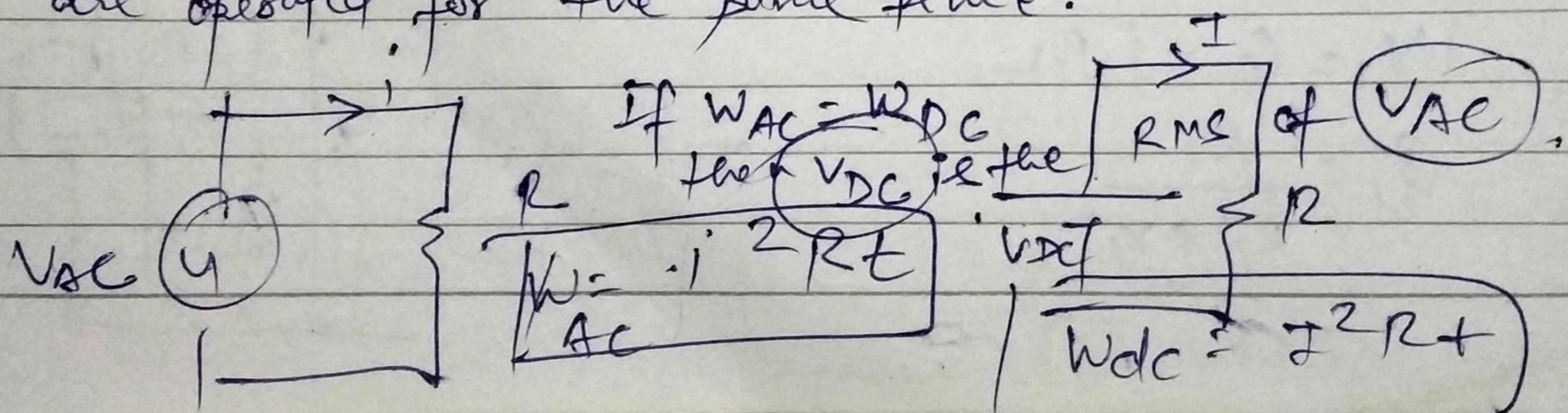


RMS value:

→ RMS value is defined based on the heating effect of the waveform.

→ The voltage of which heat dissipation in the ac ckt is equal to the heat dissipated in the dc ckt is called as RMS value provided both ac and dc have equal value of resistance and are operated for the same time.

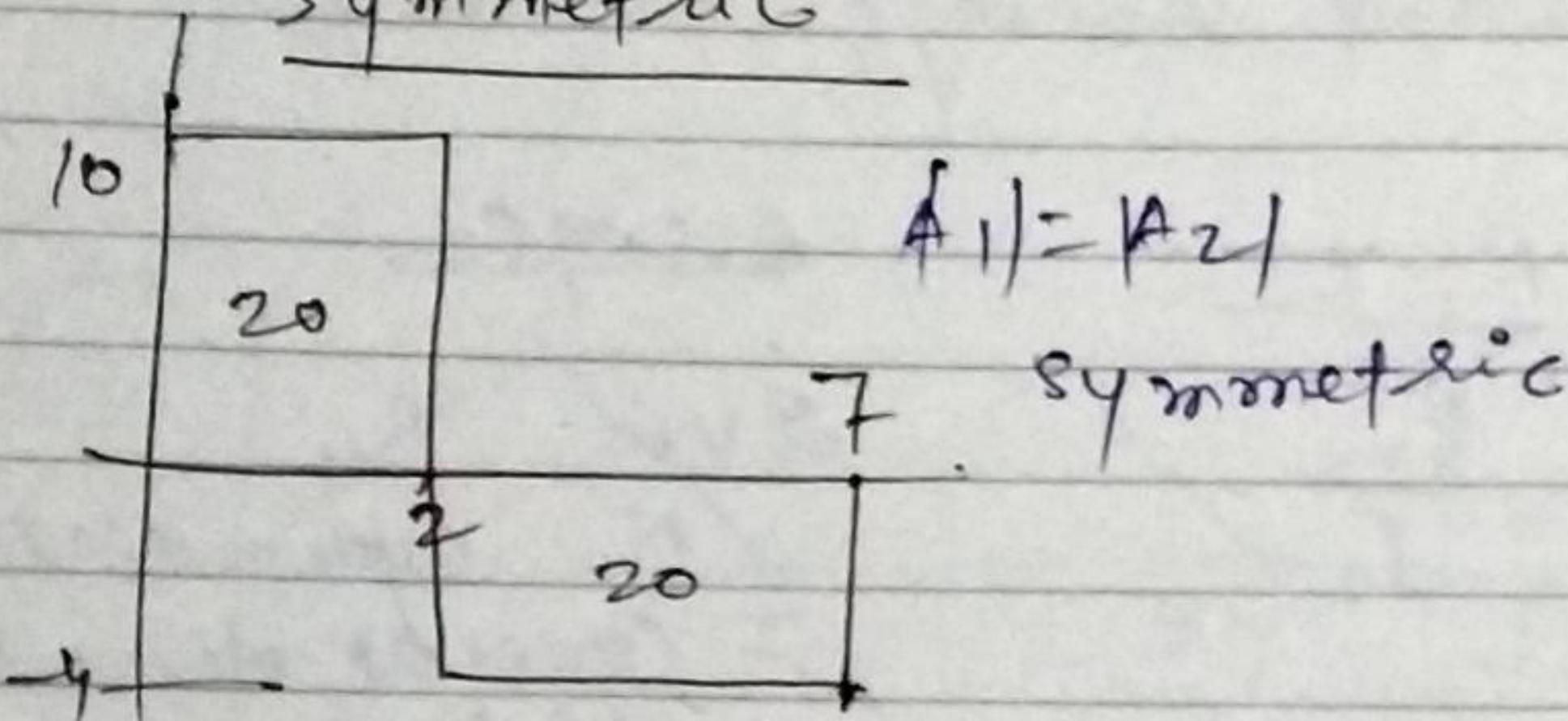


$$V_{RMS} = \sqrt{\left(\frac{1}{T} \int_0^T V^2 dt\right)} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} V^2 dt}$$

Average Value

- It is based on the charge transfer in the ckt
- The voltage at which charge transfer in an AC ckt is equal to the charge transfer in the DC ckt is called as Average value provided both AC and DC have equal value of resistances and are operated for same time interval.

Symmetric



$$V_{av} = \frac{1}{T/2} \int_0^{T/2} V dt$$

half cycle
 \rightarrow full cycle

$$V_{av} = \frac{2}{\pi} \int_0^{\pi/2} V_m \sin \omega t dt$$

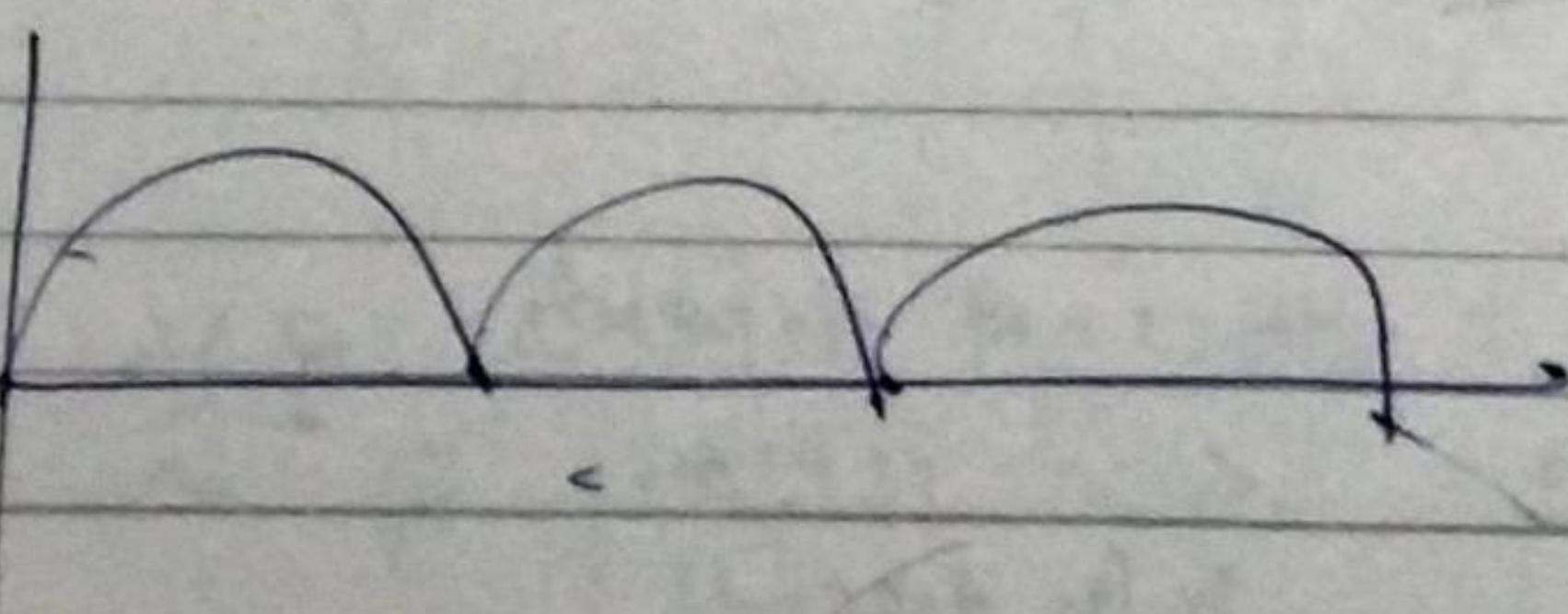
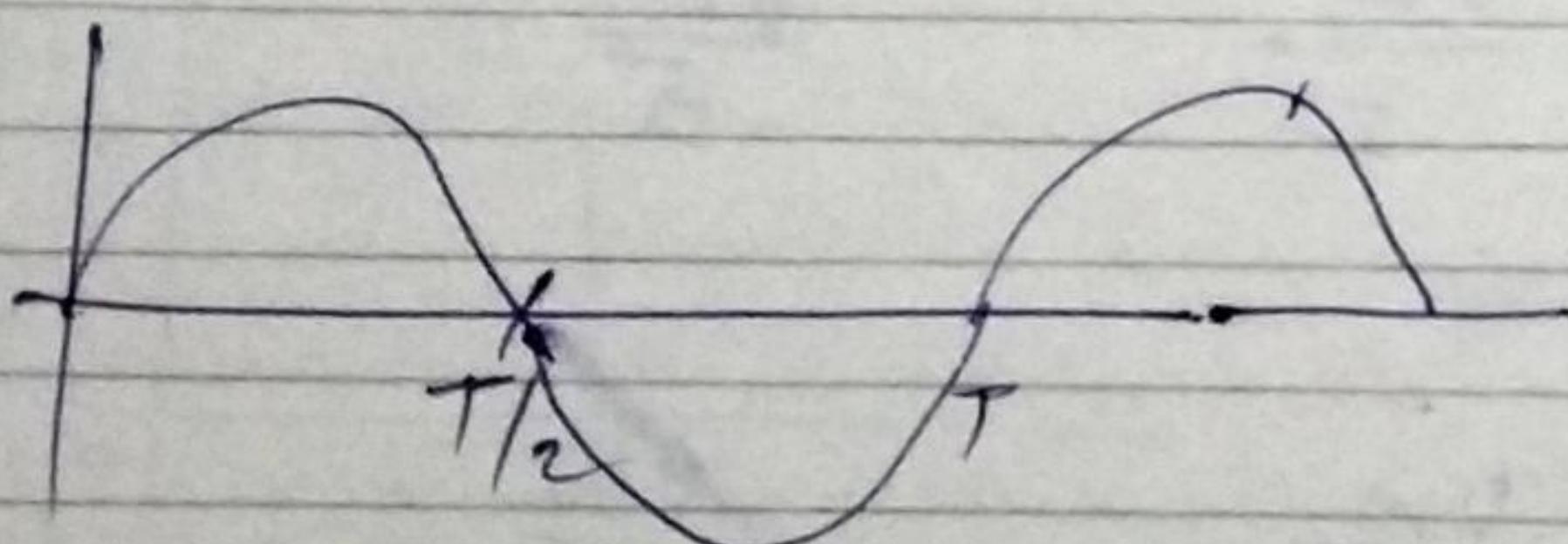
$$V_{av} = \frac{V_m}{\pi} [1 - \cos \omega t]$$

(a) $\frac{V_m}{\pi}$ ✓ $V_{av} = \frac{2V_m}{\pi}$

(b) $\frac{V_m}{\pi}$

(c) $\frac{V_m}{\pi}$

(d) $\frac{2V_m}{\pi}$ ✓



unsymmetrical -

$$V_{av} = \frac{1}{T} \int_0^T v dt$$

Form Factor:

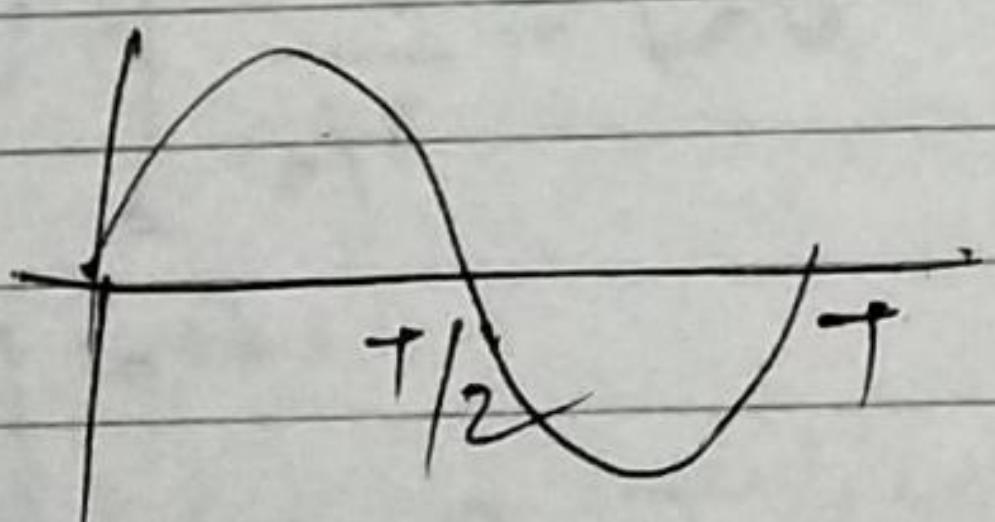
→ It is the ratio of RMS value of the waveform to the average value of the waveform.

• Peak Factor :

→ It is the ratio of mag. value of the waveform to RMS value of the waveform.

Calculate the RMS and Average values for the given w/o F

①



RMS

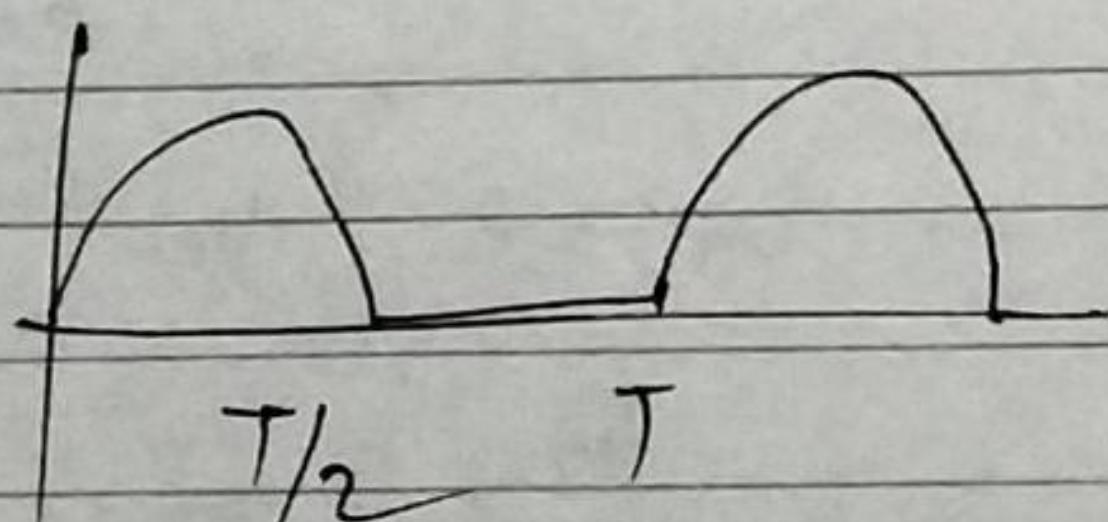
$$\frac{V_m}{\sqrt{2}}$$

Average -

$$\frac{2V_m}{T}$$

By
↑ symmetric
concept otherwise
it is 0.

