

ASEN 5070
Statistical Orbit Determination I
Fall 2012



Professor Jeffrey S. Parker
Professor George H. Born

Lecture 23: Process Noise



University of Colorado
Boulder

Announcements

- ▶ How was break?
- ▶ HW 10 due Thursday.
- ▶ HW 11 due next week.
- ▶ Grading
- ▶ You have the tools needed to finish the project. Only things missing are bonus pieces.
- ▶ I may be out tomorrow; if you need office-hour help, check the TAs or visit me Thursday.

November 2012						
Sun	Mon	Tue	Wed	Thu	Fri	Sat
4	5	6	7	8	9	10
11	12	13	14	15	16	17
18	19	20	21	22	23	24
25	26	27	28	29	30	

December 2012						
Sun	Mon	Tue	Wed	Thu	Fri	Sat
2	3	4	5	6	7	8
9	10	11	12	13	14	15
16	17	18	19	20	21	22
23	24	25	26	27	28	29
30	31					

Last Day of Classes
 Take-Home Exam Due
 Final Project Due
 All HW Due



The diagram shows four arrows originating from the text labels on the right and pointing to specific dates on the December calendar. The first arrow points to December 15th. The second arrow points to December 20th. The third arrow points to December 17th. The fourth arrow points to December 29th.

Quiz 19 Review

Information

This week we'll be talking about process noise compensation! So it will be useful to review some matrix math.

Question 1 (1 point)

Let's say A is a matrix, A^T is the transpose of A , and A^{-1} is the inverse of A (if it exists).

The relationship $(A^T)^{-1} = (A^{-1})^T$ holds for which of the following square matrices? Select all answers that are true.

- All square matrices
- All invertible square matrices
- All symmetric positive definite square matrices
- All invertible diagonal square matrices



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Quiz 19 Review

Question 2 (1 point)

Let's say that A and B are both square matrices, T is the transpose operator, and the matrices A, B, and $A^T B$ are all invertible.

What is another expression for $(A^T B)^{-1}$?

- A⁻¹B
- B⁻¹A
- B⁻¹A⁻¹T
- A⁻¹T⁻¹B⁻¹



Quiz 19 Review

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What is another expression for $(A^T B)^{-1}$?

A⁻¹B

B⁻¹A

B⁻¹·(A⁻¹)^T

A⁻¹T⁻¹B⁻¹



Quiz 19 Review

Question 3 (1 point)

Let's say A, B, and C are matrices and they are the proper size to make $A^T B^T C$ make sense. Each element of A, B, and C may be a constant or a time-varying value. What is the time derivative of (ABC) ? (i.e., what is $d(ABC)$). (For simplicity, "dM" is shorthand here for " dM/dt " or "M-dot").

- $dA^T B^T C$
- $dA + dB + dC$
- $dC + dB + dA$
- $dA^T B^T C + A^T dB^T C + A^T B^T dC$



Quiz 19 Review

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- $dA \cdot B \cdot C$
- $dA \cdot dB \cdot dC$
- $dC \cdot dB \cdot dA$
- $dA^T B^T C + A^T dB^T C + A^T B^T dC$



Quiz 19 Review

Question 4 (1 point)

Let's say we have n state parameters and m total observations, where $m \geq n$. The state estimate covariance matrix is P .

The conventional sequential time update for P looks like

$$P_{\text{bar}}(k+1) = P_k P_k^T P_k^{-1} P_k^T.$$

We're working on deriving a new time update where we add a Q matrix like this:

$$P_{\text{bar}}(k+1) = P_k P_k^T P_k^{-1} P_k^T + Q_k.$$

What size should Q_k be?

- $m \times n$
- $m \times m$
- $m \times m$
- $m \times n$



Quiz 19 Review

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What size should Q_k be?

$m \times n$

$m \times m$

$m \times m$

$m \times n$



Quiz 19 Review

Question 5 (1 point)

If Q is an $m \times m$ matrix (completely independent of Q_{-k} in the last question) and we want $L^*Q^*L^*T$ to be an $n \times n$ matrix, what should be the size of L ?

- non
- $m \times m$
- $n \times m$
- $n \times n$



Quiz 19 Review

Question 5 (1 point)

If Q is an $m \times m$ matrix (completely independent of Q_{-k} in the last question) and we want $L^*Q^*L^*T$ to be an $n \times n$ matrix, what should be the size of L ?

non

main

non

other



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HW#11

- ▶ Due the Thursday after HW10

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HW #11

1. Calculate $\Phi(18340, 0)$ using the equation $\Phi(t_k, 0) = \Phi(t_k, t_n)\Phi(t_n, 0)$
2. Report the state deviation after one pass through the sequential filter.
3. Compare the result from question 2 to your batch results.
4. Using the convention Kalman filter covariance measurement update equation, plot the trace of the position covariance on a semilog y scale (at each measurement time).
5. Using the Joseph formulation of the covariance measurement update equation, plot the trace of the position covariance on a semilog y scale (at each measurement time).
6. Plot the position error ellipsoid for your batch state error covariance. (You may use the function [plotEllipsoid.m](#))
7. Work problem 4-41. You can use this data file [hw11.dat](#), where time is in the first column and the observations are in the second column.



Contents

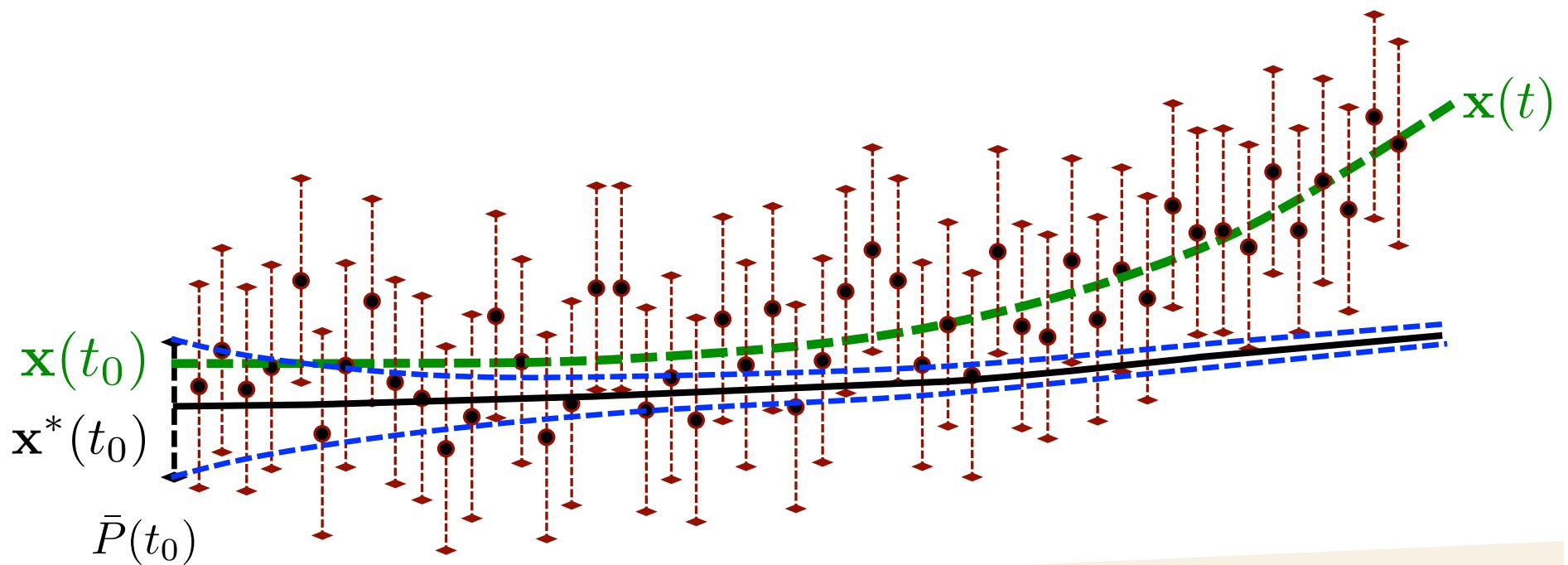
▶ Process Noise



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Issue: Filter Saturation

- ▶ Applying the sequential algorithm to a large amount of data will cause the covariance to shrink down too far.



Filter Saturation

- ▶ Sequential filter measurement update:

$$K_k = \bar{P}_k \tilde{H}_k^T \left[\tilde{H}_k \bar{P}_k \tilde{H}_k^T + R_k \right]^{-1}$$

$$\hat{\mathbf{x}}_k = \bar{\mathbf{x}}_k + K_k \left[\mathbf{y}_k - \tilde{H}_k \bar{\mathbf{x}}_k \right]$$

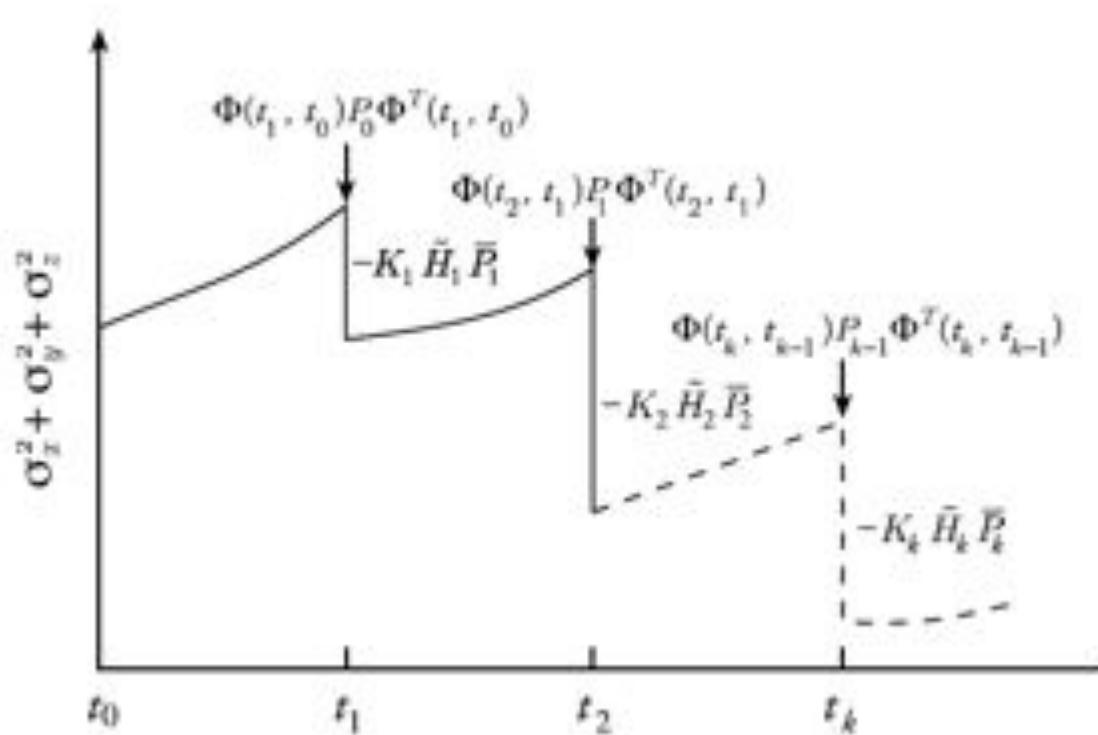
$$P_k = \left[I - K_k \tilde{H}_k^T \right] \bar{P}_k$$

- ▶ Large number of observations will drive the covariance to zero.
 - (which is great if everything is perfectly modeled! But that's never quite so.)



