirq => 1)
  port map (rst, clk, hmsti, hmsto(1));

```

The same applies to the output of each slave which includes all 32 interrupt signals in the vector ahbso.hirq. An AHB slave must therefore use a generic that specifies which HIRQ element to drive. This generic is of type integer, and typically called HIRQ (see example below).

```

component ahbslave
  generic (
    hindex : integer := 0;          -- slave index
    hirq   : integer := 0);        -- interrupt index
  port (
    rst     : in  std_ulogic;
    clk     : in  std_ulogic;
    hslvi  : in  ahb_slv_in_type;   -- AHB slave inputs
    hslvo  : out ahb_slv_out_type); -- AHB slave outputs
end component;

slave2 : ahbslave
  generic map (hindex => 2, hirq => 2)
  port map (rst, clk, hslvi, hslvo(1));

```

The AHB bus controller in the GRLIB provides interrupt combining. For each element in HIRQ, all the ahbmo.hirq signals from the AHB masters and all the ahbso.hirq signals from the AHB slaves are logically OR-ed. The combined result is output both on ahbmi.hirq (routed back to the AHB masters)

and ahbsi.hirq (routed back to the AHB slaves). Consequently, the AHB masters and slaves share the same 32 interrupt signals.

An AHB unit that implements an interrupt controller can monitor the combined interrupt vector (either ahbsi.hirq or ahbmi.hirq) and generate the appropriate processor interrupt.

## 5.4 AMBA APB on-chip bus

### 5.4.1 General

The AMBA Advanced Peripheral Bus (APB) is a single-master bus suitable to interconnect units of low complexity which require only low data rates. An APB bus is interfaced with an AHB bus by means of a single AHB slave implementing the AHB/APB bridge. The AHB/APB bridge is the only APB master on one specific APB bus. More than one APB bus can be connected to one AHB bus, by means of multiple AHB/APB bridges. A conceptual view is provided in figure 8.

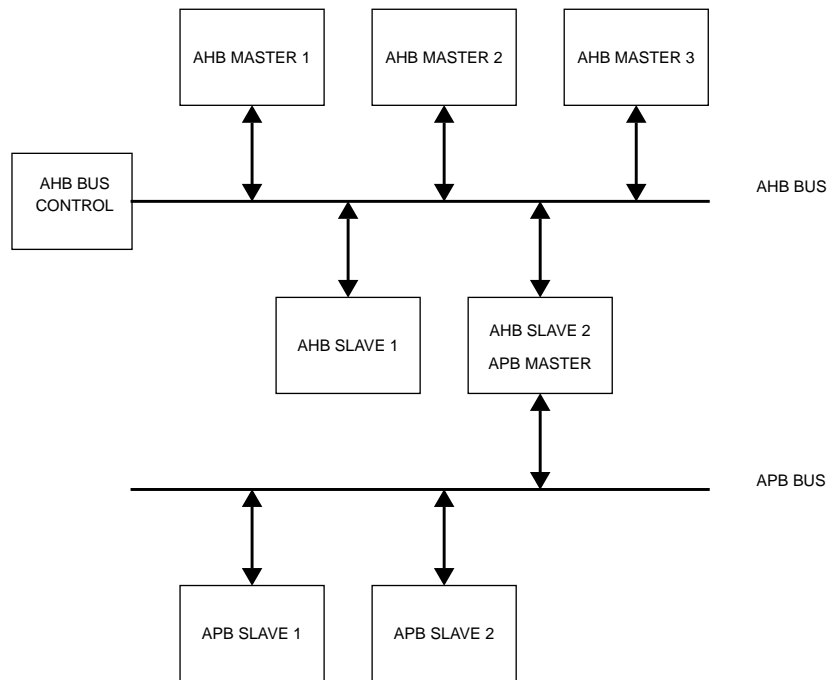


Figure 8. AMBA AHB/APB conceptual view

Since the APB bus is multiplexed (no tristate signals), a more correct view of the bus and the attached units can be seen in figure 9. The access to the AHB slave input (AHBI) is decoded and an access is made on APB bus. The APB master drives a set of signals grouped into a VHDL record called APBI which is sent to all APB slaves. The combined address decoder and bus multiplexer controls which slave is currently selected. The output record (APBO) of the active APB slave is selected by the bus multiplexer and forwarded to AHB slave output (AHBO).



### 5.4.3 AHB/APB bridge

GRLIB provides a combined AHB slave, APB bus master, address decoder and bus multiplexer. It receives the AHBI and AHBO records from the AHB bus, and generates APBI and APBO records on the APB bus. The address decoding function will drive one of the APBI.PSEL elements to indicate the selected APB slave. The bus multiplexer function will select from which APB slave data will be taken to drive the AHBI signal. A typical APB master in GRLIB has the following definition:

```
library IEEE;
use IEEE.std_logic_1164.all;
library grlib;
use grlib.amba.all;

entity apbmst is
  generic (
    hindex : integer := 0;           -- AHB slave bus index
  );
  port (
    rst      : in  std_ulogic;
    clk      : in  std_ulogic;
    ahbi     : in  ahb_slv_in_type;  -- AHB slave inputs
    ahbo     : out ahb_slv_out_type;  -- AHB slave outputs
    apbi     : out apb_slv_in_type;   -- APB master inputs
    apbo     : in  apb_slv_out_vector -- APB master outputs
  );
end;
```

### 5.4.4 APB bus index control

The APB slave output records contain the sideband signal PINDEX. This signal is used to verify that the slave is driving the correct element of the AHBPO bus. The generic PINDEX that is used to select the appropriate PSEL is driven back on APBO.PINDEX. The APB controller then checks that the value of the received PINDEX is equal to the bus index. An error is issued during simulation if a mismatch is detected.

## 5.5 APB plug&play configuration

### 5.5.1 General

The GRLIB implementation of the APB bus includes the same type of mechanism to provide plug&play support as for the AHB bus. The plug&play support consists of three parts: identification of attached slaves, address mapping, and interrupt routing. The plug&play information for each APB slave consists of a configuration record containing two 32-bit words. The first word is called the identification register and contains information on the device type and interrupt routing. The last word is the bank address register (BAR) and contains address mapping information for the APB slave. Only a single BAR is defined per APB slave. An APB slave is neither prefetchable nor cacheable.

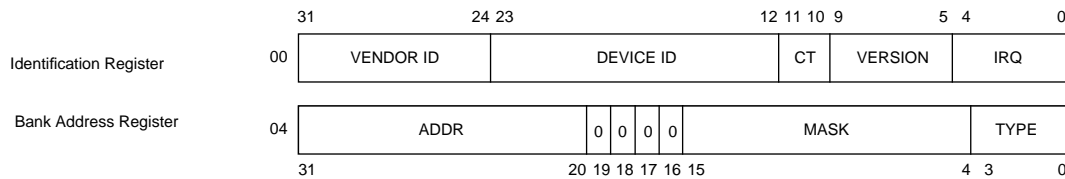


Figure 10. APB plug&play configuration layout

All addressing of the APB is referenced to the AHB address space. The 12 most significant bits of the AHB bus address are used for addressing the AHB slave of the AHB/APB bridge, leaving the 20 least significant bits for APB slave addressing.

The plug&play information for all attached APB slaves appear as a read-only table mapped on a fixed address of the AHB, typically at 0x---FF000. The configuration records of the APB slaves appear in 0x---FF000 - 0x---FFFFFF on the AHB bus. Since each record is 2 words (8 bytes), the table has space for 512 slaves on a single APB bus. A plug&play operating system (or any other application) can scan the configuration table and automatically detect which units are present on the APB bus, how they are configured, and where they are located (slaves).

The configuration record from each APB unit is sent to the APB bus controller via the PCONFIG signal. The bus controller creates the configuration table automatically, and creates a read-only memory area at the desired address (default 0x---FF000). Since the configuration information is fixed, it can be efficiently implemented as a small ROM or with relatively few gates. A special reporting module (apbreport) is provided in the WORK.DEBUG package of Grlib which can be used to print the configuration table to the console during simulation.

### 5.5.2 Device identification

The APB bus uses same type of Identification Register as previously defined for the AHB bus.

### 5.5.3 Address decoding

The address mapping of APB slaves in GRLIB is designed to be distributed, i.e. not rely on a shared static address mapping which must be modified as soon as a slave is added or removed. The GRLIB APB master, which implements the address decoder, will use the configuration information received from the slaves on PCONFIG to automatically generate the slave select signals (PSEL). When a slave is added or removed during the design, the address decoding function is automatically updated without requiring manual editing.

The APB address range for each slave is defined by its Bank Address Registers (BAR). There is one type of banks defined for the APB bus: APB I/O bank. Address decoding is performed by comparing the 12-bit ADDR field in the BAR with 12 bits in the AHB address (HADDR(19:8)). If equal, the corresponding PSEL will be generated. This means that the minimum address range occupied by an APB I/O bank is 256 Byte. To allow for larger address ranges, only the bits set in the MASK field of the BAR are compared. Consequently, PSEL will be generated when the following equation is true:

$$((\text{BAR.ADDR} \text{ xor } \text{HADDR}[19:8]) \text{ and } \text{BAR.MASK}) = 0$$

As an example, to decode an 4 kByte AHB I/O bank at address 0x---24000, the ADDR field should be set to 0x240, and the MASK to 0xFF0. Note that the 12 most significant bits of AHBI.HADDR are

used for addressing the AHB slave of the AHB/APB bridge, leaving the 20 least significant bits for APB slave addressing.

As for AHB slaves, the APB slaves in GRLIB define the value of their ADDR and MASK fields through generics. This allows to choose the address range for each slave when it is instantiated, without having to modify a central decoder or the slave itself. Below is an example of a component declaration of an APB I/O unit, and how it can be instantiated:

```
component apbio
  generic (
    pindex : integer := 0;
    paddr   : integer := 0;
    pmask   : integer := 16#fff#);
  port (
    rst      : in  std_ulogic;
    clk      : in  std_ulogic;
    apbi     : in  apb_slv_in_type;
    apbo     : out apb_slv_out_type);
end component;

io0 : apbio
  generic map (pindex => 1, paddr => 16#240#, pmask => 16#FF0#)
  port map (rst, clk, apbi, apbo(1));
```

### 5.5.4 Interrupt steering

GRLIB provides a unified interrupt handling scheme by also adding 32 interrupt signals (PIRQ) to the APB bus, both as inputs and outputs. An APB slave can drive as well as read any of the interrupts. The output of each slave includes all 32 interrupt signals in the vector APBO.PIRQ. An APB slave must therefore use a generic that specifies which PIRQ element to drive. This generic is of type integer, and typically called PIRQ (see example below).