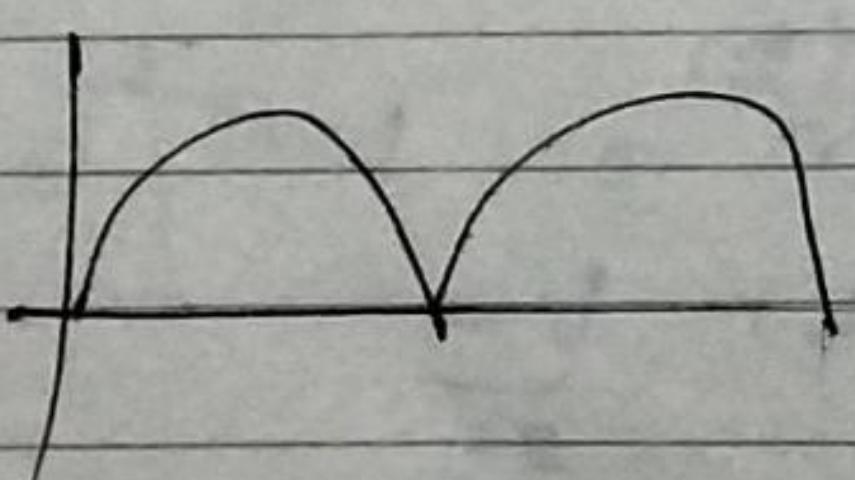
②



$$\frac{V_m}{2}$$

$$\frac{V_m}{T}$$

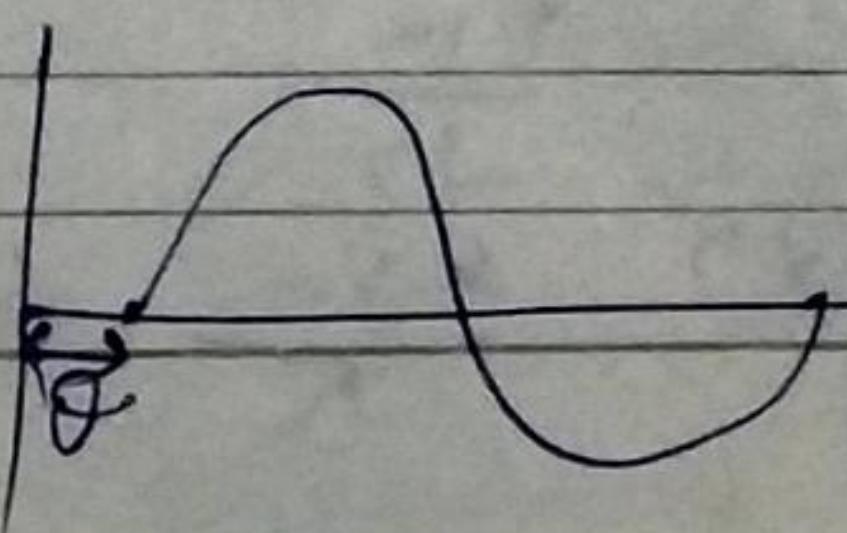
③



$$\frac{V_m}{\sqrt{2}}$$

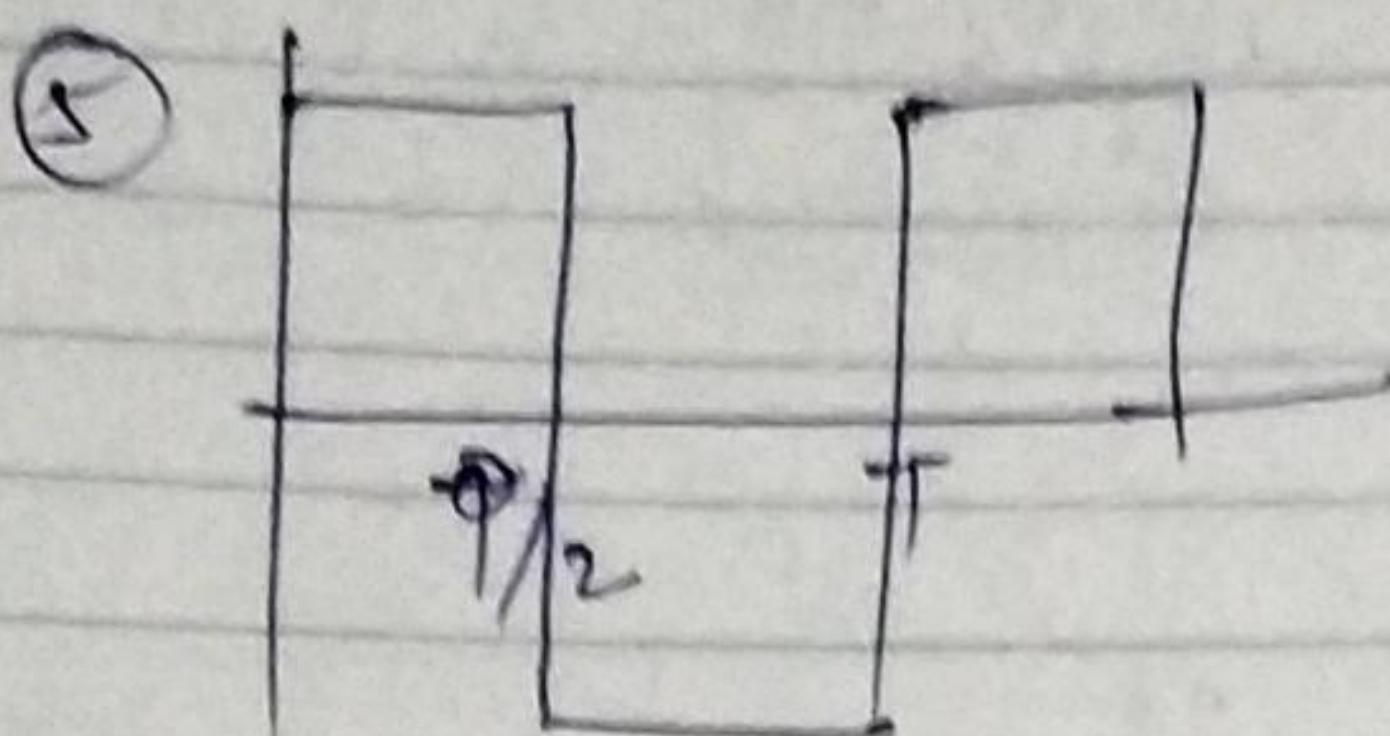
$$\frac{2V_m}{T}$$

④



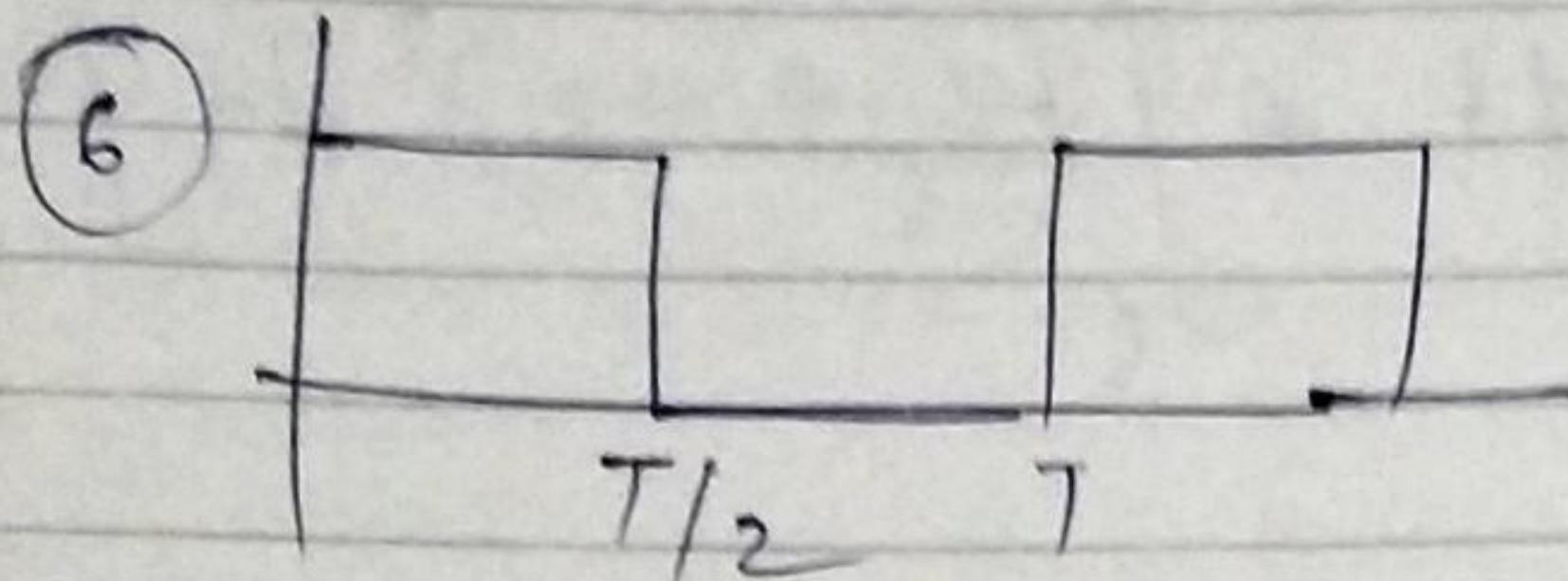
$$\frac{V_m}{\sqrt{2}}$$

does not depend. $\frac{2V_m}{T}$
One shifting
On heating. Only depends



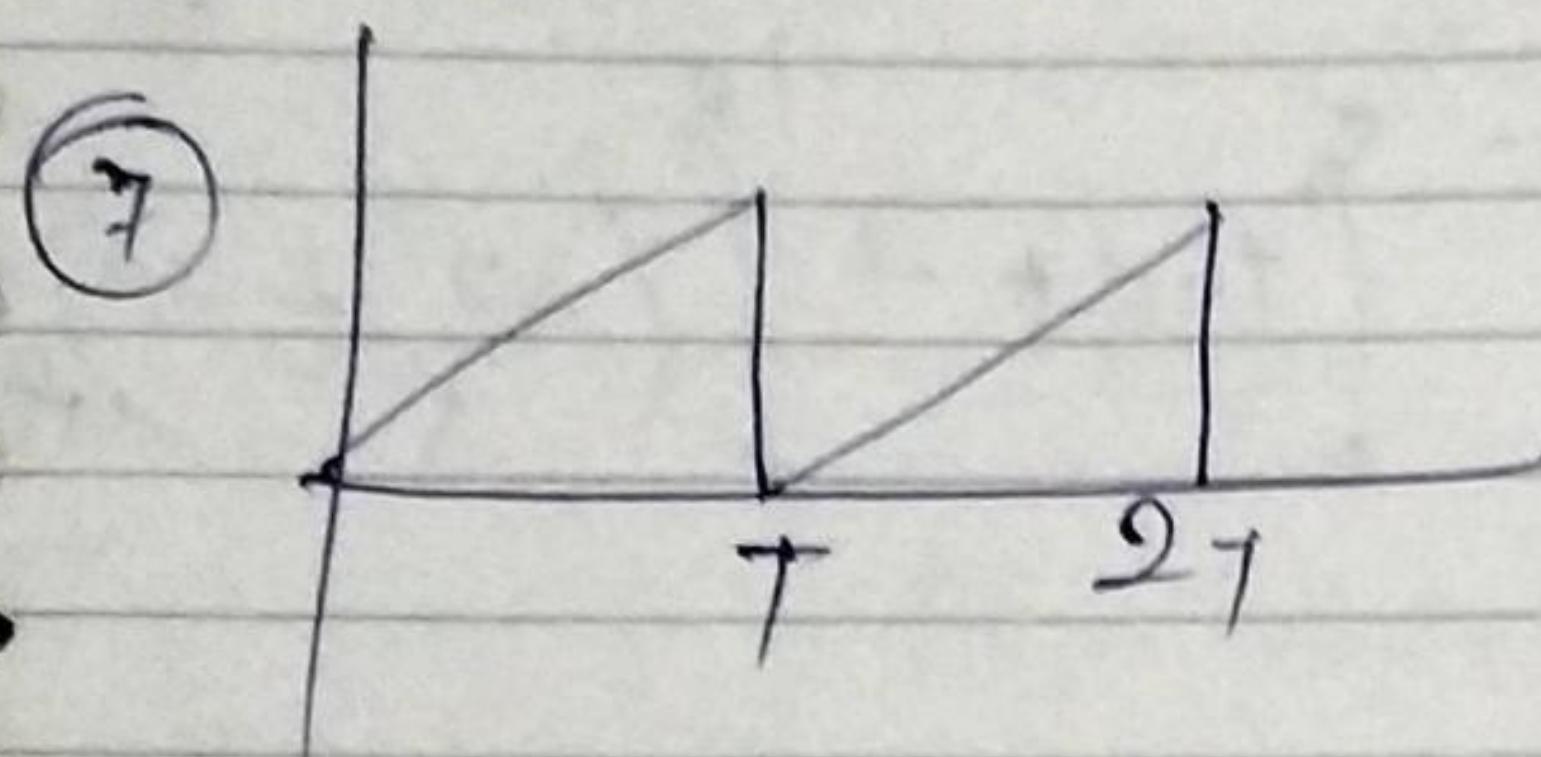
$$\text{RMS} \\ V_m$$

$$\text{Average} \\ V_m$$



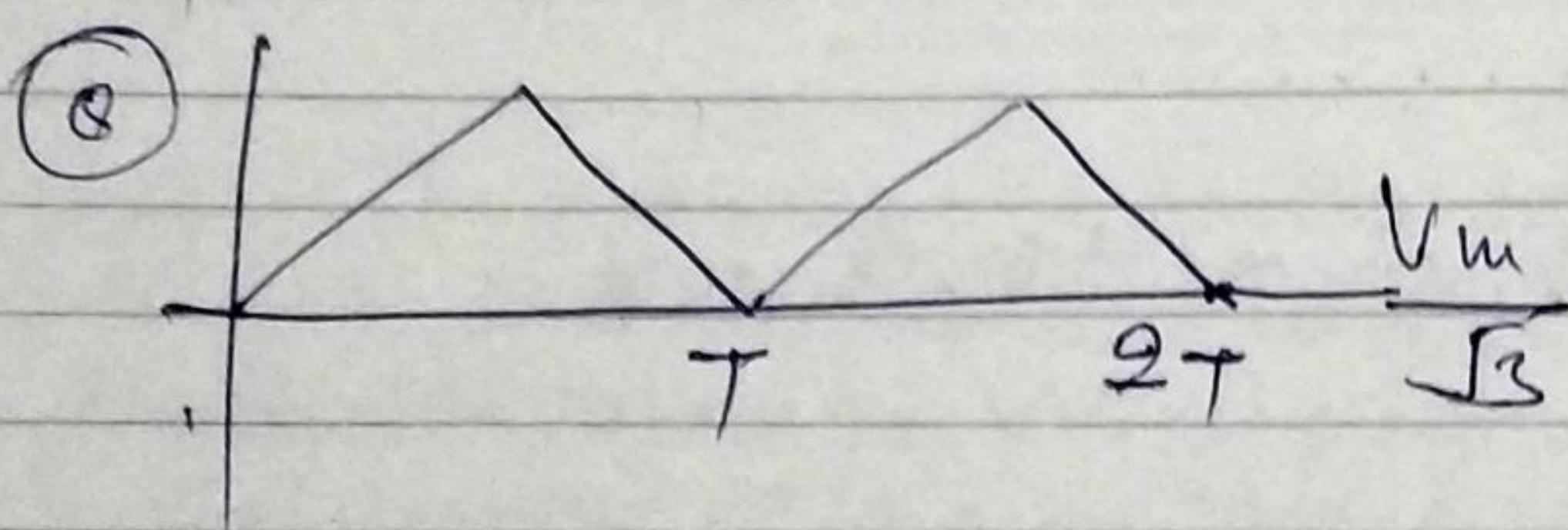
$$V_m \\ \frac{T}{2}$$

$$V_m \\ \frac{2}{2}$$

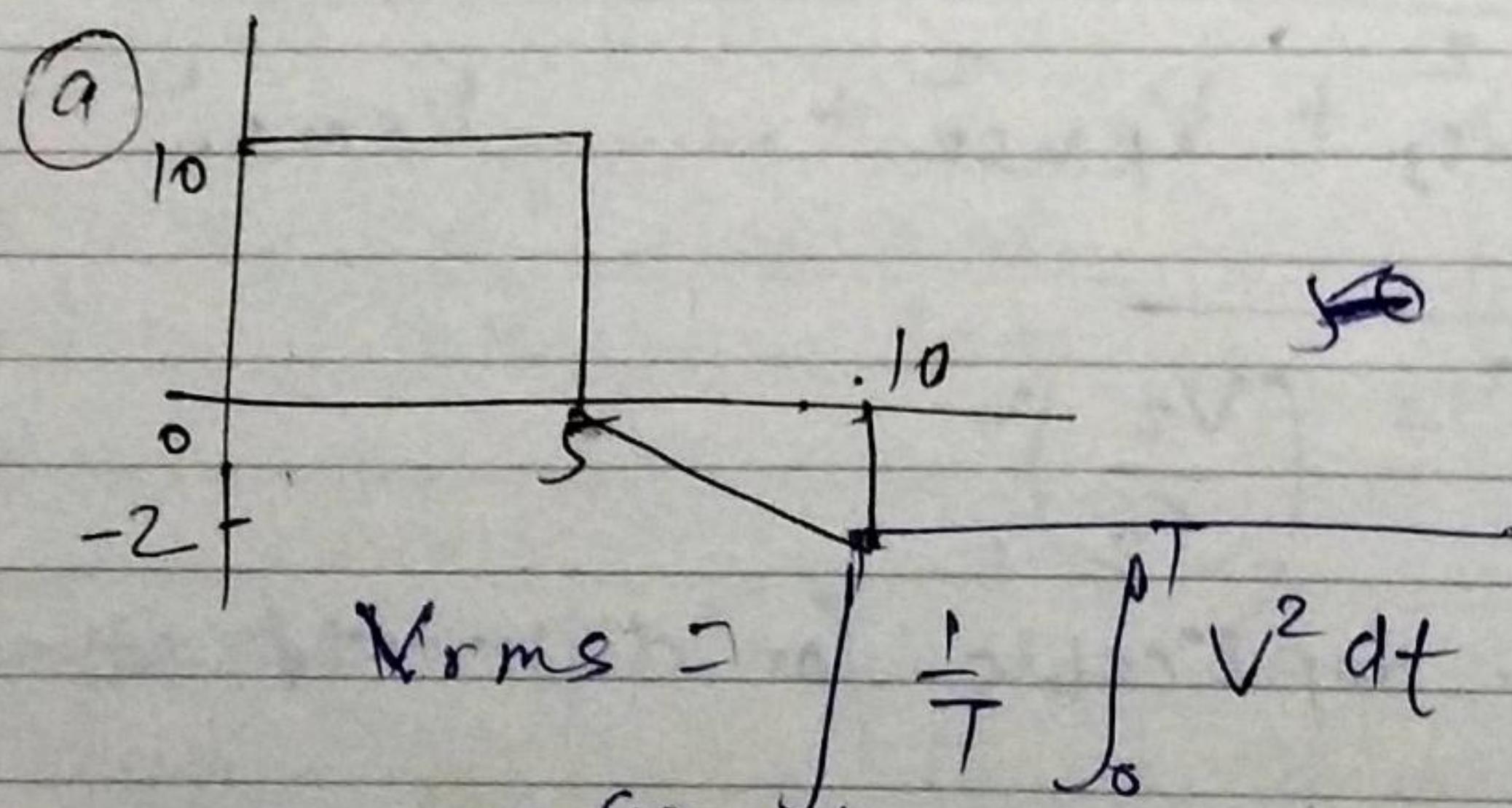


$$V_m \\ \frac{\sqrt{3}}{2}$$

$$V_m \\ \frac{2}{2}$$



$$V_m \\ \frac{2}{2}$$



$$\text{RMS} = \sqrt{\frac{1}{T} \int_0^T V^2 dt}$$

$$v(t) = 10 \quad 0 \leq t \leq 5$$

$$y = mx + c$$

$$v(t) = -\frac{2}{5}t + c \quad \text{at } t=5 \quad v(5) = 0 \\ 0 = -\frac{2}{5}(5) + c \quad c = 2$$

$$v(t) = -\frac{2}{5}t + 2$$

$$V_{rms} = \sqrt{\frac{1}{10} \left[\int_0^{10} (10)^2 dt + \int_5^{10} \left(\frac{2}{5}t + 2\right)^2 dt \right]}$$

$$V_{rms} = \underline{\underline{7.11}}$$

$$V_{av} = \frac{1}{T} \int_0^T v dt = \frac{1}{10} \left[\int_0^5 10 dt + \int_5^{10} \left(-\frac{2}{5}t + 2\right) dt \right]$$

$$= \underline{\underline{4.5}}$$

Ques: $v(t) = V_0 + V_1 \sin \omega t + V_2 \sin 3\omega t$

$$V_{av} = V_0$$

$$V_{rms} = \sqrt{V_{rms_1}^2 + V_{rms_2}^2 + \dots + V_{rms_n}^2} \\ = \sqrt{V_0^2 + \left(\frac{V_1}{\sqrt{2}}\right)^2 + \left(\frac{V_2}{\sqrt{2}}\right)^2}$$

Only applicable for different frequencies

Ques:- $v(t) = V_0 + V_1 \sin(\omega t + \phi_1) + V_2 \sin(\omega t - \phi_2)$

$V_1 < \phi_1 \times \text{change in polar}$ $V_2 < \phi_2$

$$v(t) = V_0 + (V_L \angle \phi_1 + V_B \angle \phi_2)$$

$$= V_0 + V \angle \phi$$

$$v(t) = V_0 + V \sin(\omega t + \phi)$$

$$V_{rms} = \sqrt{V_0^2 + \left(\frac{V}{\sqrt{2}}\right)^2}$$

Applicable only
same frequency
are given.

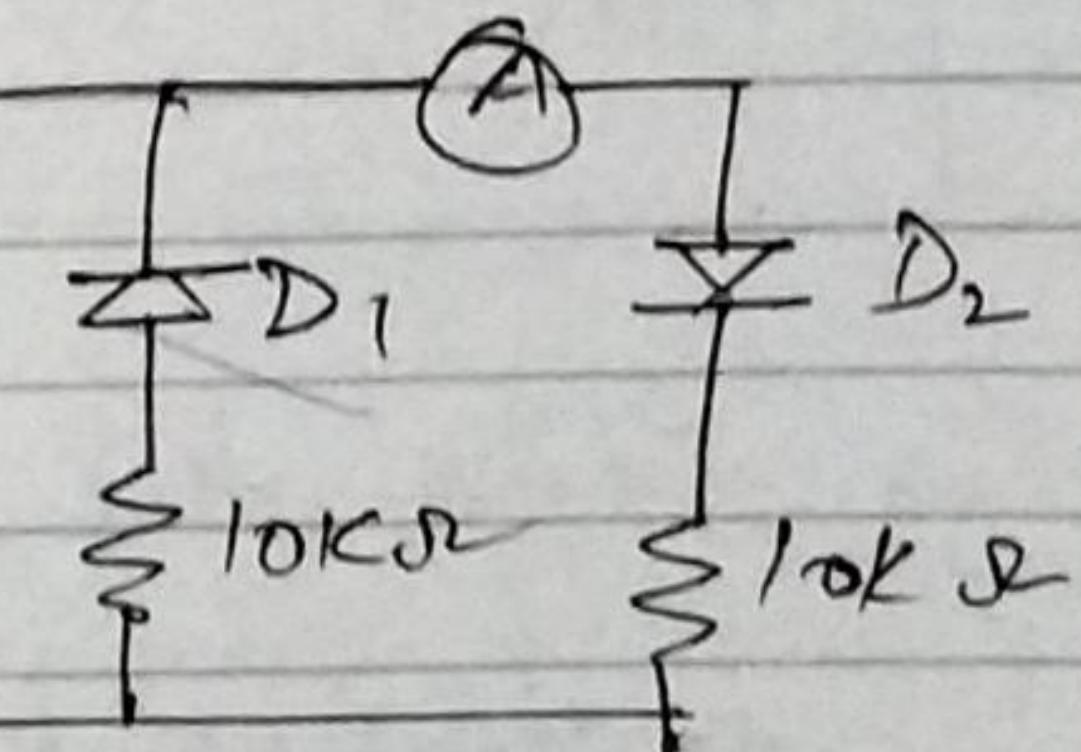
$$P_{RMS} = \frac{V_{rms}^2}{R} \Rightarrow \text{Real form system does not exist.}$$

$$P_{av} = \frac{V_{RMS}^2}{R}$$

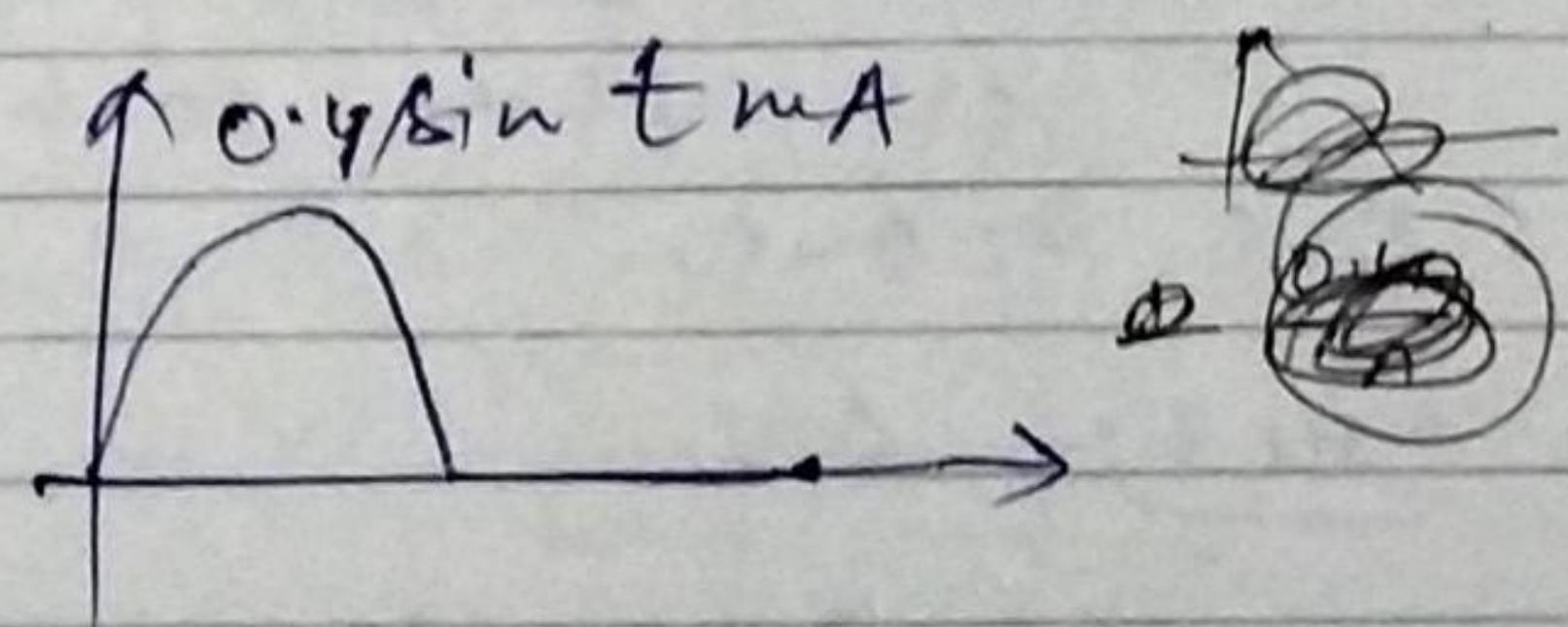
$$P_{DC} = I_{av}^2 R$$

$$P_{av} = I_{RMS}^2 R$$

Ques: The given ckt is having ideal diodes and an ~~Ammeter~~ which ~~gives~~ reads the average value find the reading of the ~~emitter~~ Ammeter.



