

$$e^{st} \longrightarrow \frac{1}{1+SRC} e^{st}$$

$$\underline{s=j\omega}$$

$$e^{j\omega t} \longrightarrow \frac{1}{1+j\omega RC} e^{j\omega t}$$

$$= C\omega t + jS\sin\omega t$$

$$= \frac{1}{\sqrt{1+(\omega RC)^2}} e^{j(\omega t - \tan^{-1}\omega RC)} = \theta$$

$$C_s \omega t = \operatorname{Re}[e^{j\omega t}] \longrightarrow \frac{1}{\sqrt{1+(\omega RC)^2}} C_s (\omega t - \theta)$$

$$S \sin\omega t \longrightarrow \frac{1}{\sqrt{1+(\omega RC)^2}} \sin(\omega t - \theta)$$

Filters

↳ Low Pass Filter (LPF)

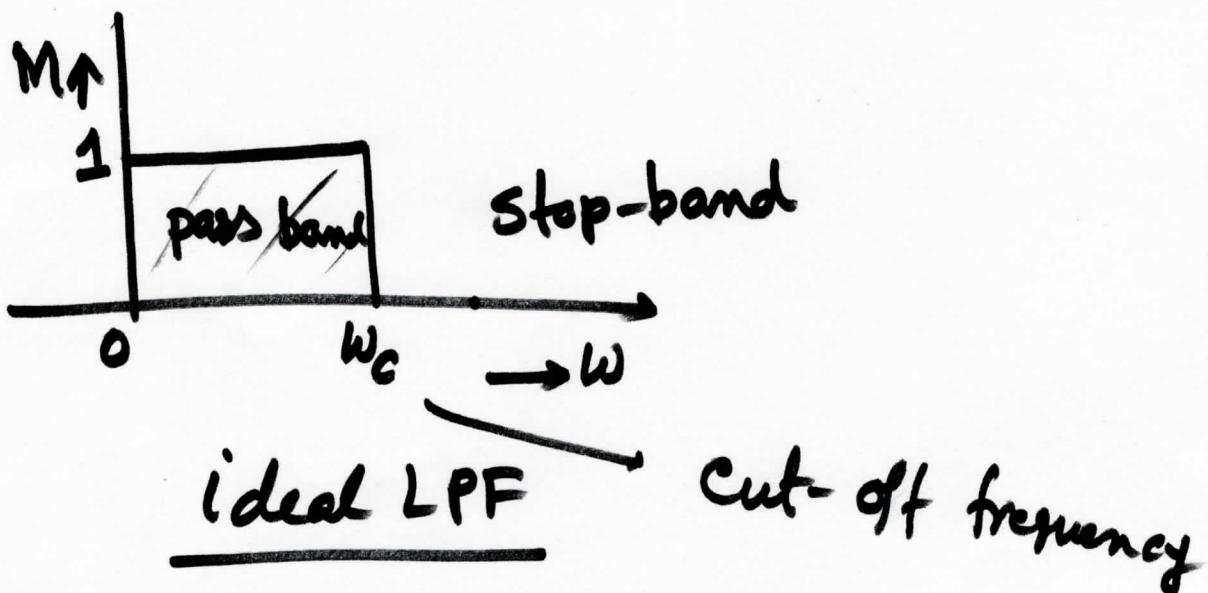


$$\frac{V_o}{V_i} = H(s)$$

$$H(s) \Big|_{s=j\omega} = H(j\omega) = M e^{j\theta}$$

where $M = |H(j\omega)|$ &

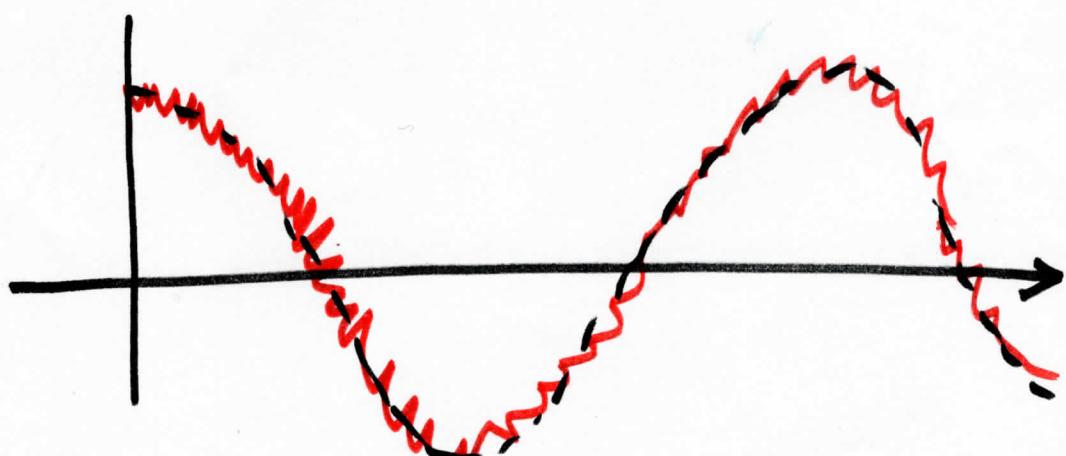
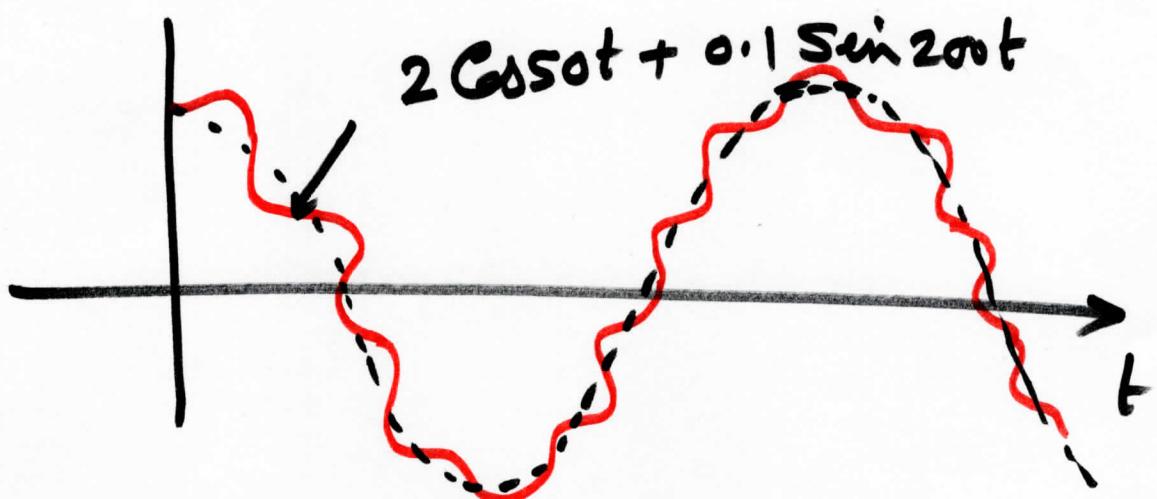
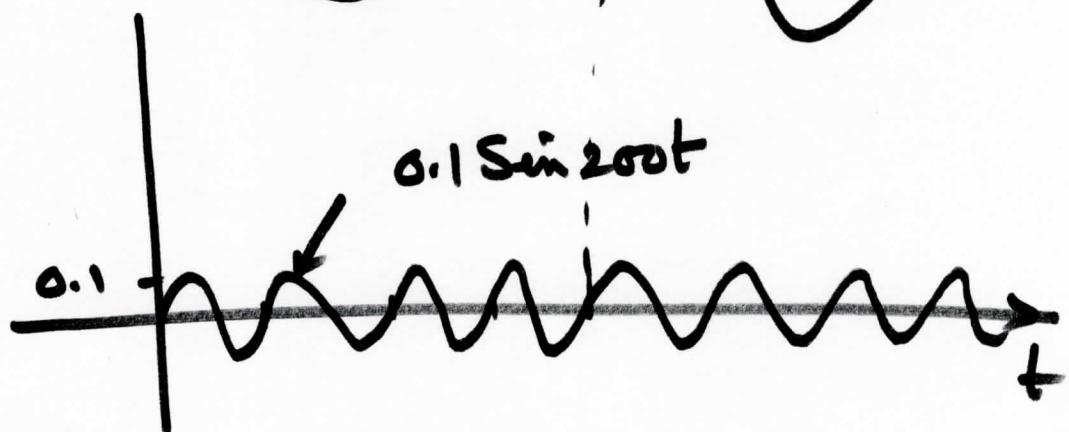
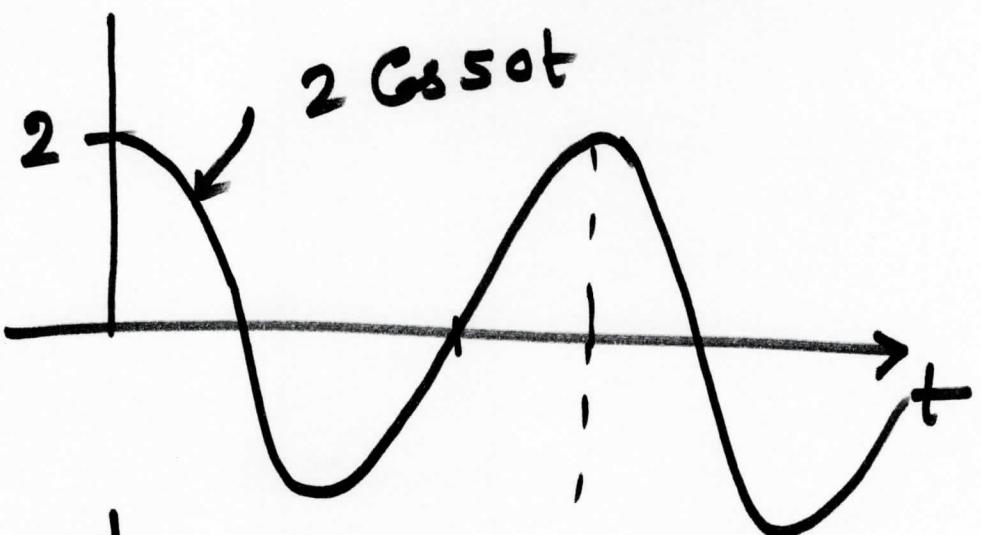
$$\theta = \angle H(j\omega)$$