Filter Saturation

- ▶ Trace of covariance over time.



Filter Saturation

► As $\bar{P}_k \rightarrow 0$

$$K_k = \bar{P}_k \tilde{H}_k^T \left[\tilde{H}_k \bar{P}_k \tilde{H}_k^T + R_k \right]^{-1}$$

$$\tilde{H}_k \bar{P}_k \tilde{H}_k^T + R_k \rightarrow R_k$$

$$K_k \rightarrow \bar{P}_k \tilde{H}_k^T R_k^{-1}$$

$$\bar{P}_k \rightarrow 0$$

$$K_k \rightarrow 0$$



Filter Saturation

► As $K_k \rightarrow 0$

$$\hat{\mathbf{x}}_k = \bar{\mathbf{x}}_k + K_k \left[\mathbf{y}_k - \tilde{H}_k \bar{\mathbf{x}}_k \right]$$

The filter will begin ignoring observations and the best estimate will remain constant over time.



Example of Filter Saturation

- ▶ This is the example given in Appendix F
- ▶ I coded it up, and you can too. It's pretty straightforward. Plus this way I could investigate different aspects of the scenario.

- ▶ Given:
 - A particle moving along the x-axis in a positive direction.
 - It is nominally moving at a constant 10 m/s velocity.
 - It is *actually* being perturbed by an unknown acceleration – unmodeled!
 - The acceleration is a small-amplitude oscillation.
 - The particle is being tracked by an observer, also on the x-axis at 10 Hz observation frequency.
 - Range with white noise $N(0,1)$ m
 - Range-rate with white noise $N(0,0.1)$ m/s
- ▶ Can we estimate the state and even the unmodeled force?



Example of Filter Saturation

- ▶ Our system

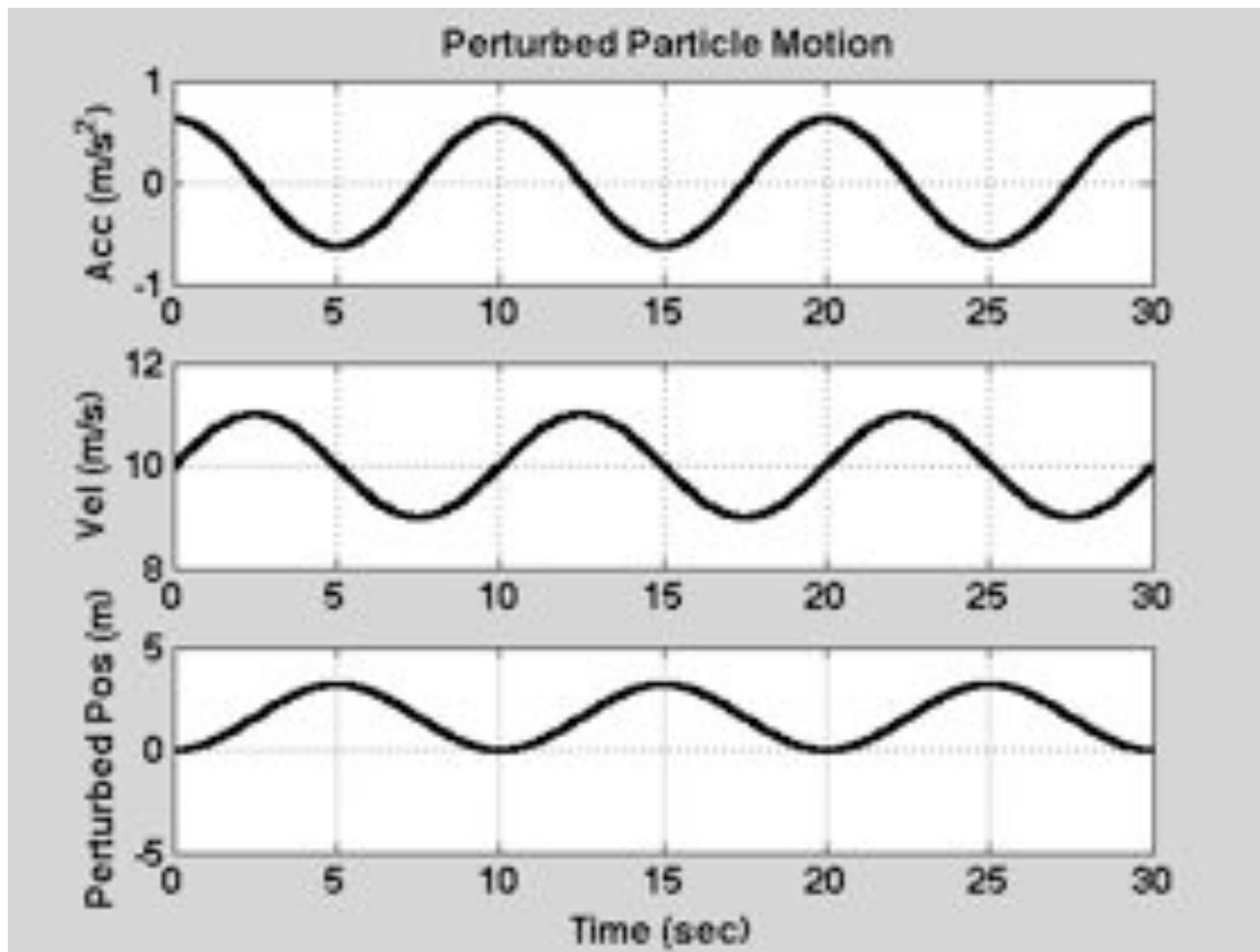
$$\mathbf{x}(t) = \begin{bmatrix} x(t) \\ \dot{x}(t) \end{bmatrix}$$

$$\begin{aligned} \dot{\mathbf{x}}(t) &= \begin{bmatrix} \dot{x}(t) \\ \ddot{x}(t) \end{bmatrix} = \begin{bmatrix} \dot{x}(t) \\ 0 \end{bmatrix} \\ &= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(t) \\ \dot{x}(t) \end{bmatrix} \end{aligned}$$

- ▶ Notice that the acceleration is not in our model.