$$P = v(t) = \frac{4 \sin \omega t}{10k}$$



$$I_{av} = \frac{0.4}{\pi} \text{ mA}$$

Ques Find the power across the element if the voltage is

$$v(t) = 100 \sin \omega t + 40 \cos(3 \omega t - 30^\circ) + 5 \sin(5 \omega t + 45^\circ)$$

and the current through it is

$$i(t) = 8 \sin(\omega t + 10^\circ) + 6 \sin(3\omega t + 40^\circ) + 2 \cos(5\omega t - 120^\circ)$$

$$v_f = 100 \sin \omega t + 40 \cos(3\omega t - 30^\circ) + 5 \sin(5\omega t + 45^\circ) \\ + 40 \sin(3\omega t + 60^\circ)$$

$$i(t) = 8 \sin(\omega t + 10^\circ) + 6 \sin(3\omega t + 40^\circ) + 2 \sin(5\omega t - 30^\circ)$$

$$P_{av} = \frac{100}{\sqrt{2}} \times \frac{8}{\sqrt{2}} \times \cos(-10^\circ) + \frac{40}{\sqrt{2}} \times \frac{6}{\sqrt{2}} \cos 20^\circ + \\ \frac{5}{\sqrt{2}} \times \frac{2}{\sqrt{2}} \cos 75^\circ$$

$$\boxed{P_{av} = 507.98 \text{ W}}$$

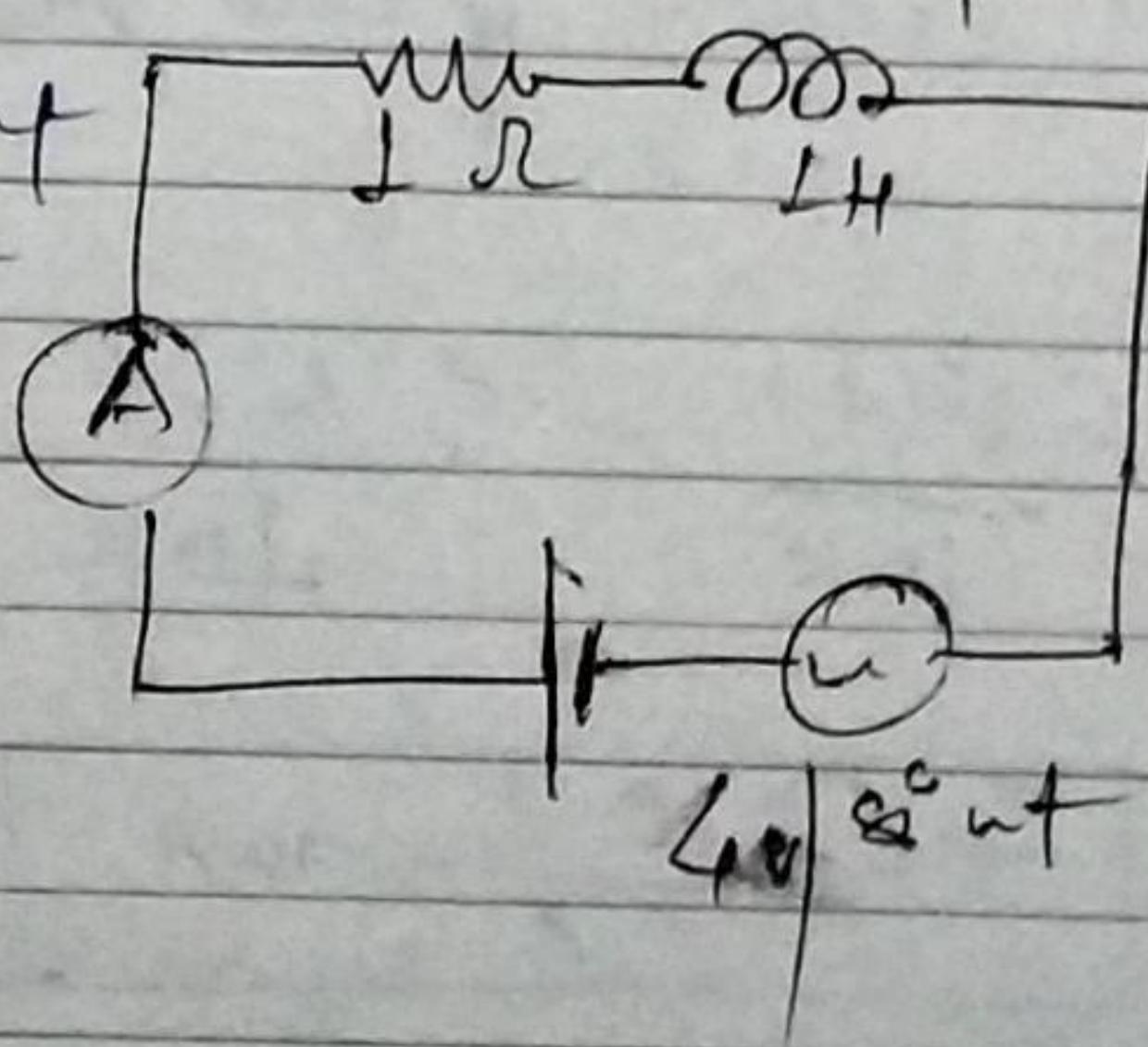
Now find the reading of the ammeter and pf of the circuit.

As the different frequency components are there so we take one by one.

Case 1 :- 4V

AS DC $L \rightarrow \infty$

$$I = \frac{4}{L} = 4 \text{ A}$$



Case 2

$$z = R + j \omega L = (1 + j) \omega r$$

* When over different frequency component are there in the circuit we should calculate $P_f = \frac{P}{S}$ we can't agree $P_f = \frac{R}{Z}$ bcz in each case Z is different.

$$I_{AC} = \frac{1}{\sqrt{2}} = \frac{1}{2} A$$

$$I_{RMS} = \sqrt{I_{RMS_1}^2 + I_{RMS_2}^2}$$

$$= \sqrt{4^2 + \frac{1}{4}} = \sqrt{16.25} A$$

$$P_f (Case) = \frac{P}{S} = \frac{I_{RMS}^2 R}{V_{RMS} I_{RMS}}$$

$$V_{RMS} = \sqrt{(4)^2 + \left(\frac{1}{2}\right)^2} = \sqrt{16.5} V$$

$$\cos \theta = \frac{\sqrt{16.25} \times 1}{\sqrt{16.5}} \quad \boxed{\cos \theta = 0.9923}$$

$$\boxed{\theta = 7.07^\circ}$$

Ans:

~~$V_{RMS} = \sqrt{\left(\frac{10}{\sqrt{2}}\right)^2 + \left(\frac{20}{\sqrt{2}}\right)^2}$~~

$10 \sin 4t = v_1(t)$

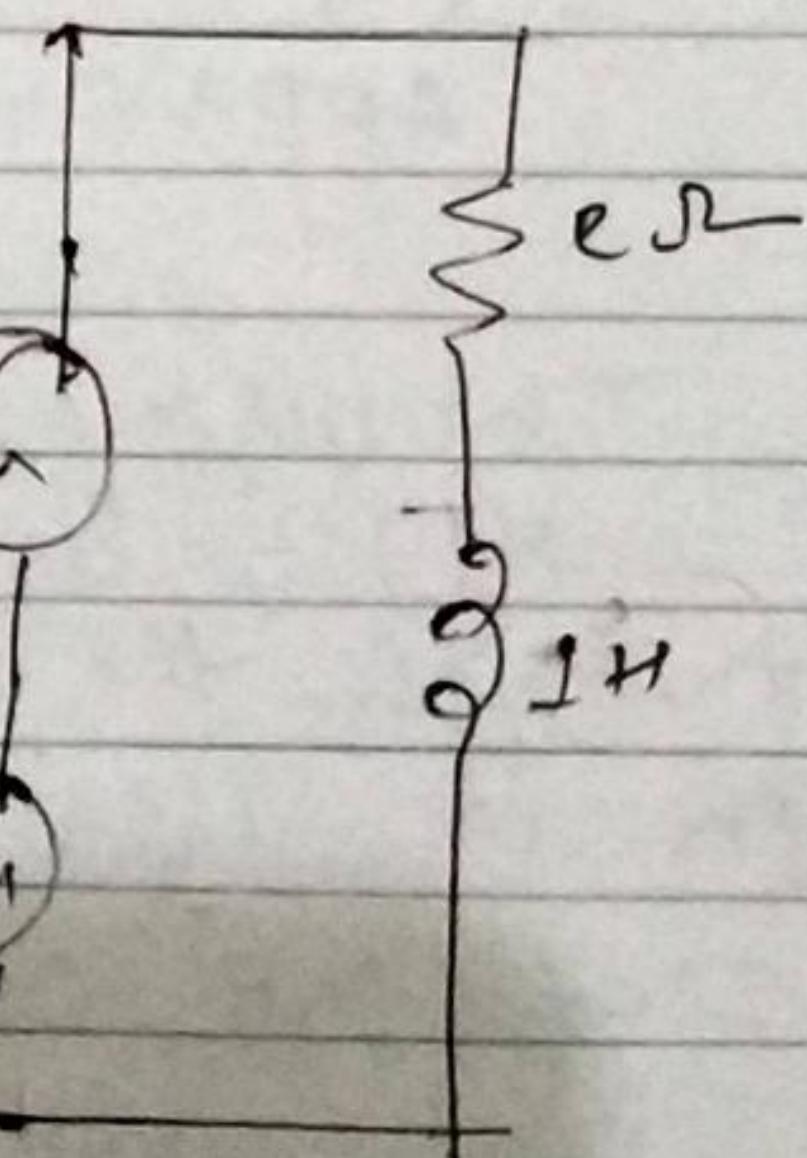
case(i)

$v_1(t) = 10 \sin 4t$

$2 \sin 2t = v_2(t)$

$Z = R + j\omega L$

$Z = 2 + 4j$



$i_1(t) = \frac{10 \sin 4t}{(2+4j)} = \frac{10 \angle 0^\circ}{4.47 \angle 63.43^\circ}$

$i_1(t) = 2.23 \sin(4t - 63.43^\circ)$

$$\text{Case iii) } v_i(t) = 20 \sin 2t$$

$$Z = R + j\omega L$$

$$= 2 + 2j \quad Z = 2\sqrt{2} \angle 45^\circ$$

$$I = \frac{20 \sin 2t}{2\sqrt{2} \angle 45^\circ} = \frac{20 \angle 0^\circ}{2\sqrt{2} \angle 45^\circ} = \frac{7.07 \angle 0^\circ}{\angle 45^\circ} = 7.07 \angle -45^\circ$$

$$P_{av} = \frac{10}{\sqrt{2}} \times \frac{2.23}{\sqrt{2}} \times \cos(-63.43) + \frac{20}{\sqrt{2}} \times \frac{7.07}{\sqrt{2}} \times \cos(-45^\circ)$$

$$= 55 \text{ W}$$

Complex power: $S = V I^*_{rms} \cos \theta$

$$S = |S| \angle \theta$$

Apparent power

Ques The voltage across the load is

$v(t) = 60 \cos(100t - 10^\circ)$ and the current through the load is in the direction of voltage drop i.e. $i(t) = 1.5 \cos(\omega t + 50^\circ)$

- ① Find the complex and apparent power.
- ② Real and reactive power.

$$V(t) = \frac{60}{\sqrt{2}} \angle -10^\circ \quad i(t) = \frac{1.5}{\sqrt{2}} \angle 50^\circ$$

$$S = V I^*_{rms} \cos \theta$$

$$S = \frac{60}{\sqrt{2}} \angle (-10^\circ) \times \frac{1.5}{\sqrt{2}} \angle (-50^\circ)$$

$$S = 45 \angle -60^\circ \quad |S| = 45 \text{ VA}$$

$$S = 45 \angle -60^\circ$$
$$= P + jQ$$

$$= \boxed{22.5 \text{ W}} + j \boxed{\frac{30.97 \text{ VAR}}{Q}}$$

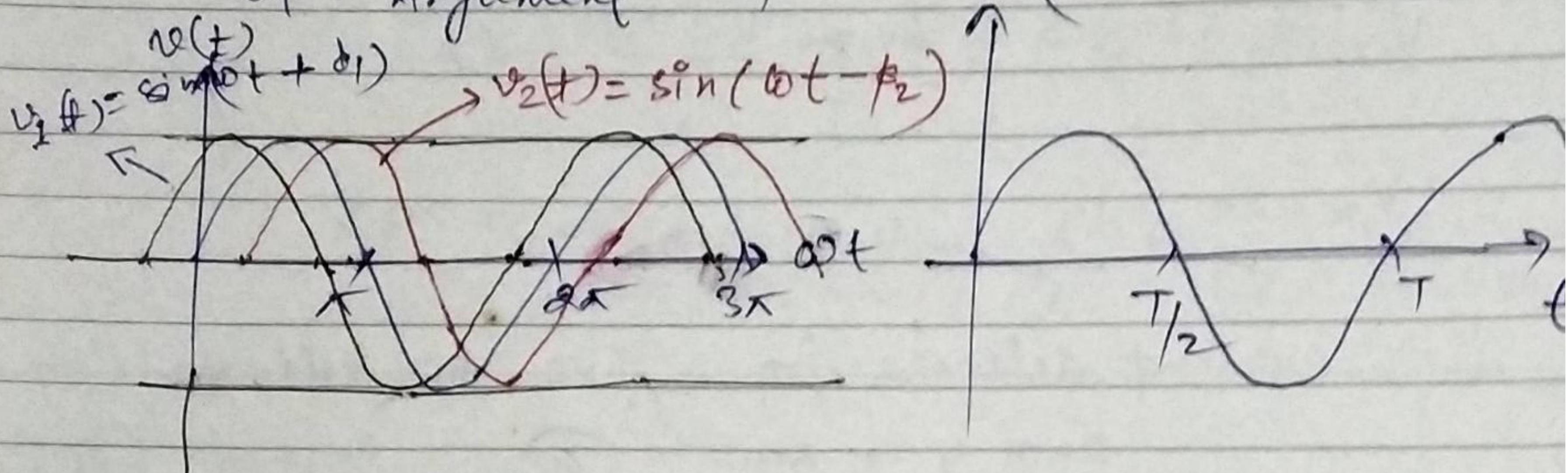
Sinusoidal Steady State Analysis :-

Sinusoidal \rightarrow Sine \rightarrow cosine

$$v(t) = V_m \sin(\omega t + \phi) \quad V_m \rightarrow \text{Maj. amplitude}$$

$\omega \rightarrow$ Angular frequency (Rad/sec)

$\phi \rightarrow$ Argument

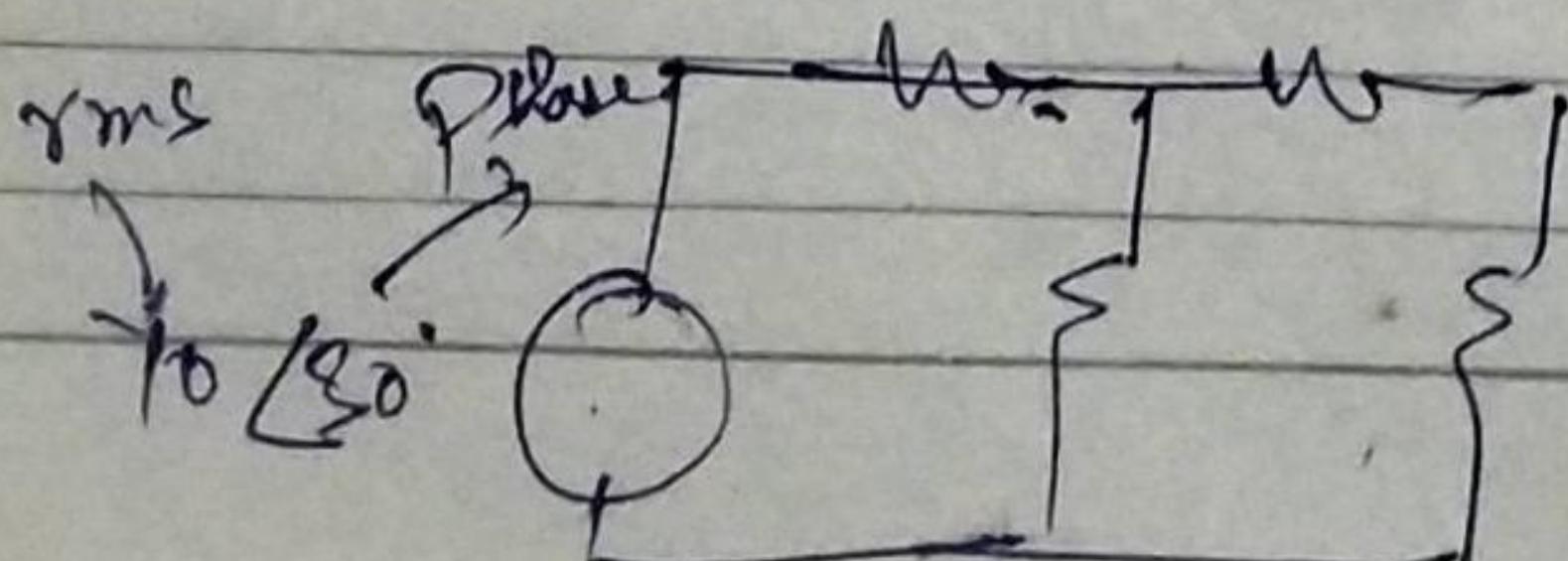


- * We can compare to sine wave providing ϕ is —
- * Both v_1 and v_2 have same frequency.
- * Both v_1 and v_2 are expressed in form of either sine function or cosine function.
- * Both v_1 and v_2 are written with +ve amplitude.

$$v(t) = V_m \cos(\omega t + \phi) = \operatorname{Re} \{ V_m e^{j(\omega t + \phi)} \}$$

$$= \operatorname{Re} \{ V_m e^{j\phi} \cdot e^{j\omega t} \}$$

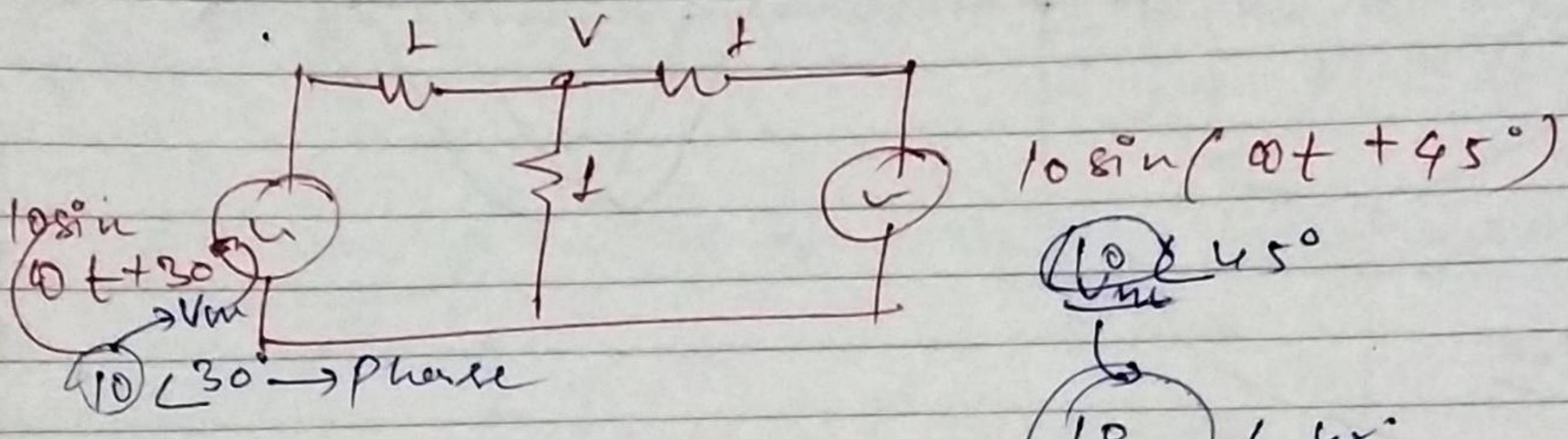
$$V_m e^{j\phi} = V_m \angle \phi$$



$$V_m \cos(\omega t + \phi)$$

$$e^{j\theta} = \cos\theta + j\sin\theta$$

$$= \sqrt{\cos^2\theta + \sin^2\theta} \angle \tan^{-1}\left(\frac{\sin\theta}{\cos\theta}\right) = +1\angle\theta$$



$$\left(\frac{10}{\sqrt{2}}\right) \angle 30^\circ \text{ up phase}$$

$$V = \underbrace{10 \angle 30^\circ}_1 + \underbrace{V}_1 + \underbrace{V - 10 \angle 45^\circ}_1 = 0$$

$$3V = 10 \angle 30^\circ + 10 \angle 45^\circ \quad V = \boxed{\frac{20}{\sqrt{3}}} \angle 37.5^\circ$$

$$\boxed{V = \frac{20}{\sqrt{3}} \sin(\omega t + 37.5^\circ)}$$

↓
 V_m ↓
phase

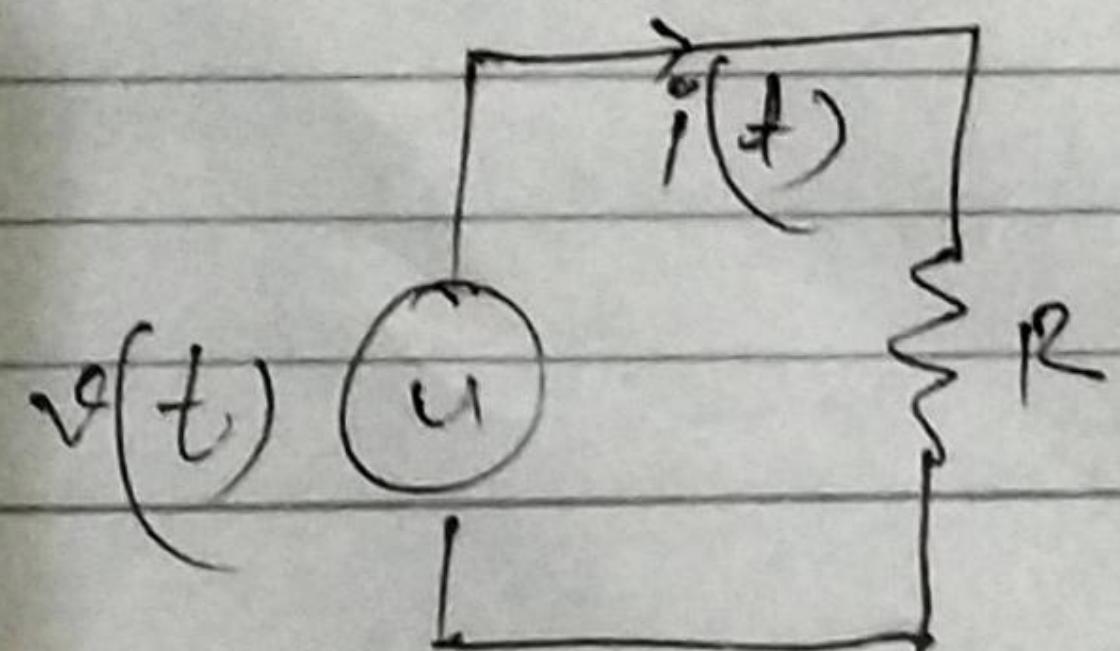
By suppressing the time factor we transform the sinusoids from the time domain to phasor domain thus the phasor transform, transforms the sinusoidal signal from the time domain to the complex domain which is also called as the frequency domain.

$$e^{j\theta} = 1 \angle 0^\circ \quad e^{j\omega t}$$

Difference b/w $v(t)$ & \bar{V} :

- $v(t)$ is the instantaneous or time domain representation while \bar{V} is the frequency or phasor domain representation.
- $v(t)$ is time dependent while \bar{V} is not.
- $v(t)$ is always real while \bar{V} is generally complex.