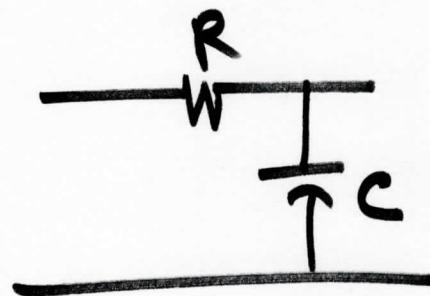
$$\omega_c = 100 \text{ rad/sec.}$$

$$v_i(t) = 2 \cos 50t + 0.10 \sin 200t$$

$$v_o(t) = 2 \cos 50t$$



Practical LPF :

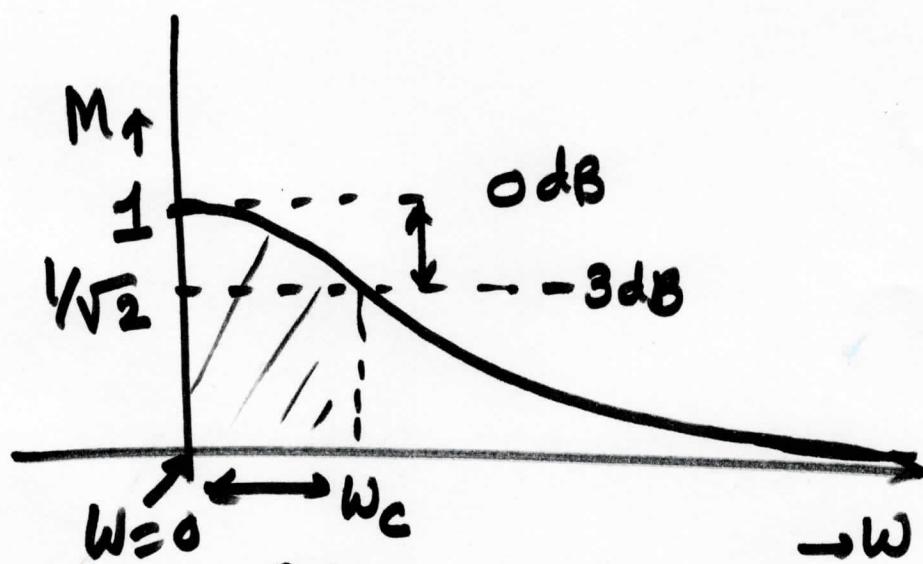


$$H(s) = \frac{1}{1 + sRC}$$

$$H(j\omega) = \frac{1}{1 + j\omega RC}$$

$$\Rightarrow M(\omega) = \frac{1}{\sqrt{1 + \omega^2 R^2 C^2}} = \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$$

where $\omega_c = \frac{1}{RC}$

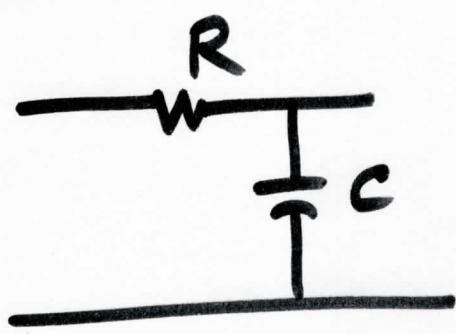


3dB BW

$20 \log_{10} M \rightarrow \text{in dB}$

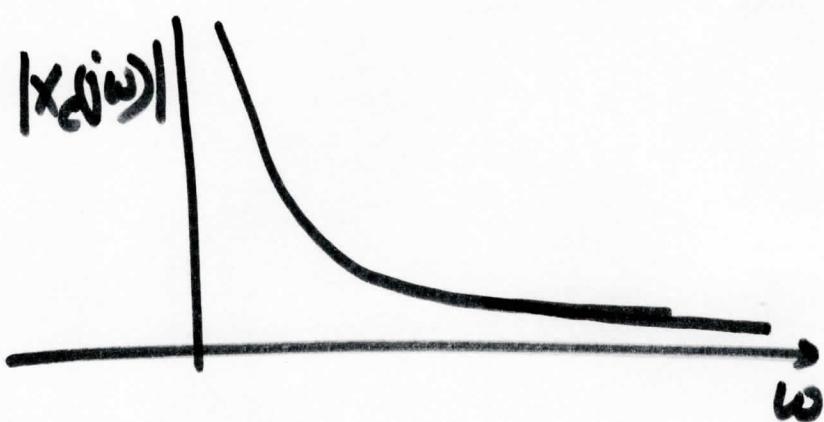
voltage gain

$$\begin{aligned}
 & 20 \log \frac{1}{\sqrt{2}} \\
 & = -10 \log 2 \\
 & = -3 \text{dB}
 \end{aligned} \quad (118)$$



$$X_C(s) = \frac{1}{Cs}$$

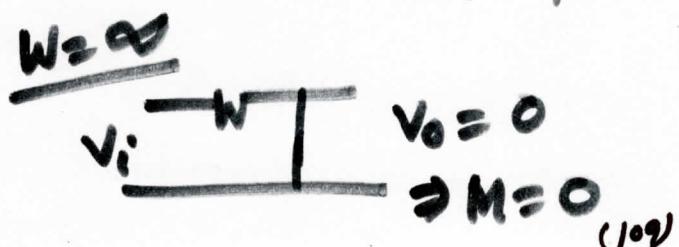
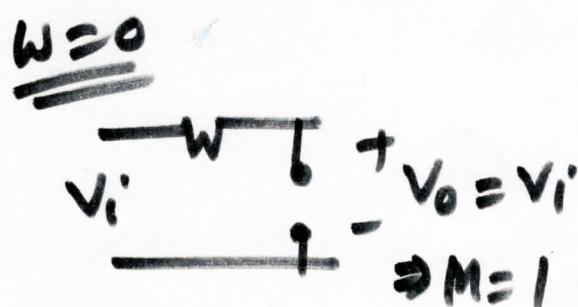
$$|X_C(j\omega)| = \left| \frac{1}{j\omega C} \right| = \frac{1}{\omega C}$$

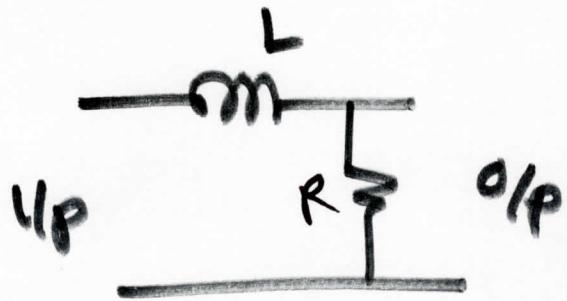


For DC : $|X_C| \rightarrow \infty$ (open ckt)

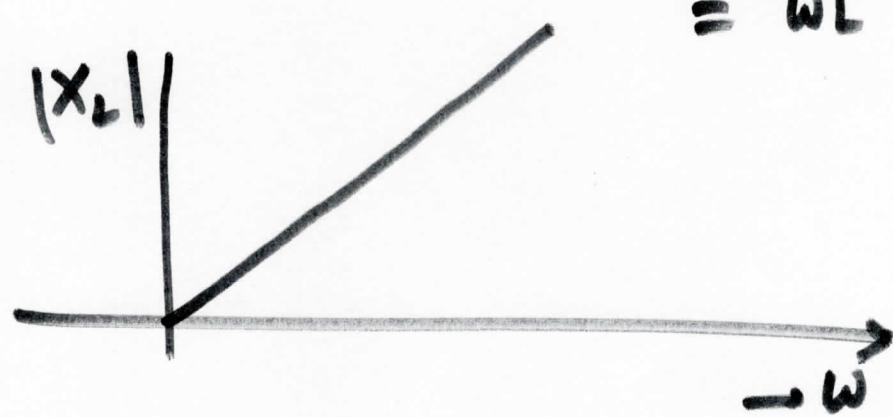
For high freq. AC: $|X_C| \rightarrow 0$ (short ckt)