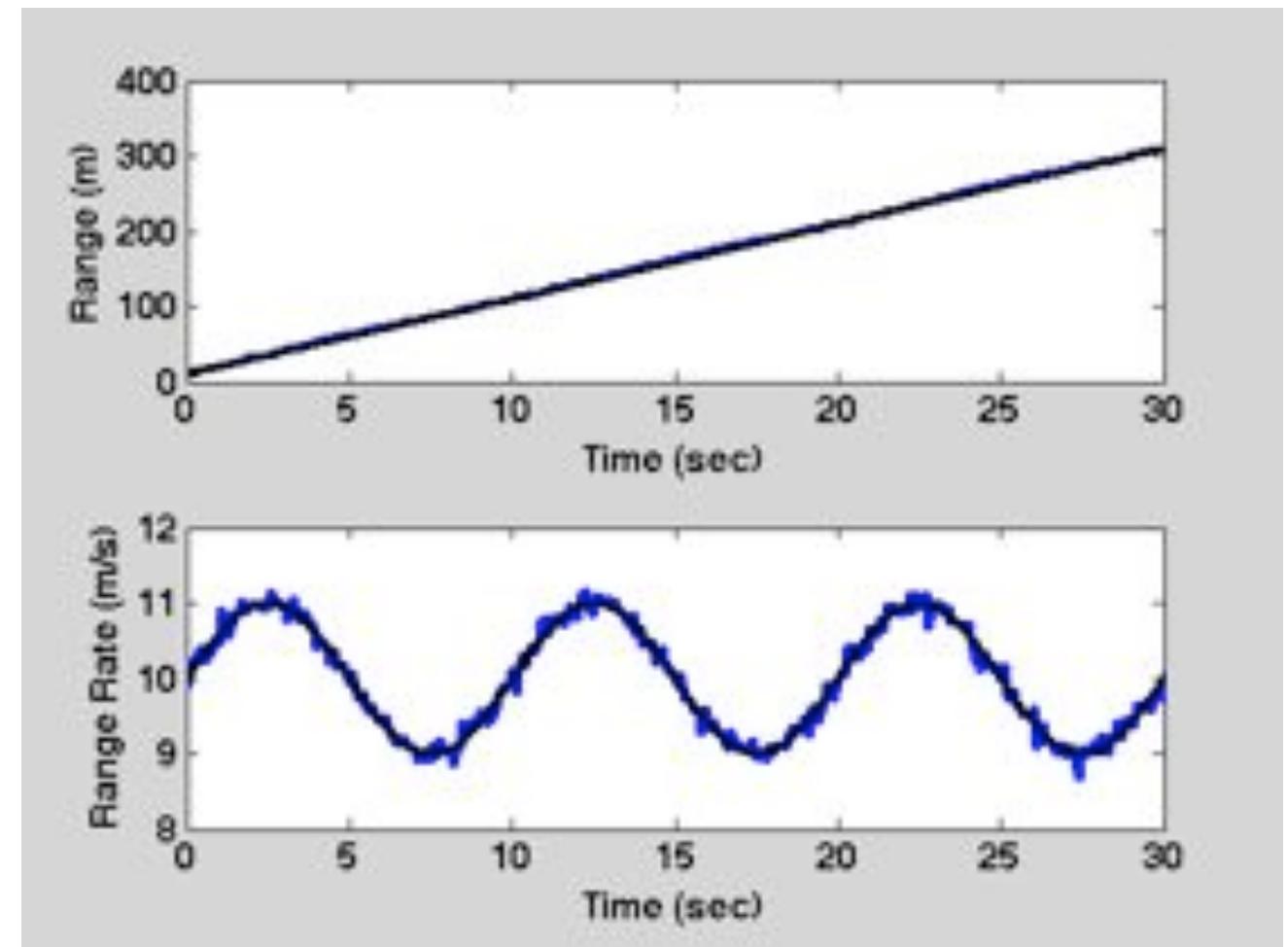
Example of Filter Saturation



Example of Filter Saturation

► Observations

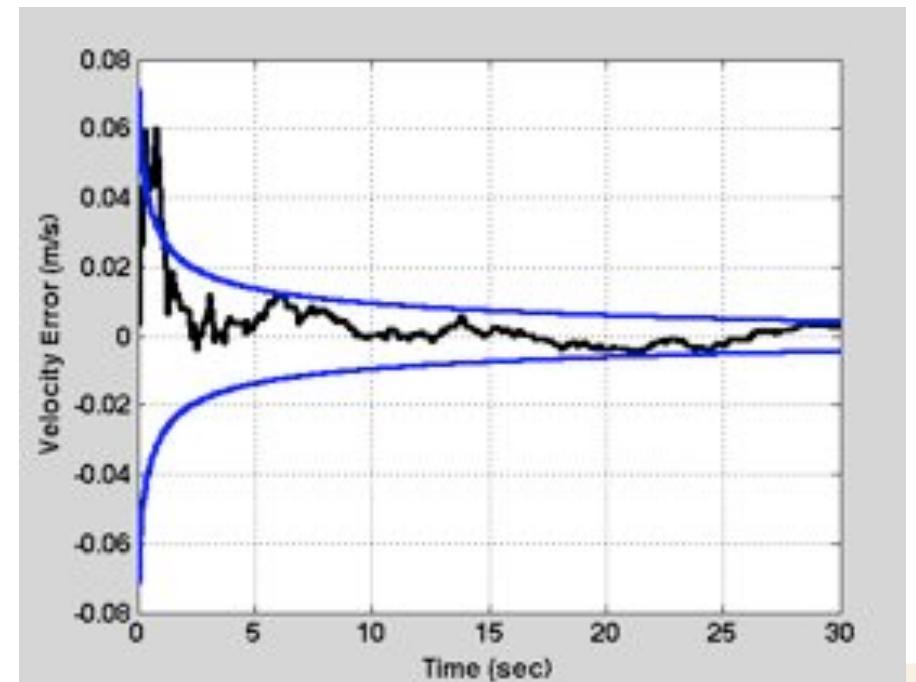
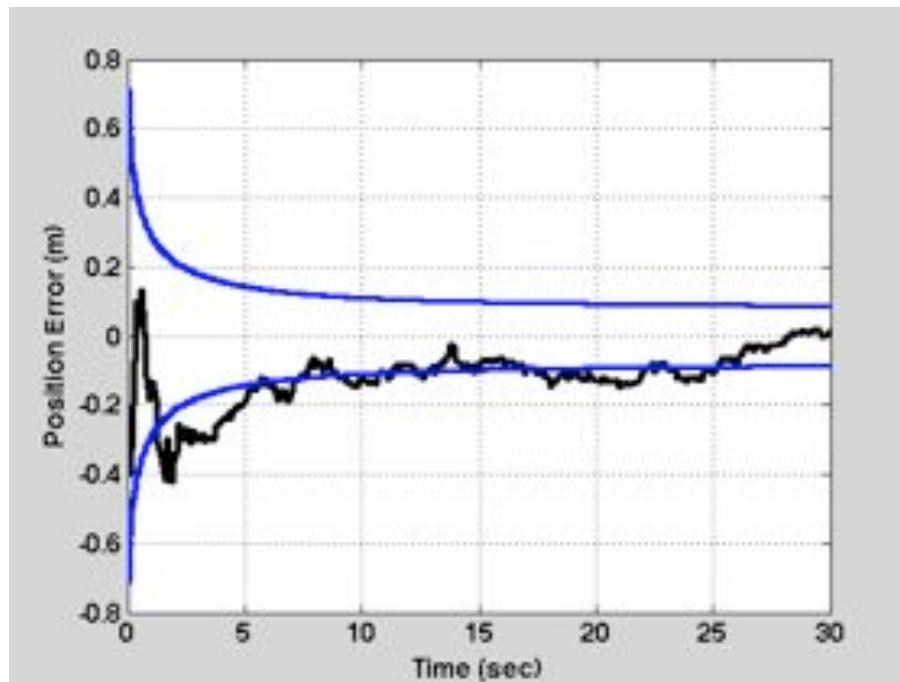
- Range
 $N(0,1)$ meter
- Range Rate
 $N(0,0.1)$ m/s



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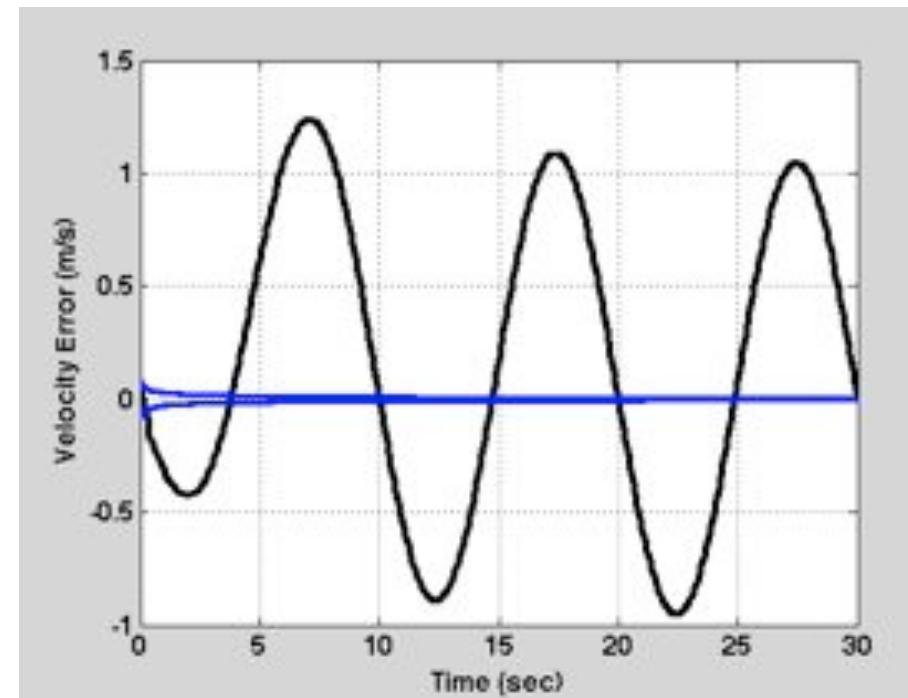
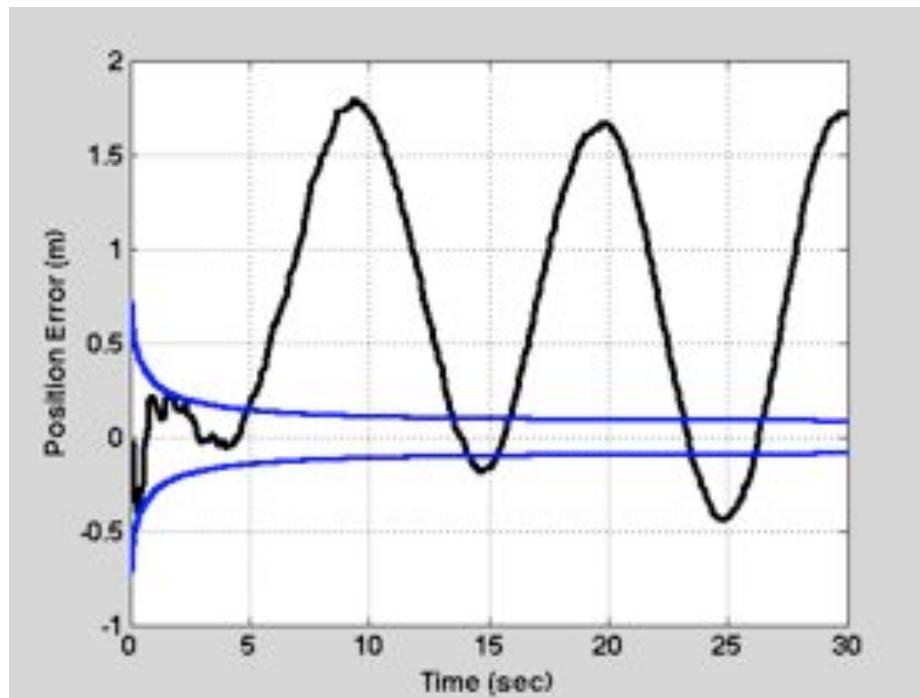
Example of Filter Saturation

- ▶ If you KNEW the acceleration, then this is what you would see using an EKF



Example of Filter Saturation

- ▶ You don't know the acceleration, so this is what you DO see with an EKF



Example of Filter Saturation

- ▶ We need a way to add expected process noise to our filter.
- ▶ Many ways to do that:
 - State Noise Compensation
 - Constant noise
 - Piecewise constant noise
 - Correlated noise
 - White noise ←———— Example
 - Dynamic Model Compensation
 - 1st order linear stochastic noise
 - More complex model compensations



Example of SNC

- ▶ Let's say that we know that our particle is being perturbed by some acceleration.
- ▶ We characterize it by simple white noise
(note: we know that it is far more structured than that!)

$$\ddot{x}(t) = \eta(t) = u(t)$$

$u(t) \equiv$ white noise

stationary Gaussian process with a mean of zero and a variance of $\sigma_u^2 \delta(t - \tau)$

(Dirac delta function)



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Example of SNC

- ▶ Adjust our system accordingly

$$\mathbf{X}(t) = \begin{bmatrix} x(t) \\ \dot{x}(t) \end{bmatrix}$$

$$\dot{\mathbf{X}}(t) = \begin{bmatrix} \dot{x}(t) \\ \ddot{x}(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(t) \\ \dot{x}(t) \end{bmatrix} \quad (\text{No perturbation})$$

$$\dot{\mathbf{X}}(t) = \begin{bmatrix} \dot{x}(t) \\ \ddot{x}(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(t) \\ \dot{x}(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

A

B



Example of SNC

- ▶ We can show that the time update for the sequential algorithm does not change:
(we'll show this later)

$$\bar{\mathbf{x}}(t) = \Phi(t, t_{k-1})\hat{\mathbf{x}}_{k-1}$$

- ▶ The time update for the covariance *does* change (we'll derive it later)

$$\bar{P}_k = \Phi_{k,k-1} P_{k-1} \Phi_{k,k-1}^T + Q_k$$



Example of SNC

- ▶ Introduce the *Process Noise Covariance Matrix Q*

$$Q_\eta(t) = \sigma_u^2 \int_{t_0}^t \Phi(t, \tau) BB^T \Phi^T(\tau) d\tau$$

(this equation is specific to this example and we'll derive the general equation later)

- ▶ For our problem, we have Phi and B. Perform the integration and we find

$$Q_\eta(t) = \sigma_u^2 \begin{bmatrix} \frac{1}{3}(t - t_0)^3 & \frac{1}{2}(t - t_0)^2 \\ \frac{1}{2}(t - t_0)^2 & t - t_0 \end{bmatrix}$$



Example of SNC

- ▶ Further, since the time update interval is always 0.1 sec (10 Hz observation frequency)

$$Q_\eta(t) = \sigma_u^2 \begin{bmatrix} \frac{1}{3}(t - t_0)^3 & \frac{1}{2}(t - t_0)^2 \\ \frac{1}{2}(t - t_0)^2 & t - t_0 \end{bmatrix}$$

$$Q_\eta(t) = \sigma_u^2 \begin{bmatrix} \frac{0.001}{3} \text{ sec}^3 & \frac{0.01}{2} \text{ sec}^2 \\ \frac{0.01}{2} \text{ sec}^2 & 0.1 \text{ sec} \end{bmatrix}$$

σ_u^2 has units of m^2/s^3



Example of SNC

$$Q_\eta(t) = \sigma_u^2 \begin{bmatrix} \frac{0.001}{3} \text{ sec}^3 & \frac{0.01}{2} \text{ sec}^2 \\ \frac{0.01}{2} \text{ sec}^2 & 0.1 \text{ sec} \end{bmatrix}$$

- ▶ Implication: original deterministic constant velocity model of the motion is modified to include
 - Random component
 - Constant-diffusion Brownian motion process
 - σ_u^2 is the *diffusion coefficient* and is a *tuning parameter*
- ▶ Remember: this is a very broad, simple process noise compensation. This assumes white noise acceleration.



