

LM833 Dual Audio Operational Amplifier

General Description

The LM833 and LM833A are dual general purpose operational amplifiers designed with particular emphasis on performance in audio systems.

These dual amplifier ICs utilize new circuit and processing techniques to deliver low noise, high speed and wide bandwidth without increasing external components or decreasing stability. The LM833 and LM833A are internally compensated for all closed loop gains and are therefore optimized for all preamp and high level stages in PCM and HiFi systems.

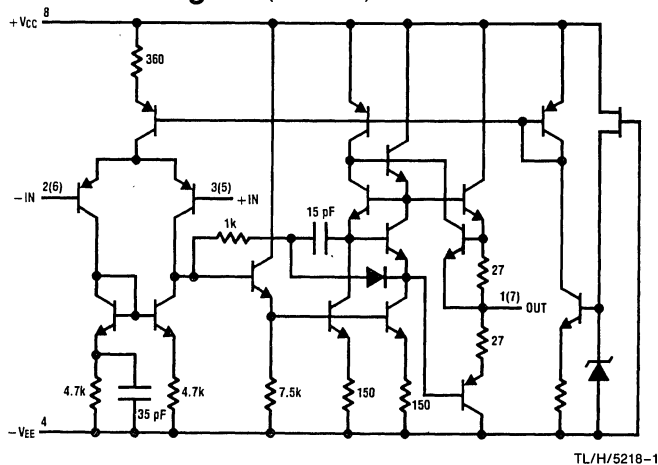
The LM833 and LM833A are pin for pin compatible with industry standard dual operational amplifiers.

The LM833A guarantees low noise for noise critical applications by 100% noise testing.

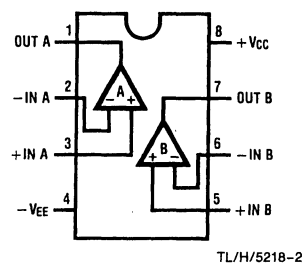
Features

- Wide dynamic range > 140 dB
- Low input noise voltage 4.5 nV/ $\sqrt{\text{Hz}}$
- High slew rate 7 V/ μs (typ)
5 V/ μs (min)
- High gain bandwidth product 15 MHz (typ)
10 MHz (min)
- Wide power bandwidth 120 kHz
- Low distortion 0.002%
- Low offset voltage 0.3 mV
- Large phase margin 60°

Schematic Diagram (1/2 LM833)

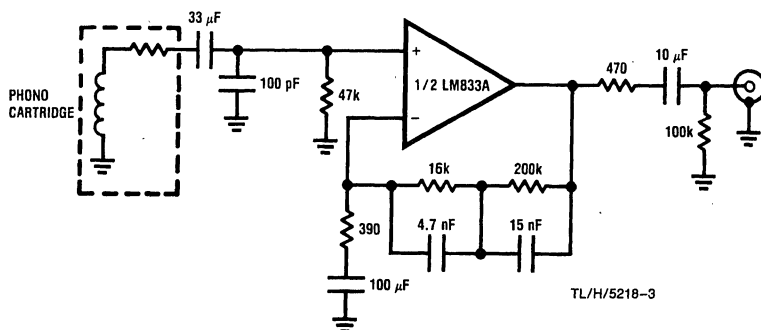


Connection Diagram



Order Number LM833N
See NS Package N08E

Typical Application RIAA Preamp



$A_v = 35 \text{ dB}$ $f = 1 \text{ kHz}$
 $E_n = 0.33 \mu\text{V}$ A Weighted
 $S/N = 90 \text{ dB}$ A Weighted, $V_{IN} = 10 \text{ mV}$
@ $f = 1 \text{ kHz}$

Absolute Maximum Ratings

Supply Voltage	V_{CC}/V_{EE}	$\pm 18V$	Power Dissipation (Note 2)	P_D	500 mW
Differential Input Voltage (Note 1)	V_{ID}	$\pm 30V$	Operating Temperature Range	T_{OPR}	$-40 \sim 85^{\circ}C$
Input Voltage Range (Note 1)	V_{IC}	$\pm 15V$	Storage Temperature Range	T_{STG}	$-60 \sim 150^{\circ}C$

