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## STM32L432KB STM3L432KC

# Ultra-low-power ARM<sup>®</sup> Cortex<sup>®</sup>-M4 32-bit MCU+FPU, 100DMIPS, up to 256KB Flash, 64KB SRAM, USB FS, analog, audio

Datasheet - production data

#### **Features**

- Ultra-low-power with FlexPowerControl
  - 1.71 V to 3.6 V power supply
  - -40 °C to 85/105/125 °C temperature range
  - 8 nA Shutdown mode (2 wakeup pins)
  - 28 nA Standby mode (2 wakeup pins)
  - 280 nA Standby mode with RTC
  - 1.0 μA Stop 2 mode, 1.28 μA Stop 2 with RTC
  - 84 μA/MHz run mode
  - Batch acquisition mode (BAM)
  - 4 µs wakeup from Stop mode
  - Brown out reset (BOR) in all modes except shutdown
  - Interconnect matrix
- Core: ARM<sup>®</sup> 32-bit Cortex<sup>®</sup>-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator<sup>™</sup>) allowing 0-wait-state execution from Flash memory, frequency up to 80 MHz, MPU, 100DMIPS/1.25DMIPS/MHz (Dhrystone 2.1), and DSP instructions
- Performance Benchmark
  - 1.25 DMIPS/MHz (Drystone 2.1)
  - 273.55 Coremark<sup>®</sup> (3.42 Coremark/MHz @ 80 MHz)
- · Energy Benchmark
  - 176.7 ULPBench<sup>®</sup> score
- Clock Sources
  - 32 kHz crystal oscillator for RTC (LSE)
  - Internal 16 MHz factory-trimmed RC (±1%)
  - Internal low-power 32 kHz RC (±5%)
  - Internal multispeed 100 kHz to 48 MHz oscillator, auto-trimmed by LSE (better than ±0.25 % accuracy)
  - Internal 48 MHz with clock recovery
  - 2 PLLs for system clock, USB, audio, ADC
- RTC with HW calendar, alarms and calibration
- · Up to 3 capacitive sensing channels



UFQFPN32 (5x5)

- 11x timers: 1x 16-bit advanced motor-control, 1x 32-bit and 2x 16-bit general purpose, 2x 16bit basic, 2x low-power 16-bit timers (available in Stop mode), 2x watchdogs, SysTick timer
- Up to 26 fast I/Os, most 5 V-tolerant
- Memories
  - Up to 256 KB single bank Flash, proprietary code readout protection
  - 64 KB of SRAM including 16 KB with hardware parity check
  - Quad SPI memory interface
- Rich analog peripherals (independent supply)
  - 1× 12-bit ADC 5 Msps, up to 16-bit with hardware oversampling, 200 μA/Msps
  - 2x 12-bit DAC, low-power sample and hold
  - 1x operational amplifier with built-in PGA
  - 2x ultra-low-power comparators
- 13x communication interfaces
  - USB 2.0 full-speed crystal less solution with LPM and BCD
  - 1x SAI (serial audio interface)
  - 2x I2C FM+(1 Mbit/s), SMBus/PMBus
  - 3x USARTs (ISO 7816, LIN, IrDA, modem)
  - 2x SPIs (3x SPIs with the Quad SPI)
  - CAN (2.0B Active)
  - SWPMI single wire protocol master I/F
  - IRTIM (Infrared interface)
- 14-channel DMA controller
- True random number generator
- CRC calculation unit, 96-bit unique ID
- Development support: serial wire debug (SWD), JTAG, Embedded Trace Macrocell™

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## 1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32L432xx microcontrollers.

This document should be read in conjunction with the STM32L4x2 reference manual (RM0393). The reference manual is available from the STMicroelectronics website <a href="https://www.st.com">www.st.com</a>.

For information on the  $\mathsf{ARM}^{\$}$   $\mathsf{Cortex}^{\$}$ -M4 core, please refer to the  $\mathsf{Cortex}^{\$}$ -M4 Technical Reference Manual, available from the www.arm.com website.





## 2 Description

The STM32L432xx devices are the ultra-low-power microcontrollers based on the high-performance ARM® Cortex®-M4 32-bit RISC core operating at a frequency of up to 80 MHz. The Cortex-M4 core features a Floating point unit (FPU) single precision which supports all ARM single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances application security.

The STM32L432xx devices embed high-speed memories (Flash memory up to 256 Kbyte, 64 Kbyte of SRAM), a Quad SPI flash memories interface and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses and a 32-bit multi-AHB bus matrix.

The STM32L432xx devices embed several protection mechanisms for embedded Flash memory and SRAM: readout protection, write protection, proprietary code readout protection and Firewall.

The devices offer a fast 12-bit ADC (5 Msps), two comparators, one operational amplifier, two DAC channels, a low-power RTC, one general-purpose 32-bit timer, one 16-bit PWM timer dedicated to motor control, four general-purpose 16-bit timers, and two 16-bit low-power timers.

In addition, up to 3 capacitive sensing channels are available.

They also feature standard and advanced communication interfaces.

- Two I2Cs
- Two SPIs
- Two USARTs and one Low-Power UART.
- One SAI (Serial Audio Interfaces)
- One CAN
- One USB full-speed device crystal less
- One SWPMI (Single Wire Protocol Master Interface)

The STM32L432xx operates in the -40 to +85  $^{\circ}$ C (+105  $^{\circ}$ C junction), -40 to +105  $^{\circ}$ C (+125  $^{\circ}$ C junction) and -40 to +125  $^{\circ}$ C (+130  $^{\circ}$ C junction) temperature ranges from a 1.71 to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

Some independent power supplies are supported: analog independent supply input for ADC, DAC, OPAMPs and comparators.

The STM32L432xx family offers a single 32-pin package.

Table 1. STM32L432Kx family device features and peripheral counts

Peripheral	STM32L432Kx					
Flash memory	256KB					
SRAM	64KB					
Quad SPI	Yes					



Table 1. STM32L432Kx family device features and peripheral counts (continued)

	Peripheral	STM32L432Kx					
Į.	Advanced control	1 (16-bit)					
	General purpose	2 (16-bit) 1 (32-bit)					
Timers	Basic	2 (16-bit)					
	_ow -power	2 (16-bit)					
5	SysTick timer	1					
	Watchdog timers (independent, window)	2					
5	SPI	2					
Ī	<sup>2</sup> C	2					
L	JSART _PUART	2 1					
Comm. interfaces	SAI	1					
	CAN	1					
	JSB FS	Yes <sup>(1)</sup>					
5	SWPMI	Yes					
RTC		Yes					
Tamper pins		1					
Random generator		Yes					
GPIOs Wakeup pins		26 2					
Capacitive sensing Number of channels		3					
12-bit ADCs Number of channels		1 10					
12-bit DAC channels	3	2					
Analog comparator		2					
Operational amplifier	rs	1					
Max. CPU frequency	У	80 MHz					
Operating voltage		1.71 to 3.6 V					
Operating temperatu	ıre	Ambient operating temperature: -40 to 85 °C / -40 to 105 °C / -40 to 125 °C  Junction temperature: -40 to 105 °C / -40 to 125 °C / -40 to 130 °C					
Packages		UFQFPN32					

<sup>1.</sup> There is no VDDUSB pin.  $V_{DDUSB}$  is connected internally at  $V_{DD}$ . To be functional,  $V_{DD}$  must be equal to 3.3 V (+/- 10%).

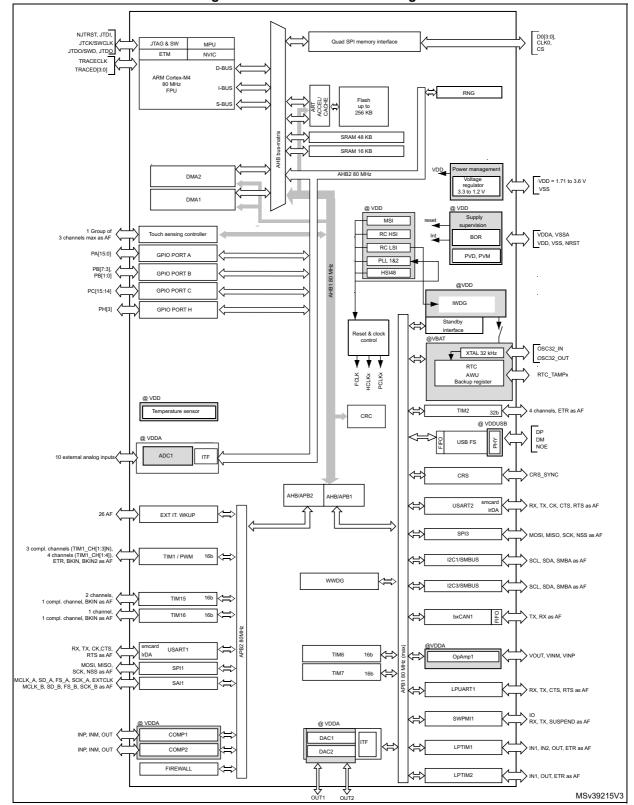


Figure 1. STM32L432xx block diagram

Note: AF: alternate function on I/O pins.

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## 3 Functional overview

## 3.1 ARM® Cortex®-M4 core with FPU

The ARM® Cortex®-M4 with FPU processor is the latest generation of ARM processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The ARM® Cortex®-M4 with FPU 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an ARM core in the memory size usually associated with 8- and 16-bit devices.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

Its single precision FPU speeds up software development by using metalanguage development tools, while avoiding saturation.

With its embedded ARM core, the STM32L432xx family is compatible with all ARM tools and software.

Figure 1 shows the general block diagram of the STM32L432xx family devices.

## 3.2 Adaptive real-time memory accelerator (ART Accelerator™)

The ART Accelerator™ is a memory accelerator which is optimized for STM32 industry-standard ARM® Cortex®-M4 processors. It balances the inherent performance advantage of the ARM® Cortex®-M4 over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor near 100 DMIPS performance at 80MHz, the accelerator implements an instruction prefetch queue and branch cache, which increases program execution speed from the 64-bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 80 MHz.

## 3.3 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.



## 3.4 Embedded Flash memory

STM32L432xx devices feature up to 256 Kbyte of embedded Flash memory available for storing programs and data in single bank architecture. The Flash memory contains 128 pages of 2 Kbyte.

Flexible protections can be configured thanks to option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels are available:
  - Level 0: no readout protection
  - Level 1: memory readout protection: the Flash memory cannot be read from or written to if either debug features are connected, boot in RAM or bootloader is selected
  - Level 2: chip readout protection: debug features (Cortex-M4 JTAG and serial wire), boot in RAM and bootloader selection are disabled (JTAG fuse). This selection is irreversible.

Area	Protection level	U	ser execution	on	Debug, boot from RAM or boot from system memory (loader)				
	ievei	Read	Write	Erase	Read	Write	Erase		
Main	1	Yes	Yes	Yes	No	No	No		
memory	2	Yes	Yes	Yes	N/A	N/A	N/A		
System memory	1	Yes	No	No	Yes	No	No		
	2	Yes	No	No	N/A	N/A	N/A		
Option	1	Yes	Yes	Yes	Yes	Yes	Yes		
bytes	2	Yes	No	No	N/A	N/A	N/A		
Backup	1	Yes	Yes	N/A <sup>(1)</sup>	No	No	N/A <sup>(1)</sup>		
registers	2	Yes	Yes	N/A	N/A	N/A	N/A		
SRAM2	1	Yes	Yes	Yes <sup>(1)</sup>	No	No	No <sup>(1)</sup>		
	2	Yes	Yes	Yes	N/A	N/A	N/A		

- 1. Erased when RDP change from Level 1 to Level 0.
- Write protection (WRP): the protected area is protected against erasing and programming. Two areas can be selected, with 2-Kbyte granularity.
- Proprietary code readout protection (PCROP): a part of the flash memory can be
  protected against read and write from third parties. The protected area is execute-only:
  it can only be reached by the STM32 CPU, as an instruction code, while all other
  accesses (DMA, debug and CPU data read, write and erase) are strictly prohibited.
  The PCROP area granularity is 64-bit wide. An additional option bit (PCROP\_RDP)
  allows to select if the PCROP area is erased or not when the RDP protection is
  changed from Level 1 to Level 0.



The whole non-volatile memory embeds the error correction code (ECC) feature supporting:

- single error detection and correction
- double error detection.
- The address of the ECC fail can be read in the ECC register

#### 3.5 Embedded SRAM

STM32L432xx devices feature 64 Kbyte of embedded SRAM. This SRAM is split into two blocks:

- 48 Kbyte mapped at address 0x2000 0000 (SRAM1)
- 16 Kbyte located at address 0x1000 0000 with hardware parity check (SRAM2).

This memory is also mapped at address 0x2000 C000, offering a contiguous address space with the SRAM1 (16 Kbyte aliased by bit band)

This block is accessed through the ICode/DCode buses for maximum performance.

These 16 Kbyte SRAM can also be retained in Standby mode.

The SRAM2 can be write-protected with 1 Kbyte granularity.

The memory can be accessed in read/write at CPU clock speed with 0 wait states.

#### 3.6 Firewall

The device embeds a Firewall which protects code sensitive and secure data from any access performed by a code executed outside of the protected areas.

Each illegal access generates a reset which kills immediately the detected intrusion.

The Firewall main features are the following:

- Three segments can be protected and defined thanks to the Firewall registers:
  - Code segment (located in Flash or SRAM1 if defined as executable protected area)
  - Non-volatile data segment (located in Flash)
  - Volatile data segment (located in SRAM1)
- The start address and the length of each segments are configurable:
  - code segment: up to 1024 Kbyte with granularity of 256 bytes
  - Non-volatile data segment: up to 1024 Kbyte with granularity of 256 bytes
  - Volatile data segment: up to 48 Kbyte with a granularity of 64 bytes
- Specific mechanism implemented to open the Firewall to get access to the protected areas (call gate entry sequence)
- Volatile data segment can be shared or not with the non-protected code
- Volatile data segment can be executed or not depending on the Firewall configuration

The Flash readout protection must be set to level 2 in order to reach the expected level of protection.

#### 3.7 Boot modes

At startup, BOOT0 pin or nSWBOOT0 option bit, and BOOT1 option bit are used to select one of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

BOOT0 value may come from the PH3-BOOT0 pin or from an option bit depending on the value of a user option bit to free the GPIO pad if needed.

A Flash empty check mechanism is implemented to force the boot from system flash if the first flash memory location is not programmed and if the boot selection is configured to boot from main flash.

The boot loader is located in system memory. It is used to reprogram the Flash memory by using USART, I2C, SPI and USB FS in Device mode through DFU (device firmware upgrade).

## 3.8 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

## 3.9 Power supply management

#### 3.9.1 Power supply schemes

- V<sub>DD</sub> = 1.71 to 3.6 V: external power supply for I/Os (V<sub>DDIO1</sub>), the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through V<sub>DD</sub> pins.
- V<sub>DDA</sub> = 1.62 V (ADCs/COMPs) / 1.8 (DACs/OPAMP) to 3.6 V: external analog power supply for ADCs, DACs, OPAMP, Comparators and Voltage reference buffer. The V<sub>DDA</sub> voltage level is independent from the V<sub>DD</sub> voltage.

Note: When the functions supplied by  $V_{DDA}$  or  $V_{DDUSB}$  are not used, these supplies should preferably be shorted to  $V_{DD}$ .

Note: If these supplies are tied to ground, the I/Os supplied by these power supplies are not 5 V tolerant (refer to Table 17: Voltage characteristics).

Note:  $V_{DDIOx}$  is the I/Os general purpose digital functions supply.  $V_{DDIOx}$  represents  $V_{DDIO1}$ , with  $V_{DDIO1} = V_{DD}$ .

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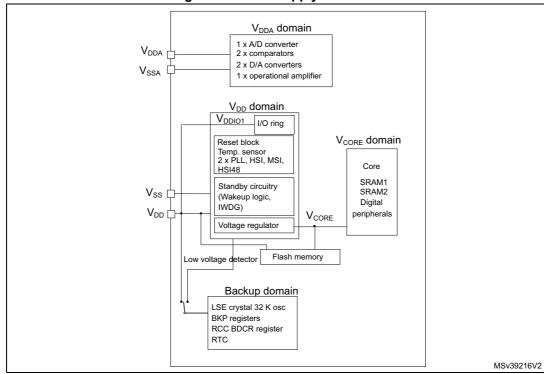


Figure 2. Power supply overview

#### 3.9.2 Power supply supervisor

The device has an integrated ultra-low-power brown-out reset (BOR) active in all modes except Shutdown and ensuring proper operation after power-on and during power down. The device remains in reset mode when the monitored supply voltage V<sub>DD</sub> is below a specified threshold, without the need for an external reset circuit.

The lowest BOR level is 1.71V at power on, and other higher thresholds can be selected through option bytes. The device features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}$  power supply and compares it to the VPVD threshold. An interrupt can be generated when  $V_{DD}$  drops below the VPVD threshold and/or when  $V_{DD}$  is higher than the VPVD threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

In addition, the devices embed a Peripheral Voltage Monitor which compares the independent supply voltage  $V_{DDA}$  with a fixed threshold in order to ensure that the peripheral is in its functional supply range.

#### 3.9.3 Voltage regulator

Two embedded linear voltage regulators supply most of the digital circuitries: the main regulator (MR) and the low-power regulator (LPR).

- The MR is used in the Run and Sleep modes and in the Stop 0 mode.
- The LPR is used in Low-Power Run, Low-Power Sleep, Stop 1 and Stop 2 modes. It is also used to supply the 16 Kbyte SRAM2 in Standby with RAM2 retention.
- Both regulators are in power-down in Standby and Shutdown modes: the regulator output is in high impedance, and the kernel circuitry is powered down thus inducing zero consumption.

The ultralow-power STM32L432xx supports dynamic voltage scaling to optimize its power consumption in run mode. The voltage from the Main Regulator that supplies the logic (VCORE) can be adjusted according to the system's maximum operating frequency.

There are two power consumption ranges:

- Range 1 with the CPU running at up to 80 MHz.
- Range 2 with a maximum CPU frequency of 26 MHz. All peripheral clocks are also limited to 26 MHz.

The VCORE can be supplied by the low-power regulator, the main regulator being switched off. The system is then in Low-power run mode.

 Low-power run mode with the CPU running at up to 2 MHz. Peripherals with independent clock can be clocked by HSI16.

#### 3.9.4 Low-power modes

The ultra-low-power STM32L432xx supports seven low-power modes to achieve the best compromise between low-power consumption, short startup time, available peripherals and available wakeup sources:

By default, the microcontroller is in Run mode after a system or a power Reset. It is up to the user to select one of the low-power modes described below:

#### • Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

#### Low-power run mode

This mode is achieved with VCORE supplied by the low-power regulator to minimize the regulator's operating current. The code can be executed from SRAM or from Flash,

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and the CPU frequency is limited to 2 MHz. The peripherals with independent clock can be clocked by HSI16.

#### • Low-power sleep mode

This mode is entered from the low-power run mode. Only the CPU clock is stopped. When wakeup is triggered by an event or an interrupt, the system reverts to the low-power run mode.

#### • Stop 0, Stop 1 and Stop 2 modes

Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the VCORE domain are stopped, the PLL, the MSI RC and the HSI16 RC are disabled. The LSE or LSI is still running.

The RTC can remain active (Stop mode with RTC, Stop mode without RTC).

Some peripherals with wakeup capability can enable the HSI16 RC during Stop mode to detect their wakeup condition.

Three Stop modes are available: Stop 0, Stop 1 and Stop 2 modes. In Stop 2 mode, most of the VCORE domain is put in a lower leakage mode.

Stop 1 offers the largest number of active peripherals and wakeup sources, a smaller wakeup time but a higher consumption than Stop 2. In Stop 0 mode, the main regulator remains ON, allowing a very fast wakeup time but with much higher consumption.

The system clock when exiting from Stop 0, Stop1 or Stop2 modes can be either MSI up to 48 MHz or HSI16, depending on software configuration.

#### Standby mode

The Standby mode is used to achieve the lowest power consumption with BOR. The internal regulator is switched off so that the VCORE domain is powered off. The PLL, the MSI RC and the HSI16 RC are also switched off.

The RTC can remain active (Standby mode with RTC, Standby mode without RTC).

The brown-out reset (BOR) always remains active in Standby mode.

The state of each I/O during standby mode can be selected by software: I/O with internal pull-up, internal pull-down or floating.

After entering Standby mode, SRAM1 and register contents are lost except for registers in the Backup domain and Standby circuitry. Optionally, SRAM2 can be retained in



Standby mode, supplied by the low-power Regulator (Standby with RAM2 retention mode).

The device exits Standby mode when an external reset (NRST pin), an IWDG reset, WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper) or a failure is detected on LSE (CSS on LSE). The system clock after wakeup is MSI up to 8 MHz.

#### • Shutdown mode

The Shutdown mode allows to achieve the lowest power consumption. The internal regulator is switched off so that the VCORE domain is powered off. The PLL, the HSI16, the MSI and the LSI oscillators are also switched off.

The RTC can remain active (Shutdown mode with RTC, Shutdown mode without RTC).

The BOR is not available in Shutdown mode. No power voltage monitoring is possible in this mode, therefore the switch to Backup domain is not supported.

SRAM1, SRAM2 and register contents are lost except for registers in the Backup domain.

The device exits Shutdown mode when an external reset (NRST pin), a WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper).

The system clock after wakeup is MSI at 4 MHz.



Table 3. Functionalities depending on the working mode<sup>(1)</sup>

					Stop	0/1	Sto	p 2	Stan	dby	Shut	down
Peripheral	Run	Sleep	Low- power run	Low- power sleep	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability
CPU	Υ	-	Υ	-	-	-	-	-	-	-	-	-
Flash memory (up to 256 KB)	O <sup>(2)</sup>	O <sup>(2)</sup>	O <sup>(2)</sup>	O <sup>(2)</sup>	-	-	-	-	-	-	-	-
SRAM1 (48 KB)	Υ	Y <sup>(3)</sup>	Υ	Y <sup>(3)</sup>	Υ	-	Υ	-	-	-	-	-
SRAM2 (16 KB)	Υ	Y <sup>(3)</sup>	Υ	Y <sup>(3)</sup>	Υ	-	Υ	-	O <sup>(4)</sup>	-	-	-
Quad SPI	0	0	0	0	ı	-	-	-	-	-	-	-
Backup Registers	Υ	Υ	Υ	Υ	Υ	-	Υ	-	Υ	-	Υ	-
Brown-out reset (BOR)	Υ	Y	Υ	Y	Y	Υ	Y	Υ	Y	Y	-	-
Programmable Voltage Detector (PVD)	0	0	0	0	0	0	0	0	-	-	-	-
Peripheral Voltage Monitor (PVMx; x=1,3,4)	0	0	0	0	0	0	0	0	-	-	-	-
DMA	0	0	0	0	-	-	-	-	-	-	-	-
High Speed Internal (HSI16)	0	0	0	0	(5)	-	(5)	-	-	-	-	-
Oscillator RC48	0	0	-	-	-	-	-	-	-	-	-	-
High Speed External (HSE)	0	0	0	0	-	-	-	-	-	-	-	-
Low Speed Internal (LSI)	0	0	0	0	0	-	0	-	0	-	-	-
Low Speed External (LSE)	0	0	0	0	0	-	0	-	0	-	0	-
Multi-Speed Internal (MSI)	0	0	0	0	-	-	-	-	-	-	-	-
Clock Security System (CSS)	0	0	0	0	-	-	-	-	-	-	-	-
Clock Security System on LSE	0	0	0	0	0	0	0	0	0	0	-	-
RTC / Auto wakeup	0	0	0	0	0	0	0	0	0	0	0	0
Number of RTC Tamper pins	1	1	1	1	1	0	1	0	1	0	1	0
USB FS	O <sup>(8)</sup>	O <sup>(8)</sup>	-	-	i	0	-	-	-	-	-	-



Table 3. Functionalities depending on the working mode<sup>(1)</sup> (continued)

				ing on t	Stop			p 2		ndby		down
Peripheral	Run	Sleep	Low- power run	Low- power sleep	-	Wakeup capability		Wakeup capability		Wakeup capability	-	Wakeup capability
USARTx (x=1,2)	0	0	0	0	O <sup>(6)</sup>	O <sup>(6)</sup>	ı	-	ı	-	-	-
Low-power UART (LPUART)	0	0	0	0	O <sup>(6)</sup>	O <sup>(6)</sup>	O <sup>(6)</sup>	O <sup>(6)</sup>	-	-	-	
I2Cx (x=1)	0	0	0	0	O <sup>(7)</sup>	O <sup>(7)</sup>	-		-	-	-	
I2C3	0	0	0	0	O <sup>(7)</sup>	O <sup>(7)</sup>	O <sup>(7)</sup>	O <sup>(7)</sup>	-	-	-	1
SPIx (x=1,3)	0	0	0	0	ı	-	ı	-	ı	-	-	-
CAN	0	0	0	0	ı	-	ı	-	ı	-	-	-
SWPMI1	0	0	0	0	-	0	-		-	-	-	1
SAIx (x=1)	0	0	0	0	-	-	-		-	-	-	
ADCx (x=1)	0	0	0	0	-	-	-		-	-	-	
DACx (x=1,2)	0	0	0	0	0	-	-		-	-	-	-
OPAMPx (x=1)	0	0	0	0	0	-	-	-	-	-	-	-
COMPx (x=1,2)	0	0	0	0	0	0	0	0	ı	-	-	-
Temperature sensor	0	0	0	0	ı	-	ı		ı	-	ı	-
Timers (TIMx)	0	0	0	0	ı	-	ı	-	ı	-	-	-
Low-power timer 1 (LPTIM1)	0	0	0	0	0	0	0	0	-	-	-	
Low-power timer 2 (LPTIM2)	0	0	0	0	0	0	-		-	-	-	,
Independent watchdog (IWDG)	0	0	0	0	0	0	0	0	0	0	-	
Window watchdog (WWDG)	0	0	0	0	-	-	-	-	-	-	-	-
SysTick timer	0	0	0	0	ı	-	1		-	-	-	-
Touch sensing controller (TSC)	0	0	0	0	-	-	-	-	-	-	-	-
Random number generator (RNG)	O <sup>(8)</sup>	O <sup>(8)</sup>	-	-	-	-	-	-	-	-	-	-
CRC calculation unit	0	0	0	0	1	-	1	1	I	-	1	-
GPIOs	0	0	0	0	0	0	0	0	(9)	2 pins (10)	(11)	2 pins (10)



- Legend: Y = Yes (Enable). O = Optional (Disable by default. Can be enabled by software). = Not available.
- 2. The Flash can be configured in power-down mode. By default, it is not in power-down mode.
- 3. The SRAM clock can be gated on or off.
- 4. SRAM2 content is preserved when the bit RRS is set in PWR\_CR3 register.
- 5. Some peripherals with wakeup from Stop capability can request HSI16 to be enabled. In this case, HSI16 is woken up by the peripheral, and only feeds the peripheral which requested it. HSI16 is automatically put off when the peripheral does not need it anymore.
- 6. UART and LPUART reception is functional in Stop mode, and generates a wakeup interrupt on Start, address match or received frame event.
- 7. I2C address detection is functional in Stop mode, and generates a wakeup interrupt in case of address match
- 8. Voltage scaling Range 1 only.
- 9. I/Os can be configured with internal pull-up, pull-down or floating in Standby mode.
- 10. The I/Os with wakeup from Standby/Shutdown capability are: PA0, PA2.
- 11. I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode but the configuration is lost when exiting the Shutdown mode.

#### 3.9.5 Reset mode

In order to improve the consumption under reset, the I/Os state under and after reset is "analog state" (the I/O schmitt trigger is disable). In addition, the internal reset pull-up is deactivated when the reset source is internal.

#### 3.10 Interconnect matrix

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources thus power supply consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep, low-power run and sleep, Stop 0, Stop 1 and Stop 2 modes.

Table 4. STM32L432xx peripherals interconnect matrix

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
	TIMx	Timers synchronization or chaining	Υ	Υ	Υ	Υ	-	-
TIMx	ADCx DACx	Conversion triggers	Υ	Υ	Υ	Y	1	-
	DMA	Memory to memory transfer trigger	Υ	Υ	Υ	Υ	-	-
	COMPx	Comparator output blanking	Υ	Υ	Υ	Υ	1	-
TIM15/TIM16	IRTIM	Infrared interface output generation	Υ	Υ	Υ	Υ	-	-

Table 4. STM32L432xx peripherals interconnect matrix (continued)

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
COMPx	TIM1 TIM2	Timer input channel, trigger, break from analog signals comparison	Υ	Υ	Υ	Υ	-	-
COIVIFX	LPTIMERx	Low-power timer triggered by analog signals comparison	Υ	Υ	Υ	Υ	Υ	Y (1)
ADCx	TIM1	Timer triggered by analog watchdog	Υ	Υ	Υ	Υ	-	-
	TIM16	Timer input channel from RTC events	Υ	Υ	Υ	Υ	-	-
RTC	LPTIMERx	Low-power timer triggered by RTC alarms or tampers	Υ	Υ	Υ	Υ	Υ	Y (1)
All clocks sources (internal and external)	TIM2 TIM15, 16	Clock source used as input channel for RC measurement and trimming			Y	Υ	-	-
USB	TIM2	Timer triggered by USB SOF	Υ	Υ	-	-	-	-
CSS CPU (hard fault) RAM (parity error) Flash memory (ECC error) COMPx PVD	TIM1 TIM15,16	Timer break	Y	Y	Y	Y	1	-
	TIMx	External trigger	Υ	Υ	Υ	Υ	-	-
GPIO	LPTIMERx	External trigger	Υ	Υ	Υ	Υ	Υ	Y (1)
	ADCx DACx	Conversion external trigger	Υ	Υ	Υ	Υ	-	-

<sup>1.</sup> LPTIM1 only.