ω	M
0	1
∞	0



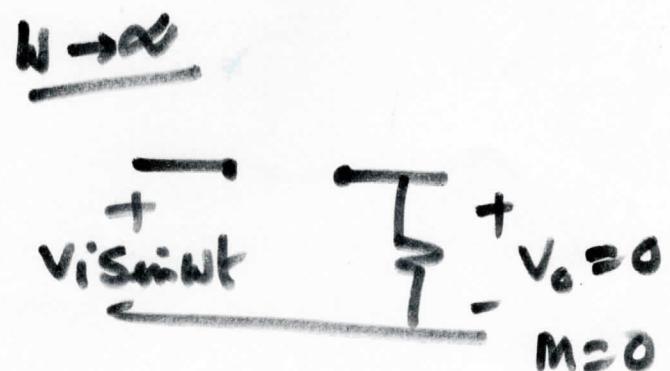
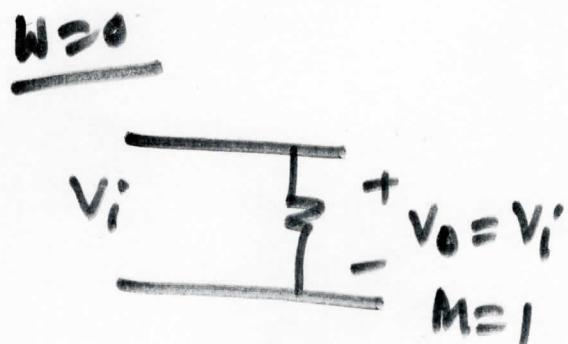
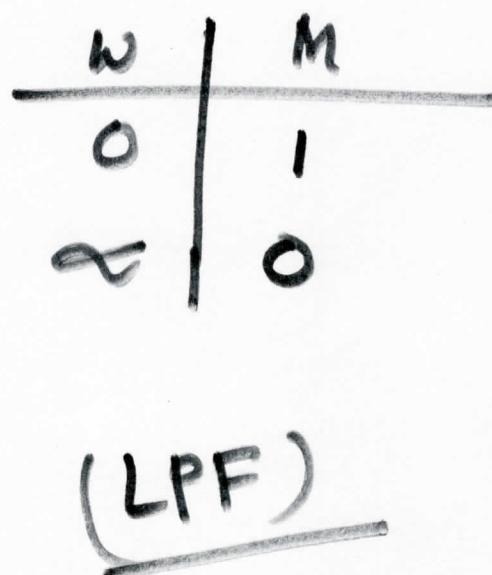


$$\frac{V_o}{V_i} = \frac{R}{R + LS}$$

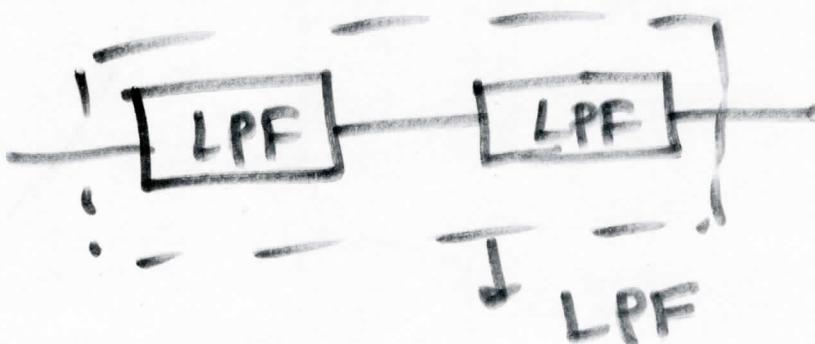
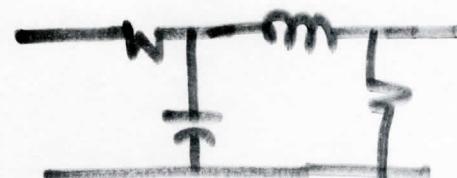
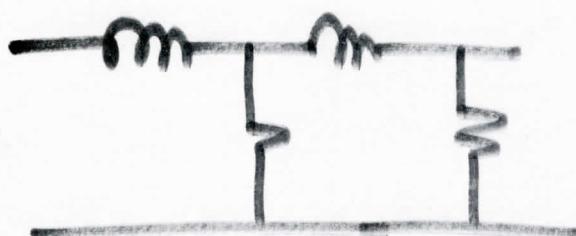
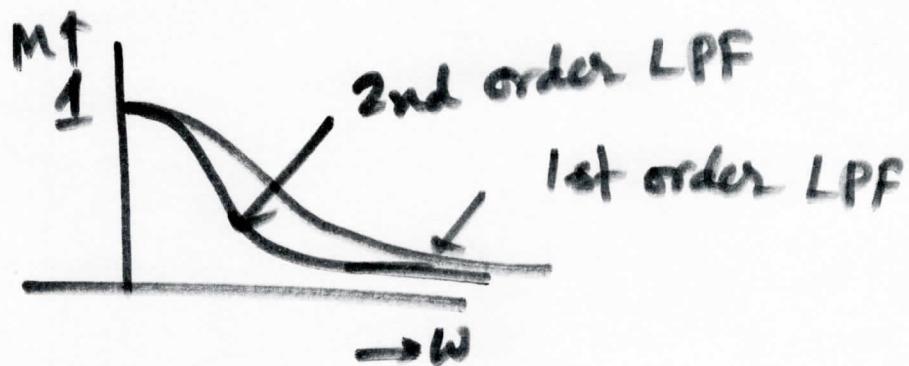
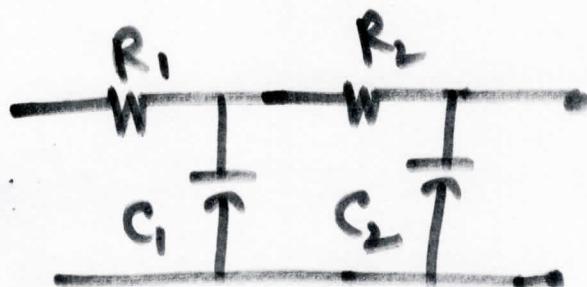


For DC : $|X_L| = 0$ (short ckt)

For high freq AC : $|X_L| \rightarrow \infty$ (open ckt)

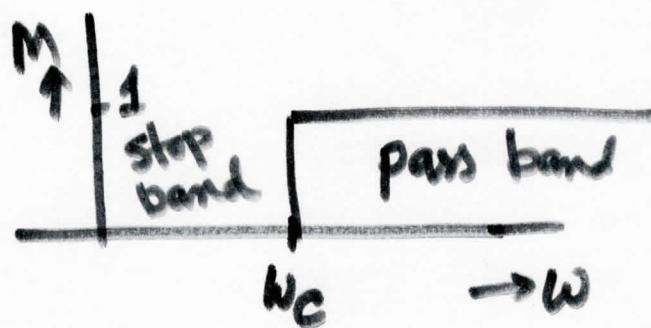


Higher Order LPF



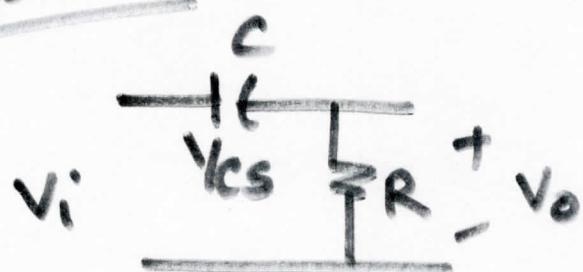
(iii)

High Pass Filter (HPF)



ideal

Practical



$$\frac{V_o}{V_i} = H(s) = \frac{R}{R + \frac{1}{Cs}}$$

$$= \frac{Rcs}{1 + Rcs}$$

$$H(s) = \frac{Rcs}{1 + Rcs}$$

$$= \frac{1}{1 + \frac{1}{Rcs}}$$

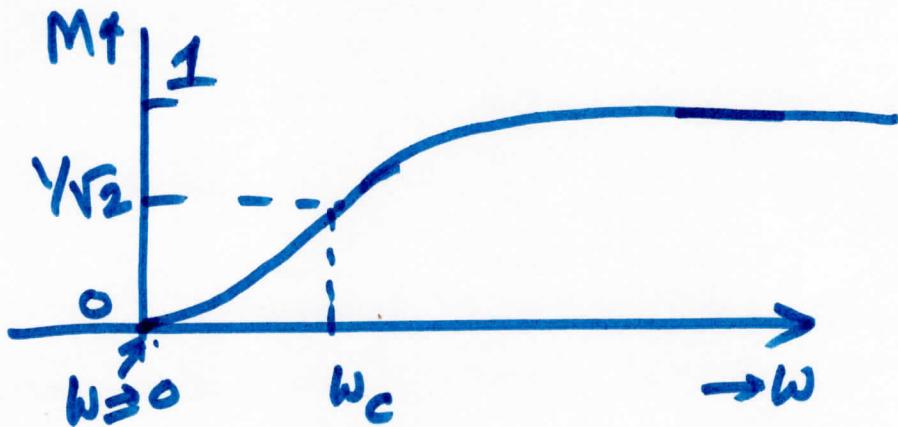
$$\Rightarrow H(j\omega) = \frac{1}{1 + \frac{1}{j\omega Rc}} = \frac{1}{1 + \frac{1}{j\omega Rc}} = \frac{1}{1 - j\left(\frac{\omega_c}{\omega}\right)} = \frac{1}{1 - j\left(\frac{\omega_c}{\omega}\right)}$$