```
component apbslave
  generic (
    pindex : integer := 0;           -- slave index
    pirq    : integer := 0);         -- interrupt index
  port (
    rst      : in  std_ulogic;
    clk      : in  std_ulogic;
    apbi     : in  apb_slv_in_type;  -- APB slave inputs
    apbo     : out apb_slv_out_type); -- APB slave outputs
end component;

slave3 : apbslave
  generic map (pindex => 1, pirq => 2)
  port map (rst, clk, pslvi, pslvo(1));
```

The AHB/APB bridge in the GRLIB provides interrupt combining, and merges the APB-generated interrupts with the interrupts bus on the AHB bus. This is done by OR-ing the 32-bit interrupt vectors from each APB slave into one joined vector, and driving the combined value on the AHB slave output bus (AHBSO.HIRQ). The APB interrupts will then be merged with the AHB interrupts. The resulting interrupt vector is available on the AHB slave input (AHBSI.HIRQ), and is also driven on the APB slave inputs (APBI.PIRQ) by the AHB/APB bridge. Each APB slave (as well as AHB slave) thus sees the combined AHB/APB interrupts. An interrupt controller can then be placed either on the AHB or APB bus and still monitor all interrupts.

## 5.6 Technology mapping

### 5.6.1 General

GRLIB provides portability support for both ASIC and FPGA technologies. The support is implemented by means of encapsulation of technology specific components such as memories, pads and clock buffers. The interface to the encapsulated component is made technology independent, not relying on any specific VHDL or Verilog code provided by the foundry or FPGA manufacturer. The interface to the component stays therefore always the same. No modification of the design is therefore required if a different technology is targeted. The following technologies are currently supported by the TECHMAP.GENCOMP package:

```
constant inferred : integer := 0;
constant virtex   : integer := 1;
constant virtex2  : integer := 2;
constant memvirage : integer := 3;
constant axcel    : integer := 4;
constant proasic  : integer := 5;
constant atcl8s   : integer := 6;
```

```

constant altera      : integer := 7;
constant umc         : integer := 8;
constant rhumc       : integer := 9;
constant apa3        : integer := 10;
constant spartan3    : integer := 11;
constant ihp25       : integer := 12;
constant rhlib18t    : integer := 13;
constant virtex4     : integer := 14;
constant lattice     : integer := 15;
constant ut25        : integer := 16;
constant spartan3e   : integer := 17;
constant peregrine   : integer := 18;
constant memartisan  : integer := 19;
constant virtex5     : integer := 20;
constant custom1     : integer := 21;
constant ihp25rh     : integer := 22;
constant stratix1    : integer := 23;
constant stratix2    : integer := 24;
constant eclipse     : integer := 25;
constant stratix3    : integer := 26;
constant cyclone3    : integer := 27;
constant memvirage90 : integer := 28;
constant tsmc90      : integer := 29;
constant easic90     : integer := 30;
constant atcl8rha    : integer := 31;
constant smic013     : integer := 32;
constant tm65gpl     : integer := 33;
constant axdsp       : integer := 34;
constant spartan6    : integer := 35;
constant virtex6     : integer := 36;
constant actfus      : integer := 37;
constant stratix4    : integer := 38;
constant st65lp      : integer := 39;
constant st65gp      : integer := 40;
constant easic45     : integer := 41;
constant cmos9sf     : integer := 42;
constant apa3e       : integer := 43;
constant apa31       : integer := 44;
constant utl30       : integer := 45;
constant ut90        : integer := 46;
constant gf65        : integer := 47;
constant virtex7     : integer := 48;
constant kintex7     : integer := 49;

```

Each encapsulating component provides a VHDL generic (normally named TECH) with which the targeted technology can be selected. The generic is used by the component to select the correct technology specific cells to instantiate in its architecture and to configure them appropriately. This method does not rely on the synthesis tool to inferring the correct cells.

For technologies not defined in GRLIB, the default “inferred” option can be used. This option relies on the synthesis tool to infer the correct technology cells for the targeted device.

A second VHDL generic (normally named MEMTECH) is used for selecting the memory cell technology. This is useful for ASIC technologies where the pads are provided by the foundry and the memory cells are provided by a different source. For memory cells, generics are also used to specify the address and data widths, and the number of ports.

The two generics TECH and MEMTECH should be defined at the top level entity of a design and be propagated to all underlying components supporting technology specific implementations.

### 5.6.2 Memory blocks

Memory blocks are often implemented with technology specific cells or macrocells and require an encapsulating component to offer a unified technology independent interface. The TECHMAP library provides such technology independent memory component, as the synchronous single-port RAM shown in the following code example. The address and data widths are fully configurable by means of the generics ABITS and DBITS, respectively.

```

component syncram
generic (
    memtech : integer := 0;           -- memory technology
    abits   : integer := 6;           -- address width
    dbits   : integer := 8);          -- data width
port (
    clk      : in  std_ulogic;
    address  : in  std_logic_vector((abits -1) downto 0);
    datain   : in  std_logic_vector((dbits -1) downto 0);
    dataout  : out std_logic_vector((dbits -1) downto 0);
    enable   : in  std_ulogic;
    write    : in  std_ulogic);
end component;

```



This synchronous single-port RAM component is used in the AHB RAM component shown in the following code example.

```
component ahb_ram
  generic (
    hindex : integer := 0;           -- AHB slave index
    haddr   : integer := 0;
    hmask   : integer := 16#fff#;
    memtech : integer := 0;         -- memory technology
    kbytes  : integer := 1);        -- memory size
  port (
    rst      : in  std_ulogic;
    clk      : in  std_ulogic;
    hslvi    : in  ahb_slv_in_type; -- AHB slave input
    hslvo    : out ahb_slv_out_type; -- AHB slave output
  );
end component;

ram0 : ahb_ram
  generic map (hindex => 1, haddr => 16#240#, hmask => 16#FF0#,
               tech => virtex, kbytes => 4)
  port map (rst, clk, hslvi, hslvo(1));
```

In addition to the selection of technology (VIRTEX in this case), the size of the AHB RAM is specified in number of kilo-bytes. The conversion from kilo-bytes to the number of address bits is performed automatically in the AHB RAM component. In this example, the data width is fixed to 32 bits and requires no generic. The VIRTEX constant used in this example is defined in the TECHMAP.GENCOMP package.

### 5.6.3 Pads

As for memory cells, the pads used in a design are always technology dependent. The TECHMAP library provides a set of encapsulated components that hide all the technology specific details from the user. In addition to the VHDL generic used for selecting the technology (normally named TECH), generics are provided for specifying the input/output technology levels, voltage levels, slew and driving strength. A typical open-drain output pad is shown in the following code example:

```
component odpad
  generic (
    tech      : integer := 0;
    level     : integer := 0;
    slew      : integer := 0;
    voltage    : integer := 0;
    strength  : integer := 0);
  port (
    pad       : out std_ulogic;
    o         : in  std_ulogic);
end component;

pad0 : odpad
  generic map (tech => virtex, level => pci33, voltage => x33v)
  port map (pad => pci_irq, o => irqn);
```

The TECHMAP.GENCOMP package defines the following constants that to be used for configuring pads:

```
-- input/output voltage
constant x18v : integer := 1;
constant x25v : integer := 2;
constant x33v : integer := 3;
constant x50v : integer := 5;

-- input/output levels
constant ttl      : integer := 0;
constant cmos     : integer := 1;
constant pci33    : integer := 2;
constant pci66    : integer := 3;
constant lvds     : integer := 4;
constant sstl2_i  : integer := 5;
constant sstl2_ii : integer := 6;
constant sstl3_i  : integer := 7;
constant sstl3_ii : integer := 8;

-- pad types
constant normal   : integer := 0;
constant pullup   : integer := 1;
constant pulldown : integer := 2;
```

```
constant opendrain: integer := 3;  
constant schmitt  : integer := 4;  
constant dci      : integer := 5;
```

The slow control and driving strength is not supported by all target technologies, or is often implemented differently between different technologies. The documentation for the IP core implementing the pad should be consulted for details.

## 5.7 Scan test support

### 5.7.1 Overview

Scan test is a method for production testing digital ASICs. A test mode is added to the design that changes all flip-flops in the design to shift registers that can be set and read out serially. This is implemented partially in RTL code and partially in the implementation flow.

In a typical GRLIB ASIC, a number of signals are added for scan test. All signals except `testen` are usually muxed with other slow I/O signals so only one pin has to be added to the design.

The signals added are:

`testen` - Enables test mode (top-level pin)

`scanen` - Muxes flip-flop data inputs to previous in chain instead of normal function

`testoen` - Controls all output-enables in test mode

`testrst` - Controls all async-resets in test mode

`scanin` - Scan chain inputs

`scanout` - Scan chain outputs

The top level of the design adds the `testen` signal to the port list and muxes in the `scanen`, `testoen` and `testrst` signals. The `scanin` and `scanout` signals are not handled at the RTL level.

At the RTL level, the test signals are connected to any hard macro that needs them, such as block RAM:s and PLL:s. Also `testoen` and `testrst` are handled fully at source code level. The RTL also contains logic so that all flip-flops are directly clocked by an input clock pin when test mode is enabled.

During synthesis, the synthesis tool implements registers using special "scan flip-flops" containing the necessary muxing for the scan chain. The actual scan chain connections are not derived until after placement, so the scan order can be selected to minimize routing.

### 5.7.2 GRLIB support

To support scan test methods, GRLIB distributes the `testen`, `scanen`, `testoen` and `testrst` signals via the AHB and APB bus records. The signals are supplied into the AHB controller which will pass them on to the AHB bus records. The APB controller and AHB-to-AHB bridges will pass them on to their bus signals. This way all IP cores connecting to an AHB or APB bus have access to the test signals without having to add extra input ports for them.

The GRLIB IP cores supporting scan test signals have a generic called `scantest` to enable this functionality. For historical reasons, this generic is on some IP cores called `scanen` or `testen` instead. Cores which use the scan signals include LEON3, MCTRL and GRGPIO.

The techmap layer handles certain test mode features. The `clkgate` component will automatically enable (pass through) the clock when test mode is enabled. The various syncram wrappers will disable the RAM:s during shifting (when `scanen` and `testen` are high).

The syncram techmaps have an input vector called `testin`, containing `testen`, `scanen`, plus two extra technology-dependent bits. The AMBA records contain a `testin` element that can be passed on directly to the syncram. The tech dependent bits can be set using the `testsig` input signal to the AHB controller. More bits can be added to the vector if necessary via a local GRLIB configuration option.

### 5.7.3 Usage for existing cores

For using the scan test support with existing cores in GRLIB, the test signals need to be supplied to the AHB controller and the scan test support needs to be enabled in the IP cores.

### 5.7.4 Usage for new cores

For adding scan test support to an IP core, a couple of changes may be needed.

- A generic called scantest should be added that enables scan test support. If the core does not have any AHB or APB interfaces, you will also need to add explicit inputs for any test signals that you need to implement the below.
- If the core has asynchronous resets, these should be tied to testrst when testen is high. This is usually done by a statement such as:
- If the core controls output enables going directly to pads, these should be tied directly to testoen when testen is high.
- If you invert or divide clocks internally, these should be bypassed in test mode so all flip-flops are clocked by the same edge on the incoming clock:

```
lnclk <= not clk;
stgen: if scantest /= 0 generate
  m1: clkmux
    generic map (tech => tech)
    port map (io => lnclk, il => clk, sel => ahbsi.testen, o => nclk);
end generate;
nstgen: if scantest = 0 generate
  nclk <= lnclk;
end generate;
```

- Pass on the scantest generic and test signals to any submodules, techmap instances and hard macros that need them.

## 6 GRLIB Design examples

### 6.1 Introduction

The template design examples described in the following sections are provided for the understanding of how to integrate the existing GRLIB IP cores into a design. The documentation for the various IP cores should be consulted for details.

### 6.2 NetCard

The NetCard design example described in this section is a simple PCI to Ethernet bridge. The design is based on IP cores from GRLIB, including the GRPCI PCI bridge and the GRETH Ethernet MAC. The VHDL code of the design is listed in its full hereafter, but has been split into sections to allow for explanations after the source code. The design is located in `grib/designs/netcard`.

```
library ieee;
use ieee.std_logic_1164.all;
library grib;
use grib.amba.all;           -- AMBA AHB/APB components
library techmap;
use techmap.gencomp.all;     -- technology
use grib.stdlib.all;         -- utilities
library gaisler;
use gaisler.uart.all;        -- AMBA AHB/APB UARTs
use gaisler.misc.all;        -- miscellaneous
use gaisler.pci.all;         -- PCI
use gaisler.net.all;         -- network cores
use work.config.all;         -- design configuration
```

The GRLIB and GAISLER VHDL libraries are used for this design. Only the most important packages are explained. The AHB bus controller and the AHB/APB bridge components are defined in the GRLIB.AMBA package. The technology selection is defined in the TECHMAP.GENCOMP package.