## 3.11 Clocks and startup

The clock controller (see *Figure 3*) distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

- Clock prescaler: to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- **Safe clock switching:** clock sources can be changed safely on the fly in run mode through a configuration register.
- **Clock management:** to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- System clock source: four different clock sources can be used to drive the master clock SYSCLK:
  - High Speed External clock (HSE) can supply a PLL.
  - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software, that can supply a PLL
  - Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 12 frequencies from 100 kHz to 48 MHz. When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be automatically trimmed by hardware to reach better than ±0.25% accuracy. In this mode the MSI can feed the USB device. The MSI can supply a PLL.
  - System PLL which can be fed by HSE, HSI16 or MSI, with a maximum frequency at 80 MHz.
- RC48 with clock recovery system (HSI48): internal RC48 MHz clock source can be used to drive the USB or the RNG peripherals. This clock can be output on the MCO.
- Auxiliary clock source: two ultralow-power clock sources that can be used to drive the real-time clock:
  - 32.768 kHz low-speed external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for an external clock.
  - 32 kHz low-speed internal RC (LSI), also used to drive the independent watchdog.
     The LSI clock accuracy is ±5% accuracy.
- Peripheral clock sources: Several peripherals (USB, RNG, SAI, USARTs, I2Cs, LPTimers, ADC, SWPMI) have their own independent clock whatever the system clock. Two PLLs, each having three independent outputs allowing the highest flexibility, can generate independent clocks for the ADC, the USB/RNG and the SAI.
- **Startup clock:** after reset, the microcontroller restarts by default with an internal 4 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- Clock security system (CSS): this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI16 and a software interrupt is generated if enabled. LSE failure can also be detected and generated an interrupt.
- Clock-out capability:
  - MCO: microcontroller clock output: it outputs one of the internal clocks for external use by the application
  - LSCO: low speed clock output: it outputs LSI or LSE in all low-power modes.



Several prescalers allow to configure the AHB frequency, the high speed APB (APB2) and the low speed APB (APB1) domains. The maximum frequency of the AHB and the APB domains is 80 MHz.

to IWDG LSI RC 32 kHz LSCO to RTC OSC32\_OUT LSE OSC /32 OSC32\_IN LSE LSI HSE мсо / 1→16 to PWR SYSCLK HSI to AHB bus, core, memory and DMA Clock HSI48 source HCLK FCLK Cortex free running clock AHB PRESC control / 1,2,..512 CK\_IN to Cortex system timer HSE Clock detector /8 MSI SYSCLK PCLK1 HSI APB1 PRESC / 1,2,4,8,16 to APB1 peripherals x1 or x2 to TIMx 16 MHz x=2,6,7 LSE HSI SYSCLK to USARTx X=2..3 to LPUART1 HSI-SYSCLK-MSI RC to I2Cx 100 kHz – 48 MHz x=1,2,3 to LPTIMx HSIto SWPMI MSI PCLK2 HSI / M PLL APR2 PRESC HSE to APB2 peripherals PLLSAI1CLK / 1,2,4,8,16 vco F<sub>vcc</sub> / P PLLUSB1CLK /Q x1 or x2 to TIMx PLLCLK x=1,15,16 to USART1 PLLSAI1 PLLSAI2CLK VCO F<sub>VCO</sub> / P PLLUSB2CLK /Q 48 MHz clock to USB, RNG PLLADC1CLK SYSCLK to ADC HSI HSI RC 48 MHz to SAI1 HSI CRS SAI1 EXTCLK MSv39217V3

Figure 3. Clock tree

## 3.12 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. Fast I/O toggling can be achieved thanks to their mapping on the AHB2 bus.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

## 3.13 Direct memory access controller (DMA)

The device embeds 2 DMAs. Refer to *Table 5: DMA implementation* for the features implementation.

Direct memory access (DMA) is used in order to provide high-speed data transfer between peripherals and memory as well as memory to memory. Data can be quickly moved by DMA without any CPU actions. This keeps CPU resources free for other operations.

The two DMA controllers have 14 channels in total, each dedicated to managing memory access requests from one or more peripherals. Each has an arbiter for handling the priority between DMA requests.

#### The DMA supports:

- 14 independently configurable channels (requests)
- Each channel is connected to dedicated hardware DMA requests, software trigger is also supported on each channel. This configuration is done by software.
- Priorities between requests from channels of one DMA are software programmable (4 levels consisting of very high, high, medium, low) or hardware in case of equality (request 1 has priority over request 2, etc.)
- Independent source and destination transfer size (byte, half word, word), emulating packing and unpacking. Source/destination addresses must be aligned on the data size.
- Support for circular buffer management
- 3 event flags (DMA Half Transfer, DMA Transfer complete and DMA Transfer Error) logically ORed together in a single interrupt request for each channel
- Memory-to-memory transfer
- Peripheral-to-memory and memory-to-peripheral, and peripheral-to-peripheral transfers
- Access to Flash, SRAM, APB and AHB peripherals as source and destination
- Programmable number of data to be transferred: up to 65536.

**Table 5. DMA implementation** 

DMA features	DMA1	DMA2
Number of regular channels	7	7



## 3.14 Interrupts and events

#### 3.14.1 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 61 maskable interrupt channels plus the 16 interrupt lines of the Cortex<sup>®</sup>-M4.

The NVIC benefits are the following:

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- · Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

#### 3.14.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 34 edge detector lines used to generate interrupt/event requests and wake-up the system from Stop mode. Each external line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently A pending register maintains the status of the interrupt requests. The internal lines are connected to peripherals with wakeup from Stop mode capability. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 26 GPIOs can be connected to the 16 external interrupt lines.

## 3.15 Analog to digital converter (ADC)

The device embeds a successive approximation analog-to-digital converter with the following features:

- 12-bit native resolution, with built-in calibration
- 5.33 Msps maximum conversion rate with full resolution
  - Down to 18.75 ns sampling time
  - Increased conversion rate for lower resolution (up to 8.88 Msps for 6-bit resolution)
- Up to 10 external channels.
- 4 internal channels: internal reference voltage, temperature sensor, DAC1 and DAC2 outputs.
- Single-ended and differential mode inputs
- Low-power design
  - Capable of low-current operation at low conversion rate (consumption decreases linearly with speed)
  - Dual clock domain architecture: ADC speed independent from CPU frequency
- Highly versatile digital interface
  - Single-shot or continuous/discontinuous sequencer-based scan mode: 2 groups of analog signals conversions can be programmed to differentiate background and high-priority real-time conversions
  - ADC supports multiple trigger inputs for synchronization with on-chip timers and external signals
  - Results stored into data register or in RAM with DMA controller support
  - Data pre-processing: left/right alignment and per channel offset compensation
  - Built-in oversampling unit for enhanced SNR
  - Channel-wise programmable sampling time
  - Three analog watchdog for automatic voltage monitoring, generating interrupts and trigger for selected timers
  - Hardware assistant to prepare the context of the injected channels to allow fast context switching

#### 3.15.1 Temperature sensor

The temperature sensor (TS) generates a voltage V<sub>TS</sub> that varies linearly with temperature.

The temperature sensor is internally connected to the ADC1\_IN17 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.



Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at a temperature of 130 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75CA - 0x1FFF 75CB

Table 6. Temperature sensor calibration values

## 3.15.2 Internal voltage reference (V<sub>REFINT</sub>)

The internal voltage reference (VREFINT) provides a stable (bandgap) voltage output for the ADC and Comparators. VREFINT is internally connected to the ADC1\_IN0 input channel. The precise voltage of VREFINT is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 7. Internal voltage reference calibration values

Calibration value name	Description	Memory address			
VREFINT	Raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75AA - 0x1FFF 75AB			

## 3.16 Digital to analog converter (DAC)

Two 12-bit buffered DAC channels can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This digital interface supports the following features:

- Up to two DAC output channels
- 8-bit or 12-bit output mode
- Buffer offset calibration (factory and user trimming)
- · Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- External triggers for conversion
- Sample and hold low-power mode, with internal or external capacitor

The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

## 3.17 Comparators (COMP)

The STM32L432xx devices embed two rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) and with selectable output polarity.

The reference voltage can be one of the following:

- External I/O
- DAC output channels
- Internal reference voltage or submultiple (1/4, 1/2, 3/4).

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers and can be also combined into a window comparator.

## 3.18 Operational amplifier (OPAMP)

The STM32L432xx embeds one operational amplifier with external or internal follower routing and PGA capability.

The operational amplifier features:

- Low input bias current
- Low offset voltage
- Low-power mode
- Rail-to-rail input

## 3.19 Touch sensing controller (TSC)

The touch sensing controller provides a simple solution for adding capacitive sensing functionality to any application. Capacitive sensing technology is able to detect finger presence near an electrode which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

The main features of the touch sensing controller are the following:

- Proven and robust surface charge transfer acquisition principle
- Supports up to 3 capacitive sensing channels
- Up to 3 capacitive sensing channels can be acquired in parallel offering a very good response time
- Spread spectrum feature to improve system robustness in noisy environments
- Full hardware management of the charge transfer acquisition sequence
- Programmable charge transfer frequency
- Programmable sampling capacitor I/O pin
- Programmable channel I/O pin
- Programmable max count value to avoid long acquisition when a channel is faulty
- Dedicated end of acquisition and max count error flags with interrupt capability
- One sampling capacitor for up to 3 capacitive sensing channels to reduce the system components
- Compatible with proximity, touchkey, linear and rotary touch sensor implementation
- Designed to operate with STMTouch touch sensing firmware library

Note:

The number of capacitive sensing channels is dependent on the size of the packages and subject to I/O availability.

## 3.20 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

## 3.21 Timers and watchdogs

The STM32L432xx includes one advanced control timers, up to five general-purpose timers, two basic timers, two low-power timers, two watchdog timers and a SysTick timer. The table below compares the features of the advanced control, general purpose and basic timers.

Table 8. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary outputs
Advanced control	TIM1	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	3
General- purpose	TIM2	32-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General- purpose	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1



Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary outputs
General- purpose	TIM16	16-bit	Up	Any integer between 1 and 65536	Yes	1	1
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

Table 8. Timer feature comparison (continued)

#### 3.21.1 Advanced-control timer (TIM1)

The advanced-control timer can each be seen as a three-phase PWM multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead-times. They can also be seen as complete general-purpose timers. The 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0-100%)
- One-pulse mode output

In debug mode, the advanced-control timer counter can be frozen and the PWM outputs disabled to turn off any power switches driven by these outputs.

Many features are shared with those of the general-purpose TIMx timers (described in Section 3.21.2) using the same architecture, so the advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

#### 3.21.2 General-purpose timers (TIM2, TIM15, TIM16)

There are up to three synchronizable general-purpose timers embedded in the STM32L432xx (see *Table 8* for differences). Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

#### TIM2

It is a full-featured general-purpose timer:

TIM2 has a 32-bit auto-reload up/downcounter and 32-bit prescaler.

This timer features 4 independent channels for input capture/output compare, PWM or one-pulse mode output. It can work with the other general-purpose timers via the Timer Link feature for synchronization or event chaining.

The counter can be frozen in debug mode.

It has independent DMA request generation and support quadrature encoder.

TIM15 and 16

They are general-purpose timers with mid-range features:

They have 16-bit auto-reload upcounters and 16-bit prescalers.

- TIM15 has 2 channels and 1 complementary channel
- TIM16 has 1 channel and 1 complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output.

The timers can work together via the Timer Link feature for synchronization or event chaining. The timers have independent DMA request generation.

The counters can be frozen in debug mode.

#### 3.21.3 Basic timers (TIM6 and TIM7)

The basic timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit timebases.

#### 3.21.4 Low-power timer (LPTIM1 and LPTIM2)

The devices embed two low-power timers. These timers have an independent clock and are running in Stop mode if they are clocked by LSE, LSI or an external clock. They are able to wakeup the system from Stop mode.

LPTIM1 is active in Stop 0, Stop 1 and Stop 2 modes.

LPTIM2 is active in Stop 0 and Stop 1 mode.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous/ one shot mode
- Selectable software/hardware input trigger
- Selectable clock source
  - Internal clock sources: LSE, LSI, HSI16 or APB clock
  - External clock source over LPTIM input (working even with no internal clock source running, used by pulse counter application).
- Programmable digital glitch filter
- Encoder mode (LPTIM1 only)

#### 3.21.5 Infrared interface (IRTIM)

The STM32L432xx includes one infrared interface (IRTIM). It can be used with an infrared LED to perform remote control functions. It uses TIM15 and TIM16 output channels to generate output signal waveforms on IR OUT pin.

#### 3.21.6 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC (LSI) and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

#### 3.21.7 System window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

#### 3.21.8 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source



## 3.22 Real-time clock (RTC) and backup registers

The RTC is an independent BCD timer/counter. It supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month.
- Two programmable alarms.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy.
- One anti-tamper detection pin with programmable filter.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event.
- 17-bit auto-reload wakeup timer (WUT) for periodic events with programmable resolution and period.

The RTC and the 32 backup registers are supplied through a switch that takes power from the  $V_{DD}$  supply.

The backup registers are 32-bit registers used to store 128 bytes of user application data when VDD power is not present. They are not reset by a system or power reset, or when the device wakes up from Standby or Shutdown mode.

The RTC clock sources can be:

- A 32.768 kHz external crystal (LSE)
- An external resonator or oscillator (LSE)
- The internal low power RC oscillator (LSI, with typical frequency of 32 kHz)
- The high-speed external clock (HSE) divided by 32.

The RTC is functional in all low-power modes when it is clocked by the LSE. When clocked by the LSI, the RTC is functional in all low-power modes except Shutdown mode.

All RTC events (Alarm, WakeUp Timer, Timestamp or Tamper) can generate an interrupt and wakeup the device from the low-power modes.

# 3.23 Inter-integrated circuit interface (I2C)

The device embeds 2 I2C. Refer to *Table 9: I2C implementation* for the features implementation.

The I<sup>2</sup>C bus interface handles communications between the microcontroller and the serial I<sup>2</sup>C bus. It controls all I<sup>2</sup>C bus-specific sequencing, protocol, arbitration and timing.

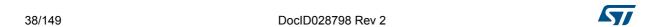
#### The I2C peripheral supports:

- I<sup>2</sup>C-bus specification and user manual rev. 5 compatibility:
  - Slave and master modes, multimaster capability
  - Standard-mode (Sm), with a bitrate up to 100 kbit/s
  - Fast-mode (Fm), with a bitrate up to 400 kbit/s
  - Fast-mode Plus (Fm+), with a bitrate up to 1 Mbit/s and 20 mA output drive I/Os
  - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
  - Programmable setup and hold times
  - Optional clock stretching
- System Management Bus (SMBus) specification rev 2.0 compatibility:
  - Hardware PEC (Packet Error Checking) generation and verification with ACK control
  - Address resolution protocol (ARP) support
  - SMBus alert
- Power System Management Protocol (PMBus<sup>TM</sup>) specification rev 1.1 compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent from the PCLK reprogramming. Refer to Figure 3: Clock tree.
- Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

Table 9. I2C implementation

I2C features <sup>(1)</sup>	I2C1	I2C3
Standard-mode (up to 100 kbit/s)	Х	Х
Fast-mode (up to 400 kbit/s)	Х	Х
Fast-mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	Х	Х
Programmable analog and digital noise filters	Х	Х
SMBus/PMBus hardware support	Х	Х
Independent clock	Х	Х
Wakeup from Stop 0 / Stop 1 mode on address match	Х	Х
Wakeup from Stop 2 mode on address match	-	Х

1. X: supported



#### 3.24 Universal synchronous/asynchronous receiver transmitter (USART)

The STM32L432xx devices have two embedded universal synchronous receiver transmitters (USART1 and USART2).

These interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. They provide hardware management of the CTS and RTS signals, and RS485 Driver Enable. They are able to communicate at speeds of up to 10Mbit/s.

USART1 and USART2 also provide Smart Card mode (ISO 7816 compliant) and SPI-like communication capability.

All USART have a clock domain independent from the CPU clock, allowing the USARTx (x=1,2) to wake up the MCU from Stop mode using baudrates up to 200 Kbaud. The wake up events from Stop mode are programmable and can be:

Table 10. STM32L432xx USART/LPUART features

- Start bit detection
- Any received data frame
- A specific programmed data frame

All USART interfaces can be served by the DMA controller.

....

USART modes/features <sup>(1)</sup>	USART1	USART2	LPUART1
Hardware flow control for modem	Х	Х	Х
Continuous communication using DMA	Х	Х	Х
Multiprocessor communication	Х	Х	Х
Synchronous mode	Х	Х	-
Smartcard mode	Х	Х	-
Single-wire half-duplex communication	Х	Х	Х
IrDA SIR ENDEC block	Х	Х	-
LIN mode	Х	Х	-
Dual clock domain	Х	Х	Х
Wakeup from Stop 0 / Stop 1 modes	Х	Х	Х
Wakeup from Stop 2 mode	-	-	Х
Receiver timeout interrupt	Х	Х	-
Modbus communication	Х	Х	-
Auto baud rate detection	X (4 n	-	
Driver Enable	Х	Х	Х
LPUART/USART data length		7, 8 and 9 bit	s

<sup>1.</sup> X = supported.



# 3.25 Low-power universal asynchronous receiver transmitter (LPUART)

The device embeds one Low-Power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock, and can wakeup the system from Stop mode using baudrates up to 220 Kbaud. The wake up events from Stop mode are programmable and can be:

- Start bit detection
- Any received data frame
- A specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

LPUART interface can be served by the DMA controller.

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# 3.26 Serial peripheral interface (SPI)

Two SPI interfaces allow communication up to 40 Mbits/s in master and up to 24 Mbits/s slave modes, in half-duplex, full-duplex and simplex modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits. The SPI interfaces support NSS pulse mode, TI mode and Hardware CRC calculation.

All SPI interfaces can be served by the DMA controller.

# 3.27 Serial audio interfaces (SAI)

The device embeds 1 SAI. Refer to *Table 11: SAI implementation* for the features implementation. The SAI bus interface handles communications between the microcontroller and the serial audio protocol.

The SAI peripheral supports:

- Two independent audio sub-blocks which can be transmitters or receivers with their respective FIFO.
- 8-word integrated FIFOs for each audio sub-block.
- Synchronous or asynchronous mode between the audio sub-blocks.
- Master or slave configuration independent for both audio sub-blocks.
- Clock generator for each audio block to target independent audio frequency sampling when both audio sub-blocks are configured in master mode.
- Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit.
- Peripheral with large configurability and flexibility allowing to target as example the following audio protocol: I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97 and SPDIF out
- Up to 16 slots available with configurable size and with the possibility to select which
  ones are active in the audio frame.
- Number of bits by frame may be configurable.
- Frame synchronization active level configurable (offset, bit length, level).
- First active bit position in the slot is configurable.
- LSB first or MSB first for data transfer.
- Mute mode.
- Stereo/Mono audio frame capability.
- Communication clock strobing edge configurable (SCK).
- Error flags with associated interrupts if enabled respectively.
  - Overrun and underrun detection.
  - Anticipated frame synchronization signal detection in slave mode.
  - Late frame synchronization signal detection in slave mode.
  - Codec not ready for the AC'97 mode in reception.
- Interruption sources when enabled:
  - Errors.
  - FIFO requests.
- DMA interface with 2 dedicated channels to handle access to the dedicated integrated FIFO of each SAI audio sub-block.



SAI features	Support <sup>(1)</sup>
I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97	X
Mute mode	X
Stereo/Mono audio frame capability.	X
16 slots	X
Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit	X
FIFO Size	X (8 Word)
SPDIF	X

Table 11. SAI implementation

# 3.28 Single wire protocol master interface (SWPMI)

The Single wire protocol master interface (SWPMI) is the master interface corresponding to the Contactless Frontend (CLF) defined in the ETSI TS 102 613 technical specification. The main features are:

- full-duplex communication mode
- automatic SWP bus state management (active, suspend, resume)
- configurable bitrate up to 2 Mbit/s
- automatic SOF, EOF and CRC handling

SWPMI can be served by the DMA controller.

# 3.29 Controller area network (CAN)

The CAN is compliant with specifications 2.0A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. It has three transmit mailboxes, two receive FIFOs with 3 stages and 14 scalable filter banks.

The CAN peripheral supports:

- Supports CAN protocol version 2.0 A, B Active
- Bit rates up to 1 Mbit/s

<sup>1.</sup> X: supported

- Transmission
  - Three transmit mailboxes
  - Configurable transmit priority
- Reception
  - Two receive FIFOs with three stages
  - 14 Scalable filter banks
  - Identifier list feature
  - Configurable FIFO overrun
- Time-triggered communication option
  - Disable automatic retransmission mode
  - 16-bit free running timer
  - Time Stamp sent in last two data bytes
- Management
  - Maskable interrupts
  - Software-efficient mailbox mapping at a unique address space

# 3.30 Universal serial bus (USB)

The STM32L432xx devices embed a full-speed USB device peripheral compliant with the USB specification version 2.0. The internal USB PHY supports USB FS signaling, embedded DP pull-up and also battery charging detection according to Battery Charging Specification Revision 1.2. The USB interface implements a full-speed (12 Mbit/s) function interface with added support for USB 2.0 Link Power Management. It has software-configurable endpoint setting with packet memory up-to 1 KB and suspend/resume support. It requires a precise 48 MHz clock which can be generated from the internal main PLL or by the internal 48 MHz oscillator in automatic trimming mode. The synchronization for this oscillator can be taken from the USB data stream itself (SOF signalization) which allows crystal less operation.

# 3.31 Clock recovery system (CRS)

The STM32L432xx devices embed a special block which allows automatic trimming of the internal 48 MHz oscillator to guarantee its optimal accuracy over the whole device operational range. This automatic trimming is based on the external synchronization signal, which could be either derived from USB SOF signalization, from LSE oscillator, from an external signal on CRS\_SYNC pin or generated by user software. For faster lock-in during startup it is also possible to combine automatic trimming with manual trimming action.

# 3.32 Quad SPI memory interface (QUADSPI)

The Quad SPI is a specialized communication interface targeting single, dual or quad SPI flash memories. It can operate in any of the three following modes:

- Indirect mode: all the operations are performed using the QUADSPI registers
- Status polling mode: the external flash status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external Flash is memory mapped and is seen by the system as if it were an internal memory

Both throughput and capacity can be increased two-fold using dual-flash mode, where two quad SPI flash memories are accessed simultaneously.

The Quad SPI interface supports:

- Three functional modes: indirect, status-polling, and memory-mapped
- SDR and DDR support
- Fully programmable opcode for both indirect and memory mapped mode
- Fully programmable frame format for both indirect and memory mapped mode
- Each of the 5 following phases can be configured independently (enable, length, single/dual/quad communication)
  - Instruction phase
  - Address phase
  - Alternate bytes phase
  - Dummy cycles phase
  - Data phase
- Integrated FIFO for reception and transmission
- 8, 16, and 32-bit data accesses are allowed
- DMA channel for indirect mode operations
- Programmable masking for external flash flag management
- Timeout management
- Interrupt generation on FIFO threshold, timeout, status match, operation complete, and access error

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## 3.33 Development support

## 3.33.1 Serial wire JTAG debug port (SWJ-DP)

The ARM SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

#### 3.33.2 Embedded Trace Macrocell™

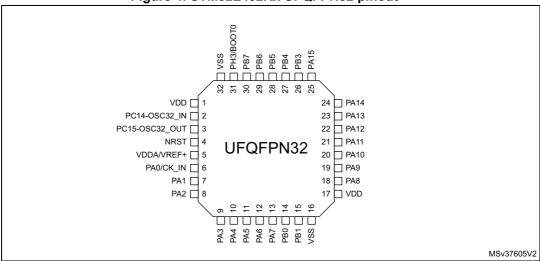
The ARM Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32L432xx through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. Real-time instruction and data flow activity be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.



# 4 Pinouts and pin description

Figure 4. STM32L432Kx UFQFPN32 pinout<sup>(1)</sup>



1. The above figure shows the package top view.

Table 12. Legend/abbreviations used in the pinout table

Na	me	Abbreviation	Definition					
Pin r	name		Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name					
		S	Supply pin					
Pin	type	I	Input only pin					
		I/O	Input / output pin					
		FT	5 V tolerant I/O					
		TT	3.6 V tolerant I/O					
		RST	Bidirectional reset pin with embedded weak pull-up resistor					
I/O str	ructure	Option for TT or FT I/Os						
		_f <sup>(1)</sup> I/O, Fm+ capable						
		_u <sup>(2)</sup>	I/O, with USB function supplied by V <sub>DDUSB</sub>					
		_a <sup>(3)</sup>	I/O, with Analog switch function supplied by V <sub>DDA</sub>					
No	otes	Unless otherwise specified by	y a note, all I/Os are set as analog inputs during and after reset.					
Pin	Alternate functions	Functions selected through 0	GPIOx_AFR registers					
functions	Additional functions	Functions directly selected/e	nabled through peripheral registers					

- 1. The related I/O structures in *Table 13* are: FT\_f, FT\_fa.
- 2. The related I/O structures in Table 13 is: FT\_u.
- 3. The related I/O structures in *Table 13* are: FT\_a, FT\_fa, TT\_a.

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Table 13. STM32L432xx pin definitions

Table 13. STM32L432xx pin definitions							
Pin Number	e ifter				Pin func	tions	
UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions	
2	PC14- OSC32_I N (PC14)	I/O	FT	(1) (2)	EVENTOUT	OSC32_IN	
3	PC15- OSC32_ OUT (PC15)	I/O	FT	(1) (2)	EVENTOUT	OSC32_OUT	
4	NRST	I/O	RST	-	-	-	
5	VDDA/VR EF+	S	-	-	-	-	
6	PA0/ CK_IN	I/O	FT_a	-	TIM2_CH1, USART2_CTS, COMP1_OUT, SAI1_EXTCLK, TIM2_ETR, EVENTOUT	OPAMP1_VINP, COMP1_INM, ADC1_IN5, RTC_TAMP2, WKUP1, CK_IN	
7	PA1	I/O	FT_a	-	TIM2_CH2, I2C1_SMBA, SPI1_SCK, USART2_RTS_DE, TIM15_CH1N, EVENTOUT	OPAMP1_VINM, COMP1_INP, ADC1_IN6	
8	PA2	I/O	FT_a	_	TIM2_CH3, USART2_TX, LPUART1_TX, QUADSPI_BK1_NCS, COMP2_OUT, TIM15_CH1, EVENTOUT	COMP2_INM, ADC1_IN7, WKUP4, LSCO	
9	PA3	I/O	TT_a	_	TIM2_CH4, USART2_RX, LPUART1_RX, QUADSPI_CLK, SAI1_MCLK_A, TIM15_CH2, EVENTOUT	OPAMP1_VOUT, COMP2_INP, ADC1_IN8	
10	PA4	I/O	TT_a	-	SPI1_NSS, SPI3_NSS, USART2_CK, SAI1_FS_B, LPTIM2_OUT, EVENTOUT	COMP1_INM, COMP2_INM, ADC1_IN9, DAC1_OUT1	
11	PA5	I/O	TT_a	-	TIM2_CH1, TIM2_ETR, SPI1_SCK, LPTIM2_ETR, EVENTOUT	COMP1_INM, COMP2_INM, ADC1_IN10, DAC1_OUT2	
12	PA6	I/O	FT_a	-	TIM1_BKIN, SPI1_MISO, COMP1_OUT, USART3_CTS, LPUART1_CTS, QUADSPI_BK1_IO3, TIM1_BKIN_COMP2, TIM16_CH1, EVENTOUT	ADC1_IN11	



Table 13. STM32L432xx pin definitions (continued)

Pin						
Number	ter				Pin func	tions
UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
13	PA7	I/O	FT_fa	ı	TIM1_CH1N, I2C3_SCL, SPI1_MOSI, QUADSPI_BK1_IO2, COMP2_OUT, EVENTOUT	ADC1_IN12
14	PB0	I/O	FT_a	-	TIM1_CH2N, SPI1_NSS,	ADC1_IN15
15	PB1	I/O	FT_a	-	TIM1_CH3N, USART3_RTS_DE, LPUART1_RTS_DE, QUADSPI_BK1_IO0, LPTIM2_IN1, EVENTOUT	COMP1_INM, ADC1_IN16
16	VSS	S	-	-	-	-
17	VDD	S	-	-	-	-
18	PA8	I/O	FT	-	MCO, TIM1_CH1, USART1_CK, SWPMI1_IO, SAI1_SCK_A, LPTIM2_OUT, EVENTOUT	-
19	PA9	I/O	FT_f	-	TIM1_CH2, I2C1_SCL, USART1_TX, SAI1_FS_A, TIM15_BKIN, EVENTOUT	-
20	PA10	I/O	FT_f	-	TIM1_CH3, I2C1_SDA, USART1_RX, USB_CRS_SYNC, SAI1_SD_A, EVENTOUT	-
21	PA11	I/O	FT_u	-	TIM1_CH4, TIM1_BKIN2, SPI1_MISO, COMP1_OUT, USART1_CTS, CAN1_RX, USB_DM, TIM1_BKIN2_COMP1, EVENTOUT	-
22	PA12	I/O	FT_u	-	TIM1_ETR, SPI1_MOSI, USART1_RTS_DE, CAN1_TX, USB_DP, EVENTOUT	-
23	PA13 (JTMS- SWDIO)	I/O	FT	(3)	JTMS-SWDIO, IR_OUT, USB_NOE, SWPMI1_TX, SAI1_SD_B, EVENTOUT	-

Table 13. STM32L432xx pin definitions (continued)

Pin					L432XX pm dennitions (con	<b>-</b>
Number	e ifter				Pin func	tions
UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
24	PA14 (JTCK- SWCLK)	I/O	FT	(3)	JTCK-SWCLK, LPTIM1_OUT, I2C1_SMBA, SWPMI1_RX, SAI1_FS_B, EVENTOUT	-
25	PA15 (JTDI)	I/O	FT	(3)	JTDI, TIM2_CH1, TIM2_ETR, USART2_RX, SPI1_NSS, SPI3_NSS, USART3_RTS_DE, TSC_G3_IO1, SWPMI1_SUSPEND, EVENTOUT	-
26	PB3 (JTDO- TRACE SWO)	I/O	FT_a	(3)	JTDO-TRACESWO, TIM2_CH2, SPI1_SCK, SPI3_SCK, USART1_RTS_DE, SAI1_SCK_B, EVENTOUT	COMP2_INM
27	PB4 (NJTRST)	I/O	FT_fa	(3)	NJTRST, I2C3_SDA, SPI1_MISO, SPI3_MISO, USART1_CTS, TSC_G2_IO1, SAI1_MCLK_B, EVENTOUT	COMP2_INP
28	PB5	I/O	FT	1	LPTIM1_IN1, I2C1_SMBA, SPI1_MOSI, SPI3_MOSI, USART1_CK, TSC_G2_IO2, COMP2_OUT, SAI1_SD_B, TIM16_BKIN, EVENTOUT	-
29	PB6	I/O	FT_fa	ı	LPTIM1_ETR, I2C1_SCL, USART1_TX, TSC_G2_IO3, SAI1_FS_B, TIM16_CH1N, EVENTOUT	COMP2_INP
30	PB7	I/O	FT_fa	-	LPTIM1_IN2, I2C1_SDA, USART1_RX, TSC_G2_IO4, EVENTOUT	COMP2_INM, PVD_IN
31	PH3/ BOOT0	I/O	FT	-	EVENTOUT	воото
32	VSS	S	-	-	-	-
1	VDD	S	-	-	-	-

PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC14 to PC15 in output mode is limited:

 The speed should not exceed 2 MHz with a maximum load of 30 pF
 These GPIOs must not be used as current sources (e.g. to drive an LED).