DC Electrical Characteristics ($T_A = 25^{\circ}C$, $V_S = \pm 15V$)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{OS}	Input Offset Voltage	$R_S = 10\Omega$		0.3	5	mV
I_{OS}	Input Offset Current			10	200	nA
I_B	Input Bias Current			500	1000	nA
A_V	Voltage Gain	$R_L = 2\text{ k}\Omega$, $V_O = \pm 10V$	90	110		dB
V_{OM}	Output Voltage Swing	$R_L = 10\text{ k}\Omega$ $R_L = 2\text{ k}\Omega$	± 12 ± 10	± 13.5 ± 13.4		V V
V_{CM}	Input Common-Mode Range		± 12	± 14.0		V
CMRR	Common-Mode Rejection Ratio	$V_{IN} = \pm 12V$	80	100		dB
PSRR	Power Supply Rejection Ratio	$V_S = 15 \sim 5V$, $-15 \sim -5V$	80	100		dB
I_Q	Supply Current	$V_O = 0V$, Both Amps		5	8	mA

AC Electrical Characteristics ($T_A = 25^{\circ}C$, $V_S = \pm 15V$, $R_L = 2\text{ k}\Omega$)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
SR	Slew Rate	$R_L = 2\text{ k}\Omega$	5	7		V/ μs
GBWP	Gain Bandwidth Product	$f = 100\text{ kHz}$	10	15		MHz
e_{n1}	LM833A Equivalent Input Noise Voltage (Note 3)	R_{IAA} , $R_S = 470\Omega$		0.5	0.8	μV

Design Electrical Characteristics ($T_A = 25^{\circ}C$, $V_S = \pm 15V$)

The following parameters are not tested or guaranteed.

Symbol	Parameter	Conditions	Typ	Units
$\Delta V_{OS}/\Delta T$	Average Temperature Coefficient of Input Offset Voltage		2	$\mu V/^{\circ}C$
THD	Distortion	$R_L = 2\text{ k}\Omega$, $f = 20 \sim 20\text{ kHz}$ $V_{OUT} = 3\text{ Vrms}$, $A_V = 1$	0.002	%
e_{n2}	Input Referred Noise Voltage 2	$R_S = 100\Omega$, JISA	0.5	μV
e_{n3}	Input Referred Noise Voltage 3	$R_S = 100\Omega$, $f = 1\text{ kHz}$	4.5	nV/\sqrt{Hz}
i_n	Input Referred Noise Current	$f = 1\text{ kHz}$	0.7	pA/\sqrt{Hz}
PBW	Power Bandwidth	$V_O = 27\text{ Vpp}$, $R_L = 2\text{ k}\Omega$, $THD \leq 1\%$	120	kHz
f_U	Unity Gain Frequency	Open Loop	9	MHz
ϕ_M	Phase Margin	Open Loop	60	deg
	Input Referred Cross Talk	$f = 20 \sim 20\text{ kHz}$	-120	dB

Note 1: If supply voltage is less than 15V, it is equal to supply voltage.

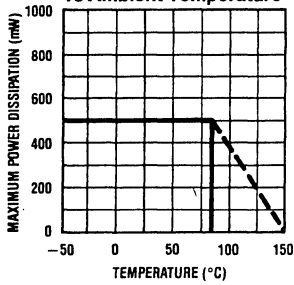
Note 2: This is the permissible value at $T_A \leq 85^{\circ}C$.

Note 3: Only the LM833A is noise tested and guaranteed.

See "Noise Measurement Circuit" for test conditions.

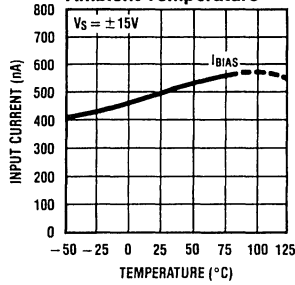
Typical Performance Characteristics

Maximum Power Dissipation vs Ambient Temperature



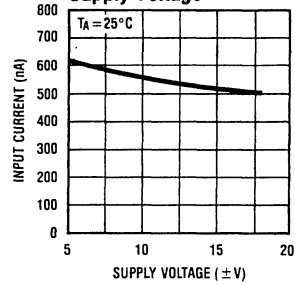
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Input Bias Current vs Ambient Temperature