Example of SNC

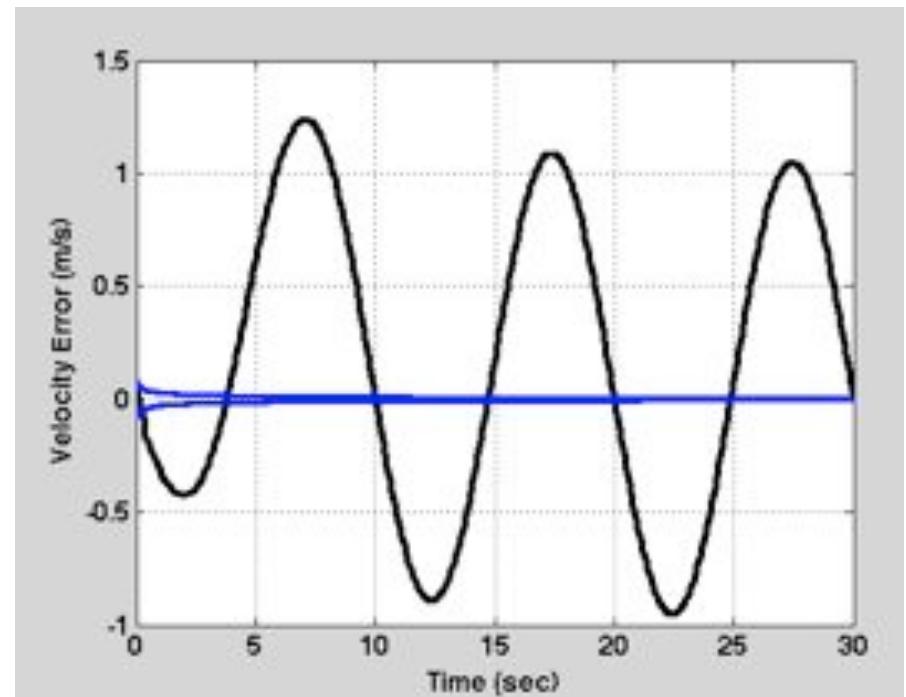
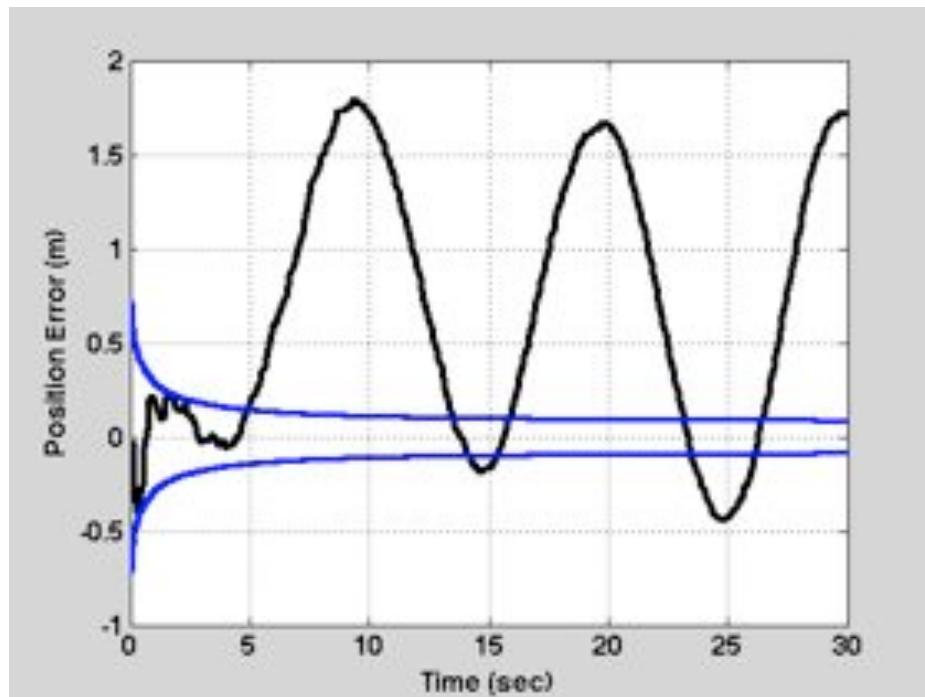
$$Q_\eta(t) = \sigma_u^2 \begin{bmatrix} \frac{0.001}{3} \text{ sec}^3 & \frac{0.01}{2} \text{ sec}^2 \\ \frac{0.01}{2} \text{ sec}^2 & 0.1 \text{ sec} \end{bmatrix}$$

- ▶ New time update:
- ▶ Old: $\bar{P}_k = \Phi(t_k, t_{k-1}) P_{k-1} \Phi^T(t_k, t_{k-1})$
- ▶ New for Example: $\bar{P}_k = \Phi_{k,k-1} P_{k-1} \Phi_{k,k-1}^T + Q_\eta(t_k)$
- ▶ General: $\bar{P}_k = \Phi_{k,k-1} P_{k-1} \Phi_{k,k-1}^T + \Gamma_{k,k-1} Q_k \Gamma_{k,k-1}^T$



Example of SNC

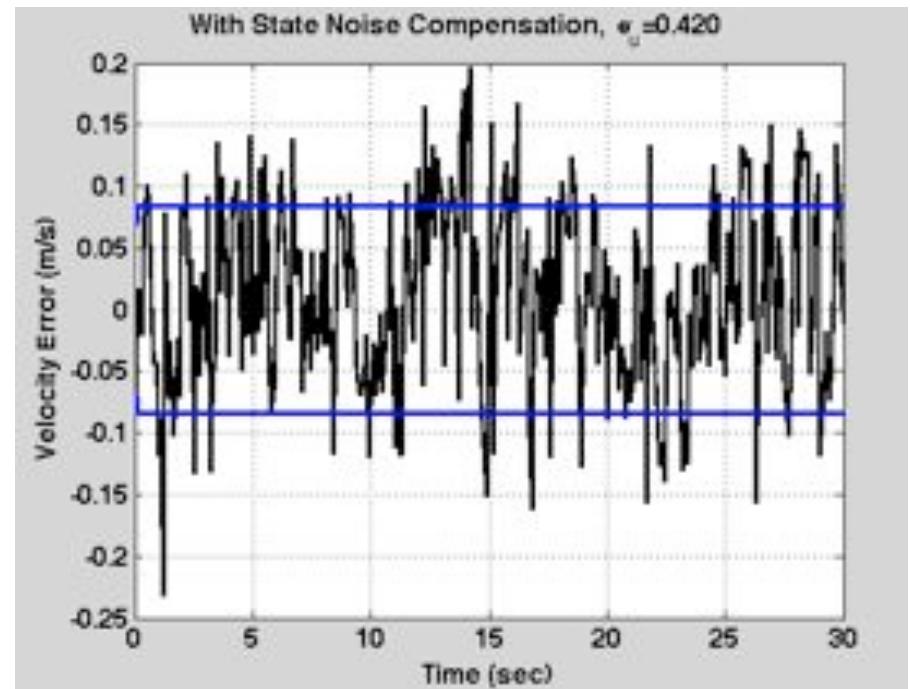
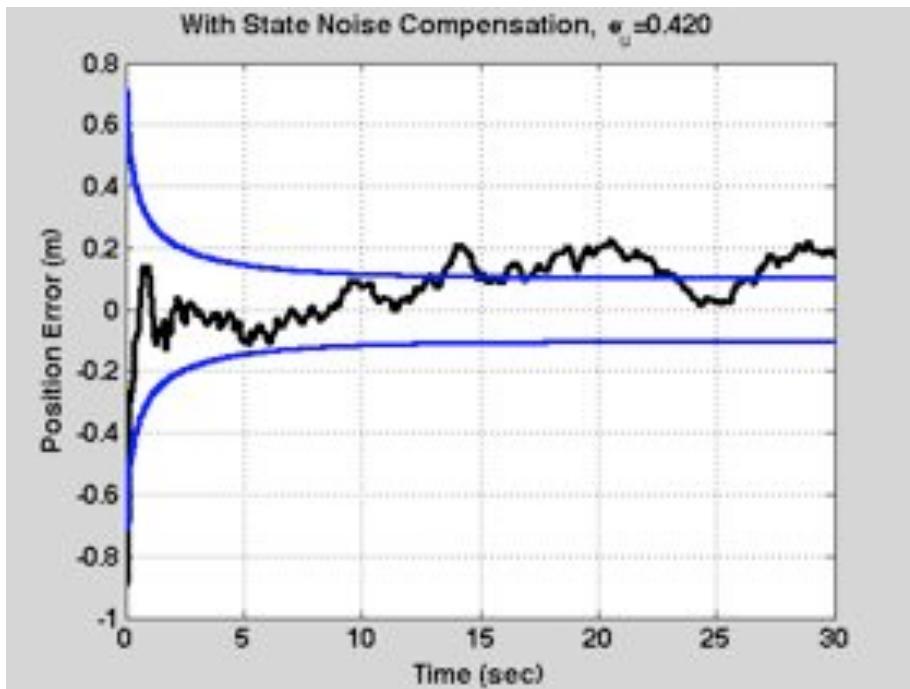
► Old Result



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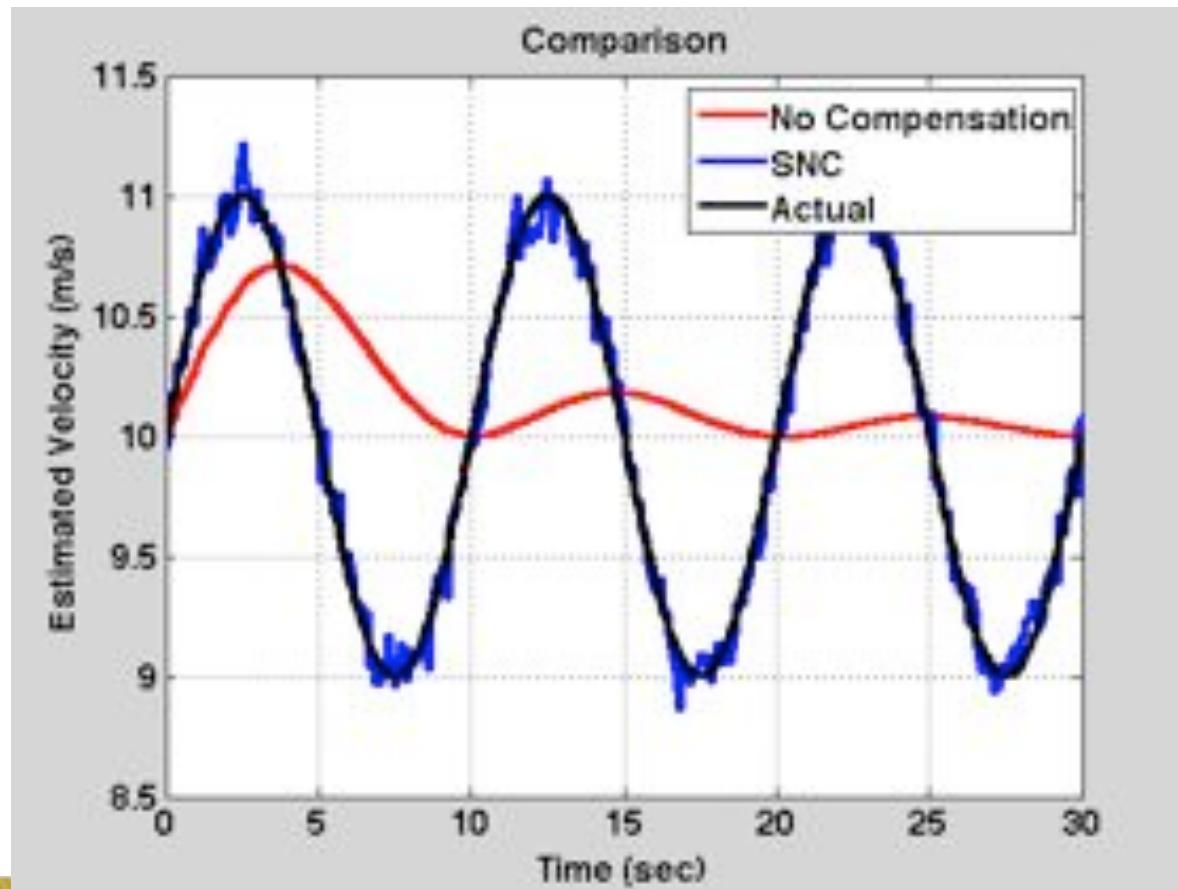
Example of SNC