After reset, these pins are configured as JTAG/SW debug alternate functions, and the internal pull-up on PA15, PA13, PB4 pins and the internal pull-down on PA14 pin are activated.



After a Backup domain power-up, PC14 and PC15 operate as GPIOs. Their function then depends on the
content of the RTC registers which are not reset by the system reset. For details on how to manage these
GPIOs, refer to the Backup domain and RTC register descriptions in the RM0393 reference manual.

Table 14. Alternate function AF0 to AF7 (for AF8 to AF15 see Table 15)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
Po	ort	SYS_AF	TIM1/TIM2/ LPTIM1	TIM1/TIM2	USART2	12C1/I2C2/I2C3	SPI1/SPI2	SPI3	USART1/ USART2/ USART3
	PA0	-	TIM2_CH1	-	-	-	-	-	USART2_CTS
	PA1	-	TIM2_CH2	-	-	I2C1_SMBA	SPI1_SCK	-	USART2_RTS_ DE
	PA2	-	TIM2_CH3	-	-	-	-	-	USART2_TX
	PA3	-	TIM2_CH4	-	-	-	-	-	USART2_RX
	PA4	-	-	-	-	-	SPI1_NSS	SPI3_NSS	USART2_CK
	PA5	-	TIM2_CH1	TIM2_ETR	-	-	SPI1_SCK	-	-
	PA6	-	TIM1_BKIN	-	-	-	SPI1_MISO	COMP1_OUT	USART3_CTS
	PA7	-	TIM1_CH1N	-	-	I2C3_SCL	SPI1_MOSI	-	-
Port A	PA8	MCO	TIM1_CH1	-	-	-	-	-	USART1_CK
	PA9	-	TIM1_CH2	-	-	I2C1_SCL	-	-	USART1_TX
	PA10	-	TIM1_CH3	-	-	I2C1_SDA	-	-	USART1_RX
	PA11	-	TIM1_CH4	TIM1_BKIN2	-	-	SPI1_MISO	COMP1_OUT	USART1_CTS
	PA12	-	TIM1_ETR	-	-	-	SPI1_MOSI	-	USART1_RTS_ DE
	PA13	JTMS-SWDIO	IR_OUT	-	-	-	-	-	-
	PA14	JTCK-SWCLK	LPTIM1_OUT	-	-	I2C1_SMBA	-	-	-
	PA15	JTDI	TIM2_CH1	TIM2_ETR	USART2_RX	-	SPI1_NSS	SPI3_NSS	USART3_RTS_ DE





Table 14. Alternate function AF0 to AF7 (for AF8 to AF15 see Table 15) (continued)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
Port		SYS_AF	TIM1/TIM2/ LPTIM1	TIM1/TIM2	USART2	I2C1/I2C2/I2C3	SPI1/SPI2	SPI3	USART1/ USART2/ USART3
	PB0	-	TIM1_CH2N	-	-	-	SPI1_NSS	-	USART3_CK
	PB1	-	TIM1_CH3N	-	-	-	-	-	USART3_RTS_ DE
Port B	PB3	JTDO- TRACESWO	TIM2_CH2	-	-	-	SPI1_SCK	SPI3_SCK	USART1_RTS_ DE
	PB4	NJTRST	-	-	-	I2C3_SDA	SPI1_MISO	SPI3_MISO	USART1_CTS
	PB5	-	LPTIM1_IN1	-	-	I2C1_SMBA	SPI1_MOSI	SPI3_MOSI	USART1_CK
	PB6	-	LPTIM1_ETR	-	-	I2C1_SCL	-	-	USART1_TX
	PB7	-	LPTIM1_IN2	-	-	I2C1_SDA	-	-	USART1_RX
Port C	PC14	-	-	-	-	-	-	-	-
POILC	PC15	-	-	-	-	-	-	-	-
Port H	PH3	-	-	-	-	-	-	-	-

Table 15. Alternate function AF8 to AF15 (for AF0 to AF7 see Table 14)

		AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
Po	ort	LPUART1	CAN1/TSC	USB/QUADSPI	-	COMP1/ COMP2/ SWPMI1	SAI1	TIM2/TIM15/ TIM16/LPTIM2	EVENTOUT
	PA0	-	-	-	-	COMP1_OUT	SAI1_EXTCLK	TIM2_ETR	EVENTOUT
	PA1	-	-	-	-	-	-	TIM15_CH1N	EVENTOUT
	PA2	LPUART1_TX	-	QUADSPI_ BK1_NCS	-	COMP2_OUT	-	TIM15_CH1	EVENTOUT
	PA3	LPUART1_RX	-	QUADSPI_CLK	-	-	SAI1_MCLK_A	TIM15_CH2	EVENTOUT
	PA4	-	-	-	-	-	SAI1_FS_B	LPTIM2_OUT	EVENTOUT
	PA5	-	-	-	-	-	-	LPTIM2_ETR	EVENTOUT
	PA6	LPUART1_CTS	-	QUADSPI_ BK1_IO3	-	TIM1_BKIN_ COMP2	-	TIM16_CH1	EVENTOUT
Port A	PA7	-	-	QUADSPI_ BK1_IO2	-	COMP2_OUT	-	-	EVENTOUT
POILA	PA8	-	-	-	-	SWPMI1_IO	SAI1_SCK_A	LPTIM2_OUT	EVENTOUT
	PA9	-	-	-	-	-	SAI1_FS_A	TIM15_BKIN	EVENTOUT
	PA10	-	-	USB_CRS_ SYNC	-	-	SAI1_SD_A	-	EVENTOUT
	PA11	-	CAN1_RX	USB_DM	-	TIM1_BKIN2_ COMP1	-	-	EVENTOUT
	PA12	-	CAN1_TX	USB_DP	-	-	-	-	EVENTOUT
	PA13	-	-	USB_NOE	-	SWPMI1_TX	SAI1_SD_B	-	EVENTOUT
	PA14	-	-	-	-	SWPMI1_RX	SAI1_FS_B	-	EVENTOUT
	PA15	-	TSC_G3_IO1	-	-	SWPMI1_ SUSPEND	-	-	EVENTOUT





Table 15. Alternate function AF8 to AF15 (for AF0 to AF7 see *Table 14*) (continued)

		AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
Po	ort	LPUART1	CAN1/TSC	USB/QUADSPI	-	COMP1/ COMP2/ SWPMI1	SAI1	TIM2/TIM15/ TIM16/LPTIM2	EVENTOUT
	PB0	-	-	QUADSPI_ BK1_IO1	-	COMP1_OUT	SAI1_EXTCLK	-	EVENTOUT
	PB1	LPUART1_RTS _DE	-	QUADSPI_ BK1_IO0	-	-	-	LPTIM2_IN1	EVENTOUT
Port B	PB3	-	-	-	-	-	SAI1_SCK_B	-	EVENTOUT
. 0	PB4	-	TSC_G2_IO1	-	-	-	SAI1_MCLK_B	-	EVENTOUT
,	PB5	-	TSC_G2_IO2	-	-	COMP2_OUT	SAI1_SD_B	TIM16_BKIN	EVENTOUT
•	PB6	-	TSC_G2_IO3	-	-	-	SAI1_FS_B	TIM16_CH1N	EVENTOUT
•	PB7	-	TSC_G2_IO4	-	-	-	-	-	EVENTOUT
Dort C	PC14	-	-	-	-	-	-	-	EVENTOUT
Port C	PC15	-	-	-	-	-	-	-	EVENTOUT
Port H	PH3	-	-	-	_	-	-	-	EVENTOUT

# 5 Memory mapping

0xFFFF FFFF 0xBFFF FFFF Reserved Cortex™-M4 0xA000 1400 with FPU 7 **QUADSPI** registers Internal 0xA000 1000 Peripherals 0xE000 0000 0x5FFF FFFF Reserved 6 0x5006 0C00 AHB2 0x4800 0000 0xC000 0000 Reserved 0x4002 4400 **QUADSPI** AHB1 registers 5 0x4002 0000 Reserved 0xA000 1000 0x4001 5800 APB2 0xA000 0000 0x4001 0000 QUADSPI Flash Reserved bank 0x4000 9800 0x9000 0000 APB1 0x4000 0000 0x1FFF FFFF 0x8000 0000 3 Reserved 0x6000 0000 0x1FFF 7810 Options Bytes 2 0x1FFF 7800 Reserved 0x1FFF 7400 Peripherals OTP area 0x4000 0000 0x1FFF 7000 System memory 1 0x1FFF 0000 SRAM2 Reserved 0x2000 C000 0x1000 4000 SRAM1 SRAM2 0x2000 0000 0x1000 0000 Reserved 0 0x0804 0000 CODE Flash memory 0x0800 0000 Reserved 0x0000 0000 0x0004 0000 Flash, system memory or SRAM, depending on BOOT configuration 0x0000 0000 Reserved MSv36892V2

Figure 5. STM32L432xx memory map



Table 16. STM32L432xx memory map and peripheral register boundary addresses (1)

Bus	Boundary address	Size(bytes)	Peripheral
	0x5006 0800 - 0x5006 0BFF	1 KB	RNG
	0x5004 0400 - 0x5006 07FF	158 KB	Reserved
	0x5004 0000 - 0x5004 03FF	1 KB	ADC
	0x5000 0000 - 0x5003 FFFF	16 KB	Reserved
AHB2	0x4800 2000 - 0x4FFF FFFF	~127 MB	Reserved
ANDZ	0x4800 1C00 - 0x4800 1FFF	1 KB	GPIOH
	0x4800 0C00 - 0x4800 1BFF	4 KB	Reserved
	0x4800 0800 - 0x4800 0BFF	1 KB	GPIOC
	0x4800 0400 - 0x4800 07FF	1 KB	GPIOB
	0x4800 0000 - 0x4800 03FF	1 KB	GPIOA
-	0x4002 4400 - 0x47FF FFFF	~127 MB	Reserved
	0x4002 4000 - 0x4002 43FF	1 KB	TSC
	0x4002 3400 - 0x4002 3FFF	1 KB	Reserved
	0x4002 3000 - 0x4002 33FF	1 KB	CRC
	0x4002 2400 - 0x4002 2FFF	3 KB	Reserved
AHB1	0x4002 2000 - 0x4002 23FF	1 KB	FLASH registers
АПВІ	0x4002 1400 - 0x4002 1FFF	3 KB	Reserved
	0x4002 1000 - 0x4002 13FF	1 KB	RCC
	0x4002 0800 - 0x4002 0FFF	2 KB	Reserved
	0x4002 0400 - 0x4002 07FF	1 KB	DMA2
	0x4002 0000 - 0x4002 03FF	1 KB	DMA1
	0x4001 5800 - 0x4001 FFFF	42 KB	Reserved
	0x4001 5400 - 0x4000 57FF	1 KB	SAI1
	0x4001 4800 - 0x4000 53FF	3 KB	Reserved
	0x4001 4400 - 0x4001 47FF	1 KB	TIM16
	0x4001 4000 - 0x4001 43FF	1 KB	TIM15
APB2	0x4001 3C00 - 0x4001 3FFF	1 KB	Reserved
	0x4001 3800 - 0x4001 3BFF	1 KB	USART1
	0x4001 3400 - 0x4001 37FF	1 KB	Reserved
	0x4001 3000 - 0x4001 33FF	1 KB	SPI1
	0x4001 2C00 - 0x4001 2FFF	1 KB	TIM1
	0x4001 2000 - 0x4001 2BFF	3 KB	Reserved

Table 16. STM32L432xx memory map and peripheral register boundary addresses

Bus	Boundary address	Size(bytes)	Peripheral
	0x4001 1C00 - 0x4001 1FFF	1 KB	FIREWALL
	0x4001 0800- 0x4001 1BFF	5 KB	Reserved
APB2	0x4001 0400 - 0x4001 07FF	1 KB	EXTI
APBZ	0x4001 0200 - 0x4001 03FF		COMP
	0x4001 0030 - 0x4001 01FF	1 KB	Reserved
	0x4001 0000 - 0x4001 002F		SYSCFG
	0x4000 9800 - 0x4000 FFFF	26 KB	Reserved
	0x4000 9400 - 0x4000 97FF	1 KB	LPTIM2
	0x4000 8C00 - 0x4000 93FF	2 KB	Reserved
	0x4000 8800 - 0x4000 8BFF	1 KB	SWPMI1
	0x4000 8400 - 0x4000 87FF	1 KB	Reserved
	0x4000 8000 - 0x4000 83FF	1 KB	LPUART1
	0x4000 7C00 - 0x4000 7FFF	1 KB	LPTIM1
	0x4000 7800 - 0x4000 7BFF	1 KB	OPAMP
	0x4000 7400 - 0x4000 77FF	1 KB	DAC
	0x4000 7000 - 0x4000 73FF	1 KB	PWR
	0x4000 6C00 - 0x4000 6FFF	1 KB	USB SRAM
	0x4000 6800 - 0x4000 6BFF	1 KB	USB FS
	0x4000 6400 - 0x4000 67FF	1 KB	CAN1
APB1	0x4000 6000 - 0x4000 63FF	1 KB	CRS
	0x4000 5C00- 0x4000 5FFF	1 KB	I2C3
	0x4000 5800 - 0x4000 5BFF	1 KB	Reserved
	0x4000 5400 - 0x4000 57FF	1 KB	I2C1
	0x4000 4800 - 0x4000 53FF	3 KB	Reserved
	0x4000 4400 - 0x4000 47FF	1 KB	USART2
	0x4000 4000 - 0x4000 43FF	1 KB	Reserved
	0x4000 3C00 - 0x4000 3FFF	1 KB	SPI3
	0x4000 3400 - 0x4000 3BFF	2 KB	Reserved
	0x4000 3000 - 0x4000 33FF	1 KB	IWDG
	0x4000 2C00 - 0x4000 2FFF	1 KB	WWDG
	0x4000 2800 - 0x4000 2BFF	1 KB	RTC
	0x4000 1800 - 0x4000 27FF	4 KB	Reserved
	0x4000 1400 - 0x4000 17FF	1 KB	TIM7

Table 16. STM32L432xx memory map and peripheral register boundary addresses

Bus	Boundary address	Size(bytes)	Peripheral
	0x4000 1000 - 0x4000 13FF	1 KB	TIM6
APB1	0x4000 0400- 0x4000 0FFF	3 KB	Reserved
	0x4000 0000 - 0x4000 03FF	1 KB	TIM2

<sup>1.</sup> The gray color is used for reserved boundary addresses.

#### 6 Electrical characteristics

#### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 6.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean  $\pm 3\sigma$ ).

#### 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25$  °C,  $V_{DD} = V_{DDA} = 3$  V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean  $\pm 2\sigma$ ).

# 6.1.3 Typical curves

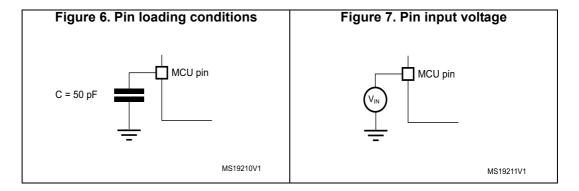
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 6.1.4 Loading capacitor

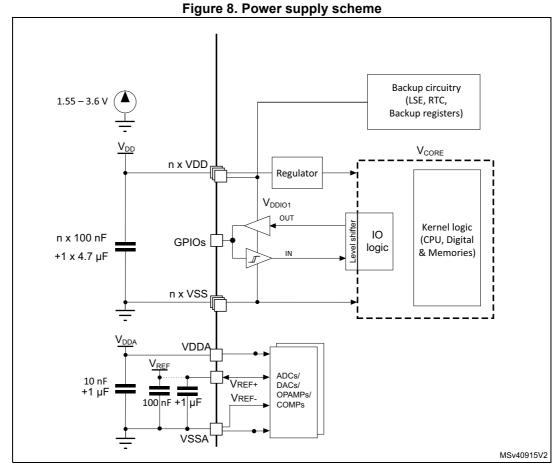
The loading conditions used for pin parameter measurement are shown in *Figure 6*.

#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 7*.



### 6.1.6 Power supply scheme



Caution:

Each power supply pair ( $V_{DD}/V_{SS}$ ,  $V_{DDA}/V_{SSA}$  etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

#### 6.1.7 Current consumption measurement

IDD\_USB
VDDUSB
VDDUSB
VDD
VDD
VDD

MSv41630V1

Figure 9. Current consumption measurement scheme

# 6.2 Absolute maximum ratings

60/149

Stresses above the absolute maximum ratings listed in *Table 17: Voltage characteristics*, *Table 18: Current characteristics* and *Table 19: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Symbol	Ratings	Min	Max	Unit
V <sub>DDX</sub> - V <sub>SS</sub>	External main supply voltage (including $V_{DD}$ , $V_{DDA}$ , $V_{DDUSB}$ )	-0.3	4.0	٧
(2)	Input voltage on FT_xxx pins	V <sub>SS</sub> -0.3	min ( $V_{DD}$ , $V_{DDA}$ , $V_{DDUSB}$ ) + $4.0^{(3)(4)}$	
V <sub>IN</sub> <sup>(2)</sup>	Input voltage on TT_xx pins	V <sub>SS</sub> -0.3	4.0	V
	Input voltage on any other pins	V <sub>SS</sub> -0.3	4.0	
$ \Delta V_{DDx} $	Variations between different V <sub>DDX</sub> power pins of the same domain	-	50	mV
V <sub>SSx</sub> -V <sub>SS</sub>	Variations between all the different ground pins <sup>(5)</sup>	-	50	mV

Table 17. Voltage characteristics<sup>(1)</sup>

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All main power (V<sub>DD</sub>, V<sub>DDA</sub>, V<sub>DDUSB</sub>,) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

V<sub>IN</sub> maximum must always be respected. Refer to Table 18: Current characteristics for the maximum allowed injected current values.

<sup>3.</sup> This formula has to be applied only on the power supplies related to the IO structure described in the pin definition table.

- 4. To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.
- 5. Include VREF- pin.

**Table 18. Current characteristics** 

Symbol	Ratings	Max	Unit
$\Sigma IV_{DD}$	Total current into sum of all V <sub>DD</sub> power lines (source) <sup>(1)</sup>	140	
ΣIV <sub>SS</sub>	Total current out of sum of all V <sub>SS</sub> ground lines (sink) <sup>(1)</sup>	140	
IV <sub>DD(PIN)</sub>	Maximum current into each V <sub>DD</sub> power pin (source) <sup>(1)</sup>	100	
IV <sub>SS(PIN)</sub>	Maximum current out of each V <sub>SS</sub> ground pin (sink) <sup>(1)</sup>	100	
	Output current sunk by any I/O and control pin except FT_f	20	
I <sub>IO(PIN)</sub>	Output current sunk by any FT_f pin	20	
	Output current sourced by any I/O and control pin	20	mA
<b>7</b> 1	Total output current sunk by sum of all I/Os and control pins <sup>(2)</sup>	100	
$\Sigma I_{IO(PIN)}$	Total output current sourced by sum of all I/Os and control pins <sup>(2)</sup>	100	
I <sub>INJ(PIN)</sub> (3)	Injected current on FT_xxx, TT_xx, RST and B pins, except PA4, PA5	-5/+0 <sup>(4)</sup>	
	Injected current on PA4, PA5	-5/0	
$\Sigma  I_{INJ(PIN)} $	Total injected current (sum of all I/Os and control pins) <sup>(5)</sup>	25	

- 1. All main power  $(V_{DD}, V_{DDA}, V_{DDUSB})$  and ground  $(V_{SS}, V_{SSA})$  pins must always be connected to the external power supplies, in the permitted range.
- 2. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count QFP packages.
- 3. Positive injection (when  $V_{IN} > V_{DDIOx}$ ) is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- A negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer also to *Table 17: Voltage characteristics* for the maximum allowed input voltage values.
- When several inputs are submitted to a current injection, the maximum ∑|I<sub>INJ(PIN)</sub>| is the absolute sum of the negative injected currents (instantaneous values).

Table 19. Thermal characteristics

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C

# 6.3 Operating conditions

## 6.3.1 General operating conditions

Table 20. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	80	
f <sub>PCLK1</sub>	Internal APB1 clock frequency	-	0	80	MHz
f <sub>PCLK2</sub>	Internal APB2 clock frequency	-	0	80	
V <sub>DD</sub>	Standard operating voltage	-	1.71	3.6	V
		ADC or COMP used	1.62		
$V_{DDA}$	Analog supply voltage	DAC or OPAMP used	1.8	3.6	V
· DDA	Thanks outpry, Tollago	ADC, DAC, OPAMP, COMP not used	0	5.5	
V	LICD complete selfore	USB used	3.0	3.6	V
$V_{DDUSB}$	USB supply voltage	USB not used	0	3.6	] V
		TT_xx I/O	-0.3	V <sub>DDIOx</sub> +0.3	
	I/O input voltage	All I/O except TT_xx	-0.3	$\begin{array}{c} {\rm MIN(MIN(V_{DD},V_{DDA},}\\ {\rm V_{DDUSB})+3.6V,}\\ {\rm 5.5V)^{(2)(3)}} \end{array}$	V
$P_D$	Power dissipation at T <sub>A</sub> = 125 °C for suffix 3 <sup>(4)</sup>	UFQFPN32	-	128	mW
P <sub>D</sub>	Power dissipation at $T_A = 85$ °C for suffix 6 or $T_A = 105$ °C for suffix $7^{(4)}$	UFQFPN32	-	523	mW
	Ambient temperature for the	Maximum power dissipation	<del>-4</del> 0	85	
	suffix 6 version	Low-power dissipation <sup>(5)</sup>	-40	105	
TA	Ambient temperature for the	Maximum power dissipation	-40	105	°C
IA	suffix 7 version	Low-power dissipation <sup>(5)</sup>	-40	125	
	Ambient temperature for the	Maximum power dissipation	-40	125	
	suffix 3 version	Low-power dissipation <sup>(5)</sup>	-40	130	
		Suffix 6 version	-40	105	
$T_J$	Junction temperature range	Suffix 7 version	-40	125	°C
		Suffix 3 version	-40	130	

<sup>1.</sup> When RESET is released functionality is guaranteed down to  $\rm V_{\rm BOR0}$  Min.



<sup>2.</sup> This formula has to be applied only on the power supplies related to the IO structure described by the pin definition table. Maximum I/O input voltage is the smallest value between MIN( $V_{DD}$ ,  $V_{DDA}$ ,  $V_{DDUSB}$ )+3.6 V and 5.5V.

<sup>3.</sup> For operation with voltage higher than Min (V<sub>DD</sub>, V<sub>DDA</sub>, V<sub>DDUSB</sub>) +0.3 V, the internal Pull-up and Pull-Down resistors must be disabled.

<sup>4.</sup> If T<sub>A</sub> is lower, higher P<sub>D</sub> values are allowed as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub> (see Section 7.2: Thermal characteristics).

 In low-power dissipation state, T<sub>A</sub> can be extended to this range as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub> (see Section 7.2: Thermal characteristics).

#### 6.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 21* are derived from tests performed under the ambient temperature condition summarized in *Table 20*.

Table 21. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
+	V <sub>DD</sub> rise time rate		0	∞	
t <sub>VDD</sub>	V <sub>DD</sub> fall time rate	-	10	∞	
+	V <sub>DDA</sub> rise time rate		0	∞	us/V
t <sub>VDDA</sub>	V <sub>DDA</sub> fall time rate	-	10	∞	μ5/ ν
t	V <sub>DDUSB</sub> rise time rate		0	∞	
t <sub>VDDUSB</sub>	V <sub>DDUSB</sub> fall time rate	-	10	∞	

# 6.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 22* are derived from tests performed under the ambient temperature conditions summarized in *Table 20: General operating conditions*.

Table 22. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Тур	Max	Unit	
t <sub>RSTTEMPO</sub> <sup>(2)</sup>	Reset temporization after BOR0 is detected	V <sub>DD</sub> rising	-	250	400	μs	
V <sub>BOR0</sub> <sup>(2)</sup>	Brown-out reset threshold 0	Rising edge	1.62	1.66	1.7	V	
VBOR0`	Brown-out reset tilleshold o	Falling edge	1.6	1.64	1.69	V	
V	Brown-out reset threshold 1	Rising edge	2.06	2.1	2.14	V	
$V_{BOR1}$	Brown-out reset tilleshold i	Falling edge	1.96	2	2.04	v	
V <sub>BOR2</sub>	Brown-out reset threshold 2	Rising edge	2.26	2.31	2.35	V	
	Brown-out reset tilleshold 2	Falling edge	2.16	2.20	2.24	V	
V	Brown-out reset threshold 3	Rising edge	2.56	2.61	2.66	V	
V <sub>BOR3</sub>	Brown-out reset threshold 3	Falling edge	2.47	2.52	2.57	V	
V	Brown-out reset threshold 4	Rising edge	2.85	2.90	2.95	W	
V <sub>BOR4</sub>	Brown-out reset threshold 4	Falling edge	2.76	2.81	2.86	V	
V	Programmable voltage	Rising edge	2.1	2.15	2.19	V	
V <sub>PVD0</sub>	detector threshold 0	Falling edge	2	2.05	2.1	V	
V	PVD threshold 1	Rising edge	2.26	2.31	2.36	V	
V <sub>PVD1</sub>	L AD IIII GUIDIO I	Falling edge	2.15	2.20	2.25		

Table 22. Embedded reset and power control block characteristics (continued)

Symbol	Parameter Parameter	Conditions <sup>(1)</sup>	Min	Тур	Max	Unit
		Rising edge	2.41	2.46	2.51	
V <sub>PVD2</sub>	PVD threshold 2	Falling edge	2.31	2.36	2.41	V
		Rising edge	2.56	2.61	2.66	
V <sub>PVD3</sub>	PVD threshold 3	Falling edge	2.47	2.52	2.57	V
	D) /D there als als 4	Rising edge	2.69	2.74	2.79	\ /
$V_{PVD4}$	PVD threshold 4	Falling edge	2.59	2.64	2.69	V
V	DVD throshold 5	Rising edge	2.85	2.91	2.96	V
$V_{PVD5}$	PVD threshold 5	Falling edge	2.75	2.81	2.86	V
V	PVD threshold 6	Rising edge	2.92	2.98	3.04	V
V <sub>PVD6</sub>	FVD tilleshold o	Falling edge	2.84	2.90	2.96	V
V <sub>hyst_BORH0</sub>	Hysteresis voltage of BORH0	Hysteresis in continuous mode	-	20	-	mV
	, c	Hysteresis in other mode	1	30	-	
V <sub>hyst_BOR_PVD</sub>	Hysteresis voltage of BORH (except BORH0) and PVD	-	-	100	-	mV
I <sub>DD</sub> (BOR_PVD) <sup>(2)</sup>	BOR <sup>(3)</sup> (except BOR0) and PVD consumption from V <sub>DD</sub>	-	-	1.1	1.6	μΑ
V <sub>PVM1</sub>	V <sub>DDUSB</sub> peripheral voltage monitoring	-	1.18	1.22	1.26	V
V	V <sub>DDA</sub> peripheral voltage	Rising edge	1.61	1.65	1.69	V
V <sub>PVM3</sub>	monitoring	Falling edge	1.6	1.64	1.68	V
V	V <sub>DDA</sub> peripheral voltage	Rising edge	1.78	1.82	1.86	٧
$V_{PVM4}$	monitoring	Falling edge	1.77	1.81	1.85	V
V <sub>hyst_PVM3</sub>	PVM3 hysteresis	-	-	10	-	mV
V <sub>hyst_PVM4</sub>	PVM4 hysteresis	-	-	10	-	mV
I <sub>DD</sub> (PVM1)	PVM1 consumption from V <sub>DD</sub>	-	-	0.2	-	μΑ
I <sub>DD</sub> (PVM3/PVM4)	PVM3 and PVM4 consumption from V <sub>DD</sub>	-	-	2	-	μΑ

Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> BOR0 is enabled in all modes (except shutdown) and its consumption is therefore included in the supply current characteristics tables.