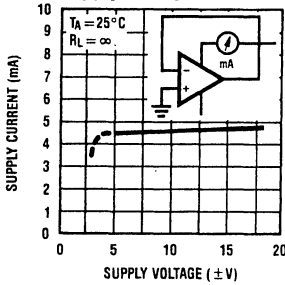
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Input Bias Current vs Supply Voltage



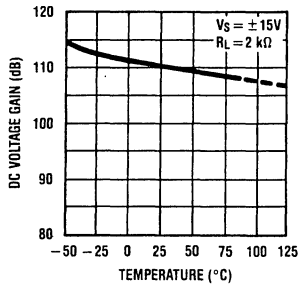
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Supply Current vs Supply Voltage



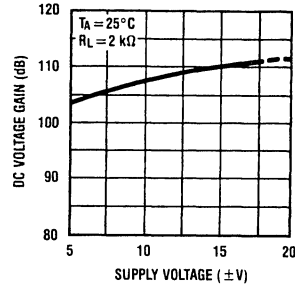
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DC Voltage Gain vs Ambient Temperature



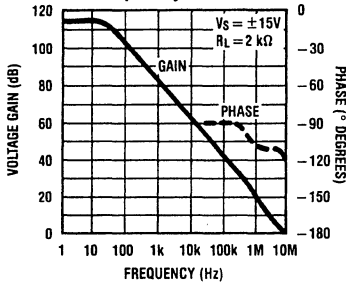
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DC Voltage Gain vs Supply Voltage



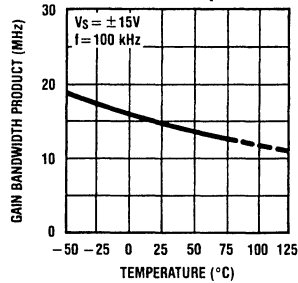
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Voltage Gain & Phase vs Frequency



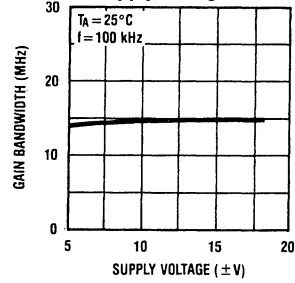
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Gain Bandwidth Product vs Ambient Temperature



TL/H/5218-11

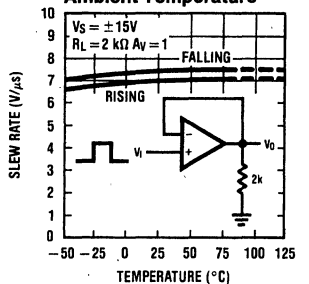
Gain Bandwidth vs Supply Voltage



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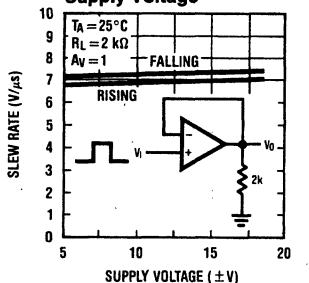
Typical Performance Characteristics (Continued)

Slew Rate vs Ambient Temperature



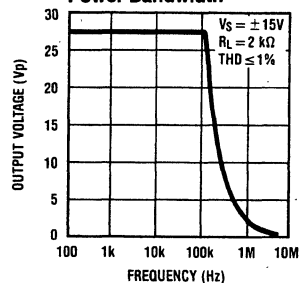
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Slew Rate vs Supply Voltage



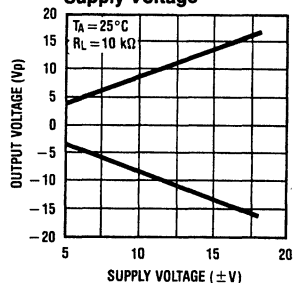
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Power Bandwidth



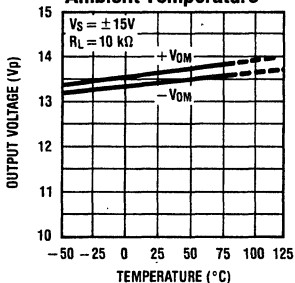
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Maximum Output Voltage vs Supply Voltage



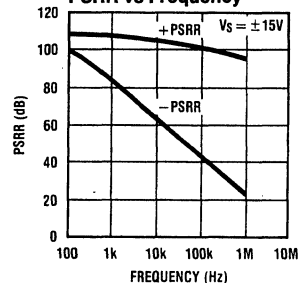
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Maximum Output Voltage vs Ambient Temperature