► New Result



Example of SNC

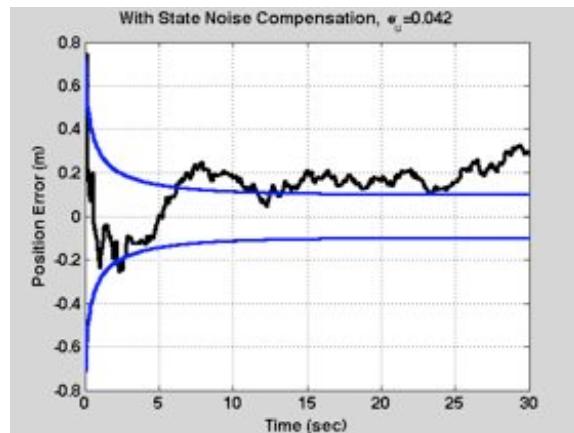
► Comparison of estimated velocity



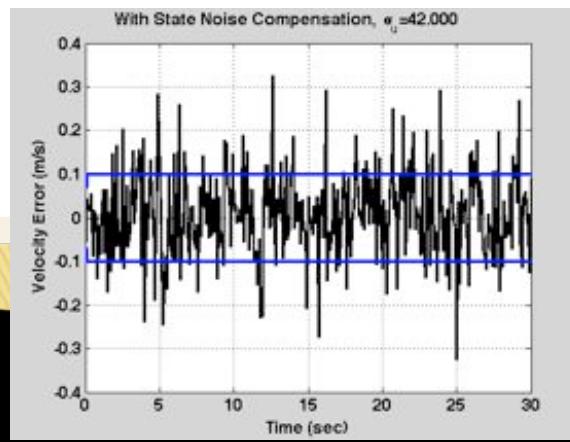
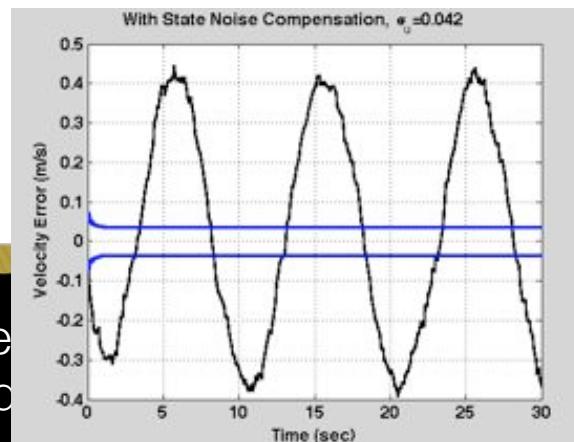
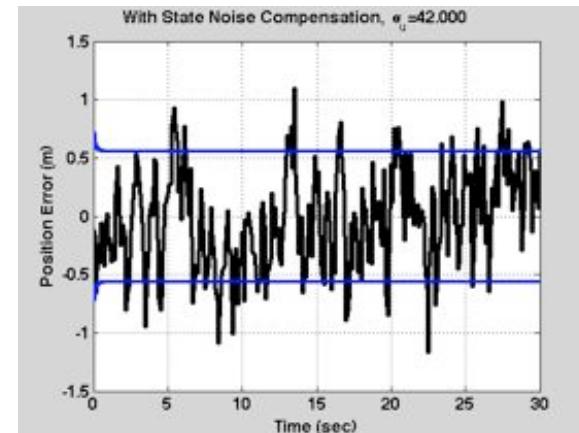
Example of SNC

- ▶ This does depend on the value of σ_u^2

Too low: not enough help



Too high: no carry-over information



Example Compensations Revisited

- ▶ We need a way to add expected process noise to our filter.
- ▶ Many ways to do that:
 - State Noise Compensation
 - Constant noise
 - Piecewise constant noise
 - Correlated noise
 - White noise ←———— Example 1
 - Dynamic Model Compensation
 - 1st order linear stochastic noise ←———— Example 2
 - More complex model compensations



Example DMC

- ▶ Let's reformulate our compensation assumption.
- ▶ Rather than assuming that our particle is under the influence of some white noise acceleration, let's assume there's some structure.
- ▶ DMC assumes that the unknown acceleration can be characterized as:
 - 1st-order linear stochastic differential equation
 - Known as the Langevin equation

$$\dot{\eta}(t) + \beta\eta(t) = u(t)$$



Example DMC

- ▶ What does this expression *mean*?

$$\dot{\eta}(t) + \beta\eta(t) = u(t)$$

- ▶ $\eta(t)$ is our estimated acceleration
- ▶ β is the inverse of a correlation time (another tuning parameter!)

$$\beta = \frac{1}{\tau}$$

- ▶ $u(t)$ is a white zero-mean Gaussian process

- ▶ Solution: $\eta(t) = \eta_0 e^{-\beta(t-t_0)} + \int_{t_0}^t e^{-\beta(t-\tau)} u(\tau) d\tau$

Deterministic

Stochastic



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Example DMC

- ▶ We can add the deterministic component of this acceleration to our state.

$$\eta(t) = \eta_0 e^{-\beta(t-t_0)} + \int_{t_0}^t e^{-\beta(t-\tau)} u(\tau) d\tau$$

Deterministic Stochastic

$$\mathbf{X}(t) = \begin{bmatrix} x(t) \\ \dot{x}(t) \\ \eta_D(t) \end{bmatrix}$$



Example DMC

- ▶ Integrate this to observe its effects on the dynamical model

$$\eta(t) = \eta_0 e^{-\beta(t-t_0)} + \int_{t_0}^t e^{-\beta(t-\tau)} u(\tau) d\tau$$

Deterministic

Stochastic

$$\mathbf{X}(t) = \begin{bmatrix} x(t) \\ \dot{x}(t) \\ \eta_D(t) \end{bmatrix} = \begin{bmatrix} x_0 + \dot{x}_0(t - t_0) + \frac{\eta_0}{\beta}(t - t_0) + \frac{\eta_0}{\beta^2} (e^{-\beta(t-t_0)} - 1) \\ \dot{x}_0 + \frac{\eta_0}{\beta} (1 - e^{-\beta(t-t_0)}) \\ \eta_0 e^{-\beta(t-t_0)} \end{bmatrix}$$



Example DMC

- ▶ Note: tau (the correlation constant) may also be added to the state to produce a 4-element state.
- ▶ This example proceeds by setting tau to be a constant.
 - Users could fiddle with tau to tune the filter.

$$\mathbf{X}(t) = \begin{bmatrix} x(t) \\ \dot{x}(t) \\ \eta_D(t) \end{bmatrix} = \begin{bmatrix} x_0 + \dot{x}_0(t - t_0) + \frac{\eta_0}{\beta}(t - t_0) + \frac{\eta_0}{\beta^2}(e^{-\beta(t-t_0)} - 1) \\ \dot{x}_0 + \frac{\eta_0}{\beta}(1 - e^{-\beta(t-t_0)}) \\ \eta_0 e^{-\beta(t-t_0)} \end{bmatrix}$$



Example DMC

- ▶ Re-derive the H-tilde, A, Phi, etc matrices to accommodate the new state parameters.
- ▶ Results:
 - A great estimate of the state
 - A much wider range of acceptable values for the tuning parameters – which is very useful when we don't know the answer!
 - A good estimate of the unmodeled acceleration profile – which may be used as information to track down missing pieces of one's dynamical model.

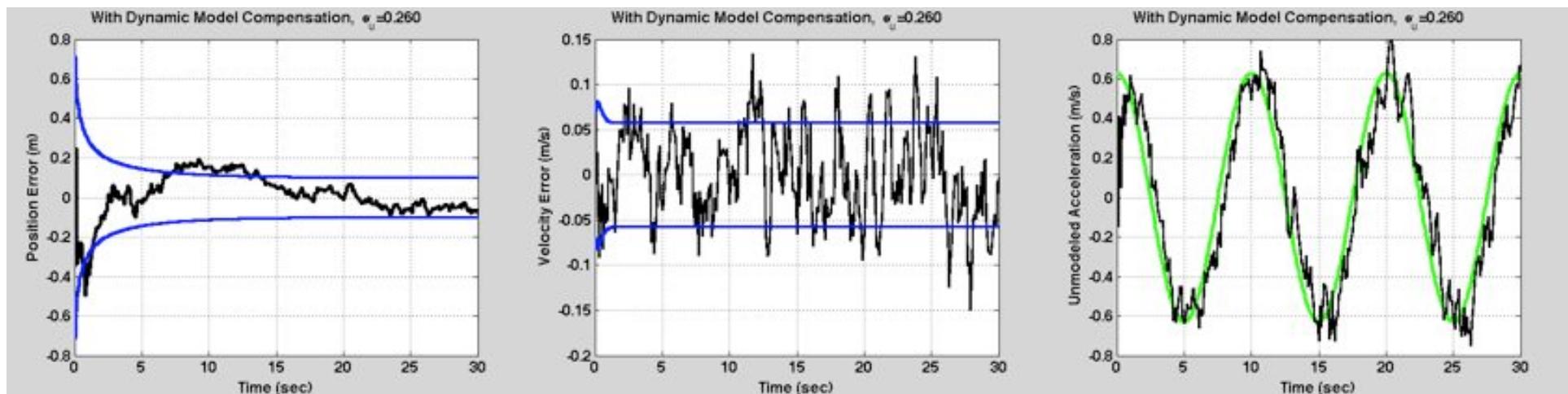


Example DMC

► Estimated State

$$\tau = 200 \text{ sec}$$

$$\sigma_u = 0.26 \text{ m/sec}^{5/2}$$



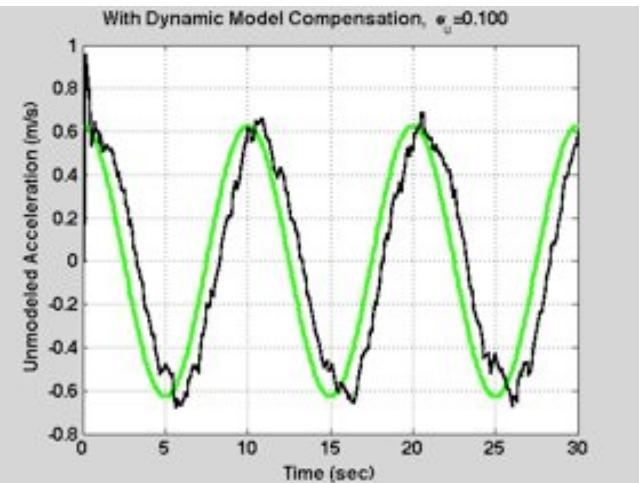
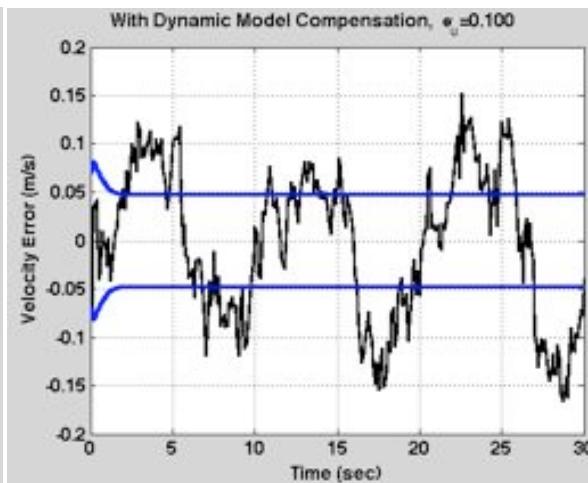
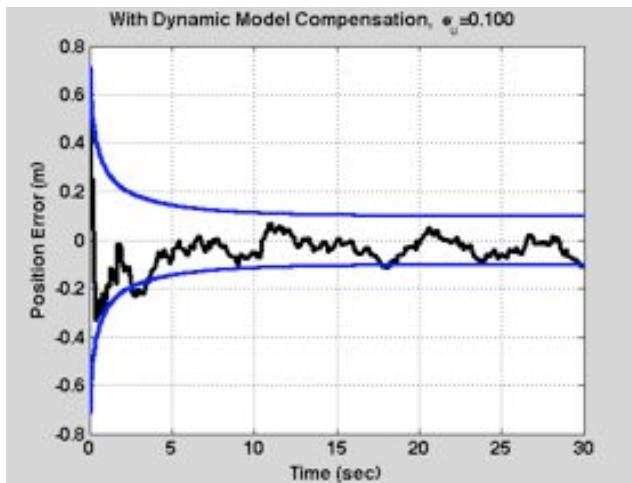
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Example DMC