# 6.3.4 Embedded voltage reference

The parameters given in *Table 23* are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*.

Table 23. Embedded internal voltage reference

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>REFINT</sub>	Internal reference voltage	-40 °C < T <sub>A</sub> < +130 °C	1.182	1.212	1.232	V
t <sub>S_vrefint</sub> (1)	ADC sampling time when reading the internal reference voltage	-	4 <sup>(2)</sup>	-	1	μs
t <sub>start_vrefint</sub>	Start time of reference voltage buffer when ADC is enable	-	-	8	12 <sup>(2)</sup>	μs
I <sub>DD</sub> (V <sub>REFINTBUF</sub> )	V <sub>REFINT</sub> buffer consumption from V <sub>DD</sub> when converted by ADC	-	-	12.5	20 <sup>(2)</sup>	μΑ
$\Delta V_{REFINT}$	Internal reference voltage spread over the temperature range	V <sub>DD</sub> = 3 V	-	5	7.5 <sup>(2)</sup>	mV
T <sub>Coeff</sub>	Temperature coefficient	-40°C < T <sub>A</sub> < +130°C	-	30	50 <sup>(2)</sup>	ppm/°C
A <sub>Coeff</sub>	Long term stability	1000 hours, T = 25°C	-	-	TBD <sup>(2)</sup>	ppm
V <sub>DDCoeff</sub>	Voltage coefficient	3.0 V < V <sub>DD</sub> < 3.6 V	-	250	1200 <sup>(2)</sup>	ppm/V
V <sub>REFINT_DIV1</sub>	1/4 reference voltage		24	25	26	
V <sub>REFINT_DIV2</sub>	1/2 reference voltage	-	49	50	51	% V <sub>REFINT</sub>
V <sub>REFINT_DIV3</sub>	3/4 reference voltage		74	75	76	INLFIINT

<sup>1.</sup> The shortest sampling time can be determined in the application by multiple iterations.

<sup>2.</sup> Guaranteed by design.

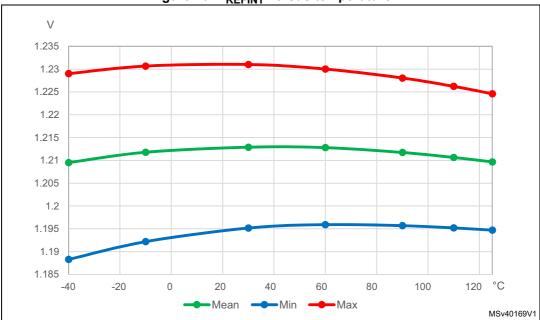


Figure 10. V<sub>REFINT</sub> versus temperature

#### 6.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 9: Current consumption measurement scheme*.

#### Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted with the minimum wait states number, depending on the f<sub>HCLK</sub> frequency (refer to the table "Number of wait states according to CPU clock (HCLK) frequency" available in the RM0393 reference manual).
- When the peripherals are enabled f<sub>PCLK</sub> = f<sub>HCLK</sub>

The parameters given in *Table 24* to *Table 36* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*.



Table 24. Current consumption in Run and Low-power run modes, code with data processing running from Flash, ART enable (Cache ON Prefetch OFF)

		Condi	itions				TYP			MAX <sup>(1)</sup>						
Symbol	Parameter	-	Voltage scaling	f <sub>HCLK</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit	
				26 MHz	2.37	2.38	2.44	2.52	2.66	2.7	2.7	2.8	2.9	3.2		
				16 MHz	1.5	1.52	1.57	1.64	1.79	1.7	1.7	1.8	2.0	2.3		
				8 MHz	0.81	0.82	0.87	0.94	1.08	0.9	0.9	1.0	1.2	1.5		
			Range 2	4 MHz	0.46	0.47	0.52	0.59	0.73	0.5	0.6	0.6	0.8	1.1		
				2 MHz	0.29	0.3	0.34	0.41	0.55	0.3	0.4	0.4	0.6	0.9		
		f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48MHz included,		1 MHz	0.2	0.21	0.25	0.32	0.46	0.2	0.3	0.3	0.5	0.8		
	Supply current in	bypass mode	bypass mode	bypass mode		100 kHz	0.12	0.13	0.17	0.24	0.38	0.1	0.2	0.2	0.4	0.7
I <sub>DD</sub> (Run)	Run mode			80 MHz	8.53	8.56	8.64	8.74	8.92	9.5	9.6	9.7	9.9	10.3		
				72 MHz	7.7	7.73	7.8	7.9	8.08	8.6	8.6	8.7	8.9	9.3		
				64 MHz	6.86	6.9	6.97	7.06	7.23	7.7	7.7	7.8	8.0	8.3		
			Range 1	48 MHz	5.13	5.16	5.23	5.32	5.49	5.8	5.8	6.0	6.1	6.5		
				32 MHz	3.46	3.48	3.55	3.64	3.8	3.9	4.0	4.1	4.2	4.6		
				24 MHz	2.63	2.64	2.71	2.79	2.96	3.0	3.0	3.1	3.3	3.6		
				16 MHz	1.8	1.81	1.87	1.96	2.12	2.0	2.1	2.2	2.3	2.7		
	Supply			2 MHz	211	230	280	355	506	273.8	301.1	360.4	502.7	815.9		
I <sub>DD</sub> (LPRun)	Supply current in	f <sub>HCLK</sub> = f <sub>MSI</sub>		1 MHz	117	134	179	254	404	154.7	184.6	249.6	398.4	712.4	μA	
IDD(LF Kull)	Low-power run mode	all peripherals disable	ole	400 kHz	58.5	70.4	116	189	338	80.2	111.5	179.7	330.8	643.4	μΑ	
	Tall Illode			100 kHz	30	41.1	85.2	159	308	46.5	76.6	147.1	299.1	611.2		

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.





Table 25. Current consumption in Run and Low-power run modes, code with data processing running from Flash, ART disable

		Condi	itions				TYP					MAX <sup>(1)</sup>				
Symbol	Parameter	-	Voltage scaling	f <sub>HCLK</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit	
				26 MHz	2.66	2.68	2.73	2.81	2.96	3.0	3.1	3.2	3.3	3.6		
				16 MHz	1.88	1.9	1.94	2.02	2.17	2.1	2.2	2.3	2.4	2.7		
				8 MHz	1.05	1.06	1.11	1.18	1.33	1.2	1.2	1.3	1.4	1.7		
			Range 2	4 MHz	0.6	0.62	0.66	0.73	0.87	0.7	0.7	0.8	0.9	1.2		
		f = f up to		2 MHz	0.36	0.37	0.34	0.48	0.62	0.4	0.4	0.5	0.6	0.9		
	0 1	TPLL ON ADOVE		1 MHz	0.23	0.25	0.25	0.36	0.5	0.3	0.3	0.4	0.5	8.0		
I <sub>DD</sub> (Run)	Supply current in			100 kHz	0.12	0.14	0.17	0.25	0.39	0.1	0.2	0.2	0.4	0.7	mA	
IDD(IXuII)	Run mode			80 MHz	8.56	8.61	8.69	8.79	8.97	9.6	9.7	9.8	10.0	10.3		
					72 MHz	7.74	7.79	7.86	7.96	8.14	8.7	8.7	8.8	9.0	9.4	
				64 MHz	7.63	7.68	7.75	7.85	8.04	8.6	8.6	8.7	8.9	9.3		
			Range 1	48 MHz	6.36	6.4	6.48	6.58	6.76	7.2	7.3	7.4	7.6	7.9		
				32 MHz	4.56	4.6	4.66	4.76	4.93	5.2	5.2	5.3	5.5	5.8		
				24 MHz	3.45	3.48	3.54	3.64	3.8	3.9	4.0	4.1	4.2	4.6		
				16 MHz	2.48	2.51	2.56	2.65	2.82	2.8	2.9	3.0	3.1	3.5		
	Supply			2 MHz	310	317	364	440	593	375.3	400.9	456.7	595.3	909.6		
L (LDDup)	Supply current in	f <sub>HCLK</sub> = f <sub>MSI</sub>		1 MHz	157	173	226	296	448	204.8	234.2	298.2	445.8	758.9		
I <sub>DD</sub> (LPRun)	Low-power run		ole	400 kHz	72.6	89	130	206	356	99.7	131.2	199.7	349.3	663.7	μA	
	Tuil			100 kHz	32.3	46	89.7	164	314	52.4	82.1	153.3	301.2	616.9		

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.

Table 26. Current consumption in Run and Low-power run modes, code with data processing running from SRAM1

Symbol	Parameter	Conditions		ТҮР				MAX <sup>(1)</sup>							
		-	Voltage scaling	f <sub>HCLK</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
	Supply current in Run mode			26 MHz	2.42	2.43	2.49	2.56	2.71	2.7	2.7	2.8	3.0	3.3	
				16 MHz	1.54	1.55	1.6	1.67	1.82	1.7	1.7	1.8	2.0	2.3	
				8 MHz	0.82	0.84	0.88	0.95	1.1	0.9	1.0	1.0	1.2	1.5	
		f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	4 MHz	0.47	0.48	0.52	0.59	0.73	0.5	0.6	0.6	0.8	1.1	- mA
				2 MHz	0.29	0.3	0.34	0.41	0.55	0.3	0.4	0.4	0.6	0.9	
I <sub>DD</sub> (Run)				1 MHz	0.2	0.21	0.25	0.32	0.46	0.2	0.3	0.3	0.5	0.8	
				100 kHz	0.12	0.13	0.17	0.24	0.38	0.1	0.2	0.2	0.4	0.7	
			Range 1	80 MHz	8.63	8.68	8.74	8.84	9.01	9.5	9.6	9.7	9.9	10.2	
				72 MHz	7.79	7.83	7.9	7.99	8.17	8.6	8.6	8.8	8.9	9.3	
				64 MHz	6.95	6.99	7.05	7.15	7.32	7.7	7.7	7.9	8.0	8.4	
				48 MHz	5.19	5.22	5.29	5.38	5.55	5.8	5.8	5.9	6.1	6.5	
				32 MHz	3.51	3.53	3.6	3.68	3.85	3.9	4.0	4.1	4.2	4.6	
				24 MHz	2.66	2.68	2.74	2.83	2.99	3.0	3.0	3.1	3.3	3.6	
				16 MHz	1.82	1.84	1.89	1.98	2.14	2.0	2.1	2.2	2.3	2.7	
I (I DDun)	Supply current in low-power run mode	f <sub>HCLK</sub> = f <sub>MSI</sub> all peripherals disable FLASH in power-down		2 MHz	205	228	275	352	501	276.5	302.3	358.4	502.5	816.4	
				1 MHz	111	126	175	248	397	151.3	180.9	245.3	390.7	703.4	uА
I <sub>DD</sub> (LPRun)				400 kHz	49.2	62.7	108	181	330	73.3	104.0	170.8	321.0	632.4	μΑ
	rummode			100 kHz	21.5	33.3	76.6	151	299	36.4	67.7	137.2	287.8	600.8	

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.



Table 27. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART enable (Cache ON Prefetch OFF)

Symbol		Conditions			TYP		TYP		
	Parameter	-	Voltage scaling	Code	25 °C	Unit	25 °C	Unit	
			Z	Reduced code <sup>(1)</sup>	2.37		91		
	Supply current in Run mode	f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 f <sub>HCLK</sub> = 26 MHz	Coremark	2.69	mA	103	μΑ/MHz	
I <sub>DD</sub> (Run)				Dhrystone 2.1	2.74		105		
				Fibonacci	2.58		99		
				While(1)	2.30		88		
			Range 1 f <sub>HCLK</sub> = 80 MHz	Reduced code <sup>(1)</sup>	8.53	mA	107	μΑ/MHz	
				Coremark	9.68		121		
				Dhrystone 2.1	9.76		122		
				Fibonacci	9.27		116		
				While(1)	8.20		103		
I <sub>DD</sub> (LPRun)				Reduced code <sup>(1)</sup>	211	μΑ	106	μΑ/MHz	
	Supply current in Low-power run			Coremark	251		126		
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2 M all peripherals dis		Dhrystone 2.1	269		135		
		, , , , , , , , , ,	-	Fibonacci	230		115		
				While(1)	286		143		

<sup>1.</sup> Reduced code used for characterization results provided in *Table 24*, *Table 25*, *Table 26*.



Table 28. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART disable

Symbol		Conditions			TYP		TYP		
	Parameter	-	Voltage scaling	Code	25 °C	Unit	25 °C	Unit	
		f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 f <sub>HCLK</sub> = 26 MHz	Reduced code <sup>(1)</sup>	2.66	mA	102	μΑ/MHz	
				Coremark	2.44		94		
	Supply current in Run mode			Dhrystone 2.1	2.46		95		
				Fibonacci	2.27		87		
I (Bun)				While(1)	2.20		84.6		
I <sub>DD</sub> (Run)			Range 1 f <sub>HCLK</sub> = 80 MHz	Reduced code <sup>(1)</sup>	8.56	mA	107	μΑ/MHz	
				Coremark	8.00		100		
				Dhrystone 2.1	7.98		100		
				Fibonacci	7.41		93		
				While(1)	7.83		98		
				Reduced code <sup>(1)</sup>	310	μА	155	μΑ/MHz	
I <sub>DD</sub> (LPRun)	Supply current in Low-power run	f -f -0.MI	-	Coremark	342		171		
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2 MI all peripherals disa		Dhrystone 2.1	324		162		
		p p		Fibonacci	324		162		
				While(1)	384		192		

<sup>1.</sup> Reduced code used for characterization results provided in *Table 24*, *Table 25*, *Table 26*.

Table 29. Typical current consumption in Run and Low-power run modes, with different codes running from SRAM1

Symbol		Conditions			TYP		TYP	
	Parameter	-	Voltage scaling	Code	25 °C	Unit	25 °C	Unit
			Hz	Reduced code <sup>(1)</sup>	2.42		93	
		f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 <sub>LK</sub> = 26 MHz	Coremark	2.18	mA	84	µA/MHz
				Dhrystone 2.1	2.40		92	
	Supply current in Run mode		Ranç f <sub>HCLK</sub> =	Fibonacci	2.40		92	
I <sub>DD</sub> (Run)			ᅸ	While(1)	2.29		88	
IDD(IXuII)			Range 1 f <sub>HCLK</sub> = 80 MHz	Reduced code <sup>(1)</sup>	8.63	mA	108	μΑ/MHz
				Coremark	7.76		97	
				Dhrystone 2.1	8.55		107	
				Fibonacci	8.56		107	
				While(1)	8.12		102	
I <sub>DD</sub> (LPRun)				Reduced code <sup>(1)</sup>	205		103	
	Supply current in Low-power run			Coremark	188		94	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2 MH   all peripherals disa		Dhrystone 2.1	222	μΑ	111	μΑ/MHz
		an penpherale alea	<b></b>	Fibonacci	204		102	
				While(1)	211		106	

<sup>1.</sup> Reduced code used for characterization results provided in *Table 24*, *Table 25*, *Table 26*.



Table 30. Current consumption in Sleep and Low-power sleep modes, Flash ON

		Cond	ditions				TYP					MAX <sup>(1)</sup>			
Symbol	Parameter	-	Voltage scaling	f <sub>HCLK</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
				26 MHz	0.68	0.69	0.74	0.81	0.95	0.8	0.8	0.9	1.0	1.3	
				16 MHz	0.46	0.48	0.52	0.59	0.73	0.5	0.6	0.6	0.8	1.1	
				8 MHz	0.29	0.30	0.34	0.41	0.55	0.3	0.4	0.4	0.6	0.9	
			Range 2	4 MHz	0.20	0.21	0.25	0.32	0.46	0.2	0.3	0.3	0.5	0.8	
		f <sub>HCLK</sub> = f <sub>HSE</sub> up to 48 MHz		2 MHz	0.16	0.17	0.21	0.28	0.42	0.2	0.2	0.3	0.4	0.7	
	Supply	included, bypass		1 MHz	0.13	0.15	0.19	0.26	0.40	0.1	0.2	0.3	0.4	0.7	
I <sub>DD</sub> (Sleep)	current in	mode		100 kHz	0.11	0.13	0.17	0.24	0.38	0.1	0.2	0.2	0.4	0.7	mA
Прр(спесь)	sleep	pll ON above		80 MHz	2.23	2.25	2.30	2.38	2.54	2.5	2.5	2.6	2.8	3.1	] ''''
	mode,	48 MHz all peripherals		72 MHz	2.02	2.04	2.10	2.18	2.34	2.2	2.3	2.4	2.5	2.9	
		disable		64 MHz	1.82	1.84	1.89	1.98	2.14	2.0	2.1	2.1	2.3	2.6	
			Range 1	48 MHz	1.34	1.36	1.42	1.50	1.66	1.5	1.6	1.7	1.8	2.2	
				32 MHz	0.93	0.95	1.01	1.09	1.25	1.1	1.1	1.2	1.4	1.7	
				24 MHz	0.73	0.75	0.80	0.88	1.04	8.0	0.9	1.0	1.1	1.4	
				16 MHz	0.53	0.55	0.60	0.68	0.84	0.6	0.6	0.7	0.9	1.2	
	Supply			2 MHz	71.8	80.7	125	200	350	91.1	122.7	191.3	341.5	653.5	
I <sub>DD</sub> (LPSleep)	current in low-power	f <sub>HCLK</sub> = f <sub>MSI</sub>		1 MHz	45.0	57.3	101	176	325	63.2	95.4	165.4	316.5	628.7	μA
iDD(rt. 2166h)	sleep	all peripherals dis	able	400 kHz	27.0	40.7	84.6	158	308	43.9	75.8	147.2	297.6	609.2	μΑ
	mode			100 kHz	22.8	30.9	63.3	113.2	207.7	35.2	67.9	140.9	290.8	602.4	

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.

Table 31. Current consumption in Low-power sleep modes, Flash in power-down

		Co	nditions				TYP					MAX <sup>(1)</sup>			
Symbol	Parameter	-	Voltage scaling	f <sub>HCLK</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
				2 MHz	58.7	70.7	103.2	153.7	248.5	80	113	180	330	641	
I <sub>DD</sub> (LPSleep		f <sub>HCLK</sub> = f <sub>MSI</sub>		1 MHz	39.4	47.2	79.3	129.6	224.8	53	86	154	304	616	μA
)	sleep mode	all peripherals	s disable	400 kHz	20.8	30.8	62.1	112.5	207.8	35	67	137	286	597	μΛ
				100 kHz	14.3	23.1	55.1	105.7	201.5	27	58	130	279	590	

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.

Table 32. Current consumption in Stop 2 mode

Cumbal	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Symbol	Parameter	-	V <sub>DD</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
			1.8 V	1	2.54	8.74	19.8	43.4	2.0	5.6	21.1	50.8	116.0	
I <sub>DD</sub> (Stop 2)	Supply current in Stop 2 mode,	_	2.4 V	1.02	2.59	8.89	20.2	44.3	2.1	5.8	21.6	52.3	119.6	μA
IDD (Glob 2)	RTC disabled	-	3 V	1.06	2.67	9.11	20.7	45.5	2.1	5.9	22.2	53.7	123.2	μΛ
			3.6 V	1.23	2.88	9.56	21.6	47.3	2.3	6.1	23.0	55.8	127.9	
			1.8 V	1.3	2.82	9.02	20.1	43.6	2.5	6.2	21.6	51.3	116.3	
		RTC clocked by LSI	2.4 V	1.39	2.95	9.24	20.5	44.6	2.8	6.4	22.3	52.8	120.0	
		TO GOORED BY LOT	3 V	1.5	3.11	9.55	21.1	45.8	3.0	6.8	23.0	54.5	123.8	
			3.6 V	1.76	3.42	10.1	22.1	47.8	3.3	7.2	24.1	56.7	128.7	
			1.8 V	1.36	2.9	9.1	20.1	43.7	-	-	-	-	-	
I <sub>DD</sub> (Stop 2	Supply current in Stop 2 mode,	RTC clocked by LSE	2.4 V	1.48	3.09	9.44	20.8	45	-	-	-	-	-	μA
with RTC)	RTC enabled	bypassed at 32768 Hz	3 V	1.83	3.67	10.4	22.3	47.3	-	-	-	-	-	μΛ
			3.6 V	3.58	6.17	13.9	26.6	53	-	-	-	-	-	
			1.8 V	1.28	2.81	9.13	20.8	-	-	-	-	-	-	
		RTC clocked by LSE quartz <sup>(2)</sup>	2.4 V	1.39	2.93	9.34	21.3	-	-	-	-	-	-	
		in low drive mode	3 V	1.59	3.1	9.64	21.8	-	-	-	-	-	-	
			3.6 V	1.86	3.45	10.2	22.8	-	-	-	-	-	-	





Table 32. Current consumption in Stop 2 mode (continued)

Cumbal	Doromotor	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Symbol	Parameter	-	V <sub>DD</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
		Wakeup clock is MSI = 48 MHz, voltage Range 1. See <sup>(3)</sup> .	3 V	1.85	-	-	-	-	-	-	-	-	-	
I <sub>DD</sub> (wakeup du from Stop2)	Supply current during wakeup from Stop 2 mode	Wakeup clock is MSI = 4 MHz, voltage Range 2. See <sup>(3)</sup> .	3 V	1.52	-	-	-	-	-	-	-	-	-	mA
		Wakeup clock is HSI16 = 16 MHz, voltage Range 1. See <sup>(3)</sup> .	3 V	1.54	-	-	-	-	-	-	-	-	-	

<sup>1.</sup> Guaranteed based on test during characterization, unless otherwise specified.

<sup>2.</sup> Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

<sup>3.</sup> Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in *Table 38: Low-power mode wakeup timings*.

Table 33. Current consumption in Stop 1 mode

Symbol	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Symbol	Farameter	-	$V_{DD}$	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Oilit
	Supply		1.8 V	4.34	12.4	43.6	96.4	204	9.3	27.4	98.9	198.7	397.5	
I <sub>DD</sub> (Stop 1)	current in	_	2.4 V	4.35	12.5	43.8	97	205	9.4	27.6	99.5	199.0	398.0	μA
	Stop 1 mode,		3 V	4.41	12.6	44.1	97.7	207	9.5	27.8	100.3	200.4	400.8	μπ
	RTC disabled		3.6 V	4.56	12.9	44.8	98.9	210	9.7	28.3	101.7	202.1	404.2	
			1.8 V	4.63	12.7	43.9	96.8	205	9.9	28.0	99.5	198.9	397.8	
		RTC clocked by LSI	2.4 V	4.78	12.8	44.2	97.4	206	10.1	28.3	100.3	199.5	399.0	
		TYTO GIOCKED BY LOT	3 V	4.93	13	44.6	98.1	207	10.4	28.7	101.2	200.9	401.9	
			3.6 V	5.05	13.4	45.3	99.5	210	10.8	29.4	102.8	202.5	405.0	
	Supply		1.8 V	4.7	12.8	44	96.9	205	-	-	-	-	-	
I <sub>DD</sub> (Stop 1	current in stop	RTC clocked by LSE	2.4 V	4.95	13	44.4	97.6	206	-	-	-	-	-	μΑ
with RTC)	1 mode,	bypassed, at 32768 Hz	3 V	5.33	13.6	45.4	99.1	209	-	-	-	-	-	μΛ
	RTC enabled		3.6 V	6.91	16.1	48.8	103	216	-	-	-	-	-	
			1.8 V	4.76	12.3	43.7	99.1	-	-	-	-	-	-	
		RTC clocked by LSE quartz <sup>(2)</sup>	2.4 V	4.95	12.4	43.8	99.3	-	-	-	-	-	-	•
		in low drive mode	3 V	5.1	12.6	44.1	99.6	-	-	-	-	-	-	
			3.6 V	5.65	13	44.8	101	-	-	-	-	-	-	
		Wakeup clock MSI = 48 MHz, voltage Range 1. See <sup>(3)</sup> .	3 V	1.14	-	-	-	-	-	-	-	-	-	
I <sub>DD</sub> (wakeup from Stop1)	Supply current during wakeup from Stop 1	Wakeup clock MSI = 4 MHz, voltage Range 2. See <sup>(3)</sup> .	3 V	1.22	-	-	-	-	-	-	-	-	-	mA
	r	Wakeup clock HSI16 = 16 MHz, voltage Range 1. See <sup>(3)</sup> .	3 V	1.20	-	-	-	-	-	-	-	-	-	

<sup>1.</sup> Guaranteed based on test during characterization, unless otherwise specified.

<sup>3.</sup> Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in *Table 38: Low-power mode wakeup timings*.



<sup>2.</sup> Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.



Table 34. Current consumption in Stop 0

			J J J			•	•	_					
Symbol	Parameter	Conditions			TYP					MAX <sup>(1)</sup>			Unit
Symbol	Parameter	V <sub>DD</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	
	Supply	1.8 V	108	119	158	221	347	133	158	244	395	704	
(Stop ())	current in	2.4 V	110	121	160	223	349	136	161	248	399	710	
I <sub>DD</sub> (Stop 0)	Stop 0 mode,	3 V	111	123	161	224	352	139	164	251	403	716	μΑ
	RTC disabled	3.6 V	114	125	163	227	355	142	167	254	408	722 <sup>(2)</sup>	

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.

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<sup>2.</sup> Guaranteed by test in production.

Electrical characteristics

Table 35. Current consumption in Standby mode

Symbol	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Symbol	Parameter	-	$V_{DD}$	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Unit
			1.8 V	27.7	144	758	2 072	5 425	119	425	2866	7524	20510	
		no independent watchdog	2.4 V	50.9	187	892	2 408	6 247	183	564	3383	8778	23768	
	Supply current in Standby	no independent waterdog	3 V	90.2	253	1 090	2 884	7 409	225	681	3912	10071	26976	
I <sub>DD</sub> (Standby)	mode (backup		3.6 V	253	459	1 474	3 575	8 836	292	877	4638	11659	30758	nA
IDD(Stariuby)	registers		1.8 V	216	-		-	-	-	-	-	-	-	IIA
	retained), RTC disabled	with independent	2.4 V	342	-	-	-	-	-	-	-	-	-	
		watchdog	3 V	416	-	-	-	-	-	-	-	-	-	
			3.6 V	551	-	-	-	-	-	-	-	-	-	
			1.8 V	287	407	989	2 230	5 396	585	944	3344	7866	20504	
		RTC clocked by LSI, no	2.4 V	386	526	1 201	2 638	6 274	811	1230	4007	9246	23824	
		independent watchdog	3 V	513	679	1 478	3 167	7 414	1022	1521	4683	10671	27124	
			3.6 V	771	978	1 963	3 992	9 039	1284	1924	5577	12383	30954 (2)	nA
			1.8 V	342	-	-	-	-	-	-	-	-	-	
	Supply current	RTC clocked by LSI, with	2.4 V	521	-	-	-	-	-	-	-	-	-	
	in Standby	independent watchdog	3 V	655	-	-	-	-	-	-	-	-	-	
I <sub>DD</sub> (Standby with RTC)	mode (backup		3.6 V	865	-	-	-	-	-	-	-	-	-	
with KTC)	registers retained),		1.8 V	142	126	865	2 220	5 650	-	-	-	-	-	
	RTC enabled	RTC clocked by LSE	2.4 V	249	219	1 090	2 660	6 600	-	-	-	-	-	
		bypassed at 32768Hz	3 V	404	364	1 410	3 260	7 850	-	-	-	-	-	
			3.6 V	742	670	2 000	4 230	9 700	-	-	-	-	-	nA
			1.8 V	281	423	1 046	2 410	5 700	-	-	-	-	-	IIA
		RTC clocked by LSE	2.4 V	388	548	1 268	2 847	6 564	-	-	-	-	-	
		quartz (3) in low drive mode	3 V	535	715	1 565	3 420	7 694	-	-	-	-	-	
			3.6 V	836	1 048	2 081	4 311	9 338	-	-	-	-	-	





Table 35. Current consumption in Standby mode (continued)

Symbol	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Cymbol	i arameter	-	V <sub>DD</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	
	Supply current		1.8 V	173	349	1 009	2 158	4 542	249	527	1604	3402	6908	
I <sub>DD</sub> (SRAM2)	to be added in Standby mode	_	2.4 V	174	345	1 015	2 163	4 535	271	589	1623	3438	6924	nA
(4)	when SRAM2		3 V	178	350	1 019	2 148	4 419	277	594	1628	3467	6935	] "" (
	is retained		3.6 V	184	352	1 033	2 208	4 610	293	611	1631	3480	6948	
I <sub>DD</sub> (wakeup from Standby)	Supply current during wakeup from Standby mode	Wakeup clock is MSI = 4 MHz. See <sup>(5)</sup> .	3 V	1.23	-	-	-	-	-	-	-	-	-	mA

- 1. Guaranteed by characterization results, unless otherwise specified.
- 2. Guaranteed by test in production.
- 3. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
- 4. The supply current in Standby with SRAM2 mode is: I<sub>DD</sub>(Standby) + I<sub>DD</sub>(SRAM2). The supply current in Standby with RTC with SRAM2 mode is: I<sub>DD</sub>(Standby + RTC) + I<sub>DD</sub>(SRAM2).
- 5. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in *Table 38: Low-power mode wakeup timings*.