where $\omega_c = \frac{1}{Rc}$

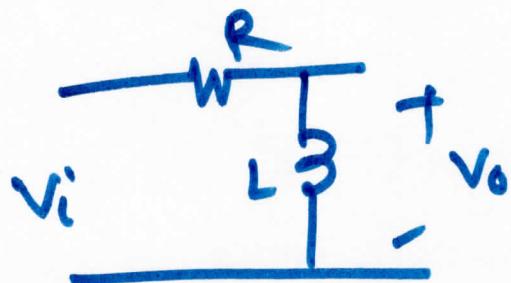
$$M = |H(j\omega)|$$

$$= \frac{1}{\sqrt{1 + \left(\frac{\omega_c}{\omega}\right)^2}}$$

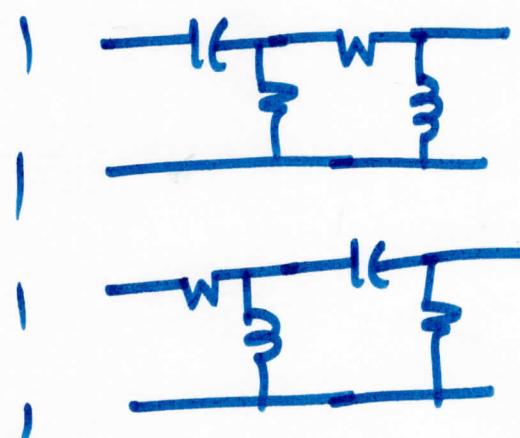
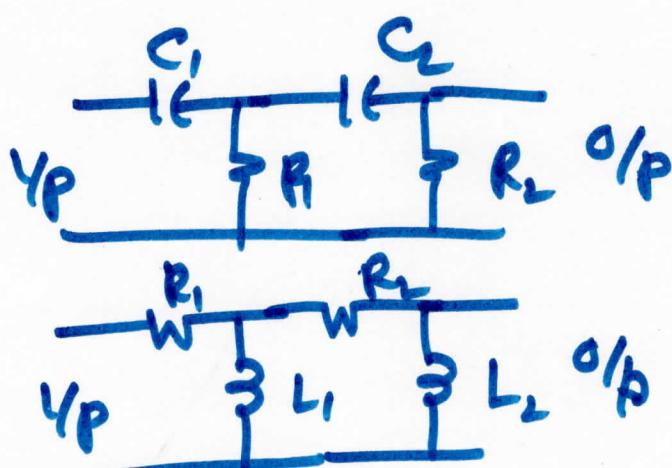
ω	M
0	0
∞	1



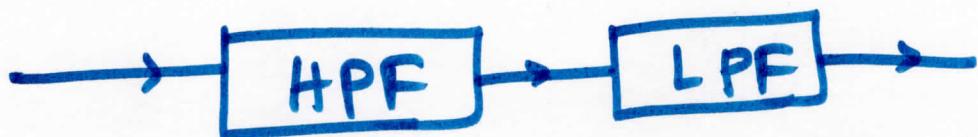
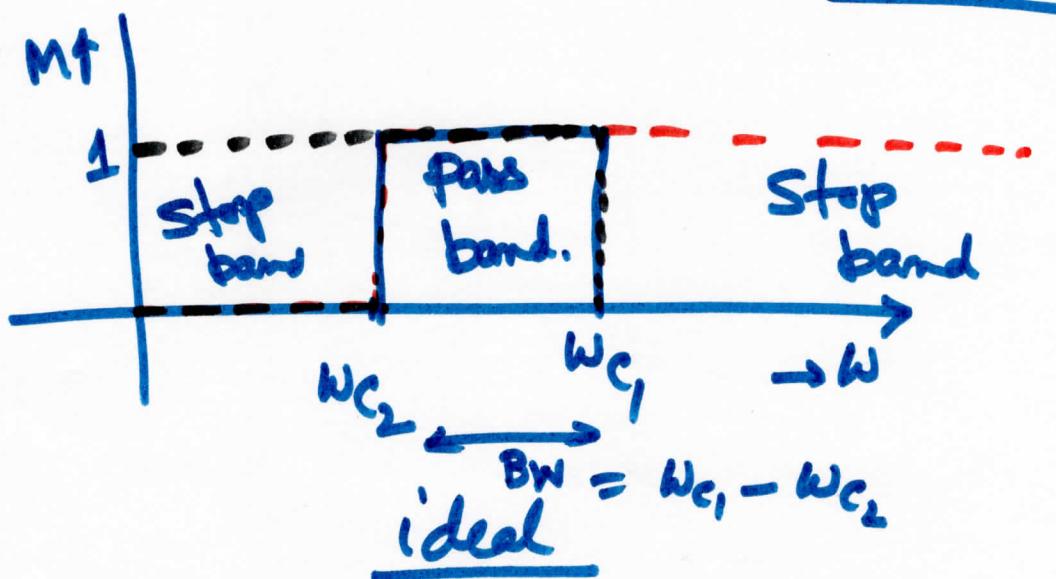
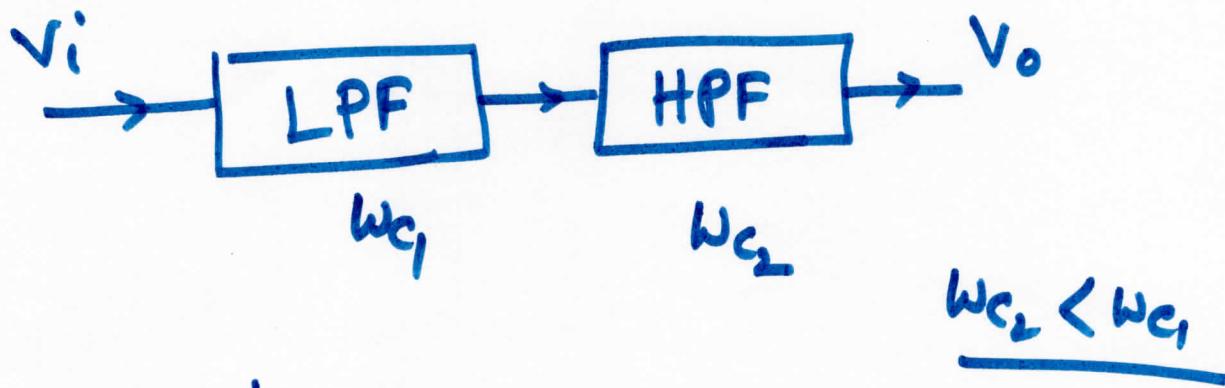
with $L \neq R$



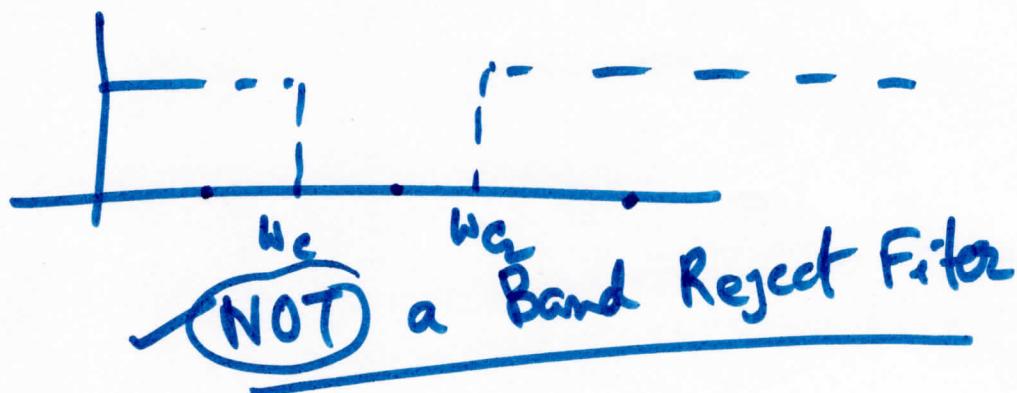
ω	M
0	0
∞	1

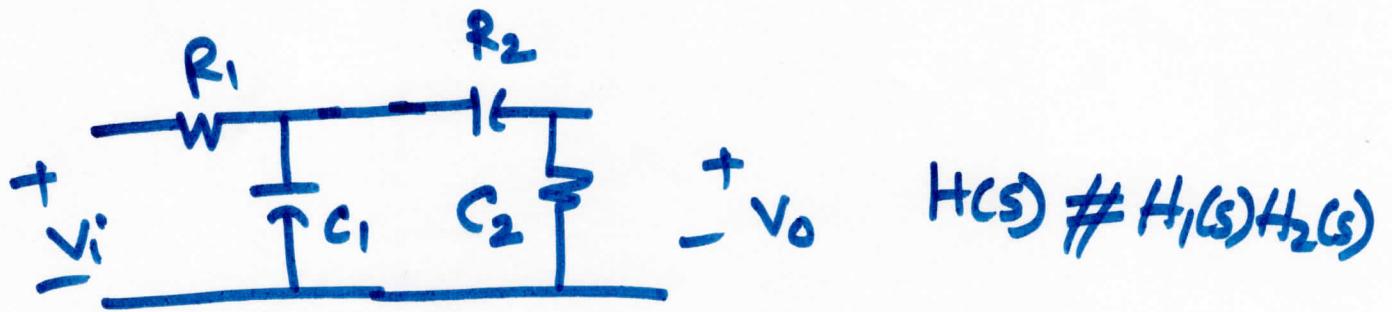


Band Pass Filter (BPF)



$$w_{c_2} > w_{c_1} \rightarrow v_o = 0$$





$$\frac{R_1}{W} \frac{1}{C_1 s} \Rightarrow H_1(s) = \frac{\frac{1}{C_1 s}}{R_1 + \frac{1}{C_1 s}} = \frac{1}{1 + R_1 C_1 s}$$

$$\frac{C_2}{R_2} \Rightarrow H_2(s) = \frac{R_2}{R_2 + \frac{1}{C_2 s}} = \frac{R_2 C_2 s}{1 + R_2 C_2 s}$$