► Estimated State

$$\tau = 200 \text{ sec}$$

$$\sigma_u = 0.1 \text{ m/sec}^{5/2}$$



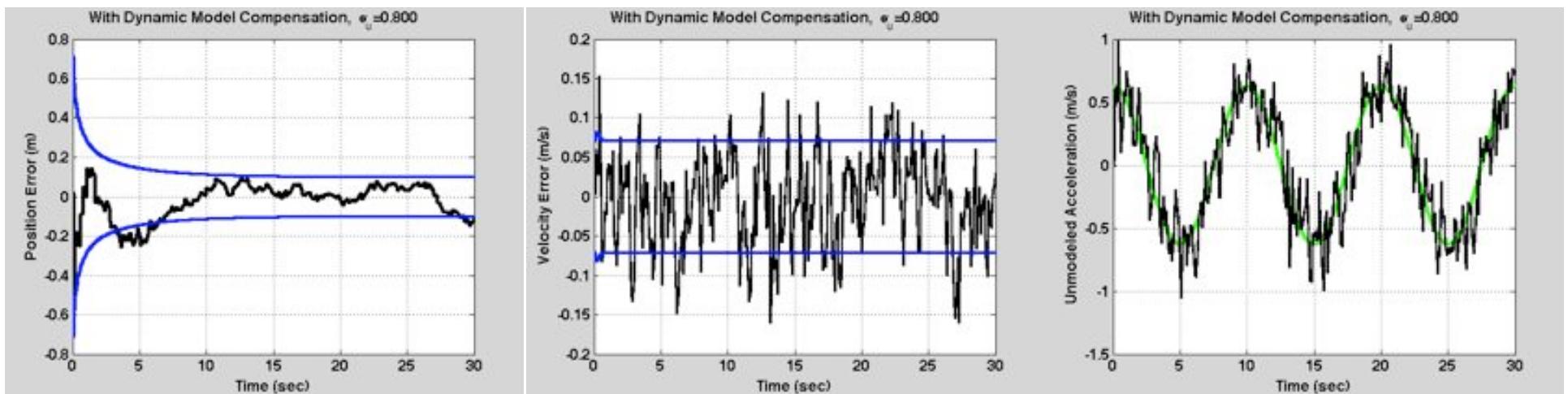
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Example DMC

► Estimated State

$$\tau = 200 \text{ sec}$$

$$\sigma_u = 0.8 \text{ m/sec}^{5/2}$$



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Example DMC

- ▶ DMC is less sensitive to tuning than SNC

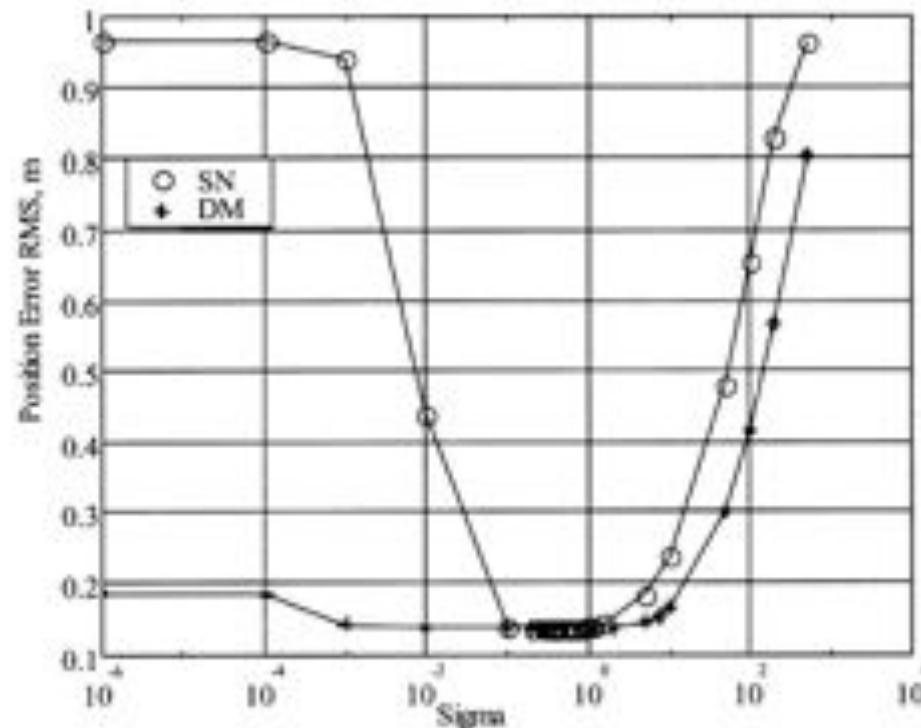


Figure F.3.2: RMS Position error as a function of σ_u for the two-state SNC and three-state DMC filters.



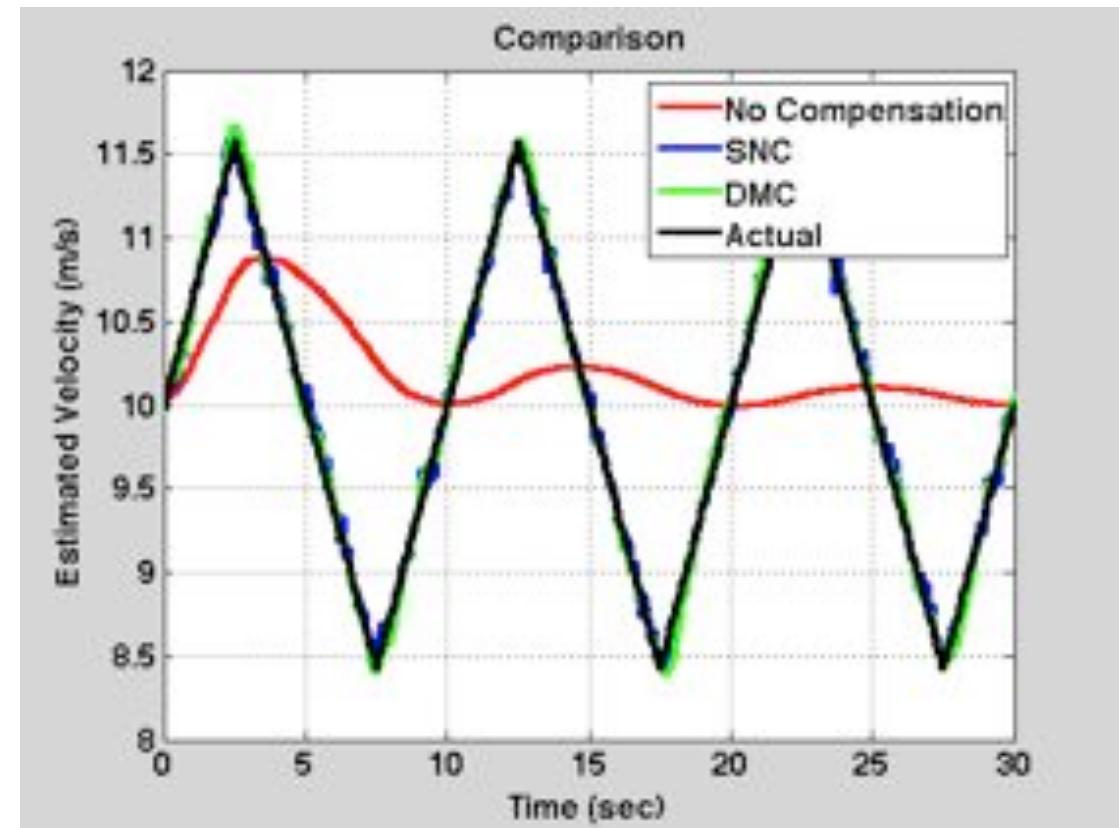
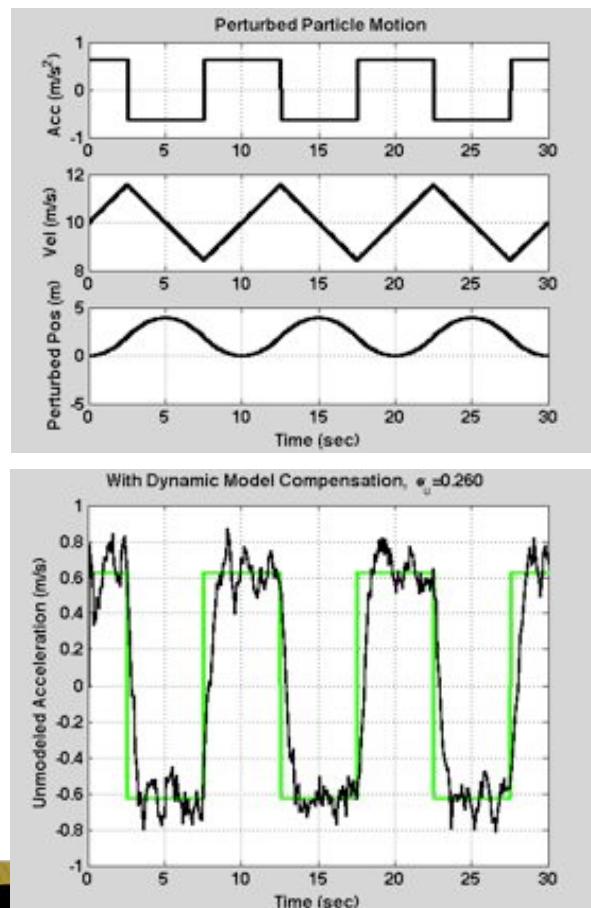
Example SNC / DMC

- ▶ What if we changed the perturbing acceleration?
 - Triangle wave
 - Square wave
 - Constant acceleration
 - White noise