1. Resistor :-



$$v(t) = V_m \sin \omega t$$

$$i(t) = \frac{V_m \sin \omega t}{R}$$

$$i(t) = \frac{V_m}{R} \sin \omega t$$

$$\xrightarrow{v(t)} \xrightarrow{i(t)}$$

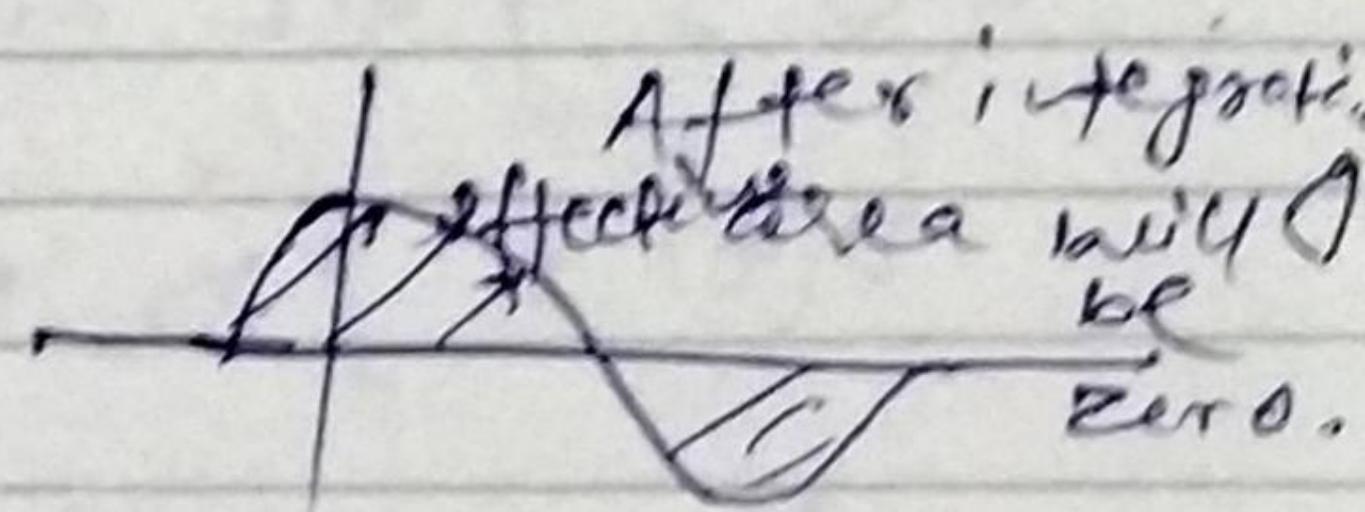
$$P_{av} = \frac{1}{T} \int_0^T v(t) \cdot i(t) dt = \frac{1}{T} \int_0^T (V_m \sin \omega t) \cdot (I_m \sin \omega t) dt$$

$$= V_m I_m \frac{1}{T} \int_0^T \sin^2 \omega t dt = V_m I_m \frac{1}{2T} \int_0^T (1 - \cos 2\omega t) dt$$

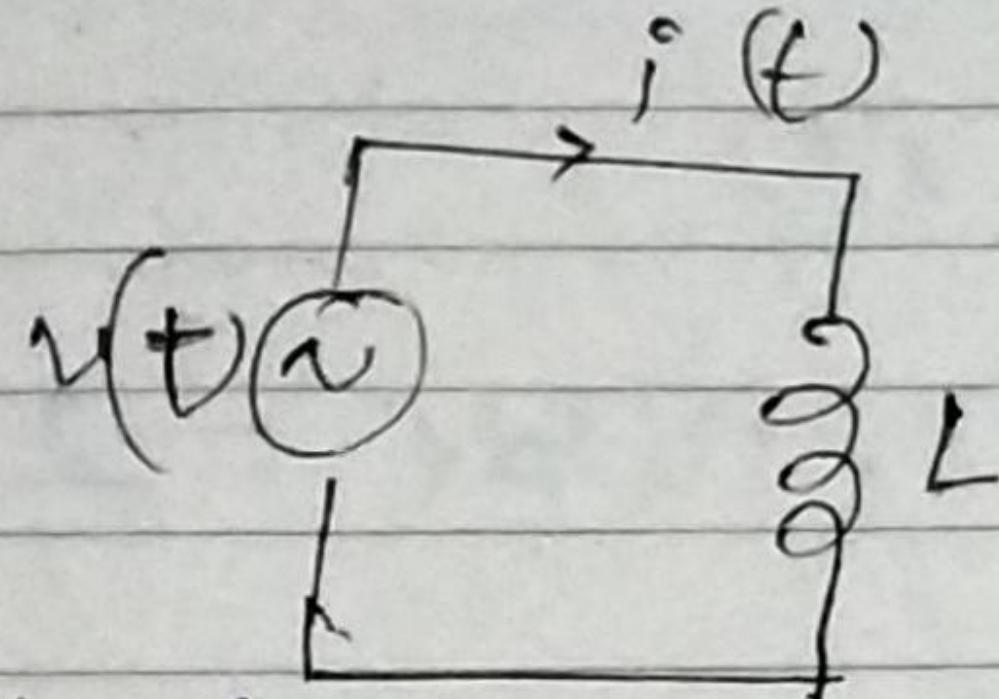
$$= V_m I_m \frac{1}{T} \left[\int_0^T 1 dt - \int_0^T \frac{\cos 2\omega t}{2} dt \right]$$

$$P_{av} = \frac{V_m I_m}{2} = \frac{V_m}{\sqrt{2}} \cdot \frac{I_m}{\sqrt{2}}$$

$$\boxed{P_{av} = V_{rms} \cdot I_{rms}}$$



a. Inductor:



$$i(t) = I_m \sin \omega t$$

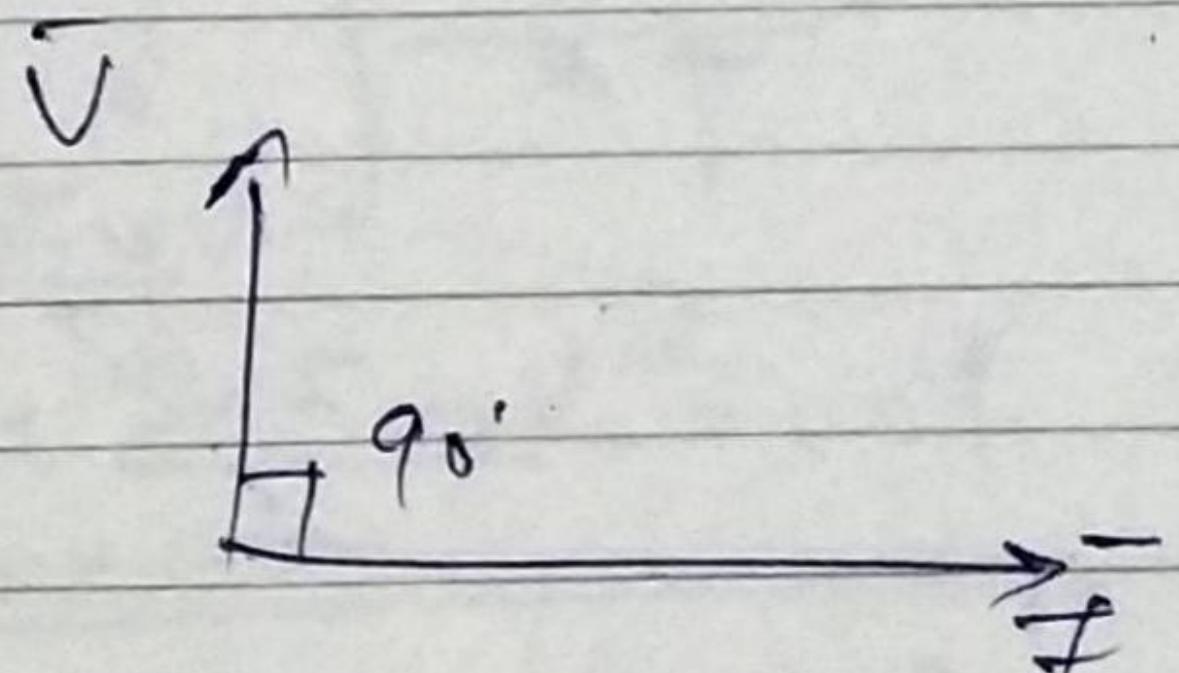
$$v(t) = L \frac{di}{dt} = L \frac{d}{dt} (I_m \sin \omega t)$$

$$= L I_m \omega \cos \omega t = \omega L I_m \sin(\omega t + 90^\circ)$$

$$= \omega L I_m \sin(\omega t) \cdot j$$

$$v(t) = j \omega L I_m \sin \omega t$$

$$v(t) = j X_L I_m \sin \omega t$$



$X_L \rightarrow$ Inductive Reactance

$$X_L = \omega L$$

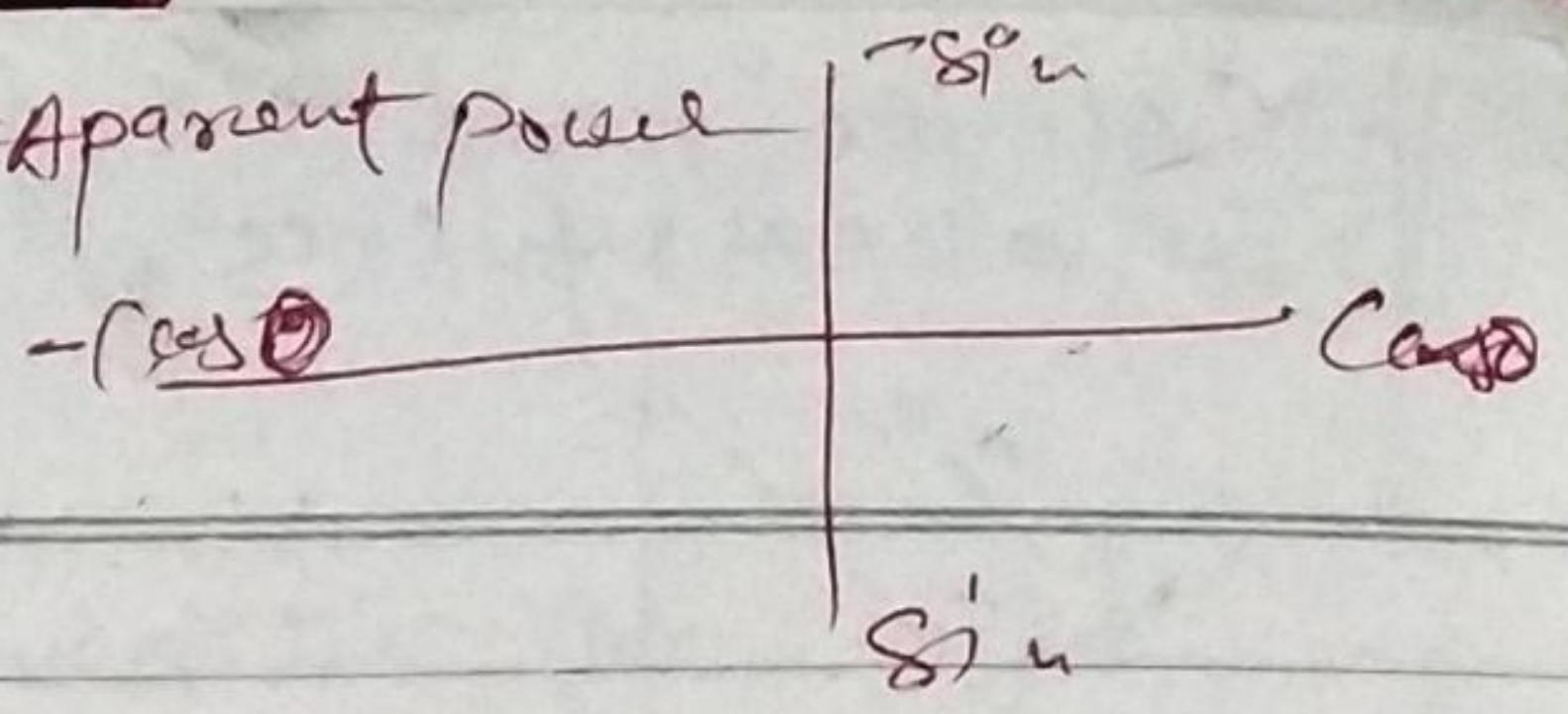
$$P_{av} = \frac{1}{T} \int_0^T v(t) i(t) dt$$

$$= \frac{1}{T} \int_0^T V_m \cos \omega t \cdot I_m \sin \omega t dt$$

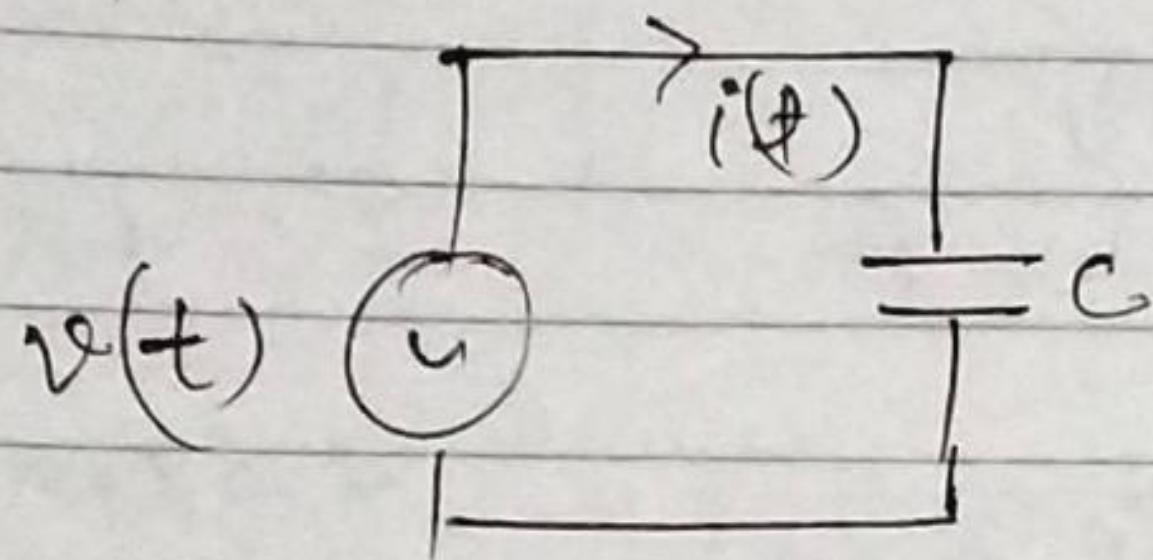
$$= \frac{V_m I_m}{2T} \int_0^T \sin 2\omega t dt$$

$$\boxed{P_{av} = 0}$$

magnitude of complex power = Apparent power



(3) Capacitor



$$v(t) = V_m \sin \omega t$$

$$i = C \frac{d}{dt} (V_m \sin \omega t)$$

$$i = \omega C V_m \cos \omega t$$

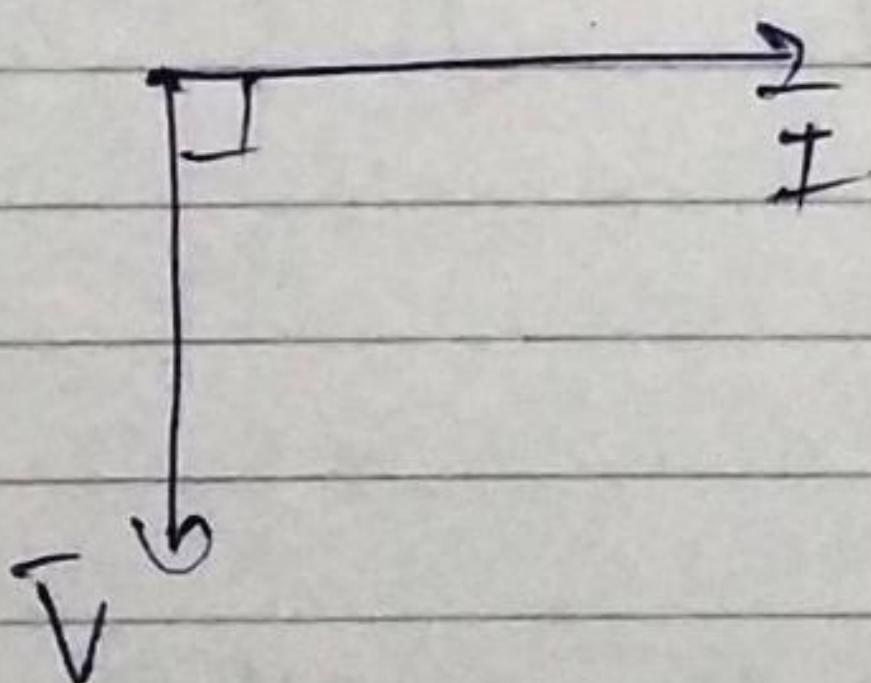
$$i = \omega C V_m \sin (\omega t + 90^\circ)$$

$$i = j \omega C V_m \sin \omega t$$

$$i = j \frac{1}{\omega C} V_m \sin \omega t, Z = \frac{1}{\omega C}$$

$$X_C = \frac{1}{\omega C} \text{ Capacitive Reactance}$$

$$P_{av} = 0$$



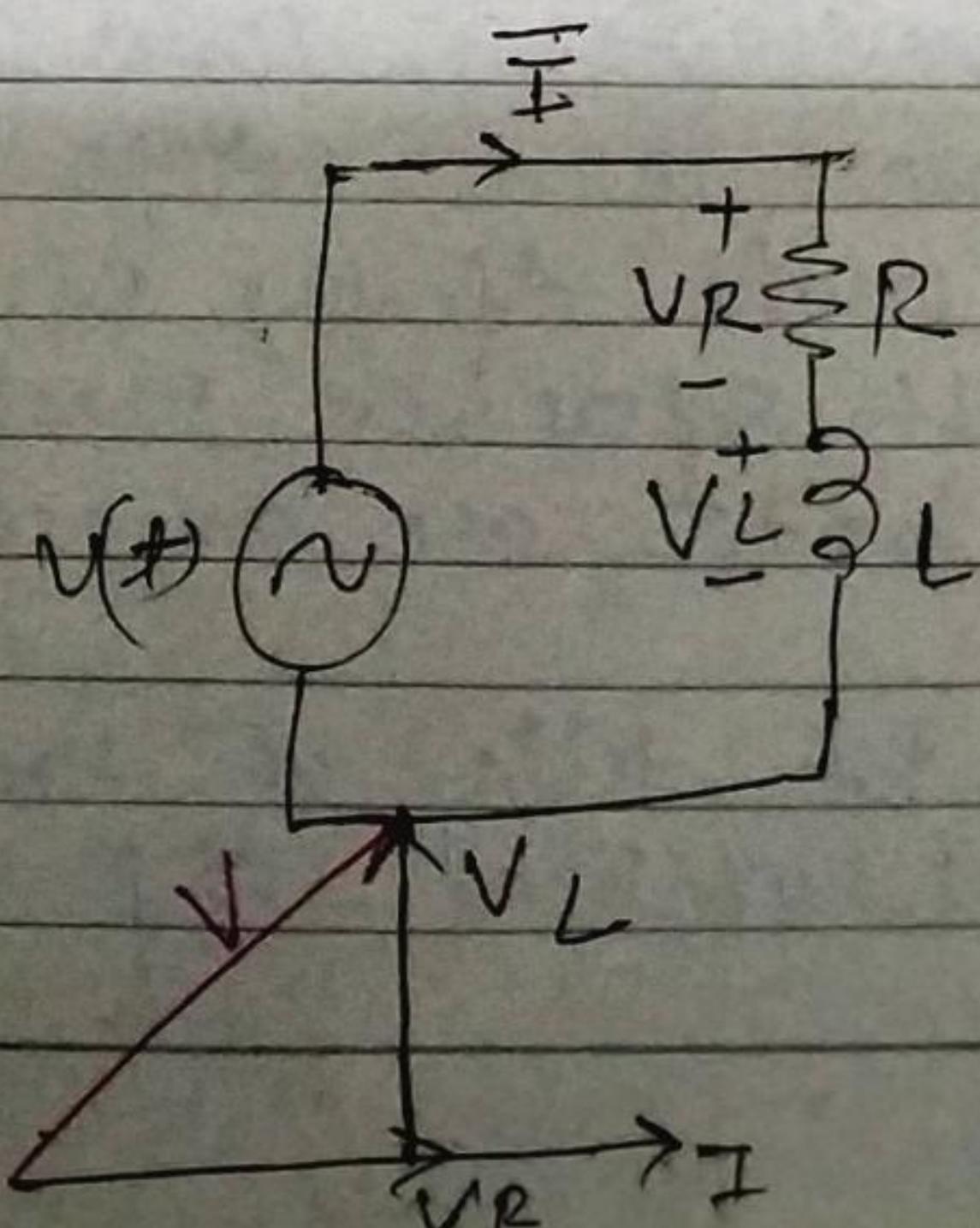
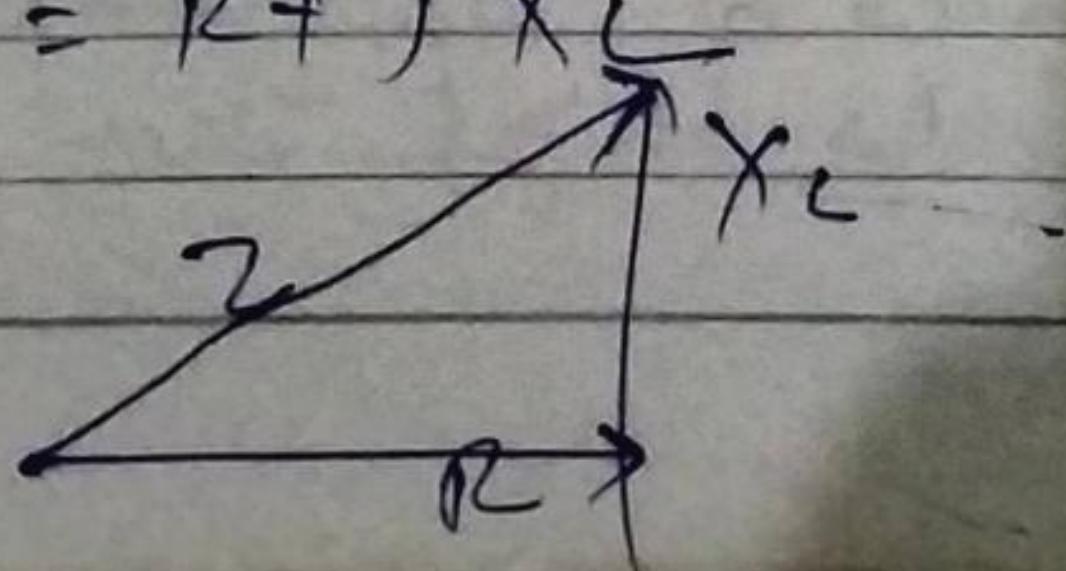
$$f_p = 2 f_v / i$$