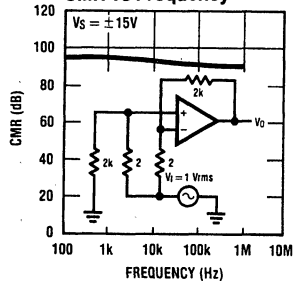
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PSRR vs Frequency



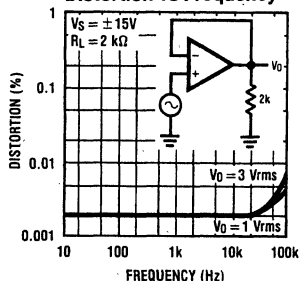
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CMR vs Frequency



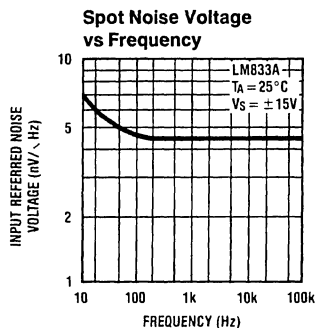
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Distortion vs Frequency

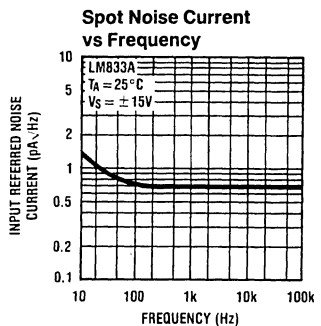


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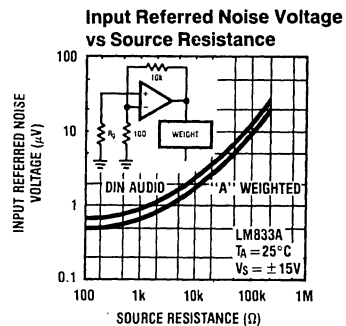
Typical Performance Characteristics (Continued)



TL/H/5218-21

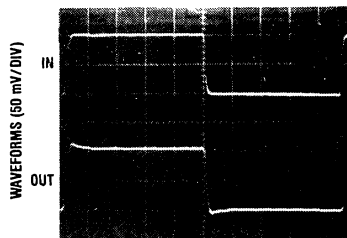


TL/H/5218-22



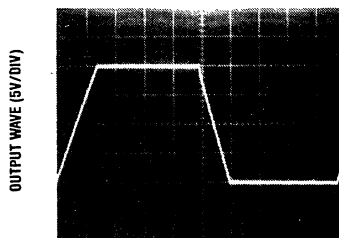
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Noninverting Amp

TIME (0.2 μs /DIV)

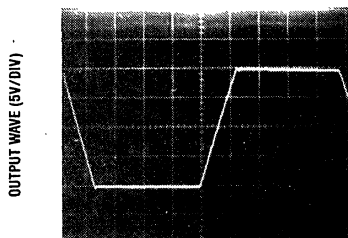
TL/H/5218-24

Noninverting Amp

TIME (2 μs /DIV)

TL/H/5218-25

Inverting Amp

TIME (2 μs /DIV)