```
entity netcard is
  generic (
    fabtech    : integer := CFG_FABTECH;
    memtech    : integer := CFG_MEMTECH;
    padtech    : integer := CFG_PADTECH;
    clktech    : integer := CFG_CLKTECH
  );
```

The TECH and MEMTECH generics are used for selecting the overall technology and the memory technology. It is possible to include optionally a debugger and a PCI signal tracer. It is possible to select the functionality of the PCI bridge, either as target only or as combined initiator/target.

```
port (
  resetn      : in    std_ulogic;
  clk         : in    std_ulogic;

  dsutx       : out   std_ulogic;           -- DSU tx data
  dsurx       : in    std_ulogic;           -- DSU rx data

  emdio       : inout std_logic;             -- ethernet
  etx_clk     : in    std_logic;
  erx_clk     : in    std_logic;
  erxd        : in    std_logic_vector(3 downto 0);
  erx_dv      : in    std_logic;
  erx_er      : in    std_logic;
  erx_col     : in    std_logic;
  erx_crs     : in    std_logic;
  etxd        : out   std_logic_vector(3 downto 0);
  etx_en      : out   std_logic;
  etx_er      : out   std_logic;
  emdc        : out   std_logic;

  pci_rst     : in    std_ulogic;           -- PCI
  pci_clk     : in    std_ulogic;
  pci_gnt     : in    std_ulogic;
  pci_idsel   : in    std_ulogic;
  pci_lock    : inout std_ulogic;
  pci_ad      : inout std_logic_vector(31 downto 0);
  pci_cbe     : inout std_logic_vector(3 downto 0);
  pci_frame   : inout std_ulogic;
  pci_irdy    : inout std_ulogic;
  pci_trdy    : inout std_ulogic;
  pci_devsel  : inout std_ulogic;
  pci_stop    : inout std_ulogic;
  pci_perr    : inout std_ulogic;
  pci_par     : inout std_ulogic;
  pci_req     : inout std_ulogic;
```

```

    pci_serr    : inout std_ulogic;
    pci_irq     : out   std_ulogic;
    pci_host    : in    std_ulogic;
    pci_66      : in    std_ulogic;
end;
```

The interface ports of the design are all defined as standard IEEE 1164 types.

```

architecture rtl of netcard is
    signal apbi : apb_slv_in_type;
    signal apbo : apb_slv_out_vector := (others => apb_none);
```

Local signal declarations for the APB slave inputs and outputs. The outputs are contained in a vector and each APB slave drives it own element. Note that a default value is given to the APB output vector in the architecture declarative part. This is generally not supported for synthesis, but all synthesis tools supported by GRLIB generate all-zero values which makes the outcome deterministic. If this design style is not accepted by a tool or user, the unused entries in the vector should be assigned the default value explicitly in the architecture statement part.

```

    signal ahbsi : ahb_slv_in_type;
    signal ahbso : ahb_slv_out_vector := (others => ahbs_none);
```

Local signal declarations for the AHB slave inputs and outputs. The outputs are contained in a vector, and each AHB slave drives it own element.

```

    signal ahbmi : ahb_mst_in_type;
    signal ahbmo : ahb_mst_out_vector := (others => ahbm_none);
```

Local signal declarations for the AHB masters inputs and outputs. The outputs are contained in a vector, and each AHB masters drives it own element.

```

    signal clk_m, rstn, pciclk : std_ulogic;
    signal cgi : clkgen_in_type;
    signal cgo : clkgen_out_type;
    signal dui : uart_in_type;
    signal duo : uart_out_type;
    signal pcii : pci_in_type;
    signal pcio : pci_out_type;
    signal ethi : eth_in_type;
    signal etho : eth_out_type;
    signal irqn : std_logic;
```

The rest of the local signal declarations are used for the clock generation, debugger, PCI and Ethernet interfaces.

```

begin

-----
---  Reset and Clock generation  -----
-----

    cgi.pllctrl <= "00";
    cgi.pllrst  <= resetn;
    cgi.pllref  <= '0';

    clkgen0 : clkgen -- clock generator
        generic map (clk_mul => 4, clk_div => 2, pcien => pci, tech => tech)
        port map (clk, pci_clk, clk_m, open, open, open, pciclk, cgi, cgo);
    rst0 : rstgen -- reset generator
        port map (resetn, clk_m, cgo.clkclock, rstn);
```

The clock generator can be implemented using technology specific cells, which is controlled by the CLKTECH generic.

```

-----
---  AHB CONTROLLER  -----
-----

    ahb0 : ahbctrl -- AHB arbiter/multiplexer
        port map (rstn, clk_m, ahbmi, ahbmo, ahbsi, ahbso);
```

The GRLIB GAISLER AHB bus controller is used for implementing the AHB arbiter, address decoder and multiplexer. All AHB master and slave inputs/outputs are route through the controller.

```

-----
---  ETHERNET  -----
-----

e0 : greth generic map(hindex => log2x(CFG_PCI),
    pindex => 0, paddr => 11, pirq => 11, memtech => memtech)
    port map( rst => rstn, clk => clk, ahbmi => ahbmi, ahbmo => ahbmo(log2x(CFG_PCI)),
```

```
apbi => apbi, apbo => apbo(0), ethi => ethi, etho => etho);
```

The GRETH Ethernet interface is an AHB master and an APB slave. The generic `hindex` defines its AHB master number and the generic `pindex` defines its APB slave index. Note that `hindex` and the index used for selecting the correct element in the AHBMO vector must be the same. The same applies to `pindex` and `apbo`. The two indices have no relation to the address mapping of the slave. The address of the APB bank is specified by the `paddr` generic, and in this case its starting address will be 0x80000B00. The `IRQ` generic specifies that the device will generate interrupts on interrupt vector element 11.

```
emdio_pad : iopad generic map (tech => padtech)
port map (emdio, etho.mdio_o, etho.mdio_oe, ethi.mdio_i);
etxc_pad : clkpad generic map (tech => padtech, arch => 1)
port map (etx_clk, ethi.tx_clk);
erxc_pad : clkpad generic map (tech => padtech, arch => 1)
port map (erx_clk, ethi.rx_clk);
erxd_pad : inpadv generic map (tech => padtech, width => 4)
port map (erxd, ethi.rxd(3 downto 0));
erxdv_pad : inpad generic map (tech => padtech)
port map (erx_dv, ethi.rx_dv);
erxer_pad : inpad generic map (tech => padtech)
port map (erx_er, ethi.rx_er);
erxco_pad : inpad generic map (tech => padtech)
port map (erx_col, ethi.rx_col);
erxcr_pad : inpad generic map (tech => padtech)
port map (erx_crs, ethi.rx_crs);

etxd_pad : outpadv generic map (tech => padtech, width => 4)
port map (etxd, etho.txd(3 downto 0));
etxen_pad : outpad generic map (tech => padtech)
port map (etx_en, etho.tx_en);
etxer_pad : outpad generic map (tech => padtech)
port map (etx_er, etho.tx_er);
emdc_pad : outpad generic map (tech => padtech)
port map (emdc, etho.mdc);

irqn      <= ahbso(3).hirq(11);

irq_pad : odpad generic map (tech => padtech, level => pci33)
port map (pci_irq, irqn);
```

All Ethernet interface signals are mapped pads with tech mapping, selecting the appropriate pads for the selected target technology. A pad is explicitly instantiated for the interrupt output, ensuring that an open-drain output with PCI33 levels is being used.

```
-----
---  AHB/APB Bridge  -----
-----
apb0 : apbctrl -- AHB/APB bridge
generic map (hindex => 0, haddr => 16#800#)
port map (rstn, clk, ahbsi, ahbso(0), apbi, apbo );
```

The GRLIB AHB/APB bridge is instantiated as a slave on the AHB bus. The `HINDEX` generic specifies its index on the AHB slave bus, and the `HADDR` generic specifies that the corresponding APB bus address area will be starting from AHB address 0x80000000.

```
-----
---  AHB RAM  -----
-----
ram0 : ahbram
generic map (hindex => 2, haddr => 0, hmask => 16#FFF#,
tech => memtech, kbytes => 8)
port map (rstn, clk, ahbsi, ahbso(2));
```

A local RAM is implemented as a slave on the AHB bus. The technology selection is made with the `MEMTECH` generic. The size is specified to be 8 kbytes with the `KBYTES` generic, and the memory is located at address 0x00000000 as specified by `HADDR`. The `HMASK` generic allocates a minimum 1 Mbyte address space on the AHB bus.

```
-----
---  PCI  -----
-----
pp : if pci /= 0 generate
pci_gr0 : if pci = 1 generate
pci0 : pci_target
generic map (hindex => 0,
device_id => 16#0210#, vendor_id => 16#16E3#)
port map (rstn, clk, pciclk, pcii, pcio, ahbmi, ahbmo(0));
end generate;
pci_mtf0 : if pci = 2 generate
pci0 : pci_mtf
generic map (memtech => memtech, hmstndx => 0,
```

```

        fifodepth => 6, device_id => 16#0210#,
        vendor_id => 16#16E3#, hslvndx => 1,
        pindex => 6, paddr => 2, haddr => 16#E00#,
        loaddr => 16#400#, nsync => 2)
    port map (rstn, clk, pciclk, pcii, pcio, apbi, apbo(2),
        ahbmi, ahbmo(0), ahbsi, ahbso(1));
end generate;
pci_trc0 : if pcitrc /= 0 generate
    pt0 : pcitrace
        generic map (memtech => memtech, pindex => 3,
            paddr => 16#100#, pmask => 16#f00#)
        port map (rstn, clk, pciclk, pcii, apbi, apbo(3));
end generate;
pcipads0 : pcipads
    generic map (tech)
    port map (pci_rst, pci_gnt, pci_idsel, pci_lock, pci_ad, pci_cbe,
        pci_frame, pci_irdy, pci_trdy, pci_devsel, pci_stop,
        pci_perr, pci_par, pci_req, pci_serr, pci_host, pci_66,
        pcii, pcio);
end generate;

```

If the PCI interface is implemented as a target only, the device is only implemented as a master on AHB. This option does not require any on-chip memory and no technology selection is required. The PCI device and vendor ID is specified by means of generics.

For an initiator/target PCI interface, the device is implemented as both master and slave on AHB. This option implements on-chip memory for which the technology is selected with the MEMTECH generic. The size of the memory is selected with the FIFODEPTH generic and it is located at 0xE0000000 as specified by HADDR. The I/O bank of the device is located at AHB address 0x40000000. This option also implements a APB slave, and the PINDEX generic is used for specifying its APB bus number.

Not shown in this example is that there are several other generics specified for the PCI IP cores for which default values are being used. What should be noted is that most of the generics are hard coded in this example, not allowing the design to be changed by means of top level entity generics.

The pads for the PCI interface are implemented in the PCIPADS component, which only uses the TECH generic since the signal levels are already determined.

As an option, a PCI signal trace buffer can be included in the design. The trace buffer samples PCI signal activity and stores the data in a local on-chip memory. The trace buffer is accessible as an APB slave I/O bank of 4 kBytes at AHB address 0x80010000 as specified by the PADDR and PMASK generics. The 0x800 part of the address is specified by the AHB/APB bridge HADDR generic as explained above.

```

-----
--- Optional DSU UART -----
-----
dcomgen : if dbg = 1 generate
    dcom0 : ahbuart -- Debug UART
        generic map (ahbndx => 2, apbndx => 1, apbaddr => 1)
        port map (rstn, clk, dui, duo, apbi, apbo(1), ahbmi, ahbmo(2));
        dui.rxd <= dsurx; dsutx <= duo.txd;
    end generate;

```

An option debug support unit serial interface can be included in the design. The DSU acts as an AHB master and as an APB slave.