(4) Series R L Circ:-

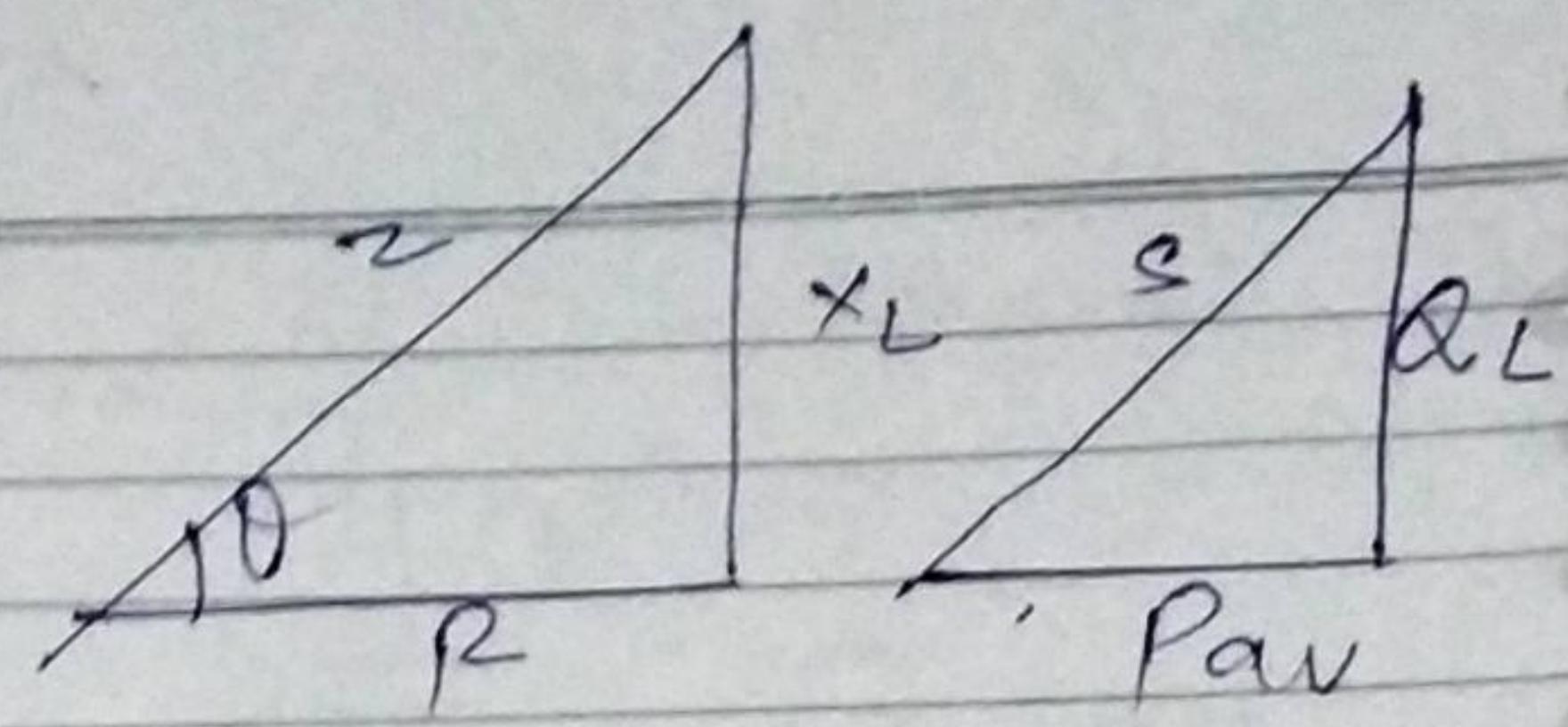
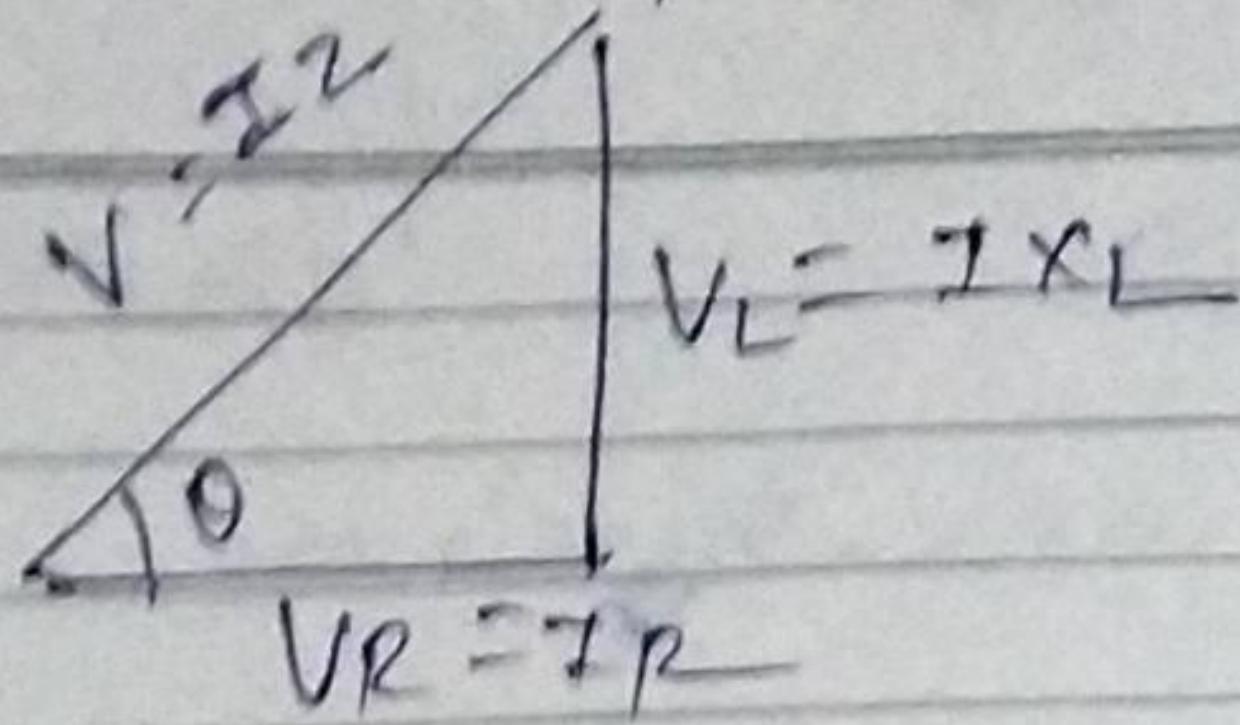
$$V = V_R + V_L$$

$$\bar{I}Z = \bar{I}R + \bar{I}(jX_L)$$

$$Z = R + jX_L$$



∴ All the loads are in II so voltage is same.
We take V_R as reference.



$$V = \sqrt{V_R^2 + V_L^2}$$

$$\theta = \tan^{-1} \left(\frac{V_L}{V_R} \right)$$

$$\cos \theta = \frac{V_R}{V} \quad (\text{lagging})$$

$$P_{av} = \frac{1}{T} \int_0^T V(t) i(t) dt$$

$$= \frac{1}{T} \int_0^T V_m \sin(\omega t + \phi) \cdot I_m \sin(\omega t) dt$$

$$= \frac{V_m I_m \cos \theta}{2} = \frac{V_m}{\sqrt{2}} \times \frac{I_m}{\sqrt{2}} \cos \theta$$

$$z = \sqrt{R^2 + X_L^2}$$

$$\theta = \tan^{-1} \left(\frac{X_L}{R} \right)$$

\rightarrow Apparent power

$$S = \sqrt{P_{av}^2 + Q_{av}^2}$$

$$\cos \theta = \frac{P}{S}$$

$$\theta = \tan^{-1} \frac{Q_{av}}{P}$$

$$\cos \theta = \frac{P}{S}$$

$$(\text{lagging})$$

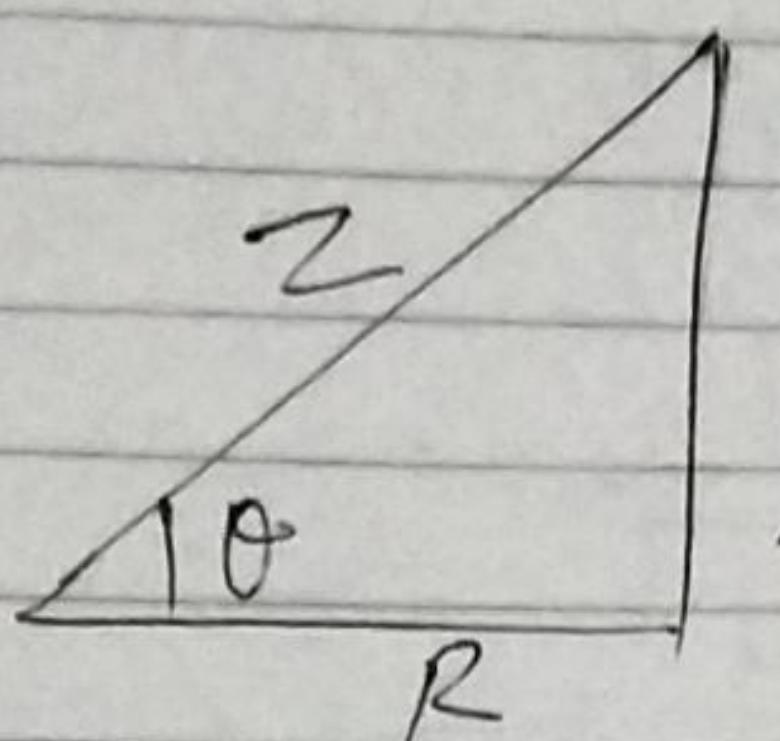
$$P_{av} = V_{rms} I_{rms} \cos \theta$$

Power factor :- power factor indicates position of current phasor w.r.t voltage phasor.

* While defining P.F. voltage is always taken as the reference phasor bcz in real time system loads are connected in parallel.

* P.F tells us how much of Apparent power is utilized. If P.F is high then Active

power delivered to the load is also high.



$$\tan \theta = \frac{X_L}{R}$$

$$Z = 10 \angle (-45^\circ)$$

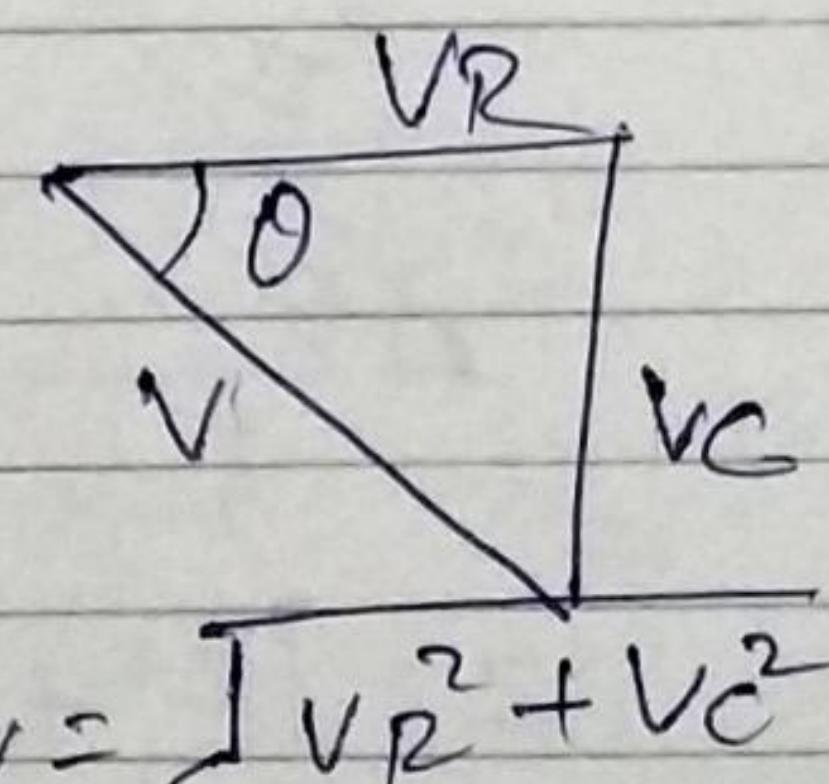
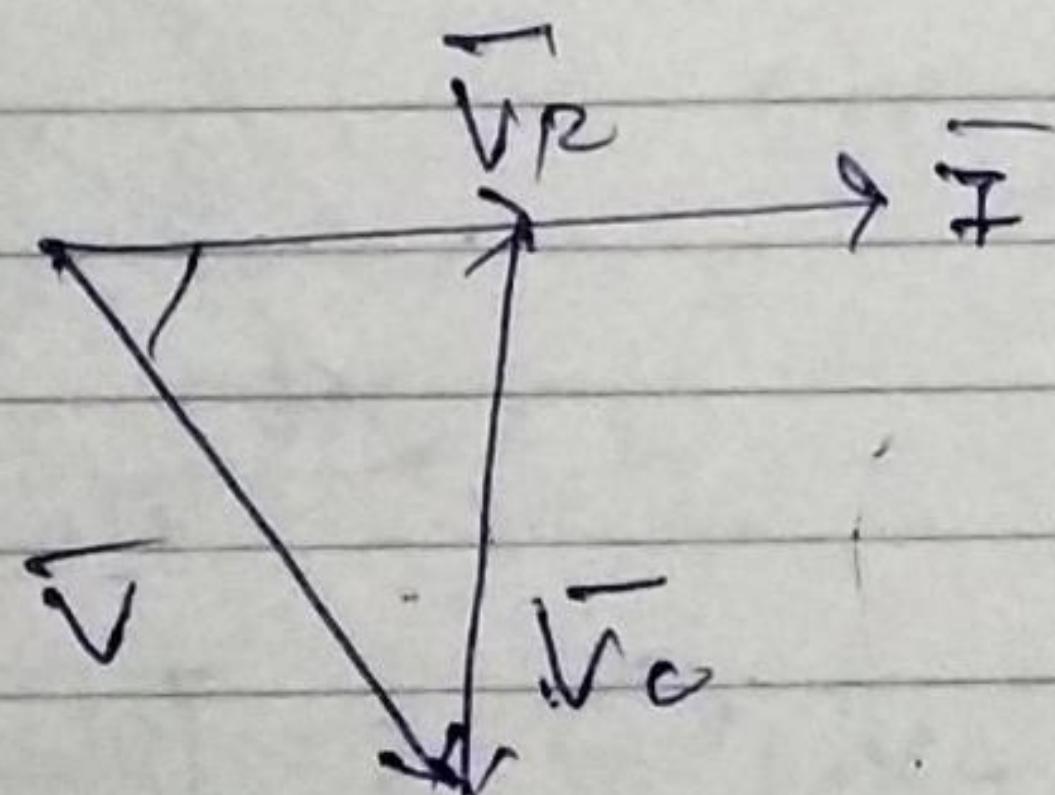
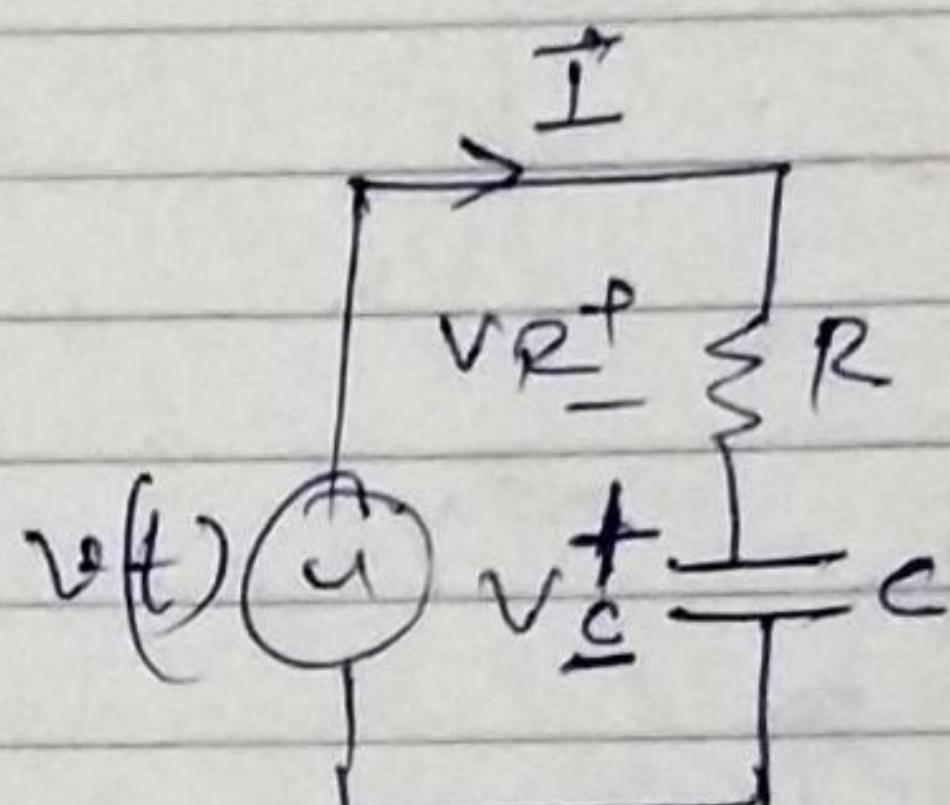
$$I = \frac{V}{|Z|L_0} = \frac{V}{10} \angle (-\theta)$$

impedance angle and pf angle shows opposite nature.

(5) Series RC Ckt

$$V = V_R + V_C$$

$$I \equiv I_R + j X_C I$$



$$\tan \theta = -\frac{V_C}{V_R}$$

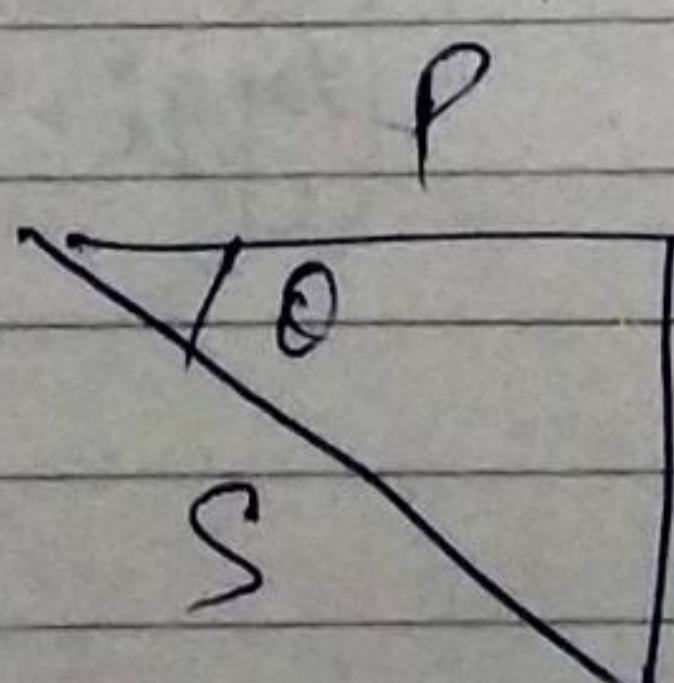
$$\tan \theta = -\frac{X_C}{R}$$

$$\cos \theta = \frac{V_R}{V}$$

(leading)

$$\cos \theta = \frac{R}{Z}$$

(leading)