Table 36. Current consumption in Shutdown mode

Symbol	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
Symbol	Faiailletei	-	$V_{DD}$	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	
	Supply current		1.8 V	7.82	190	386	1 286	3 854	25.0	255	1721	5052	15543	
	in Shutdown mode		2.4 V	23	229	485	1 517	4 431	34.9	270	2085	5878	17639	
I <sub>DD</sub> (Shutdown)		-	3 V	44.3	290	634	1 878	5 310	70.1	345	2454	6755	19984	nA
	registers retained) RTC disabled		3.6 V	212	397	977	2 516	6 656	119.1	496	2992	7939	22860	

Table 36. Current consumption in Shutdown mode (continued)

Symbol	Parameter	Conditions				TYP					MAX <sup>(1)</sup>			Unit
- Oyiiiboi	i arameter	-	V <sub>DD</sub>	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	55 °C	85 °C	105 °C	125 °C	Oint
			1.8 V	63	133	522	1 490	4 270	-	-	-	-	-	
	Supply current	RTC clocked by LSE	2.4 V	165	253	710	1 830	4 980	-	-	-	-	-	
	in Shutdown	bypassed at 32768 Hz	3 V	316	423	990	2 340	6 050	-	-	-	-	-	
I <sub>DD</sub> (Shutdown	mode (backup		3.6 V	649	787	1 530	3 220	7 710	1	-	-	-	-	nA
with RTC)	registers		1.8 V	203	293	700	1 675	-	-	-	-	-	-	IIA
	registers retained) RTC	RTC clocked by LSE quartz <sup>(2)</sup> in low drive	2.4 V	303	411	880	2 001	-	1	-	-	-	-	
	enabled	mode	3 V	448	567	1 136	2 479	-	-	-	-	-	-	
			3.6 V	744	887	1 609	3 256	-	1	-	-	-	-	
I <sub>DD</sub> (wakeup from Shutdown)	Supply current during wakeup from Shutdown mode	Wakeup clock is MSI = 4 MHz. See <sup>(3)</sup> .	3 V	0.780	-	-	-	-	-	-	-	-	-	mA

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.



<sup>2.</sup> Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

<sup>3.</sup> Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in *Table 38: Low-power mode wakeup timings*.

#### I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

#### I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 56: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

#### Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

#### I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 37: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIOx} \times f_{SW} \times C$$

where

 $I_{\mbox{SW}}$  is the current sunk by a switching I/O to charge/discharge the capacitive load

V<sub>DDIOx</sub> is the I/O supply voltage

 $f_{\mbox{SW}}$  is the I/O switching frequency

C is the total capacitance seen by the I/O pin:  $C = C_{INT} + C_{EXT} + C_{S}$ 

C<sub>S</sub> is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

#### On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 37*. The MCU is placed under the following conditions:

- All I/O pins are in Analog mode
- The given value is calculated by measuring the difference of the current consumptions:
  - when the peripheral is clocked on
  - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in Table 17: Voltage characteristics
- The power consumption of the digital part of the on-chip peripherals is given in *Table 37*. The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

Table 37. Peripheral current consumption

	Peripheral	Range 1	Range 2	Low-power run and sleep	Unit
	Bus Matrix <sup>(1)</sup>	3.2	2.9	3.1	
	ADC independent clock domain	0.4	0.1	0.2	
	ADC clock domain	2.1	1.9	1.9	
	CRC	0.4	0.2	0.3	
	DMA1	1.4	1.3	1.4	
	DMA2	1.5	1.3	1.4	
	FLASH	6.2	5.2	5.8	
	GPIOA <sup>(2)</sup>	1.7	1.4	1.6	
AHB	GPIOB <sup>(2)</sup> )	1.6	1.3	1.6	
АПВ	GPIOC <sup>(2)</sup>	1.7	1.5	1.6	
	GPIOH <sup>(2)</sup>	0.6	0.6	0.5	
	QSPI	7.0	5.8	7.3	µA/MHz
	RNG independent clock domain	2.2	NA	NA	
	RNG clock domain	0.5	NA	NA	
	SRAM1	0.8	0.9	0.7	
	SRAM2	1.0	0.8	0.8	
	TSC	1.6	1.3	1.3	
	All AHB Peripherals	21.7	18.5	20.3	
	AHB to APB1 bridge <sup>(3)</sup>	0.9	0.7	0.9	
	CAN1	4.1	3.2	3.9	
APB1	DAC1	2.4	1.8	2.2	
	RTCA	1.7	1.1	2.1	
	CRS	0.3	0.3	0.6	



Table 37. Peripheral current consumption (continued)

	Peripheral	Range 1	Range 2	Low-power run and sleep	Unit
	USB FS independent clock domain	2.9	NA	NA	
	USB FS clock domain	2.3	NA	NA	
	I2C1 independent clock domain	3.5	2.8	3.4	
	I2C1 clock domain	1.1	0.9	1.0	
	I2C3 independent clock domain	2.9	2.3	2.5	
	I2C3 clock domain	0.9	0.4	0.8	
	LPUART1 independent clock domain	1.9	1.6	1.8	
	LPUART1 clock domain	0.6	0.6	0.6	
	LPTIM1 independent clock domain	2.9	2.4	2.8	
	LPTIM1 clock domain	0.8	0.4	0.7	
	LPTIM2 independent clock domain	3.1	2.7	3.9	
APB1	LPTIM2 clock domain	0.8	0.7	0.8	
	OPAMP	0.4	0.2	0.4	
	PWR	0.4	0.1	0.4	
	SPI3	1.7	1.3	1.6	μΑ/MHz
	SWPMI1 independent clock domain	1.9	1.6	1.9	
	SWPMI1 clock domain	0.9	0.7	0.8	
	TIM2	6.2	5.0	5.9	
	TIM6	1.0	0.6	0.9	
	TIM7	1.0	0.6	0.6	
	USART2 independent clock domain	4.1	3.6	3.8	
	USART2 clock domain	1.3	0.9	1.1	
	WWDG	0.5	0.5	0.5	
	All APB1 on	40.2	26.7	37.9	
	AHB to APB2 <sup>(4)</sup>	1.0	0.9	0.9	
	FW	0.2	0.2	0.2	
ADDO	SAI1 independent clock domain	2.3	1.8	1.9	
APB2	SAI1 clock domain	2.1	1.8	2.0	
	SPI1	1.8	1.6	1.7	
	SYSCFG/COMP	0.6	0.5	0.6	



	Peripheral	Range 1	Range 2	Low-power run and sleep	Unit
	TIM1	8.1	6.5	7.6	
	TIM15	3.7	3.0	3.4	
	TIM16	2.7	2.1	2.6	
APB2	USART1 independent clock domain	4.8	4.2	4.6	μΑ/MHz
	USART1 clock domain	1.5	1.3	1.7	
	All APB2 on	24.2	19.9	22.6	
	ALL	86.1	65.1	80.9	

Table 37. Peripheral current consumption (continued)

# 6.3.6 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in *Table 38* are the latency between the event and the execution of the first user instruction.

The device goes in low-power mode after the WFE (Wait For Event) instruction.

Table 38. Low-power mode wakeup timings<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>WUSLEEP</sub>	Wakeup time from Sleep mode to Run mode	-	6	6	Nb of
t <sub>WULPSLEEP</sub>	Wakeup time from Low- power sleep mode to Low- power run mode	Wakeup in Flash with Flash in power-down during low-power sleep mode (SLEEP_PD=1 in FLASH_ACR) and with clock MSI = 2 MHz	6	8.3	CPU cycles

<sup>1.</sup> The BusMatrix is automatically active when at least one master is ON (CPU, DMA).

<sup>2.</sup> The GPIOx (x= A...H) dynamic current consumption is approximately divided by a factor two versus this table values when the GPIO port is locked thanks to LCKK and LCKy bits in the GPIOx\_LCKR register. In order to save the full GPIOx current consumption, the GPIOx clock should be disabled in the RCC when all port I/Os are used in alternate function or analog mode (clock is only required to read or write into GPIO registers, and is not used in AF or analog modes).

<sup>3.</sup> The AHB to APB1 Bridge is automatically active when at least one peripheral is ON on the APB1.

<sup>4.</sup> The AHB to APB2 Bridge is automatically active when at least one peripheral is ON on the APB2.

Table 38. Low-power mode wakeup timings<sup>(1)</sup> (continued)

Symbol	Parameter		Conditions	Тур	Max	Unit	
		Range 1	Wakeup clock MSI = 48 MHz	3.8	5.7		
	Wake up time from Stop 0	Range	Wakeup clock HSI16 = 16 MHz	4.1	6.9		
	mode to Run mode in Flash		Wakeup clock MSI = 24 MHz	4.07	6.2		
		Range 2	Wakeup clock HSI16 = 16 MHz	4.1	6.8		
<b>+</b>			Wakeup clock MSI = 4 MHz	8.45	11.8	110	
t <sub>WUSTOP0</sub>		Range 1	Wakeup clock MSI = 48 MHz	1.5	2.9	μs	
	Wake up time from Stop 0	Range	Wakeup clock HSI16 = 16 MHz	2.4	2.76		
	mode to Run mode in		Wakeup clock MSI = 24 MHz	2.4	3.48		
	SRAM1	Range 2	Wakeup clock HSI16 = 16 MHz	2.4	2.76	6	
			Wakeup clock MSI = 4 MHz	8.16	10.94		
	Wake up time from Stop 1 mode to Run in Flash	Range 1	Wakeup clock MSI = 48 MHz	6.34	7.86		
		range i	Wakeup clock HSI16 = 16 MHz	6.84	8.23		
		Range 2	Wakeup clock MSI = 24 MHz	6.74	8.1		
			Wakeup clock HSI16 = 16 MHz	6.89	8.21		
			Wakeup clock MSI = 4 MHz	10.47	12.1		
		Range 1	Wakeup clock MSI = 48 MHz	4.7	5.97		
	Wake up time from Stop 1	Trange i	Wakeup clock HSI16 = 16 MHz	5.9	6.92		
t <sub>WUSTOP1</sub>	mode to Run mode in		Wakeup clock MSI = 24 MHz	5.4	6.51	μs	
	SRAM1	Range 2	Wakeup clock HSI16 = 16 MHz	5.9	6.92		
			Wakeup clock MSI = 4 MHz	11.1	12.2		
	Wake up time from Stop 1 mode to Low-power run mode in Flash	Regulator in low-power	Makeup alaak MSI = 2 MI I-	16.4	17.73		
	Wake up time from Stop 1 mode to Low-power run mode in SRAM1	mode (LPR=1 in PWR_CR1)	Wakeup clock MSI = 2 MHz	17.3	18.82		

Table 38. Low-power mode wakeup timings<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions			Max	Unit	
		Dongs 1	Wakeup clock MSI = 48 MHz	8.02	9.24		
	Wake up time from Stop 2	Range 1	Wakeup clock HSI16 = 16 MHz	7.66	8.95		
twustop2	mode to Run mode in		Wakeup clock MSI = 24 MHz	8.5	9.54		
	Flash	Range 2	Wakeup clock HSI16 = 16 MHz	7.75	8.95		
			Wakeup clock MSI = 4 MHz	12.06	13.16		
	Wake up time from Stop 2 mode to Run mode in SRAM1	Dance 4	Wakeup clock MSI = 48 MHz	5.45	6.79	μs	
		Range 1	Wakeup clock HSI16 = 16 MHz	6.9	7.98		
		Range 2	Wakeup clock MSI = 24 MHz	6.3 7.36			
			Wakeup clock HSI16 = 16 MHz	6.9	7.9		
			Wakeup clock MSI = 4 MHz	13.1	13.31		
	Wakeup time from Standby	Dance 4	Wakeup clock MSI = 8 MHz	12.2	18.35		
twustby	mode to Run mode	Range 1	Wakeup clock MSI = 4 MHz	19.14	25.8	μs	
t <sub>WUSTBY</sub>	Wakeup time from Standby	Dance 4	Wakeup clock MSI = 8 MHz	12.1	18.3		
SRAM2	with SRAM2 to Run mode	Range 1	Wakeup clock MSI = 4 MHz	19.2	25.87	μs	
twushdn	Wakeup time from Shutdown mode to Run mode	Range 1	Wakeup clock MSI = 4 MHz	261.5	315.7	μs	

<sup>1.</sup> Guaranteed by characterization results.

Table 39. Regulator modes transition times<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>WULPRUN</sub>	Wakeup time from Low-power run mode to Run mode <sup>(2)</sup>	Code run with MSI 2 MHz	5	7	
t <sub>VOST</sub>	Regulator transition time from Range 2 to Range 1 or Range 1 to Range 2 <sup>(3)</sup>	Code run with MSI 24 MHz	20	40	μs

- 1. Guaranteed by characterization results.
- 2. Time until REGLPF flag is cleared in PWR\_SR2.
- 3. Time until VOSF flag is cleared in PWR\_SR2.

Table 40. Wakeup time using USART/LPUART<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
•	Wakeup time needed to calculate the	Stop mode 0	-	1.7	
t <sub>WUUSART</sub> t <sub>WULPUART</sub>	maximum USART/LPUART baudrate allowing to wakeup up from stop mode when USART/LPUART clock source is HSI	Stop mode 1/2	1	8.5	μs

<sup>1.</sup> Guaranteed by design.

#### 6.3.7 **External clock source characteristics**

#### High-speed external user clock generated from an external source

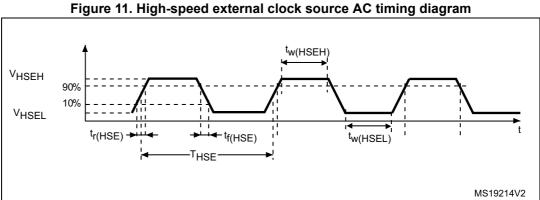
In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in Section 6.3.14. However, the recommended clock input waveform is shown in Figure 11: High-speed external clock source AC timing diagram.

Table 41. High-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	User external clock source frequency	Voltage scaling Range 1	-	8	48	MHz
f <sub>HSE_ext</sub>	Osci external clock source frequency	Voltage scaling Range 2	-	8	26	IVIITZ
V <sub>HSEH</sub>	CK_IN input pin high level voltage	-	0.7 V <sub>DDIOx</sub>	-	$V_{DDIOx}$	V
V <sub>HSEL</sub>	CK_IN input pin low level voltage	-	V <sub>SS</sub>	-	0.3 V <sub>DDIOx</sub>	
t <sub>w(HSEH)</sub>	CK_IN high or low time	Voltage scaling Range 1	7	-	-	no
t <sub>w(HSEL)</sub>		Voltage scaling Range 2	18	-	-	ns

<sup>1.</sup> Guaranteed by design.



#### Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

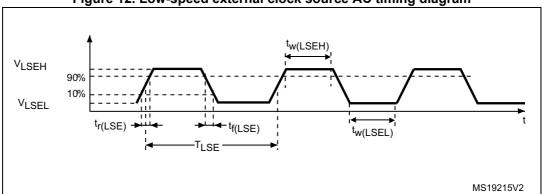
The external clock signal has to respect the I/O characteristics in *Section 6.3.14*. However, the recommended clock input waveform is shown in *Figure 12*.

Table 42. Low-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User external clock source frequency	-	-	32.768	1000	kHz
$V_{LSEH}$	OSC32_IN input pin high level voltage	-	0.7 V <sub>DDIOx</sub>	-	$V_{DDIOx}$	V
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage	-	$V_{SS}$	-	0.3 V <sub>DDIOx</sub>	
$t_{w(LSEH)}$ $t_{w(LSEL)}$	OSC32_IN high or low time	-	250	-	-	ns

<sup>1.</sup> Guaranteed by design.

Figure 12. Low-speed external clock source AC timing diagram



#### Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 43*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
		LSEDRV[1:0] = 00 Low drive capability	-	250	-	
I <sub>DD(LSE)</sub>	LSE current consumption	LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	nA
	LSE current consumption	LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	I IIA
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
	Maximum critical crystal gm	LSEDRV[1:0] = 00 Low drive capability	-	-	0.5	
Gm		LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	μΑ/V
Gm <sub>critmax</sub>		LSEDRV[1:0] = 10 Medium high drive capability	-	-	1.7	μΑνν
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
t <sub>SU(LSE)</sub> (3)	Startup time	V <sub>DD</sub> is stabilized	-	2	-	s

Table 43. LSE oscillator characteristics  $(f_{LSE} = 32.768 \text{ kHz})^{(1)}$ 

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

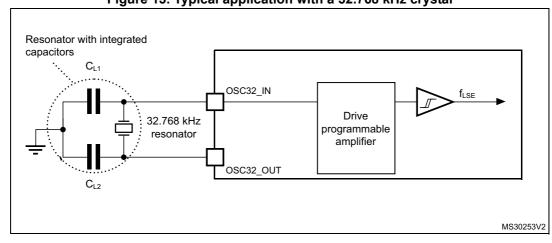


Figure 13. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.

<sup>1.</sup> Guaranteed by design.

Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".

<sup>3.</sup>  $t_{SU(LSE)}$  is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

## 6.3.8 Internal clock source characteristics

The parameters given in *Table 44* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*. The provided curves are characterization results, not tested in production.

#### High-speed internal (HSI16) RC oscillator

Table 44. HSI16 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI16</sub>	HSI16 Frequency	V <sub>DD</sub> =3.0 V, T <sub>A</sub> =30 °C	15.88	-	16.08	MHz
TRIM	HSI16 user trimming step	Trimming code is not a multiple of 64	0.2	0.3	0.4	%
	riorro user tilillilling step	Trimming code is a multiple of 64	-4	-6	-8	70
DuCy(HSI16) <sup>(2)</sup>	Duty Cycle	-	45	-	55	%
$\Delta_{Temp}(HSI16)$	HSI16 oscillator frequency drift over temperature	T <sub>A</sub> = 0 to 85 °C	-1	-	1	%
		T <sub>A</sub> = -40 to 125 °C	-2	-	1.5	%
Δ <sub>VDD</sub> (HSI16)	HSI16 oscillator frequency drift over V <sub>DD</sub>	V <sub>DD</sub> =1.62 V to 3.6 V	-0.1	-	0.05	%
t <sub>su</sub> (HSI16) <sup>(2)</sup>	HSI16 oscillator start-up time	-	-	0.8	1.2	μs
t <sub>stab</sub> (HSI16) <sup>(2)</sup>	HSI16 oscillator stabilization time	-	-	3	5	μs
I <sub>DD</sub> (HSI16) <sup>(2)</sup>	HSI16 oscillator power consumption	-	-	155	190	μΑ

<sup>1.</sup> Guaranteed by characterization results.



<sup>2.</sup> Guaranteed by design.

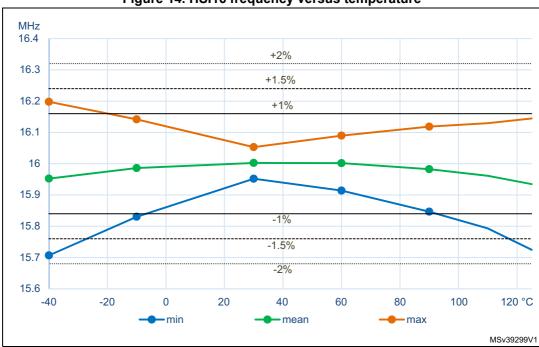


Figure 14. HSI16 frequency versus temperature

## Multi-speed internal (MSI) RC oscillator

Table 45. MSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter		Conditions	Min	Тур	Max	Unit
			Range 0	98.7	100	101.3	
			Range 1	197.4	200	202.6	kHz
			Range 2	394.8	400	405.2	NI IZ
			Range 3	789.6	800	810.4	
f			Range 4	0.987	1	1.013	
		MSI mode	Range 5	1.974	2	2.026	
		WiSi mode	Range 6	3.948	4	4.052	
			Range 7	7.896	8	8.104	MHz
			Range 8	15.79	16	16.21	1011 12
	MSI frequency after factory calibration, done at V <sub>DD</sub> =3 V and T <sub>A</sub> =30 °C		Range 9	23.69	24	24.31	
			Range 10	31.58	32	32.42	
			Range 11	47.38	48	48.62	
f <sub>MSI</sub>			Range 0	-	98.304	-	- kHz
			Range 1	-	196.608	-	
			Range 2	-	393.216	-	
			Range 3	-	786.432	-	
			Range 4	-	1.016	-	
		PLL mode XTAL=	Range 5	-	1.999	-	
		32.768 kHz	Range 6	-	3.998	-	
			Range 7	-	7.995	-	MHz
			Range 8	-	15.991	-	IVIITZ
			Range 9	-	23.986	-	
			Range 10	-	32.014	-	
			Range 11	-	48.005	-	
(1401)(2)	MSI oscillator	MOL	T <sub>A</sub> = -0 to 85 °C	-3.5	-	3	0/
$\Delta_{TEMP}(MSI)^{(2)}$	frequency drift over temperature	MSI mode	T <sub>A</sub> = -40 to 125 °C	-8	-	6	%

Table 45. MSI oscillator characteristics<sup>(1)</sup> (continued)

Symbol	Parameter		Conditions	(0000000	Min	Тур	Max	Unit
			D 0 1 . 0	V <sub>DD</sub> =1.62 V to 3.6 V	-1.2	-	0.5	
			Range 0 to 3	V <sub>DD</sub> =2.4 V to 3.6 V	-0.5	-	0.5	
A (MGI)(2)	MSI oscillator frequency drift	MSI mode	Dange 4 to 7	V <sub>DD</sub> =1.62 V to 3.6 V	-2.5	-	0.7	%
$\Delta_{\text{VDD}}(\text{MSI})^{(2)}$	over V <sub>DD</sub> (reference is 3 V)		Range 4 to 7	V <sub>DD</sub> =2.4 V to 3.6 V	-0.8	-	0.7	%
				V <sub>DD</sub> =1.62 V to 3.6 V	-5	-	1	
			Kange o to 11	V <sub>DD</sub> =2.4 V to 3.6 V	-1.6	-	1	
AFCAMPLING	Frequency		$T_A = -40 \text{ to } 85^\circ$	°C	-	1	2	
$\Delta F_{SAMPLING} \ (MSI)^{(2)(6)}$	variation in sampling mode <sup>(3)</sup>	MSI mode	T <sub>A</sub> = -40 to 125 °C - 2		4	%		
P_USB Jitter(MSI) <sup>(6)</sup>	Period jitter for USB clock <sup>(4)</sup>	eriod jitter for PLL mode	for next transition	-	-	-	3.458	
		Range 11	for paired transition	-	-	-	3.916	ns
MT_USB	Medium term jitter for USB clock <sup>(5)</sup>		for next transition	-	-	-	2	no
Jitter(MSI) <sup>(6)</sup>		Range 11	for paired transition	-	-	-	1	ns
CC jitter(MSI) <sup>(6)</sup>	RMS cycle-to- cycle jitter	PLL mode R	lange 11	-	-	60	-	ps
P jitter(MSI) <sup>(6)</sup>	RMS Period jitter	PLL mode R	ange 11	-	-	50	-	ps
		Range 0		-	-	10	20	
		Range 1		-	-	5	10	
(1.401)(6)	MSI oscillator	Range 2		-	-	4	8	
t <sub>SU</sub> (MSI) <sup>(6)</sup>	start-up time	Range 3		-	-	3	7	us
		Range 4 to 7	7	-	-	3	6	
		Range 8 to 11 -	-	-	2.5	6		
			10 % of final frequency	-	-	0.25	0.5	
t <sub>STAB</sub> (MSI) <sup>(6)</sup>	MSI oscillator stabilization time	PLL mode Range 11	5 % of final frequency	-	-	0.5	1.25	ms
		_		-	-	-	2.5	



Symbol	Parameter		Conditions		Min	Тур	Max	Unit
			Range 0	-	-	0.6	1	
			Range 1	-	-	0.8	1.2	
		Range 2	-	-	1.2	1.7		
		Range 3	-	-	1.9	2.5		
		Range 4	-	-	4.7	6		
L (MCI)(6)	MSI oscillator power consumption	MSI and	Range 5	-	-	6.5	9	
IDD(INIQI)		PLL mode	Range 6	-	-	11	15	μA
·			Range 7	-	-	18.5	25	
			Range 8	-	-	62	80	1
			Range 9	-	-	85	110	
			Range 10	-	-	110	130	
		Range 11	-	-	155	190		

Table 45. MSI oscillator characteristics<sup>(1)</sup> (continued)

6. Guaranteed by design.

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> This is a deviation for an individual part once the initial frequency has been measured.

<sup>3.</sup> Sampling mode means Low-power run/Low-power sleep modes with Temperature sensor disable.

Average period of MSI @48 MHz is compared to a real 48 MHz clock over 28 cycles. It includes frequency tolerance + jitter of MSI @48 MHz clock.

<sup>5.</sup> Only accumulated jitter of MSI @48 MHz is extracted over 28 cycles. For next transition: min. and max. jitter of 2 consecutive frame of 28 cycles of the MSI @48 MHz, for 1000 captures over 28 cycles. For paired transitions: min. and max. jitter of 2 consecutive frame of 56 cycles of the MSI @48 MHz, for 1000 captures over 56 cycles.

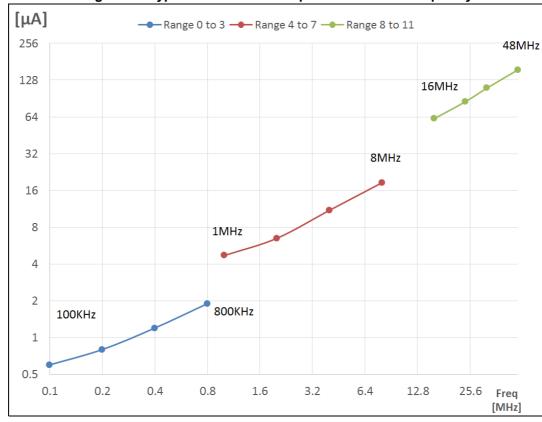


Figure 15. Typical current consumption versus MSI frequency

High-speed internal 48 MHz (HSI48) RC oscillator

Table 46. HSI48 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI48</sub>	HSI48 Frequency	V <sub>DD</sub> =3.0V, T <sub>A</sub> =30°C	-	48	-	MHz
TRIM	HSI48 user trimming step	-	-	0.11 <sup>(2)</sup>	0.18 <sup>(2)</sup>	%
USER TRIM COVERAGE	HSI48 user trimming coverage	±32 steps	±3 <sup>(3)</sup>	±3.5 <sup>(3)</sup>	-	%
DuCy(HSI48)	Duty Cycle	-	45 <sup>(2)</sup>	-	55 <sup>(2)</sup>	%
ACC	Accuracy of the HSI48 oscillator	V <sub>DD</sub> = 3.0 V to 3.6 V, T <sub>A</sub> = -15 to 85 °C	-	-	±3 <sup>(3)</sup>	%
ACC <sub>HSI48_REL</sub>	over temperature (factory calibrated)	V <sub>DD</sub> = 1.65 V to 3.6 V, T <sub>A</sub> = -40 to 125 °C	-	-	±4.5 <sup>(3)</sup>	70
D (HCI40)	HSI48 oscillator frequency drift	V <sub>DD</sub> = 3 V to 3.6 V	-	0.025 <sup>(3)</sup>	0.05 <sup>(3)</sup>	%
D <sub>VDD</sub> (HSI48)	with V <sub>DD</sub>	V <sub>DD</sub> = 1.65 V to 3.6 V	-	0.05 <sup>(3)</sup>	0.1 <sup>(3)</sup>	70
t <sub>su</sub> (HSI48)	HSI48 oscillator start-up time	-	-	2.5 <sup>(2)</sup>	6 <sup>(2)</sup>	μs
I <sub>DD</sub> (HSI48)	HSI48 oscillator power consumption	-	-	340 <sup>(2)</sup>	380 <sup>(2)</sup>	μA

Table 46. HSI48 oscillator	characteristics(1)	(continued)
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Symbol	Parameter	Conditions	Min	Тур	Max	Unit
N <sub>T</sub> jitter	Next transition jitter Accumulated jitter on 28 cycles <sup>(4)</sup>	-	-	+/-0.15 <sup>(2)</sup>	-	ns
P <sub>T</sub> jitter	Paired transition jitter Accumulated jitter on 56 cycles <sup>(4)</sup>	-	-	+/-0.25 <sup>(2)</sup>	-	ns

- 1.  $V_{DD}$  = 3 V,  $T_A$  = -40 to 125°C unless otherwise specified.
- 2. Guaranteed by design.
- 3. Guaranteed by characterization results.
- 4. Jitter measurement are performed without clock source activated in parallel.