```

-----
--- Boot message -----
-----
-- pragma translate_off

apbrep : apbreport -- APB reporting module
generic map (haddr => 16#800#)
port map (apbo);

ahbrep : ahbreport -- AHB reporting module
port map (ahbmo, ahbso);

x : report_version
generic map (
    msg1 => "Network Card Demonstration design",
    msg2 => "GRLIB Version " & tost(LIBVHDL_VERSION/100) &
        "." & tost(LIBVHDL_VERSION mod 100),
    msg3 => "Target technology: " & tech_table(tech) &
        ", memory library: " & tech_table(memtech),
    mdel => 1
);
-- pragma translate_on
end;

```

Finally, a component is added to the design which generates a report during simulation regarding the GRLIB version and technology selections. The component is not included in synthesis, as indicated by the pragma usage.

To simulate the default design, move to the `grlib/designs/netcard` directory and execute the ‘`vsim`’ command.

```
$ vsim -c netcard
```

Simulate the first 100 ns by writing ‘`run`’.

```
# Ethernet/PCI Network Card Demonstration design
# GRLIB Version 1.0.15, build 2194
# Target technology: virtex2 , memory library: virtex2
# ahbctrl: AHB arbiter/multiplexer rev 1
# ahbctrl: Common I/O area disabled
# ahbctrl: AHB masters: 3, AHB slaves: 4
# ahbctrl: Configuration area at 0xfffff000, 4 kbyte
# ahbctrl: mst0: Gaisler Research      Fast 32-bit PCI Bridge
# ahbctrl: mst1: Gaisler Research      GR Ethernet MAC
# ahbctrl: mst2: Gaisler Research      AHB Debug UART
# ahbctrl: slv0: Gaisler Research      AHB/APB Bridge
# ahbctrl:      memory at 0x80000000, size 1 Mbyte
# ahbctrl: slv1: Gaisler Research      Fast 32-bit PCI Bridge
# ahbctrl:      memory at 0xe0000000, size 256 Mbyte
# apbctrl: APB Bridge at 0x80000000 rev 1
# apbctrl: slv0: Gaisler Research      GR Ethernet MAC
# apbctrl:      I/O ports at 0x80000b00, size 256 byte
# apbctrl: slv1: Gaisler Research      AHB Debug UART
# apbctrl:      I/O ports at 0x80000100, size 256 byte
# apbctrl: slv3: Gaisler Research      32-bit PCI Trace Buffer
# apbctrl:      I/O ports at 0x80010000, size 64 kbyte
# apbctrl: slv6: Gaisler Research      Fast 32-bit PCI Bridge
# apbctrl:      I/O ports at 0x80000200, size 256 byte
# ahbuart1: AHB Debug UART rev 0
# pci_mtf1: 32-bit PCI/AHB bridge rev 0, 2 Mbyte PCI memory BAR, 64-word FIFOs
# greth1: 10/100 Mbit Ethernet MAC rev 01, EDCL 0, buffer 0 kbyte 8 txfifo
# clkgen_virtex2: virtex-2 sdram/pci clock generator, version 1
# clkgen_virtex2: Frequency 25000 KHz, DCM divisor 2/2
```

The report shows that the Xilinx Virtex-2 technology is used for pads, clock generation and memories. The PCI initiator/target bridge is implemented, and the optional PCI trace buffer is included.

Generics can be provided as command line arguments to ‘`vsim`’. It is simple to simulate an ASIC instead of an Xilinx Virtex-2 implementation.

```
$ vsim -gtech=6 -gmementech=3 -gclktch=0 -c netcard
```

Simulate the first 100 ns by writing ‘`run`’.

```
# Ethernet/PCI Network Card Demonstration design
# GRLIB Version 1.0.15, build 2194
# Target technology: atcl8, memory library: virage
# ahbctrl: AHB arbiter/multiplexer rev 1
# ahbctrl: Common I/O area disabled
# ahbctrl: AHB masters: 3, AHB slaves: 4
# ahbctrl: Configuration area at 0xfffff000, 4 kbyte
# ahbctrl: mst0: Gaisler Research      Fast 32-bit PCI Bridge
# ahbctrl: mst1: Gaisler Research      GR Ethernet MAC
# ahbctrl: mst2: Gaisler Research      AHB Debug UART
# ahbctrl: slv0: Gaisler Research      AHB/APB Bridge
# ahbctrl:      memory at 0x80000000, size 1 Mbyte
# ahbctrl: slv1: Gaisler Research      Fast 32-bit PCI Bridge
# ahbctrl:      memory at 0xe0000000, size 256 Mbyte
# apbctrl: APB Bridge at 0x80000000 rev 1
# apbctrl: slv0: Gaisler Research      GR Ethernet MAC
# apbctrl:      I/O ports at 0x80000b00, size 256 byte
# apbctrl: slv1: Gaisler Research      AHB Debug UART
# apbctrl:      I/O ports at 0x80000100, size 256 byte
# apbctrl: slv3: Gaisler Research      32-bit PCI Trace Buffer
# apbctrl:      I/O ports at 0x80010000, size 64 kbyte
# apbctrl: slv6: Gaisler Research      Fast 32-bit PCI Bridge
# apbctrl:      I/O ports at 0x80000200, size 256 byte
# ahbuart1: AHB Debug UART rev 0
# pci_mtf1: 32-bit PCI/AHB bridge rev 0, 2 Mbyte PCI memory BAR, 64-word FIFOs
# greth1: 10/100 Mbit Ethernet MAC rev 01, EDCL 0, buffer 0 kbyte 8 txfifo
```

The report shows that the ACT18 technology is used for pads and Virage technology for the memories.



### 6.3 LEON3MP

The LEON3MP design example described in this section is a multi-processor system based on LEON3MP. The design is based on IP cores from GRLIB. Only part of the VHDL code is listed hereafter, with comments after each excerpt. The design and the full source code is located in `grib/designs/leon3mp`.

```
entity leon3mp is
  generic (
    ncpu      : integer := 1;
```

The number of LEON3 processors in this design example can be selected by means of the NCPU generic shown in the entity declaration excerpt above.

```
signal leon3i : l3_in_vector(0 to NCPU-1);
signal leon3o : l3_out_vector(0 to NCPU-1);
signal irqi   : irq_in_vector(0 to NCPU-1);
signal irqo   : irq_out_vector(0 to NCPU-1);
signal l3dbg_i : l3_debug_in_vector(0 to NCPU-1);
signal l3dbg_o : l3_debug_out_vector(0 to NCPU-1);
```

The debug support and interrupt handling is implemented separately for each LEON3 instantiation in a multi-processor system. The above signals are therefore declared in numbers corresponding to the NCPU generic.

```
signal apbi : apb_slv_in_type;
signal apbo : apb_slv_out_vector := (others => apb_none);
signal ahbsi : ahb_slv_in_type;
signal ahbso : ahb_slv_out_vector := (others => ahbs_none);
signal ahbmi : ahb_mst_in_type;
signal ahbmo : ahb_mst_out_vector := (others => ahbm_none);
```

The multiple LEON AMBA interfaces do not need any special handling in this example, and the AHB master/slave are therefore declared in the same way as in the previous example.

```
-----
---  LEON3 processor and DSU  -----
-----
cpu : for i in 0 to NCPU-1 generate
  u0 : leon3s      -- LEON3 processor
    generic map (hindex => i, fabtech => FABTECH, memtech => MEMTECH,
      fpu => fpu, dsu => dbg, disas => disas,
      pslow => pslow, tbuf => 8*dbg,
      v8 => 2, mac => 1, nwp => 2, lddel => 1,
      isetsize => 1, ilinesize => 8, dsetsize => 1,
      dlinesize => 8, dsnoop => 0)
    port map (clk, rstn, ahbmi, ahbmo(i), ahbsi, leon3i(i), leon3o(i));

    irqi(i)      <= leon3o(i).irq;
    leon3i(i).irq <= irqo(i);
    leon3i(i).debug <= l3dbg_i(i);
    l3dbg_o(i)    <= leon3o(i).debug;
  end generate;
```

The multiple LEON3 processors are instantiated using a generate statement. Note that the AHB index generic is incremented with the generate statement. Note also that the complete AHB slave input is fed to the processor, to allow for cache snooping.

```
dcomgen : if dbg = 1 generate
  dsu0 : dsu      -- LEON3 Debug Support Unit
    generic map (hindex => 2, ncpu => ncpu, tech => memtech, kbytes => 2)
    port map (rstn, clk, ahbmi, ahbsi, ahbso(2), l3dbg_o, l3dbg_i, dsui, dsuo);

  dsui.enable <= dsuen;
  dsui.break  <= dsubre;
  dsuact      <= dsuo.active;

  dcom0 : ahbuart      -- Debug UART
    generic map (ahbndx => NCPU, pindex => 7, paddr => 7)
    port map (rstn, clk, dui, duo, apbi, apbo(7), ahbmi, ahbmo(NCPU));

  dui.rxd <= dsurx;
  dsutx <= duo.txd;
end generate;
```

There is only one debug support unit (DSU) in the design, supporting multiple LEON3 processors.

```
irqctrl0 : irqmp -- interrupt controller
  generic map (pindex => 2, paddr => 2, ncpu => NCPU)
  port map (rstn, clk, apbi, apbo(2), irqi, irqo);
```

There is also only one interrupt controller, supporting multiple LEON3 processors.

To prepare the design for simulation with ModelSim, move to the grlib/designs/leon3mp directory and execute the 'make vsim' command.

```
$ make vsim
```

To simulate the default design execute the 'vsim' command.