$$S = \sqrt{P^2 + Q_C^2}$$

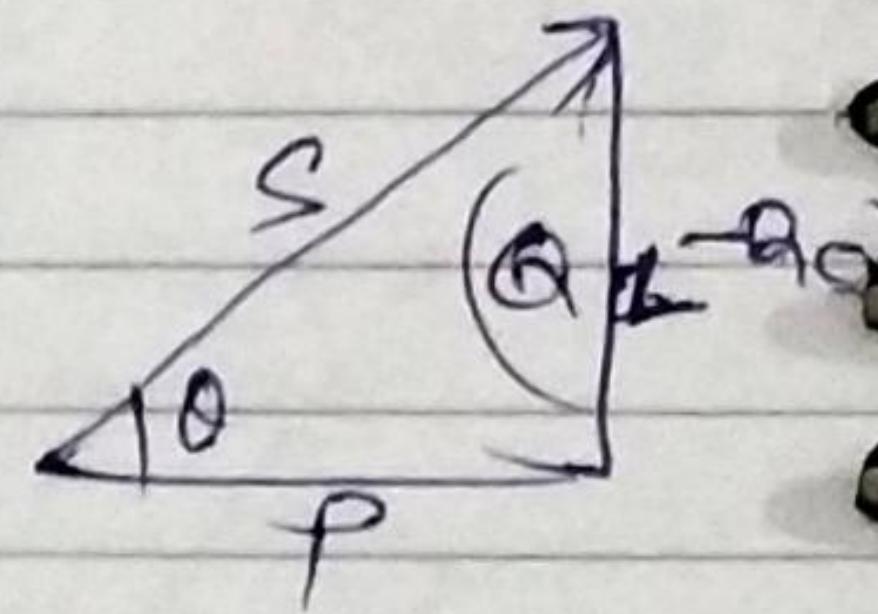
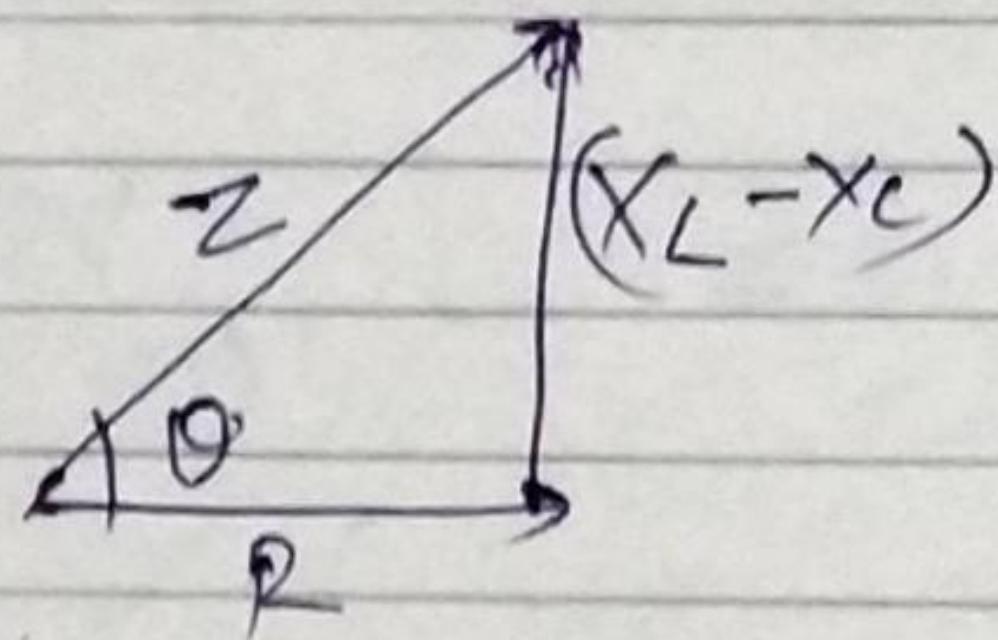
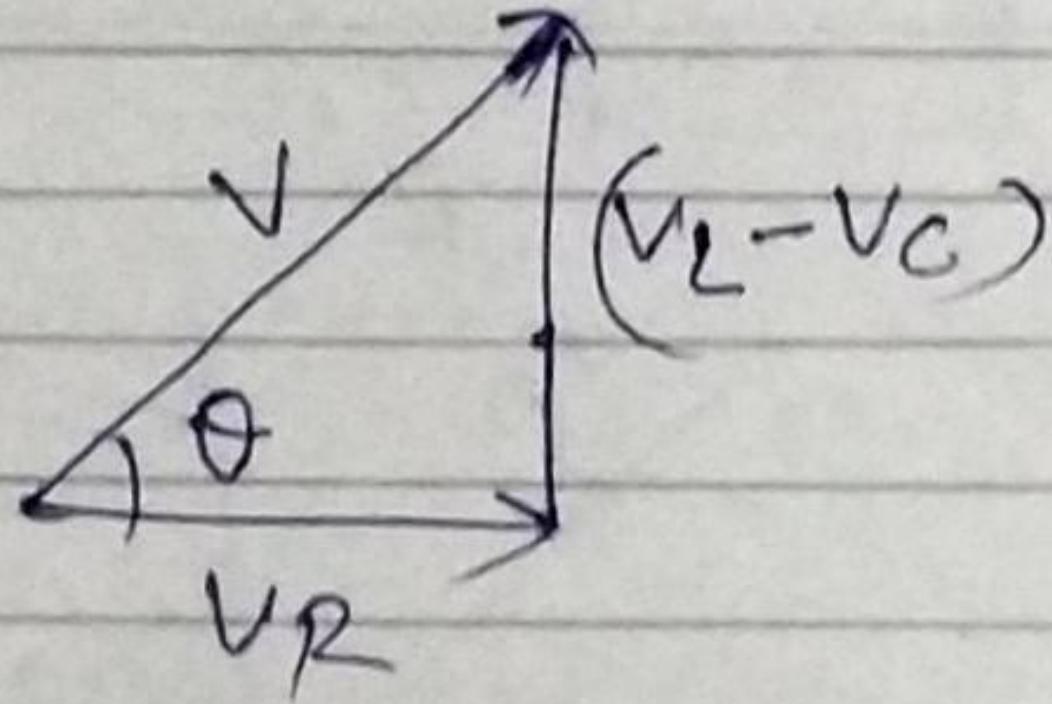
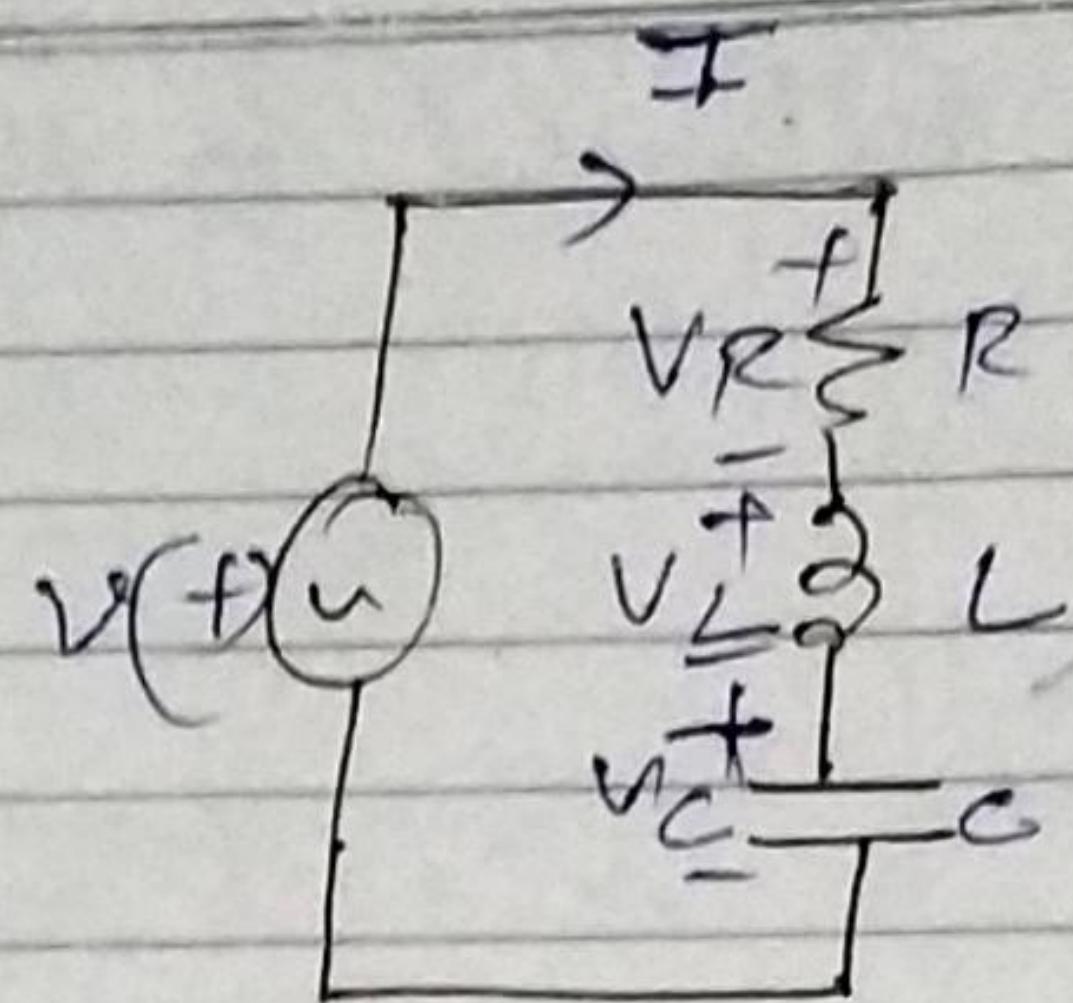
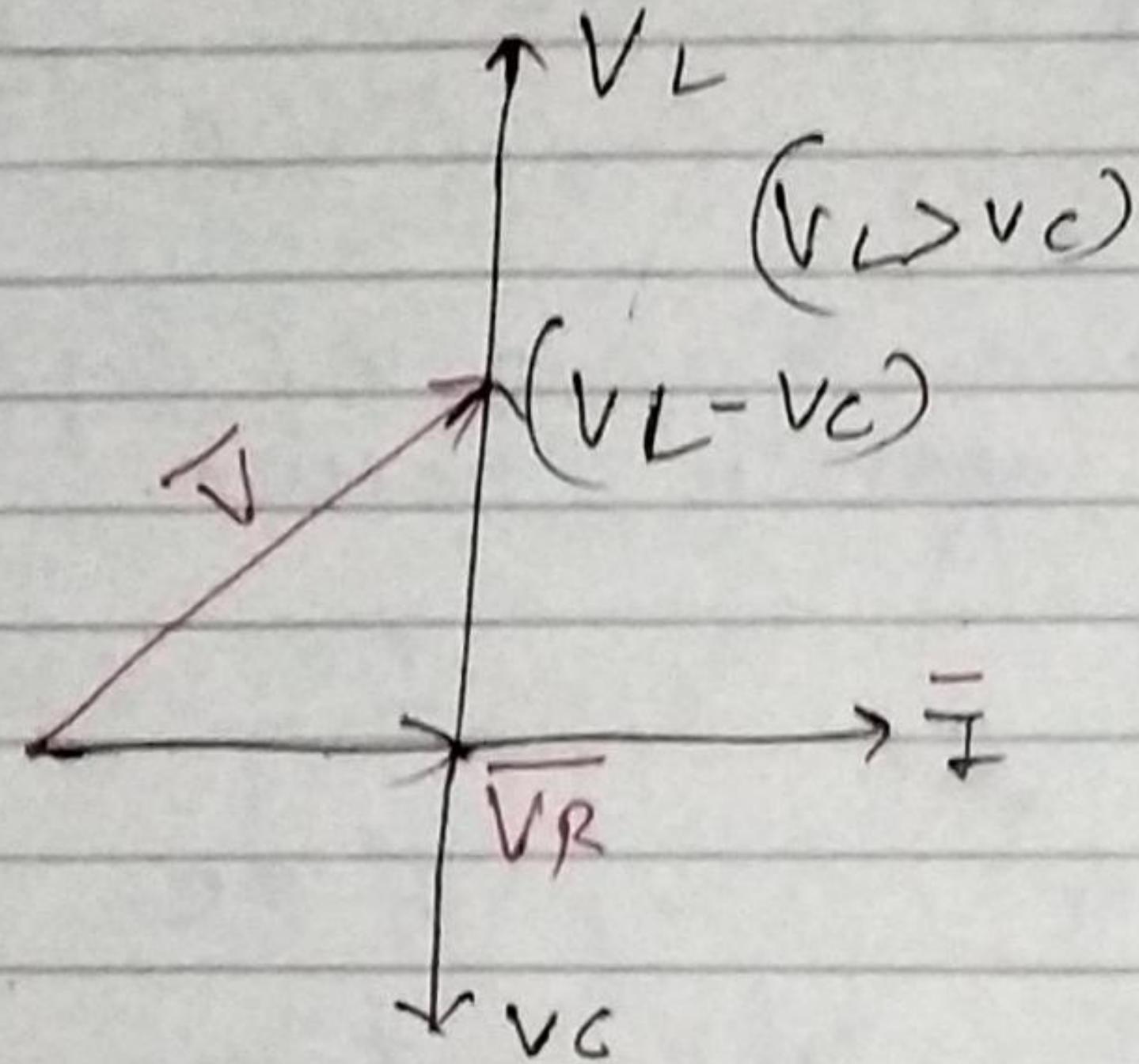
$$\tan \theta = -\frac{Q_C}{P}$$

$$\cos \theta = \frac{P}{S}$$

(leading)

⑥ Series RLC Ckt:-

$$\bar{V} = \bar{V}_R + \bar{V}_L + \bar{V}_C$$



$$V = \sqrt{\bar{V}_R^2 + (V_L - V_C)^2}$$

$$Z = \sqrt{P^2 + (X_L - X_C)^2}$$

$$S = \sqrt{P^2 + (Q_L - Q_C)^2}$$

$$\tan \theta = \frac{V_L - V_C}{\bar{V}_R}$$

$$\tan \theta = \frac{X_L - X_C}{P}$$

$$\tan \theta = \frac{Q_L - Q_C}{P}$$

$$\cos \theta = \frac{\bar{V}_R}{V}$$

(lagging)

$$\cos \theta = \frac{P}{Z}$$

(lagging)

$$\cos \theta = \frac{P}{S}$$

(lagging)

⑦ Parallel RL Ckt:-

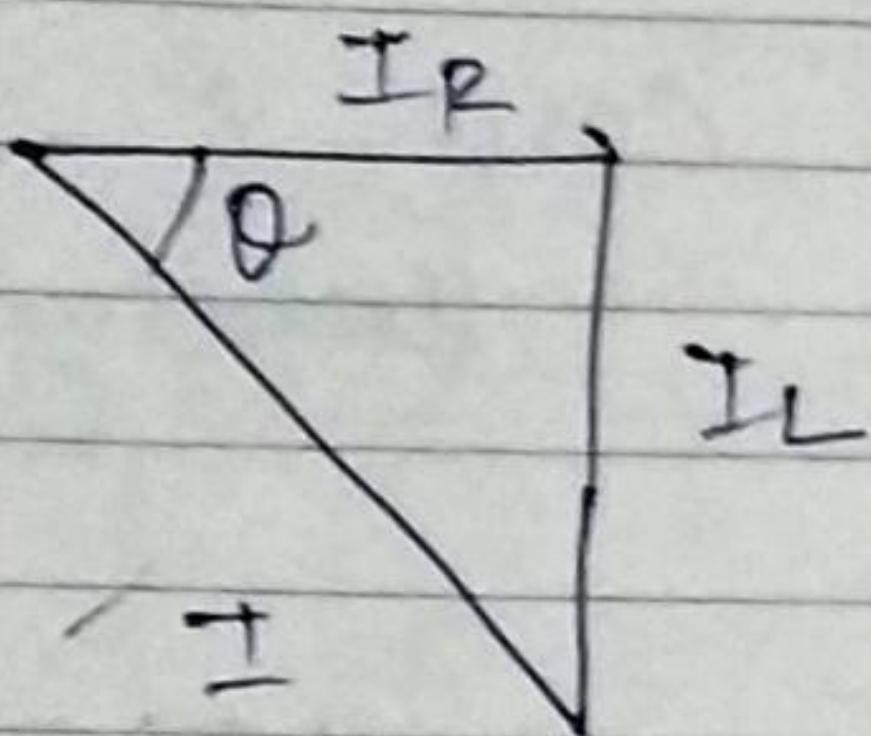
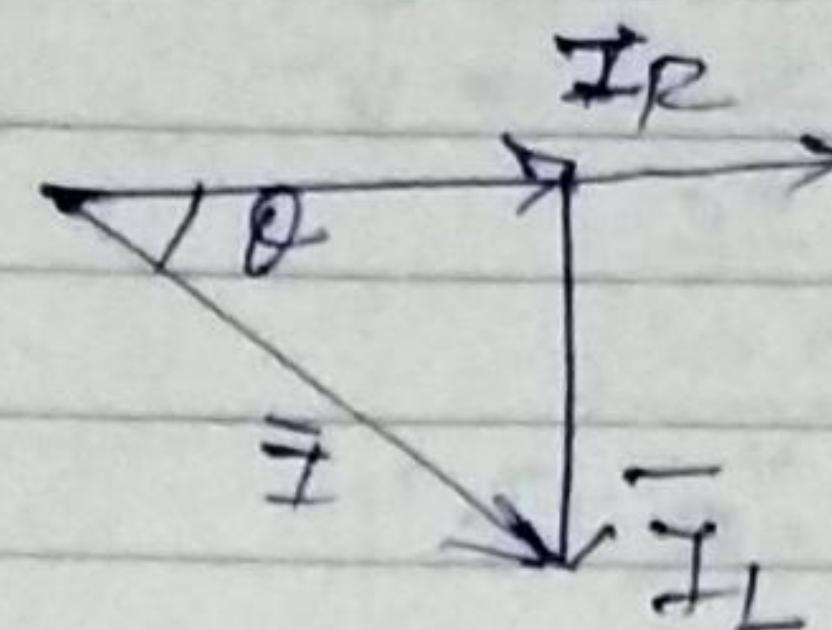
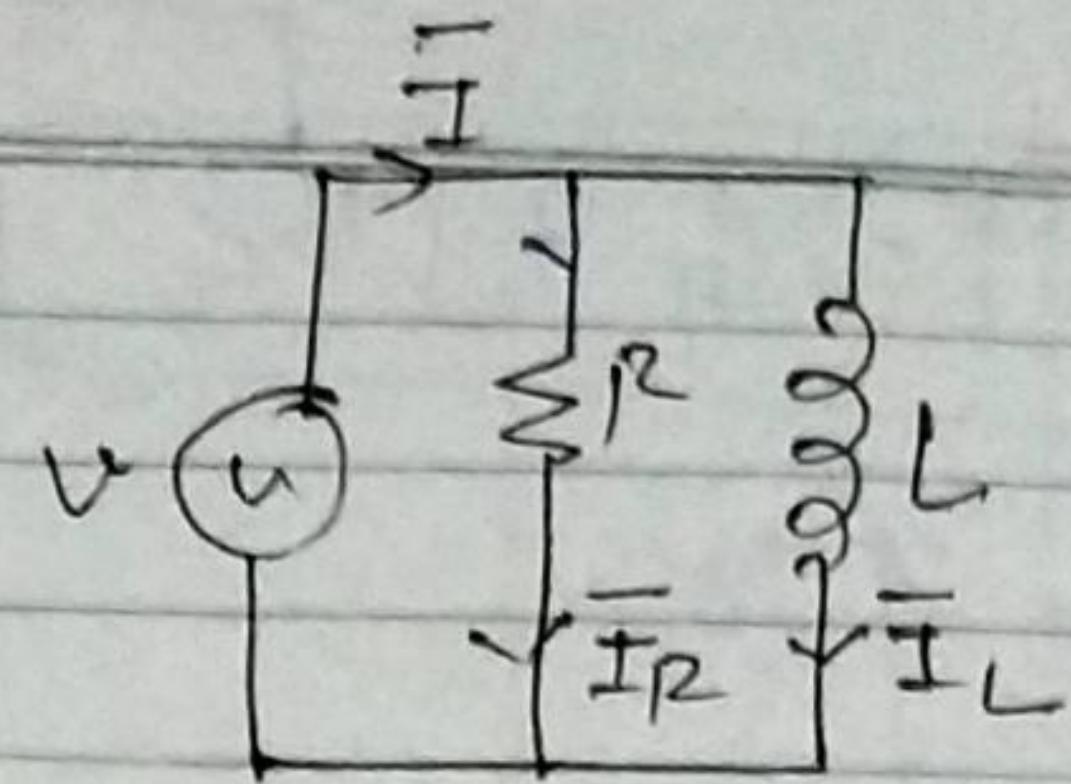
$$I = I_R + I_L$$

$$\frac{V}{Z} = \frac{V}{R} + \frac{V}{jX_L}$$

$$\frac{1}{Z} = \frac{1}{R} - j\frac{1}{X_L}$$

$$Y = G - jB_L$$

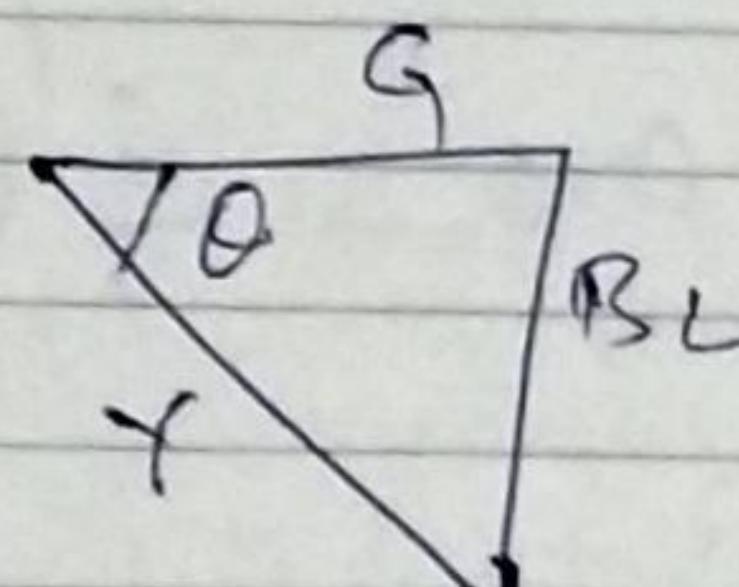
\hookrightarrow Inductive conductance susceptibility



$$I = \sqrt{I_R^2 + I_L^2}$$

$$\tan \theta = -\frac{I_L}{I_R}$$

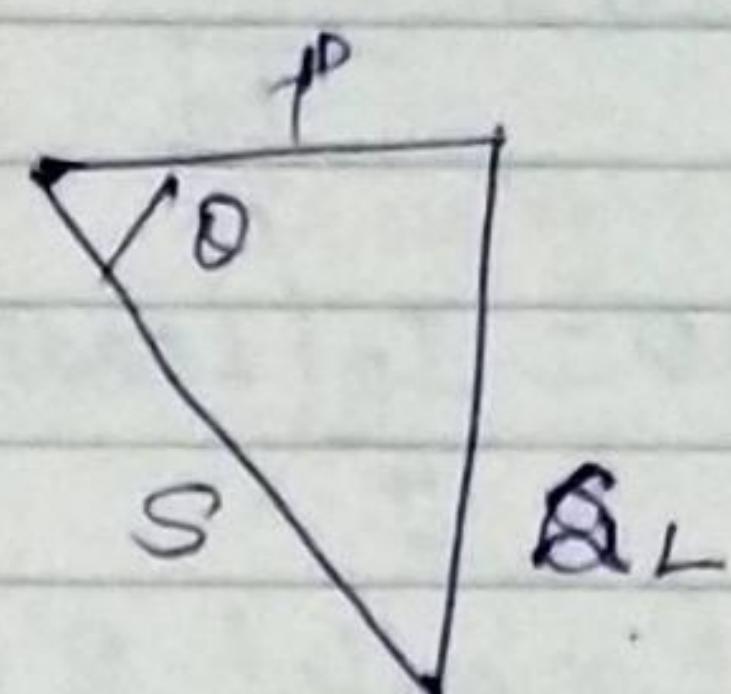
$$\cos \theta = \frac{I_R}{I} \text{ (lagging)}$$