Figure 16. HSI48 frequency versus temperature 6 4 2 0 -2 -6 -50 -30 70 -10 10 30 50 90 110 130 °C --- Avg --- min - max MSv40989V1

Low-speed internal (LSI) RC oscillator

Table 47. LSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	I SI Fraguency	V <sub>DD</sub> = 3.0 V, T <sub>A</sub> = 30 °C	31.04	-	32.96	kHz
f <sub>LSI</sub>	LSI Frequency	$V_{DD}$ = 1.62 to 3.6 V, TA = -40 to 125 °C	29.5	-	34	KI IZ
t <sub>SU</sub> (LSI) <sup>(2)</sup>	LSI oscillator start- up time	-	1	80	130	μs
t <sub>STAB</sub> (LSI) <sup>(2)</sup>	LSI oscillator stabilization time	5% of final frequency	1	125	180	μs
I <sub>DD</sub> (LSI) <sup>(2)</sup>	LSI oscillator power consumption	-	-	110	180	nA

- 1. Guaranteed by characterization results.
- 2. Guaranteed by design.

#### 6.3.9 PLL characteristics

The parameters given in *Table 48* are derived from tests performed under temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 20: General operating conditions*.

Table 48. PLL, PLLSAI1 characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
f.	PLL input clock <sup>(2)</sup>	-	4	-	16	MHz	
f <sub>PLL_IN</sub>	PLL input clock duty cycle	-	45	-	55	%	
£	DLL multiplier output pleak D	Voltage scaling Range 1	3.0968	-	80	MHz	
f <sub>PLL_P_OUT</sub>	PLL multiplier output clock P	Voltage scaling Range 2	3.0968	-	26	IVITZ	
f	DLL multiplior output clock O	Voltage scaling Range 1	12	-	80	MHz	
f <sub>PLL_Q_OUT</sub>	PLL multiplier output clock Q	Voltage scaling Range 2	12	-	26	IVIITZ	
f	PLL multiplier output clock R	Voltage scaling Range 1	12	-	80	MHz	
f <sub>PLL_R_OUT</sub>		Voltage scaling Range 2	12	-	26	IVIITZ	
f	DLL VCC output	Voltage scaling Range 1	96	-	344	MHz	
f <sub>VCO_OUT</sub>	PLL VCO output	Voltage scaling Range 2	96	-	128	IVITZ	
t <sub>LOCK</sub>	PLL lock time	-	-	15	40	μs	
littor	RMS cycle-to-cycle jitter	System sleek 90 MHz	-	40	-	±00	
Jitter	RMS period jitter	- System clock 80 MHz	-	30	-	±ps	
I <sub>DD</sub> (PLL)	PLL power consumption on $V_{DD}^{(1)}$	VCO freq = 96 MHz	-	200	260		
		VCO freq = 192 MHz	-	300	380	μΑ	
	טט	VCO freq = 344 MHz	-	520	650	1	

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between the 2 PLLs.

## 6.3.10 Flash memory characteristics

Table 49. Flash memory characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit	
t <sub>prog</sub>	64-bit programming time	-	81.69	90.76	μs	
	one row (32 double	normal programming	2.61	2.90		
t <sub>prog_row</sub>	word) programming time	fast programming	1.91	2.12		
+	one page (2 Kbyte)	normal programming	20.91	23.24	ms	
t <sub>prog_page</sub>	programming time	fast programming	15.29	16.98		
t <sub>ERASE</sub>	Page (2 KB) erase time	-	22.02	24.47		
+	one bank (512 Kbyte) programming time	normal programming	5.35	5.95	s	
t <sub>prog_bank</sub>		fast programming	3.91	4.35	5	
t <sub>ME</sub>	Mass erase time (one or two banks)	-	22.13	24.59	ms	
	Average consumption	Write mode	3.4	-		
	from V <sub>DD</sub>	Erase mode	3.4	-		
I <sub>DD</sub>	Maximum ourrant (noak)	Write mode	7 (for 2 μs)	-	mA	
	Maximum current (peak)	Erase mode	7 (for 41 μs)	-		

<sup>1.</sup> Guaranteed by design.

Table 50. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Unit
N <sub>END</sub>	Endurance	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	10	kcycles
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30	
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	15	
	Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 125 °C	7	Years
t <sub>RET</sub>		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 55 °C	30	Tears
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 85 °C	15	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 105 °C	10	

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Cycling performed over the whole temperature range.

#### 6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 51*. They are based on the EMS levels and classes defined in application note AN1709.

Level/ **Symbol Parameter Conditions** Class  $V_{DD} = 3.3 \text{ V}, T_A = +25 ^{\circ}\text{C},$ Voltage limits to be applied on any I/O pin  $f_{HCLK} = 80 \text{ MHz}.$ 2B  $V_{FESD}$ to induce a functional disturbance conforming to IEC 61000-4-2 Fast transient voltage burst limits to be  $V_{DD} = 3.3 \text{ V}, T_A = +25 ^{\circ}\text{C},$  $f_{HCLK} = 80 MHz$ ,  $\mathsf{V}_{\mathsf{EFTB}}$ applied through 100 pF on V<sub>DD</sub> and V<sub>SS</sub> 5A pins to induce a functional disturbance conforming to IEC 61000-4-4

Table 51. EMS characteristics

#### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

#### Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

#### **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f <sub>HSE</sub> /f <sub>HCLK</sub> ]	Unit
			moquonoy bana	8 MHz/ 80 MHz	
		V <sub>DD</sub> = 3.6 V, T <sub>A</sub> = 25 °C, UFQFPN32 package compliant with IEC	0.1 MHz to 30 MHz	1	
			30 MHz to 130 MHz	0	dBµV
S <sub>EMI</sub>	Peak level		130 MHz to 1 GHz	-1	αБμν
		61967-2	1 GHz to 2 GHz	7	
			EMI Level	1	-

Table 52. EMI characteristics

#### 6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	T <sub>A</sub> = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001	2	2000	V
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C, conforming to ANSI/ESD STM5.3.1	C3	250	V

Table 53. ESD absolute maximum ratings



<sup>1.</sup> Guaranteed by characterization results.

#### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 54. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +105 °C conforming to JESD78A	П

#### 6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DDIOX}$  (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the -5  $\mu$ A/+0  $\mu$ A range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in *Table 55*.

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

Table 55. I/O current injection susceptibility<sup>(1)</sup>

Symbol	Description		tional ptibility	Unit
Symbol	Description	Negative injection	Positive injection	Oilit
l	Injected current on all pins except PA4, PA5	-5	NA	mA
INJ	Injected current on PA4, PA5 pins	-5	0	111/4

1. Guaranteed by characterization results.



## 6.3.14 I/O port characteristics

### General input/output characteristics

Unless otherwise specified, the parameters given in *Table 56* are derived from tests performed under the conditions summarized in *Table 20: General operating conditions*. All I/Os are designed as CMOS- and TTL-compliant.

Table 56. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
V <sub>IL</sub> <sup>(1)</sup>	I/O input low level voltage	1.62 V <v<sub>DDIOx&lt;3.6 V</v<sub>	-	-	0.3xV <sub>DDIOx</sub> (2)		
	I/O input low level voltage	1.62 V <v<sub>DDIOx&lt;3.6 V</v<sub>	-	-	0.39xV <sub>DDIOx</sub> -0.06 <sup>(3)</sup>	V	
	I/O input low level voltage	1.08 V <v<sub>DDIOx&lt;1.62 V</v<sub>	-	ı	0.43xV <sub>DDIOx</sub> -0.1 <sup>(3)</sup>		
	I/O input high level voltage	1.62 V <v<sub>DDIOx&lt;3.6 V</v<sub>	0.7xV <sub>DDIOx</sub> <sup>(2)</sup>	-	-		
V <sub>IH</sub> <sup>(1)</sup>	I/O input high level voltage	1.62 V <v<sub>DDIOx&lt;3.6 V</v<sub>	0.49xV <sub>DDIOX</sub> +0.26 <sup>(3)</sup>	-	-	V	
	I/O input high level voltage	1.08 V <v<sub>DDIOX&lt;1.62 V</v<sub>	0.61xV <sub>DDIOX</sub> +0.05 <sup>(3)</sup>	3)			
V <sub>hys</sub> <sup>(3)</sup>	TT_xx, FT_xxx and NRST I/O input hysteresis	1.62 V <v<sub>DDIOx&lt;3.6 V - 200</v<sub>		-	mV		
	FT_sx	_sx 1.08 V <v<sub>DDIOx&lt;1.62 V - 150 -</v<sub>		-			
	FT_xx input leakage current <sup>(3)</sup>	$V_{IN} \le Max(V_{DDXXX})^{(4)}$	-	ı	±100		
		$ \begin{aligned} & Max(V_{DDXXX}) \leq V_{IN} \leq \\ & Max(V_{DDXXX}) + 1 \ V^{(4)(5)} \end{aligned} $	-	ı	650 <sup>(3)(6)</sup>		
		Max( $V_{DDXXX}$ )+1 V < VIN ≤ 5.5 V <sup>(3)(5)</sup>		-	200 <sup>(6)</sup>		
	FT_u and PC3 IO	$V_{IN} \le Max(V_{DDXXX})^{(4)}$	-	ı	±150		
I <sub>lkg</sub>		$Max(V_{DDXXX}) \le V_{IN} \le Max(V_{DDXXX})+1 V^{(4)}$	-	-	2500 <sup>(3)(7)</sup>	nA	
		$Max(V_{DDXXX})+1 V < VIN \le 5.5 V^{(4)(5)(7)}$	-	-	250 <sup>(7)</sup>		
	TT_xx input leakage current	$V_{IN} \le Max(V_{DDXXX})^{(6)}$	-	-	±150		
		$ \text{Max}(V_{\text{DDXXX}}) \le V_{\text{IN}} < $ $ 3.6 \text{ V}^{(6)} $	-	-	2000 <sup>(3)</sup>		
R <sub>PU</sub>	Weak pull-up equivalent resistor (8)	V <sub>IN</sub> = V <sub>SS</sub>	25	40	55	kΩ	
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(8)</sup>	V <sub>IN</sub> = V <sub>DDIOx</sub>	25	40	55	kΩ	
C <sub>IO</sub>	I/O pin capacitance	-	-	5	-	pF	

<sup>1.</sup> Refer to Figure 17: I/O input characteristics.

- 2. Tested in production.
- 3. Guaranteed by design.
- 4. Max(V<sub>DDXXX</sub>) is the maximum value of all the I/O supplies. Refer to Table: Legend/Abbreviations used in the pinout table.
- 5. All TX\_xx IO except FT\_u and PC3.
- This value represents the pad leakage of the IO itself. The total product pad leakage is provided by this formula: I<sub>Total\_Ileak\_max</sub> = 10 μA + [number of IOs where V<sub>IN</sub> is applied on the pad] x I<sub>Ikg</sub>(Max).
- To sustain a voltage higher than MIN(V<sub>DD</sub>, V<sub>DDA</sub>, V<sub>DDUSB</sub>) +0.3 V, the internal Pull-up and Pull-Down resistors must be disabled.
- 8. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 17* for standard I/Os, and in *Figure 17* for 5 V tolerant I/Os.

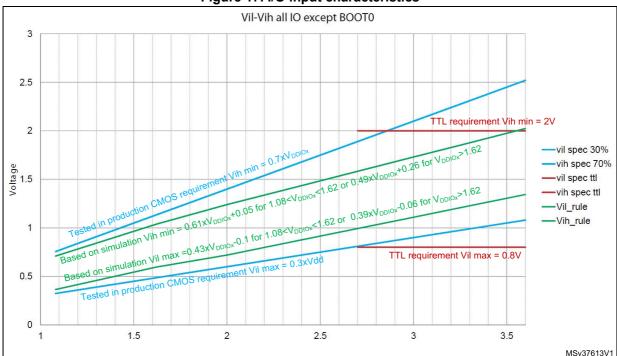


Figure 17. I/O input characteristics

#### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 20$  mA (with a relaxed  $V_{OL}/V_{OH}$ ).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2:

- The sum of the currents sourced by all the I/Os on V<sub>DDIOX</sub>, plus the maximum consumption of the MCU sourced on V<sub>DD</sub>, cannot exceed the absolute maximum rating ΣI<sub>VDD</sub> (see *Table 17: Voltage characteristics*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub>, plus the maximum consumption of the MCU sunk on V<sub>SS</sub>, cannot exceed the absolute maximum rating ΣI<sub>VSS</sub> (see Table 17: Voltage characteristics).

#### **Output voltage levels**

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*. All I/Os are CMOS- and TTL-compliant (FT OR TT unless otherwise specified).

Table 57. Output voltage characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>OL</sub>	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup>	-	0.4	
V <sub>OH</sub>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 8 mA  V <sub>DDIOx</sub> ≥ 2.7 V	V <sub>DDIOx</sub> -0.4	-	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	TTL port <sup>(2)</sup>	-	0.4	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 8 mA  V <sub>DDIOx</sub> ≥ 2.7 V	2.4	-	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 20 mA	-	1.3	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	V <sub>DDIOx</sub> ≥ 2.7 V	V <sub>DDIOx</sub> -1.3	-	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 4 mA	-	0.45	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	V <sub>DDIOx</sub> ≥ 1.62 V	V <sub>DDIOx</sub> -0.45	-	V
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 2 mA	-	$0.35_xV_{DDIOx}$	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	1.62 V ≥ V <sub>DDIOx</sub> ≥ 1.08 V	$0.65_xV_{DDIOx}$	-	
	Output low level voltage for an FT I/O pin in FM+ mode (FT I/O with "f" option)	I <sub>IO</sub>   = 20 mA V <sub>DDIOx</sub> ≥ 2.7 V	-	0.4	
V <sub>OLFM+</sub>		I <sub>IO</sub>   = 10 mA V <sub>DDIOx</sub> ≥ 1.62 V	-	0.4	
	. ,	$ I_{IO}  = 2 \text{ mA}$ 1.62 V $\ge$ V <sub>DDIOx</sub> $\ge$ 1.08 V	-	0.4	

The I<sub>IO</sub> current sourced or sunk by the device must always respect the absolute maximum rating specified in Table 17:
 Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI<sub>IO</sub>.

#### Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 18* and *Table 58*, respectively.



<sup>2.</sup> TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

<sup>3.</sup> Guaranteed by design.

Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*.

Table 58. I/O AC characteristics<sup>(1)(2)</sup>

Speed	Symbol	Parameter	Conditions	Min	Max	Unit			
	Fmax	Maximum frequency	C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	5				
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	- 1					
			C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	0.1	MHz			
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	10	IVITZ			
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	1.5				
00			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	0.1				
00			C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	25				
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	52				
	Tr/Tf	Output rise and fall time	C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	140	no			
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	17	ns			
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	37				
			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	110				
	Fmax	nax Maximum frequency	C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	25	- MHz			
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	10				
			C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	1				
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	50				
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	15				
01			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	1				
01		Tr/Tf Output rise and fall time	C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	9				
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	16				
	Tr/Tf		C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	40				
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	4.5	ns			
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	9				
			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	21				

Table 58. I/O AC characteristics<sup>(1)(2)</sup> (continued)

Speed	Symbol	Parameter	Conditions	Min	Max	Unit			
		Maximum frequency	C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	50	MHz			
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	25				
	Fmax		C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	5				
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V - 1					
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	37.5	37.5 5			
10			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	5				
10	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	5.8				
			C=50 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	11				
			C=50 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	28				
			C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	2.5	ns			
			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	5	1			
			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	12				
	Fmax		C=30 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	120 <sup>(3)</sup>	. NALL-			
			C=30 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	50				
		Maximum fraguancy	C=30 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	10				
		rmax	Maximum frequency	C=10 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	180 <sup>(3)</sup>	MHz		
11			C=10 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	75				
			C=10 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	10				
	Tr/Tf	Tf Output rise and fall time	C=30 pF, 2.7 V≤V <sub>DDIOx</sub> ≤3.6 V	-	- 3.3				
			C=30 pF, 1.62 V≤V <sub>DDIOx</sub> ≤2.7 V	-	6	ns			
			C=30 pF, 1.08 V≤V <sub>DDIOx</sub> ≤1.62 V	-	16				
Emi	Fmax	Maximum frequency	C-50 pE 16 \/<\/>	-	1	MHz			
Fm+	Tf	Output fall time <sup>(4)</sup>	C=50 pF, 1.6 V≤V <sub>DDIOx</sub> ≤3.6 V	-	5	ns			

The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG\_CFGR1 register. Refer to the RM0393 reference manual for a description of GPIO Port configuration register.

<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> This value represents the I/O capability but the maximum system frequency is limited to 80 MHz.

<sup>4.</sup> The fall time is defined between 70% and 30% of the output waveform accordingly to  $I^2C$  specification.

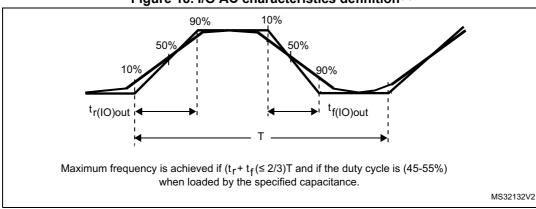


Figure 18. I/O AC characteristics definition<sup>(1)</sup>

1. Refer to Table 58: I/O AC characteristics.

## 6.3.15 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor,  $R_{\text{PU}}$ .

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 20: General operating conditions*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
V <sub>IL(NRST)</sub>	NRST input low level voltage	-	-	-	0.3 <sub>x</sub> V <sub>DDIOx</sub>	V	
V <sub>IH(NRST)</sub>	NRST input high level voltage	-	0.7 <sub>x</sub> V <sub>DDIOx</sub>	-	-	]	
V <sub>hys(NRST)</sub>	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV	
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(2)</sup>	V <sub>IN</sub> = V <sub>SS</sub>	25	40	55	kΩ	
V <sub>F(NRST)</sub>	NRST input filtered pulse	-	-	-	70	ns	
V <sub>NF(NRST)</sub>	NRST input not filtered pulse	1.71 V ≤ V <sub>DD</sub> ≤ 3.6 V	350	-	-	ns	

Table 59. NRST pin characteristics<sup>(1)</sup>

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).

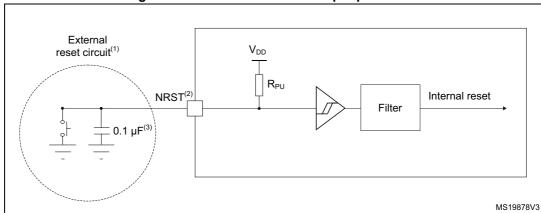


Figure 19. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- 2. The user must ensure that the level on the NRST pin can go below the V<sub>IL(NRST)</sub> max level specified in *Table 59: NRST pin characteristics*. Otherwise the reset will not be taken into account by the device.
- 3. The external capacitor on NRST must be placed as close as possible to the device.

## 6.3.16 Analog switches booster

Table 60. Analog switches booster characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Тур	Max	Unit	
$V_{DD}$	Supply voltage	1.62	-	3.6	V	
t <sub>SU(BOOST)</sub>	Booster startup time	-	-	240	μs	
I <sub>DD(BOOST)</sub>	Booster consumption for $1.62 \text{ V} \le \text{V}_{DD} \le 2.0 \text{ V}$	-	-	250	μА	
	Booster consumption for 2.0 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	-	500		
	Booster consumption for $2.7 \text{ V} \leq \text{V}_{\text{DD}} \leq 3.6 \text{ V}$	-	-	900		

1. Guaranteed by design.

## 6.3.17 Analog-to-Digital converter characteristics

Unless otherwise specified, the parameters given in *Table 61* are preliminary values derived from tests performed under ambient temperature,  $f_{PCLK}$  frequency and  $V_{DDA}$  supply voltage conditions summarized in *Table 20: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

Table 61. ADC characteristics<sup>(1)</sup> (2)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
$V_{DDA}$	Analog supply voltage	-	1.62	-	3.6	V	
\/	Docitive reference veltage	V <sub>DDA</sub> ≥ 2 V	2	-	$V_{DDA}$	V	
$V_{REF+}$	Positive reference voltage	V <sub>DDA</sub> < 2 V		$V_{DDA}$		V	
V <sub>REF-</sub>	Negative reference voltage	-		V <sub>SSA</sub>		V	
f	ADC clock frequency	Range 1	-	-	80	MHz	
f <sub>ADC</sub>	ADC clock frequency	Range 2	-	-	26	IVITZ	
		Resolution = 12 bits	-	-	5.33		
	Sampling rate for FAST	Resolution = 10 bits	-	-	6.15		
	channels	Resolution = 8 bits	-	-	7.27		
f		Resolution = 6 bits	-	-	8.88	Msps	
f <sub>s</sub>	Sampling rate for SLOW channels	Resolution = 12 bits	-	-	4.21	ivisps	
		Resolution = 10 bits	-	-	4.71		
		Resolution = 8 bits	-	-	5.33		
		Resolution = 6 bits	-	-	6.15		
$f_{TRIG}$	External trigger frequency	f <sub>ADC</sub> = 80 MHz Resolution = 12 bits	-	-	5.33	MHz	
		Resolution = 12 bits	-	-	15	1/f <sub>ADC</sub>	
V <sub>CMIN</sub>	Input common mode	Differential mode	(V <sub>REF+</sub> + V <sub>REF-</sub> )/2 - 0.18	(V <sub>REF+</sub> + V <sub>REF-</sub> )/2	(V <sub>REF+</sub> + V <sub>REF-</sub> )/2 + 0.18	V	
V <sub>AIN</sub> (3)	Conversion voltage range(2)	-	0	-	V <sub>REF+</sub>	V	
R <sub>AIN</sub>	External input impedance	-	-	-	50	kΩ	
C <sub>ADC</sub>	Internal sample and hold capacitor	-	-	5	-	pF	
t <sub>STAB</sub>	Power-up time	-		1		conversion cycle	
+	Calibration time	f <sub>ADC</sub> = 80 MHz	= 80 MHz 1.45			μs	
$t_{CAL}$	Calibration time	-		116			

Table 61. ADC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		CKMODE = 00	1.5	2	2.5	
	Trigger conversion latency Regular and	CKMODE = 01	-	-	2.0	
t <sub>LATR</sub>	injected channels without	CKMODE = 10	-	-	2.25	1/f <sub>ADC</sub>
	conversion abort	CKMODE = 11	-	-	2.125	
		CKMODE = 00	2.5	3	3.5	
	Trigger conversion latency Injected channels	CKMODE = 01	-	-	3.0	4 15
t <sub>LATRINJ</sub>	aborting a regular	CKMODE = 10	-	-	3.25	1/f <sub>ADC</sub>
	conversion	CKMODE = 11	-	-	3.125	
1	O a manalism as this as	f <sub>ADC</sub> = 80 MHz	0.03125	-	8.00625	μs
t <sub>s</sub>	Sampling time	-	- 2.5		640.5	1/f <sub>ADC</sub>
t <sub>ADCVREG_STUP</sub>	ADC voltage regulator start-up time	-			20	μs
	Total conversion time (including sampling time)	f <sub>ADC</sub> = 80 MHz Resolution = 12 bits	0.1875	-	8.1625	μs
t <sub>CONV</sub>		Resolution = 12 bits	ts + 12.5 cycles for successive approximation = 15 to 653			1/f <sub>ADC</sub>
		fs = 5 Msps	-	730	830	
I <sub>DDA</sub> (ADC)	ADC consumption from the V <sub>DDA</sub> supply	fs = 1 Msps	-	160	220	μΑ
	THE TODA CUPP.	fs = 10 ksps	-	16	50	
	ADC consumption from	fs = 5 Msps	-	130	160	
I <sub>DDV_S</sub> (ADC)	the V <sub>REF+</sub> single ended	fs = 1 Msps	-	30	40	μΑ
	mode	fs = 10 ksps	-	0.6	2	
	ADC consumption from	fs = 5 Msps	-	260	310	
I <sub>DDV_D</sub> (ADC)	the V <sub>REF+</sub> differential	fs = 1 Msps	-	60	70	μΑ
	mode	fs = 10 ksps	-	1.3	3	

<sup>1.</sup> Guaranteed by design

<sup>2.</sup> The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4V). It is disable when  $V_{DDA} \ge 2.4$  V.

<sup>3.</sup>  $V_{REF+}$  can be internally connected to  $V_{DDA}$  and  $V_{REF-}$  can be internally connected to  $V_{SSA}$ , depending on the package. Refer to Section 4: Pinouts and pin description for further details.

## Equation 1: R<sub>AIN</sub> max formula

$$R_{AIN} < \frac{T_{S}}{f_{ADC} \times C_{ADC} \times \ln(2^{N+2})} - R_{ADC}$$

The formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).

Table 62. Maximum ADC RAIN<sup>(1)(2)</sup>

D I d	Sampling cycle	Sampling time [ns]	RAIN	nax (Ω)
Resolution	@80 MHz	@80 MHz	Fast channels <sup>(3)</sup>	Slow channels <sup>(4)</sup>
	2.5	31.25	100	N/A
	6.5	81.25	330	100
	12.5	156.25	680	470
40 bits	24.5	306.25	1500	1200
12 bits	47.5	593.75	2200	1800
	92.5	1156.25	4700	3900
	247.5	3093.75	12000	10000
	640.5	8006.75	39000	33000
	2.5	31.25	120	N/A
	6.5	81.25	390	180
	12.5	156.25	820	560
40.1%	24.5	306.25	1500	1200
10 bits	47.5	593.75	2200	1800
	92.5	1156.25	5600	4700
	247.5	3093.75	12000	10000
	640.5	8006.75	47000	39000
	2.5	31.25	180	N/A
	6.5	81.25	470	270
	12.5	156.25	1000	680
8 bits	24.5	306.25	1800	1500
SJIQ Ø	47.5	593.75	2700	2200
	92.5	1156.25	6800	5600
	247.5	3093.75	15000	12000
	640.5	8006.75	50000	50000



Table 62. Maximum ADC RAIN<sup>(1)(2)</sup> (continued)

Resolution	Sampling cycle	Sampling time [ns]	RAIN max (Ω)			
Resolution	@80 MHz	@80 MHz	Fast channels <sup>(3)</sup>	Slow channels <sup>(4)</sup>		
	2.5	31.25	220	N/A		
	6.5	81.25	560	330		
	12.5	156.25	1200	1000		
6 bits	24.5	306.25	2700	2200		
O DIIS	47.5	593.75	3900	3300		
	92.5	1156.25	8200	6800		
	247.5	3093.75	18000	15000		
	640.5	8006.75	50000	50000		

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4V). It is disable when  $V_{DDA} \ge 2.4$  V.

<sup>3.</sup> Fast channels are: PC0, PC1, PC2, PC3, PA0, PA1.

<sup>4.</sup> Slow channels are: all ADC inputs except the fast channels.

Table 63. ADC accuracy - limited test conditions 1<sup>(1)(2)(3)</sup>

Sym- bol	Parameter	(	Conditions <sup>(4</sup>	)	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	4	5	
ET	Total unadjusted		ended	Slow channel (max speed)	-	4	5	
	error		Differential	Fast channel (max speed)	-	3.5	4.5	
			Billerential	Slow channel (max speed)	-	3.5	4.5	
			Single ended	Fast channel (max speed)	-	1	2.5	
EO	Offset			Slow channel (max speed)	-	1	2.5	
	error		Differential	Fast channel (max speed)	-	1.5	2.5	
			Dillerential	Slow channel (max speed)	-	1.5	2.5	
			Single	Fast channel (max speed)	-	2.5	4.5	
EG	Gain error		ended	Slow channel (max speed)	-	2.5	4.5	LSB
EG	Gain enoi		Differential	Fast channel (max speed)	-	2.5	3.5	LOD
			Dillerential	Slow channel (max speed)	-	2.5	3.5	
			Single	Fast channel (max speed)	-	1	1.5	
ED	ED Differential linearity error	ADC clock frequency ≤ 80 MHz, Sampling rate ≤ 5.33 Msps, V <sub>DDA</sub> = VREF+ = 3 V, TA = 25 °C	ended	Slow channel (max speed)	-	1	1.5	-
			Differential	Fast channel (max speed)	-	1	1.2	
			Dillerential	Slow channel (max speed)	-	1	1.2	
			Single	Fast channel (max speed)	-	1.5	2.5	
EL	Integral		ended	Slow channel (max speed)	-	1.5	2.5	1
EL	linearity error		Differential	Fast channel (max speed)	-	1	2	
			Dillerential	Slow channel (max speed)	-	1	2	
			Single	Fast channel (max speed)	10.4	10.5	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10.4	10.5	-	bits
LINOB	bits		Differential	Fast channel (max speed)	10.8	10.9	-	טונס
			Dillerential	Slow channel (max speed)	10.8	10.9	-	
	Signal to		Single	Fast channel (max speed)	64.4	65	-	
SINAD	Signal-to- noise and		ended	Slow channel (max speed)	64.4	65	-	
SINAD	distortion		Differential	Fast channel (max speed)	66.8	67.4	-	
	ratio		Dillerential	Slow channel (max speed)	66.8	67.4	-	dB
			Single	Fast channel (max speed)	65	66	-	ub
SNR	Signal-to-		ended	Slow channel (max speed)	65	66	-	
SINK	noise ratio		Differential	Fast channel (max speed)	67	68	-	-
			Dillerential	Slow channel (max speed)	67	68	1	



Table 63. ADC accuracy - limited test conditions  $1^{(1)(2)(3)}$  (continued)

Sym- bol	Parameter	C	Conditions <sup>(4)</sup>					
		ADC clock frequency ≤	Single	Fast channel (max speed)	-	-74	-73	
THD	Total harmonic		ended	Slow channel (max speed)	-	-74	-73	dB
distortion	stortion \( \stortion \)	Differential	Fast channel (max speed)	-	-79	-76	uВ	
		TA = 25 °C	Dillerential	Slow channel (max speed)	-	-79	-76	

- 1. Guaranteed by design.
- 2. ADC DC accuracy values are measured after internal calibration.
- 3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4 V). It is disable when  $V_{DDA} \ge 2.4$  V. No oversampling.