$$H(s) = \frac{P(s)}{Q(s)}$$

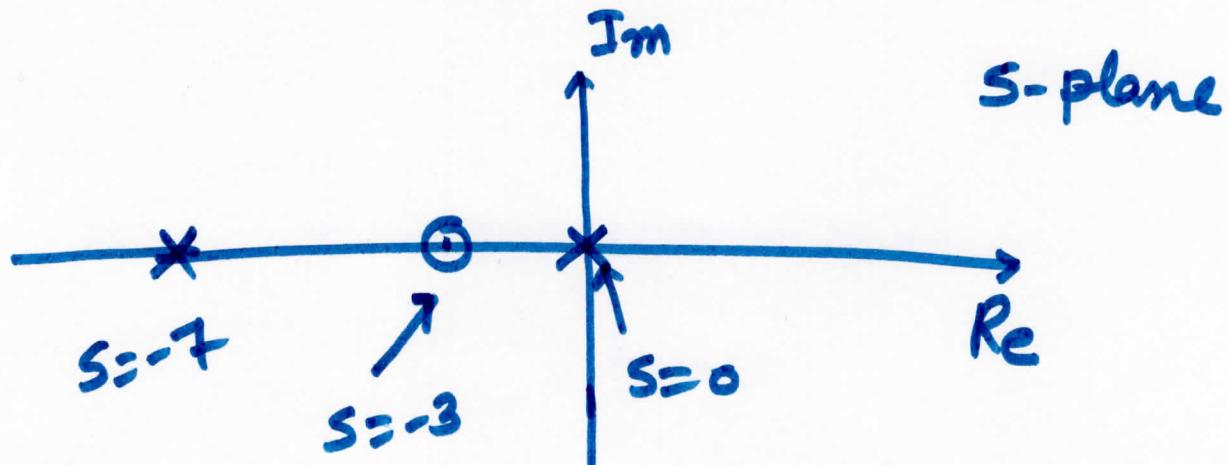
$P(s) = 0 \Rightarrow s = ?$ zeros. ('o')

$Q(s) = 0 \Rightarrow s = ?$ poles ('x')

ex

$$H(s) = \frac{2(s+3)}{s(s+7)}$$

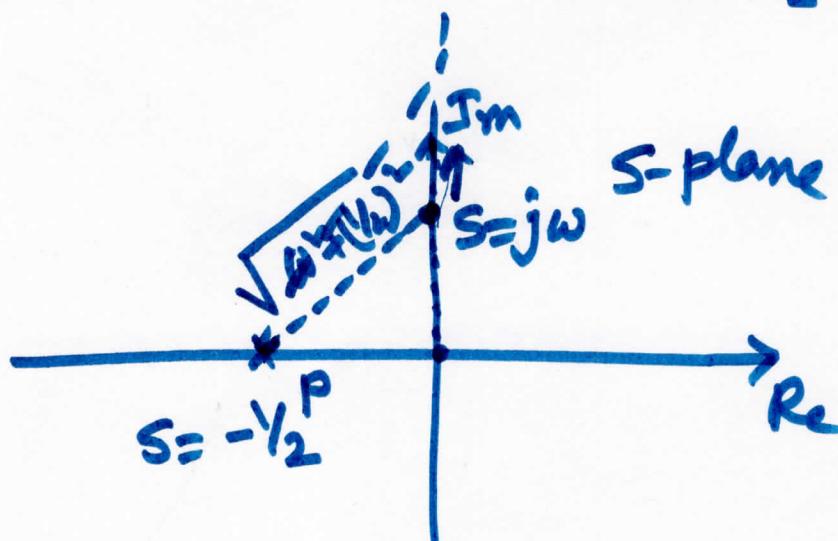
$$s = \sigma + j\omega$$



ex

LPF

$$H(s) = \frac{1}{1+2s} = \frac{\gamma_2}{s+\gamma_2} \quad \begin{cases} M = |H(j\omega)| \\ \quad \quad \quad = |H(s)|_{s=j\omega} \end{cases}$$

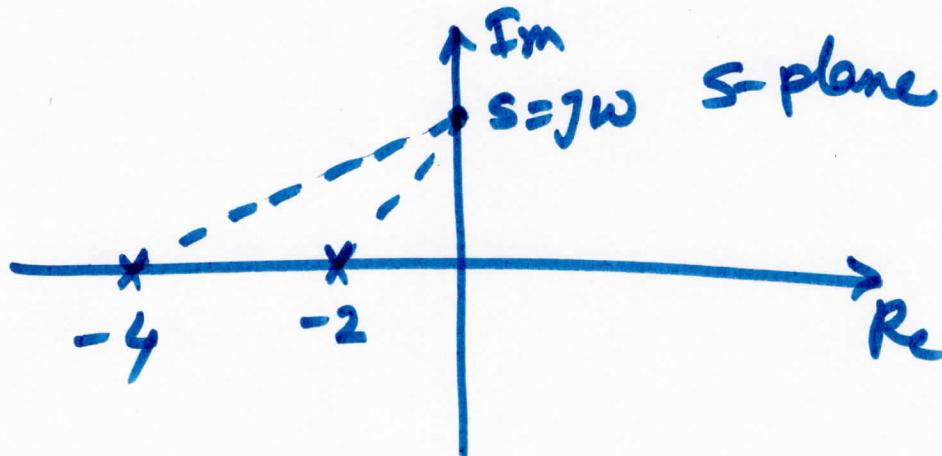


$$\begin{aligned} |H(j\omega)| &= \frac{\gamma_2}{\sqrt{\omega^2 + (\gamma_2)^2}} \\ &= \frac{\gamma_2}{\omega \sqrt{1 + (\omega/\gamma_2)^2}} \end{aligned}$$

$$H(s) = \frac{K(s+z_1)(s+z_2)\dots}{(s+p_1)(s+p_2)\dots}$$

$\hat{H}(s)$

$$H(s) = \frac{1}{(s+2)(s+4)}$$



$$M = |H(j\omega)|$$

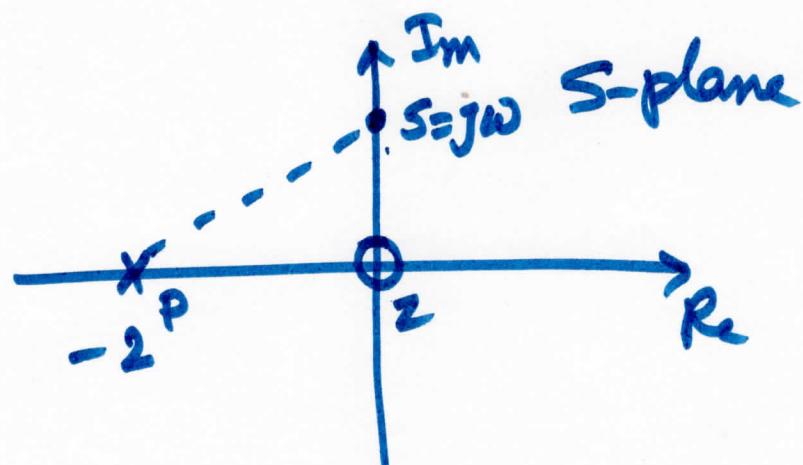
$$= \frac{1}{\sqrt{2^2 + \omega^2} \cdot \sqrt{4^2 + \omega^2}}$$

$$\omega \uparrow \rightarrow M \downarrow \quad (\text{LPF})$$

All poles system : \rightarrow LPF

Ex

$$H(s) = \frac{s}{s+2}$$



$$M = |H(j\omega)| = \frac{\sqrt{s^2}}{\sqrt{s^2 + P^2}} = \frac{\omega}{\sqrt{2^2 + \omega^2}}$$



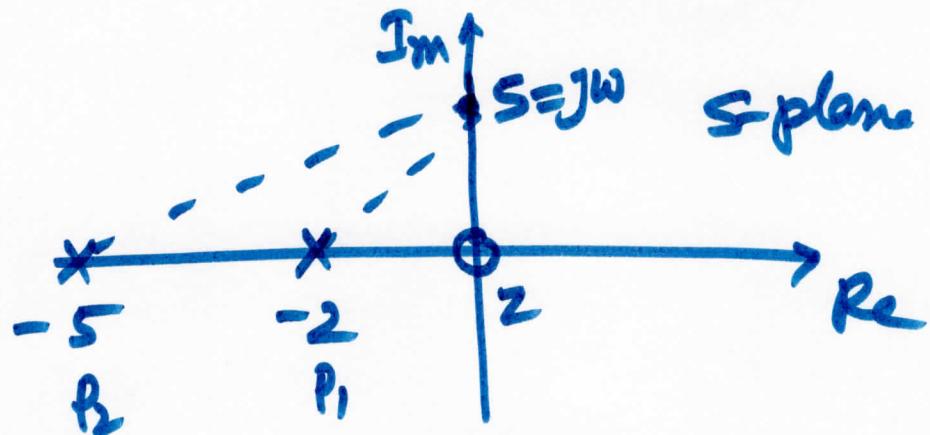
(HPF)

$$H(s) = \frac{s^2}{(s+2)(s+4)}$$

No. of zeros at origin = No. of poles
on the LHS of S-plane

↓
HPF

$$\text{Ex} \quad H(s) = \frac{s}{(s+2)(s+5)} = \left(\frac{s}{s+2}\right) \times \frac{1}{(s+5)}$$