TL/H/5218-26

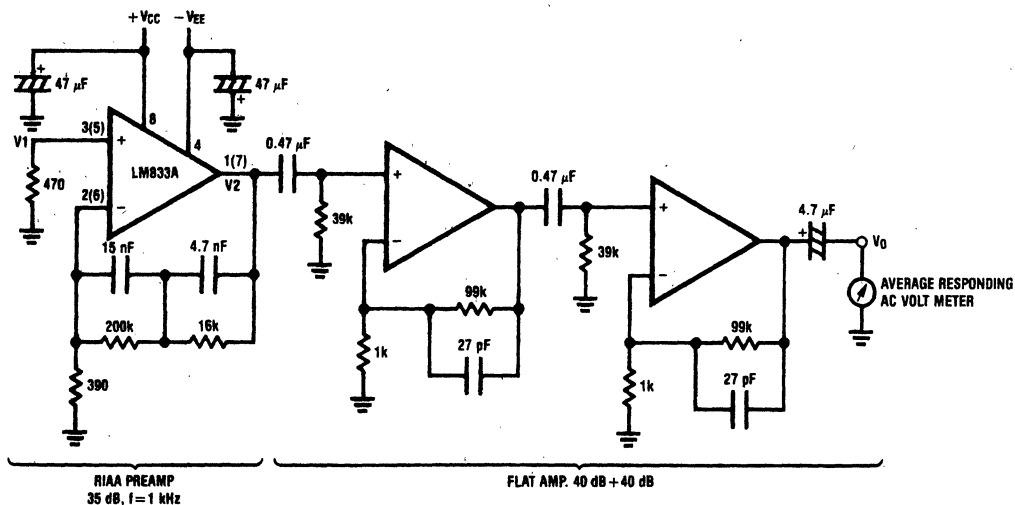
Application Hints

The LM833 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 50 pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 50 pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

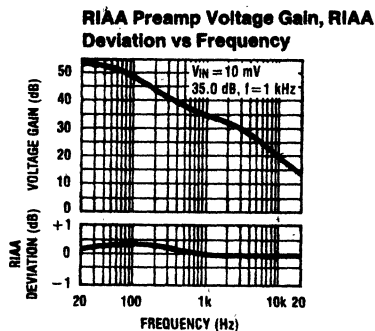
Noise Measurement Circuit

Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

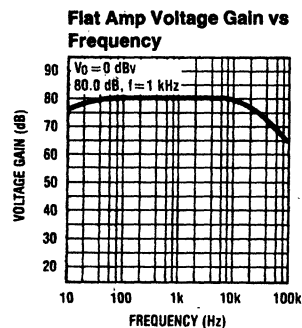


TL/H/5218-27

Total Gain: 115 dB @ $f = 1$ kHz
Input Referred Noise Voltage: $e_n = V_0/560,000$ (V)



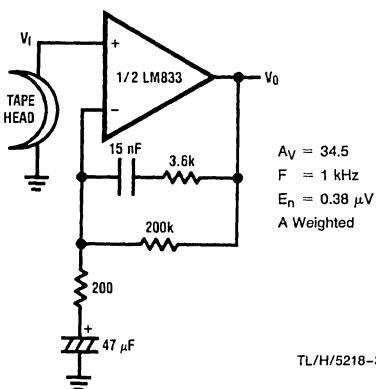
TL/H/5218-28



TL/H/5218-29

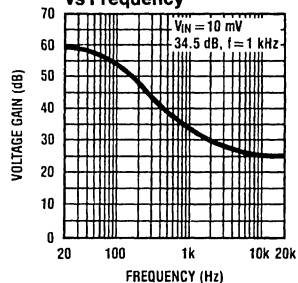
Typical Applications

NAB Preamp



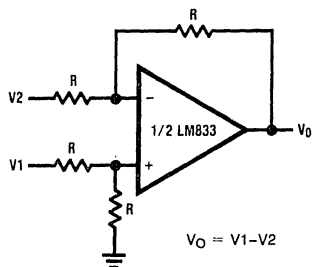
TL/H/5218-30

NAB Preamp Voltage Gain vs Frequency



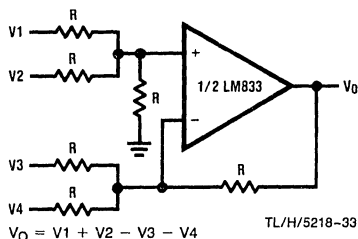
TL/H/5218-31

Balanced to Single Ended Converter



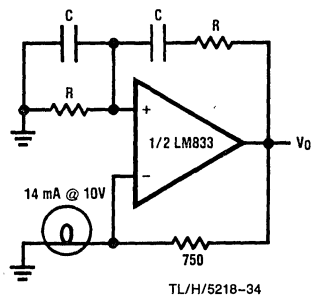
TL/H/5218-32

Adder/Subtractor



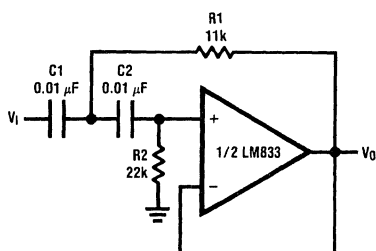
TL/H/5218-33

Sine Wave Oscillator



TL/H/5218-34

Second Order High Pass Filter (Butterworth)