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Table 64. ADC accuracy - limited test conditions  $2^{(1)(2)(3)}$ 

Sym- bol	Parameter		Conditions <sup>(4</sup>	)	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	4	6.5	
	Total		ended	Slow channel (max speed)	-	4	6.5	
ET	unadjusted error		Differential	Fast channel (max speed)	-	3.5	5.5	
			Differential	Slow channel (max speed)	-	3.5	5.5	
			Single	Fast channel (max speed)	-	1	4.5	
EO	Offset		ended	Slow channel (max speed)	-	1	5	
	error		Differential	Fast channel (max speed)	-	1.5	3	
			Dillerential	Slow channel (max speed)	-	1.5	3	
			Single	Fast channel (max speed)	-	2.5	6	
EG	Gain error		ended	Slow channel (max speed)	-	2.5	6	LSB
EG	Gain enoi		Differential	Fast channel (max speed)	-	2.5	3.5	LOD
			Dillerential	Slow channel (max speed)	-	2.5	3.5	
			Single	Fast channel (max speed)	-	1	1.5	
Differential ED linearity		ended	Slow channel (max speed)	-	1	1.5		
	ED linearity error	ADC clock frequency ≤ 80 MHz,  Sampling rate ≤ 5.33 Msps, 2 V ≤ V <sub>DDA</sub>	Differential	Fast channel (max speed)	-	1	1.2	
			Dillerential	Slow channel (max speed)	-	1	1.2	
			Single	Fast channel (max speed)	-	1.5	3.5	
EL	Integral		ended	Slow channel (max speed)	-	1.5	3.5	-
	linearity error		Differential	Fast channel (max speed)	-	1	3	
			Dillerential	Slow channel (max speed)	-	1	2.5	
			Single	Fast channel (max speed)	10	10.5	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10	10.5	-	bits
ENOB	bits		Differential	Fast channel (max speed)	10.7	10.9	-	DILS
			Dillerential	Slow channel (max speed)	10.7	10.9	-	
	Cianal to		Single	Fast channel (max speed)	62	65	-	
CINIAD	Signal-to- noise and		ended	Slow channel (max speed)	62	65	-	
SINAD	distortion		Differential	Fast channel (max speed)	66	67.4	-	
ratio		Dillerential	Slow channel (max speed)	66	67.4	-	٩D	
			Single	Fast channel (max speed)	64	66	-	dB
SNR	Signal-to-	<u> </u>	Olligic	Slow channel (max speed)	64	66	-	
SINK	noise ratio			Fast channel (max speed)	66.5	68	-	_
			Differential	Slow channel (max speed)	66.5	68	-	



Table 64. ADC accuracy - limited test conditions  $2^{(1)(2)(3)}$  (continued)

Sym- bol	Parameter	C	Conditions <sup>(4)</sup>					
		ADC clock frequency ≤	Single	Fast channel (max speed)	-	-74	-65	
THD	Total	otal 80 MHz,	ended	Slow channel (max speed)	-	-74	-67	dB
distortion	stortion   Sampling rate ≤ 5.33 Msps,	Differential	Fast channel (max speed)	-	-79	-70	uБ	
		2 V ≤ V <sub>DDA</sub>	Dillerential	Slow channel (max speed)	-	-79	-71	

- 1. Guaranteed by design.
- 2. ADC DC accuracy values are measured after internal calibration.
- 3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4 V). It is disable when  $V_{DDA} \ge 2.4$  V. No oversampling.

Table 65. ADC accuracy - limited test conditions 3<sup>(1)(2)(3)</sup>

Sym- bol	Parameter	(	Conditions <sup>(4</sup>	)	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	5.5	7.5	
ET	Total unadjusted		ended	Slow channel (max speed)	-	4.5	6.5	
	error		Differential	Fast channel (max speed)	-	4.5	7.5	
			Dillerential	Slow channel (max speed)	-	4.5	5.5	
			Single	Fast channel (max speed)	-	2	5	
EO	Offset		ended	Slow channel (max speed)	-	2.5	5	
	error		Differential	Fast channel (max speed)	-	2	3.5	
			Dillerential	Slow channel (max speed)	-	2.5	3	
			Single	Fast channel (max speed)	-	4.5	7	
EG	Gain error		ended	Slow channel (max speed)	-	3.5	6	LSB
EG	Gairrenoi		Differential -	Fast channel (max speed)	-	3.5	4	LSB
		L	Dillerential	Slow channel (max speed)	-	3.5	5	
	- In the second		Single	Fast channel (max speed)	-	1.2	1.5	
Differential ED linearity		ended	Slow channel (max speed)	-	1.2	1.5		
	error	ADC clock frequency ≤ 80 MHz,  Sampling rate ≤ 5.33 Msps,  1.65 V ≤ V <sub>DDA</sub> = V <sub>REF+</sub> ≤	Differential	Fast channel (max speed)	-	1	1.2	
			Dillerential	Slow channel (max speed)	-	1	1.2	
			Single	Fast channel (max speed)	-	3	3.5	
EL	Integral linearity	3.6 V, Voltage scaling Range 1	ended	Slow channel (max speed)	-	2.5	3.5	
	error		Differential	Fast channel (max speed)	-	2	2.5	
			Dillerential	Slow channel (max speed)	-	2	2.5	
			Single	Fast channel (max speed)	10	10.4	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10	10.4	1	bits
LINOD	bits		Differential	Fast channel (max speed)	10.6	10.7	-	Dita
			Dilicicitiai	Slow channel (max speed)	10.6	10.7	1	
	Signal-to-		Single	Fast channel (max speed)	62	64	1	
SINAD	noise and		ended	Slow channel (max speed)	62	64	-	
SINAD	distortion		Differential	Fast channel (max speed)	65	66	-	
	ratio		Dillerential	Slow channel (max speed)	65	66	-	dB
			Single	Fast channel (max speed)	63	65	ı	ub
SNR	Signal-to-		Olligic	Slow channel (max speed)	63	65	-	
SINIX	noise ratio		Differential -	Fast channel (max speed)	66	67	-	
			Dillorential	Slow channel (max speed)	66	67	-	



Table 65. ADC accuracy - limited test conditions  $3^{(1)(2)(3)}$  (continued)

Sym- bol	Parameter	C	Min	Тур	Max	Unit		
		ADC clock frequency ≤	Single	Fast channel (max speed)	-	-69	-67	
Т	Total	Sampling rate ≤ 5.33 Msps, _	ended	Slow channel (max speed)	-	-71	-67	
THD	harmonic distortion			Fast channel (max speed)	-	-72	-71	dB
	distortion	3.6 V, Voltage scaling Range 1	Differential	Slow channel (max speed)	-	-72	-71	

- 1. Guaranteed by design.
- 2. ADC DC accuracy values are measured after internal calibration.
- ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this
  significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a
  Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4 V). It is disable when  $V_{DDA} \ge 2.4$  V. No oversampling.

Table 66. ADC accuracy - limited test conditions 4<sup>(1)(2)(3)</sup>

Sym- bol	Parameter	(	Conditions <sup>(4</sup>	)	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	5	5.4	
ET	Total		ended	Slow channel (max speed)	-	4	5	
E1	unadjusted error		Differential	Fast channel (max speed)	-	4	5	
			Dillerential	Slow channel (max speed)	-	3.5	4.5	
			Single ended	Fast channel (max speed)	-	2	4	
EO	Offset			Slow channel (max speed)	-	2	4	
	error		Differential	Fast channel (max speed)	-	2	3.5	
			Dillerential	Slow channel (max speed)	-	2	3.5	
			Single	Fast channel (max speed)	-	4	4.5	
EG	Gain error		ended	Slow channel (max speed)	-	4	4.5	LSB
EG	Gain enoi		Differential	Fast channel (max speed)	-	3	4	LSB
			Dillerential	Slow channel (max speed)	-	3	4	
		ADC clock frequency ≤ 26 MHz,  1.65 V ≤ V <sub>DDA</sub> = VREF+ ≤ 3.6 V,	Single	Fast channel (max speed)	-	1	1.5	
ED	ED Differential linearity error		ended	Slow channel (max speed)	-	1	1.5	-
			Differential	Fast channel (max speed)	-	1	1.2	
			Dillerential	Slow channel (max speed)	-	1	1.2	
			Single ended  Differential	Fast channel (max speed)	-	2.5	3	
EL	Integral linearity	Voltage scaling Range 2		Slow channel (max speed)	-	2.5	3	
	error			Fast channel (max speed)	-	2	2.5	
			Dillerential	Slow channel (max speed)	-	2	2.5	
			Single	Fast channel (max speed)	10.2	10.5	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10.2	10.5	-	bits
LINOB	bits		Differential	Fast channel (max speed)	10.6	10.7	ı	Dita
			Dillerential	Slow channel (max speed)	10.6	10.7	-	
	Signal-to-		Single	Fast channel (max speed)	63	65	-	
SINAD	noise and		ended	Slow channel (max speed)	63	65	ı	
OIIVAD	distortion		Differential	Fast channel (max speed)	65	66	ı	
	ratio		Dillerential	Slow channel (max speed)	65	66	-	dB
			Single	Fast channel (max speed)	64	65	-	ub
SNR	Signal-to-		ended	Slow channel (max speed)	64	65	ı	<u> </u>
SINK	noise ratio		Differential	Fast channel (max speed)	66	67	ı	
			Dilletetilial	Slow channel (max speed)	66	67	-	



Sym- bol	Parameter	C	Conditions <sup>(4</sup>	)	Min	Тур	Max	Unit
		ADC clock frequency ≤	Single	Fast channel (max speed)	-	-71	-69	
THD	Total		ended	Slow channel (max speed)	-	-71	-69	dB
THD harmonic distortion	$1.65 \text{ V} \le \text{V}_{\text{DDA}} = \text{VREF+} \le$ 3.6  V,	Differential	Fast channel (max speed)	-	-73	-72	uБ	
		Voltage scaling Range 2	Dillerential	Slow channel (max speed)	1	-73	-72	

Slow channel (max speed)

Table 66. ADC accuracy - limited test conditions  $4^{(1)(2)(3)}$  (continued)

- Guaranteed by design.
- ADC DC accuracy values are measured after internal calibration.
- ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- The I/O analog switch voltage booster is enable when  $V_{DDA}$  < 2.4 V (BOOSTEN = 1 in the SYSCFG\_CFGR1 when  $V_{DDA}$  < 2.4 V). It is disable when  $V_{DDA} \ge 2.4$  V. No oversampling.

Figure 20. ADC accuracy characteristics Vssa Eg (1) Example of an actual transfer curve 4095 (2) The ideal transfer curve (3) End point correlation line 4094 4093 ET = Total Unajusted Error: maximum deviation between the actual and ideal transfer curves. Eo = Offset Error: maximum deviation between the first actual transition and the first ideal one. 6 Eg = Gain Error: deviation between the last 5 ideal transition and the last actual one. ED = Differential Linearity Error: maximum deviation between actual steps and the ideal ones. 3 EL = Integral Linearity Error: maximum deviation 2 between any actual transition and the end point 1 LSB IDEAL correlation line. Vdda 0 4093 4094 4095 4096 6 MS19880V2

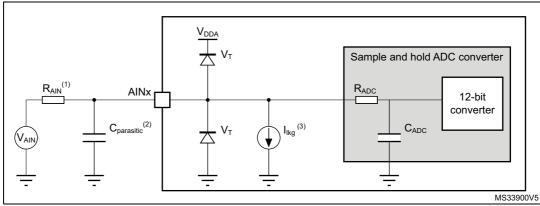


Figure 21. Typical connection diagram using the ADC

- 1. Refer to Table 61: ADC characteristics for the values of  $R_{AIN}$ ,  $R_{ADC}$  and  $C_{ADC}$ .
- C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to *Table 56: I/O static characteristics* for the value of the pad capacitance). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.
- 3. Refer to Table 56: I/O static characteristics for the values of Ilkg.

### General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 8: Power supply scheme*. The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.

# 6.3.18 Digital-to-Analog converter characteristics

Table 67. DAC characteristics<sup>(1)</sup>

Symbol	Parameter	Co	onditions	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage for DAC ON	-		1.8	-	3.6	
V <sub>REF+</sub>	Positive reference voltage	-		1.8	-	$V_{DDA}$	V
V <sub>REF-</sub>	Negative reference voltage		-		V <sub>SSA</sub>		
$R_{L}$	Resistive load	DAC output	connected to V <sub>SSA</sub>	5	-	-	kΩ
, , _	Troolotivo loud	buffer ON	connected to V <sub>DDA</sub>	25	-	-	K77
R <sub>O</sub>	Output Impedance	DAC output bu	ffer OFF	9.6	11.7	13.8	kΩ
	Output impedance sample	V <sub>DD</sub> = 2.7 V		-	-	2	
R <sub>BON</sub>	and hold mode, output buffer ON	V <sub>DD</sub> = 2.0 V		-	-	3.5	kΩ
	Output impedance sample	V <sub>DD</sub> = 2.7 V		-	-	16.5	
R <sub>BOFF</sub>	R <sub>BOFF</sub> and hold mode, output buffer OFF			-	-	18.0	kΩ
C <sub>L</sub>	Conscitive load	DAC output buffer ON Sample and hold mode		-	-	50	pF
C <sub>SH</sub>	Capacitive load			-	0.1	1	μF
V <sub>DAC_OUT</sub>	Voltage on DAC_OUT	DAC output bu	ffer ON	0.2	-	V <sub>REF+</sub> - 0.2	V
_	output	DAC output buffer OFF		0	-	V <sub>REF+</sub>	
	Cattling time (full scale) for		±0.5 LSB	-	1.7	3	
	Settling time (full scale: for a 12-bit code transition	Normal mode DAC output	±1 LSB	-	1.6	2.9	
	between the lowest and the highest input codes	buffer ON	±2 LSB	-	1.55	2.85	
t <sub>SETTLING</sub>	when DAC_OUT reaches	CL ≤ 50 pF, RL ≥ 5 kΩ	±4 LSB	-	1.48	2.8	μs
	final value ±0.5LSB, ±1 LSB, ±2 LSB, ±4 LSB,		±8 LSB	-	1.4	2.75	
	±8 LSB)	Normal mode I OFF, ±1LSB, C	DAC output buffer CL = 10 pF	-	2	2.5	
t <sub>WAKEUP</sub> (2) (s	Wakeup time from off state (setting the ENx bit in the	Normal mode DAC output buffer ON CL $\leq$ 50 pF, RL $\geq$ 5 k $\Omega$		-	4.2	7.5	
	DAC Control register) until final value ±1 LSB	Normal mode DAC output buffer OFF, CL ≤ 10 pF		-	2	5	μs
PSRR	V <sub>DDA</sub> supply rejection ratio	Normal mode I CL ≤ 50 pF, RL	DAC output buffer ON . = 5 kΩ, DC	-	-80	-28	dB

Table 67. DAC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter		enditions	Min	Тур	Max	Unit
		DAC_OUT	DAC output buffer ON, C <sub>SH</sub> = 100 nF	-	0.7	3.5	mo
	Sampling time in sample and hold mode (code transition between the	pin connected	DAC output buffer OFF, C <sub>SH</sub> = 100 nF	ı	10.5	18	ms
t <sub>SAMP</sub>	t <sub>SAMP</sub> lowest input code and the highest input code when DACOUT reaches final value ±1LSB)	DAC_OUT pin not connected (internal connection only)	DAC output buffer OFF	1	2	3.5	μs
I <sub>leak</sub>	Output leakage current	Sample and ho DAC_OUT pin		-	-	_(3)	nA
Cl <sub>int</sub>	Internal sample and hold capacitor		-	5.2	7	8.8	pF
t <sub>TRIM</sub>	Middle code offset trim time	DAC output bu	ffer ON	50	-	-	μs
V	Middle code offset for 1 V <sub>REF+</sub> = 3.6 V		-	1500	-	μV	
V <sub>offset</sub>	trim code step	V <sub>REF+</sub> = 1.8 V		-	750	-	μν
		DAC output	No load, middle code (0x800)	-	315	500	
		buffer ON	No load, worst code (0xF1C)	ı	450	670	
I <sub>DDA</sub> (DAC)	DAC consumption from V <sub>DDA</sub>	DAC output buffer OFF	No load, middle code (0x800)	-	-	0.2	μA
		Sample and hold mode, C <sub>SH</sub> = 100 nF		-	315 <sub>x</sub> Ton/(Ton +Toff) (4)	670 x Ton/(Ton +Toff) (4)	
		DAC output	No load, middle code (0x800)	-	185	240	
		buffer ON	No load, worst code (0xF1C)	-	340	400	
		DAC output buffer OFF	No load, middle code (0x800)	-	155	205	
I <sub>DDV</sub> (DAC)	DAC consumption from V <sub>REF+</sub>	Sample and hold mode, buffer ON, C <sub>SH</sub> = 100 nF, worst case		-	185 <sub>x</sub> Ton/(Ton +Toff) (4)	400 x Ton/(Ton +Toff) (4)	μА
		Sample and ho C <sub>SH</sub> = 100 nF,	old mode, buffer OFF, worst case	-	155 x Ton/(Ton +Toff) (4)	205 x Ton/(Ton +Toff) (4)	

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> In buffered mode, the output can overshoot above the final value for low input code (starting from min value).



- 3. Refer to Table 56: I/O static characteristics.
- 4. Ton is the Refresh phase duration. Toff is the Hold phase duration. Refer to RM0393 reference manual for more details.

Buffered/non-buffered DAC

Buffer (1)

12-bit digital to analog converter

DACX\_OUT

RLOAD

CLOAD

ai17157d

Figure 22. 12-bit buffered / non-buffered DAC

The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly
without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the
DAC\_CR register.

Table 68. DAC accuracy<sup>(1)</sup>

Symbol	Parameter	Conditio	ns	Min	Тур	Max	Unit
DNII	Differential non	DAC output buffer ON		-	-	±2	
DNL	linearity (2)	DAC output buffer OFF		-	-	±2	
-	monotonicity	10 bits		g	guarantee	d	
INL	Integral non	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ			-	±4	
IINL	linearity <sup>(3)</sup>	DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±4	
	DAC output buffer ON	V <sub>REF+</sub> = 3.6 V	-	-	±12		
Offset	Offset error at code 0x800 <sup>(3)</sup>	CL ≤ 50 pF, RL ≥ 5 kΩ	V <sub>REF+</sub> = 1.8 V	-	-	±25	LSB
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±8	
Offset1	Offset error at code 0x001 <sup>(4)</sup>	DAC output buffer OFF CL ≤ 50 pF, no RL		_	-	±5	
OffsetCal	Offset Error at	at DAC output buffer ON	V <sub>REF+</sub> = 3.6 V	-	-	±5	
OffsetCal code 0x800 after calibration	CL ≤ 50 pF, RL ≥ 5 k $\Omega$	V <sub>REF+</sub> = 1.8 V	-	-	±7		

Table 68. DAC accuracy<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Gain	Gain error <sup>(5)</sup>	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±0.5	%
Gaiii	Gain endiv	DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±0.5	70
Total TUE unadjusted error	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±30	LSB	
		DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±12	LOB
TUECal	Total unadjusted error after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±23	LSB
SNR	Signal-to-noise	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ 1 kHz, BW 500 kHz	-	71.2	-	dB
SINK	ratio	ratio  DAC output buffer OFF  CL ≤ 50 pF, no RL, 1 kHz  BW 500 kHz	-	71.6	-	uв
THD	Total harmonic	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	-78	-	dB
IIID	distortion	DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	-79	-	uБ
SINAD	Signal-to-noise and distortion	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	70.4	-	dB
SINAD	ratio	DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	71	-	uБ
ENOB	Effective	DAC output buffer ON $CL \le 50 \text{ pF, } RL \ge 5 \text{ k}\Omega, 1 \text{ kHz}$	-	11.4	-	bits
ENOB	number of bits	DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	11.5	-	มแจ

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> Difference between two consecutive codes - 1 LSB.

<sup>3.</sup> Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.

<sup>4.</sup> Difference between the value measured at Code (0x001) and the ideal value.

Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFF when buffer is OFF, and from code giving 0.2 V and (V<sub>REF+</sub> – 0.2) V when buffer is ON.

## 6.3.19 Comparator characteristics

Table 69. COMP characteristics<sup>(1)</sup>

Symbol	Parameter	Co	onditions	Min	Тур	Max	Unit
V <sub>DDA</sub>	Analog supply voltage		-	1.62	-	3.6	
V <sub>IN</sub>	Comparator input voltage range	-		0	-	$V_{DDA}$	V
V <sub>BG</sub> <sup>(2)</sup>	Scaler input voltage		-		V <sub>REFINT</sub>		
V <sub>SC</sub>	Scaler offset voltage		-	-	±5	±10	mV
L (SCALED)	Scaler static consumption BRG_EN=0 (bridge disable)		-	200	300	nA	
I <sub>DDA</sub> (SCALER)	from V <sub>DDA</sub>	BRG_EN=1 (bridge enable)		-	0.8	1	μΑ
t <sub>START_SCALER</sub>	Scaler startup time		-	-	100	200	μs
		High-speed	V <sub>DDA</sub> ≥ 2.7 V	-	-	5	
	Comparator startup time to	mode	V <sub>DDA</sub> < 2.7 V	-	-	7	
t <sub>START</sub>	reach propagation delay specification	Ma di una una ala	V <sub>DDA</sub> ≥ 2.7 V	-	-	15	μs
		Medium mode	V <sub>DDA</sub> < 2.7 V	-	-	25	
		Ultra-low-powe	Ultra-low-power mode		-	40	
t <sub>D</sub> <sup>(3)</sup>	Propagation delay with	High-speed V <sub>DDA</sub> ≥ 2.7 V	-	55	80		
		mode	V <sub>DDA</sub> < 2.7 V	-	65	100	ns 00
	100 mV overdrive	Medium mode	l	-	0.55	0.9	
		Ultra-low-powe	r mode	-	4	7	μs
V <sub>offset</sub>	Comparator offset error	Full common mode range	-	-	±5	±20	mV
		No hysteresis		-	0	-	
.,		Low hysteresis		-	8	-	.,
$V_{hys}$	Comparator hysteresis	Medium hyster	esis	-	15	-	mV
		High hysteresis	3	-	27	-	
			Static	-	400	600	
		Ultra-low- power mode	With 50 kHz ±100 mV overdrive square signal	-	1200	-	nA
			Static	-	5	7	
I <sub>DDA</sub> (COMP)	Comparator consumption from V <sub>DDA</sub>	Medium mode	With 50 kHz ±100 mV overdrive square signal	-	6	-	
			Static	-	70	100	μA
		High-speed M mode	With 50 kHz ±100 mV overdrive square signal	-	75	-	

<sup>1.</sup> Guaranteed by design, unless otherwise specified.



- 2. Refer to Table 23: Embedded internal voltage reference.
- 3. Guaranteed by characterization results.

# 6.3.20 Operational amplifiers characteristics

Table 70. OPAMP characteristics<sup>(1)</sup>

Symbol	Parameter	Con	ditions	Min	Тур	Max	Unit	
V <sub>DDA</sub>	Analog supply voltage <sup>(2)</sup>		-	1.8	-	3.6	V	
CMIR	Common mode input range		-	0	-	$V_{DDA}$	V	
\/I	Input onoct	25 °C, No Load on	output.	-	-	±1.5	mV	
VI <sub>OFFSET</sub>	voltage	Itage All voltage/Temp	-	-	±3	IIIV		
ΔVI <sub>OFFSET</sub>	Input offset	Normal mode		-	±5	-	μV/°C	
AVIOFFSET	voltage drift	Low-power mode		-	±10	-	μν/ Ο	
TRIMOFFSETP TRIMLPOFFSETP	Offset trim step at low common input voltage (0.1 x V <sub>DDA</sub> )		-	-	0.8	1.1	mV	
TRIMOFFSETN TRIMLPOFFSETN	Offset trim step at high common input voltage (0.9 x V <sub>DDA</sub> )		-		1	1.35	1111	
I <sub>LOAD</sub>	Drive current	Normal mode	- V <sub>DDA</sub> ≥ 2 V	-	-	500		
LOAD	Brive current	ow-power mode	-	-	100	μA		
I <sub>LOAD_PGA</sub>	Drive current in	Normal mode	ormal mode V <sub>DDA</sub> ≥ 2 V	-	-	450	μπ	
'LOAD_PGA	PGA mode	Low-power mode	VDDA - 2 V	-	-	50		
R <sub>LOAD</sub>	Resistive load (connected to	Normal mode	- V <sub>DDA</sub> < 2 V	4	-	-		
LLOAD	VSSA or to VDDA)	Low-power mode	334 01 10		20	-	-	kΩ
D	Resistive load in PGA mode (connected to	Normal mode	V ~2V	4.5	-	-	K\$2	
R <sub>LOAD_PGA</sub>	VSSA or to V <sub>DDA</sub> )	Low-power mode	- V <sub>DDA</sub> < 2 V	40	-	-		
C <sub>LOAD</sub>	Capacitive load		-	-	-	50	pF	
	Common mode	Normal mode		-	-85	-	-10	
CMRR	rejection ratio	Low-power mode		-	-90	-	dB	
PSRR	Power supply rejection ratio	Normal mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 4 \text{ k}\Omega \text{ DC}$	70	85	-	dB	
TORK		Low-power mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 20 \text{ k}\Omega \text{ DC}$	72	90	-	uБ	

Table 70. OPAMP characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Con	ditions	Min	Тур	Max	Unit
		Normal mode	V <sub>DDA</sub> ≥ 2.4 V	550	1600	2200	
GBW	Gain Bandwidth	Low-power mode	(OPA_RANGE = 1)	100	420	600	kHz
GBVV	Product	Normal mode	V <sub>DDA</sub> < 2.4 V	250	700	950	KI IZ
		Low-power mode	(OPA_RANGE = 0)	40	180	280	
	01	Normal mode	V >24V	-	700	-	
SR <sup>(3)</sup>	Slew rate (from 10 and	Low-power mode	- V <sub>DDA</sub> ≥ 2.4 V	-	180	-	) //
5K(*)	90% of output	Normal mode	V	-	300	-	V/ms
	voltage)	Low-power mode	− V <sub>DDA</sub> < 2.4 V	-	80	-	
4.0	0	Normal mode		55	110	-	Ē
AO	Open loop gain	Low-power mode		45	110	-	dB
V (3)	$V_{OHSAT}^{(3)}$ High saturation voltage Normal mode $I_{load}$ = max or $R_{load}$ = min Input at $V_{DDA}$ .	I <sub>load</sub> = max or R <sub>load</sub> =	V <sub>DDA</sub> - 100	-	-		
VOHSAT` ′		min Input at V <sub>DDA</sub> .	V <sub>DDA</sub> - 50	-	-	mV	
V <sub>OLSAT</sub> <sup>(3)</sup>	Low saturation		I <sub>load</sub> = max or R <sub>load</sub> =	-	-	100	
VOLSAT`	voltage Low-power m	Low-power mode	min Input at 0.	-	-	50	
	Dhago marain	Normal mode	Normal mode		74	-	0
$\Phi_{m}$	Phase margin	Low-power mode		-	66	-	
GM	Coin margin	Normal mode		-	13	-	dB
Givi	Gain margin	Low-power mode		-	20	-	uБ
	Wake up time	Normal mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 4 \text{ k}\Omega$ follower configuration	-	5	10	
<sup>t</sup> WAKEUP	from OFF state.	Low-power mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 20 \text{ k}\Omega$ follower configuration	-	10	30	μs
I <sub>bias</sub>	OPAMP input bias current	General purpose input		-	-	_(4)	nA
				-	2	-	
PGA gain <sup>(3)</sup>	Non inverting	-		-	4	-	_
PGA gain <sup>(3)</sup>	gain value			-	8	-	
				-	16	-	

Table 70. OPAMP characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Con	ditions	Min	Тур	Max	Unit
		PGA Gain = 2		-	80/80	1	
	R2/R1 internal resistance	PGA Gain = 4		-	120/ 40	-	
R <sub>network</sub>	values in PGA mode <sup>(5)</sup>	PGA Gain = 8		-	140/ 20	-	kΩ/kΩ
		PGA Gain = 16		-	150/ 10	-	
Delta R	Resistance variation (R1 or R2)	-		-15	-	15	%
PGA gain error	PGA gain error		-	-1	-	1	%
	PGA bandwidth for different non inverting gain	Gain = 2	-	-	GBW/ 2	-	
DCA DW		Gain = 4	-	-	GBW/ 4	-	MHz
PGA BW		Gain = 8	-	-	GBW/ 8	-	IVITZ
		Gain = 16	-	-	GBW/ 16	-	
		Normal mode	at 1 kHz, Output loaded with 4 kΩ	-	500	-	
on	Voltage noise	Low-power mode	at 1 kHz, Output loaded with 20 kΩ	-	600	-	nV/√Hz
en	density	Normal mode	at 10 kHz, Output loaded with 4 kΩ	-	180	-	TIIV/ VIIZ
		Low-power mode	at 10 kHz, Output loaded with 20 kΩ	-	290	-	
(0000000)(3)	OPAMP	Normal mode no	no Load, quiescent mode	-	120	260	
I <sub>DDA</sub> (OPAMP) <sup>(3)</sup>	consumption from V <sub>DDA</sub>	Low-power mode		-	45	100	μA

- 1. Guaranteed by design, unless otherwise specified.
- 2. The temperature range is limited to 0 °C-125 °C when  $V_{DDA}$  is below 2 V
- 3. Guaranteed by characterization results.
- 4. Mostly I/O leakage, when used in analog mode. Refer to  $I_{lkg}$  parameter in Table 56: I/O static characteristics.
- R2 is the internal resistance between OPAMP output and OPAMP inverting input. R1 is the internal resistance between OPAMP inverting input and ground. The PGA gain =1+R2/R1

## 6.3.21 Temperature sensor characteristics

Table 71. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>TS</sub> linearity with temperature	-	±1	±2	°C
Avg_Slope <sup>(2)</sup>	Average slope	2.3	2.5	2.7	mV/°C
V <sub>30</sub>	Voltage at 30°C (±5 °C) <sup>(3)</sup>	0.742	0.76	0.785	V
t <sub>START</sub> (TS_BUF) <sup>(1)</sup>	Sensor Buffer Start-up time in continuous mode <sup>(4)</sup>	-	8	15	μs
t <sub>START</sub> (1)	Start-up time when entering in continuous mode <sup>(4)</sup>	-	70	120	μs
t <sub>S_temp</sub> <sup>(1)</sup>	ADC sampling time when reading the temperature		-	-	μs
I <sub>DD</sub> (TS) <sup>(1)</sup>	Temperature sensor consumption from $V_{DD}$ , when selected by ADC	-	4.7	7	μΑ

<sup>1.</sup> Guaranteed by design.