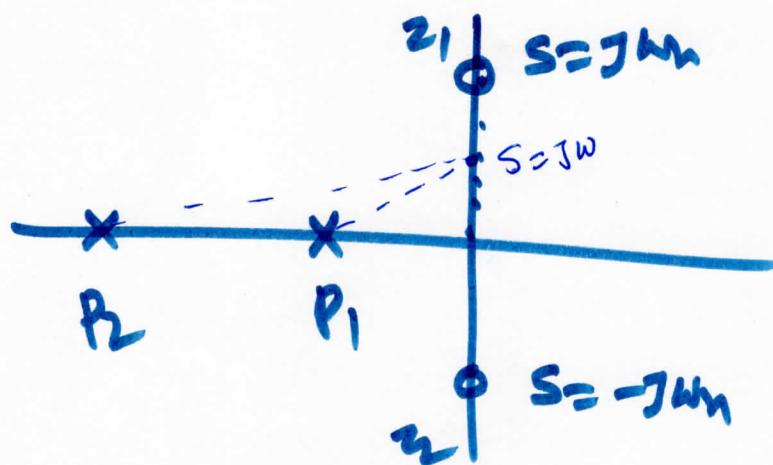
$$M = |H(j\omega)| \\ = \frac{\overline{sz}}{\overline{sp_1} \cdot \overline{sp_2}}$$

$$= \frac{\omega}{\sqrt{2^2 + \omega^2} \sqrt{5^2 + \omega^2}}$$

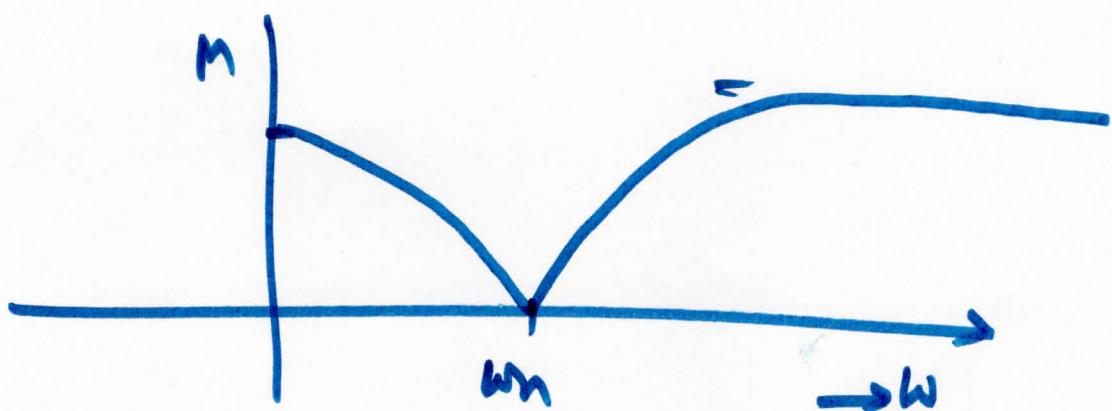


Ex

$$H(s) = \frac{s^2 + \omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$



$$M = \frac{\overline{sz_1} \cdot \overline{sz_2}}{\overline{sp_1} \cdot \overline{sp_2}}$$



Notch Filter
(Notch)

$$H(s) = \frac{Q(s)}{s^2 + 2j\omega_n s + \omega_n^2}$$

LPF : $Q(s) = \omega_n^2$

HPF : $Q(s) = s^2$

BPF : $Q(s) = s$

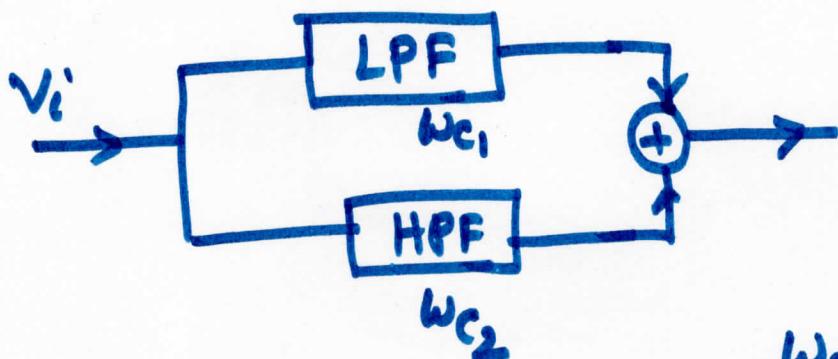
BRF : $Q(s) = s^2 + \omega_n^2$

APF : $Q(s) = s^2 - 2j\omega_n s + \omega_n^2$
(All pass Filter)

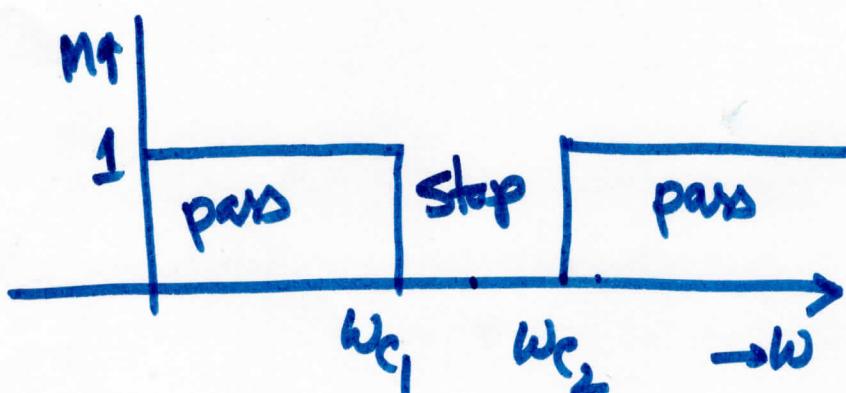
BRF

$$\begin{aligned} s^2 + 2j\omega_n s + \omega_n^2 &= 0 \\ s = -2j\omega_n \pm \sqrt{4j^2\omega_n^2 - 4\omega_n^2} &= \frac{-2j\omega_n \pm \omega_n\sqrt{j^2-1}}{2} \\ &= -j\omega_n \pm \omega_n\sqrt{-1} \end{aligned}$$

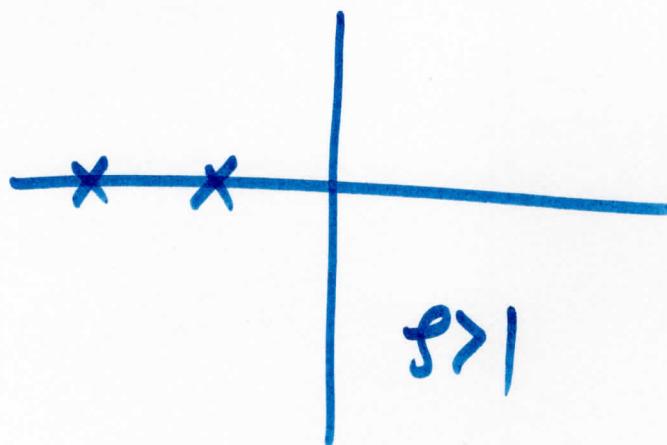
$j > 1 \rightarrow$ real & unequal
over-damped
 $j = 1 \rightarrow$ real & equal
critically "