$$Y = \sqrt{G^2 + B_L^2}$$

$$\tan \theta = -\frac{B_L}{G}$$

$$\cos \theta = \frac{G}{Y} \quad (\text{lagging})$$



$$S = \sqrt{P^2 + Q_L^2}$$

$$\tan \theta = -\frac{Q_L}{P}$$

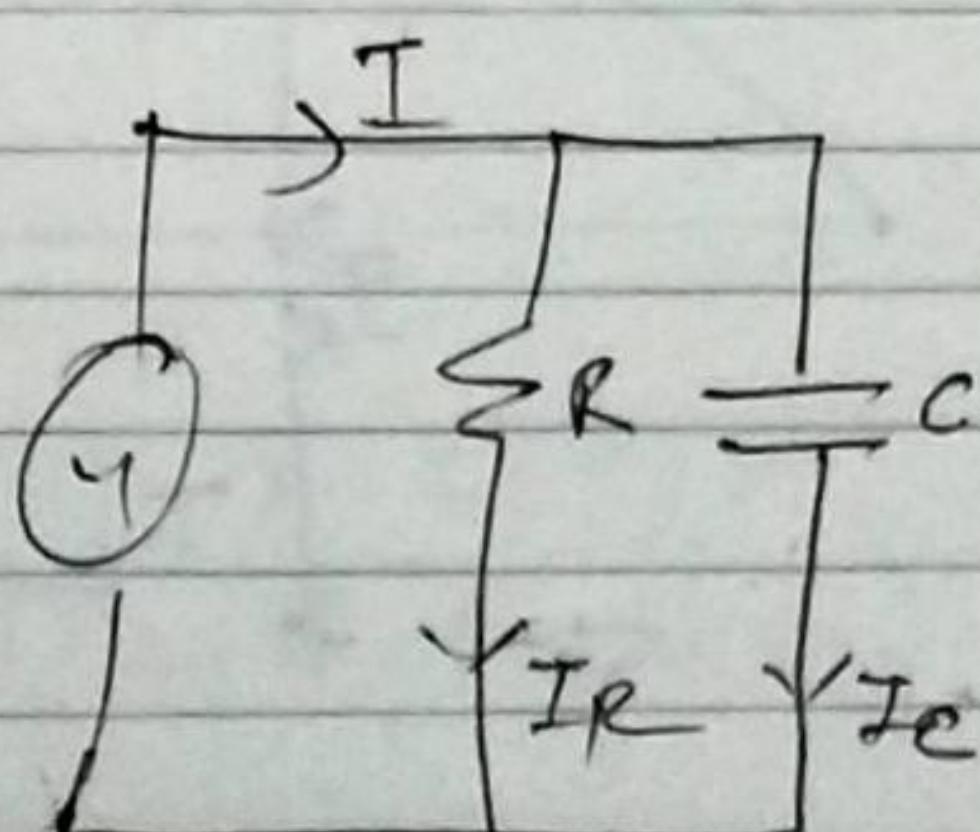
$$\cos \theta = \frac{P}{S}$$

Parallel RC Ckt:

$$I = I_R + I_C \text{ (phasor sum)}$$

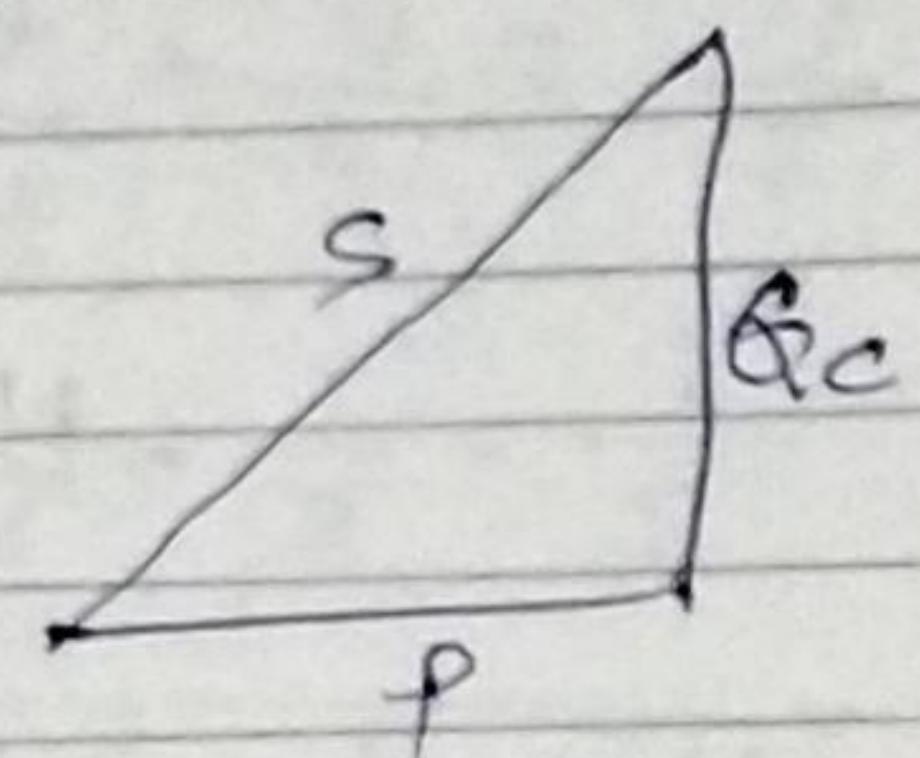
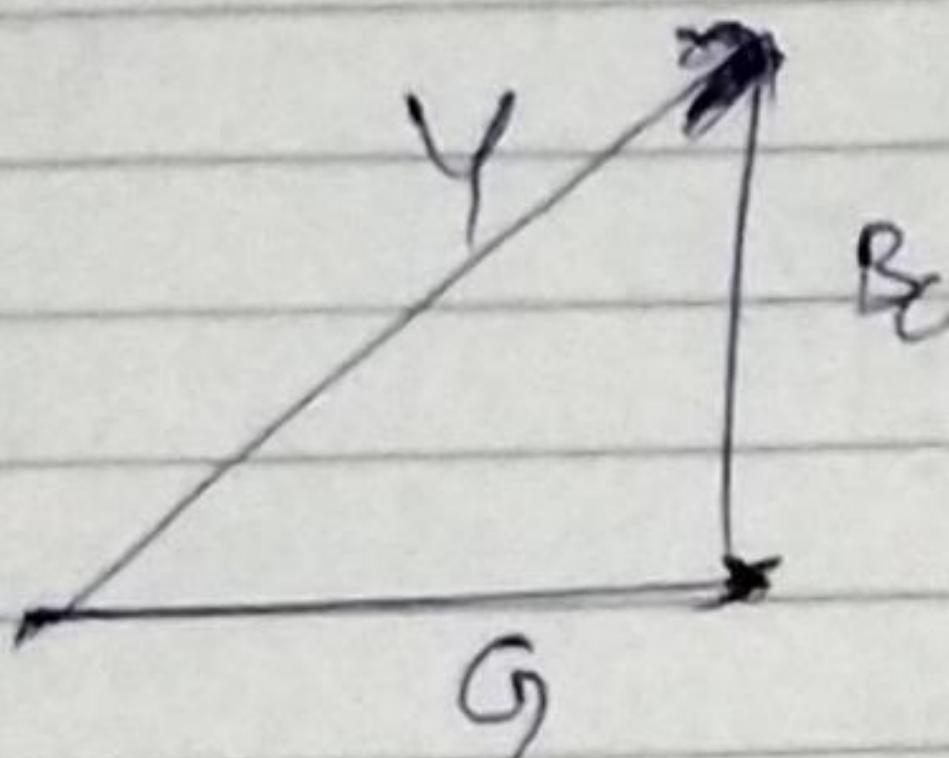
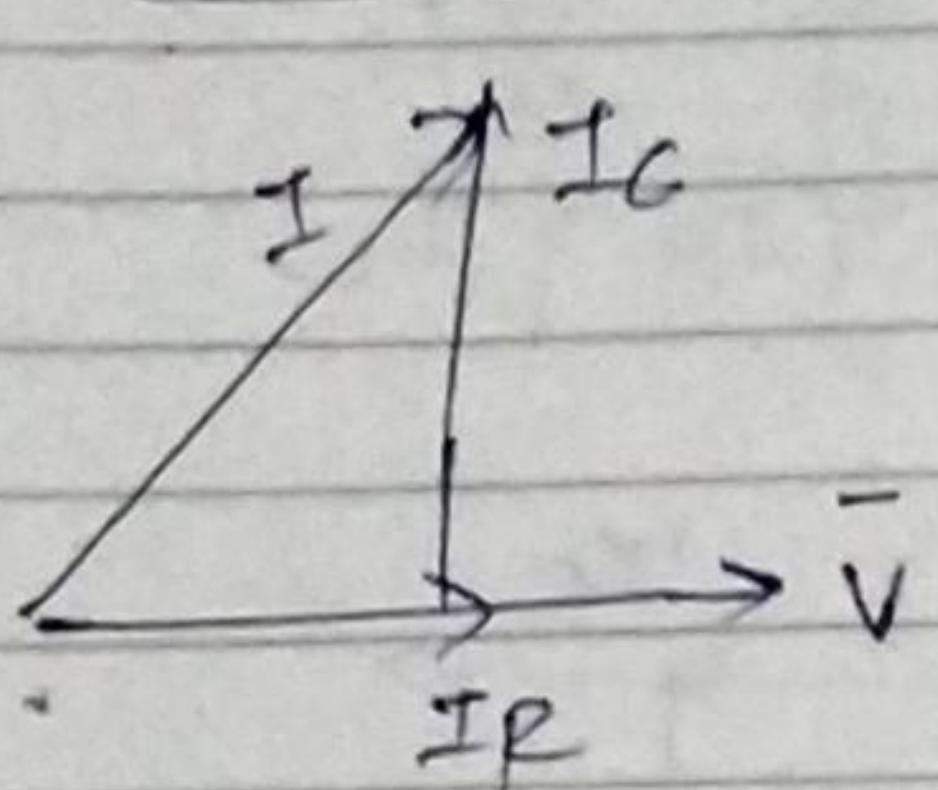
$$\frac{V}{Z} = \frac{V}{R} + \frac{V}{-jX_C}$$

$$\frac{1}{Z} = \frac{1}{R} + j\frac{1}{\omega X_C}$$



capacitive Susceptance

$$Y = G + jB_C$$



$$I = \sqrt{I_R^2 + I_C^2}$$

$$Y = \sqrt{G^2 + B_C^2}$$

$$\tan \theta = \frac{B_C}{G}$$

$$\tan \theta = \frac{Q_C}{P}$$

$$\tan \theta = \frac{I_C}{I_R}$$

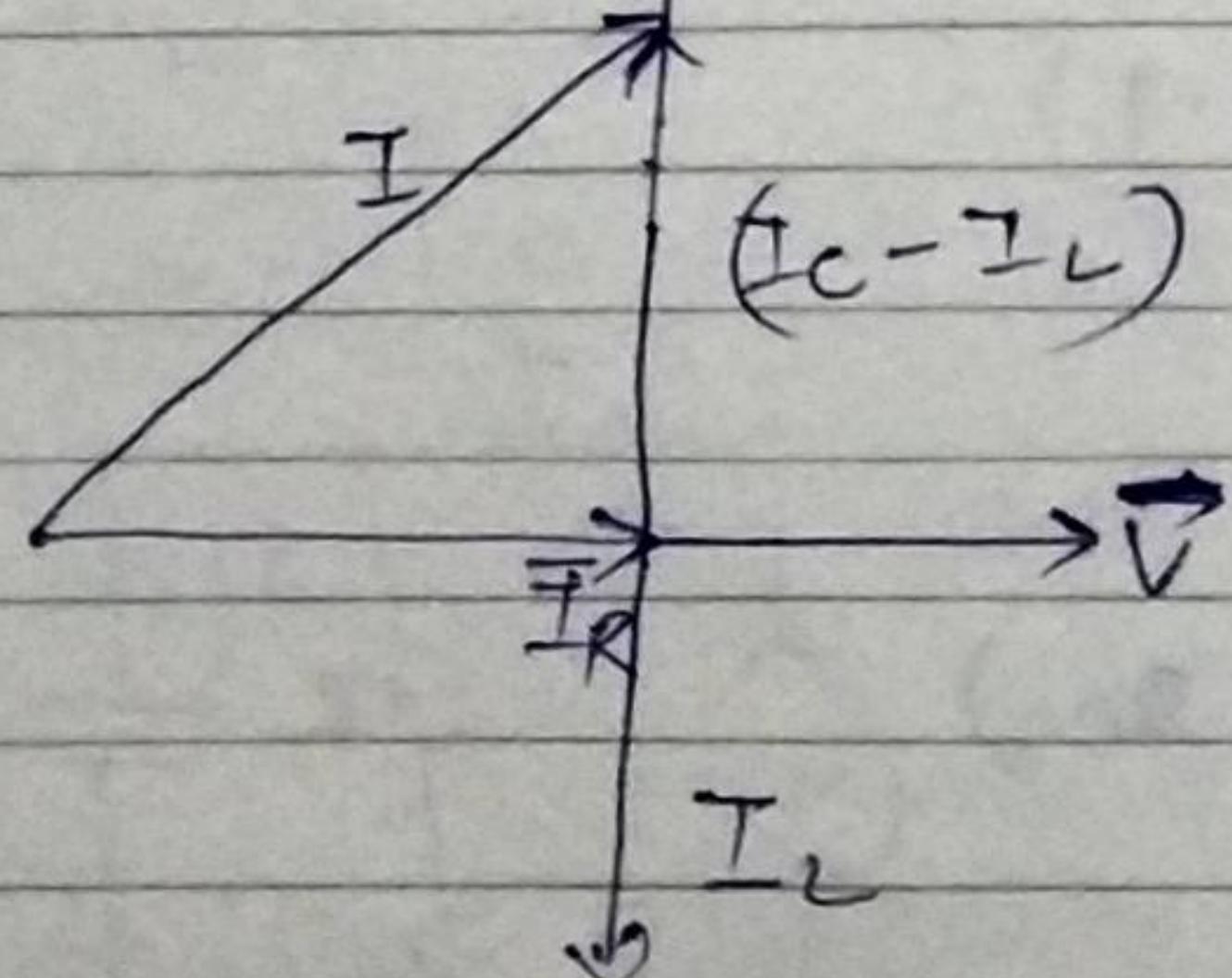
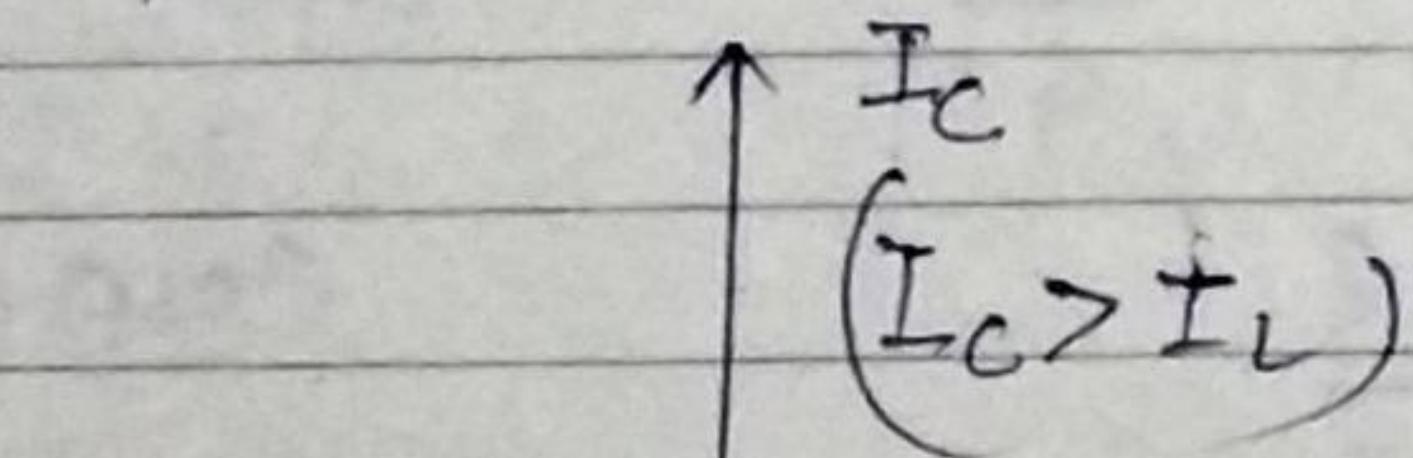
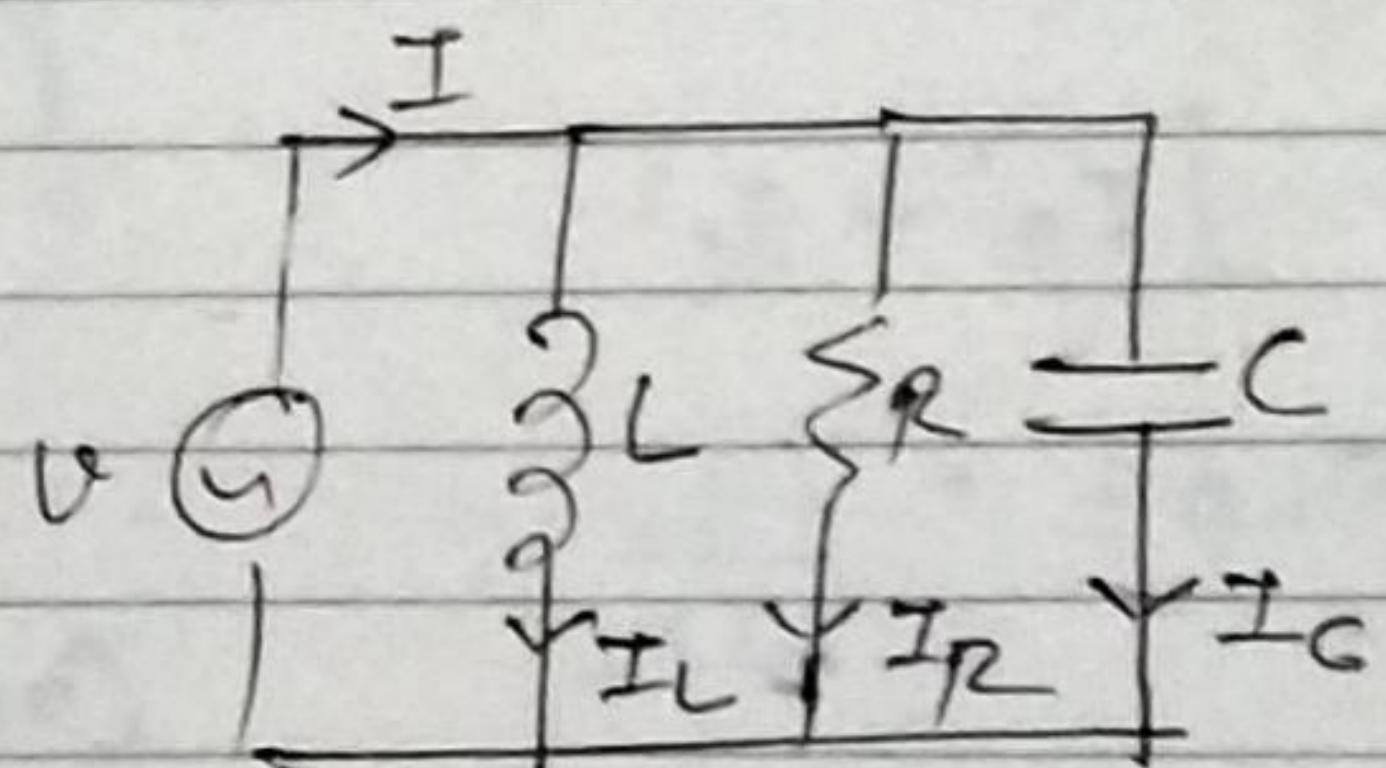
$$\cos \theta = \frac{G}{Y} \quad (\text{leading})$$

$$\cos \theta = \frac{P}{S} \quad (\text{leading})$$

$$\cos \theta = \frac{I_R}{I} \quad (\text{leading})$$

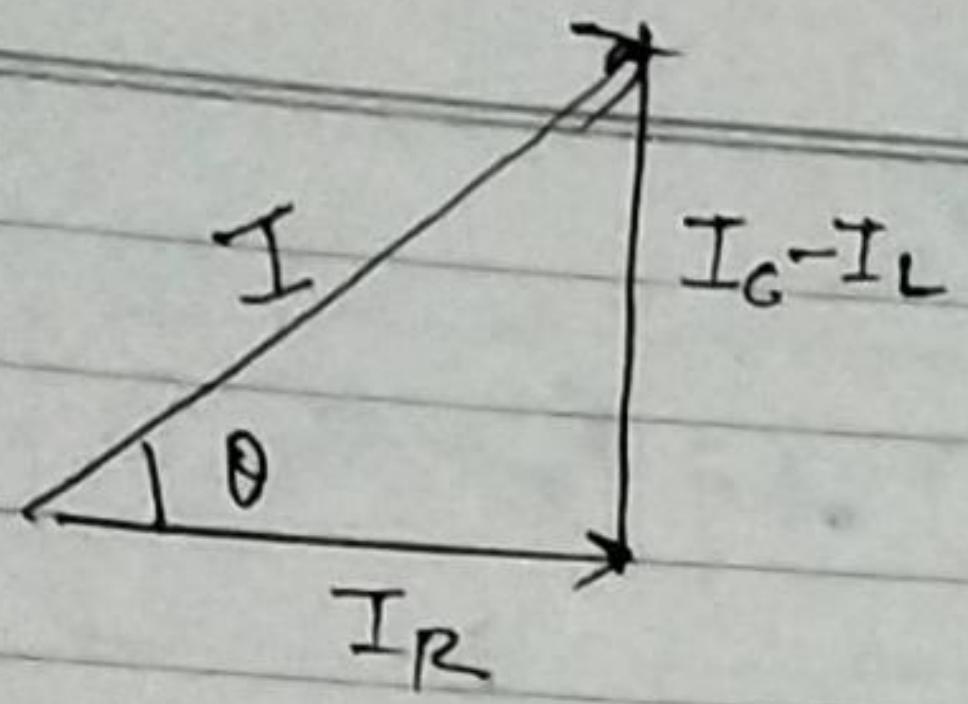
Parallel RLC Ckt

$$I = I_R + I_L + I_C$$



$$\frac{V}{z} = \frac{V}{R} + \frac{V}{jX_L} + \frac{V}{-jX_C}$$

$$Y = G + j(B_C - B_L)$$



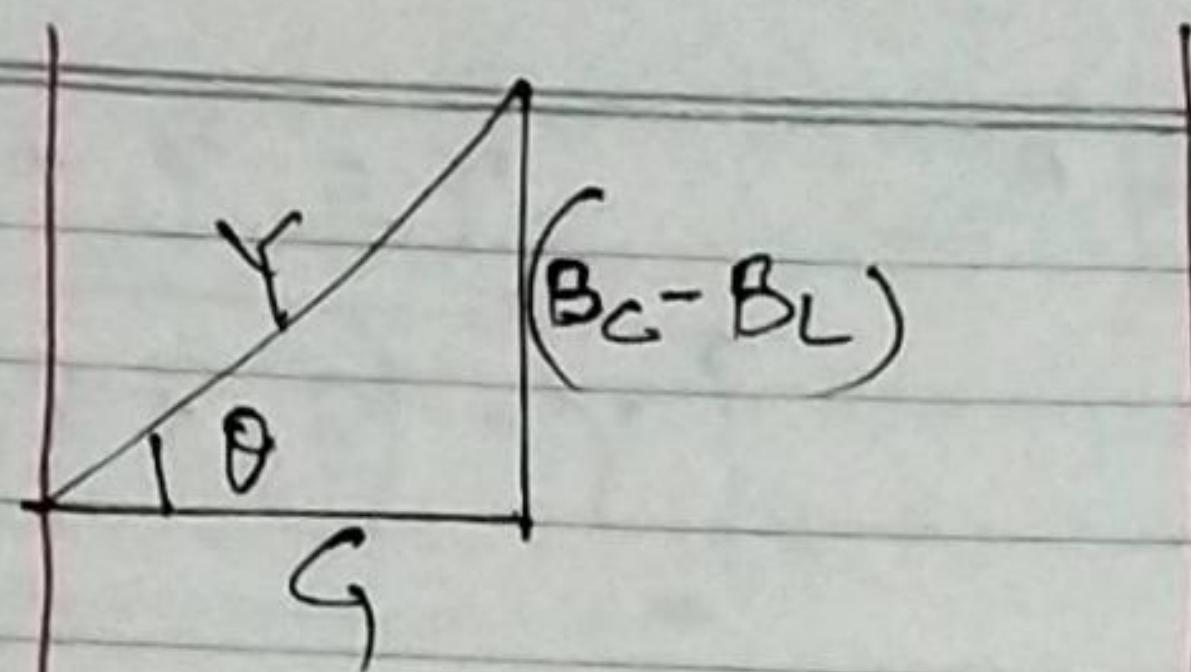
$$I = \sqrt{I_R^2 + (I_C - I_L)^2}$$

$$\tan \theta = \frac{I_C - I_L}{I_R}$$

$$\cos \theta = \frac{I_R}{I}$$

(leading)