## 6.3.22 Timer characteristics

The parameters given in the following tables are guaranteed by design.

Refer to Section 6.3.14: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 72. TIMx<sup>(1)</sup> characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
t	Timer resolution time	-	1	-	t <sub>TIMxCLK</sub>
t <sub>res(TIM)</sub>	Time resolution time	f <sub>TIMxCLK</sub> = 80 MHz	12.5	-	ns
f	Timer external clock	-	0	f <sub>TIMxCLK</sub> /2	MHz
'EXT	frequency on CH1 to CH4	f <sub>TIMxCLK</sub> = 80 MHz	0	40	MHz
Res <sub>TIM</sub>	Timer resolution	TIMx (except TIM2)	-	16	bit
		TIM2	-	32	
4	16-bit counter clock	-	1	65536	t <sub>TIMxCLK</sub>
<sup>t</sup> COUNTER	period	f <sub>TIMxCLK</sub> = 80 MHz	0.0125	819.2	μs
t	Maximum possible count	-	-	65536 × 65536	t <sub>TIMxCLK</sub>
t <sub>MAX_COUNT</sub>	with 32-bit counter	f <sub>TIMxCLK</sub> = 80 MHz	-	53.68	s

<sup>1.</sup> TIMx is used as a general term in which x stands for 1,2,3,4,5,6,7,8,15,16 or 17.

<sup>2.</sup> Guaranteed by characterization results.

Measured at V<sub>DDA</sub> = 3.0 V ±10 mV. The V<sub>30</sub> ADC conversion result is stored in the TS\_CAL1 byte. Refer to Table 6: Temperature sensor calibration values.

<sup>4.</sup> Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFF	Unit
/4	0	0.125	512	
/8	1	0.250	1024	
/16	2	0.500	2048	
/32	3	1.0	4096	ms
/64	4	2.0	8192	
/128	5	4.0	16384	
/256	6 or 7	8.0	32768	

Table 73. IWDG min/max timeout period at 32 kHz (LSI)<sup>(1)</sup>

The exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there
is always a full RC period of uncertainty.

Prescaler	WDGTB Min timeout value Max timeout value		Unit						
1	0	0.0512	3.2768						
2	1	0.1024	6.5536	me					
4	2	0.2048	13.1072	ms					
8	3	0.4096	26.2144						

Table 74. WWDG min/max timeout value at 80 MHz (PCLK)

#### 6.3.23 Communication interfaces characteristics

## I<sup>2</sup>C interface characteristics

The I2C interface meets the timings requirements of the I<sup>2</sup>C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The I2C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to RM0393 reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and  $V_{\rm DDIOx}$  is disabled, but is still present. Only FT\_f I/O pins support Fm+ low level output current maximum requirement. Refer to Section 6.3.14: I/O port characteristics for the I2C I/Os characteristics.

All I2C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

## Table 75. I2C analog filter characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>AF</sub>	Maximum pulse width of spikes that are suppressed by the analog filter	50 <sup>(2)</sup>	260 <sup>(3)</sup>	ns

- 1. Guaranteed by design.
- 2. Spikes with widths below  $t_{\mbox{\scriptsize AF}(\mbox{\scriptsize min})}$  are filtered.
- 3. Spikes with widths above  $t_{\text{AF}(\text{max})}$  are not filtered

#### **SPI** characteristics

Unless otherwise specified, the parameters given in *Table 76* for SPI are derived from tests performed under the ambient temperature, f<sub>PCLKx</sub> frequency and supply voltage conditions summarized in *Table 20: General operating conditions*.

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V<sub>DD</sub>

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

Table 76. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode receiver/full duplex 2.7 < V <sub>DD</sub> < 3.6 V Voltage Range 1			40	
		Master mode receiver/full duplex 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 1			16	
٠		Master mode transmitter 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 1			40	
f <sub>SCK</sub> 1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode receiver 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 1	-	-		MHz
	Slave mode transmitter/full duplex 2.7 < V <sub>DD</sub> < 3.6 V Voltage Range 1			37 <sup>(2)</sup>		
		Slave mode transmitter/full duplex 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 1			20 <sup>(2)</sup>	
		Voltage Range 2			13	
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI prescaler = 2	4 <sub>x</sub> T <sub>PCLK</sub>	-	-	ns
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI prescaler = 2	2 <sub>x</sub> T <sub>PCLK</sub>	ı	-	ns
$\begin{matrix} t_{w(\text{SCKH})} \\ t_{w(\text{SCKL})} \end{matrix}$	SCK high and low time	Master mode	T <sub>PCLK</sub> -2	T <sub>PCLK</sub>	T <sub>PCLK</sub> +2	ns
t <sub>su(MI)</sub>	Data input setup time	Master mode	4	-	-	ns
t <sub>su(SI)</sub>	Data input setup time	Slave mode	1.5	-	-	113
t <sub>h(MI)</sub>	Data input hold time	Master mode	6.5	ı	-	ns
t <sub>h(SI)</sub>	Data input noid time	Slave mode	1.5	ı	-	113
t <sub>a(SO)</sub>	Data output access time	Slave mode	9	-	36	ns
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	9	-	16	ns



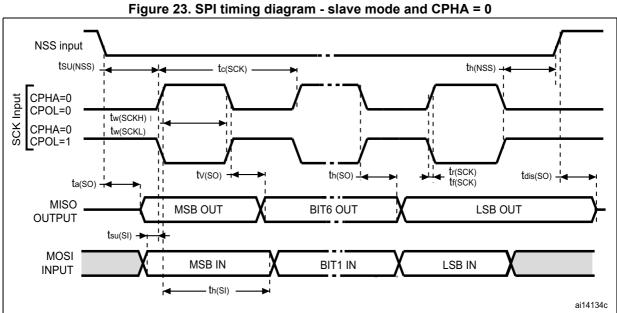
0

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	t <sub>v(SO)</sub> Data output valid time	Slave mode 2.7 < V <sub>DD</sub> < 3.6 V Voltage Range 1	-	12.5	13.5	
t <sub>v(SO)</sub>		Slave mode 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 1	-	12.5	24	ns
		Slave mode 1.71 < V <sub>DD</sub> < 3.6 V Voltage Range 2	-	12.5	33	
t <sub>v(MO)</sub>		Master mode	-	4.5	6	
t <sub>h(SO)</sub>	Data output hold time	Slave mode	7	-	-	ne

Table 76. SPI characteristics<sup>(1)</sup> (continued)

t<sub>h(MO)</sub>

Master mode



<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Maximum frequency in Slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty(SCK) = 50 %.

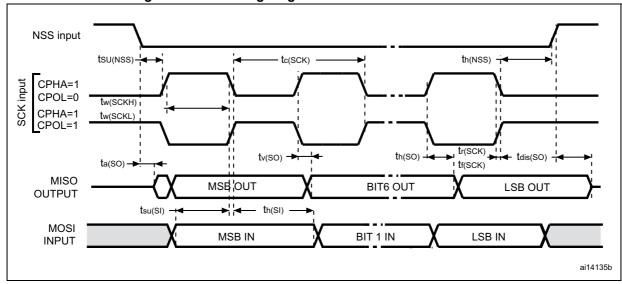
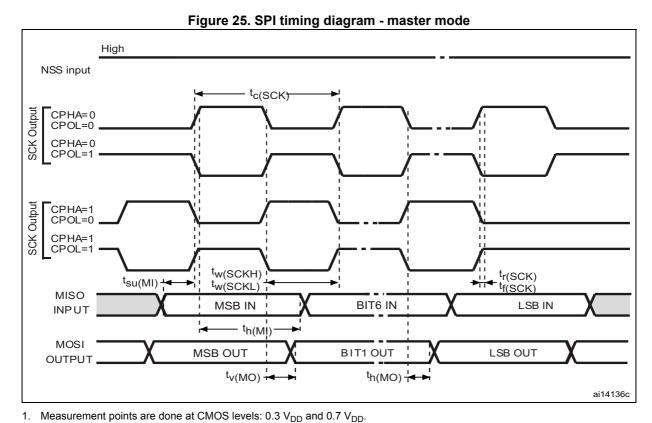


Figure 24. SPI timing diagram - slave mode and CPHA = 1

1. Measurement points are done at CMOS levels: 0.3  $\rm V_{DD}$  and 0.7  $\rm V_{DD}$ .



#### **Quad SPI characteristics**

Unless otherwise specified, the parameters given in *Table 77* and *Table 78* for Quad SPI are derived from tests performed under the ambient temperature,  $f_{AHB}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 20: General operating conditions*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 15 or 20 pF
- Measurement points are done at CMOS levels: 0.5 x V<sub>DD</sub>

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics.

Table 77. Quad SPI characteristics in SDR mode<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		1.71 < V <sub>DD</sub> < 3.6 V, C <sub>LOAD</sub> = 20 pF Voltage Range 1	-	-	40	
F <sub>CK</sub>	Quad SPI clock frequency	1.71 < V <sub>DD</sub> < 3.6 V, C <sub>LOAD</sub> = 15 pF Voltage Range 1	-	-	48	MHz
1/t <sub>(CK)</sub>	Quad of Follock frequency	2.7 < V <sub>DD</sub> < 3.6 V, C <sub>LOAD</sub> = 15 pF Voltage Range 1	-	-	60	IVII IZ
		1.71 < V <sub>DD</sub> < 3.6 V C <sub>LOAD</sub> = 20 pF Voltage Range 2	-	-	26	
t <sub>w(CKH)</sub>	Quad SPI clock high and	f= 48 MHz, presc=0	48 MHz, presc=0 $ \frac{t_{(CK)}/2-2}{t_{(CK)}/2} $	-	t <sub>(CK)</sub> /2	
t <sub>w(CKL)</sub>	low time	TAHBCLK - 40 Mil 12, presc-0		-	t <sub>(CK)</sub> /2+2	
<b>+</b>	Data input setup time	Voltage Range 1	2	-	-	
t <sub>s(IN)</sub>	Data input setup time	Voltage Range 2 3.	3.5	-	-	
+	Data input hold time	Voltage Range 1	5	-	-	ns
t <sub>h(IN)</sub>	Data input hold time	Voltage Range 2	6.5	-	-	115
	Data output valid time	Voltage Range 1	-	1	5	
t <sub>v(OUT)</sub>	Data output valid time	Voltage Range 2	-	3	5	
+	Data output hold time	Voltage Range 1	0	-	-	
t <sub>h(OUT)</sub>	Data output hold time	Voltage Range 2	0	-	-	

<sup>1.</sup> Guaranteed by characterization results.

Table 78. QUADSPI characteristics in DDR mode<sup>(1)</sup>

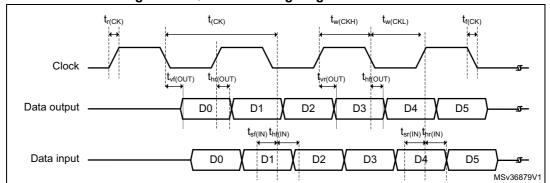
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		$1.71 < V_{DD} < 3.6 \text{ V, } C_{LOAD} = 20 \text{ pF}$ Voltage Range 1	-	-	40	
F <sub>CK</sub>	Quad SPI clock	2 < V <sub>DD</sub> < 3.6 V, C <sub>LOAD</sub> = 20 pF Voltage Range 1	-	-	48	MHz
1/t <sub>(CK)</sub>	frequency	1.71 < V <sub>DD</sub> < 3.6 V, C <sub>LOAD</sub> = 15 pF Voltage Range 1	-	-	48	IVITZ
		1.71 < V <sub>DD</sub> < 3.6 V C <sub>LOAD</sub> = 20 pF Voltage Range 2	-	-	26	
t <sub>w(CKH)</sub>	Quad SPI clock high	f <sub>AHBCLK</sub> = 48 MHz, presc=0	t <sub>(CK)</sub> /2-2	-	t <sub>(CK)</sub> /2	
t <sub>w(CKL)</sub>	and low time	IAHBCLK - 40 WII 12, presc-0	t <sub>(CK)</sub> /2	-	t <sub>(CK)</sub> /2+2	
4	Data input setup time	Voltage Range 1	1 3.5			
t <sub>sr(IN)</sub>	on rising edge	Voltage Range 2			-	
	Data input setup time Voltag	Voltage Range 1	1			
t <sub>sf(IN)</sub>	on falling edge	Voltage Range 2	1.5	-	-	
	Data input hold time	Voltage Range 1	6			
t <sub>hr(IN)</sub>	on rising edge	Voltage Range 2	6.5	] -	_	
	Data input hold time	Voltage Range 1	5.5			
t <sub>hf(IN)</sub>	on falling edge	Voltage Range 2	5.5	-	-	ns
4	Data output valid time	Voltage Range 1		5	5.5	
t <sub>vr(OUT)</sub>	on rising edge	Voltage Range 2	-	9.5	14	
1	Data output valid time	Voltage Range 1		5	8.5	
t <sub>vf(OUT)</sub>	on falling edge	Voltage Range 2	-	15	19	
1	Data output hold time	Voltage Range 1	3.5	-		
t <sub>hr(OUT)</sub>	on rising edge	Voltage Range 2	8	-	-	
1	Data output hold time	Voltage Range 1	3.5	-	1	
t <sub>hf(OUT)</sub>	on falling edge	Voltage Range 2	13	-	-	

<sup>1.</sup> Guaranteed by characterization results.

 $t_{(\mathsf{CK})}$  $t_{\text{w}(\text{CKH})}$  $t_{\text{w}(\text{CKL})}$  $t_{\text{f(CK)}}$ Clock t<sub>v(OUT)</sub>  $\overset{t_{h(OUT)}}{\longleftrightarrow}$ Data output D0 D1 D2  $t_{\text{s}(\text{IN})}$  $t_{h(IN)}$ Data input D0 D1 D2 MSv36878V1

Figure 26. Quad SPI timing diagram - SDR mode





#### **SAI** characteristics

Unless otherwise specified, the parameters given in *Table 79* for SAI are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 20: General operating conditions*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V<sub>DD</sub>

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CK,SD,FS).

Table 79. SAI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>MCLK</sub>	SAI Main clock output	-	-	50	MHz
		Master transmitter 2.7 ≤ V <sub>DD</sub> ≤ 3.6 Voltage Range 1	-	18.5	
		Master transmitter 1.71 ≤ V <sub>DD</sub> ≤ 3.6 Voltage Range 1	-	12.5	
		Master receiver Voltage Range 1	-	25	
f <sub>CK</sub>	SAI clock frequency <sup>(2)</sup>	Slave transmitter 2.7 ≤ V <sub>DD</sub> ≤ 3.6 Voltage Range 1	-	22.5	MHz
	Slave transmitter 1.71 ≤ V <sub>DD</sub> ≤ 3.6 Voltage Range 1 Slave receiver Voltage Range 1	1.71 ≤ V <sub>DD</sub> ≤ 3.6	-	14.5	
			-	25	
		Voltage Range 2	-	12.5	
	FS valid time	Master mode 2.7 ≤ V <sub>DD</sub> ≤ 3.6	-	22	20
t <sub>v(FS)</sub>	rs valid time	Master mode 1.71 ≤ V <sub>DD</sub> ≤ 3.6	-	40	ns
t <sub>h(FS)</sub>	FS hold time	Master mode	10	-	ns
t <sub>su(FS)</sub>	FS setup time	Slave mode	1	-	ns
t <sub>h(FS)</sub>	FS hold time	Slave mode	2	-	ns
t <sub>su(SD_A_MR)</sub>	Data input setup time	Master receiver	2	-	ns
t <sub>su(SD_B_SR)</sub>	Data input setup tille	Slave receiver	1.5	-	113
t <sub>h(SD_A_MR)</sub>	Data input hold time	Master receiver	5	-	ns
t <sub>h(SD_B_SR)</sub>	Data input noid time	Slave receiver	2.5	-	113



Table 73. OAI characteristics (continued)						
Symbol	Parameter	Conditions		Max	Unit	
t	Data output valid time	Slave transmitter (after enable edge) $2.7 \le V_{DD} \le 3.6$		22	ns	
t <sub>v(SD_B_ST)</sub> Data output valid time		Slave transmitter (after enable edge) $1.71 \le V_{DD} \le 3.6$	i	34	113	
t <sub>h(SD_B_ST)</sub>	Data output hold time	Slave transmitter (after enable edge)	10	-	ns	
		Master transmitter (after enable edge) $2.7 \le V_{DD} \le 3.6$	ı	27	ns	
t <sub>v(SD_A_MT)</sub>	Data output valid time	Master transmitter (after enable edge) $1.71 \le V_{DD} \le 3.6$		40	115	
t <sub>h(SD_A_MT)</sub>	Data output hold time	Master transmitter (after enable edge)		-	ns	

Table 79. SAI characteristics<sup>(1)</sup> (continued)

- 1. Guaranteed by characterization results.
- 2. APB clock frequency must be at least twice SAI clock frequency.

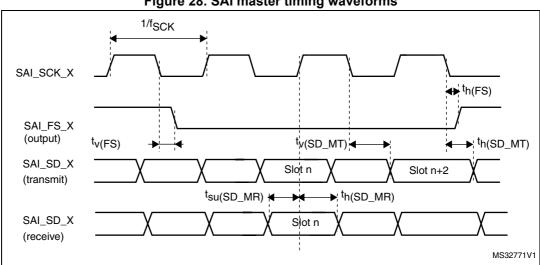


Figure 28. SAI master timing waveforms

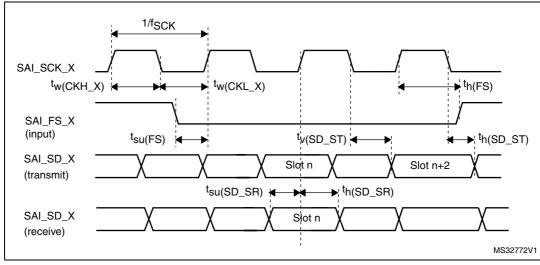


Figure 29. SAI slave timing waveforms

### **USB** characteristics

The STM32L432xx USB interface is fully compliant with the USB specification version 2.0 and is USB-IF certified (for Full-speed device operation).

	Table ov. OSB electrical characteristics.					
Symbol	Parameter	Parameter Conditions		Тур	Max	Unit
V <sub>DDUSB</sub>	USB transceiver operating voltage		3.0 <sup>(2)</sup>	-	3.6	V
T <sub>crystal_less</sub>	USB crystal less operation temp	USB crystal less operation temperature		-	85	°C
R <sub>PUI</sub>	Embedded USB_DP pull-up value during idle		900	1250	1600	
R <sub>PUR</sub>	Embedded USB_DP pull-up value during reception		1400	2300	3200	Ω
Z <sub>DRV</sub> <sup>(3)</sup>	Output driver impedance <sup>(4)</sup>	Driving high and low	28	36	44	Ω

Table 80. USB electrical characteristics<sup>(1)</sup>

### CAN (controller area network) interface

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CAN\_TX and CAN\_RX).

<sup>1.</sup>  $T_A$  = -40 to 125 °C unless otherwise specified.

<sup>2.</sup> The STM32L432xx USB functionality is ensured down to 2.7 V but not the full USB electrical characteristics which are degraded in the 2.7-to-3.0 V voltage range.

<sup>3.</sup> Guaranteed by design.

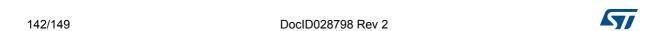
No external termination series resistors are required on USB\_DP (D+) and USB\_DM (D-); the matching impedance is already included in the embedded driver.

#### **SWPMI** characteristics

The Single Wire Protocol Master Interface (SWPMI) and the associated SWPMI\_IO transceiver are compliant with the ETSI TS 102 613 technical specification.

Table 81. SWPMI electrical characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>SWPSTART</sub>	SWPMI regulator startup time	SWP Class B 2.7 V ≤ V <sub>DD</sub> ≤ 3,3V	-	-	300	μs
towns	SWP bit duration	V <sub>CORE</sub> voltage range 1	500	•	-	ns
<sup>I</sup> SWPBIT	SVVF DIL GUIALIOIT	V <sub>CORE</sub> voltage range 2	620	-	-	113

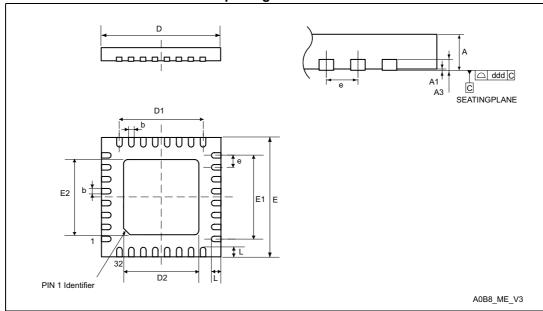


# 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status are available at: <a href="https://www.st.com">www.st.com</a>. ECOPACK<sup>®</sup> is an ST trademark.

## 7.1 UFQFPN32 package information

Figure 30. UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package outline



- 1. Drawing is not to scale.
- 2. There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this backside pad to PCB ground.

inches<sup>(1)</sup> millimeters **Symbol** Min Min Тур Max Тур Max 0.500 0.550 0.600 0.0197 0.0217 0.0236 Α Α1 0.050 0.0020 А3 0.152 0.0060 0.280 0.0071 0.0110 b 0.180 0.230 0.0091 4.900 5.000 5.100 0.1929 D 0.1969 0.2008 D1 3.400 3.500 3.600 0.1339 0.1378 0.1417 D2 3.400 3.500 3.600 0.1339 0.1378 0.1417 Ε 4.900 5.000 5.100 0.1929 0.1969 0.2008 E1 3.400 3.500 3.600 0.1339 0.1378 0.1417 E2 3.400 3.500 3.600 0.1339 0.1378 0.1417 е 0.500 0.0197 \_ \_ L 0.300 0.400 0.500 0.0118 0.0157 0.0197 ddd 0.080 0.0031

Table 82. UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package mechanical data

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

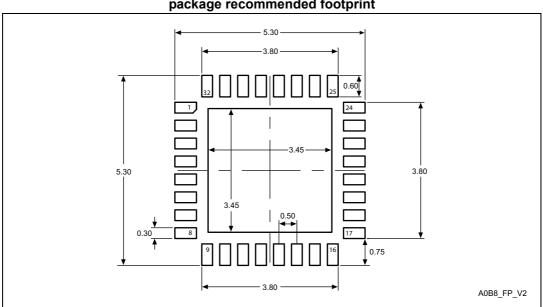


Figure 31. UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package recommended footprint

1. Dimensions are expressed in millimeters.

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.



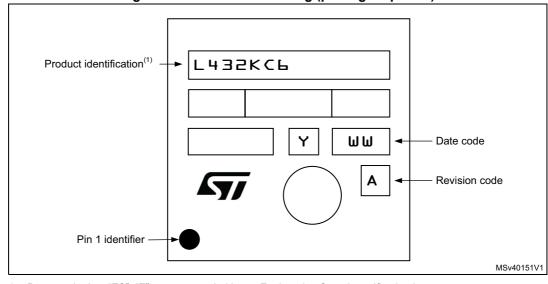


Figure 32. UFQFPN32 marking (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.



### 7.2 Thermal characteristics

The maximum chip junction temperature (T<sub>J</sub>max) must never exceed the values given in *Table 20: General operating conditions*.

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

$$T_J \max = T_A \max + (P_D \max x \Theta_{JA})$$

#### Where:

- T<sub>A</sub> max is the maximum ambient temperature in °C,
- Θ<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- P<sub>D</sub> max is the sum of P<sub>INT</sub> max and P<sub>I/O</sub> max (P<sub>D</sub> max = P<sub>INT</sub> max + P<sub>I/O</sub>max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

 $P_{I\!/O}$  max represents the maximum power dissipation on output pins where:

$$P_{I/O}$$
 max =  $\Sigma (V_{OL} \times I_{OL}) + \Sigma ((V_{DDIOx} - V_{OH}) \times I_{OH})$ ,

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Table 83. Package thermal characteristics

Symbol	Parameter	Value	Unit
$\Theta_{JA}$	Thermal resistance junction-ambient UFQFPN32 - 5 × 5 mm / 0.5 mm pitch	39	°C/W

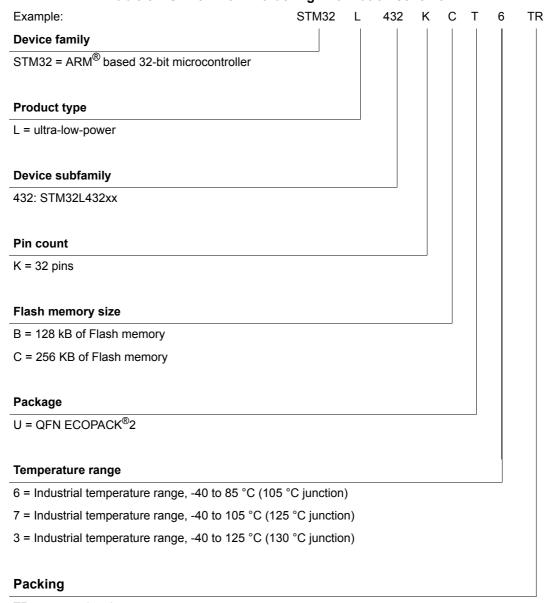
### 7.2.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org



# 8 Part numbering

Table 84. STM32L432xx ordering information scheme



TR = tape and reel

xxx = programmed parts

# 9 Revision history

Table 85. Document revision history

Date	Revision	Changes
08-Feb-2016	1	Initial release.
31-May-2016	2	Updated Table 1: STM32L432Kx family device features and peripheral counts. Updated Section 3.24: Universal synchronous/asynchronous receiver transmitter (USART). Updated Table 13: STM32L432xx pin definitions. Updated Table 15: Alternate function AF8 to AF15 (for AF0 to AF7 see Table 14). Updated Table 20: General operating conditions. Added Figure 10: VREFINT versus temperature. Updated Table 22: Embedded reset and power control block characteristics. Updated Table 24 to Table 26 and Table 30 to Table 38. Updated Table 38: Low-power mode wakeup timings. Added Table 40: Wakeup time using USART/LPUART. Updated Table 45: MSI oscillator characteristics. Added Figure 16: HSI48 frequency versus temperature. Updated Table 48: PLL, PLLSAI1 characteristics. Updated Table 51: EMS characteristics. Updated Table 52: EMI characteristics. Updated Table 52: EMI characteristics. Updated Table 52: EMI characteristics. Updated Table 60: Analog switches booster characteristics. Updated Table 60: Analog switches booster characteristics. Updated Table 61: ADC characteristics. Updated Table 69: COMP characteristics. Updated Table 80: USB electrical characteristics. Added Section: SWPMI characteristics. Updated Table 83: Package thermal characteristics.

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