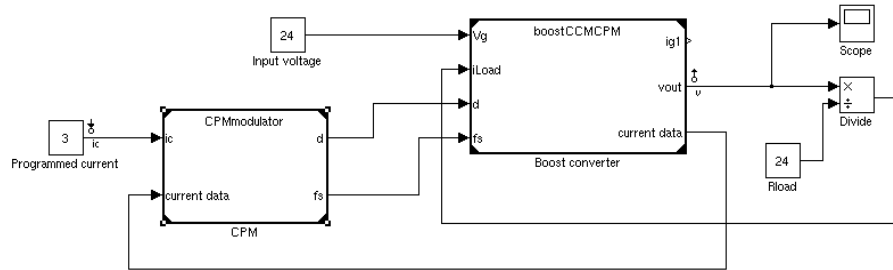


# Current mode control model: Simulink



- Use same power stage models previously described
- Need new model for CPM controller
- A single CPM controller model can be used with any power stage

## Simulink model: CPM controller block

From Tan model: equation of the CPM modulator

$$\langle i_L \rangle = d' i_1 + d i_2 = i_c - m_a d T_s - \frac{m_1 + m_2}{2} d d' T_s$$

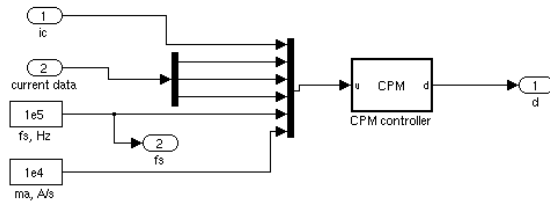
The CPM modulator effectively solves this equation for  $d(t)$ ; the following is the CCM solution:

$$d(t) = b \left[ 1 - \sqrt{1 - \frac{2f_s(i_c - \langle i_L \rangle)}{b^2(m_1 + m_2)}} \right]$$

with  $b = \frac{m_a}{(m_1 + m_2)} + \frac{1}{2}$

The modulator model must therefore evaluate this equation.

# CPM controller model, Simulink



$$d(t) = b \left[ 1 - \sqrt{1 - \frac{2f_s(i_c - \langle i_L \rangle)}{b^2(m_1 + m_2)}} \right]$$

with  $b = \frac{m_a}{(m_1 + m_2)} + \frac{1}{2}$

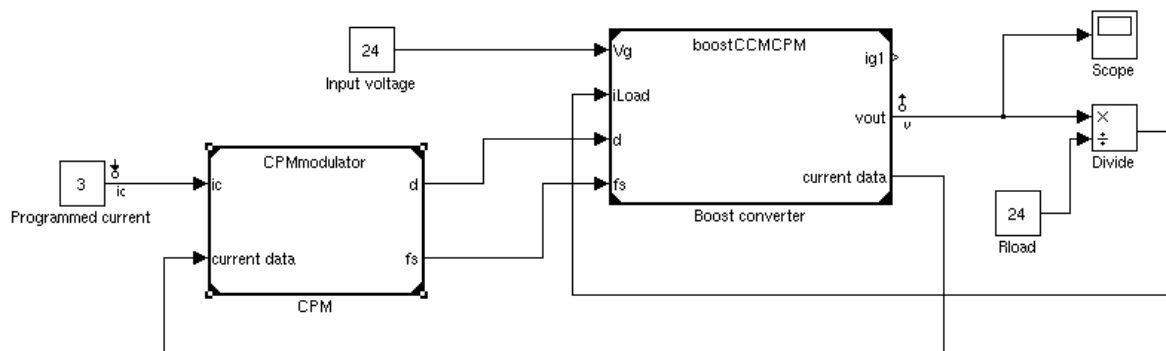
```

1 function d = CPM(u)
2 %eml
3 % Current mode controller, using Tan's model
4 % No sampling dynamics included
5 ic = u(1); % control current
6 i = u(2); % average inductor current
7 m1 = u(3); % inductor up slope
8 m2 = u(4); % inductor down slope
9 fs = u(5); % switching frequency
10 ma = u(6); % slope of artificial ramp
11 b = ma/(m1+m2) + 0.5;
12 d = max(0,min(0.9,b*(1-sqrt(1-(ic-i)*2*fs/(m1+m2)/b^2)))); % duty cycle computation, including modulator saturation limits

```

- The slopes  $m_a$ ,  $m_1$ ,  $m_2$  are supplied by the converter power stage, via the vector “current data”
- Lines 11-12 evaluate the solution for  $d(t)$  from previous slide
- Line 12 also includes saturation limits for  $d(t)$

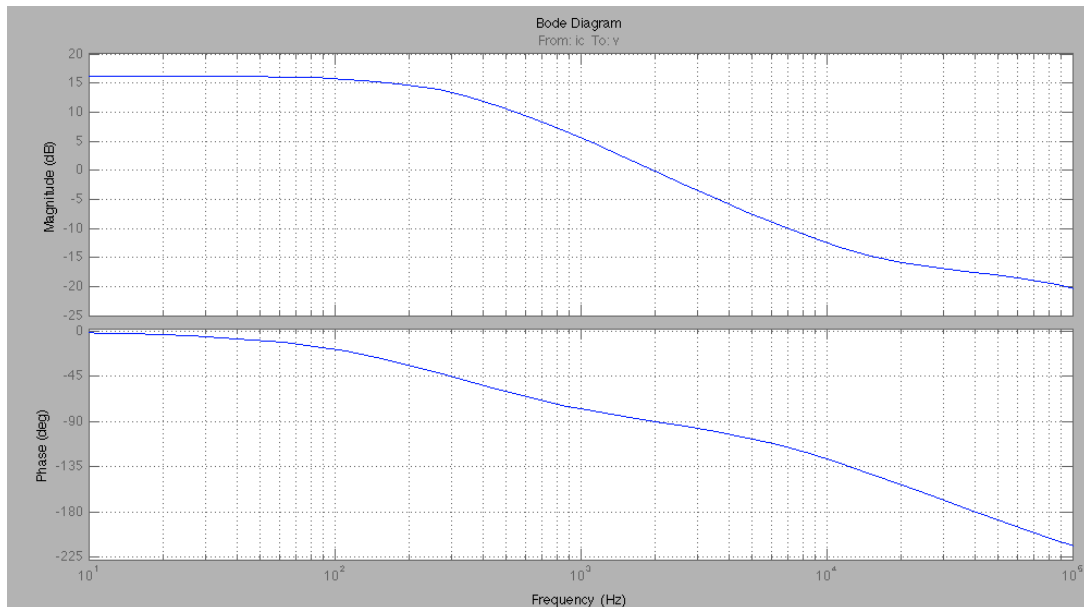
## Example: a current-mode boost converter



Setup to plot control-to-output transfer function is shown; results on next slide

# Control-to-output transfer function / Simulink

## CPM boost, previous slide



# Plotting the output impedance / Simulink

## CPM boost, previous slides