```
$ vsim -c leon3mp
```

```
Simulate the first 100 ns by writing 'run'.
# LEON3 Demonstration design
# GRLIB Version 0.10
# Target technology: virtex , memory library: virtex
# ahbctrl: AHB arbiter/multiplexer rev 1
# ahbctrl: Common I/O area at 0xffff0000, 1 Mbyte
# ahbctrl: Configuration area at 0xfffff000, 4 kbyte
# ahbctrl: mst0: Gaisler Research      Leon3 SPARC V8 Processor
# ahbctrl: mst1: Gaisler Research      AHB Debug UART
# ahbctrl: slv0: European Space Agency Leon2 Memory Controller
# ahbctrl:      memory at 0x00000000, size 512 Mbyte, cacheable, prefetch
# ahbctrl:      memory at 0x20000000, size 512 Mbyte
# ahbctrl:      memory at 0x40000000, size 1024 Mbyte, cacheable, prefetch
# ahbctrl: slv1: Gaisler Research      AHB/APB Bridge
# ahbctrl:      memory at 0x80000000, size 16 Mbyte
# ahbctrl: slv2: Gaisler Research      Leon3 Debug Support Unit
# ahbctrl:      memory at 0x90000000, size 256 Mbyte
# ahbctrl: slv6: Gaisler Research      AMBA Trace Buffer
# ahbctrl:      I/O port at 0xffff4000, size 128kbyte
# apbmst: APB Bridge at 0x80000000 rev 1
# apbmst: slv0: European Space Agency Leon2 Memory Controller
# apbmst:      I/O ports at 0x80000000, size 256 byte
# apbmst: slv1: Gaisler Research      Generic UART
# apbmst:      I/O ports at 0x80000100, size 256 byte
# apbmst: slv2: Gaisler Research      Multi-processor Interrupt Ctrl.
# apbmst:      I/O ports at 0x80000200, size 256 byte
# apbmst: slv3: Gaisler Research      Modular Timer Unit
# apbmst:      I/O ports at 0x80000300, size 256 byte
# apbmst: slv7: Gaisler Research      AHB Debug UART
# apbmst:      I/O ports at 0x80000700, size 256 byte
# ahbtrace6: AHB Trace Buffer, 2 kbytes
# gptimer3: GR Timer Unit rev 0, 16-bit scaler, 2 32-bit timers, irq 8
# apbictrl: Multi-processor Interrupt Controller rev 1, #cpu 1
# apbuart1: Generic UART rev 1, irq 2
# ahbuart7: AHB Debug UART rev 0
# dsu2: LEON3 Debug support unit + AHB Trace Buffer, 2 kbytes
# leon3_0: LEON3 SPARC V8 processor rev 0
# leon3_0: icache 1*1 kbyte, dcache 1*1 kbyte
```

## 7 Using netlists

### 7.1 Introduction

GRLIB supports the usage of mapped netlists in the implementation flow. The netlists can be included in the flow at two different points; during synthesis or during place&route. The netlists can have two basic formats: mapped VHDL (.vhd) or a technology-specific netlist format (.ngo, .vqm, .edf). The sections below outline how the different formats are handled.

### 7.2 Mapped VHDL

A core provided in mapped VHDL format is included during synthesis, and treated the same as any RTL VHDL code. To use such netlist, the core must be configured to incorporate the netlist rather than the RTL VHDL code. This can be done in the xconfig configuration menu, or by setting the 'netlist' generic on core. The benefit of VHDL netlists is that the core (and whole design) can be simulated and verified without special simulation libraries.

The following Gaisler cores support the VHDL netlist format: GRFPU, GRFPU-Lite, GRSPW, LEON3FT. The netlists are available for the following technologies:

- Xilinx: GRFPU, GRFPU-Lite, GRSPW, GRUSBHC
- Actel: GRSPW, LEON3FT
- Altera: GRFPU, GRFPU-lite

The Gaisler netlists have the following default configurations:

- GRFPU-Lite: simple FPC controller
- GRSPW: 16 word AHB FIFO, 16 byte TX/RX FIFO, no RMAP support, RMAP CRC support enabled. FIFO protection can be enabled/disabled through xconfig.
- LEON3FT: 8 reg windows, SVT, no hardware MUL/DIV, 2 watch-points, power-down enabled, 8 Kbyte icache with 32 bytes/line, 4 Kbyte dcache with 16/byte/line, GRFPU-Lite enabled
- GRUSBHC: One port configuration with only universal controller, one port with only enhanced controller, one port with both controllers, and two ports with both controllers. All other generics are set to their default values (see GRUSBHC section of *grip.pdf*).

Contact Aeroflex Gaisler if other settings or technologies for the netlists are required. Netlists can also be delivered for other cores, such as GRETH\_GBIT.

Note that when implementing Xilinx systems, the standard (non-FT) LEON3 core is always used, even if LEON3FT is selected in xconfig. This allows the user to change the parameters to the core since the standard version of LEON3 is provided in source code.

### 7.3 Xilinx netlist files

To use Xilinx netlist files (.ngo or .edf), the netlist should be placed in the 'netlists/xilinx/tech' directories. During place&route, the ISE mapper will look in this location and replace and black-boxes in the design with the corresponding netlist. Note that when using .ngo or .edf files, the 'netlist' generic on the cores should NOT be set.

A special case exists for GRFPU and GRFPU-lite netlists. In GRLIB distributions that lack FPU source code, the netlist version of the selected FPU core will always be instantiated. When the design is simulated a VHDL netlist will be used (if available) and when the design is synthesized an EDIF netlist will be used. This is done in order to speed up synthesis. Parsing and performing synthesis on VHDL netlists is time consuming and using an EDIF netlist instead decreases the time required to run the tools.

Some tool versions have bugs that prevent them from using EDIF netlists. In order to work around such issues, convert the EDIF netlist to a .ngo netlist using the *edif2ngd* application in the ISE suite. After a netlist has been converted to .ngo format the EDIF version can be removed from the library.

## 7.4 Altera netlists

To use Altera netlist files (.vqm), the netlist should be placed in the 'netlists/altera/tech' directories, or in the current design directory. During place&route, the Altera mapper will look in these location and replace and black-boxes in the design with the corresponding netlist. Note that when using .vqm files, the 'netlist' generic on the cores should NOT be set.

A special case exists for GRFPU and GRFPU-lite netlists. In GRLIB distributions that lack FPU source code, the netlist version of the selected FPU core will always be instantiated. When the design is simulated a VHDL netlist will be used (if available) and when the design is synthesized a .vqm netlist will be used. This is done in order to speed up synthesis and due to the synthesis tools not always being able to handle VHDL netlists correctly.

## 7.5 Known limitations

Some tool versions have bugs that prevent them from using EDIF netlists. In order to work around such issues, convert the EDIF netlist to a .ngo netlist using the *edif2ngd* application in the ISE suite. After a netlist has been converted to .ngo format the EDIF version can be removed from the library

When synthesizing with Xilinx XST, the tool can crash when the VHDL netlist of GRFPU is used. This is not an issue with recent GRLIB versions since the VHDL netlists are currently only used for simulation.

## 8 Extending GRLIB

### 8.1 Introduction

GRLIB consists of a number of VHDL libraries, each one providing a specific set of interfaces or IP cores. The libraries are used to group IP cores according to the vendor, or to provide shared data structures and functions. Extension of GRLIB can be done by adding cores to an existing library, adding a new library and associated cores/packages, adding portability support for a new target technology, adding support for a new simulator or synthesis tool, or adding a board support package for a new FPGA board.

### 8.2 GRLIB organisation

The automatic generation of compile scripts searches for VHDL libraries in the file `lib/libs.txt`, and in `lib/*/libs.txt`. The `libs.txt` files contains paths to directories containing IP cores to be compiled into the same VHDL library. The name of the VHDL library is the same as the directory. The main `libs.txt` (`lib/libs.txt`) provides mappings to libraries that are always present in GRLIB, or which depend on a specific compile order (the libraries are compiled in the order they appear in `libs.txt`):

```
$ cat lib/libs.txt
grlib
tech/atc18
tech/apa
tech/unisim
tech/virage
fpu
gaisler
esa
opencores
```

Relative paths are allowed as entries in the `libs.txt` files. The path depth is unlimited. The leaf of each path corresponds to a VHDL library name (e.g. 'grlib' and 'unisim').

Each directory specified in the `libs.txt` contains the file `dirs.txt`, which contains paths to sub-directories containing the actual VHDL code. In each of the sub-directories appearing in `dirs.txt` should contain the files `vhdlsyn.txt` and `vhdlsim.txt`. The file `vhdlsyn.txt` contains the names of the files which should be compiled for synthesis (and simulation), while `vhdlsim.txt` contains the name of the files which only should be used for simulation. The files are compiled in the order they appear, with the files in `vhdlsyn.txt` compiled before the files in `vhdlsim.txt`.

The example below shows how the AMBA package in the GRLIB VHDL library is constructed:

```
$ ls lib/grlib
amba/  dirs.txt  modgen/  sparc/  stdlib/  tech/  util/

$ cat lib/grlib/dirs.txt
stdlib util sparc modgen amba tech

$ ls lib/grlib/amba
ahbctrl.vhd amba.vhd  apbctrl.vhd vhdlsyn.txt

$ cat grlib/lib/grlib/amba/vhdlsyn.txt
amba.vhd apbctrl.vhd ahbctrl.vhd
```

The libraries listed in the `grlib/lib/libs.txt` file are scanned first, and the VHDL files are added to the automatically generated compile scripts. Then all sub-directories in `lib` are scanned for additional `libs.txt` files, which are then also scanned for VHDL files. It is therefore possible to add a VHDL library (= sub-directory to `lib`) without having to edit `lib/libs.txt`, just by inserting into `lib`.

When all `libs.txt` files have been scanned, the `dirs.txt` file in `lib/work` is scanned and any cores in the VHDL work library are added to the compile scripts. The work directory must be treated last to avoid circular references between work and other libraries. The work directory is always scanned as does not appear in `lib/libs.txt`.

## 8.3 Adding an AMBA IP core to GRLIB

### 8.3.1 Example of adding an existing AMBA AHB slave IP core

An IP core with AMBA interfaces can be easily adapted to fit into GRLIB. If the AMBA signals are declared as standard IEEE-1164 signals, then it is simple a matter of assigning the IEEE-1164 signal to the corresponding field of the AMBA record types declared in GRLIB, and to define the plug&play configuration information, as shown in the example hereafter.

The plug&play configuration utilizes the constants and functions declared in the GRLIB AMBA ‘types’ package, and the HADDR and HMASK generics.

Below is the resulting entity for the adapted component:

```
library ieee; use ieee.std_logic_1164.all;
library grlib; use grlib.amba.all;

entity ahb_example is
  generic (
    hindex : integer := 0;
    haddr : integer := 0;
    hmask : integer := 16#fff#);
  port (
    rst : in std_ulogic;
    clk : in std_ulogic;
    ahbsi : in ahb_slv_in_type;
    ahbso : out ahb_slv_out_type);
end;

architecture rtl of ahb_example is
  -- component to be interfaced to GRLIB
  component ieee_example
    port (
      rst : in std_ulogic;
      clk : in std_ulogic;
      hsel : in std_ulogic;
      haddr : in std_logic_vector(31 downto 0);
      hwrite : in std_ulogic;
      htrans : in std_logic_vector(1 downto 0);
      hsize : in std_logic_vector(2 downto 0);
      hburst : in std_logic_vector(2 downto 0);
      hwdata : in std_logic_vector(31 downto 0);
      hprot : in std_logic_vector(3 downto 0);
      hreadyi : in std_ulogic;
      hmaster : in std_logic_vector(3 downto 0);
      hmastlock : in std_ulogic;
      hreadyo : out std_ulogic;
      hresp : out std_logic_vector(1 downto 0);
      hrdata : out std_logic_vector(31 downto 0);
      hsplite : out std_logic_vector(15 downto 0));
  end component;

  -- plug&play configuration
  constant HCONFIG: ahb_config_type := (
    0 => ahb_device_reg (VENDOR_EXAMPLE, EXAMPLE_AHBRAM, 0, 0, 0),
    4 => ahb_membar(memaddr, '0', '0', memmask), others => X"00000000");

begin
  ahbso.hconfig <= HCONFIG; -- Plug&play configuration
  ahbso.hirq <= (others => '0'); -- No interrupt line used
  -- original component
  e0: ieee_example
    port map(
      rst, clk, ahbsi.hsel(ahbndx), ahbsi.haddr, ahbsi.hwrite, ahbsi.htrans, ahbsi.hsize,
      ahbsi.hburst, ahbsi.hwdata, ahbsi.hprot, ahbsi.hready, ahbsi.hmaster,
      ahbsi.hmastlock, ahbso.hready, ahbso.hresp, ahbso.hrdata, ahbso.hsplite);
end;
```

The files containing the entity *ahb\_example* the entity for *ieee\_example* should be added to GRLIB by listing the files in a *vhdsyn.txt* file located in a directory that will be scanned by the GRLIB scripts, as described in section 8.2. The paths in *vhdsyn.txt* can be relative, allowing the VHDL files to be placed outside the GRLIB tree. The entities and packages will be compiled into a library with the same name as the directory that holds the *vhdsyn.txt* file.

In the *ahb\_example* example, the core does not have the ability to assert an interrupt. In order to assert an interrupt, an AHB core must drive the *hirq* vector in the *ahb\_slv\_out\_type* (or *ahb\_mst\_out\_type*) output record. If the core is an APB slave, it should drive the *apb\_slv\_out\_type* record's *pirq* vector. Position *n* of *hirq*/*pirq* corresponds to interrupt line *n*. All unused interrupt lines must be driven to '0'.

### 8.3.2 AHB Plug&play configuration

As described in section 5.3, the configuration record from each AHB unit is sent to the AHB bus controller via the *HCONFIG* signal. From this information, the bus controller automatically creates the read-only plug&play area.

In the *ahb\_example* example in the previous section, the plug&play configuration is held in the constant *HCONFIG*, which is assigned to the output *ahbso.hconfig*. The constant is created with:

```
-- plug&play configuration
constant HCONFIG : ahb_config_type := (
  0    => ahb_device_reg (VENDOR_EXAMPLE, EXAMPLE_AHBRAM, 0, 0, 0),
  4    => ahb_membar(memaddr, '0', '0', memmask), others => X"00000000");
```

The *ahb\_config\_type* is an array of 32-bit vectors. Each position in this array corresponds to the same word in the core's plug&play information. Section 5.3.1 describes the plug&play information in the following way: The first word is called the identification register and contains information on the device type and interrupt routing. The last four words are called bank address registers, and contain address mapping information for AHB slaves. The remaining three words are currently not assigned and could be used to provide core-specific configuration information.

The AMBA package (*lib/grlib/amba/amba.vhd*) in GRLIB provides functions that help users create proper plug&play information. Two of these functions are used above. The *ahb\_device\_reg* function creates the identification register value for an AHB slave or master:

```
ahb_device_reg (vendor, device, cfgver, version, interrupt)
```

The parameters are explained in the table below:

**TABLE 36. ahb\_device\_reg parameters**

Parameter	Comments
vendor	Integer Vendor ID. Typically defined in <i>lib/grlib/amba/devices.vhd</i> . It is recommended that new cores be added under a new vendor ID or under the contrib vendor ID.
device	Integer Device ID. Typically defined in <i>lib/grlib/amba/devices.vhd</i> . The combination of vendor and device ID must not match any existing core as this may lead to your IP core being initialized by drivers for another core.
cfgver	Plug&play information version, only supported value is 0.
version	Core version/revision. Assigned to 5-bit wide field in plug&plat information.
interrupt	Set this value to the first interrupt line that the core drives. Set to 0 if core does not make use of interrupts.

If an IP core only has an AHB master interface, the only position in *HCONFIG* that needs to be specified is the first word:

```
constant hconfig : ahb_config_type := (
  0 => ahb_device_reg (venid, devid, 0, version, 0),
  others => X"00000000");
```

If an IP core has an AHB slave interface, as in the *ahb\_example* example, we also need to specify the memory area(s) that the slave will map. Again, the *HCONFIG* constant from *ahb\_example* is:

```
-- plug&play configuration
constant HCONFIG : ahb_config_type := (
  0    => ahb_device_reg (VENDOR_EXAMPLE, EXAMPLE_AHBRAM, 0, 0, 0),
  4    => ahb_membar(memaddr, '0', '0', memmask), others => X"00000000");
```

The last four words of *ahb\_config\_type* (positions 4 - 7) are called bank address registers (BARs), and contain memory map information. This information determines address decoding in the AHB controller (AHBCTRL core). Address decoding is described in detail under section 5.3.3. When creating an AHB memory bank, the *ahb\_membar* function can be used to automatically generate the correct layout for a BAR:

```
ahb_membar(memaddr, prefetch, cache, memmask)
```

To create an AHB I/O bank, the *ahb\_iobar* function can be used:

```
ahb_iobar(memaddr, memmask)
```

The parameters of these functions are described in the table below:

**TABLE 37. ahb\_membar/ahb\_iobar parameters**

Parameter	Comments
memaddr	Integer value propagated to BAR.ADDR
memmask	Integer value propagated to BAR.MASK
prefetch	Std_Logic value propagated to prefetchable field (P) in bank address register. Only applicable for AHB memory bars ( <i>ahb_membar</i> function).
cache	Std_Logic value propagated to cacheable field (C) in bank address register. Only applicable for AHB memory bars ( <i>ahb_membar</i> function).

An AHB slave can map up to four address areas (it has four bank address registers). Typically, an IP core has one AHB I/O bank with registers and zero or several AHB memory banks that map a larger memory area. One example is the GRLIB DDR2 controller (DDR2SPA) that has the following HCONFIG:

```
constant hconfig : ahb_config_type := (
  0 => ahb_device_reg ( VENDOR_GAISLER, GAISLER_DDR2SP, 0, REVISION, 0),
  4 => ahb_membar(haddr, '1', '1', hmask),
  5 => ahb_iobar(ioaddr, iomask),
  others => zero32);
```

Position four, the first bank address register, defines an AHB memory bank which maps external DDR2 SDRAM memory. Position five, the second bank address register, defines an AHB I/O bank that holds the memory controller's register interface. On this core, the *haddr*, *hmask*, *ioaddr* and *iomask* values are set via VHDL generics.

For IP cores that map multiple memory areas, there is no need for the IP core to decode the address in order to determine which bank that is accessed. The AHB controller decodes the incoming address and selects the correct AHB slave via the HSEL vector. The AHB controller also indicates which bank that is being accessed via the HMBSEL vector, when bank *n* is accessed HMBSEL(*n*) will be asserted.

### 8.3.3 Example of creating an APB slave IP core

The next page contains an APB slave example core. The IP core has one memory mapped 32-bit register that will be reset to zero. The register can be read or written from register address offset 0. The core's base address, mask and bus index settings are configurable via VHDL generics (*pindex*, *paddr*, *pmask*). The *paddr* and *pmask* VHDL generics are propagated to the APB bridge via the *apbo.pconfig* signal and the index is propagated via the *apbo.pindex* signal. These values are then used by the APB bridge to generate the APB address decode and slave select logic.

Example of APB slave IP core with one 32-bit register that can be read and written:

```
library ieee; use ieee.std_logic_1164.all;
library grlib; use grlib.amba.all; use grlib.devices.all;
library gaisler; use gaisler.misc.all;

entity apb_example is
  generic (
    pindex : integer := 0;
    paddr   : integer := 0;
    pmask   : integer := 16#fff#);
  port (
    rst     : in  std_ulogic;
    clk     : in  std_ulogic;
    apbi    : in  apb_slv_in_type;
    apbo    : out apb_slv_out_type);
end;

architecture rtl of apb_example is

  constant REVISION : integer := 0;

  constant PCONFIG : apb_config_type := (
    0 => ahb_device_reg (VENDOR_ID, DEVICE_ID, 0, REVISION, 0),
    1 => apb_iobar(paddr, pmask));

  type registers is record
```



```

    reg : std_logic_vector(31 downto 0);
end record;

signal r, rin : registers;

begin

    comb : process(rst, r, apbi)
        variable readdata : std_logic_vector(31 downto 0);
        variable v        : registers;
    begin
        v := r;

        -- read register
        readdata := (others => '0');
        case apbi.paddr(4 downto 2) is
            when "000" => readdata := r.reg(31 downto 0);
            when others => null;
        end case;

        -- write registers
        if (apbi.psel(pindex) and apbi.penable and apbi.pwrite) = '1' then
            case apbi.paddr(4 downto 2) is
                when "000" => v.reg := apbi.pwdata;
                when others => null;
            end case;
        end if;

        -- system reset
        if rst = '0' then v.reg := (others => '0'); end if;

        rin <= v;
        apbo.prdata <= readdata; -- drive apb read bus
    end process;

    apbo.pirq <= (others => '0');          -- No IRQ
    apbo.pindex <= pindex;                -- VHDL generic
    apbo.pconfig <= PCONFIG;              -- Config constant

-- registers
regs : process(clk)
begin
    if rising_edge(clk) then r <= rin; end if;
end process;

-- boot message

-- pragma translate_off
bootmsg : report_version
    generic map ("apb_example" & tost(pindex) & ": Example core rev " & tost(REVISION));
-- pragma translate_on

end;
```

The steps required to instantiate the *apb\_example* IP core in a system are:

- Add the file to a directory covered by the GRLIB scripts (via *libs.txt* and *dirs.txt*)
- Add the file to *vhdlsyn.txt* in the current directory
- Modify the example to use a unique vendor and device ID (see creation of PCONFIG constant)
- Create a component for the *apb\_example* core in a package that is also synthesized.
- Include the package in your design top-level
- Instantiate the component in your design top-level

For a complete example, see the General Purpose Register (GRGPREG) IP core located in *lib/gaisler/misc/grgpreg.vhd*. That core is very similar to the example given in this section. The GRGPREG core has a component declaration in the *glib.misc* package located at *lib/gaisler/misc/misc.vhd*. Note that both of these files are listed in the *vhdlsyn.txt* file located in the same directory.

### 8.3.4 APB plug&play configuration

APB slave plug&play configuration is propagated via the *apb\_slv\_out\_type* record's *pconfig* member. The configuration is very similar to that of an AHB slave. The main difference is that APB slaves only have one type of BAR and each APB slave only has one bank. The creation of the PCONFIG array in the previous section looked like:

```

constant PCONFIG : apb_config_type := (
    0 => ahb_device_reg (VENDOR_ID, DEVICE_ID, 0, REVISION, 0),
    1 => apb_iobar(paddr, pmask));
```

The *ahb\_device\_reg* function has been described in section 8.3.2. The *apb\_iobar* function takes the same arguments as the *ahb\_iobar* function, also described in section 8.3.2.

## 8.4 Using verilog code

Verilog does not have the notion of libraries, and although some CAD tools supports the compilation of verilog code into separate libraries, this feature is not provided in all tools. Most CAD tools however support mixing of verilog and VHDL, and it is therefore possible to add verilog code to the work library. Adding verilog files is done in the same way as VHDL files, except that the verilog file names should appear in *vlogsyn.txt* and *vlogsim.txt*.

The basic steps for adding a synthesizable verilog core are:

- Create a directory and add it to *libs.txt* and *dirs.txt* as described in section 8.2, or use an existing directory.
- List the verilog files in a *vlogsyn.txt* file located in the selected directory
- Create a VHDL component declaration for the verilog top-level

In case the verilog IP core will be instantiated directly in the design, the component can be added to a package. This package can then be referenced in the design's top-level and the verilog core can be instantiated using the VHDL component.

In case the verilog IP core has an AMBA interface, it will likely require wrapping in order to add the GRLIB AMBA plug&play signals. To do this, the procedure described in section 8.3.1 can be used, where the *ieee\_example* component declaration would be the VHDL component for the verilog IP core.

As mentioned above, all CAD tools may not support compiling verilog code into a library. Should the strategy above not work, another option is to list the verilog files in the *VERILOGSYNFILES* variable defined in the (template) design's Makefile and to create the VHDL component of the verilog IP core in the design's top-level.

Other issues that may arise include propagation problems of VHDL generics to Verilog parameters (issues crossing the language barrier). Many tools handle propagation of integer and string values correctly. Should there be any problems, it is recommended to change the Verilog code to remove the parameters.

## 8.5 Adding portability support for new target technologies

### 8.5.1 General

New technologies to support portability can be added to GRLIB without the need to modify any previously developed designs. This is achieved by technology independent encapsulation of components such as memories, pads and clock buffers. The technology mapping is organized as follows:

- A VHDL library with the technology simulation models is placed in *lib/tech/library*
- Wrappers for memory, pads, PLL and other cells are placed under *lib/techmap/library*
- All 'virtual' components with technology mapping are placed in *lib/techmap/maps*
- Declaration of all 'virtual' components and technologies is made in *lib/techmap/gencomp/gencomp.vhd*

An entity that uses a technology independent component needs only to make the *techmap.gencomp* package visible, and can then instantiate any of the mapped components.

### 8.5.2 Adding a new technology

A new technology is added in four steps. First, a VHDL library is created in the *lib/tech/library* location. Secondly, a package containing all technology specific component declarations is created and the source code file name is added to the '*vhdlsyn.txt*' or '*vlogsyn.txt*' file. Third, simulation models are created for all the components and the source file names are added to the '*vhdlsim.txt*' or '*vlogsim.txt*' file. A technology constant is added to the GENCOMP package defined in the TECHMAP library. The library name is not put in *lib/libs.txt* but added either to the FPGALIBS or ASICLIBS in *bin/Makfile*.

The technology library part is completed and the components need to be encapsulated as described in the next section. As an example, the ASIC memories from Virage are defined in the VIRAGE library, located in the *lib/virage* directory. The component declarations are defined in the VCOMPONENTS

package in the `virage_vcomponents.vhd` file. The simulation models are defined in `virage_simpri.ms.vhd`.

### 8.5.3 Encapsulation

Memories, pads and clock buffers used in GRLIB are defined in the TECHMAP library. The encapsulation of technology specific components is done in two levels.

The lower level handles the technology dependent interfacing to the specific memory cells or macro cells. This lower level is implemented separately for each technology as described hereafter.

For each general type of memory, pad or clock buffer, an entity/architecture is created at the lower level. The entity declarations are technology independent and have similar interfaces with only minor functional variations between technologies. The architectures are used for instantiating, configuring and interfacing the memory cells or macro cells defined for the technology.

A package is created for each component type containing component declarations for the aforementioned entities. Currently there is a separate memory, pad and clock buffer package for each technology. The components in these packages are only used in the higher level, never directly in the designs or IP cores.

The higher level defines a technology independent interface to the memory, pad or clock buffer. This higher level is implemented only once and is common to all technologies.

For each general type of memory, pad or clock buffer, an entity/architecture is created at the higher level. The entity declarations are technology independent. The architectures are used for selecting the relevant lower level component depending on the value of the `tech` and `memtech` generics.

A package is created for each component type containing component declarations for the aforementioned entities. Currently there is a separate memory, pad and clock buffer package. The components declared in these packages are used in the designs or by other IP cores. The two level approach allows each technology to be maintained independently of other technologies.

### 8.5.4 Memories

The currently defined memory types are single-port, dual-port, two-port and triple-port synchronous RAM. The encapsulation method described in the preceding section is applied to include a technology implementing one of these memory types.

For example, the ASIC memory models from Virage are encapsulated at the lower level in the `lib/tech/techmap/virage/mem_virage_gen.vhd` file. Specifically, the single-port RAM is defined in the `VIRAGE_SYNCRAM` entity:

```
entity virage_syncram is
  generic (
    abits   : integer := 10;
    dbits   : integer := 8 );
  port (
    clk      : in  std_ulogic;
    address  : in  std_logic_vector(abits -1 downto 0);
    datain   : in  std_logic_vector(dbits -1 downto 0);
    dataout  : out std_logic_vector(dbits -1 downto 0);
    enable   : in  std_ulogic;
    write    : in  std_ulogic);
end;
```

The corresponding architecture instantiates the Virage specific technology specific memory cell, e.g. `hdss1_256x32cm4sw0` shown hereafter:

```
architecture rtl of virage_syncram is
  signal d, q, gnd : std_logic_vector(35 downto 0);
  signal a : std_logic_vector(17 downto 0);
  signal vcc : std_ulogic;
  constant synopsys_bug : std_logic_vector(37 downto 0) := (others => '0');
begin

  gnd <= (others => '0'); vcc <= '1';
  a(abits -1 downto 0) <= address;
  d(dbits -1 downto 0) <= datain(dbits -1 downto 0);
  a(17 downto abits) <= synopsys_bug(17 downto abits);
  d(35 downto dbits) <= synopsys_bug(35 downto dbits);
  dataout <= q(dbits -1 downto 0);
  q(35 downto dbits) <= synopsys_bug(35 downto dbits);

  a8d32 : if (abits = 8) and (dbits <= 32) generate
    id0 : hdss1_256x32cm4sw0
      port map (a(7 downto 0), gnd(7 downto 0), clk,
```

```

        d(31 downto 0), gnd(31 downto 0), q(31 downto 0),
        enable, vcc, write, gnd(0), gnd(0), gnd(0), gnd(0), gnd(0));
    end generate;
    ...
end rtl;

```

The `lib/tech/techmap/virage/mem_virage.vhd` file contains the corresponding component declarations in the MEM\_VIRAGE package.

```

package mem_virage is
  component virage_syncram
    generic (
      abits    : integer := 10;
      dbits    : integer := 8 );
    port (
      clk       : in  std_ulogic;
      address   : in  std_logic_vector(abits -1 downto 0);
      datain    : in  std_logic_vector(dbits -1 downto 0);
      dataout   : out std_logic_vector(dbits -1 downto 0);
      enable    : in  std_ulogic;
      write     : in  std_ulogic);
    end component;
  ...
end;

```

The higher level single-port RAM model SYNCRAM is defined in the `lib/gaisler/maps/syncram.vhd` file. The entity declaration is technology independent:

```

entity syncram is
  generic (
    tech    : integer := 0;
    abits   : integer := 6;
    dbits   : integer := 8 );
  port (
    clk      : in  std_ulogic;
    address  : in  std_logic_vector((abits -1) downto 0);
    datain   : in  std_logic_vector((dbits -1) downto 0);
    dataout  : out std_logic_vector((dbits -1) downto 0);
    enable   : in  std_ulogic;
    write    : in  std_ulogic);
end;

```

The corresponding architecture implements the selection of the lower level components based on the MEMTECH or TECH generic:

```

architecture rtl of syncram is
begin
  inf : if tech = infered generate
    u0 : generic_syncram generic map (abits, dbits)
      port map (clk, address, datain, dataout, write);
    end generate;
  ...
  vir : if tech = memvirage generate
    u0 : virage_syncram generic map (abits, dbits)
      port map (clk, address, datain, dataout, enable, write);
    end generate;
  ...
end;

```

The `lib/tech/techmap/gencomp/gencomp.vhd` file contains the corresponding component declaration in the GENCOMP package:

```

package gencomp is
  component syncram
    generic (
      tech    : integer := 0;
      abits   : integer := 6;
      dbits   : integer := 8);
    port (
      clk      : in  std_ulogic;
      address  : in  std_logic_vector((abits -1) downto 0);
      datain   : in  std_logic_vector((dbits -1) downto 0);
      dataout  : out std_logic_vector((dbits -1) downto 0);
      enable   : in  std_ulogic;
      write    : in  std_ulogic);
    end component;
  ...
end;

```

The GENCOMP package contains component declarations for all portable components, i.e. SYNCRAM, SYNCRAM\_DP, SYNCRAM\_2P and REGFILE\_3P.

### 8.5.5 Pads

The currently defined pad types are in-pad, out-pad, open-drain out-pad, I/O-pad, open-drain I/O pad, tri-state output-pad and open-drain tri-state output-pad. Each pad type comes in a discrete and a vectorized version.

The encapsulation method described in the preceding sections is applied to include a technology implementing these pad types.

The file structure is similar to the one used in the memory example above. The pad related files are located in `grib/lib/tech/techmap/maps`. The `grib/lib/tech/techmap/gen-comp/gencomp.vhd` file contains the component declarations in the GENCOMP package.

### 8.5.6 Clock generators

There is currently only one defined clock generator types named CLKGEN.

The encapsulation method described in the preceding sections is applied to include a technology implementing clock generators and buffers.

The file structure is similar to the one used in the memory example above. The clock generator related files are located in `grib/lib/tech/techmap/maps`. The CLKGEN component is declared in the GENCOMP package.

## 8.6 Extending the xconfig GUI configuration

### 8.6.1 Introduction

Each template design has a simple graphical configuration interface that can be started by issuing *make xconfig* in the template design directory. The tool presents the user with configuration options and generates the file *config.vhd* that contains configuration constants used in the design.

The subsections below describe how to create configuration menus for a core and then how to include these new options in xconfig for an existing template design.

### 8.6.2 IP core xconfig files

Each core has a set of files that are used to generate the core's xconfig menu entries. As an example we will look at the GRGPIO core's menu. The xconfig files are typically located in the same directory as the core's HDL files (but this is not a requirement). For the GRGPIO core the xconfig files are:

```
$ ls lib/gaisler/misc/grgpio.in.*
lib/gaisler/misc/grgpio.in
lib/gaisler/misc/grgpio.in.h
lib/gaisler/misc/grgpio.in.help
lib/gaisler/misc/grgpio.in.vhd
```

We will start by looking at the *grgpio.in* file. This file defines the menu structure and options for the GRGPIO core:

```
bool 'Enable generic GPIO port' CONFIG_GRGPIO_ENABLE
if [ "$CONFIG_GRGPIO_ENABLE" = "y" ]; then
  int 'GPIO width' CONFIG_GRGPIO_WIDTH 8
  hex 'GPIO interrupt mask' CONFIG_GRGPIO_IMASK 0000
fi
```

The first line defines a boolean option that will be saved in the variable *CONFIG\_GRGPIO\_ENABLE*. This will be rendered as a yes/no question in the menu. If this constant is set to yes ('y') then the user will be able to select two more configuration options. First the width, which is defined as an integer (int), and the interrupt mask which is defined as a hexadecimal value (hex).

The GUI has a help option for each item in the menu. When a user clicks on the help button a help text can be optionally displayed. The contents of the help text boxes is defined in the file that ends with *.in.help*, in this case *grgpio.in.help*:

```
GPIO port
CONFIG_GRGPIO_ENABLE
Say Y here to enable a general purpose I/O port. The port can be
configured from 1 - 32 bits, with each port signal individually
programmable as input or output. The port signals can also serve
```

```

    as interrupt inputs.

GPIO port width
CONFIG_GRGPIO_WIDTH
    Number of bits in the I/O port. Must be in the range of 1 - 32.

GPIO interrupt mask
CONFIG_GRGPIO_IMASK
    The I/O port interrupt mask defines which bits in the I/O port
    should be able to create an interrupt.

```

As can be seen above, each help entry consists of a topic, the name of the variable used in the menu and the help text.

The two remaining files (*grgpio.in.h* and *grgpio.in.vhd*) are used when generating the *config.vhd* file for a design. *config.vhd* typically consists of a set of lines for each core where the first line decides if the core should be instantiated in the design and the following lines contain configuration options. For the GRGPIO core, the file *grgpio.in.vhd* defines that the following constants should be included in *config.vhd*:

```

-- GPIO port
constant CFG_GRGPIO_ENABLE : integer := CONFIG_GRGPIO_ENABLE;
constant CFG_GRGPIO_IMASK  : integer := 16#CONFIG_GRGPIO_IMASK#;
constant CFG_GRGPIO_WIDTH  : integer := CONFIG_GRGPIO_WIDTH;

```

In the listing above, we see a mix of VHDL and the constants defined in the menus (see listing for *grgpio.in* above). The value we select for *CONFIG\_GRGPIO\_ENABLE* will be assigned to the VHDL constant *CFG\_GRGPIO\_ENABLE*. In the menu we defined *CONFIG\_GRGPIO\_IMASK* as a hexadecimal value. The VHDL notation for this is to enclose the value in *16#..#* and this is done for the *CFG\_GRGPIO\_IMASK* constant.

When exiting the xconfig tool, the *.in.vhd* files for all cores will be concatenated into one file. Then a pre-processor will be used to replace all the variables defined in the menus (for instance *CONFIG\_GRGPIO\_ENABLE*) into the values they represent. In this process, additional information is inserted via the *.in.vhd* files. The contents of *grgpio.in.h* is:

```

#ifndef CONFIG_GRGPIO_ENABLE
#define CONFIG_GRGPIO_ENABLE 0
#endif
#ifndef CONFIG_GRGPIO_IMASK
#define CONFIG_GRGPIO_IMASK 0000
#endif
#ifndef CONFIG_GRGPIO_WIDTH
#define CONFIG_GRGPIO_WIDTH 1
#endif

```

This file is used to guarantee that the *CONFIG\_* variable always exist and are defined to sane values. If a user has disabled *CONFIG\_GRGPIO\_ENABLE* via the configuration menu, then this variable and all the other GRGPIO variables will be undefined. This would result in a *config.vhd* entry that looks like:

```

-- GPIO port
constant CFG_GRGPIO_ENABLE : integer := ;
constant CFG_GRGPIO_IMASK  : integer := 16##;
constant CFG_GRGPIO_WIDTH  : integer := ;

```

... and lead to errors during compilation. This is prevented by *grgpio.in.h* above, where all undefined variables are defined to sane values. It is also possible to place additional intelligence in the *.in.h* file where dependencies between variables can be expressed in ways that would be complicated in the menu definition in the *.in* file.

### 8.6.3 xconfig menu entries

The menu entries to include in xconfig is defined for each template design in the file *config.in*. As an example we will look at the *config.in* file for the design *leon3-gr-xc3s-1500*. In *designs/leon3-gr-xc3s-1500/config.in* we find the entry for the GRGPIO port (described in the previous section) as part of one of the submenus:

```

mainmenu_option next_comment
comment 'UART, timer, I/O port and interrupt controller'
source lib/gaisler/uart/uart1.in
if [ "$CONFIG_DSU_UART" != "y" ]; then
    source lib/gaisler/uart/uart2.in
fi

```

```

source lib/gaisler/leon3/irqmp.in
source lib/gaisler/misc/gptimer.in
source lib/gaisler/misc/grgpio.in
endmenu

```

These lines will create a submenu named *UART, timer, I/O port and interrupt controller* and under this submenu include the options for the two UART cores, interrupt controller, timer unit and GPIO port. When the *.in* file for a core is specified in *config.in*, the xconfig tool will automatically also use the corresponding *.in.h* and *.in.vhd* files when generating the *config.vhd* file.

#### 8.6.4 Adding new xconfig entries

In this section we will extend the menu in the *leon3-gr-xc3s-1500* design to include configuration options for one additional core. Note that adding xconfig entries does not include IP core HDL files in the list of files to be synthesized for a design. See section 8.3 for information on adding the HDL files of an IP core to GRLIB.

When we start, the *config.in* file for *leon3-gr-xc3s-1500* has the following contents around the inclusion of GRGPIO:

```

mainmenu_option next_comment
comment 'UART, timer, I/O port and interrupt controller'
source lib/gaisler/uart/uart1.in
if [ "$CONFIG_DSU_UART" != "y" ]; then
source lib/gaisler/uart/uart2.in
fi
source lib/gaisler/leon3/irqmp.in
source lib/gaisler/misc/gptimer.in
source lib/gaisler/misc/grgpio.in
endmenu

```

and the *config.vhd* file has the following entries (also just around the GRGPIO port):

```

-- GPIO port
constant CFG_GRGPIO_ENABLE : integer := 1;
constant CFG_GRGPIO_IMASK : integer := 16#0000#;
constant CFG_GRGPIO_WIDTH : integer := (8);

-- Spacewire interface
....

```

The core that we will add support for is the I2C2AHB core. We start by making copies of the existing configuration files for the GRGPIO core (described in section 8.6.2) and modify them for I2C2AHB. The resulting files are listed below:

*i2c2ahb.in*:

```

bool 'Enable I2C to AHB bridge ' CONFIG_I2C2AHB
if [ "$CONFIG_I2C2AHB" = "y" ]; then
bool 'Enable APB interface ' CONFIG_I2C2AHB_APB
hex 'AHB protection address (high) ' CONFIG_I2C2AHB_ADDRH 0000
hex 'AHB protection address (low) ' CONFIG_I2C2AHB_ADDRL 0000
hex 'AHB protection mask (high) ' CONFIG_I2C2AHB_MASKH 0000
hex 'AHB protection mask (low) ' CONFIG_I2C2AHB_MASKL 0000
bool 'Enable after reset ' CONFIG_I2C2AHB_APB
hex 'I2C memory address ' CONFIG_I2C2AHB_SADDR 50
hex 'I2C configuration address ' CONFIG_I2C2AHB_CADDR 51
fi

```

*i2c2ahb.in.help*:

```

GRLIB I2C2AHB core
CONFIG_I2C2AHB
Say Y here to enable I2C2AHB

CONFIG_I2C2AHB_APB
Say Y here to configure the core's APB interface

CONFIG_I2C2AHB_ADDRH
Defines address bits 31:16 of the core's AHB protection area

... and so on ..

```

*i2c2ahb.in.vhd*:

```

-- I2C to AHB bridge

```

```

constant CFG_I2C2AHB      : integer := CONFIG_I2C2AHB;
constant CFG_I2C2AHB_APB  : integer := CONFIG_I2C2AHB_APB;
constant CFG_I2C2AHB_ADDRH : integer := 16#CONFIG_I2C2AHB_ADDRH#;
constant CFG_I2C2AHB_ADDRL : integer := 16#CONFIG_I2C2AHB_ADDRL#;
constant CFG_I2C2AHB_MASKH : integer := 16#CONFIG_I2C2AHB_MASKH#;
constant CFG_I2C2AHB_MASKL : integer := 16#CONFIG_I2C2AHB_MASKL#;
constant CFG_I2C2AHB_RESEN : integer := CONFIG_I2C2AHB_RESEN;
constant CFG_I2C2AHB_SADDR : integer := 16#CONFIG_I2C2AHB_SADDR#;
constant CFG_I2C2AHB_CADDR : integer := 16#CONFIG_I2C2AHB_CADDR#;
constant CFG_I2C2AHB_FILTER : integer := CONFIG_I2C2AHB_FILTER;

```

*i2c2ahb.in.h*:

```

#ifndef CONFIG_I2C2AHB
#define CONFIG_I2C2AHB 0
#endif
#ifndef CONFIG_I2C2AHB_APB
#define CONFIG_I2C2AHB_APB 0
#endif
#ifndef CONFIG_I2C2AHB_ADDRH
#define CONFIG_I2C2AHB_ADDRH 0
#endif
#ifndef CONFIG_I2C2AHB_ADDRL
#define CONFIG_I2C2AHB_ADDRL 0
#endif
#ifndef CONFIG_I2C2AHB_MASKH
#define CONFIG_I2C2AHB_MASKH 0
#endif
#ifndef CONFIG_I2C2AHB_MASKL
#define CONFIG_I2C2AHB_MASKL 0
#endif
#ifndef CONFIG_I2C2AHB_RESEN
#define CONFIG_I2C2AHB_RESEN 0
#endif
#ifndef CONFIG_I2C2AHB_SADDR
#define CONFIG_I2C2AHB_SADDR 50
#endif
#ifndef CONFIG_I2C2AHB_CADDR
#define CONFIG_I2C2AHB_CADDR 51
#endif
#ifndef CONFIG_I2C2AHB_FILTER
#define CONFIG_I2C2AHB_FILTER 2
#endif

```

Once we have the above files in place, we will modify *designs/leon3-gr-emaxc3s-1500/config.in* so that I2C2AHB is also included. The resulting entries in *config.in* looks like:

```

mainmenu_option next_comment
comment 'UART, timer, I/O port and interrupt controller'
source lib/gaisler/uart/uart1.in
if [ "$CONFIG_DSU_UART" != "y" ]; then
source lib/gaisler/uart/uart2.in
fi
source lib/gaisler/leon3/irqmp.in
source lib/gaisler/misc/gptimer.in
source lib/gaisler/misc/grgpio.in
source lib/gaisler/misc/i2c2ahb.in
endmenu

```

Where the inclusion of *i2c2ahb.in* is made just before the *endmenu* statement.

We can now issue *make xconfig* in the template design directory to rebuild the graphical menu:

```

user@host:~/GRLIB/designs/leon3-gr-xc3s-1500$ make xconfig
make main.tk
make[1]: Entering directory `/home/user/GRLIB/designs/leon3-gr-xc3s-1500'
gcc -g -c ../../bin/tkconfig/tkparse.c
gcc -g -c ../../bin/tkconfig/tkcond.c
gcc -g -c ../../bin/tkconfig/tkgen.c
gcc -g tkparse.o tkcond.o tkgen.o -o tkparse.exe
./tkparse.exe config.in ../../ > main.tk
make[1]: Leaving directory `/home/user/GRLIB/designs/leon3-gr-xc3s-1500'
cat ../../bin/tkconfig/header.tk main.tk ../../bin/tkconfig/tail.tk > lconfig.tk
chmod a+x lconfig.tk

```

As can be seen from the output above, the change of *config.in* triggered a re-build of *tkparse.exe* and *lconfig.tk*. *tkparse.exe* is used to parse the *.in* files and *lconfig.tk* is what is executed when issuing *make xconfig*. In order to rebuild *tkparse.exe* the system must have a working copy of the GNU C compiler installed.

Under some circumstances the menus may not be rebuilt after *config.in* has been modified. If this happens try to issue *touch config.in* or remove the file *lconfig.tk*.



Now that the xconfig menus have been re-built we can check under *Peripherals > UART, timer, I/O port and interrupt controller* to see our newly added entries for the I2C2AHB core. Once we save and exit the xconfig tool a new *config.vhd* file will be generated that now also contains the constants defined in *i2c2ahb.in.vhd*:

```
-- GPIO port
constant CFG_GRGPIO_ENABLE : integer := 1;
constant CFG_GRGPIO_IMASK : integer := 16#0000#;
constant CFG_GRGPIO_WIDTH : integer := (8);

-- I2C to AHB bridge
constant CFG_I2C2AHB : integer := 0;
constant CFG_I2C2AHB_APB : integer := 0;
constant CFG_I2C2AHB_ADDRH : integer := 16#0#;
constant CFG_I2C2AHB_ADDRL : integer := 16#0#;
constant CFG_I2C2AHB_MASKH : integer := 16#0#;
constant CFG_I2C2AHB_MASKL : integer := 16#0#;
constant CFG_I2C2AHB_RESEN : integer := 0;
constant CFG_I2C2AHB_SADDR : integer := 16#50#;
constant CFG_I2C2AHB_CADDR : integer := 16#51#;
constant CFG_I2C2AHB_FILTER : integer := 2;

-- Spacewire interface
```

These constants can now be used in all files that include the *work.config* VHDL package.

### 8.6.5 Other uses and limitations

There is nothing IP core specific in xconfig. Local copies of configuration files (\*.in\*) can be created in the template design directory to create constants that are used to control other aspects of the design and not just IP core configuration.

The graphical interface provided by xconfig can ease configuration but the tool has several limitations that designers must be aware of:

1. When configuration options are saved and xconfig is exited, the *config.vhd* file is overwritten.
2. When a core is disabled, the present configuration is not restored when the core is re-enabled.
3. The tool does not provide a good solution for multiple instances of the same core.

The last item means that xconfig can not be used to configure two separate instances of the same core (unless the cores should have the exact same configuration, if this is the case the same set of *config.vhd* constants can be used in several instantiations). It is not possible to just include the same *.in* file several times in *config.in*. This will lead to constants with the same name being created in *config.vhd*. One option is to make a local copy of a core's configuration files (\*.in\*) and place them in the template design directory. The local copies can then be edited to have all their variable names changed (for instance by adding a 2 to the end of the variable names) and a reference to the local files can be added to *config.in*. This way a separate set of menu items, that will affect a separate set of constants in *config.vhd*, can be included.

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