TL/H/5218-35

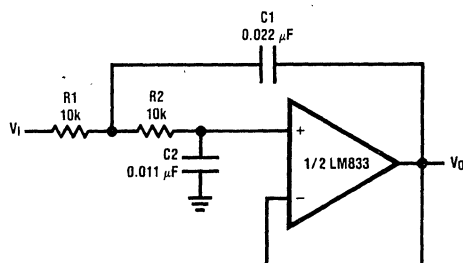
if $C_1 = C_2 = C$

$$R_1 = \frac{\sqrt{2}}{2\omega_0 C}$$

$$R_2 = 2 \cdot R_1$$

Illustration is $f_0 = 1 \text{ kHz}$

Second Order Low Pass Filter (Butterworth)



TL/H/5218-36

if $R_1 = R_2 = R$

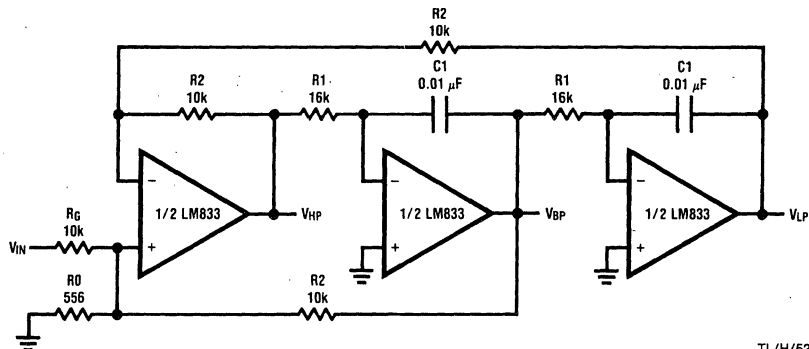
$$C_1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C_2 = \frac{C_1}{2}$$

Illustration is $f_0 = 1 \text{ kHz}$

Typical Applications (Continued)

State Variable Filter

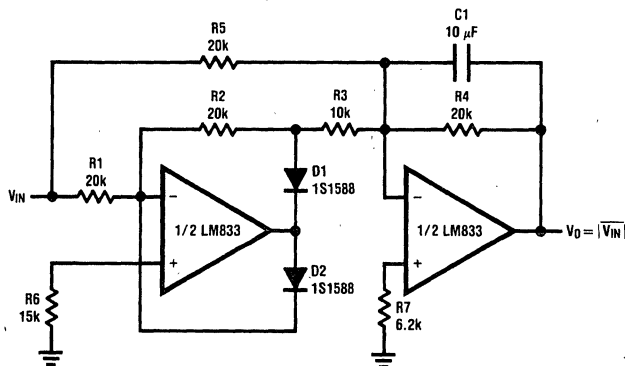


TL/H/5218-37

$$f_0 = \frac{1}{2\pi C_1 R_1}, Q = \frac{1}{2} \left(1 + \frac{R_2}{R_Q} + \frac{R_2}{R_Q} \right), A_{BP} = Q A_{LP} = Q A_{LH} = \frac{R_2}{R_Q}$$

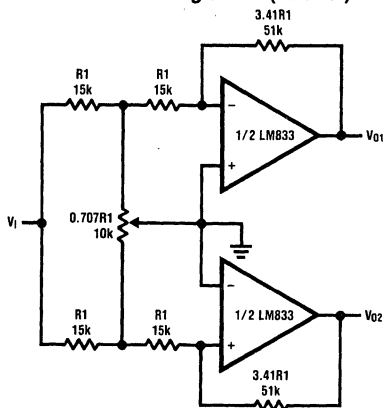
Illustration is $f_0 = 1$ kHz, $Q = 10$, $A_{BP} = 1$

AC/DC Converter



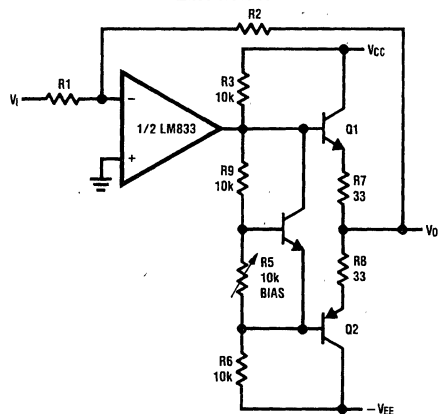
TL/H/5218-38

2 Channel Panning Circuit (Pan Pot)



TL/H/5218-39

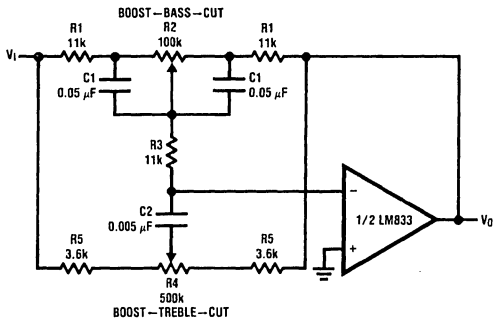
Line Driver



TL/H/5218-40

Typical Application (Continued)