equivalent admittance

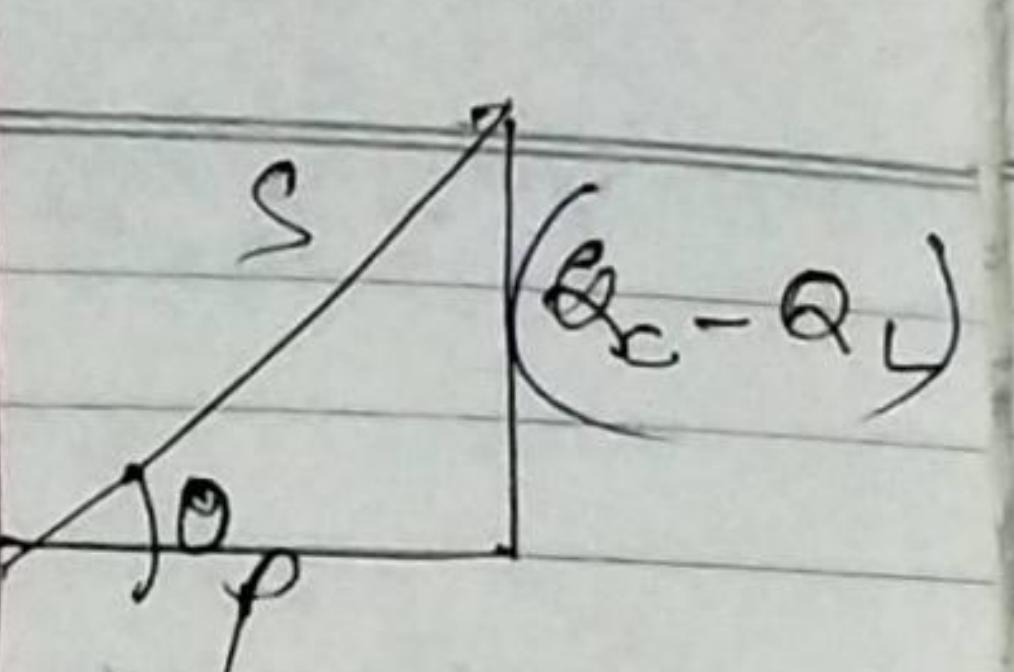


$$Y = \sqrt{G^2 + (B_C - B_L)^2}$$

$$\tan \theta = \frac{B_C - B_L}{G}$$

$$\cos \theta = \frac{G}{Y}$$

(leading)

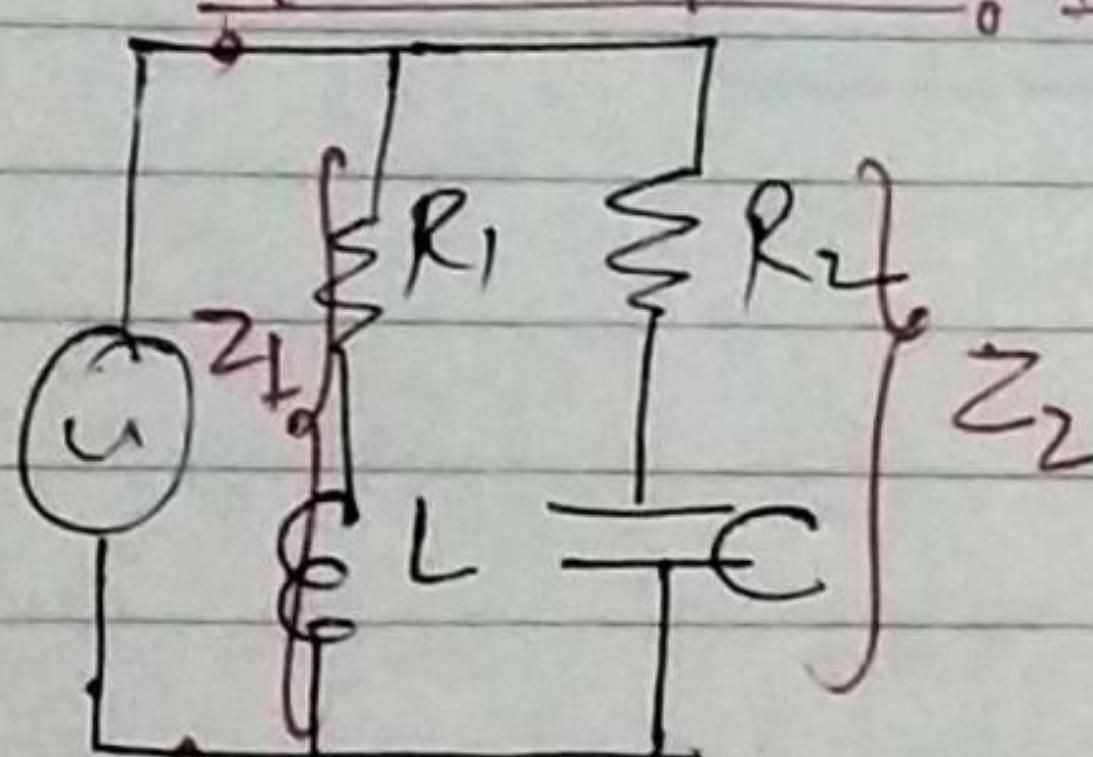


$$S = \sqrt{P^2 + (Q_C - Q_L)^2}$$

$$\tan \theta = \frac{Q_C - Q_L}{P}$$

$$\cos \theta = \frac{P}{S}$$

(leading)



$$Z_1 = R_L + jX_L$$

$$Y_1 = \frac{1}{R_1 + jX_L} \times \frac{(R_1 - jX_L)}{(R_1 - jX_L)}$$

$$Y_1 = \frac{R_1 - jX_L}{R_1^2 + X_L^2}$$

$$Y_1 = \frac{R_1}{(R_1^2 + X_L^2)} - j \frac{X_L}{(R_1^2 + X_L^2)}$$

$$Y_1 = G_1 - jB_L$$

$$Z_2 = R_2 - jX_C \quad Y_2 = \frac{1}{R_2 - jX_C} \times \frac{R_2 + jX_C}{R_2 + jX_C}$$

$$Y_2 = \frac{R_2 + jX_C}{R_2^2 + X_C^2}$$

$$Y_2 = \frac{R_2}{R_2^2 + X_C^2} + j \frac{X_C}{R_2^2 + X_C^2}$$

$$\gamma_2 = G_2 + jB_C \quad \gamma = \gamma_1 + \gamma_2$$

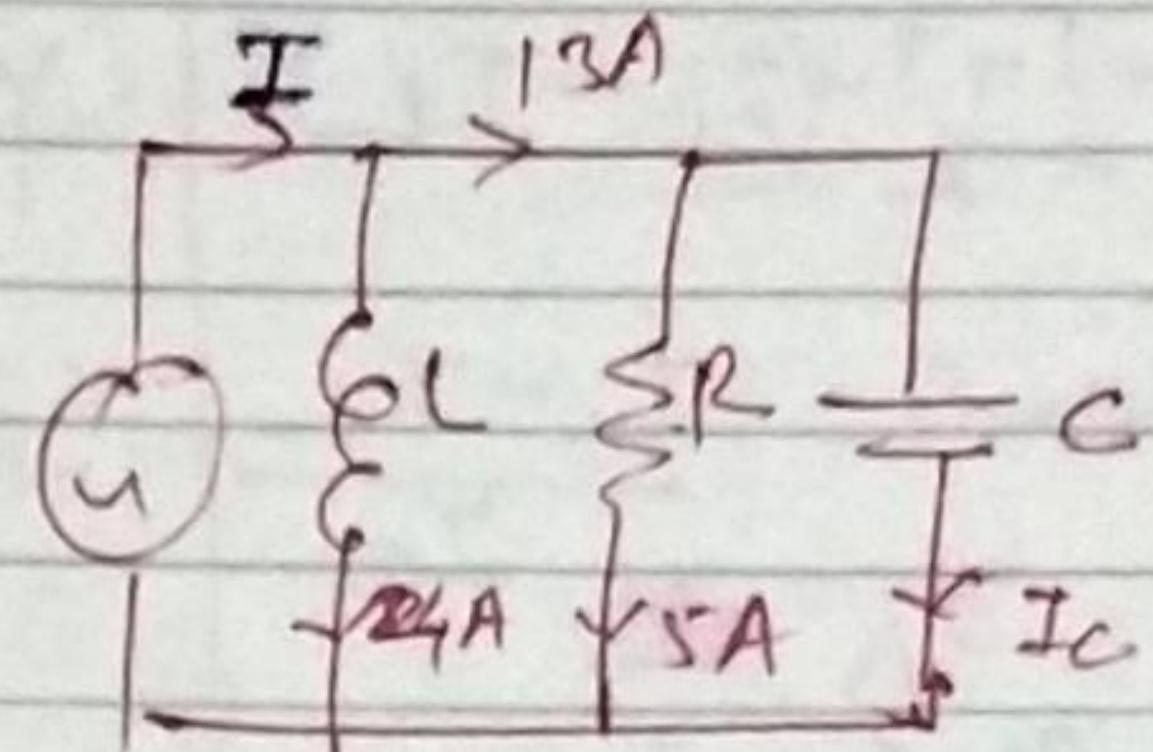
$$\boxed{\gamma = (G_1 + G_2) + j(B_C - B_L)}$$

Ans:

$$IB = \sqrt{I^2 + I_C^2}$$

$$169 - 25 = I_C^2$$

$$\boxed{I_C = 12}$$

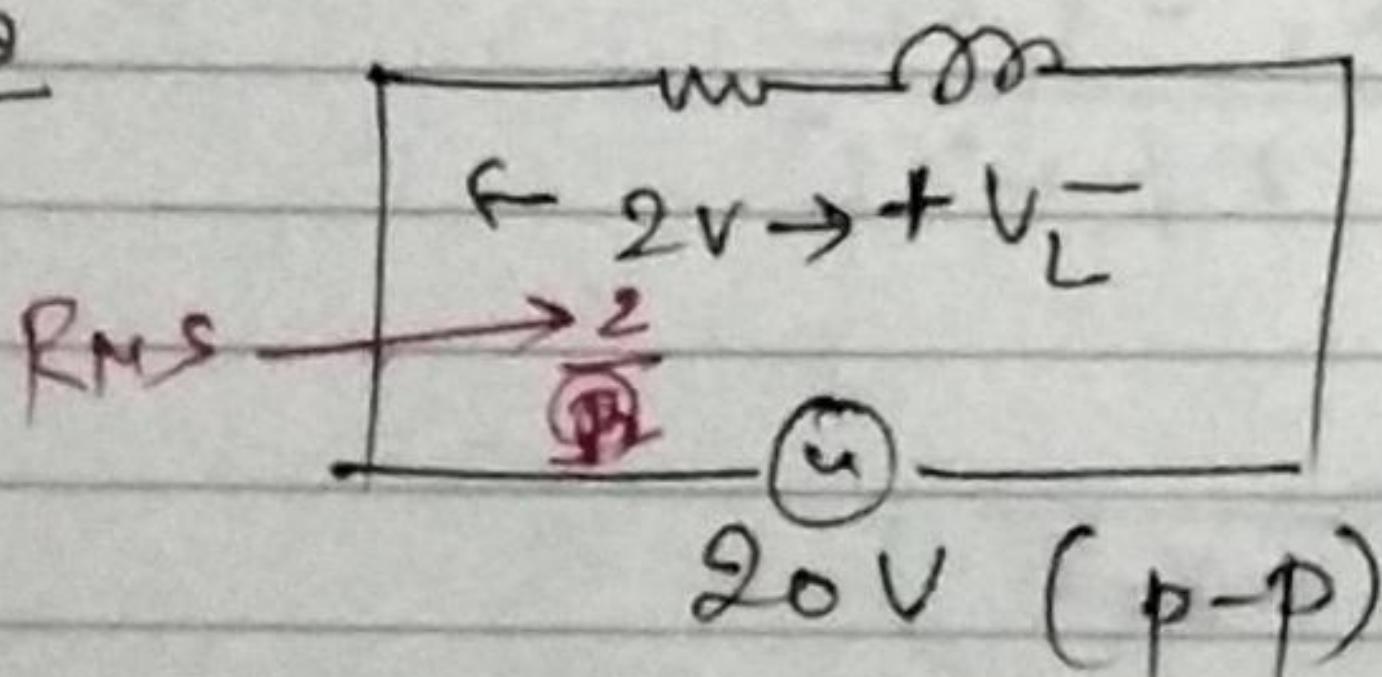


$$Z = \sqrt{25 + (12-24)^2}$$

$$\boxed{Z = 21.37}$$

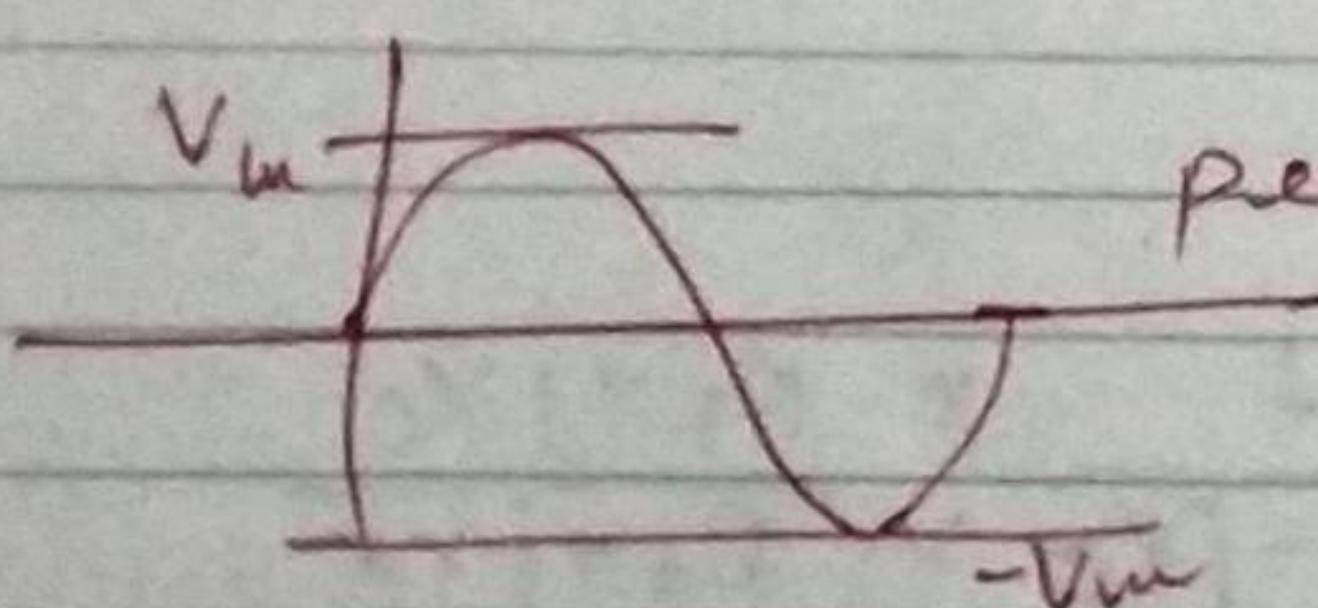
$$\boxed{I = 15A}$$

Ques



$$\cancel{20^2 = \epsilon^2 + V_L^2}$$

$$\cancel{19.9V}$$



peak-topeak

$$2V_m = 80V$$

$$\boxed{V_m = 40V}$$

$$V_{rms} = \frac{10}{\sqrt{2}}$$

$$V = \sqrt{V_R^2 + V_L^2} \quad \left(\frac{10}{\sqrt{2}} \right)^2 = 4 + V_L^2$$

$$\boxed{V_L = \sqrt{46} \text{ Volts}}$$

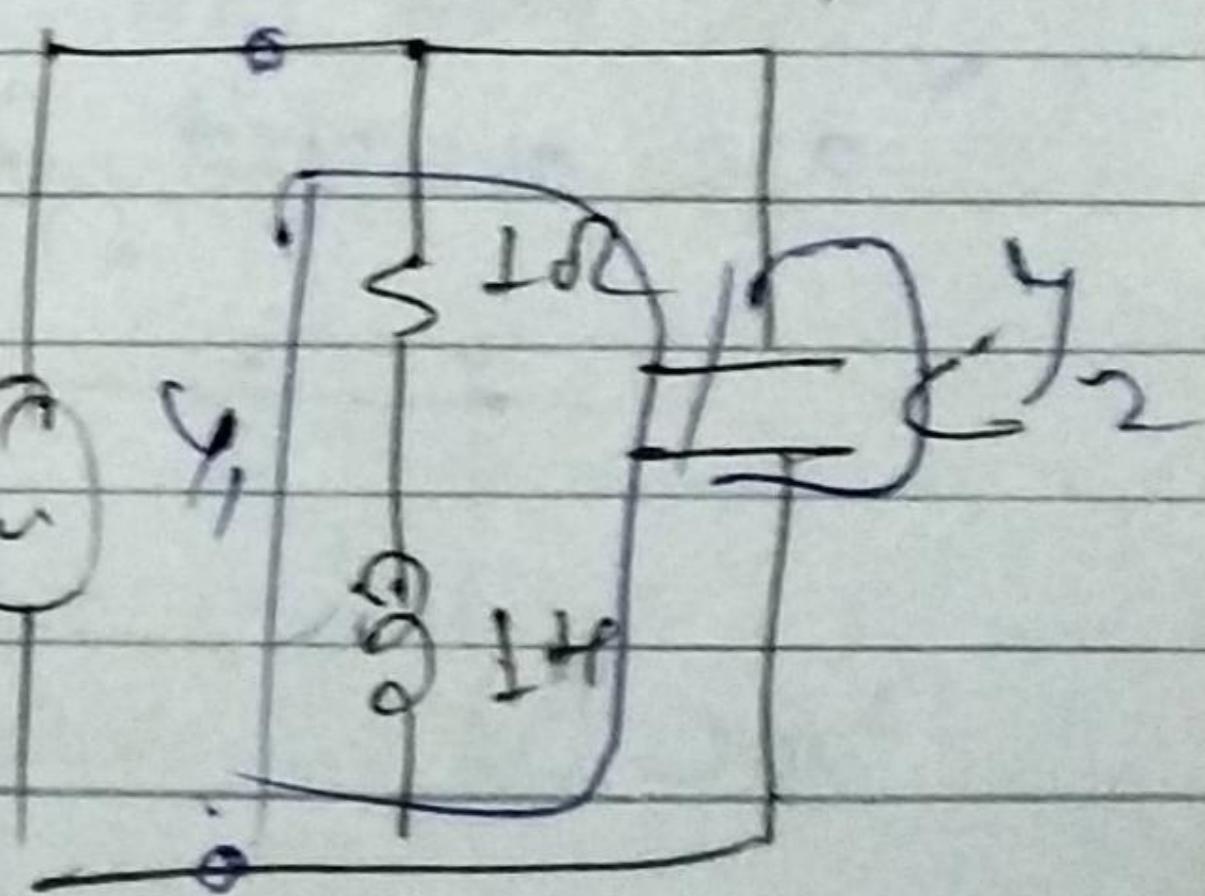
$R_1 \leftarrow 5$

Ques : Find the value of 'C' when the pf of the
CKT is 0.8 lagging.

$$Y_1 = G_1 - j B_L$$

$$G_1 = \frac{R_1}{R_1^2 + X_L^2}$$

$$S_{out} = u(t) \circ Y_1$$



$$X_L = \omega L = (1)(1) = j \Omega \quad G_1 = \frac{j}{1+j} = \frac{1}{2}$$

$$B_1 = \frac{X_L}{R_1^2 + X_L^2} = \frac{1}{2} \quad Y_1 = \frac{1}{2} - j \frac{1}{2}$$

$$B_C = \frac{1}{X_C} = \frac{1}{j/\omega_C} \quad Y_2 = j B_C = j C$$

$$Y = Y_1 + Y_2 \Rightarrow \frac{1}{2} - j \frac{1}{2} + j C$$

$$Y = \frac{1}{2} + j \left(C - \frac{1}{2}\right) \quad \cos \phi = \frac{G}{|Y|}$$

$$0.8 = \frac{|Y|}{\sqrt{\frac{1}{4} + \left(\frac{C-1}{2}\right)^2}} \quad C = \frac{7}{8}, \frac{1}{8}$$

$C = \frac{1}{8}$ F : Pflogging