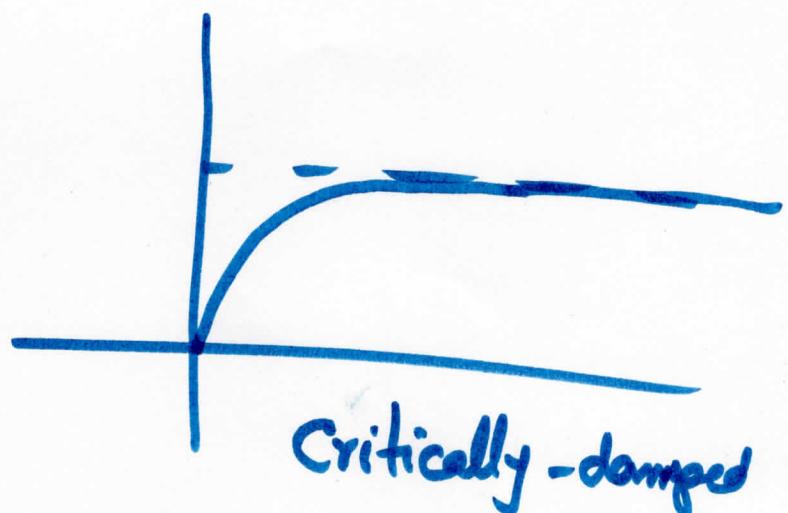
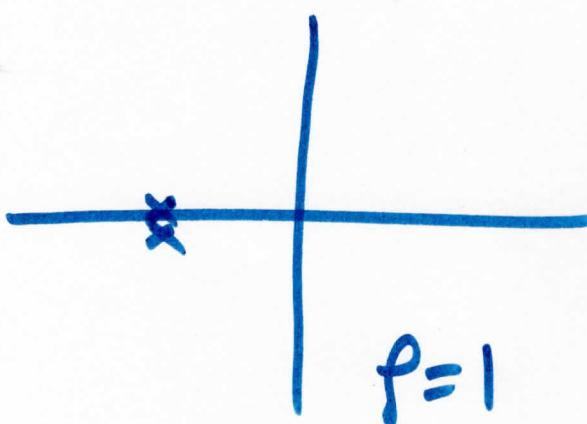
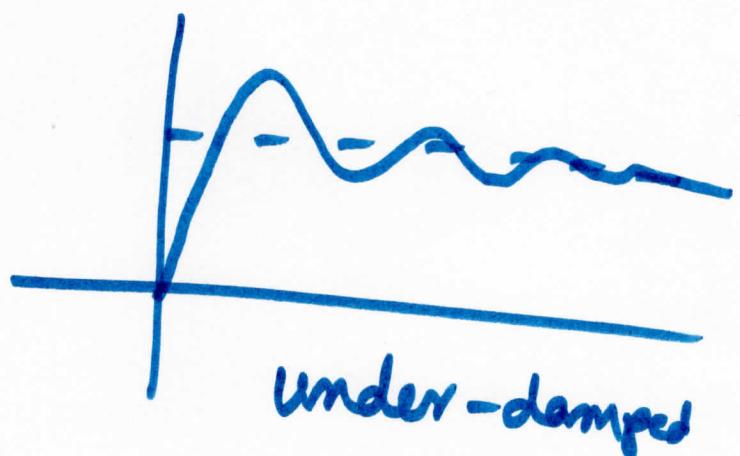
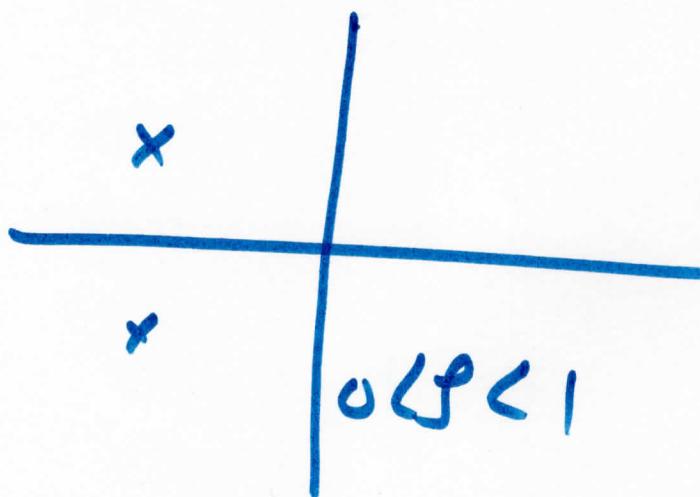
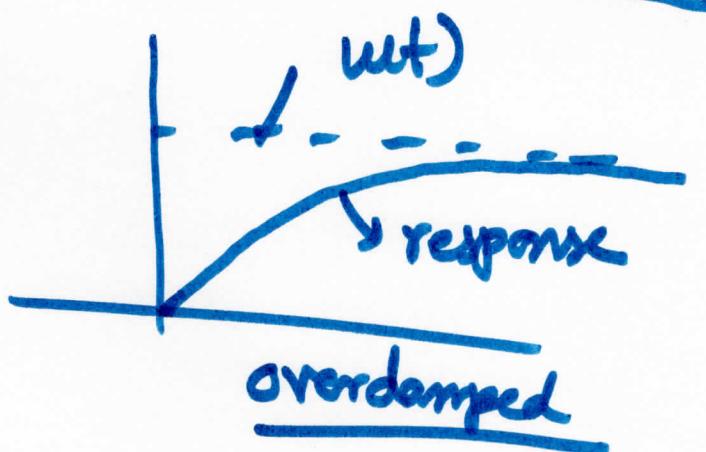
$0 < j < 1 \rightarrow$ Complex Conjugate
 $\omega_{c_2} > \omega_{c_1} \rightarrow$ under damped



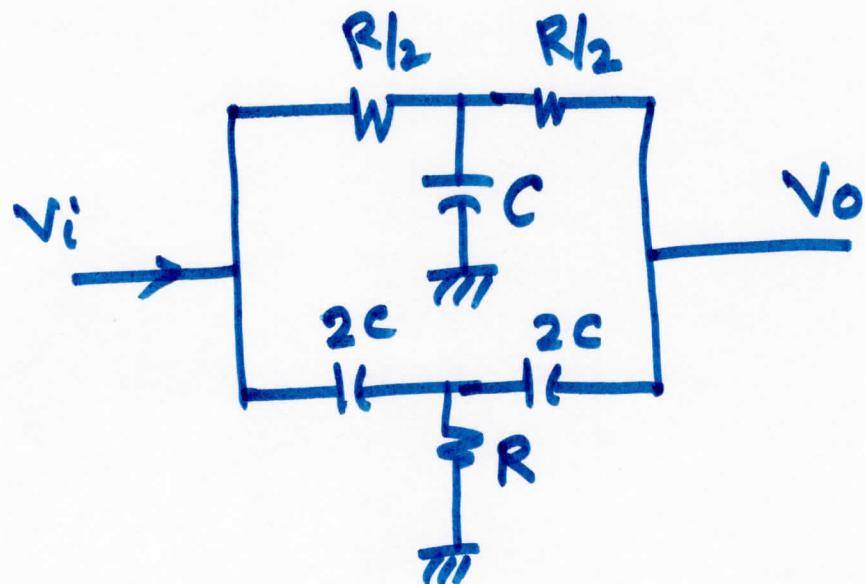
LPF



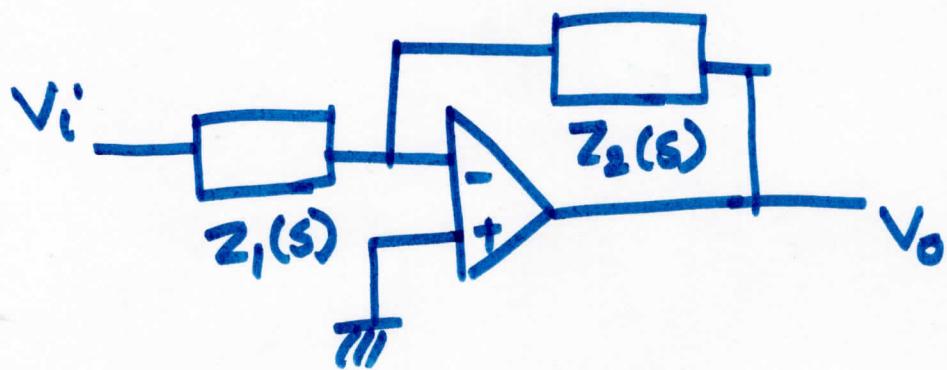
Step-response



Design of BRF



Active Filters



$$\frac{V_o(s)}{V_i(s)} = - \frac{Z_2(s)}{Z_1(s)}$$

For $Z_1(s)$ or $Z_2(s)$:

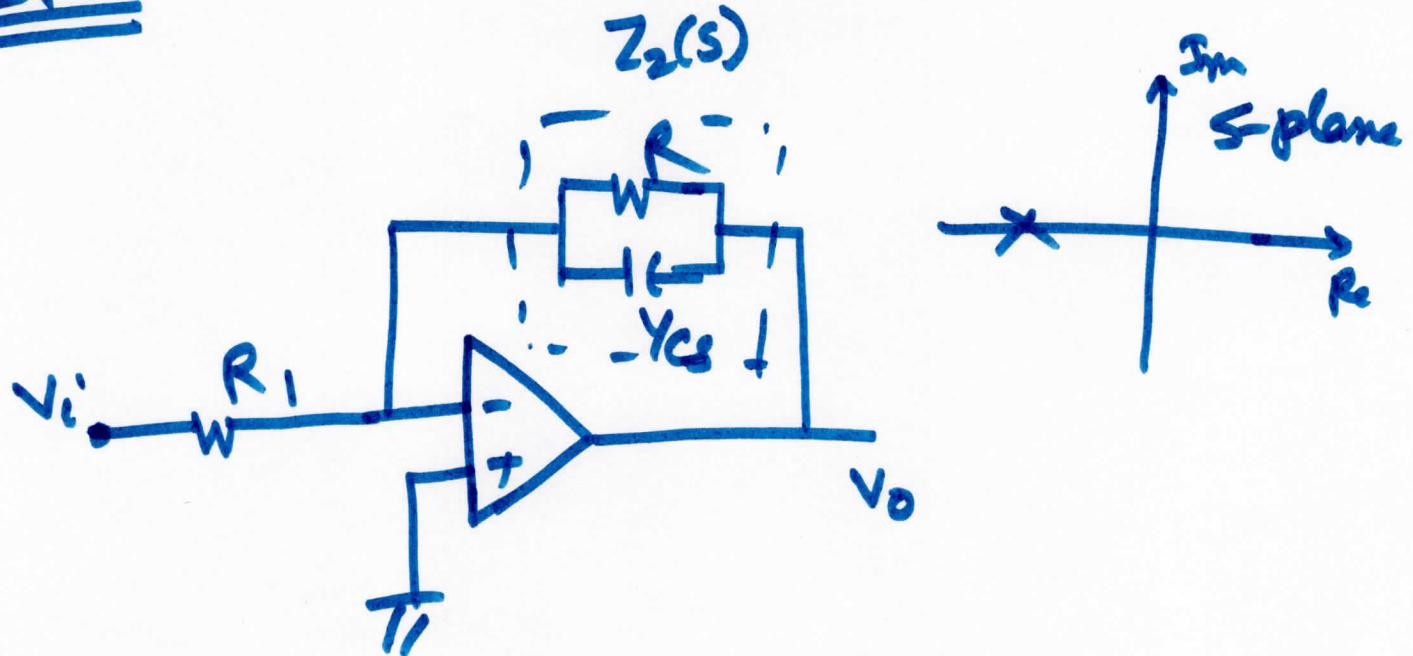
1) $\Rightarrow Z(s) = R$

2) $\Rightarrow Z(s) = 1/sC$

3) $\Rightarrow Z(s) = \frac{1+RCS}{CS}$

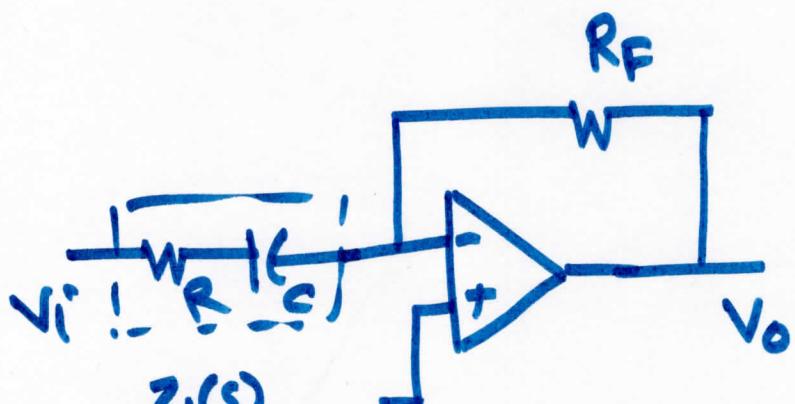
4) $\Rightarrow Z(s) = \frac{R}{1+RCS}$

LPF

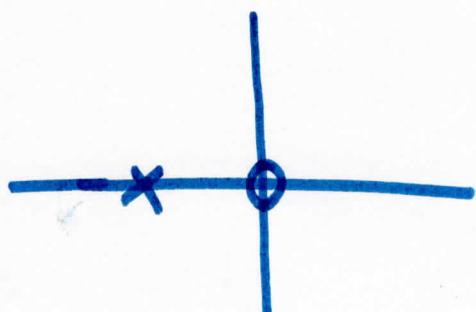


$$\begin{aligned}\frac{V_o}{V_i}(s) &= -\frac{Z_2(s)}{Z_1(s)} = -\frac{R/(1+RCS)}{R_1} \\ &= -\frac{(R/R_1)}{(1+RCS)}\end{aligned}$$

HPF

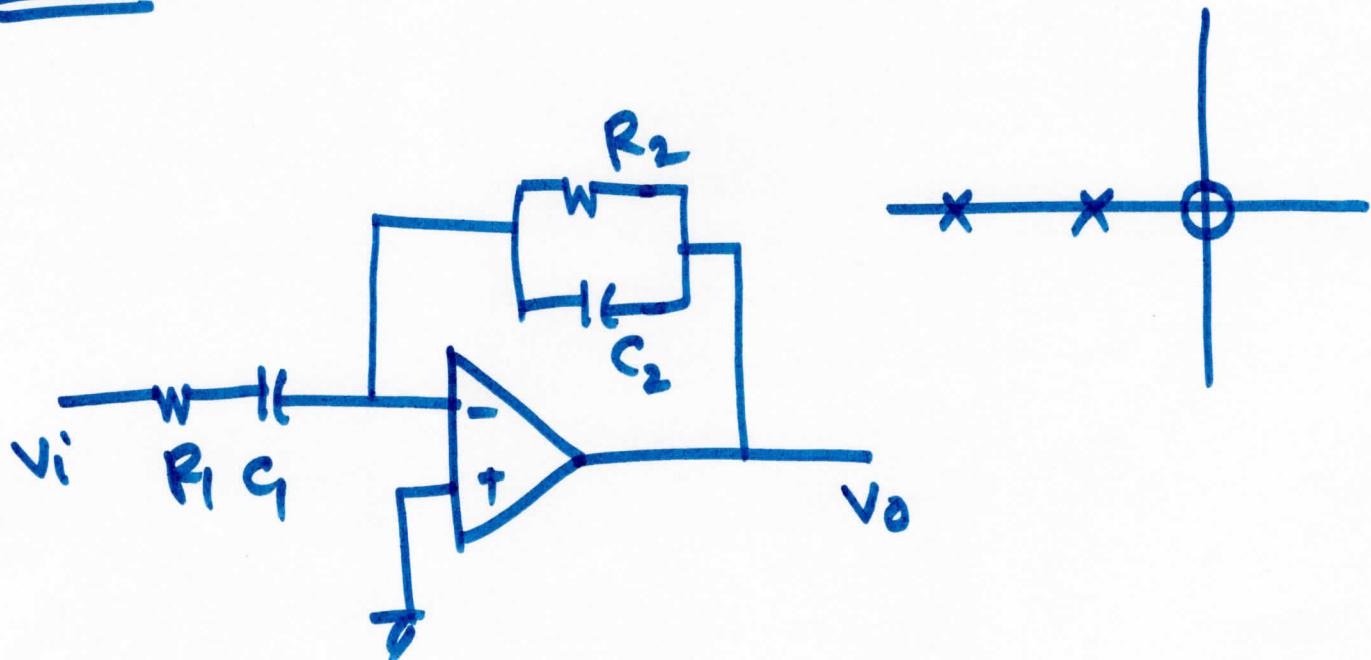


$$Z_1(s) = \frac{1+RCS}{Cs}$$



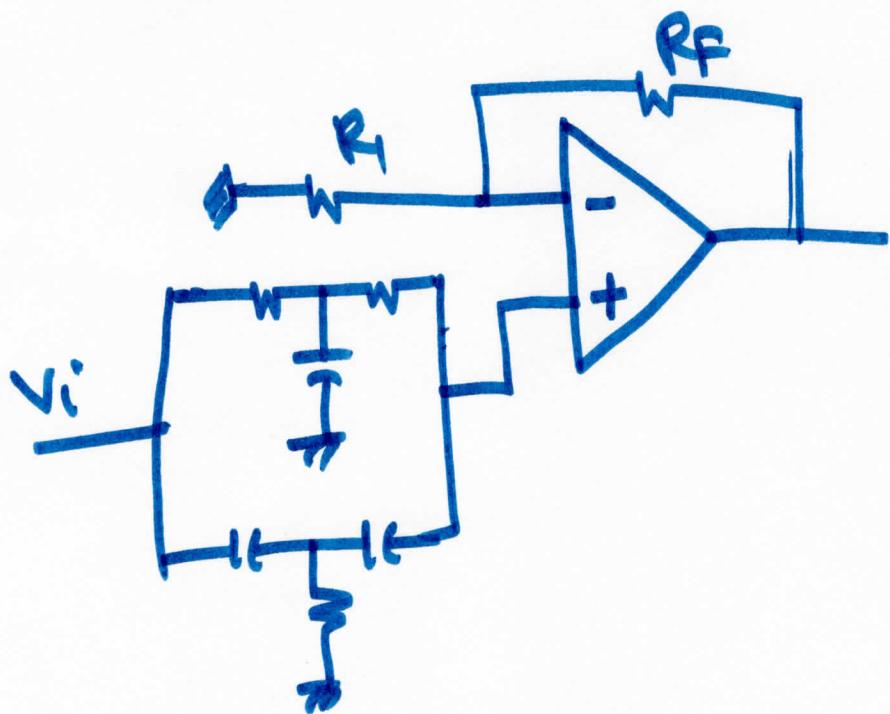
$$\frac{V_o}{V_i} = -\frac{R_F}{\frac{(1+RCS)}{Cs}} = \frac{-R_F C S}{1+RCS}$$

BPF

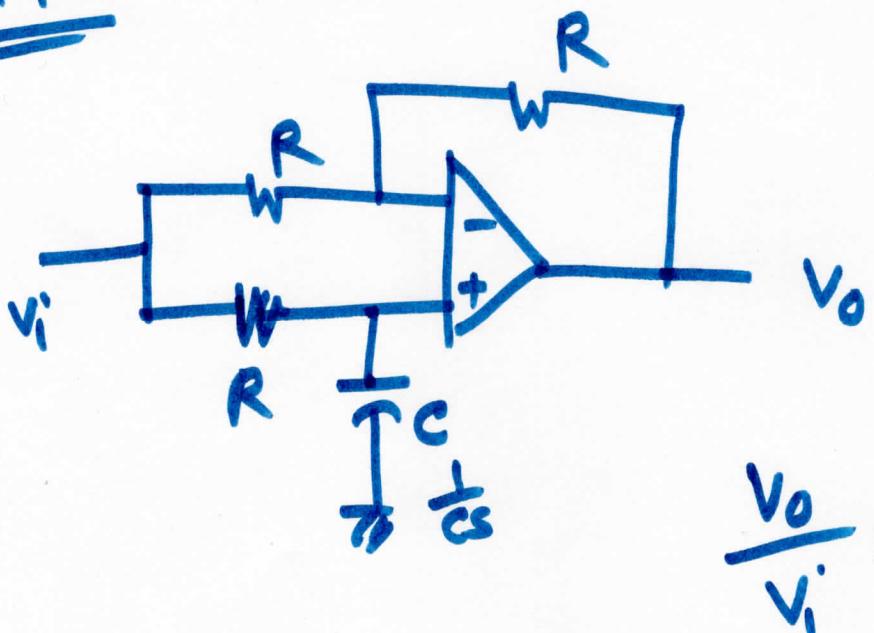


$$\begin{aligned}\frac{v_o}{v_i} &= - \frac{Z_2(s)}{Z_1(s)} \\ &= - \frac{R_2/(1+R_2C_2s)}{(1+R_1C_1s)/C_1s} \\ &= - \frac{R_2 C_1 s}{(1+R_1C_1s)(1+R_2C_2s)}\end{aligned}$$

B NOTCH Filter



APF



$$\frac{V_o}{V_i} = ? = H(s)$$