Example SNC / DMC

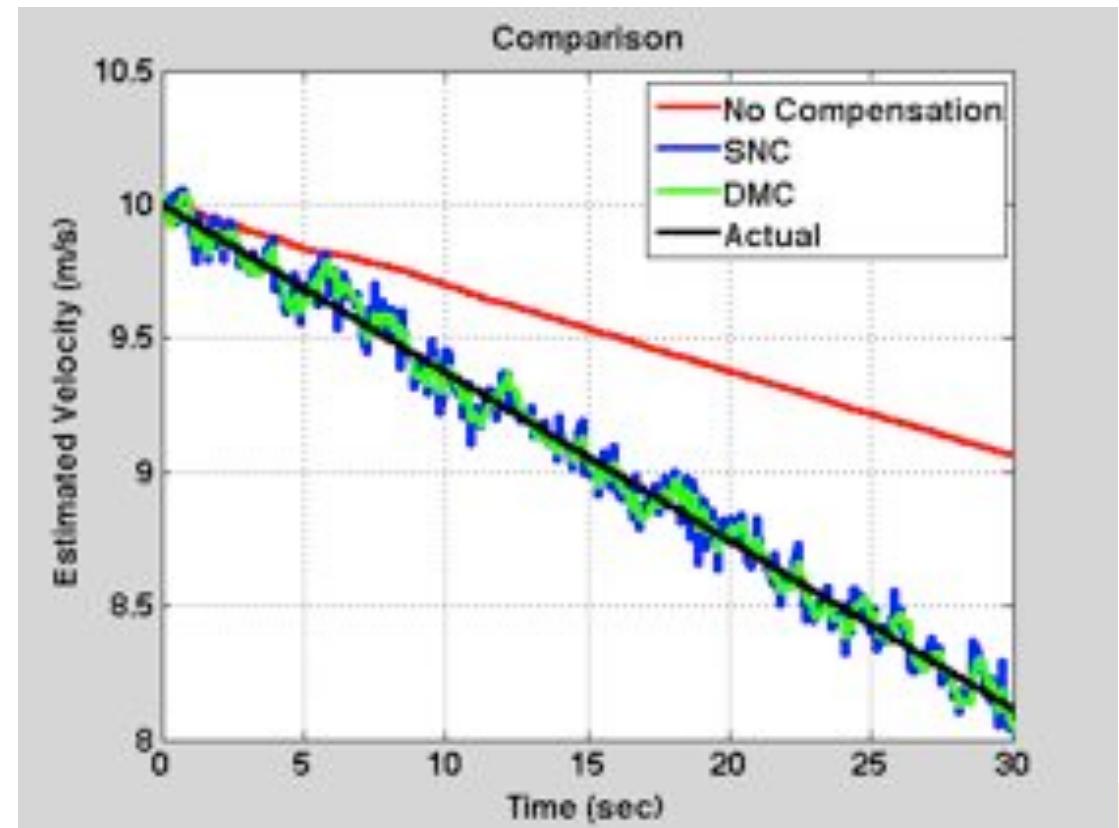
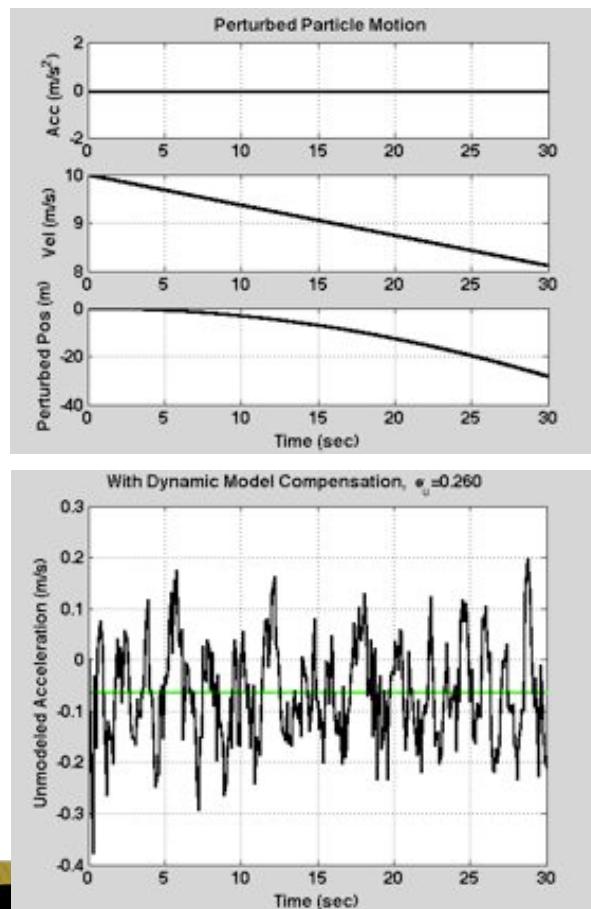
► Square Wave



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Example SNC / DMC

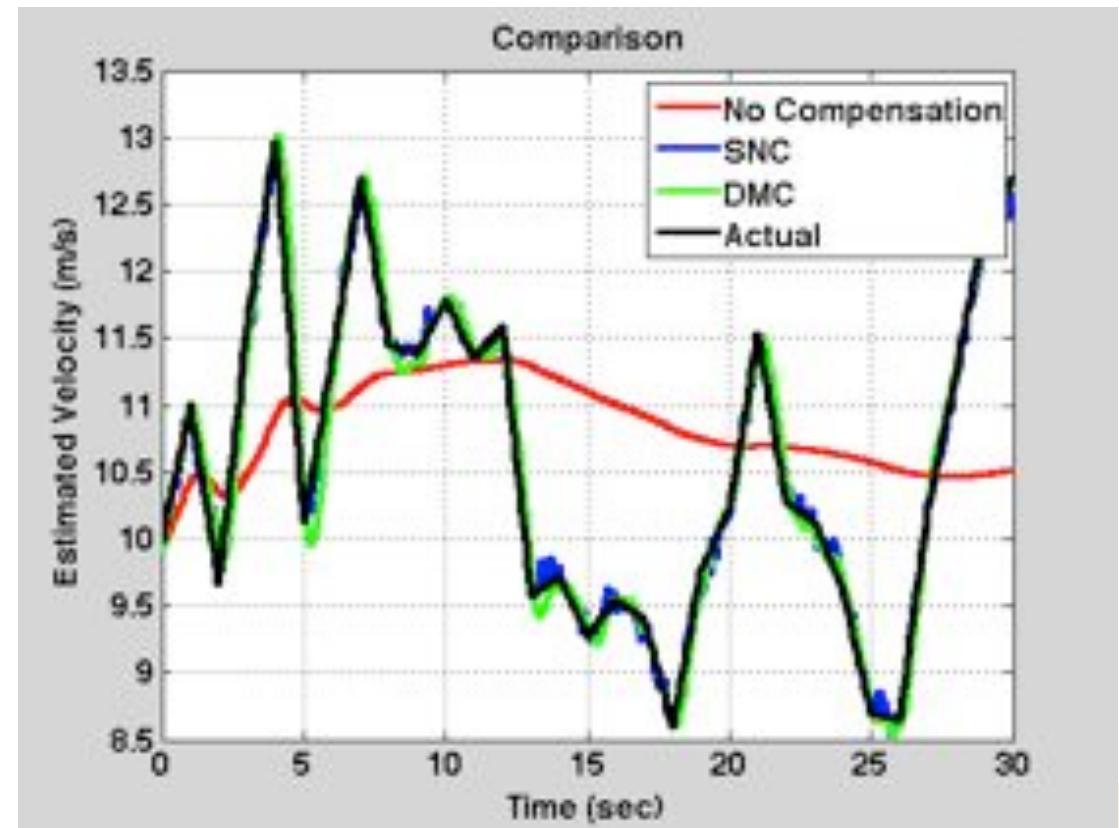
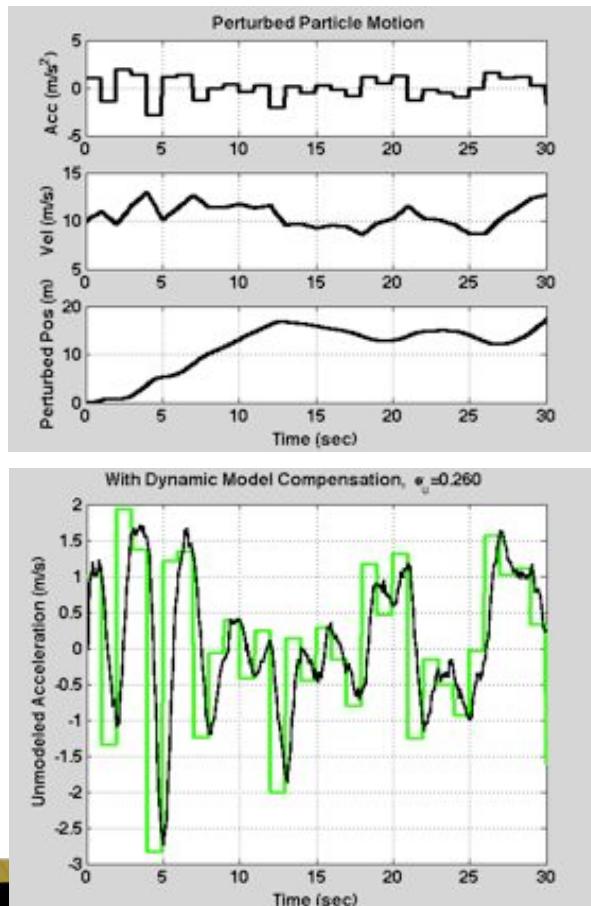
Constant Acceleration