Tone Control



TL/H/5218-41

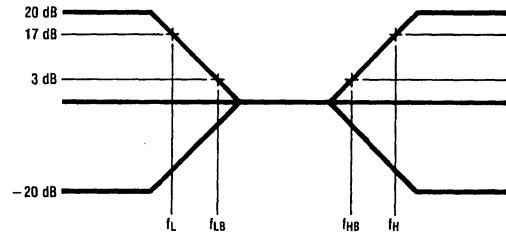
$$f_L = \frac{1}{2\pi R_2 C_1}, f_{LB} = \frac{1}{2\pi R_1 C_1}$$

$$f_H = \frac{1}{2\pi R_5 C_2}, f_{HB} = \frac{1}{2\pi (R_1 + 2R_3) C_2}$$

Illustration is:

$$f_L = 32 \text{ Hz}, f_{LB} = 320 \text{ Hz}$$

$$f_H = 11 \text{ kHz}, f_{HB} = 1.1 \text{ kHz}$$



TL/H/5218-42

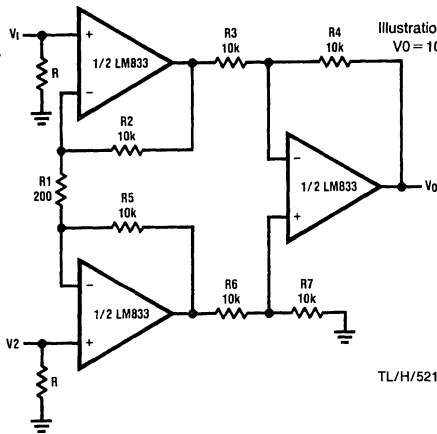
Balanced Input Mic Amp

$$\text{If } R_2 = R_5, R_3 = R_6, R_4 = R_7$$

$$V_0 = \left(1 + \frac{2R_2}{R_1}\right) \frac{R_4}{R_3} (V_2 - V_1)$$

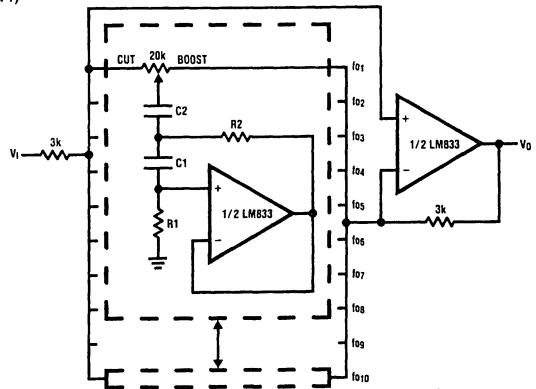
Illustration is:

$$V_0 = 101(V_2 - V_1)$$



TL/H/5218-43

10 Band Graphic Equalizer



TL/H/5218-44

fo(Hz)	C ₁	C ₂	R ₁	R ₂
32	0.12μF	4.7μF	75kΩ	500Ω
64	0.056μF	3.3μF	68kΩ	510Ω
125	0.033μF	1.5μF	62kΩ	510Ω
250	0.015μF	0.82μF	68kΩ	470Ω
500	8200pF	0.39μF	62kΩ	470Ω
1k	3900pF	0.22μF	68kΩ	470Ω
2k	2000pF	0.1μF	68kΩ	470Ω
4k	1100pF	0.056μF	62kΩ	470Ω
8k	510pF	0.022μF	68kΩ	510Ω
16k	330pF	0.012μF	51kΩ	510Ω

At volume of change = ±12dB

$$Q = 1.7$$

Reference: "AUDIO/RADIO HANDBOOK", National Semiconductor, 1980, Page 2-61