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Example SNC / DMC

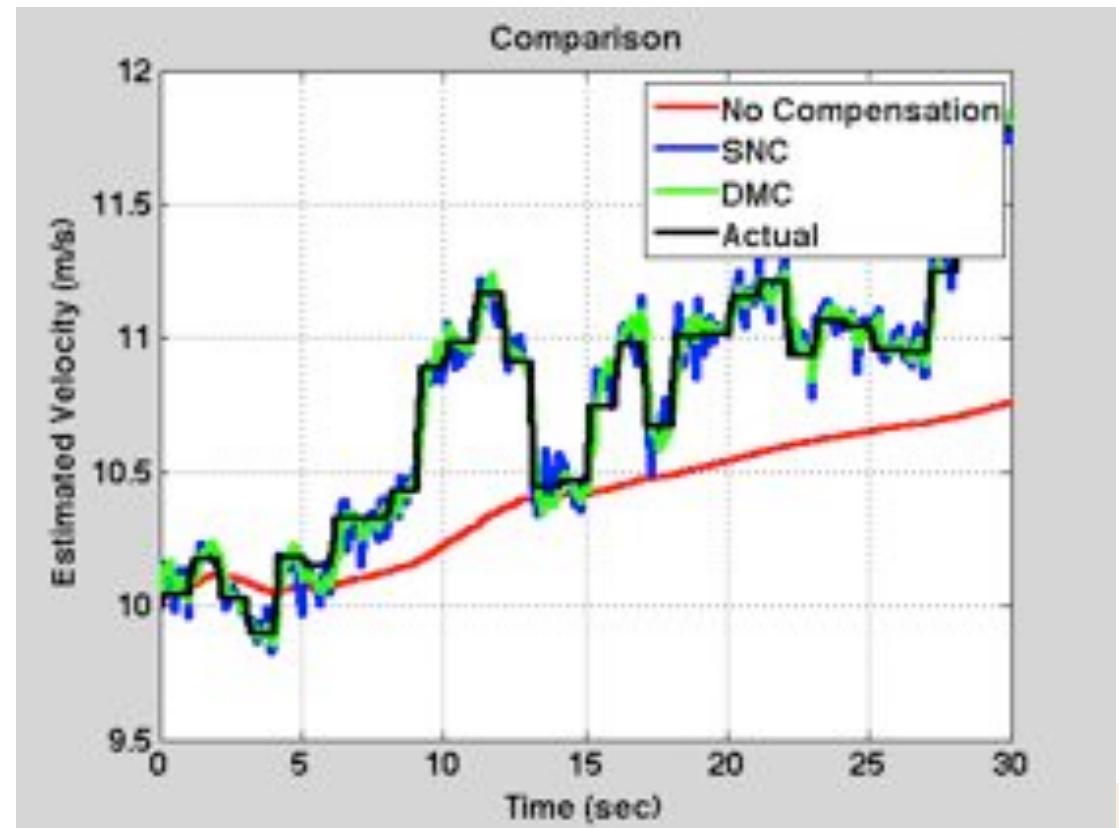
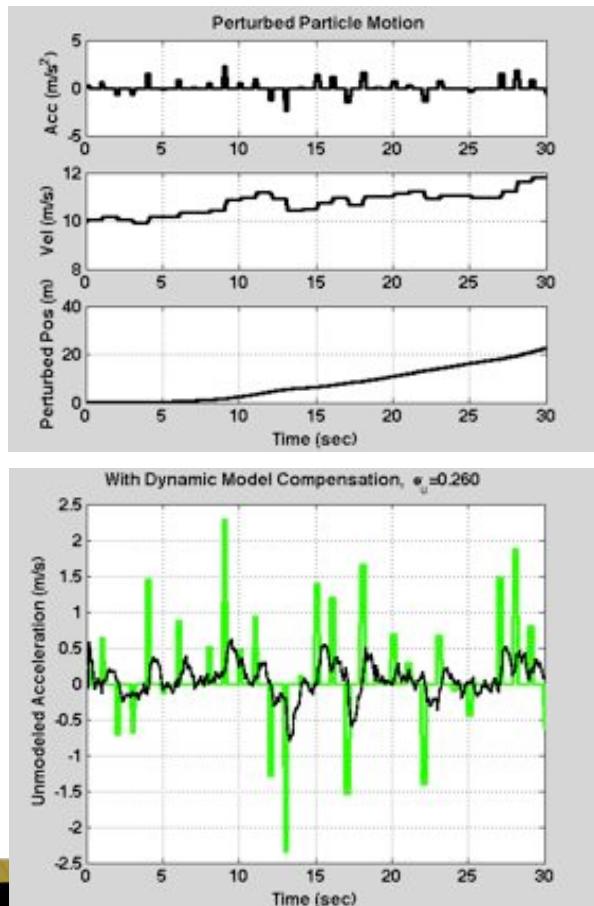
► ~White Noise



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Example SNC / DMC

► ~Impulses



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Status

- ▶ Shown an example
- ▶ Now we'll derive the SNC and DMC formulations
- ▶ Then we'll apply them to our satellite state estimation problem.
- ▶ But first! A quick break.



Derivation of State Noise Compensation

- ▶ Our setting:
 - We are not modeling everything perfectly – and we either suspect it or know it.
- ▶ State dynamics of a linear system under the influence of process noise:

$$\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$$

- $A(t)$ and $B(t)$ are known.
- $\mathbf{u}(t)$ is $m \times 1$
- $B(t)$ is $n \times m$



Derivation of State Noise Compensation

- ▶ State dynamics of a linear system under the influence of process noise:

$$\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$$

- $\mathbf{u}(t)$ can include all kinds of processes.
 - White noise, constant noise, piecewise constant, correlated, etc.
- The standard *State Noise Compensation* (SNC) algorithm assumes that $\mathbf{u}(t)$ is white noise
- Dynamic Model Compensation assumes that $\mathbf{u}(t)$ is more complex, as we saw earlier.
- SNC: $E[\mathbf{u}(t)] = 0$

$$E[\mathbf{u}(t)\mathbf{u}^T(t)] = Q(t)\delta(t - \tau)$$



Derivation of State Noise Compensation

- ▶ Solve the linear system via a method of variation of parameters.
- ▶ Homogeneous equation: $\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t)$
- ▶ Solution of the form: $\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}_0$
- ▶ Select \mathbf{C}_0 as a function of time so that
$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}(t)$$



Derivation of State Noise Compensation

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}(t)$$

► Then we can take the derivative to find

$$\dot{\mathbf{x}}(t) = \dot{\Phi}(t, t_0)\mathbf{C}(t) + \Phi(t, t_0)\dot{\mathbf{C}}(t)$$



Derivation of State Noise Compensation

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}(t)$$

- ▶ Then we can take the derivative to find

$$\dot{\mathbf{x}}(t) = \dot{\Phi}(t, t_0)\mathbf{C}(t) + \Phi(t, t_0)\dot{\mathbf{C}}(t)$$

- ▶ Plugging this into $\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$

$$\dot{\Phi}(t, t_0)\mathbf{C}(t) + \Phi(t, t_0)\dot{\mathbf{C}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$$



Derivation of State Noise Compensation

$$\dot{\Phi}(t, t_0) \mathbf{C}(t) + \Phi(t, t_0) \dot{\mathbf{C}}(t) = A(t) \mathbf{x}(t) + B(t) \mathbf{u}(t)$$

- ▶ Recall our old friend: $\dot{\Phi}(t, t_0) = A(t)\Phi(t, t_0)$

$$A(t)\Phi(t, t_0) \mathbf{C}(t) + \Phi(t, t_0) \dot{\mathbf{C}}(t) = A(t) \mathbf{x}(t) + B(t) \mathbf{u}(t)$$



Derivation of State Noise Compensation

$$A(t)\Phi(t, t_0)\mathbf{C}(t) + \Phi(t, t_0)\dot{\mathbf{C}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$$

► Substitute the solution from above:

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}(t)$$

~~$$A(t)\Phi(t, t_0)\mathbf{C}(t) + \Phi(t, t_0)\dot{\mathbf{C}}(t) = A(t)\Phi(t, t_0)\mathbf{C}(t) + B(t)\mathbf{u}(t)$$~~

~~$$\Phi(t, t_0)\dot{\mathbf{C}}(t) = B(t)\mathbf{u}(t)$$~~



Derivation of State Noise Compensation

$$\Phi(t, t_0) \dot{\mathbf{C}}(t) = B(t) \mathbf{u}(t)$$

► Hence $\dot{\mathbf{C}}(t) = \Phi^{-1}(t, t_0) B(t) \mathbf{u}(t)$

► Integrating yields

$$\mathbf{C}(t) = \mathbf{C}_0 + \int_{t_0}^t \Phi^{-1}(\tau, t_0) B(\tau) \mathbf{u}(\tau) d\tau$$



Derivation of State Noise Compensation

► Substituting $\mathbf{C}(t) = \mathbf{C}_0 + \int_{t_0}^t \Phi^{-1}(\tau, t_0) B(\tau) \mathbf{u}(\tau) d\tau$

into $\mathbf{x}(t) = \Phi(t, t_0) \mathbf{C}(t)$

yields $\mathbf{x}(t) = \Phi(t, t_0) \mathbf{C}_0 + \int_{t_0}^t \Phi(t, t_0) \Phi^{-1}(\tau, t_0) B(\tau) \mathbf{u}(\tau) d\tau$

$$\underbrace{\Phi(t, \tau)}$$

$$\mathbf{x}(t) = \Phi(t, t_0) \mathbf{C}_0 + \int_{t_0}^t \Phi(t, \tau) B(\tau) \mathbf{u}(\tau) d\tau$$



Derivation of State Noise Compensation

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{C}_0 + \int_{t_0}^t \Phi(t, \tau)B(\tau)\mathbf{u}(\tau)d\tau$$

- ▶ Initial conditions $t = t_0, \quad \mathbf{x}(t_0) = \mathbf{x}_0$

We can show that $\mathbf{C}_0 = \mathbf{x}_0$

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{x}_0 + \int_{t_0}^t \Phi(t, \tau)B(\tau)\mathbf{u}(\tau)d\tau$$



Derivation of State Noise Compensation

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{x}_0 + \int_{t_0}^t \Phi(t, \tau)B(\tau)\mathbf{u}(\tau)d\tau$$

- ▶ This equation is the general solution for the inhomogeneous equation

$$\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$$

- ▶ It indicates how the true state propagates under the influence of process noise.



Derivation of State Noise Compensation



Next, we wish to understand how the estimate of the state $\bar{\mathbf{x}}(t)$ propagates in the presence of process noise.

$$\mathbf{x}(t) = \Phi(t, t_0)\mathbf{x}_0 + \int_{t_0}^t \Phi(t, \tau)B(\tau)\mathbf{u}(\tau)d\tau$$

is a stochastic integral and cannot be evaluated in a deterministic sense.

$$\bar{\mathbf{x}}(t) = E[\mathbf{x}(t)|\mathbf{y}_{k-1}] \quad \text{for} \quad t \geq t_{k-1}$$

$$\dot{\bar{\mathbf{x}}}(t) = E[\dot{\mathbf{x}}(t)|\mathbf{y}_{k-1}] = E[\{A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)\}|\mathbf{y}_{k-1}]$$

$$\text{but} \quad E[\mathbf{u}(t)|\mathbf{y}_{k-1}] = 0$$



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Derivation of State Noise Compensation



so

$$\dot{\bar{\mathbf{x}}}(t) = A(t)E[\mathbf{x}(t)|\mathbf{y}_{k-1}] = A(t)\bar{\mathbf{x}}(t)$$

$$\dot{\bar{\mathbf{x}}}(t) = A(t)\bar{\mathbf{x}}(t)$$

We have our familiar solution

$$\bar{\mathbf{x}}(t) = \Phi(t, t_{k-1})\hat{\mathbf{x}}_{k-1}$$

This is good news! It means that we don't have to change our methods to update $\bar{\mathbf{x}}$ in our sequential algorithm.



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Derivation of State Noise Compensation

- ▶ Since $E[\mathbf{u}(t)] = 0$

we have shown that the time update of the state is unchanged

$$\bar{\mathbf{x}}(t_k) = \Phi(t_k, t_{k-1})\hat{\mathbf{x}}(t_{k-1})$$

- ▶ If the mean of the process noise is nonzero

$$E[\mathbf{u}(t)] = \bar{\mathbf{u}}$$

$$\bar{\mathbf{x}}(t_k) = \Phi(t_k, t_{k-1})\hat{\mathbf{x}}(t_{k-1}) + \Gamma(t_k, t_{k-1})\bar{\mathbf{u}}$$

(defined later)



The End

- ▶ How was break?
- ▶ HW 10 due Thursday.
- ▶ HW 11 due next week.
- ▶ Grading
- ▶ You have the tools needed to finish the project. Only things missing are bonus pieces.
- ▶ I may be out tomorrow; if you need office-hour help, check the TAs or visit me Thursday.

Sun	Mon	Tue	Wed	Thu	Fri	Sat
				1	2	3
4	5	6	7	8	9	10
11	12	13	14	15	16	17
18	19	20	21	22	23	24
25	26	27	28	29	30	

Sun	Mon	Tue	Wed	Thu	Fri	Sat
						1
2	3	4	5	6	7	8
9	10	11	12	13	14	15
16	17	18	19	20	21	22
23	24	25	26	27	28	29
30	31					

Last Day of Classes

Take-Home Exam Due

Final Project Due
All HW Due